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### Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

### Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

#### Details

Product Status	Active
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	66MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (4)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TJ)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mpc860srvr66d4">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mpc860srvr66d4</a>

## 2 Features

The following list summarizes the key MPC860 features:

- Embedded single-issue, 32-bit core (implementing the Power Architecture technology) with thirty-two 32-bit general-purpose registers (GPRs)
  - The core performs branch prediction with conditional prefetch without conditional execution.
  - 4- or 8-Kbyte data cache and 4- or 16-Kbyte instruction cache (see [Table 1](#))
    - 16-Kbyte instruction caches are four-way, set-associative with 256 sets; 4-Kbyte instruction caches are two-way, set-associative with 128 sets.
    - 8-Kbyte data caches are two-way, set-associative with 256 sets; 4-Kbyte data caches are two-way, set-associative with 128 sets.
    - Cache coherency for both instruction and data caches is maintained on 128-bit (4-word) cache blocks.
    - Caches are physically addressed, implement a least recently used (LRU) replacement algorithm, and are lockable on a cache block basis.
  - MMUs with 32-entry TLB, fully-associative instruction, and data TLBs
  - MMUs support multiple page sizes of 4-, 16-, and 512-Kbytes, and 8-Mbytes; 16 virtual address spaces and 16 protection groups
  - Advanced on-chip-emulation debug mode
- Up to 32-bit data bus (dynamic bus sizing for 8, 16, and 32 bits)
- 32 address lines
- Operates at up to 80 MHz
- Memory controller (eight banks)
  - Contains complete dynamic RAM (DRAM) controller
  - Each bank can be a chip select or  $\overline{\text{RAS}}$  to support a DRAM bank.
  - Up to 15 wait states programmable per memory bank
  - Glueless interface to DRAM, SIMMS, SRAM, EPROM, Flash EPROM, and other memory devices
  - DRAM controller programmable to support most size and speed memory interfaces
  - Four  $\overline{\text{CAS}}$  lines, four  $\overline{\text{WE}}$  lines, and one  $\overline{\text{OE}}$  line
  - Boot chip-select available at reset (options for 8-, 16-, or 32-bit memory)
  - Variable block sizes (32 Kbytes to 256 Mbytes)
  - Selectable write protection
  - On-chip bus arbitration logic
- General-purpose timers
  - Four 16-bit timers or two 32-bit timers
  - Gate mode can enable/disable counting
  - Interrupt can be masked on reference match and event capture.

## 7 Thermal Calculation and Measurement

For the following discussions,  $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$ , where  $P_{I/O}$  is the power dissipation of the I/O drivers.

### 7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature,  $T_J$ , in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

$T_A$  = ambient temperature (°C)

$R_{\theta JA}$  = package junction-to-ambient thermal resistance (°C/W)

$P_D$  = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity  $T_J - T_A$ ) are possible.

### 7.2 Estimation with Junction-to-Case Thermal Resistance

Historically, the thermal resistance has frequently been expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$  = junction-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$  = junction-to-case thermal resistance (°C/W)

$R_{\theta CA}$  = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$  is device related and cannot be influenced by the user. The user adjusts the thermal environment to affect the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For instance, the user can change the airflow around the device, add a heat sink, change the mounting arrangement on the printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device. This thermal model is most useful for ceramic packages with heat sinks where some 90% of the heat flows through the case and the heat sink to the ambient environment. For most packages, a better model is required.

### 7.3 Estimation with Junction-to-Board Thermal Resistance

A simple package thermal model which has demonstrated reasonable accuracy (about 20%) is a two-resistor model consisting of a junction-to-board and a junction-to-case thermal resistance. The junction-to-case thermal resistance covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed-circuit board. It has been observed that the thermal performance of most plastic packages, especially PBGA packages, is strongly dependent on the board temperature; see [Figure 2](#).

where:

$\Psi_{JT}$  = thermal characterization parameter

$T_T$  = thermocouple temperature on top of package

$P_D$  = power dissipation in package

The thermal characterization parameter is measured per JEDEC JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

## 7.6 References

Semiconductor Equipment and Materials International (415) 964-5111  
 805 East Middlefield Rd.  
 Mountain View, CA 94043

MIL-SPEC and EIA/JESD (JEDEC) Specifications 800-854-7179 or  
 (Available from Global Engineering Documents) 303-397-7956

JEDEC Specifications <http://www.jedec.org>

1. C.E. Triplett and B. Joiner, “An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module,” Proceedings of SemiTherm, San Diego, 1998, pp. 47–54.
2. B. Joiner and V. Adams, “Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling,” Proceedings of SemiTherm, San Diego, 1999, pp. 212–220.

## 8 Layout Practices

Each  $V_{DD}$  pin on the MPC860 should be provided with a low-impedance path to the board’s supply. Each GND pin should likewise be provided with a low-impedance path to ground. The power supply pins drive distinct groups of logic on the chip. The  $V_{DD}$  power supply should be bypassed to ground using at least four 0.1  $\mu$ F-bypass capacitors located as close as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip  $V_{DD}$  and GND should be kept to less than half an inch per capacitor lead. A four-layer board employing two inner layers as  $V_{CC}$  and GND planes is recommended.

All output pins on the MPC860 have fast rise and fall times. Printed circuit (PC) trace interconnection length should be minimized in order to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data buses. Maximum PC trace lengths of 6 inches are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the  $V_{CC}$  and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins.

