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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	16
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	384 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 12x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VQFN Exposed Pad
Supplier Device Package	28-QFN (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1827-i-ml

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

PIC16(L)F1826/27 Family Types

	Program Memory	Da Men	ata nory		ch)	ch)	ſS	-bit)			idge)	idge)		
Device	Words	SRAM (bytes)	Data EEPROM (bytes)	ا/O's ⁽¹⁾	10-bit ADC (CapSense (Comparato	Timers (8/16-	EUSART	ASSM	ECCP (Full-Br	ECCP (Half-Br	ССР	SR Latch
PIC16LF1826	2K	256	256	16	12	12	2	2/1	1	1	1	_	-	Yes
PIC16F1826	2K	256	256	16	12	12	2	2/1	1	1	1			Yes
PIC16LF1827	4K	384	256	16	12	12	2	4/1	1	2	1	1	2	Yes
PIC16F1827	4K	384	256	16	12	12	2	4/1	1	2	1	1	2	Yes

Note 1: One pin is input only.

Pin Diagram – 18-Pin PDIP, SOIC (PIC16(L)F1826/27)

PDIP, SOIC



Pin Diagram – 20-Pin SSOP (PIC16(L)F1826/27)

SSOP



1.0 DEVICE OVERVIEW

The PIC16(L)F1826/27 are described within this data sheet. They are available in 18/20/28-pin packages. Figure 1-1 shows a block diagram of the PIC16(L)F1826/27 devices. Table 1-2 shows the pinout descriptions.

Reference Table 1-1 for peripherals available per device.

TABLE 1-1:DEVICE PERIPHERALSUMMARY

Peripheral	PIC16F/LF1826	PIC16(L)F1827	
ADC		•	•
Capacitive Sensing Mod	dule	•	•
Digital-to-Analog Conve	erter (DAC)	•	•
Digital Signal Modulator	r (DSM)	•	•
EUSART		•	•
Fixed Voltage Referenc	e (FVR)	•	•
Reference Clock Modul	e	•	•
SR Latch	•	•	
Capture/Compare/PWN	-		
	ECCP1	•	•
	ECCP2		•
	CCP3		•
	CCP4		•
Comparators		-	
	C1	•	•
	C2	•	•
Master Synchronous Se	erial Ports	-	
	MSSP1	•	•
	MSSP2		•
Timers	•	-	
	Timer0	•	•
	Timer1	•	•
	Timer2	•	•
	Timer4		•
	Timer6		•

3.4 Stack

All devices have a 16-level x 15-bit wide hardware stack (refer to Figures 3-5 through 3-8). The stack space is not part of either program or data space. The PC is PUSHed onto the stack when CALL or CALLW instructions are executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer if the STVREN bit is programmed to '0' (Configuration Word 2). This means that after the stack has been PUSHed sixteen times, the seventeenth PUSH overwrites the value that was stored from the first PUSH. The eighteenth PUSH overwrites the second PUSH (and so on). The STKOVF and STKUNF flag bits will be set on an Overflow/Underflow, regardless of whether the Reset is enabled.

Note 1: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, CALLW, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

3.4.1 ACCESSING THE STACK

The stack is available through the TOSH, TOSL and STKPTR registers. STKPTR is the current value of the Stack Pointer. TOSH:TOSL register pair points to the TOP of the stack. Both registers are read/writable. TOS is split into TOSH and TOSL due to the 15-bit size of the PC. To access the stack, adjust the value of STKPTR, which will position TOSH:TOSL, then read/write to TOSH:TOSL. STKPTR is 5 bits to allow detection of overflow and underflow.

Note:	Care should be taken when modifying the
	STKPTR while interrupts are enabled.

During normal program operation, CALL, CALLW and Interrupts will increment STKPTR while RETLW, RETURN, and RETFIE will decrement STKPTR. At any time STKPTR can be inspected to see how much stack is left. The STKPTR always points at the currently used place on the stack. Therefore, a CALL or CALLW will increment the STKPTR and then write the PC, and a return will unload the PC and then decrement STKPTR.

