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### Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

### Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

#### Details

Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	80MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (2), 10/100Mbps (1)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc860dtvr80d4">https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc860dtvr80d4</a>

- System integration unit (SIU)
  - Bus monitor
  - Software watchdog
  - Periodic interrupt timer (PIT)
  - Low-power stop mode
  - Clock synthesizer
  - Decrementer, time base, and real-time clock (RTC)
  - Reset controller
  - IEEE 1149.1™ Std. test access port (JTAG)
- Interrupts
  - Seven external interrupt request (IRQ) lines
  - 12 port pins with interrupt capability
  - 23 internal interrupt sources
  - Programmable priority between SCCs
  - Programmable highest priority request
- 10/100 Mbps Ethernet support, fully compliant with the IEEE 802.3u® Standard (not available when using ATM over UTOPIA interface)
- ATM support compliant with ATM forum UNI 4.0 specification
  - Cell processing up to 50–70 Mbps at 50-MHz system clock
  - Cell multiplexing/demultiplexing
  - Support of AAL5 and AAL0 protocols on a per-VC basis. AAL0 support enables OAM and software implementation of other protocols.
  - ATM pace control (APC) scheduler, providing direct support for constant bit rate (CBR) and unspecified bit rate (UBR) and providing control mechanisms enabling software support of available bit rate (ABR)
  - Physical interface support for UTOPIA (10/100-Mbps is not supported with this interface) and byte-aligned serial (for example, T1/E1/ADSL)
  - UTOPIA-mode ATM supports level-1 master with cell-level handshake, multi-PHY (up to four physical layer devices), connection to 25-, 51-, or 155-Mbps framers, and UTOPIA/system clock ratios of 1/2 or 1/3.
  - Serial-mode ATM connection supports transmission convergence (TC) function for T1/E1/ADSL lines, cell delineation, cell payload scrambling/descrambling, automatic idle/unassigned cell insertion/stripping, header error control (HEC) generation, checking, and statistics.
- Communications processor module (CPM)
  - RISC communications processor (CP)
  - Communication-specific commands (for example, GRACEFUL STOP TRANSMIT, ENTER HUNT MODE, and RESTART TRANSMIT)
  - Supports continuous mode transmission and reception on all serial channels

## 7 Thermal Calculation and Measurement

For the following discussions,  $P_D = (V_{DD} \times I_{DD}) + PI/O$ , where PI/O is the power dissipation of the I/O drivers.

### 7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature,  $T_J$ , in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

$T_A$  = ambient temperature (°C)

$R_{\theta JA}$  = package junction-to-ambient thermal resistance (°C/W)

$P_D$  = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity  $T_J - T_A$ ) are possible.

### 7.2 Estimation with Junction-to-Case Thermal Resistance

Historically, the thermal resistance has frequently been expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$  = junction-to-ambient thermal resistance (°C/W)

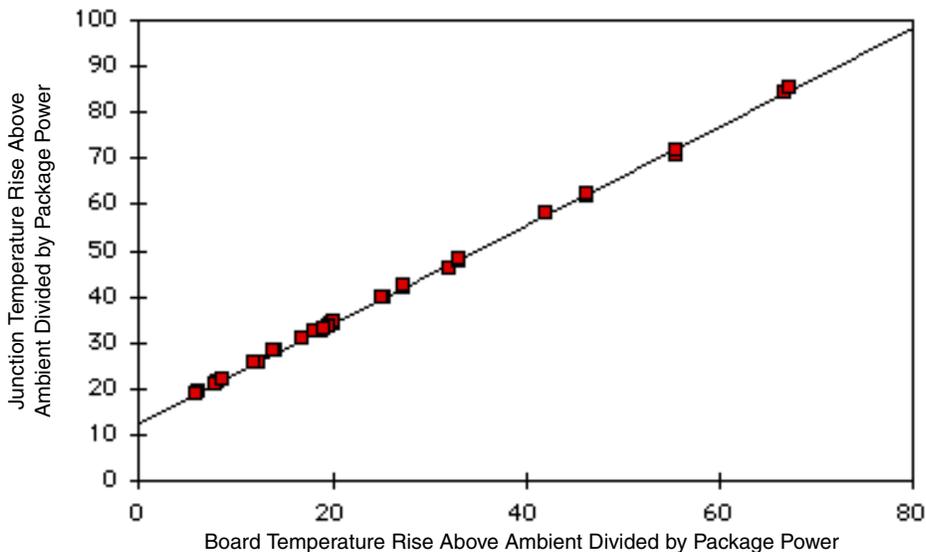
$R_{\theta JC}$  = junction-to-case thermal resistance (°C/W)

$R_{\theta CA}$  = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$  is device related and cannot be influenced by the user. The user adjusts the thermal environment to affect the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For instance, the user can change the airflow around the device, add a heat sink, change the mounting arrangement on the printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device. This thermal model is most useful for ceramic packages with heat sinks where some 90% of the heat flows through the case and the heat sink to the ambient environment. For most packages, a better model is required.

### 7.3 Estimation with Junction-to-Board Thermal Resistance

A simple package thermal model which has demonstrated reasonable accuracy (about 20%) is a two-resistor model consisting of a junction-to-board and a junction-to-case thermal resistance. The junction-to-case thermal resistance covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed-circuit board. It has been observed that the thermal performance of most plastic packages, especially PBGA packages, is strongly dependent on the board temperature; see [Figure 2](#).



**Figure 2. Effect of Board Temperature Rise on Thermal Behavior**

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$R_{\theta JB}$  = junction-to-board thermal resistance ( $^{\circ}\text{C}/\text{W}$ )

$T_B$  = board temperature ( $^{\circ}\text{C}$ )

$P_D$  = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and by attaching the thermal balls to the ground plane.

## 7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two-resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

## 7.5 Experimental Determination

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter ( $\Psi_{JT}$ ) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

Figure 5 provides the timing for the synchronous output signals.

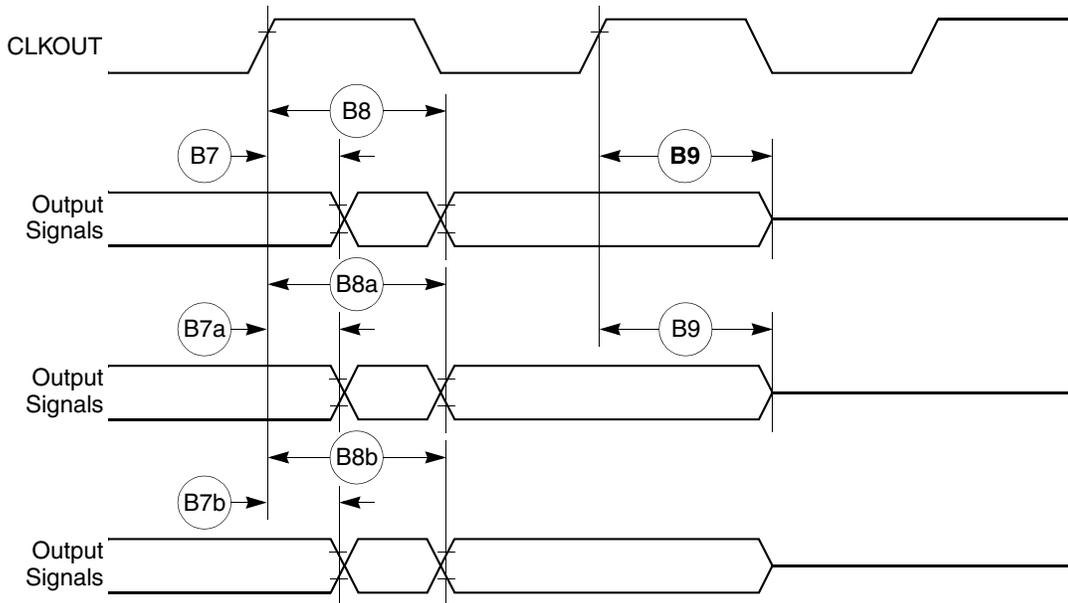


Figure 5. Synchronous Output Signals Timing

Figure 6 provides the timing for the synchronous active pull-up and open-drain output signals.