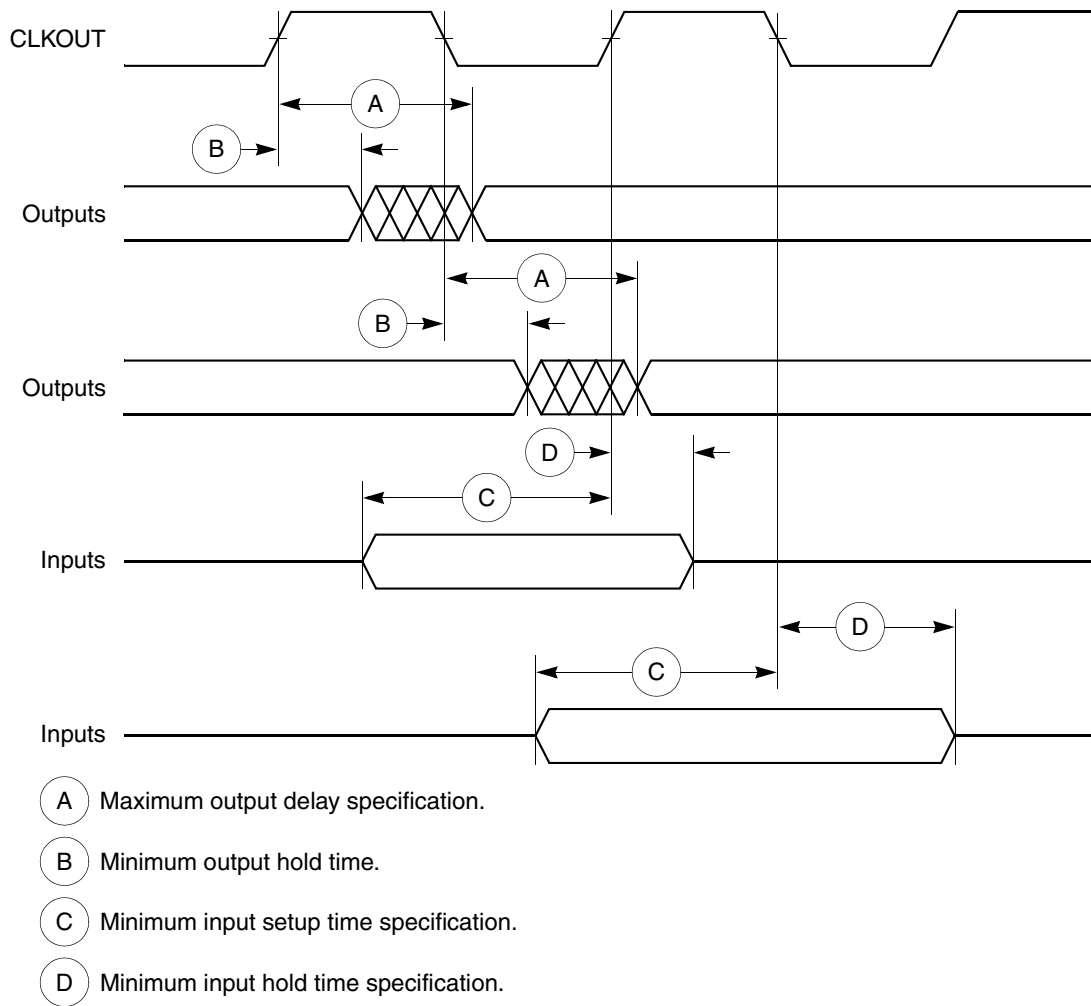
Table 7. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B9	CLKOUT to A(0:31), BADDR(28:30), RD/WR, BURST, D(0:31), DP(0:3), TSIZ(0:1), REG, RSV, AT(0:3), PTR High-Z	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns
B11	CLKOUT to $\overline{TS}$ , $\overline{BB}$ assertion	7.58	13.58	6.25	12.25	5.00	11.00	3.80	11.29	ns
B11a	CLKOUT to $\overline{TA}$ , $\overline{BI}$ assertion (when driven by the memory controller or PCMCIA interface)	2.50	9.25	2.50	9.25	2.50	9.25	2.50	9.75	ns
B12	CLKOUT to $\overline{TS}$ , $\overline{BB}$ negation	7.58	14.33	6.25	13.00	5.00	11.75	3.80	8.54	ns
B12a	CLKOUT to $\overline{TA}$ , $\overline{BI}$ negation (when driven by the memory controller or PCMCIA interface)	2.50	11.00	2.50	11.00	2.50	11.00	2.50	9.00	ns
B13	CLKOUT to $\overline{TS}$ , $\overline{BB}$ High-Z	7.58	21.58	6.25	20.25	5.00	19.00	3.80	14.04	ns
B13a	CLKOUT to $\overline{TA}$ , $\overline{BI}$ High-Z (when driven by the memory controller or PCMCIA interface)	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B14	CLKOUT to $\overline{TEA}$ assertion	2.50	10.00	2.50	10.00	2.50	10.00	2.50	9.00	ns
B15	CLKOUT to $\overline{TEA}$ High-Z	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B16	$\overline{TA}$ , $\overline{BI}$ valid to CLKOUT (setup time)	9.75	—	9.75	—	9.75	—	6.00	—	ns
B16a	$\overline{TEA}$ , $\overline{KR}$ , $\overline{RETRY}$ , $\overline{CR}$ valid to CLKOUT (setup time)	10.00	—	10.00	—	10.00	—	4.50	—	ns
B16b	$\overline{BB}$ , $\overline{BG}$ , $\overline{BR}$ , valid to CLKOUT (setup time) <sup>5</sup>	8.50	—	8.50	—	8.50	—	4.00	—	ns
B17	CLKOUT to $\overline{TA}$ , $\overline{TEA}$ , $\overline{BI}$ , $\overline{BB}$ , $\overline{BG}$ , $\overline{BR}$ valid (hold time)	1.00	—	1.00	—	1.00	—	2.00	—	ns
B17a	CLKOUT to $\overline{KR}$ , $\overline{RETRY}$ , $\overline{CR}$ valid (hold time)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B18	D(0:31), DP(0:3) valid to CLKOUT rising edge (setup time) <sup>6</sup>	6.00	—	6.00	—	6.00	—	6.00	—	ns
B19	CLKOUT rising edge to D(0:31), DP(0:3) valid (hold time) <sup>6</sup>	1.00	—	1.00	—	1.00	—	2.00	—	ns
B20	D(0:31), DP(0:3) valid to CLKOUT falling edge (setup time) <sup>7</sup>	4.00	—	4.00	—	4.00	—	4.00	—	ns
B21	CLKOUT falling edge to D(0:31), DP(0:3) valid (hold time) <sup>7</sup>	2.00	—	2.00	—	2.00	—	2.00	—	ns
B22	CLKOUT rising edge to $\overline{CS}$ asserted GPCM ACS = 00	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns
B22a	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 10, TRLX = 0	—	8.00	—	8.00	—	8.00	—	8.00	ns
B22b	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 11, TRLX = 0, EBDF = 0	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.54	ns
B22c	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 11, TRLX = 0, EBDF = 1	10.86	17.99	8.88	16.00	7.00	14.13	5.18	12.31	ns

Table 7. Bus Operation Timings (continued)

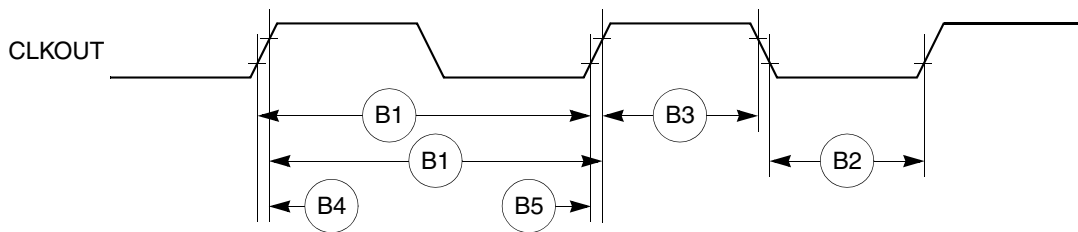
Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B31a	CLKOUT falling edge to $\overline{CS}$ valid—as requested by control bit CST1 in the corresponding word in UPM	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.54	ns
B31b	CLKOUT rising edge to $\overline{CS}$ valid—as requested by control bit CST2 in the corresponding word in UPM	1.50	8.00	1.50	8.00	1.50	8.00	1.50	8.00	ns
B31c	CLKOUT rising edge to $\overline{CS}$ valid—as requested by control bit CST3 in the corresponding word in UPM	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns
B31d	CLKOUT falling edge to $\overline{CS}$ valid—as requested by control bit CST1 in the corresponding word in UPM, EBDF = 1	13.26	17.99	11.28	16.00	9.40	14.13	7.58	12.31	ns
B32	CLKOUT falling edge to $\overline{BS}$ valid—as requested by control bit BST4 in the corresponding word in UPM	1.50	6.00	1.50	6.00	1.50	6.00	1.50	6.00	ns
B32a	CLKOUT falling edge to $\overline{BS}$ valid—as requested by control bit BST1 in the corresponding word in UPM, EBDF = 0	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.54	ns
B32b	CLKOUT rising edge to $\overline{BS}$ valid—as requested by control bit BST2 in the corresponding word in UPM	1.50	8.00	1.50	8.00	1.50	8.00	1.50	8.00	ns
B32c	CLKOUT rising edge to $\overline{BS}$ valid—as requested by control bit BST3 in the corresponding word in UPM	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.54	ns
B32d	CLKOUT falling edge to $\overline{BS}$ valid—as requested by control bit BST1 in the corresponding word in UPM, EBDF = 1	13.26	17.99	11.28	16.00	9.40	14.13	7.58	12.31	ns
B33	CLKOUT falling edge to $\overline{GPL}$ valid—as requested by control bit GxT4 in the corresponding word in UPM	1.50	6.00	1.50	6.00	1.50	6.00	1.50	6.00	ns
B33a	CLKOUT rising edge to $\overline{GPL}$ valid—as requested by control bit GxT3 in the corresponding word in UPM	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.54	ns
B34	A(0:31), BADDR(28:30), and D(0:31) to $\overline{CS}$ valid—as requested by control bit CST4 in the corresponding word in UPM	5.58	—	4.25	—	3.00	—	1.79	—	ns
B34a	A(0:31), BADDR(28:30), and D(0:31) to $\overline{CS}$ valid—as requested by control bit CST1 in the corresponding word in UPM	13.15	—	10.50	—	8.00	—	5.58	—	ns
B34b	A(0:31), BADDR(28:30), and D(0:31) to $\overline{CS}$ valid—as requested by control bit CST2 in the corresponding word in UPM	20.73	—	16.75	—	13.00	—	9.36	—	ns

Figure 3 is the control timing diagram.



