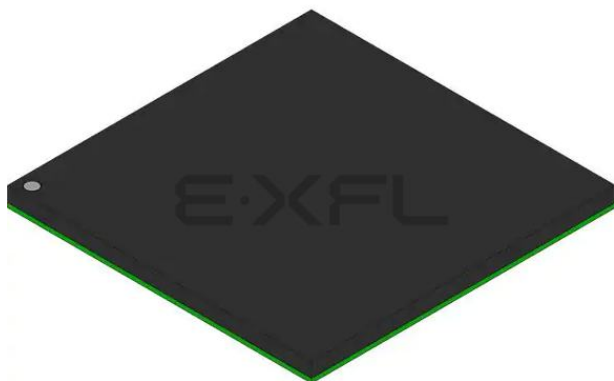




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Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Active
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	80MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (4)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc860srvr80d4

1 Overview

The MPC860 power quad integrated communications controller (PowerQUICC™) is a versatile one-chip integrated microprocessor and peripheral combination designed for a variety of controller applications. It particularly excels in communications and networking systems. The PowerQUICC unit is referred to as the MPC860 in this hardware specification.

The MPC860 implements Power Architecture™ technology and contains a superset of Freescale's MC68360 quad integrated communications controller (QUICC), referred to here as the QUICC, RISC communications processor module (CPM). The CPU on the MPC860 is a 32-bit core built on Power Architecture technology that incorporates memory management units (MMUs) and instruction and data caches. The CPM from the MC68360 QUICC has been enhanced by the addition of the inter-integrated controller (I²C) channel. The memory controller has been enhanced, enabling the MPC860 to support any type of memory, including high-performance memories and new types of DRAMs. A PCMCIA socket controller supports up to two sockets. A real-time clock has also been integrated.

Table 1 shows the functionality supported by the MPC860 family.

Table 1. MPC860 Family Functionality

Part	Cache (Kbytes)		Ethernet		ATM	SCC	Reference ¹
	Instruction Cache	Data Cache	10T	10/100			
MPC860DE	4	4	Up to 2	—	—	2	1
MPC860DT	4	4	Up to 2	1	Yes	2	1
MPC860DP	16	8	Up to 2	1	Yes	2	1
MPC860EN	4	4	Up to 4	—	—	4	1
MPC860SR	4	4	Up to 4	—	Yes	4	1
MPC860T	4	4	Up to 4	1	Yes	4	1
MPC860P	16	8	Up to 4	1	Yes	4	1
MPC855T	4	4	1	1	Yes	1	2

¹ Supporting documentation for these devices refers to the following:

1. MPC860 PowerQUICC Family User's Manual (MPC860UM, Rev. 3)
2. MPC855T User's Manual (MPC855TUM, Rev. 1)

- System integration unit (SIU)
 - Bus monitor
 - Software watchdog
 - Periodic interrupt timer (PIT)
 - Low-power stop mode
 - Clock synthesizer
 - Decrementer, time base, and real-time clock (RTC)
 - Reset controller
 - IEEE 1149.1™ Std. test access port (JTAG)
- Interrupts
 - Seven external interrupt request (IRQ) lines
 - 12 port pins with interrupt capability
 - 23 internal interrupt sources
 - Programmable priority between SCCs
 - Programmable highest priority request
- 10/100 Mbps Ethernet support, fully compliant with the IEEE 802.3u® Standard (not available when using ATM over UTOPIA interface)
- ATM support compliant with ATM forum UNI 4.0 specification
 - Cell processing up to 50–70 Mbps at 50-MHz system clock
 - Cell multiplexing/demultiplexing
 - Support of AAL5 and AAL0 protocols on a per-VC basis. AAL0 support enables OAM and software implementation of other protocols.
 - ATM pace control (APC) scheduler, providing direct support for constant bit rate (CBR) and unspecified bit rate (UBR) and providing control mechanisms enabling software support of available bit rate (ABR)
 - Physical interface support for UTOPIA (10/100-Mbps is not supported with this interface) and byte-aligned serial (for example, T1/E1/ADSL)
 - UTOPIA-mode ATM supports level-1 master with cell-level handshake, multi-PHY (up to four physical layer devices), connection to 25-, 51-, or 155-Mbps framers, and UTOPIA/system clock ratios of 1/2 or 1/3.
 - Serial-mode ATM connection supports transmission convergence (TC) function for T1/E1/ADSL lines, cell delineation, cell payload scrambling/descrambling, automatic idle/unassigned cell insertion/stripping, header error control (HEC) generation, checking, and statistics.
- Communications processor module (CPM)
 - RISC communications processor (CP)
 - Communication-specific commands (for example, GRACEFUL STOP TRANSMIT, ENTER HUNT MODE, and RESTART TRANSMIT)
 - Supports continuous mode transmission and reception on all serial channels

