### Microchip Technology - PIC16LF18426-I/P Datasheet





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### Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	12
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x12b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	14-DIP (0.300", 7.62mm)
Supplier Device Package	14-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf18426-i-p

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- Multiple gate modes
- Time base for capture/compare function
- Timer2/4/6 with Hardware Limit Timer:
  - 8-bit timers
  - Programmable prescaler/postscaler
  - Time base for PWM function
  - Hardware Limit (HLT) and one-shot extensions
  - Selectable clock sources
- Signal Measurement Timer (SMT)
  - 1 SMT(s)
  - 24-bit timer/counter with programmable prescaler

## **Analog Peripherals**

- Analog-to-Digital Converter with Computation (ADC<sup>2</sup>):
  - 12-bit with up to 17 external channels
  - Conversion available during Sleep
  - Automated post-processing
  - Automated math functions on input signals:
    - Averaging, filter calculations, oversampling and threshold comparison
  - Integrated charge pump for low-voltage operation
  - CVD support
- Zero-Cross Detect (ZCD):
  - AC high voltage zero-crossing detection for simplifying TRIAC control
  - Synchronized switching control and timing
- Temperature Sensor Circuit
- Comparator:
  - 2 Comparators
  - Fixed Voltage Reference at (non)inverting input(s)
  - Comparator outputs externally accessible
- Digital-to-Analog Converter (DAC):
  - 5-bit resolution, rail-to-rail
  - Positive Reference Selection
  - Unbuffered I/O pin output
  - Internal connections to ADCs and comparators
- Fixed Voltage Reference (FVR) module:
  - 1.024V, 2.048V and 4.096V output levels

## **Flexible Oscillator Structure**

- High-Precision Internal Oscillator:
  - Software-selectable frequency range up to 32 MHz
  - ±2% at calibration (nominal)
- 4x PLL for use with external sources
  - up to 32 MHz (4-8 MHz input)

## 9. Oscillator Module (with Fail-Safe Clock Monitor)

## 9.1 Overview

The oscillator module has multiple clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. The following figure illustrates a block diagram of the oscillator module.

Clock sources can be supplied from external oscillators, quartz-crystal resonators and ceramic resonators. In addition, the system clock source can be supplied from one of two internal oscillators and PLL circuits, with a choice of speeds selectable via software. Additional clock features include:

- Selectable system clock source between external or internal sources via software.
- Fail-Safe Clock Monitor (FSCM) designed to detect a failure of the external clock source (LP, XT, HS, ECH, ECM, ECL) and switch automatically to the internal oscillator.
- Oscillator Start-up Timer (OST) ensures stability of crystal oscillator sources.

The RSTOSC bits of Configuration Word 1 determine the type of oscillator that will be used when the device runs after Reset, including when it is first powered up.

The internal clock modes, LFINTOSC, HFINTOSC (set at 1 MHz), or HFINTOSC (set at 32 MHz) can be set through the RSTOSC bits.

If an external clock source is selected, the FEXTOSC bits of Configuration Word 1 must be used in conjunction with the RSTOSC bits to select the External Clock mode.

The external oscillator module can be configured in one of the following clock modes, by setting the FEXTOSC bits of Configuration Word 1:

- ECL External Clock Low-Power mode (≤ 500 kHz)
- ECM External Clock Medium Power mode (≤ 8 MHz)
- ECH External Clock High-Power mode (≤ 32 MHz)
- 4. LP 32 kHz Low-Power Crystal mode.
- 5. XT Medium Gain Crystal or Ceramic Resonator Oscillator mode (between 100 kHz and 4 MHz)
- 6. HS High Gain Crystal or Ceramic Resonator mode (above 4 MHz)

The ECH, ECM, and ECL Clock modes rely on an external logic level signal as the device clock source. The LP, XT, and HS Clock modes require an external crystal or resonator to be connected to the device. Each mode is optimized for a different frequency range. The internal oscillator block produces low and high-frequency clock sources, designated LFINTOSC and HFINTOSC. Multiple device clock frequencies may be derived from these clock sources.