**Figure 3. Control Timing**

Figure 4 provides the timing for the external clock.



**Figure 4. External Clock Timing**

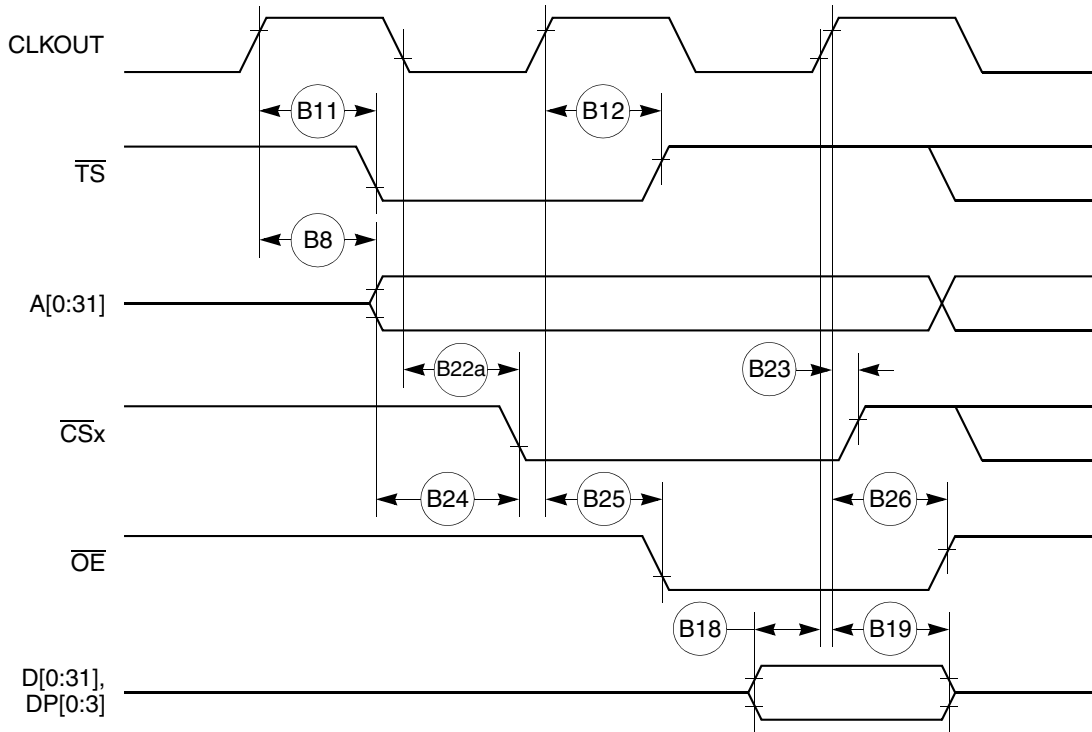


Figure 11. External Bus Read Timing (GPCM Controlled—TRLX = 0, ACS = 10)

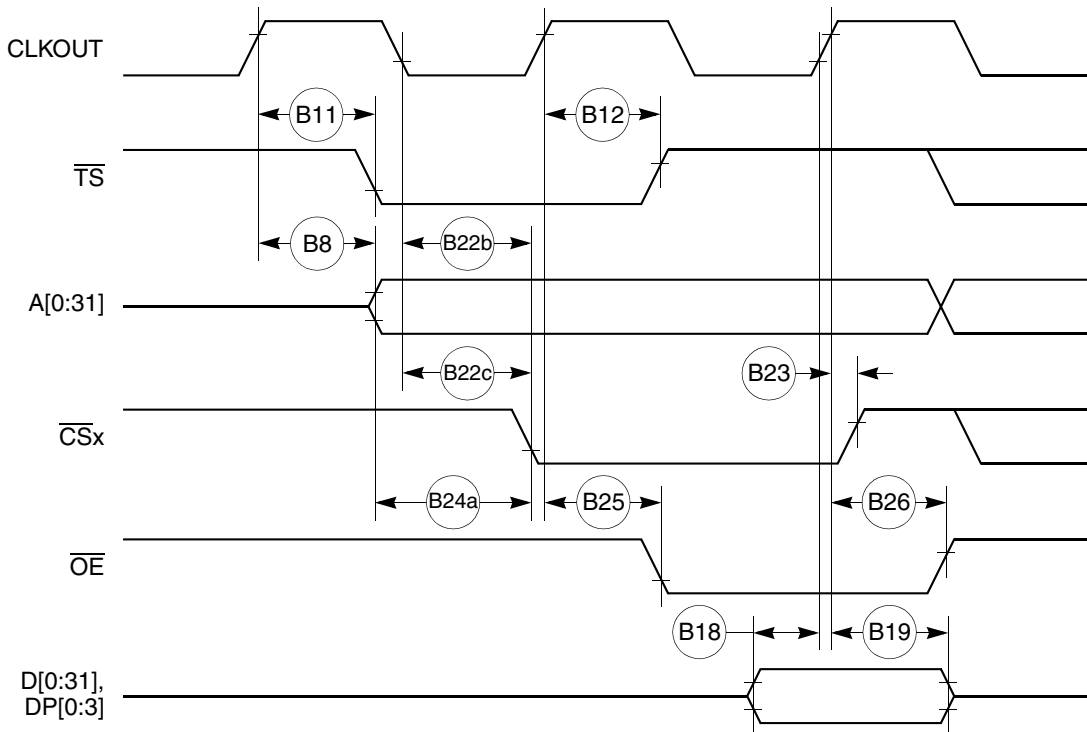


Figure 12. External Bus Read Timing (GPCM Controlled—TRLX = 0, ACS = 11)