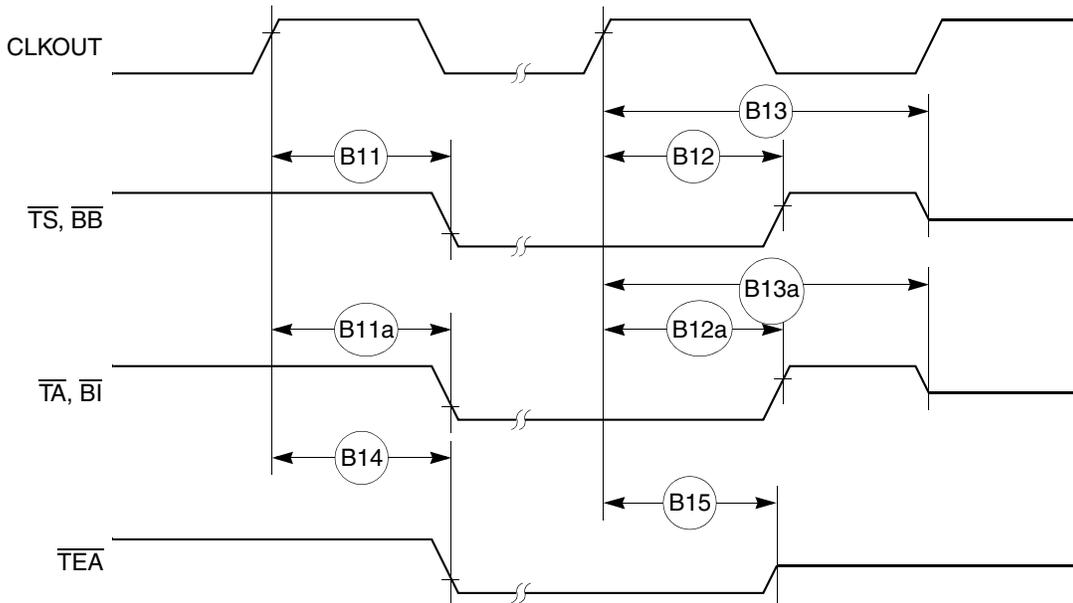


Figure 6. Synchronous Active Pull-Up Resistor and Open-Drain Outputs Signals Timing

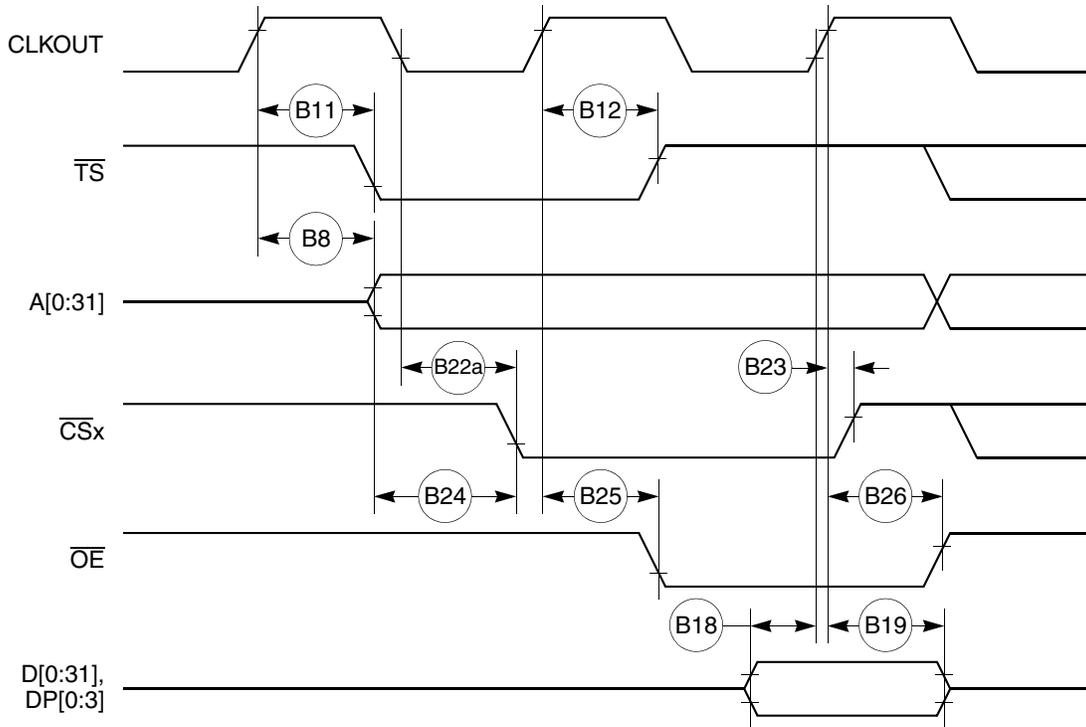


Figure 11. External Bus Read Timing (GPCM Controlled—TRLX = 0, ACS = 10)

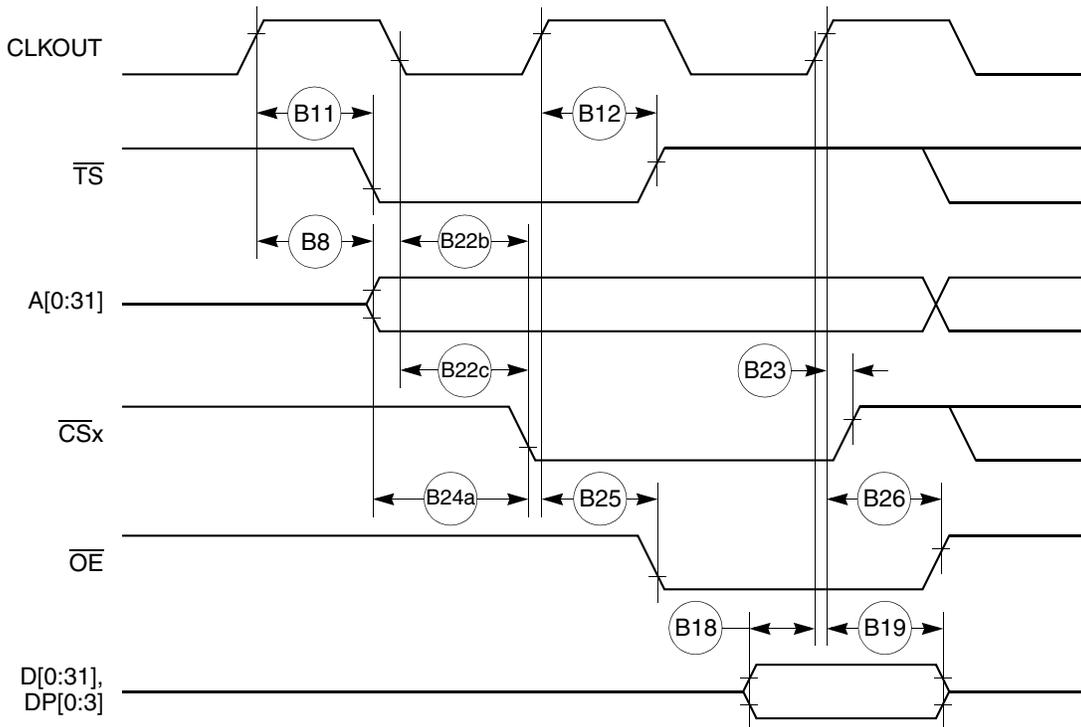


Figure 12. External Bus Read Timing (GPCM Controlled—TRLX = 0, ACS = 11)

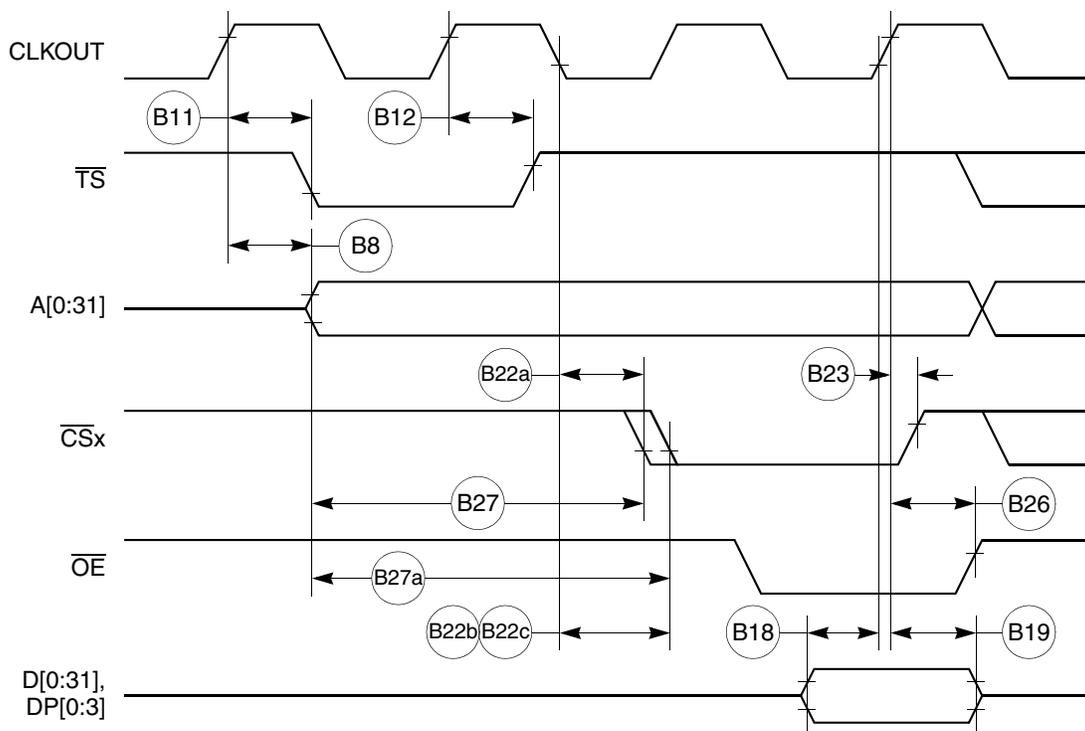


Figure 13. External Bus Read Timing (GPCM Controlled—TRLX = 0 or 1, ACS = 10, ACS = 11)

Table 9 shows the PCMCIA timing for the MPC860.