## 11.4 Register Summary - Power Savings Control

Offset	Name	Bit Pos.							
0x0812	VREGCON	7:0						VREGPM	
0x0813  0x088B	Reserved								
0x088C	CPUDOZE	7:0	IDLEN	DOZEN	ROI	DOE		DOZE[2:0]	

## 11.5 Register Definitions: Power Savings Control

### 12.8.4 WDTPSL

Name:WDTPSLAddress:0x80E

WWDT Prescale Select Low Register (Read-Only)

Bit	7	6	5	4	3	2	1	0
				PSCN	TL[7:0]			
Access	RO	RO	RO	RO	RO	RO	RO	RO
Reset	0	0	0	0	0	0	0	0

Bits 7:0 – PSCNTL[7:0] Prescale Select Low Byte bits<sup>(1)</sup>

### Note:

1. The 18-bit WDT prescale value, PSCNT<17:0> includes the WDTPSL, WDTPSH and the lower bits of the WDTTMR registers. PSCNT<17:0> is intended for debug operations and should be read during normal operation.

### 17.6.6 IOCBF

Name:IOCBFAddress:0x1F4A

PORTB Interrupt-on-Change Flag Register

Bit	7	6	5	4	3	2	1	0
	IOCBF7	IOCBF6	IOCBF5	IOCBF4				
Access	R/W/HS	R/W/HS	R/W/HS	R/W/HS				
Reset	0	0	0	0				

Bits 4, 5, 6, 7 – IOCBFn Interrupt-on-Change Flag bits

Value	Condition	Description
1	IOCBP[n]=1	A positive edge was detected on the RB[n] pin
1	IOCBN[n]=1	A negative edge was detected on the RB[n] pin
0	IOCBP[n]=x and	No change was detected, or the user cleared the detected change
	IOCBN[n]=x	

Note: PORTB associated registers are available on 20-pin or higher pin-count devices only.

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## (ADC2) Analog-to-Digital Converter with Comp...

ADC CI	ock Period (T <sub>AD</sub> )	Device Frequency (F <sub>OSC</sub> )						
ADC Clock Source	ADCLK	64 MHz	32 MHz	20 MHz	16 MHz	8 MHz	4 MHz	1 MHz
F <sub>OSC</sub> /6	000010	93.75 ns <sup>(2)</sup>	187.5 ns <sup>(2)</sup>	300 ns <sup>(2)</sup>	375 ns <sup>(2)</sup>	750 ns <sup>(2)</sup>	1.5 µs	6.0 µs
F <sub>OSC</sub> /8	000011	125 ns <sup>(2)</sup>	250 ns <sup>(2)</sup>	400 ns <sup>(2)</sup>	500 ns <sup>(2)</sup>	1.0 µs	2.0 µs	8.0 µs
F <sub>OSC</sub> /16	000100	250 ns <sup>(2)</sup>	500 ns <sup>(2)</sup>	800 ns <sup>(2)</sup>	1.0 µs	2.0 µs	4.0 µs	16.0 μs <sup>(3)</sup>
F <sub>OSC</sub> /128	111111	2.0 µs	4.0 µs	6.4 µs	8.0 µs	16.0 μs <sup>(3)</sup>	32.0 µs <sup>(2)</sup>	128.0 μs <sup>(2)</sup>
FRC	CS(ADCON0<4>) = 1	1.0-6.0 μs	1.0-6.0 µs	1.0-6.0 μs	1.0-6.0 µs	1.0-6.0 μs	1.0-6.0 µs	1.0-6.0 μs

### Note:

- 1. See T<sub>AD</sub> parameter in the "Electrical Specifications" section for FRC source typical T<sub>AD</sub> value.
- 2. These values violate the required T<sub>AD</sub> time.
- 3. Outside the recommended T<sub>AD</sub> time.
- The ADC clock period (T<sub>AD</sub>) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock F<sub>OSC</sub>. However, the FRC oscillator source must be used when conversions are to be performed with the device in Sleep mode.

### **Related Links**

### ADCON0

Analog-to-Digital Converter (ADC) Conversion Timing Specifications

### 20.1.5 Interrupts

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital Conversion. The ADC Interrupt Flag is the ADIF bit in the PIRx register. The ADC Interrupt Enable is the ADIE bit in the PIEx register. The ADIF bit must be cleared in software.