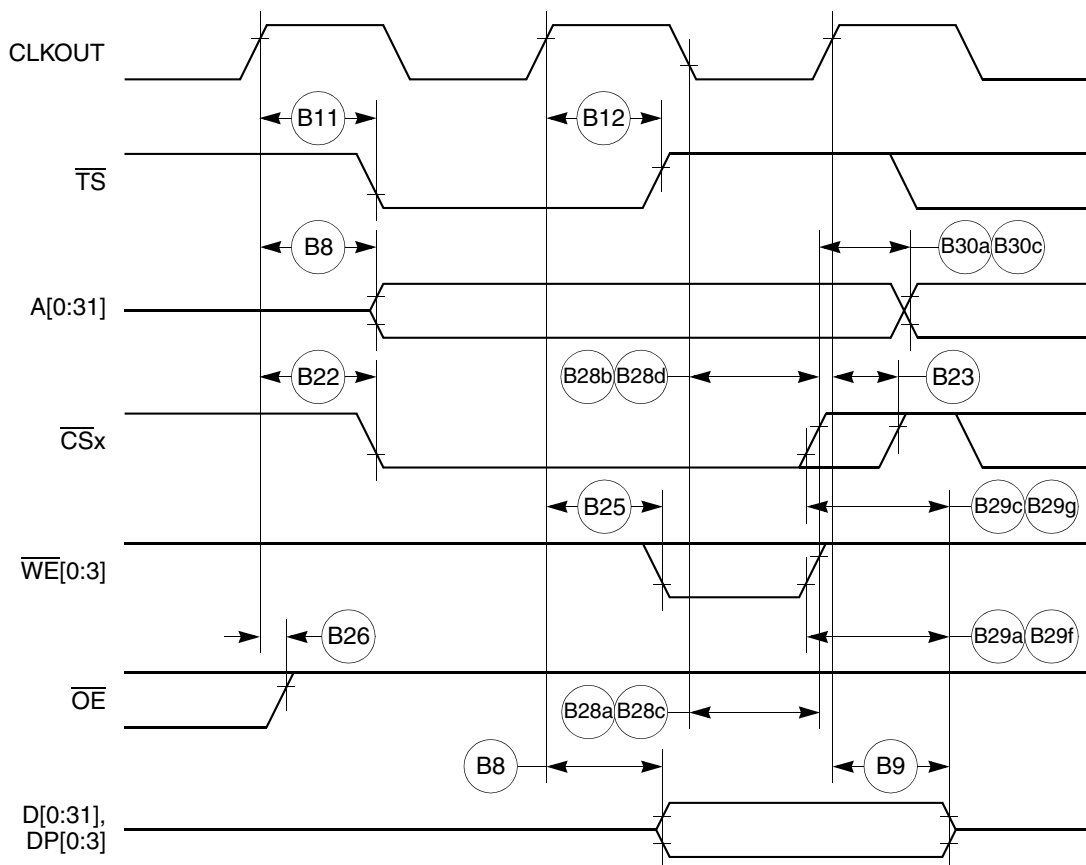
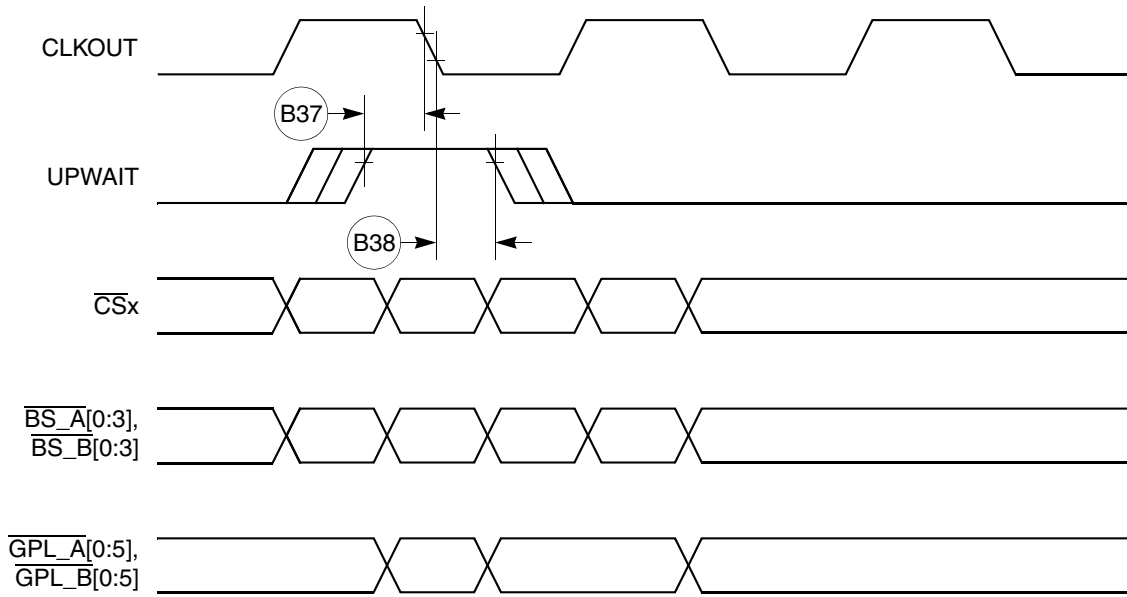


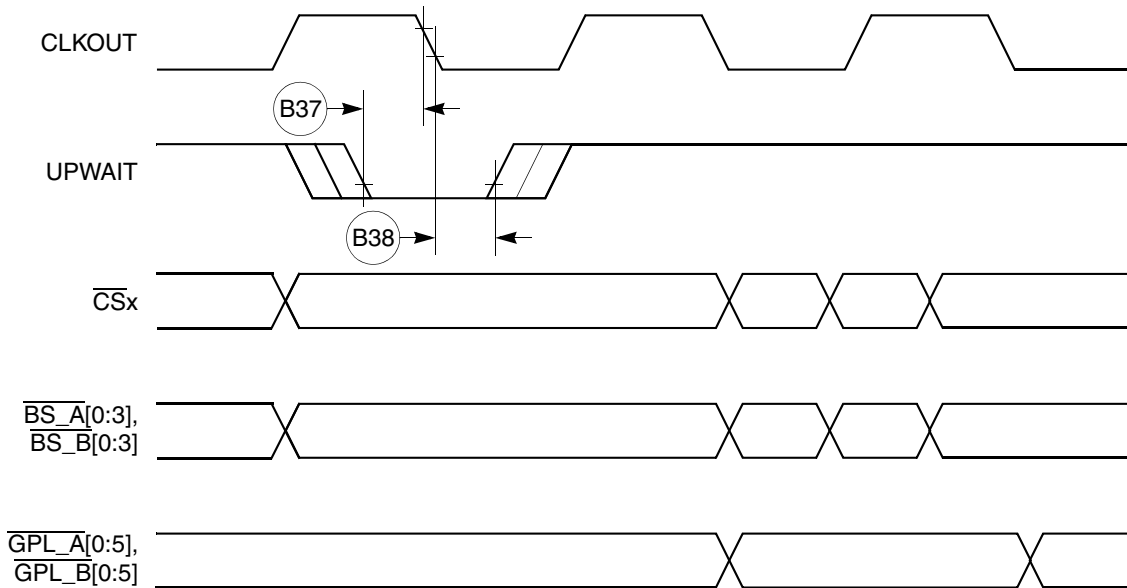
Figure 15. External Bus Write Timing (GPCM Controlled—TRLX = 0 or 1, CSNT = 1)

Figure 18 provides the timing for the asynchronous asserted UPWAIT signal controlled by the UPM.



**Figure 18. Asynchronous UPWAIT Asserted Detection in UPM Handled Cycles Timing**

Figure 19 provides the timing for the asynchronous negated UPWAIT signal controlled by the UPM.



**Figure 19. Asynchronous UPWAIT Negated Detection in UPM Handled Cycles Timing**

Table 9 shows the PCMCIA timing for the MPC860.

**Table 9. PCMCIA Timing**

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P44	A(0:31), $\overline{REG}$ valid to PCMCIA Strobe asserted <sup>1</sup>	20.73	—	16.75	—	13.00	—	9.36	—	ns
P45	A(0:31), $\overline{REG}$ valid to ALE negation <sup>1</sup>	28.30	—	23.00	—	18.00	—	13.15	—	ns
P46	CLKOUT to $\overline{REG}$ valid	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P47	CLKOUT to $\overline{REG}$ invalid	8.58	—	7.25	—	6.00	—	4.84	—	ns
P48	CLKOUT to $\overline{CE1}$ , $\overline{CE2}$ asserted	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P49	CLKOUT to $\overline{CE1}$ , $\overline{CE2}$ negated	7.58	15.58	6.25	14.25	5.00	13.00	3.79	11.84	ns
P50	CLKOUT to $\overline{PCOE}$ , $\overline{IORD}$ , $\overline{PCWE}$ , $\overline{IOWR}$ assert time	—	11.00		11.00	—	11.00	—	11.00	ns
P51	CLKOUT to $\overline{PCOE}$ , $\overline{IORD}$ , $\overline{PCWE}$ , $\overline{IOWR}$ negate time	2.00	11.00	2.00	11.00	2.00	11.00	2.00	11.00	ns
P52	CLKOUT to ALE assert time	7.58	15.58	6.25	14.25	5.00	13.00	3.79	10.04	ns
P53	CLKOUT to ALE negate time	—	15.58		14.25	—	13.00	—	11.84	ns
P54	$\overline{PCWE}$ , $\overline{IOWR}$ negated to D(0:31) invalid <sup>1</sup>	5.58	—	4.25	—	3.00	—	1.79	—	ns
P55	$\overline{WAITA}$ and $\overline{WAITB}$ valid to CLKOUT rising edge <sup>1</sup>	8.00	—	8.00	—	8.00	—	8.00	—	ns
P56	CLKOUT rising edge to $\overline{WAITA}$ and $\overline{WAITB}$ invalid <sup>1</sup>	2.00	—	2.00	—	2.00	—	2.00	—	ns

<sup>1</sup> PSST = 1. Otherwise add PSST times cycle time.  
 PSHT = 0. Otherwise add PSHT times cycle time.