**Table 9. PCMCIA Timing**

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P44	A(0:31), $\overline{\text{REG}}$ valid to PCMCIA Strobe asserted <sup>1</sup>	20.73	—	16.75	—	13.00	—	9.36	—	ns
P45	A(0:31), $\overline{\text{REG}}$ valid to ALE negation <sup>1</sup>	28.30	—	23.00	—	18.00	—	13.15	—	ns
P46	CLKOUT to $\overline{\text{REG}}$ valid	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P47	CLKOUT to $\overline{\text{REG}}$ invalid	8.58	—	7.25	—	6.00	—	4.84	—	ns
P48	CLKOUT to $\overline{\text{CE1}}$ , $\overline{\text{CE2}}$ asserted	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P49	CLKOUT to $\overline{\text{CE1}}$ , $\overline{\text{CE2}}$ negated	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P50	CLKOUT to $\overline{\text{PCOE}}$ , $\overline{\text{IORD}}$ , $\overline{\text{PCWE}}$ , $\overline{\text{IOWR}}$ assert time	—	11.00		11.00	—	11.00	—	11.00	ns
P51	CLKOUT to $\overline{\text{PCOE}}$ , $\overline{\text{IORD}}$ , $\overline{\text{PCWE}}$ , $\overline{\text{IOWR}}$ negate time	2.00	11.00	2.00	11.00	2.00	11.00	2.00	11.00	ns
P52	CLKOUT to ALE assert time	7.58	15.58	6.25	14.25	5.00	13.00	3.79	10.04	ns
P53	CLKOUT to ALE negate time	—	15.58		14.25	—	13.00	—	11.84	ns
P54	$\overline{\text{PCWE}}$ , $\overline{\text{IOWR}}$ negated to D(0:31) invalid <sup>1</sup>	5.58	—	4.25	—	3.00	—	1.79	—	ns
P55	$\overline{\text{WAITA}}$ and $\overline{\text{WAITB}}$ valid to CLKOUT rising edge <sup>1</sup>	8.00	—	8.00	—	8.00	—	8.00	—	ns
P56	CLKOUT rising edge to $\overline{\text{WAITA}}$ and $\overline{\text{WAITB}}$ invalid <sup>1</sup>	2.00	—	2.00	—	2.00	—	2.00	—	ns

<sup>1</sup> PSST = 1. Otherwise add PSST times cycle time.

PSHT = 0. Otherwise add PSHT times cycle time.

These synchronous timings define when the  $\overline{\text{WAITx}}$  signals are detected in order to freeze (or relieve) the PCMCIA current cycle. The  $\overline{\text{WAITx}}$  assertion will be effective only if it is detected 2 cycles before the PSL timer expiration. See Chapter 16, “PCMCIA Interface,” in the *MPC860 PowerQUICC™ Family User’s Manual*.

Figure 26 provides the PCMCIA access cycle timing for the external bus write.

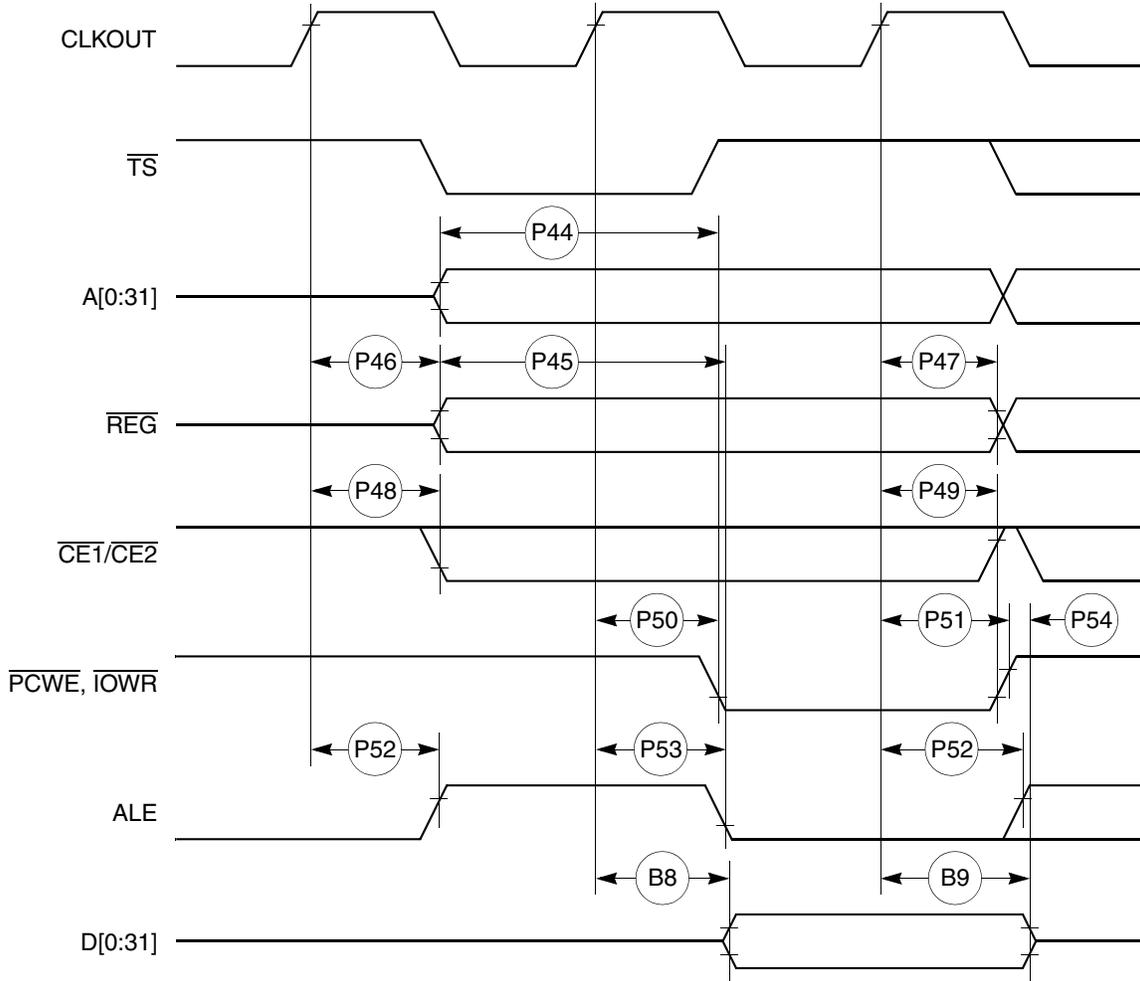


Figure 26. PCMCIA Access Cycle Timing External Bus Write

Figure 27 provides the PCMCIA  $\overline{\text{WAIT}}$  signal detection timing.