Reference Figure 3-5 through Figure 3-8 for examples of accessing the stack.





4.0 DEVICE CONFIGURATION

Device Configuration consists of Configuration Word 1 and Configuration Word 2, Code Protection and Device ID.

4.1 Configuration Words

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word 1 at 8007h and Configuration Word 2 at 8008h.

Note:	The DEBUG bit in Configuration Word is
	managed automatically by device
	development tools including debuggers
	and programmers. For normal device
	operation, this bit should be maintained as
	a '1'.

FIGURE 5-3:

QUARTZ CRYSTAL OPERATION (LP, XT OR HS MODE)



- Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
 - **2:** Always verify oscillator performance over the VDD and temperature range that is expected for the application.
 - **3:** For oscillator design assistance, reference the following Microchip Applications Notes:
 - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices" (DS00826)
 - AN849, "Basic PIC[®] Oscillator Design" (DS00849)
 - AN943, "Practical PIC[®] Oscillator Analysis and Design" (DS00943)
 - AN949, "Making Your Oscillator Work" (DS00949)

FIGURE 5-4: CERAMIC RESONATOR OPERATION

(XT OR HS MODE)



3: An additional parallel feedback resistor (RP) may be required for proper ceramic resonator operation.

5.2.1.3 Oscillator Start-up Timer (OST)

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see **Section 5.4 "Two-Speed Clock Start-up Mode"**).

5.2.1.4 4X PLL

The oscillator module contains a 4X PLL that can be used with both external and internal clock sources to provide a system clock source. The input frequency for the 4X PLL must fall within specifications. See the PLL Clock Timing Specifications in **Section 30.0 "Electrical Specifications"**.

The 4X PLL may be enabled for use by one of two methods:

- 1. Program the PLLEN bit in Configuration Word 2 to a '1'.
- Write the SPLLEN bit in the OSCCON register to a '1'. If the PLLEN bit in Configuration Word 2 is programmed to a '1', then the value of SPLLEN is ignored.

5.2.1.5 TIMER1 Oscillator

The Timer1 Oscillator is a separate crystal oscillator that is associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the T1OSO and T1OSI device pins.

The Timer1 Oscillator can be used as an alternate system clock source and can be selected during run-time using clock switching. Refer to **Section 5.3 "Clock Switching"** for more information.

FIGURE 5-5: QUARTZ CRYSTAL OPERATION (TIMER1 OSCILLATOR)



- Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
 - 2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.
 - **3:** For oscillator design assistance, reference the following Microchip Applications Notes:
 - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices" (DS00826)
 - AN849, "Basic PIC[®] Oscillator Design" (DS00849)
 - AN943, "Practical PIC[®] Oscillator Analysis and Design" (DS00943)
 - AN949, "Making Your Oscillator Work" (DS00949)
 - TB097, "Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS" (DS91097)
 - AN1288, "Design Practices for Low-Power External Oscillators" (DS01288)

5.2.1.6 External RC Mode

The external Resistor-Capacitor (RC) modes support the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required.

The RC circuit connects to OSC1. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. The function of the OSC2/CLKOUT pin is determined by the state of the CLKOUTEN bit in Configuration Word 1.

Figure 5-6 shows the external RC mode connections.



The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- · threshold voltage variation
- component tolerances
- · packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

FIGURE 5-7:	INTERNAL OSCILLATOR SWITCH TIMING
SENETORC/	UFBNTOSC (FBCSS and WDT (Fbablad)
IRCF <3:0>	$\neq 0$ $X = 0$
System Clock	
nfintosc/ i Mfintosc	LENETCISC (ENNET FISCHI es WIDT enabled)
HFINTOSC/ MEINTOSC	
LFINTOSC	
IRCF <3:0>	$\neq 0$ $\chi = 0$
System Clock	
LEBITOSC 1	EPIRTOSCARFINITOSC LPREVOIC fume of univer WST or PSCOR is enabled
KUT DIR EKARDKA	University of the second se
MENTOSC/	
\$2CF <330>	
System Grook	

5.4 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device. This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC internal oscillator block as the clock source and go back to Sleep without waiting for the external oscillator to become stable.