These synchronous timings define when the  $\overline{WAITx}$  signals are detected in order to freeze (or relieve) the PCMCIA current cycle. The  $\overline{WAITx}$  assertion will be effective only if it is detected 2 cycles before the PSL timer expiration. See Chapter 16, “PCMCIA Interface,” in the *MPC860 PowerQUICC™ Family User’s Manual*.

Figure 34 provides the reset timing for the debug port configuration.

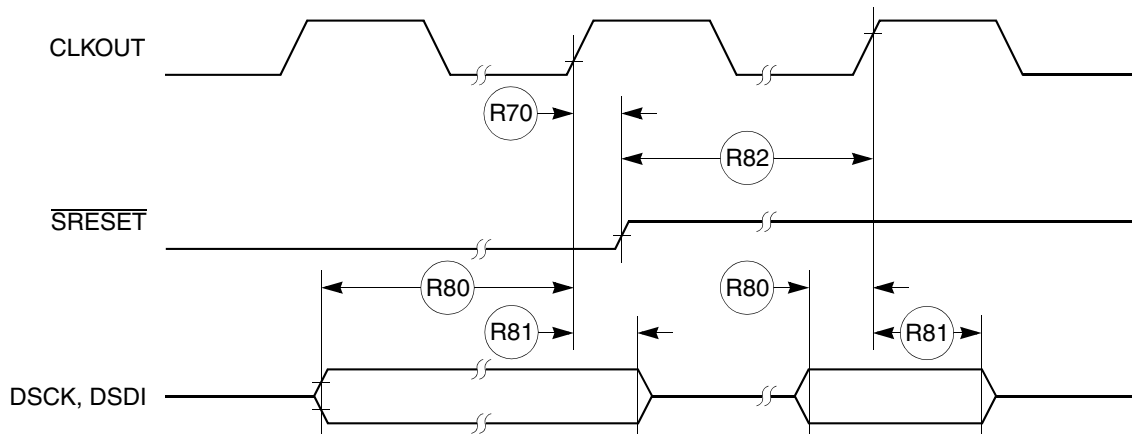


Figure 34. Reset Timing—Debug Port Configuration

## 10 IEEE 1149.1 Electrical Specifications

Table 13 provides the JTAG timings for the MPC860 shown in Figure 35 through Figure 38.

Table 13. JTAG Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
J82	TCK cycle time	100.00	—	ns
J83	TCK clock pulse width measured at 1.5 V	40.00	—	ns
J84	TCK rise and fall times	0.00	10.00	ns
J85	TMS, TDI data setup time	5.00	—	ns
J86	TMS, TDI data hold time	25.00	—	ns
J87	TCK low to TDO data valid	—	27.00	ns
J88	TCK low to TDO data invalid	0.00	—	ns
J89	TCK low to TDO high impedance	—	20.00	ns
J90	$\overline{\text{TRST}}$ assert time	100.00	—	ns
J91	$\overline{\text{TRST}}$ setup time to TCK low	40.00	—	ns
J92	TCK falling edge to output valid	—	50.00	ns
J93	TCK falling edge to output valid out of high impedance	—	50.00	ns
J94	TCK falling edge to output high impedance	—	50.00	ns
J95	Boundary scan input valid to TCK rising edge	50.00	—	ns
J96	TCK rising edge to boundary scan input invalid	50.00	—	ns

# 11 CPM Electrical Characteristics

This section provides the AC and DC electrical specifications for the communications processor module (CPM) of the MPC860.

## 11.1 PIP/PIO AC Electrical Specifications

Table 14 provides the PIP/PIO AC timings as shown in Figure 39 through Figure 43.

Table 14. PIP/PIO Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
21	Data-in setup time to STBI low	0	—	ns
22	Data-in hold time to STBI high	$2.5 - t_3^1$	—	CLK
23	STBI pulse width	1.5	—	CLK
24	STBO pulse width	1 CLK – 5 ns	—	ns
25	Data-out setup time to STBO low	2	—	CLK
26	Data-out hold time from STBO high	5	—	CLK
27	STBI low to STBO low (Rx interlock)	—	2	CLK
28	STBI low to STBO high (Tx interlock)	2	—	CLK
29	Data-in setup time to clock high	15	—	ns
30	Data-in hold time from clock high	7.5	—	ns
31	Clock low to data-out valid (CPU writes data, control, or direction)	—	25	ns

<sup>1</sup>  $t_3$  = Specification 23.