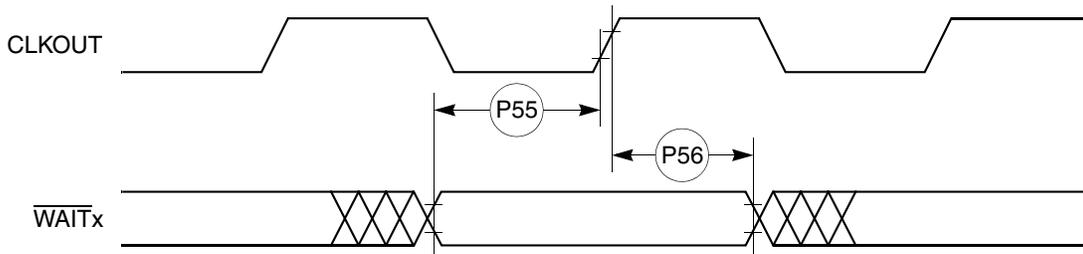


Figure 27. PCMCIA  $\overline{\text{WAIT}}$  Signal Detection Timing

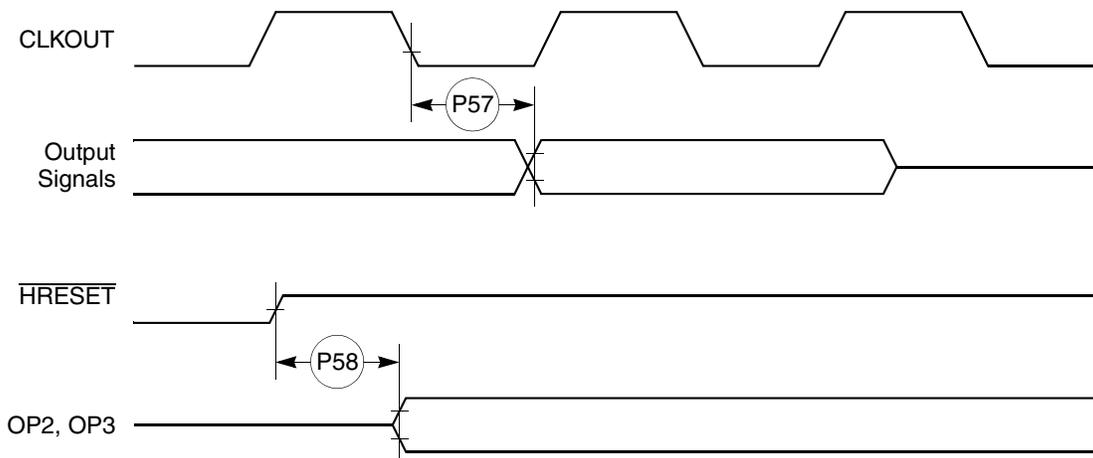
Table 10 shows the PCMCIA port timing for the MPC860.

**Table 10. PCMCIA Port Timing**

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P57	CLKOUT to OPx valid	—	19.00	—	19.00	—	19.00	—	19.00	ns
P58	$\overline{\text{HRESET}}$ negated to OPx drive <sup>1</sup>	25.73	—	21.75	—	18.00	—	14.36	—	ns
P59	IP_Xx valid to CLKOUT rising edge	5.00	—	5.00	—	5.00	—	5.00	—	ns
P60	CLKOUT rising edge to IP_Xx invalid	1.00	—	1.00	—	1.00	—	1.00	—	ns

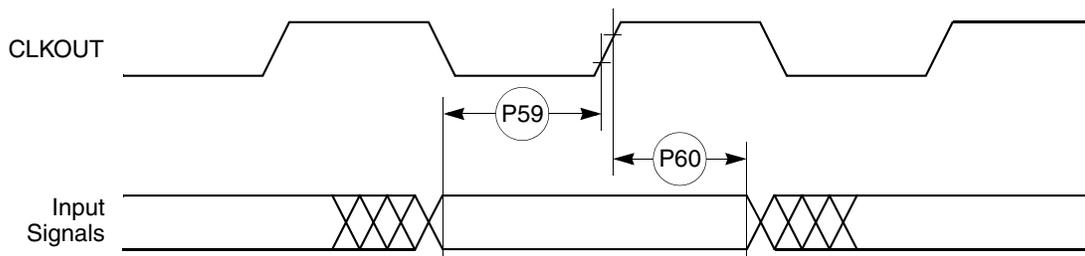
<sup>1</sup> OP2 and OP3 only.

Figure 28 provides the PCMCIA output port timing for the MPC860.



**Figure 28. PCMCIA Output Port Timing**

Figure 29 provides the PCMCIA output port timing for the MPC860.