Two-Speed Start-up provides benefits when the oscillator module is configured for LP, XT, or HS modes. The Oscillator Start-up Timer (OST) is enabled for these modes and must count 1024 oscillations before the oscillator can be used as the system clock source.

If the oscillator module is configured for any mode other than LP, XT or HS mode, then Two-Speed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

If the OST count reaches 1024 before the device enters Sleep mode, the OSTS bit of the OSCSTAT register is set and program execution switches to the external oscillator. However, the system may never operate from the external oscillator if the time spent awake is very short.

Note:	Executing a SLEEP instruction will abort
	the oscillator start-up time and will cause
	the OSTS bit of the OSCSTAT register to
	remain clear.

5.4.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Word 1) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 00.
- FOSC<2:0> bits in the Configuration Word 1 configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

TABLE 5-1: OS	CILLATOR SWITCHING DELAYS
---------------	---------------------------

Switch From	Switch To	Frequency	Oscillator Delay
Sleep/POR	LFINTOSC ⁽¹⁾ MFINTOSC ⁽¹⁾ HFINTOSC ⁽¹⁾	31 kHz 31.25 kHz-500 kHz 31.25 kHz-16 MHz	Oscillator Warm-up Delay (Twarm)
Sleep/POR	EC, RC ⁽¹⁾	DC – 32 MHz	2 cycles
LFINTOSC	EC, RC ⁽¹⁾	DC – 32 MHz	1 cycle of each
Sleep/POR	Timer1 Oscillator LP, XT, HS ⁽¹⁾	32 kHz-20 MHz	1024 Clock Cycles (OST)
Any clock source	MFINTOSC ⁽¹⁾ HFINTOSC ⁽¹⁾	31.25 kHz-500 kHz 31.25 kHz-16 MHz	2 μs (approx.)
Any clock source	LFINTOSC ⁽¹⁾	31 kHz	1 cycle of each
Any clock source	Timer1 Oscillator	32 kHz	1024 Clock Cycles (OST)
PLL inactive	PLL active	16-32 MHz	2 ms (approx.)

Note 1: PLL inactive.

8.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to **Section 9.0 "Power-Down Mode (Sleep)"** for more details.

8.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

8.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the Shadow registers:

- W register
- STATUS register (except for TO and PD)
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding Shadow register should be modified and the value will be restored when exiting the ISR. The Shadow registers are available in Bank 31 and are readable and writable. Depending on the user's application, other registers may also need to be saved.

10.0 WATCHDOG TIMER

The Watchdog Timer is a system timer that generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The Watchdog Timer is typically used to recover the system from unexpected events.

The WDT has the following features:

- · Independent clock source
- Multiple operating modes
 - WDT is always on
 - WDT is off when in Sleep
 - WDT is controlled by software
 - WDT is always off
- Configurable time-out period is from 1 ms to 256 seconds (typical)
- Multiple Reset conditions
- Operation during Sleep

FIGURE 10-1: WATCHDOG TIMER BLOCK DIAGRAM



NOTES:

FIGURE 17-1: DIGITAL-TO-ANALOG CONVERTER BLOCK DIAGRAM







R/W-x/u	R/W-x/u	R/W-x/u	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
MDCLODIS	MDCLPOL	MDCLSYNC			MDCL	_<3:0>		
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable I	oit	U = Unimplen	nented bit, read	l as '0'		
u = Bit is unch	anged	x = Bit is unkn	own	-n/n = Value a	t POR and BO	R/Value at all c	ther Resets	
'1' = Bit is set		'0' = Bit is clea	ared					
bit 7 MDCLODIS: Modulator Low Carrier Output Disable bit 1 = Output signal driving the peripheral output pin (selected by MDCL<3:0> of the MDCARL I								
is disabled 0 = Output signal driving the peripheral output pin (selected by MDCL<3:0> of the MDCARL regis is enabled							CARL register)	
bit 6	Dit 6 MDCLPOL: Modulator Low Carrier Polarity Select bit							
	1 = Selected low carrier signal is inverted							
bit 5		• Modulator Low	Corrier Syn	chronization En	able hit			
bit 5	1 = Modulate	or waits for a fall	ing edge on t	the low time carr	ier signal before	e allowing a swi	itch to the high	
	time car	rier					iter te tre rigri	
	0 = Modulate	or Output is not	synchronize	d to the low time	carrier signal ^{(*}	1)		
bit 4	Unimplemer	nted: Read as 'd)'					
bit 3-0	MDCL<3:0>	Modulator Data	High Carrie	r Selection bits ⁽	1)			
	1111 = Res	erved. No char	nel connect	ed.				
	•							
	•							
	1000 = Res	erved. No char	nel connect	ed.				
	0111 = CCF	P3 output (PWN	l Output mod	de only)				
	0101 = CCF	P2 output (PWN	Output mod	de only)				
	0100 = CC	P1 output (PWN	Output mod	de only)				
	0011 = Ref	erence Clock m	odule signal					
	001 = MD	CIN2 port pin						
	0000 = Vss							

REGISTER 23-4: MDCARL: MODULATION LOW CARRIER CONTROL REGISTER

Note 1: Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.

TABLE 23-1:	SUMMARY OF REGISTERS	S ASSOCIATED WITH DATA	SIGNAL MODULATOR MODE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
MDCARH	MDCHODIS	MDCHPOL	MDCHSYNC		MDCH<3:0>				200
MDCARL	MDCLODIS	MDCLPOL	MDCLSYNC	—		201			
MDCON	MDEN	MDOE	MDSLR	MDOPOL	MDOUT MDBIT		198		
MDSRC	MDMSODIS	_	—	—	MDMS<3:0>			199	

Legend: — = unimplemented, read as '0'. Shaded cells are not used in the Data Signal Modulator mode.

24.3 PWM Overview

Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined.

PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 24-3 shows a typical waveform of the PWM signal.

24.3.1 STANDARD PWM OPERATION

The standard PWM function described in this section is available and identical for CCP modules ECCP1, ECCP2, CCP3 and CCP4.

The standard PWM mode generates a Pulse-Width Modulation (PWM) signal on the CCPx pin with up to 10 bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PRx registers
- TxCON registers
- · CCPRxL registers
- · CCPxCON registers

Figure 24-4 shows a simplified block diagram of PWM operation.

- Note 1: The corresponding TRIS bit must be cleared to enable the PWM output on the CCPx pin.
 - 2: Clearing the CCPxCON register will relinquish control of the CCPx pin.

FIGURE 24-3: CCP PWM OUTPUT SIGNAL





SIMPLIFIED PWM BLOCK DIAGRAM





FIGURE 25-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE

28.0 IN-CIRCUIT SERIAL PROGRAMMING[™] (ICSP[™])

ICSP[™] programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP[™] programming:

- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- Vss

In Program/Verify mode the Program Memory, User IDs and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ICSPCLK pin is the clock input. For more information on ICSPTM refer to the "PIC16(L)F182X/PIC12(L)F1822 Memory Programming Specification" (DS41390).

28.1 High-Voltage Programming Entry Mode

The device is placed into High-Voltage Programming Entry mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP to VIHH.

Some programmers produce VPP greater than VIHH (9.0V), an external circuit is required to limit the VPP voltage. See Figure 28-1 for example circuit.



FIGURE 28-1: VPP LIMITER EXAMPLE CIRCUIT

PIC16(L)F1826/27.