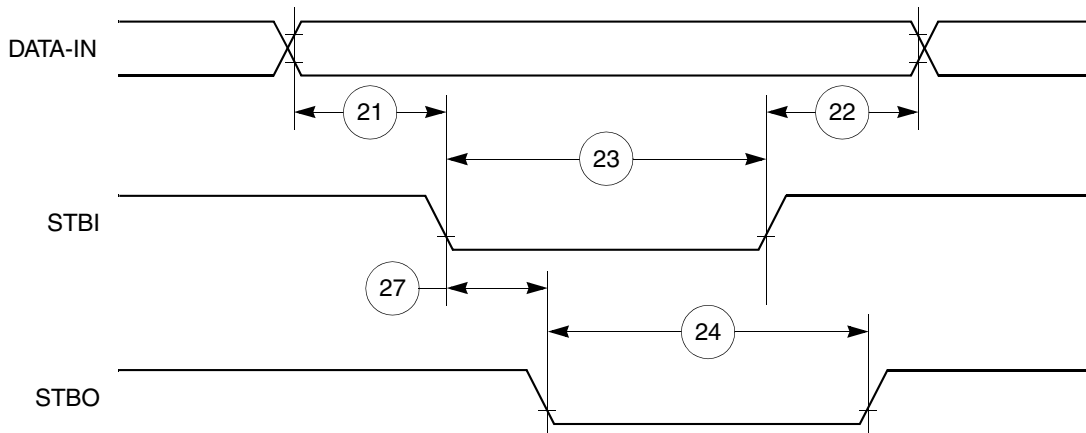


Figure 39. PIP Rx (Interlock Mode) Timing Diagram

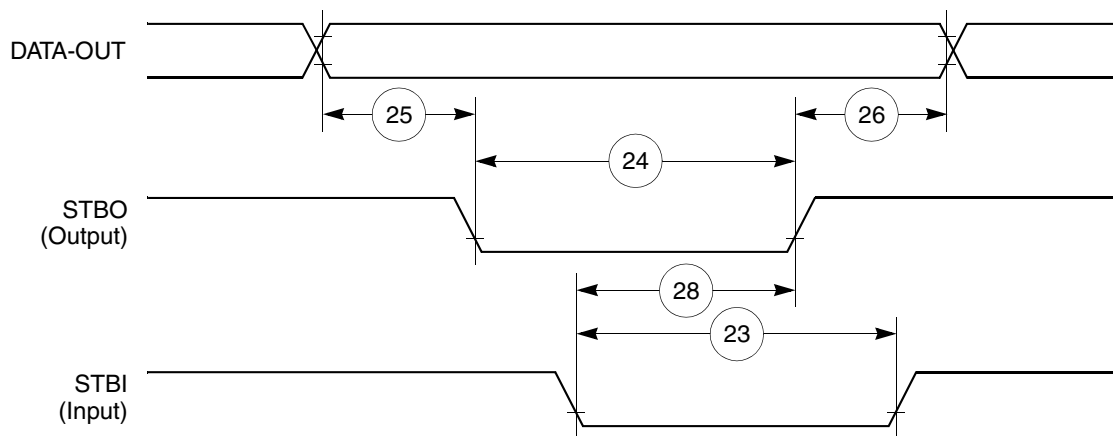


Figure 40. PIP Tx (Interlock Mode) Timing Diagram

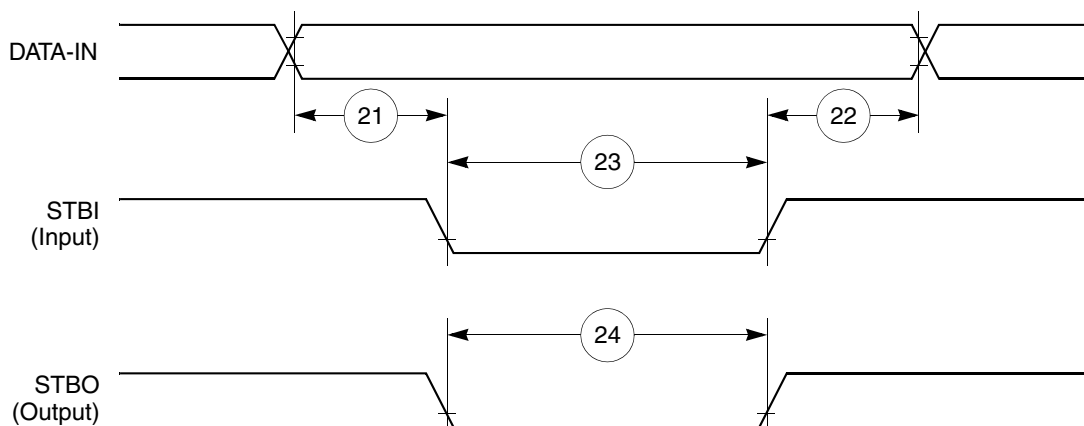


Figure 41. PIP Rx (Pulse Mode) Timing Diagram

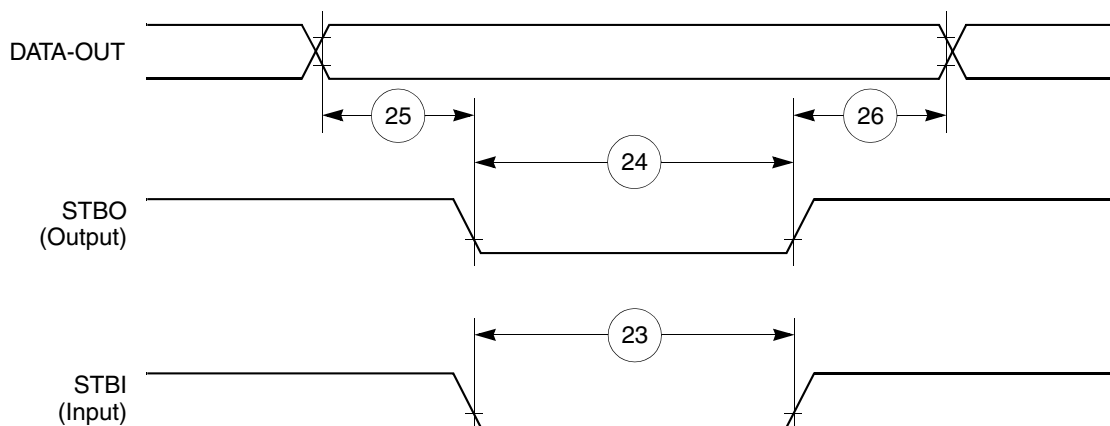


Figure 42. PIP TX (Pulse Mode) Timing Diagram

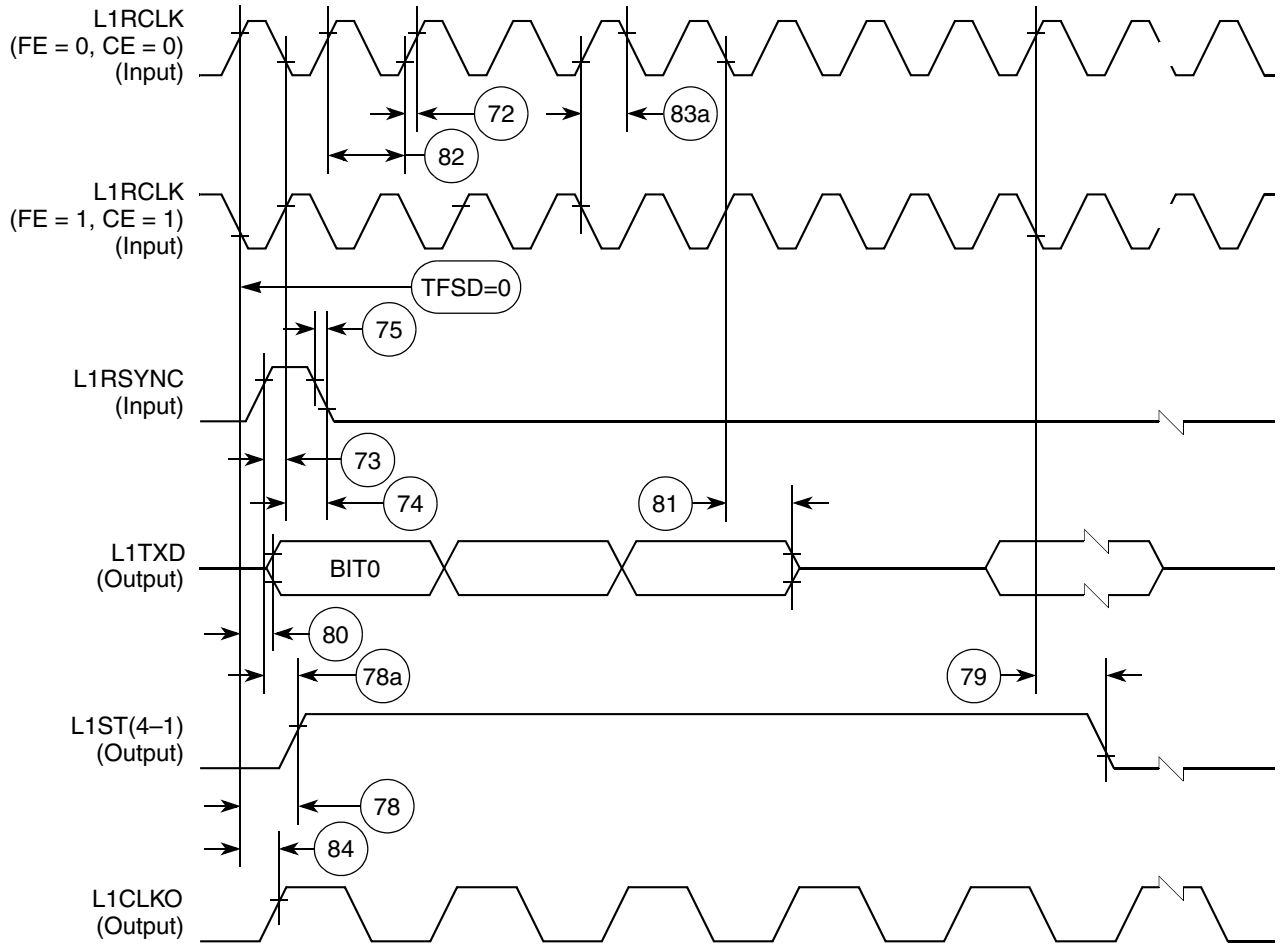


Figure 54. SI Transmit Timing with Double Speed Clocking (DSC = 1)