**Figure 29. PCMCIA Input Port Timing**

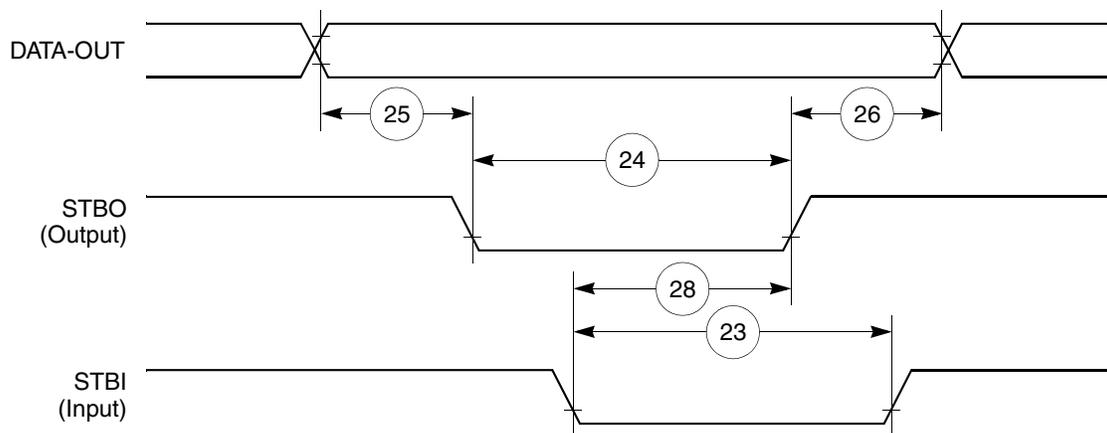


Figure 40. PIP Tx (Interlock Mode) Timing Diagram

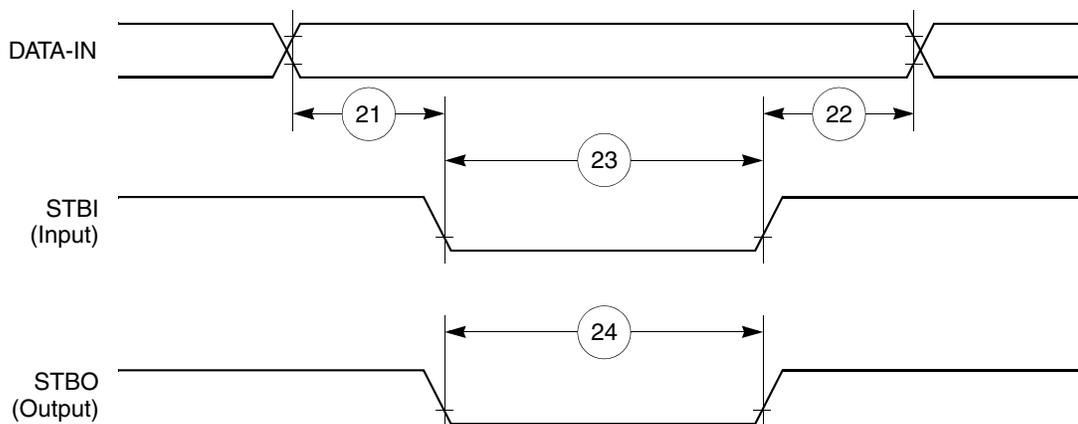


Figure 41. PIP Rx (Pulse Mode) Timing Diagram

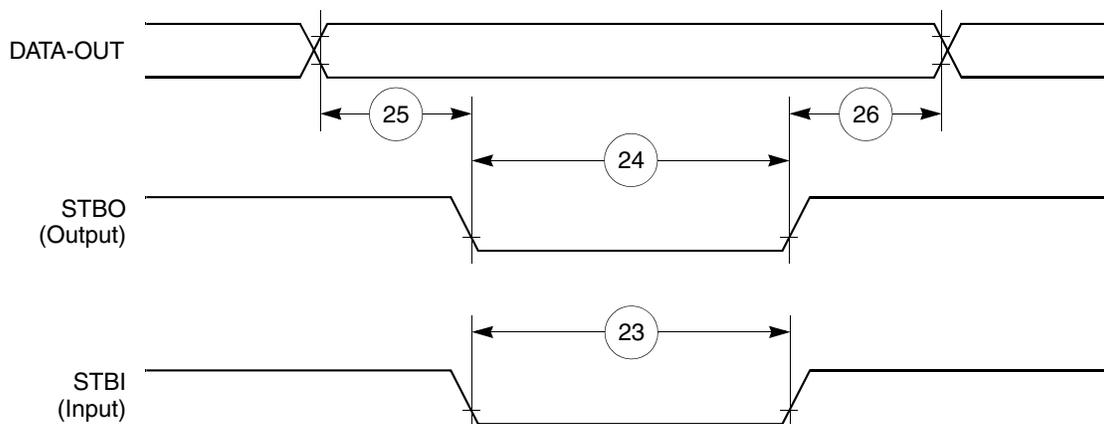


Figure 42. PIP TX (Pulse Mode) Timing Diagram

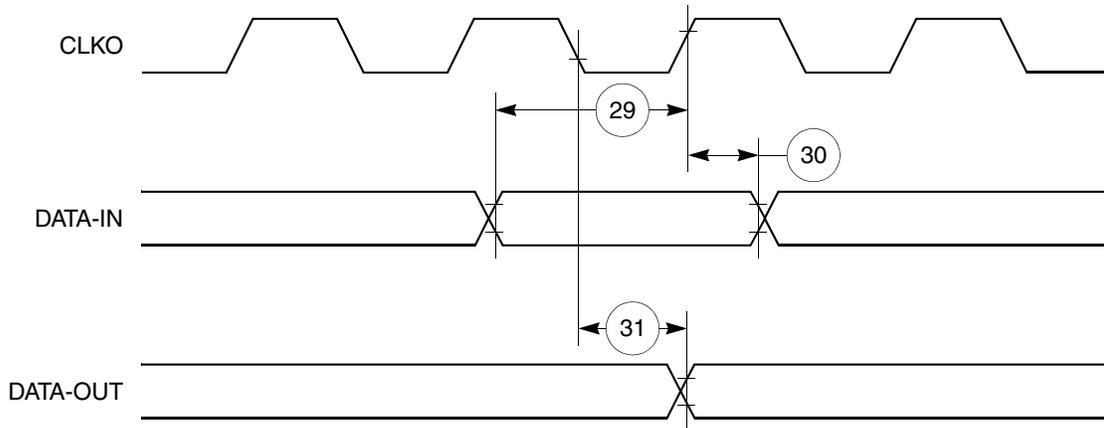


Figure 43. Parallel I/O Data-In/Data-Out Timing Diagram

## 11.2 Port C Interrupt AC Electrical Specifications

Table 15 provides the timings for port C interrupts.

Table 15. Port C Interrupt Timing

Num	Characteristic	≥ 33.34 MHz <sup>1</sup>		Unit
		Min	Max	
35	Port C interrupt pulse width low (edge-triggered mode)	55	—	ns
36	Port C interrupt minimum time between active edges	55	—	ns

<sup>1</sup> External bus frequency of greater than or equal to 33.34 MHz.

Figure 44 shows the port C interrupt detection timing.

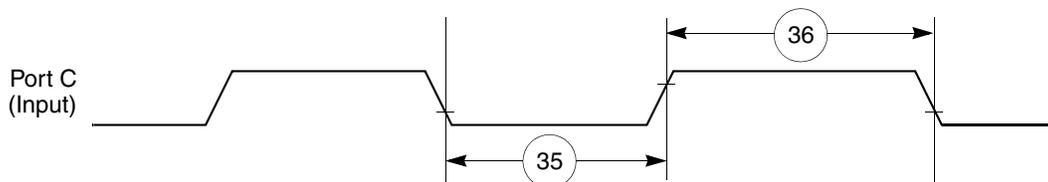


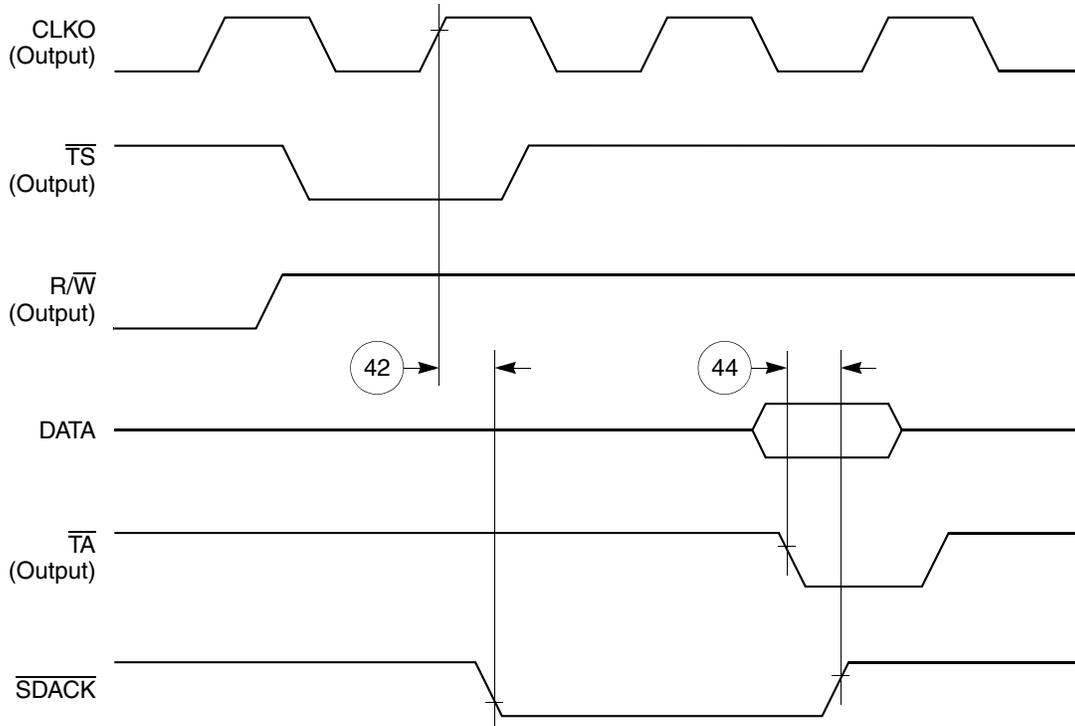
Figure 44. Port C Interrupt Detection Timing

## 11.3 IDMA Controller AC Electrical Specifications

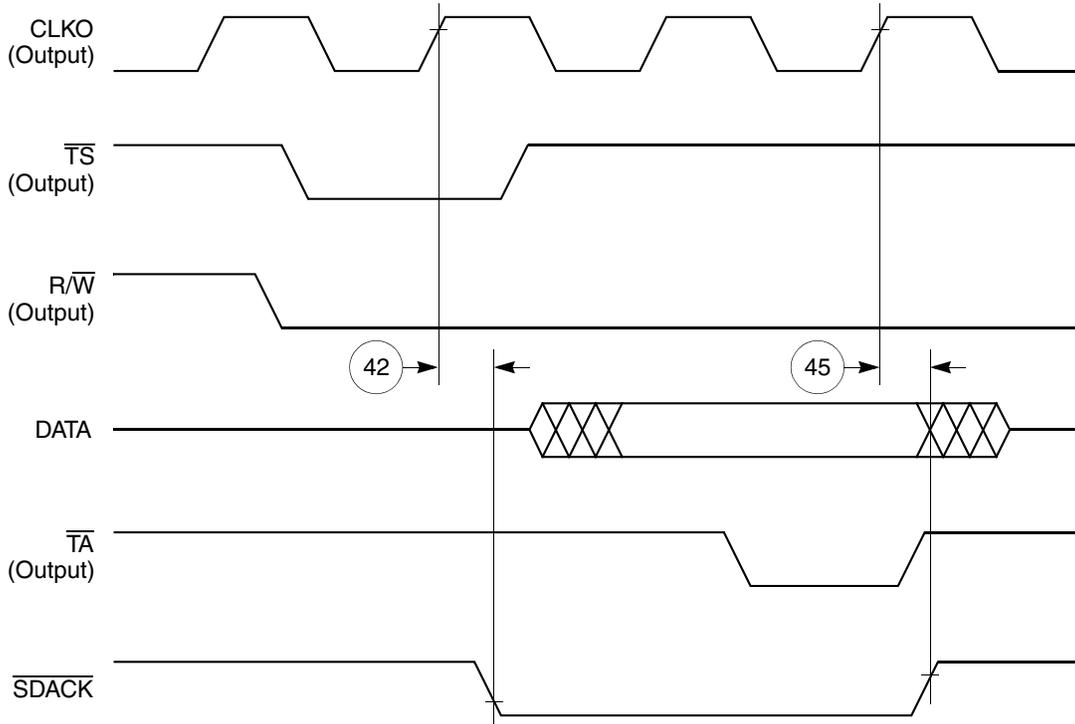
Table 16 provides the IDMA controller timings as shown in Figure 45 through Figure 48.