### Important:

- 1. The ADIF bit is set at the completion of every conversion, regardless of whether or not the ADC interrupt is enabled.
- 2. The ADC operates during Sleep only when the FRC oscillator is selected.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the ADIE bit and the PEIE bit of the INTCON register must both be set and the GIE bit of the

- The result of the calculation (ADERR) is compared to the upper and lower thresholds, ADUTH<UTHH:UTHL> and ADLTH<LTHH:LTHL> registers, to set the UTHR and LTHR flag bits. The threshold logic is selected by MD bits. The threshold trigger option can be one of the following:
  - Never interrupt
  - Error is less than lower threshold
  - Error is greater than or equal to lower threshold
  - Error is between thresholds (inclusive)
  - Error is outside of thresholds
  - Error is less than or equal to upper threshold
  - Error is greater than upper threshold
  - Always interrupt regardless of threshold test results
  - If the threshold condition is met, the threshold interrupt flag ADTIF is set.

### Note:

- 1. The threshold tests are signed operations.
- 2. If OV is set, a threshold interrupt is signaled. It is good practice for threshold interrupt handlers to verify the validity of the threshold by checking ADAOV.

### Table 20-6. ADC Error Calculation Mode

	Action During 1s		
CALC	DSEN = 0 Single-Sample Mode	DSEN = 1 CVD Double- Sample Mode <sup>(1)</sup>	Application
111	Reserved	Reserved	Reserved
110	Reserved	Reserved	Reserved
101	ADLFTR-ADSTPT	ADFLTR-ADSTPT	Average/filtered value vs. setpoint
100	ADPREV-ADFLTR	ADPREV-ADFLTR	First derivative of filtered value <sup>(3)</sup> (negative)
011	Reserved	Reserved	Reserved
010	ADRES-ADFLTR	(ADRES-ADPREV)-ADFLTR	Actual result vs. averaged/filtered value
001	ADRES-ADSTPT	(ADRES-ADPREV)-ADSTPT	Actual result vs. setpoint
000	ADRES-ADPREV	ADRES-ADPREV	First derivative of single measurement <sup>(2)</sup>
			Actual CVD result in CVD mode <sup>(2)</sup>

### Note:

- 1. When PSIS = 0, the value of ADRES-ADPREV) is the value of (S2-S1) from Computation Modes.
- 2. When **PSIS** = 0
- 3. When PSIS = 1.

### 20.6.8 Continuous Sampling Mode

Setting the CONT bit automatically retriggers a new conversion cycle after updating the ADACC register. The GO bit remains set and re-triggering occurs automatically.

## 23. (CMP) Comparator Module

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. Comparators are very useful mixed-signal building blocks because they provide analog functionality independent of program execution. The PIC16(L)F18426/46 devices have 2 comparators (C1/C2).

The analog comparator module includes the following features:

- Programmable input selection
- Programmable output polarity
- Rising/falling output edge interrupts
- Wake-up from Sleep
- CWG Auto-shutdown source
- Selectable voltage reference
- ADC Auto-trigger
- Odd numbered timers (Timer1, Timer3, etc.) Gate
- Even numbered timers (Timer2, Timer4, etc.) Reset
- CCP Capture Mode Input
- DSM Modulator Source
- Input and Window Signal-to-Signal Measurement Timer

### 23.1 Comparator Overview

A single comparator is shown in Figure 23-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at  $V_{IN}$ + is less than the analog voltage at  $V_{IN}$ -, the output of the comparator is a digital low level. When the analog voltage at  $V_{IN}$ + is greater than the analog voltage at  $V_{IN}$ +, the output of the comparator is a digital low level. When the analog voltage at  $V_{IN}$ + is greater than the analog voltage at  $V_{IN}$ -, the output of the comparator is a digital high level.