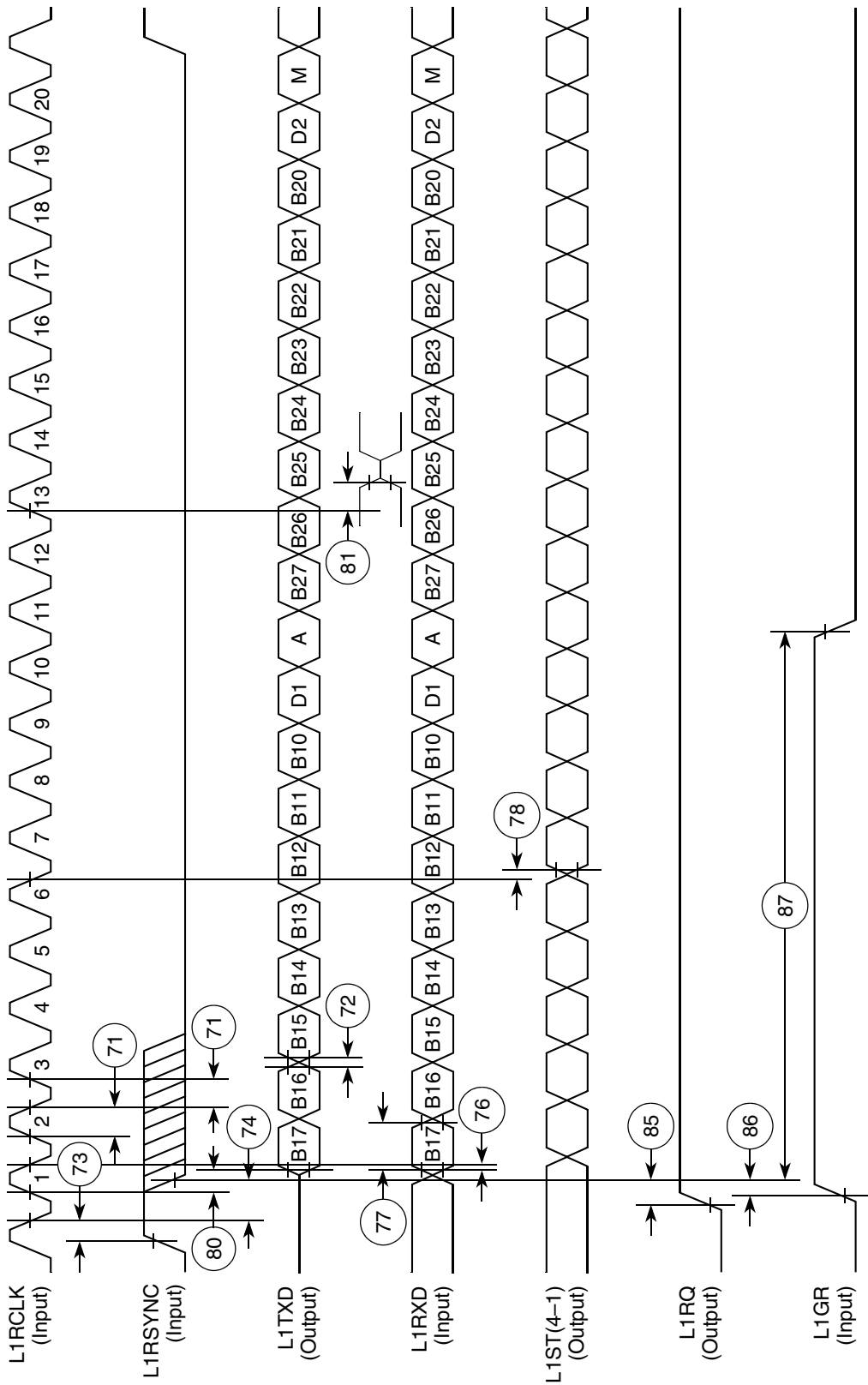


Figure 55. IDL Timing



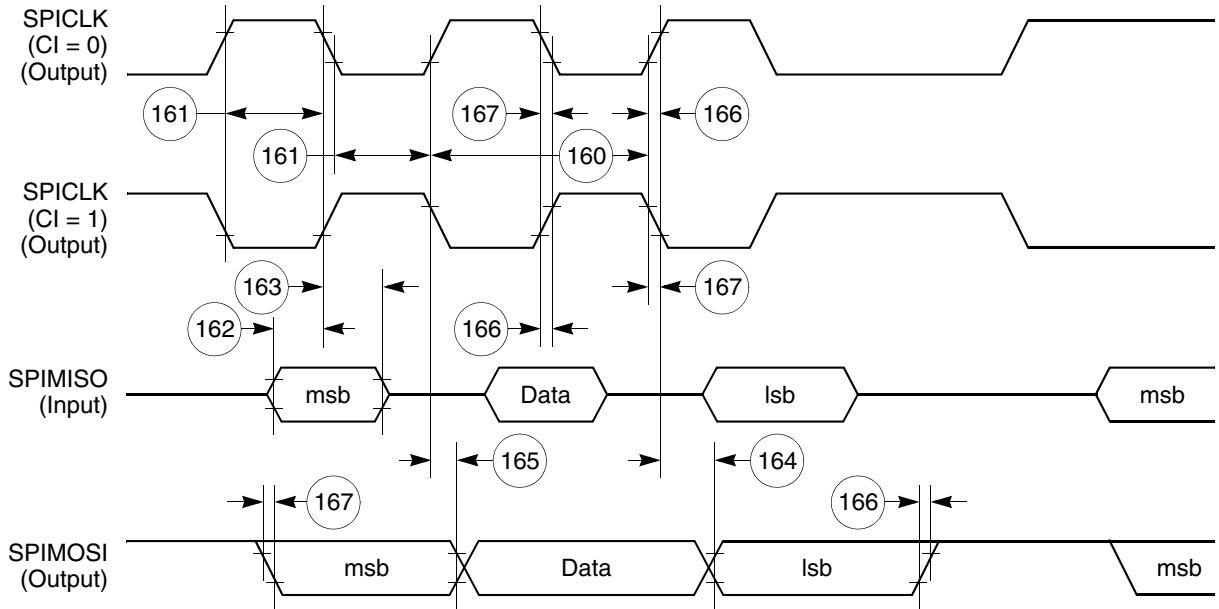


Figure 66. SPI Master (CP = 1) Timing Diagram

## 11.11 SPI Slave AC Electrical Specifications

Table 25 provides the SPI slave timings as shown in Figure 67 and Figure 68.

Table 25. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	$t_{cyc}$
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	$t_{cyc}$
174	Slave sequential transfer delay (does not require deselect)	1	—	$t_{cyc}$
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

## 13 FEC Electrical Characteristics

This section provides the AC electrical specifications for the Fast Ethernet controller (FEC). Note that the timing specifications for the MII signals are independent of system clock frequency (part speed designation). Also, MII signals use TTL signal levels compatible with devices operating at either 5.0 V or 3.3 V.

### 13.1 MII Receive Signal Timing (MII\_RXD[3:0], MII\_RX\_DV, MII\_RX\_ER, MII\_RX\_CLK)

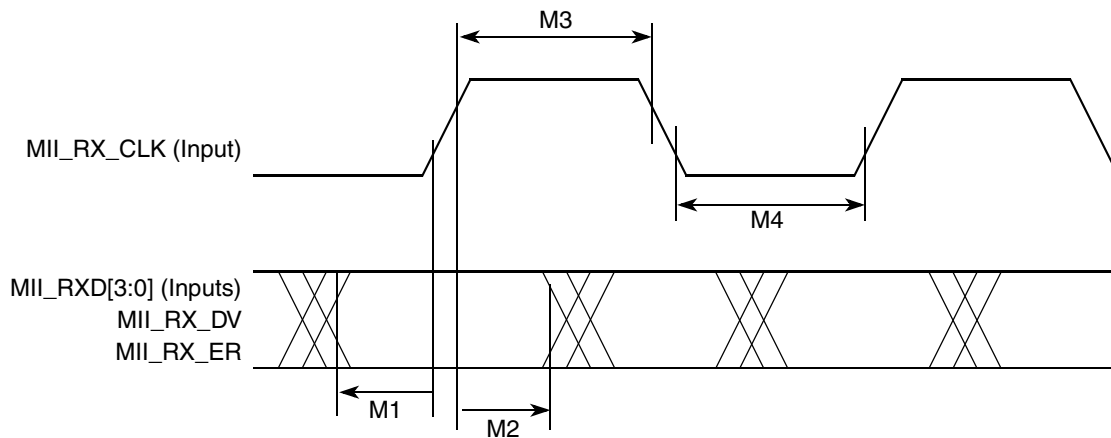
The receiver functions correctly up to a MII\_RX\_CLK maximum frequency of 25 MHz + 1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII\_RX\_CLK frequency – 1%.

Table 29 provides information on the MII receive signal timing.