- Up to 8 Kbytes of dual-port RAM
- 16 serial DMA (SDMA) channels
- Three parallel I/O registers with open-drain capability
- Four baud-rate generators (BRGs)
 - Independent (can be tied to any SCC or SMC)
 - Allows changes during operation
 - Autobaud support option
- Four serial communications controllers (SCCs)
 - Ethernet/IEEE 802.3® standard optional on SCC1–4, supporting full 10-Mbps operation (available only on specially programmed devices)
 - HDLC/SDLC (all channels supported at 2 Mbps)
 - HDLC bus (implements an HDLC-based local area network (LAN))
 - Asynchronous HDLC to support point-to-point protocol (PPP)
 - AppleTalk
 - Universal asynchronous receiver transmitter (UART)
 - Synchronous UART
 - Serial infrared (IrDA)
 - Binary synchronous communication (BISYNC)
 - Totally transparent (bit streams)
 - Totally transparent (frame-based with optional cyclic redundancy check (CRC))
- Two SMCs (serial management channels)
 - UART
 - Transparent
 - General circuit interface (GCI) controller
 - Can be connected to the time-division multiplexed (TDM) channels
- One SPI (serial peripheral interface)
 - Supports master and slave modes
 - Supports multimaster operation on the same bus
- One I²C (inter-integrated circuit) port
 - Supports master and slave modes
 - Multiple-master environment support
- Time-slot assigner (TSA)
 - Allows SCCs and SMCs to run in multiplexed and/or non-multiplexed operation
 - Supports T1, CEPT, PCM highway, ISDN basic rate, ISDN primary rate, user defined
 - 1- or 8-bit resolution
 - Allows independent transmit and receive routing, frame synchronization, and clocking

3 Maximum Tolerated Ratings

This section provides the maximum tolerated voltage and temperature ranges for the MPC860. [Table 2](#) provides the maximum ratings.

This device contains circuitry protecting against damage due to high-static voltage or electrical fields; however, it is advised that normal precautions be taken to avoid application of any voltages higher than maximum-rated voltages to this high-impedance circuit. Reliability of operation is enhanced if unused inputs are tied to an appropriate logic voltage level (for example, either GND or V_{DD}).

Table 2. Maximum Tolerated Ratings

(GND = 0 V)

Rating	Symbol	Value	Unit
Supply voltage ¹	V_{DDH}	−0.3 to 4.0	V
	V_{DDL}	−0.3 to 4.0	V
	KAPWR	−0.3 to 4.0	V
	V_{DDSYN}	−0.3 to 4.0	V
Input voltage ²	V_{in}	GND − 0.3 to V_{DDH}	V
Temperature ³ (standard)	$T_{A(min)}$	0	°C
	$T_{j(max)}$	95	°C
Temperature ³ (extended)	$T_{A(min)}$	−40	°C
	$T_{j(max)}$	95	°C
Storage temperature range	T_{stg}	−55 to 150	°C

¹ The power supply of the device must start its ramp from 0.0 V.

² Functional operating conditions are provided with the DC electrical specifications in [Table 6](#). Absolute maximum ratings are stress ratings only; functional operation at the maxima is not guaranteed. Stress beyond those listed may affect device reliability or cause permanent damage to the device.

Caution: All inputs that tolerate 5 V cannot be more than 2.5 V greater than the supply voltage. This restriction applies to power-up and normal operation (that is, if the MPC860 is unpowered, voltage greater than 2.5 V must not be applied to its inputs).

³ Minimum temperatures are guaranteed as ambient temperature, T_A . Maximum temperatures are guaranteed as junction temperature, T_j .

7 Thermal Calculation and Measurement

For the following discussions, $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$, where $P_{I/O}$ is the power dissipation of the I/O drivers.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J , in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (°C)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

7.2 Estimation with Junction-to-Case Thermal Resistance

Historically, the thermal resistance has frequently been expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)

$R_{\theta CA}$ = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ is device related and cannot be influenced by the user. The user adjusts the thermal environment to affect the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the airflow around the device, add a heat sink, change the mounting arrangement on the printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device. This thermal model is most useful for ceramic packages with heat sinks where some 90% of the heat flows through the case and the heat sink to the ambient environment. For most packages, a better model is required.