Table 16. IDMA Controller Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
40	$\overline{DREQ}$ setup time to clock high	7	—	ns
41	$\overline{DREQ}$ hold time from clock high	3	—	ns



**Figure 47.  $\overline{\text{SDACK}}$  Timing Diagram—Peripheral Write, Internally-Generated  $\overline{\text{TA}}$**



**Figure 48.  $\overline{\text{SDACK}}$  Timing Diagram—Peripheral Read, Internally-Generated  $\overline{\text{TA}}$**

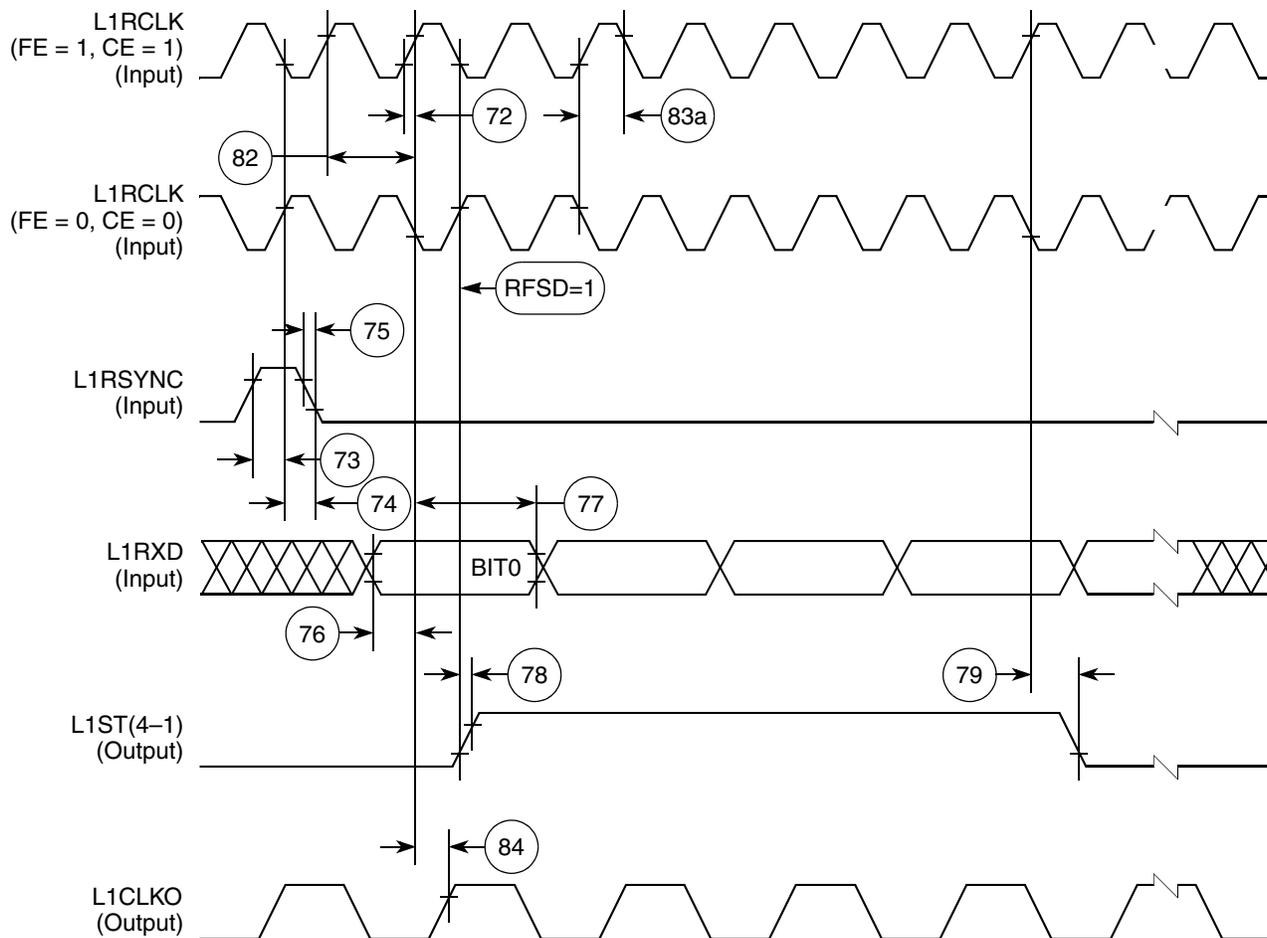


Figure 52. SI Receive Timing with Double-Speed Clocking (DSC = 1)

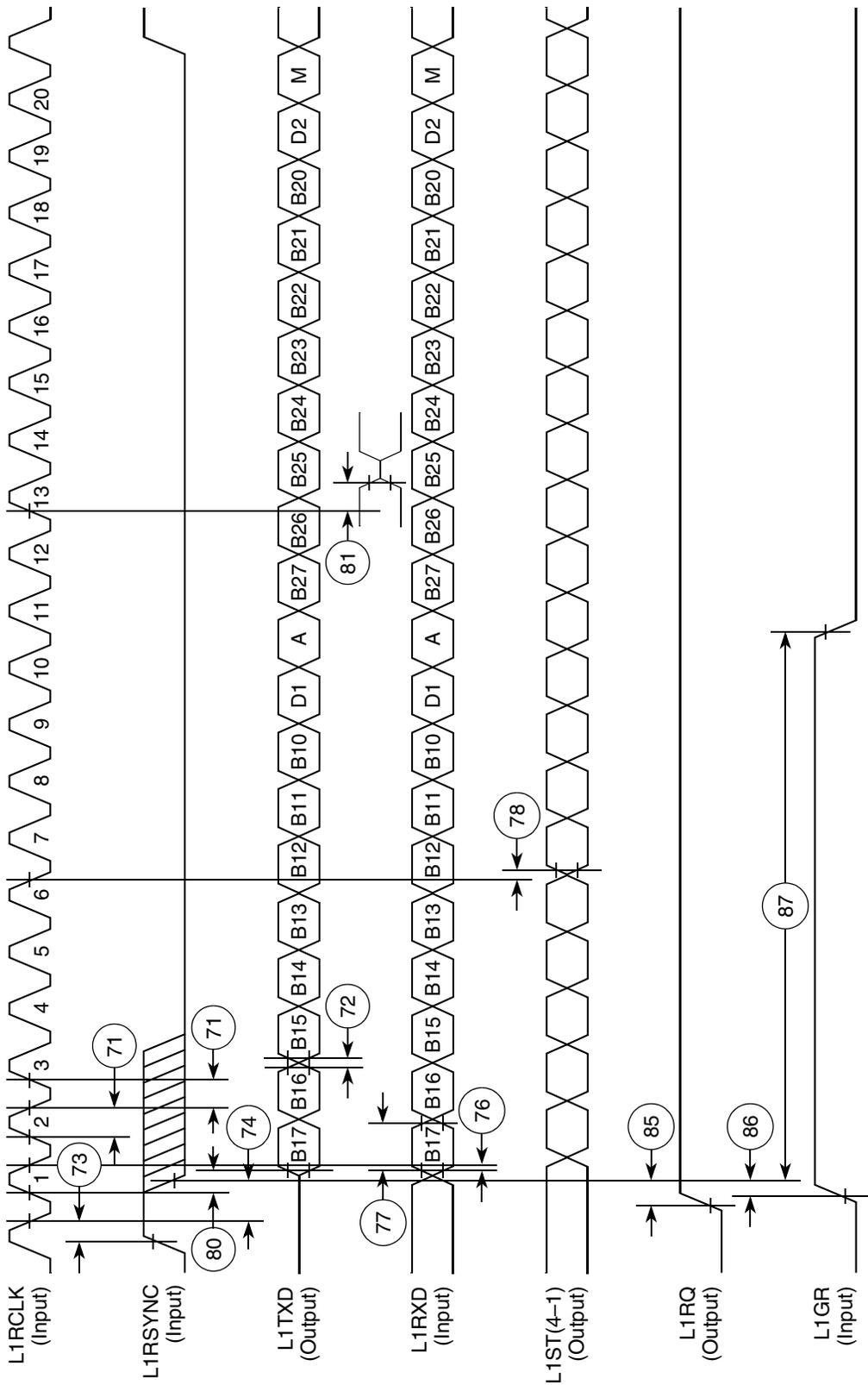


Figure 55. IDL Timing

Figure 56 through Figure 58 show the NMSI timings.