### Figure 23-1. Single Comparator



### Note:

1. The black areas of the output of the comparator represent the uncertainty due to input offsets and response time.

### 25.6.2 T0CON1

Name:T0CON1Address:0x59F

Timer0 Control Register 1

Bit	7	6	5	4	3	2	1	0
		T0CS[2:0]		TOASYNC		TOCK	PS[3:0]	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

**Bits 7:5 – T0CS[2:0]** Timer0 Clock Source Select bits Refer the clock source selection table

### Bit 4 – TOASYNC TMR0 Input Asynchronization Enable bit

Value	Description
1	The input to the TMR0 counter is not synchronized to system clocks
0	The input to the TMR0 counter is synchronized to Fosc/4

### Bits 3:0 - T0CKPS[3:0] Prescaler Rate Select bit

Value	Description
1111	1:32768
1110	1:16384
1101	1:8192
1100	1:4096
1011	1:2048
1010	1:1024
1001	1:512
1000	1:256
0111	1:128
0110	1:64
0101	1:32
0100	1:16
0011	1:8
0010	1:4
0001	1:2
0000	1:1

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### **Timer2 Module**

0843.05	Clock Source						
0353.02	Timer2	Timer4	Timer6				
0110	MFINTOSC(31.25 kHz)	MFINTOSC(31.25 kHz)	MFINTOSC(31.25 kHz)				
0101	MFINTOSC(500 kHz)	MFINTOSC(500 kHz)	MFINTOSC(500 kHz)				
0100	LFINTOSC	LFINTOSC	LFINTOSC				
0011	HFINTOSC(32 MHz)	HFINTOSC(32 MHz)	HFINTOSC(32 MHz)				
0010	F <sub>OSC</sub>	F <sub>OSC</sub>	F <sub>OSC</sub>				
0001	F <sub>OSC</sub> /4	F <sub>OSC</sub> /4	F <sub>OSC</sub> /4				
0000	T2CKIPPS	T4CKIPPS	T6CKIPPS				

## 27.1 Timer2 Operation

Timer2 operates in three major modes:

- Free Running Period
- One-shot
- Monostable

Within each mode there are several options for starting, stopping, and reset. Table 27-3 lists the options.

In all modes, the T2TMR count register is incremented on the rising edge of the clock signal from the programmable prescaler. When T2TMR equals T2PR, a high level is output to the postscaler counter. T2TMR is cleared on the next clock input.

An external signal from hardware can also be configured to gate the timer operation or force a T2TMR count Reset. In Gate modes the counter stops when the gate is disabled and resumes when the gate is enabled. In Reset modes the T2TMR count is reset on either the level or edge from the external source.

The T2TMR and T2PR registers are both directly readable and writable. The T2TMR register is cleared and the T2PR register initializes to FFh on any device Reset. Both the prescaler and postscaler counters are cleared on the following events:

- A write to the T2TMR register
- A write to the T2CON register
- Any device Reset
- External Reset Source event that resets the timer.



**Important:** T2TMR is not cleared when T2CON is written.

### 27.1.1 Free Running Period Mode

The value of T2TMR is compared to that of the Period register, T2PR, on each clock cycle. When the two values match, the comparator resets the value of T2TMR to 00h on the next cycle and increments the output postscaler counter. When the postscaler count equals the value in the OUTPS bits of the T2CON

### 27.9.2 TxPR

Name:	TxPR
Address:	0x28D,0x293,0x299

Timer Period Register

Bit	7	6	5	4	3	2	1	0
	TxPR[7:0]							
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	1	1	1	1	1	1	1	1

### Bits 7:0 – TxPR[7:0] Timer Period Register bits

Value	Description
0 - 255	The timer restarts at '0' when TxTMR reaches TxPR value

CWG module. When connected, as shown in Figure 31-5, the outputs are appropriate for a full-bridge motor driver. Each CWG output signal has independent polarity control, so the circuit can be adapted to high-active and low-active drivers. A simplified block diagram for the Full-Bridge modes is shown in Figure 31-6.