Mnemonic, Operands		Description	0	14-Bit Opcode)	Status	Natar
		Description		MSb			LSb	Affected	Notes
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	2
ADDWFC	f, d	Add with Carry W and f	1	11	1101	dfff	ffff	C, DC, Z	2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	2
ASRF	f, d	Arithmetic Right Shift	1	11	0111	dfff	ffff	C, Z	2
LSLF	f, d	Logical Left Shift	1	11	0101	dfff	ffff	C, Z	2
LSRF	f, d	Logical Right Shift	1	11	0110	dfff	ffff	C, Z	2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	_	Clear W	1	00	0001	0000	00xx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	2
INCF	f. d	Increment f	1	00	1010	dfff	ffff	z	2
IORWE	f. d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	2
MOVE	fd	Move f	1	00	1000	dfff	ffff	7	2
MOVWE	., ⊆ f	Move W to f	1	00	0000	1fff	ffff	-	2
RIF	fd	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	2
RRF	f d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	2
SUBWE	f d	Subtract W from f	1	00	0010	dfff	ffff		2
SUBWEB	fd	Subtract with Borrow W from f	1	11	1011	dfff	ffff	C DC Z	2
SWAPE	f d	Swan nibbles in f	1	00	1110	dfff	ffff	0, 00, 2	2
XORWE	f d	Exclusive OR W with f	1	00	0110	dfff	ffff	7	2
XOIT			0110	uIII	LLLL	2	2		
	fd	Decrement f Skin if 0			1011	4fff	<i>f f f f f f f f f f </i>		1 2
DECESZ	r, u f d	Increment f. Skip if 0	1(2)	00	1111	JEEE			1, 2
INCFSZ	1, u		1(2)	00		aiii	IIII		1, 2
		BIT-ORIENTED FILE REGIST	ER OPER	RATION	IS				
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		2
	1	BIT-ORIENTED SKIP O	PERATIO	NS					
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		1, 2
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		1, 2
LITERAL	OPERATIO	NS							
ADDLW	k	Add literal and W	1	11	1110	kkkk	kkkk	C, DC, Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLB	k	Move literal to BSR	1	00	0000	001k	kkkk		
MOVLP	k	Move literal to PCLATH	1	11	0001	1kkk	kkkk		
MOVLW	k	Move literal to W	Move literal to W 1 11 0000 k ²		kkkk	kkkk			
SUBLW k Subtract W from literal 1		1	11	1100	kkkk	kkkk	C, DC, Z		
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	
L									

TABLE 29-3: PIC16(L)F1826/27 ENHANCED INSTRUCTION SET

Note 1:If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

2: If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.

RRF	Rotate Right f through Carry				
Syntax:	[<i>label</i>] RRF f,d				
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$				
Operation:	See description below				
Status Affected:	С				
Description:	The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.				
	C Register f				

SUBLW	Subtract W from literal				
Syntax:	[label] SUBLW k				
Operands:	$0 \le k \le 255$				
Operation:	$k - (W) \to (W)$				
Status Affected:	C, DC, Z				
Description:	The W register is subtracted (2's com- plement method) from the eight-bit literal 'k'. The result is placed in the W register.				
	C = 0	W > k			
	C = 1	$W \leq k$			
	DC = 0	W<3:0> > k<3:0>			

DC = 1

 $W<3:0> \le k<3:0>$

 $W<3:0> \le f<3:0>$

SLEEP	Enter Sleep mode				
Syntax:	[label] SLEEP				
Operands:	None				
Operation:	$\begin{array}{l} 00h \rightarrow WDT, \\ 0 \rightarrow \underline{W}DT \text{ prescaler}, \\ 1 \rightarrow \underline{TO}, \\ 0 \rightarrow \overline{PD} \end{array}$				
Status Affected:	TO, PD				
Description:	The power-down Status bit, $\overline{\text{PD}}$ is cleared. Time-out Status bit, $\overline{\text{TO}}$ is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.				