**Table 29. MII Receive Signal Timing**

Num	Characteristic	Min	Max	Unit
M1	MII_RXD[3:0], MII_RX_DV, MII_RX_ER to MII_RX_CLK setup	5	—	ns
M2	MII_RX_CLK to MII_RXD[3:0], MII_RX_DV, MII_RX_ER hold	5	—	ns
M3	MII_RX_CLK pulse width high	35%	65%	MII_RX_CLK period
M4	MII_RX_CLK pulse width low	35%	65%	MII_RX_CLK period

Figure 72 shows MII receive signal timing.



**Figure 72. MII Receive Signal Timing Diagram**

Figure 75 shows the MII serial management channel timing diagram.

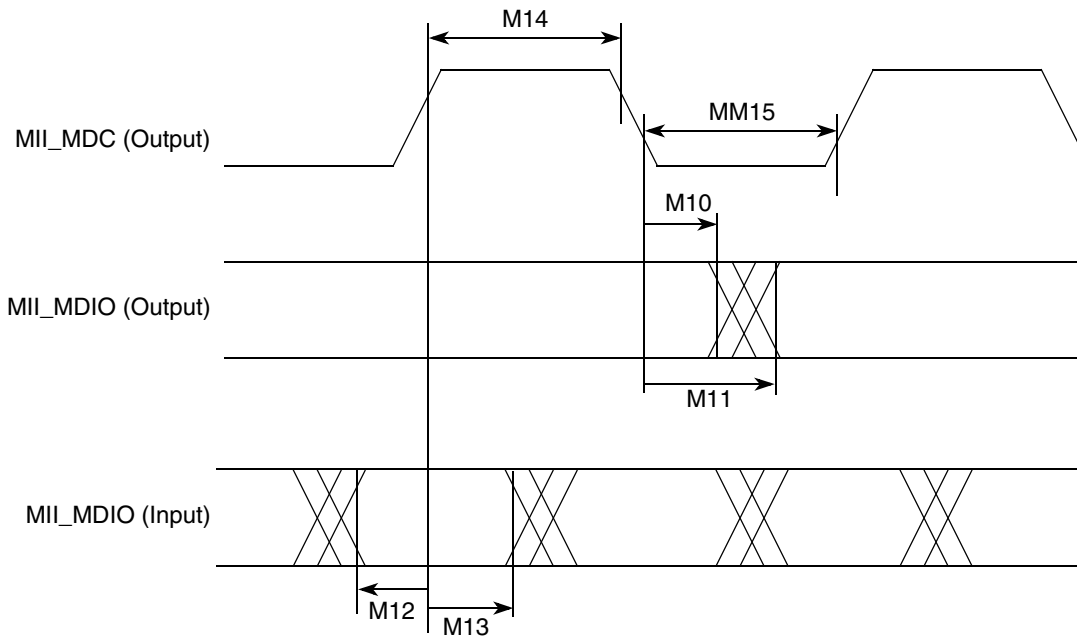


Figure 75. MII Serial Management Channel Timing Diagram

## 14 Mechanical Data and Ordering Information

### 14.1 Ordering Information

Table 33 provides information on the MPC860 Revision D.4 derivative devices.

Table 33. MPC860 Family Revision D.4 Derivatives

Device	Number of SCCs <sup>1</sup>	Ethernet Support <sup>2</sup> (Mbps)	Multichannel HDLC Support	ATM Support
MPC855T	1	10/100	Yes	Yes
MPC860DE	2	10	N/A	N/A
MPC860DT		10/100	Yes	Yes
MPC860DP		10/100	Yes	Yes
MPC860EN	4	10	N/A	N/A
MPC860SR		10	Yes	Yes
MPC860T		10/100	Yes	Yes
MPC860P		10/100	Yes	Yes

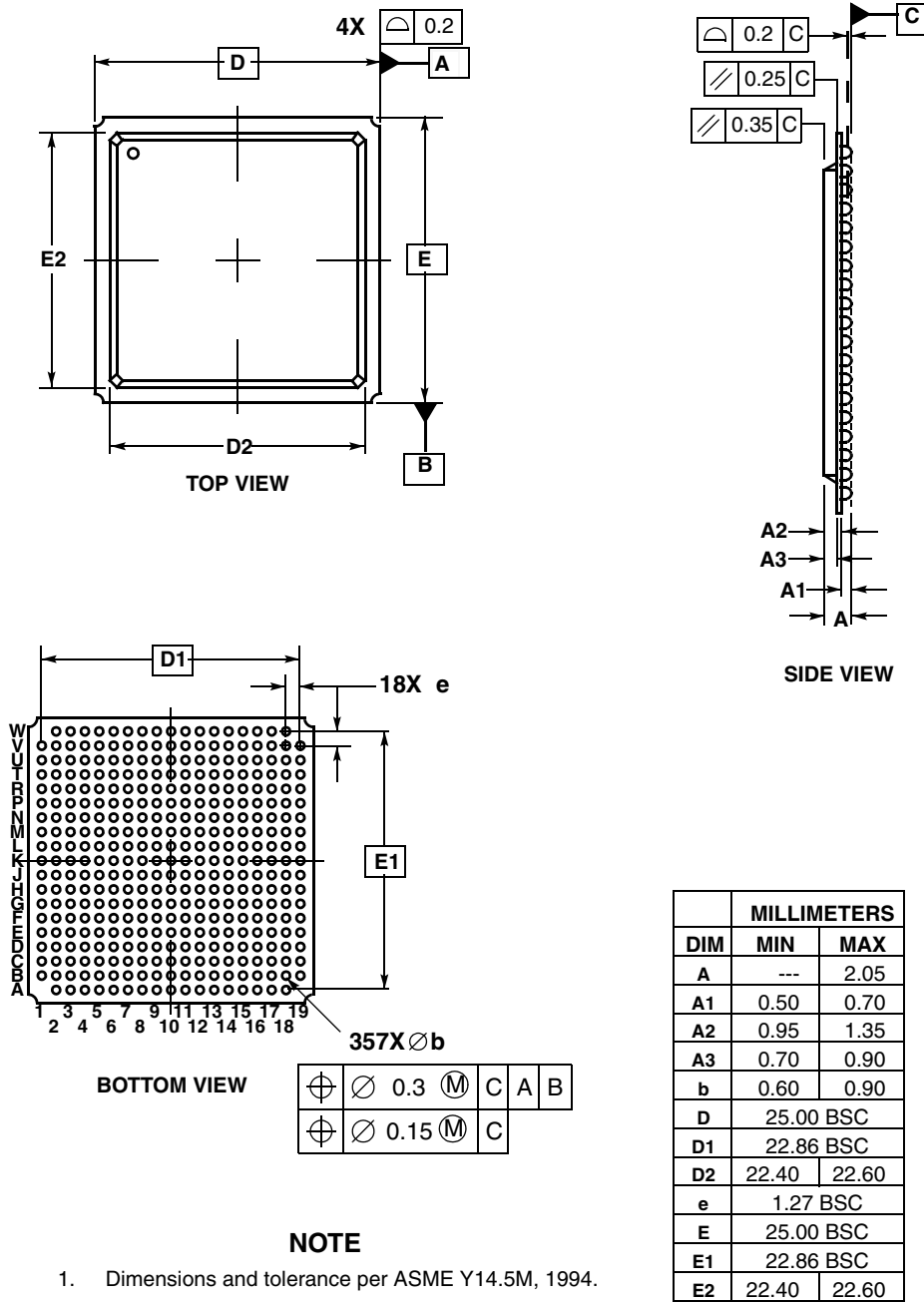
<sup>1</sup> Serial communications controller (SCC)

<sup>2</sup> Up to 4 channels at 40 MHz or 2 channels at 25 MHz



### 14.3 Mechanical Dimensions of the PBGA Package

Figure 77 shows the mechanical dimensions of the ZP PBGA package.



**NOTE**

1. Dimensions and tolerance per ASME Y14.5M, 1994.
2. Dimensions in millimeters.
3. Dimension b is the maximum solder ball diameter measured parallel to data C.

**Figure 77. Mechanical Dimensions and Bottom Surface Nomenclature of the ZP PBGA Package**