### Figure 31-5. Example of Full-Bridge Application



### 33.8.1 CLCxCON

Name:	CLCxCON
Address:	0x1E10,0x1E1A,0x1E24,0x1E2E

Configurable Logic Cell Control Register

Bit	7	6	5	4	3	2	1	0
	EN		OUT	INTP	INTN		MODE[2:0]	
Access	R/W		RO	R/W	R/W	R/W	R/W	R/W
Reset	0		0	0	0	0	0	0

### Bit 7 – EN

CLC Enable bit

Value	Description
1	Configurable logic cell is enabled and mixing signals
0	Configurable logic cell is disabled and has logic zero output

### Bit 5 – OUT

Logic cell output data, after LCPOL. Sampled from CLCxOUT

### Bit 4 – INTP

Configurable Logic Cell Positive Edge Going Interrupt Enable bit

Value	Description
1	CLCxIF will be set when a rising edge occurs on CLCxOUT
0	Rising edges on CLCxOUT have no effect on CLCxIF

### Bit 3 – INTN

Configurable Logic Cell Negative Edge Going Interrupt Enable bit

Value	Description
1	CLCxIF will be set when a falling edge occurs on CLCxOUT
0	Falling edges on CLCxOUT have no effect on CLCxIF

### Bits 2:0 - MODE[2:0]

Configurable Logic Cell Functional Mode Selection bits

Value	Description
111	Cell is 1-input transparent latch with Set and Reset
110	Cell is J-K flip-flop with Reset
101	Cell is 2-input D flip-flop with Reset
100	Cell is 1-input D flip-flop with Set and Reset
011	Cell is S-R latch
010	Cell is 4-input AND
001	Cell is OR-XOR
000	Cell is AND-OR

### 34.1 Clock Source

The clock source of the reference clock peripheral is selected with the CLK bits. The available clock sources are listed in the following table:

CLK	Clock Source				
1111-1011	Reserved				
1010	CLC4 OUT				
1001	CLC3 OUT				
1000	CLC2 OUT				
0111	CLC1 OUT				
0110	NCO1 OUT				
0101	SOSC				
0100	MFINTOSC (32 kHz)				
0011	MFINTOSC (500 kHz)				
0010	LFINTOSC				
0001	HFINTOSC (32 MHz)				
0000	F <sub>OSC</sub>				

### Table 34-1. CLKR Clock Sources

### 34.1.1 Clock Synchronization

The CLKR output signal is ensured to be glitch-free when the EN bit is set to start the module and enable the CLKR output.

When the reference clock output is disabled, the output signal will be disabled immediately.

Clock dividers and clock duty cycles can be changed while the module is enabled but doing so may cause glitches to occur on the output. To avoid possible glitches, clock dividers and clock duty cycles should be changed only when the EN bit is clear.

## 34.2 Programmable Clock Divider

The module takes the clock input and divides it based on the value of the DIV bits.

The following configurations are available:

- Base Fosc value
- F<sub>OSC</sub> divided by 2
- F<sub>OSC</sub> divided by 4
- F<sub>OSC</sub> divided by 8
- F<sub>OSC</sub> divided by 16
- F<sub>OSC</sub> divided by 32
- F<sub>OSC</sub> divided by 64
- F<sub>OSC</sub> divided by 128

- Master Transmit mode (master is transmitting data to a slave)
- Master Receive mode (master is receiving data from a slave)
- Slave Transmit mode (slave is transmitting data to a master)
- Slave Receive mode (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an ACK. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

The Acknowledge bit (ACK) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an  $\overline{ACK}$  bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an  $\overline{ACK}$  bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it sends the Stop bit in place of the last ACK bit. A Stop bit is indicated by a low-to-high transition of the SDA line while the SCL line is held high.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last  $\overline{ACK}$  bit when it is in receive mode.

The I<sup>2</sup>C bus specifies three message protocols;

- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.

When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a

### 35.4.2 Arbitration

Each master device must monitor the bus for Start and Stop bits. If the device detects that the bus is busy, it cannot begin a new message until the bus returns to an Idle state.

However, two master devices may try to initiate a transmission on or about the same time. When this occurs, the process of arbitration begins. Each transmitter checks the level of the SDA data line and compares it to the level that it expects to find. The first transmitter to observe that the two levels do not match, loses arbitration, and must stop transmitting on the SDA line.