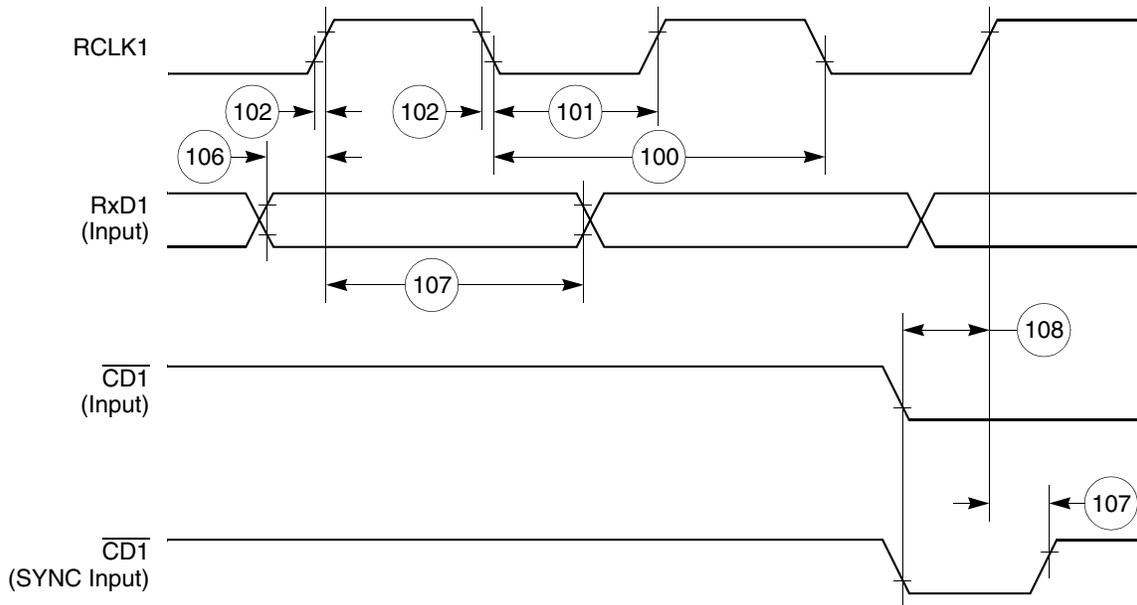


Figure 56. SCC NMSI Receive Timing Diagram

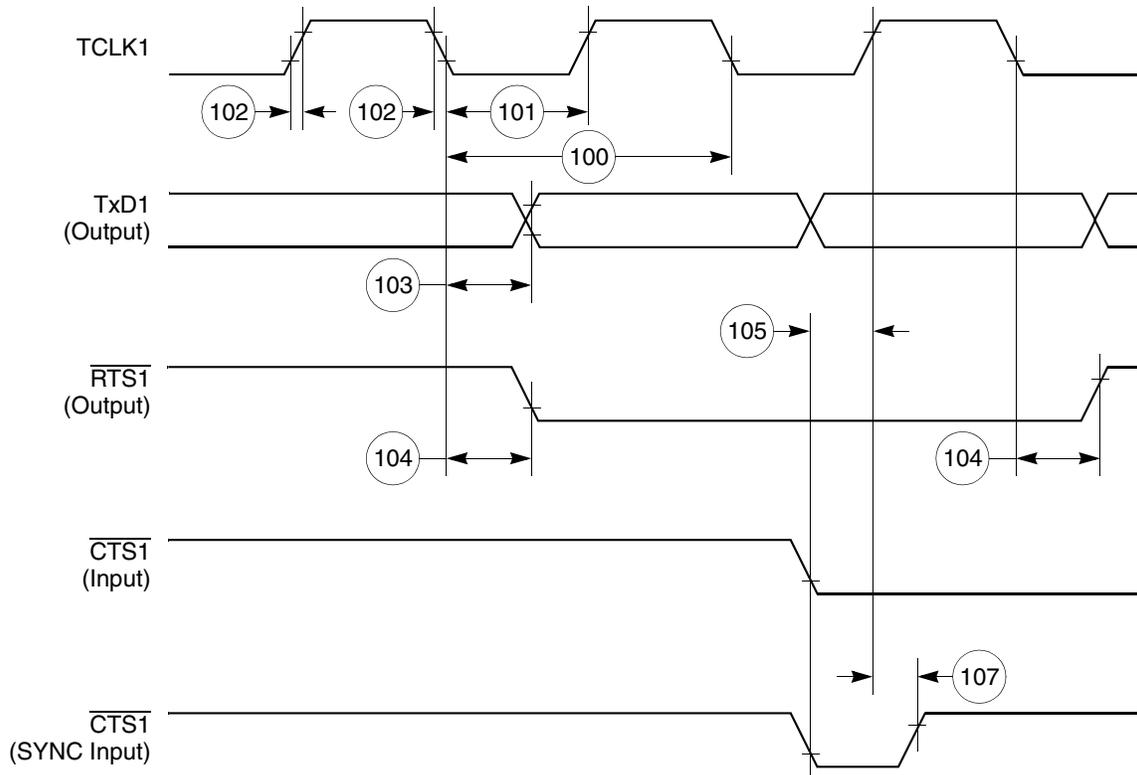


Figure 57. SCC NMSI Transmit Timing Diagram

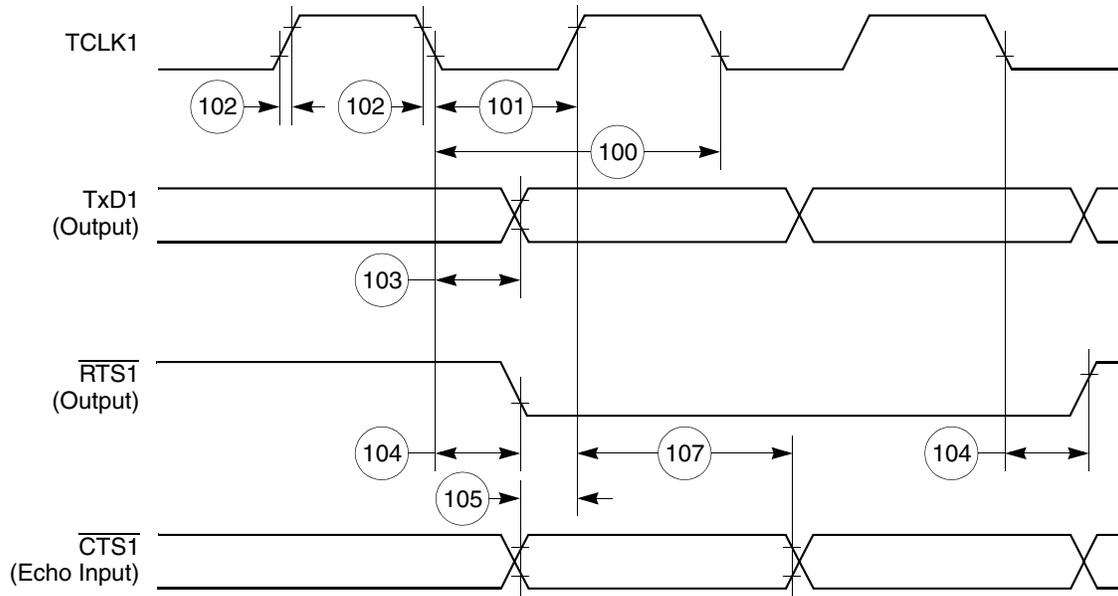


Figure 58. HDLC Bus Timing Diagram

## 11.8 Ethernet Electrical Specifications

Table 22 provides the Ethernet timings as shown in Figure 59 through Figure 63.

Table 22. Ethernet Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
120	CLSN width high	40	—	ns
121	RCLK1 rise/fall time	—	15	ns
122	RCLK1 width low	40	—	ns
123	RCLK1 clock period <sup>1</sup>	80	120	ns
124	RXD1 setup time	20	—	ns
125	RXD1 hold time	5	—	ns
126	RENA active delay (from RCLK1 rising edge of the last data bit)	10	—	ns
127	RENA width low	100	—	ns
128	TCLK1 rise/fall time	—	15	ns
129	TCLK1 width low	40	—	ns
130	TCLK1 clock period <sup>1</sup>	99	101	ns
131	TXD1 active delay (from TCLK1 rising edge)	10	50	ns
132	TXD1 inactive delay (from TCLK1 rising edge)	10	50	ns
133	TENA active delay (from TCLK1 rising edge)	10	50	ns
134	TENA inactive delay (from TCLK1 rising edge)	10	50	ns