SUBWF	Subtract W from f				
Syntax:	[label] SU	JBWF f,d			
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$				
Operation:	(f) - (W) \rightarrow (destination)				
Status Affected:	C, DC, Z				
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f.				
	C = 0	W > f			
	C = 1	$W \leq f$			
	DC = 0	W<3:0> > f<3:0>			

	Out (as a) W (as a fueld, Demonstra
SORMLR	Subtract w from f with Borrow
Syntax:	SUBWFB f {,d}
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(f) - (W) - (\overline{B}) \rightarrow dest$
Status Affected:	C, DC, Z
Description:	Subtract W and the BORROW flag (CARRY) from register 'f' (2's comple- ment method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

DC = 1

SWAPF	Swap Nibbles in f				
Syntax:	[label] SWAPF f,d				
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in [0,1] \end{array}$				
Operation:	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$				
Status Affected:	None				
Description:	The upper and lower nibbles of regis- ter 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.				

XORLW	Exclusive OR literal with W				
Syntax:	[label] XORLW k				
Operands:	$0 \le k \le 255$				
Operation:	(W) .XOR. $k \rightarrow (W)$				
Status Affected:	Z				
Description:	The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.				

TRIS	Load TRIS Register with W	XORWF	Exclusive OR W with f			
Syntax:	[label] TRIS f	Syntax:	[<i>label</i>] XORWF f,d			
Operands:	$5 \leq f \leq 7$	Operands:	$0 \le f \le 127$			
Operation: $(W) \rightarrow TRIS \text{ register 'f'}$			$u \in [0, \bot]$			
Status Affected:	None	Operation:	(W) .XOR. (f) \rightarrow (destination) Z			
Description:	Move data from W register to TRIS	Status Affected:				
Description: Move data from W register to TRIS register. When 'f' = 5, TRISA is loaded. When 'f' = 6, TRISB is loaded. When 'f' = 7, TRISC is loaded.		Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.			

30.2 DC Characteristics: PIC16(L)F1826/27-I/E (Industrial, Extended) (Continued)

PIC16LF1826/27			$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array} $				
PIC16F1826/27			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended				less otherwise stated) $A \le +85^{\circ}C$ for industrial $A \le +125^{\circ}C$ for extended
Param Device		Typt	Max	Unite	Conditions		
No.	Characteristics	WIIII.	турт	Wax.	Units	Vdd	Note
	Supply Current (IDD) ^{(1,}	2)					
D014		_	260	475	μA	1.8	Fosc = 4 MHz
		—	550	800	μA	3.0	EC Oscillator mode, Medium-power mode
D014		—	375	655	μA	1.8	Fosc = 4 MHz
		_	600	800	μA	3.0	EC Oscillator mode
		—	650	930	μA	5.0	
D015		—	3.6	10	μΑ	1.8	Fosc = 31 kHz
		—	7.0	15	μA	3.0	LFINTOSC mode
D015		_	21	42	μΑ	1.8	Fosc = 31 kHz
		—	27	55	μΑ	3.0	LFINTOSC mode
		—	28	60	μA	5.0	
D016		—	110	210	μA	1.8	Fosc = 500 kHz
		—	150	250	μA	3.0	MFINTOSC mode
D016		-	150	250	μA	1.8	Fosc = 500 kHz
		-	210	345	μA	3.0	MFINTOSC mode
		-	270	425	μA	5.0	
D017*		—	0.8	1.5	mA	1.8	Fosc = 8 MHz
		—	1.3	2.4	mA	3.0	HFINTOSC mode
D017*		_	1.0	2.0	mA	1.8	Fosc = 8 MHz
		_	1.5	2.6	mA	3.0	
		—	1.7	2.8	mA	5.0	
D018		_	1.2	2.5	mA	1.8	Fosc = 16 MHz
		—	2.5	3.75	mA	3.0	
D018		_	1.7	2.23	mA	1.8	Fosc = 16 MHz
		—	2.7	4.3	mA	3.0	
		-	3.0	4.6	mA	5.0	

These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins as inputs, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

- 3: 8 MHz internal RC oscillator with 4x PLL enabled.
- 4: 8 MHz crystal oscillator with 4x PLL enabled.

5: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in kΩ.