For example, if one transmitter holds the SDA line to a logical one (lets it float) and a second transmitter holds it to a logical zero (pulls it low), the result is that the SDA line will be low. The first transmitter then observes that the level of the line is different than expected and concludes that another transmitter is communicating.

The first transmitter to notice this difference is the one that loses arbitration and must stop driving the SDA line. If this transmitter is also a master device, it also must stop driving the SCL line. It then can monitor the lines for a Stop condition before trying to reissue its transmission. In the meantime, the other device that has not noticed any difference between the expected and actual levels on the SDA line continues with its original transmission. It can do so without any complications, because so far, the transmission appears exactly as expected with no other transmitter disturbing the message.

Slave Transmit mode can also be arbitrated, when a master addresses multiple slaves, but this is less common.

If two master devices are sending a message to two different slave devices at the address stage, the master sending the lower slave address always wins arbitration. When two master devices send messages to the same slave address, and addresses can sometimes refer to multiple slaves, the arbitration process must continue into the data stage.

Arbitration usually occurs very rarely, but it is a necessary process for proper multi-master support.

### 35.4.3 Byte Format

All communication in I<sup>2</sup>C is done in 9-bit segments. A byte is sent from a master to a slave or vice-versa, followed by an Acknowledge bit sent back. After the eighth falling edge of the SCL line, the device outputting data on the SDA changes that pin to an input and reads in an acknowledge value on the next clock pulse.

The clock signal, SCL, is provided by the master. Data is valid to change while the SCL signal is low, and sampled on the rising edge of the clock. Changes on the SDA line while the SCL line is high define special conditions on the bus, explained below.

### 35.4.4 Definition of I<sup>2</sup>C Terminology

There is language and terminology in the description of  $I^2C$  communication that have definitions specific to  $I^2C$ . That word usage is defined below and may be used in the rest of this document without explanation. This table was adapted from the Philips  $I^2C$  specification.

TERM	Description			
Transmitter	The device which shifts data out onto the bus.			
Receiver	The device which shifts data in from the bus.			
Master	The device that initiates a transfer, generates clock signals and terminates a transfer.			

- Fully customizable interface: toolbars, toolbar buttons, windows, window placement, etc.
- Call graph window

Project-Based Workspaces:

- Multiple projects
- Multiple tools
- Multiple configurations
- Simultaneous debugging sessions

File History and Bug Tracking:

- Local file history feature
- Built-in support for Bugzilla issue tracker

## 41.2 MPLAB XC Compilers

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip's 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

### 41.3 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel<sup>®</sup> standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code

areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM<sup>™</sup> and dsPICDEM<sup>™</sup> demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KeeLoq<sup>®</sup> security ICs, CAN, IrDA<sup>®</sup>, PowerSmart battery management, SEEVAL<sup>®</sup> evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

## 41.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent<sup>®</sup> and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika<sup>®</sup>

# PIC16(L)F18426/46

## **Electrical Specifications**

Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C $\leq$ TA $\leq$ +125°C							
Param No. Sym. Characteristic Typ. Units Conditions							
TH04	PD	Power Dissipation		W	PD=P <sub>INTERNAL</sub> +P <sub>I/O</sub>		
TH05	PINTERNAL	Internal Power Dissipation		W	P <sub>INTERNAL</sub> =I <sub>DD</sub> XV <sub>DD</sub> <sup>(1)</sup>		
TH06	P <sub>I/O</sub>	I/O Power Dissipation		W	$P_{I/O} = \Sigma(I_{OL} * V_{OL}) + \Sigma(I_{OH} * (V_{DD} - V_{OH}))$		
TH07	P <sub>DER</sub>	Derated Power		W	$P_{DER}=PD_{MAX} (T_J-T_A)/\theta_{JA}^{(2)}$		
Note:							

- 1.  $I_{DD}$  is current to run the chip alone without driving any load on the output pins.
- 2.  $T_A$  = Ambient Temperature,  $T_J$  = Junction Temperature.

## 42.4 AC Characteristics

Figure 42-4. Load Conditions



Legend: CL=50 pF for all pins

## 42.4.1 External Clock/Oscillator Timing Requirements Figure 42-5. Clock Timing





## PIC16(L)F18426/46 Packaging Information