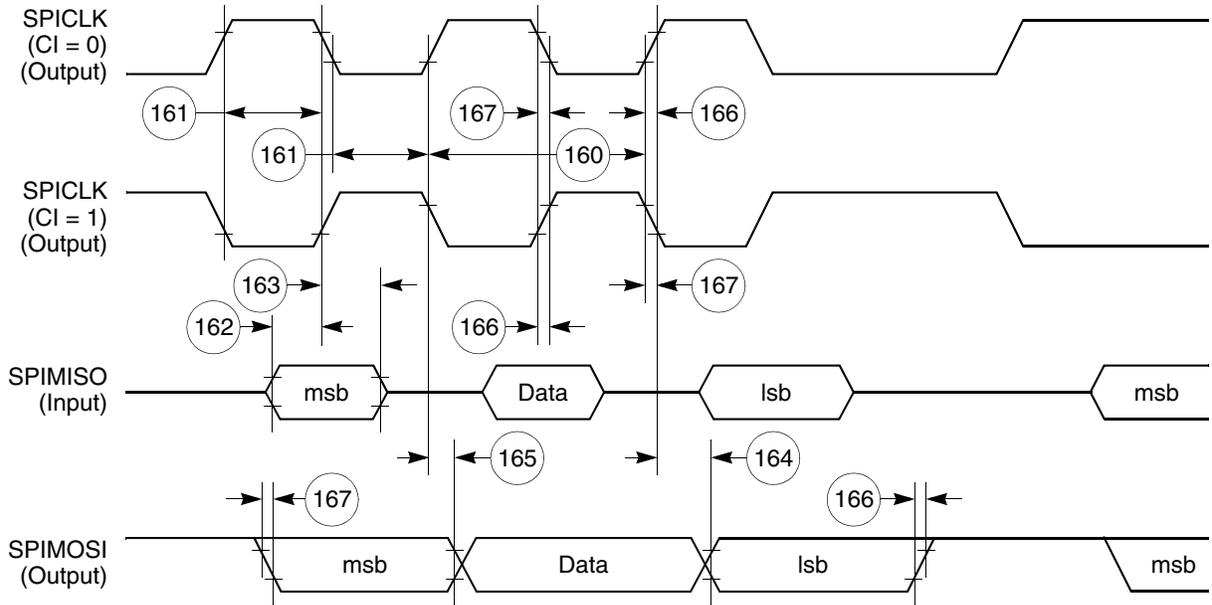


Figure 66. SPI Master (CP = 1) Timing Diagram

## 11.11 SPI Slave AC Electrical Specifications

Table 25 provides the SPI slave timings as shown in Figure 67 and Figure 68.

Table 25. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	$t_{cyc}$
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	$t_{cyc}$
174	Slave sequential transfer delay (does not require deselect)	1	—	$t_{cyc}$
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

## 13 FEC Electrical Characteristics

This section provides the AC electrical specifications for the Fast Ethernet controller (FEC). Note that the timing specifications for the MII signals are independent of system clock frequency (part speed designation). Also, MII signals use TTL signal levels compatible with devices operating at either 5.0 V or 3.3 V.

### 13.1 MII Receive Signal Timing (MII\_RXD[3:0], MII\_RX\_DV, MII\_RX\_ER, MII\_RX\_CLK)

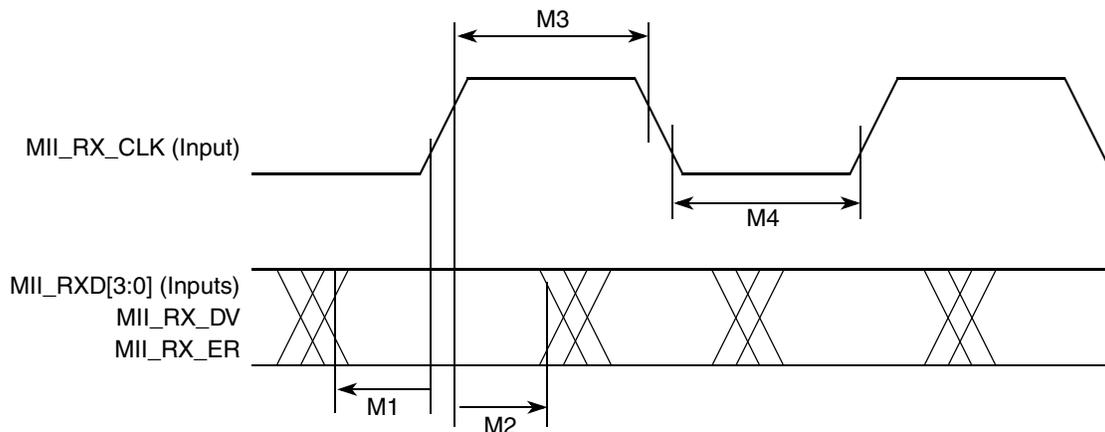
The receiver functions correctly up to a MII\_RX\_CLK maximum frequency of 25 MHz + 1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII\_RX\_CLK frequency – 1%.

Table 29 provides information on the MII receive signal timing.

**Table 29. MII Receive Signal Timing**

Num	Characteristic	Min	Max	Unit
M1	MII_RXD[3:0], MII_RX_DV, MII_RX_ER to MII_RX_CLK setup	5	—	ns
M2	MII_RX_CLK to MII_RXD[3:0], MII_RX_DV, MII_RX_ER hold	5	—	ns
M3	MII_RX_CLK pulse width high	35%	65%	MII_RX_CLK period
M4	MII_RX_CLK pulse width low	35%	65%	MII_RX_CLK period

Figure 72 shows MII receive signal timing.



**Figure 72. MII Receive Signal Timing Diagram**

### 14.3 Mechanical Dimensions of the PBGA Package

Figure 77 shows the mechanical dimensions of the ZP PBGA package.

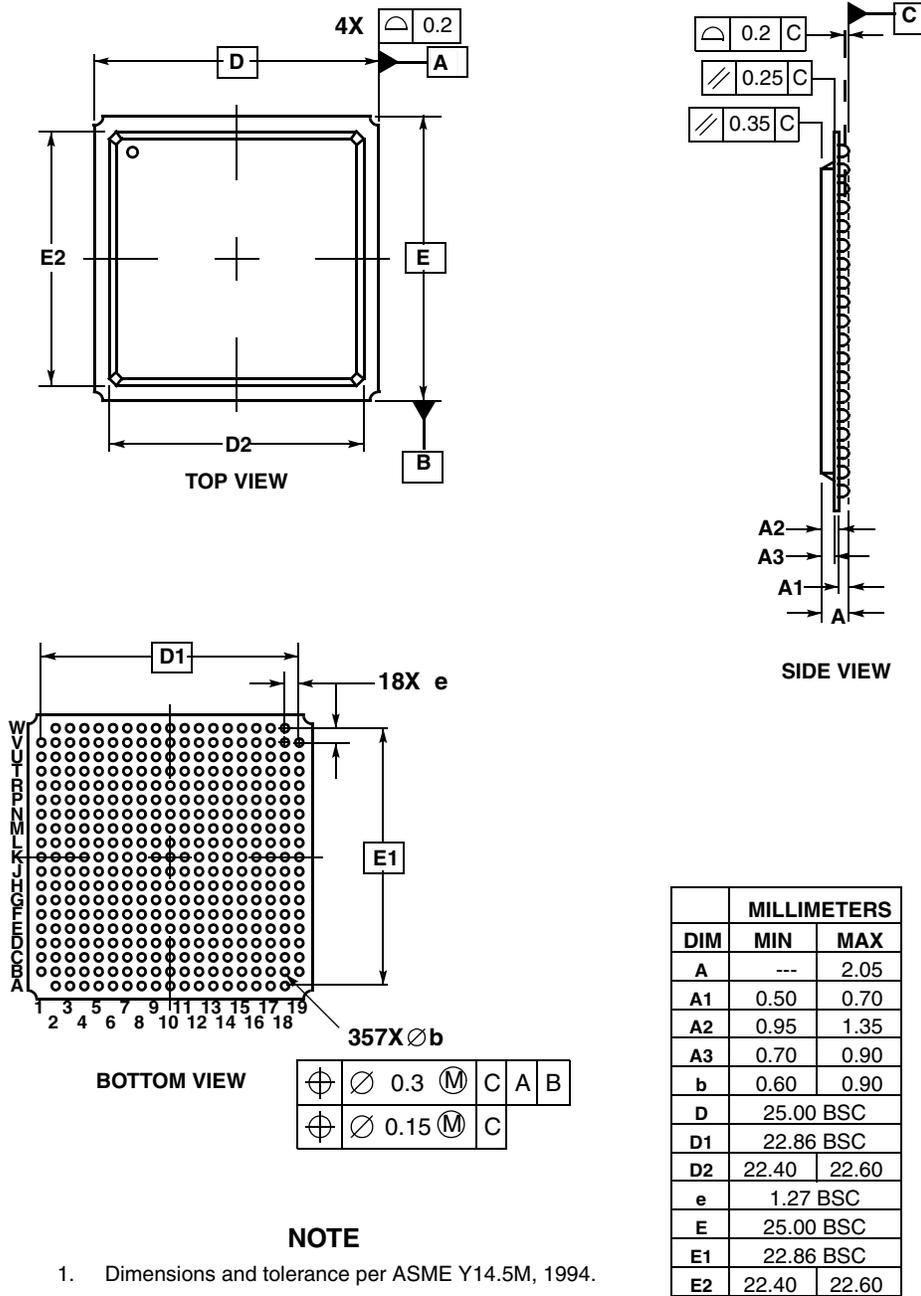


Figure 77. Mechanical Dimensions and Bottom Surface Nomenclature of the ZP PBGA Package

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Rev. 10  
09/2015