7.3 Estimation with Junction-to-Board Thermal Resistance

A simple package thermal model which has demonstrated reasonable accuracy (about 20%) is a two-resistor model consisting of a junction-to-board and a junction-to-case thermal resistance. The junction-to-case thermal resistance covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed-circuit board. It has been observed that the thermal performance of most plastic packages, especially PBGA packages, is strongly dependent on the board temperature; see [Figure 2](#).

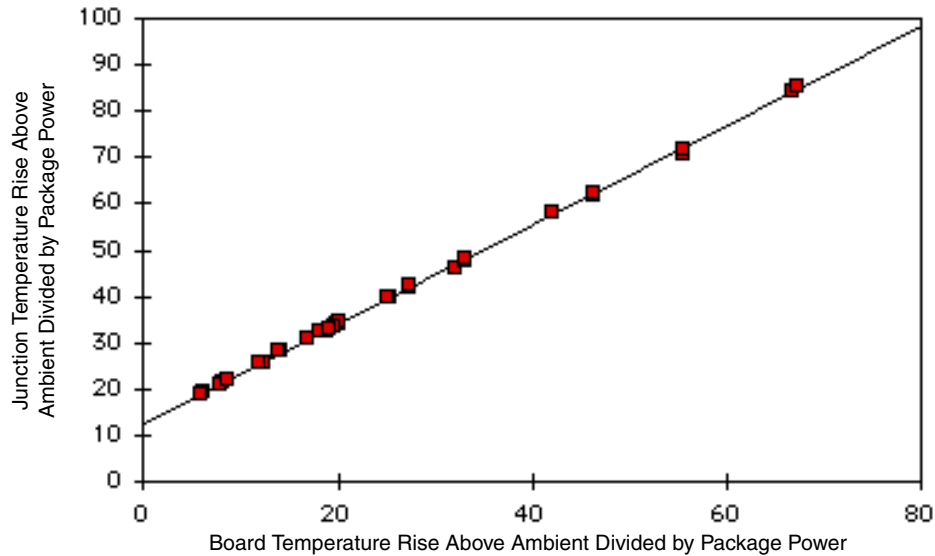


Figure 2. Effect of Board Temperature Rise on Thermal Behavior

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$R_{\theta JB}$ = junction-to-board thermal resistance ($^{\circ}\text{C}/\text{W}$)

T_B = board temperature ($^{\circ}\text{C}$)

P_D = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and by attaching the thermal balls to the ground plane.

7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two-resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

7.5 Experimental Determination

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

9 Bus Signal Timing

Table 7 provides the bus operation timing for the MPC860 at 33, 40, 50, and 66 MHz.

The maximum bus speed supported by the MPC860 is 66 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC860 used at 80 MHz must be configured for a 40-MHz bus).

The timing for the MPC860 bus shown assumes a 50-pF load for maximum delays and a 0-pF load for minimum delays.

Table 7. Bus Operation Timings

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B1	CLKOUT period	30.30	30.30	25.00	30.30	20.00	30.30	15.15	30.30	ns
B1a	EXTCLK to CLKOUT phase skew (EXTCLK > 15 MHz and MF <= 2)	−0.90	0.90	−0.90	0.90	−0.90	0.90	−0.90	0.90	ns
B1b	EXTCLK to CLKOUT phase skew (EXTCLK > 10 MHz and MF < 10)	−2.30	2.30	−2.30	2.30	−2.30	2.30	−2.30	2.30	ns
B1c	CLKOUT phase jitter (EXTCLK > 15 MHz and MF <= 2) ¹	−0.60	0.60	−0.60	0.60	−0.60	0.60	−0.60	0.60	ns
B1d	CLKOUT phase jitter ¹	−2.00	2.00	−2.00	2.00	−2.00	2.00	−2.00	2.00	ns
B1e	CLKOUT frequency jitter (MF < 10) ¹	—	0.50	—	0.50	—	0.50	—	0.50	%
B1f	CLKOUT frequency jitter (10 < MF < 500) ¹	—	2.00	—	2.00	—	2.00	—	2.00	%
B1g	CLKOUT frequency jitter (MF > 500) ¹	—	3.00	—	3.00	—	3.00	—	3.00	%
B1h	Frequency jitter on EXTCLK ²	—	0.50	—	0.50	—	0.50	—	0.50	%
B2	CLKOUT pulse width low	12.12	—	10.00	—	8.00	—	6.06	—	ns
B3	CLKOUT width high	12.12	—	10.00	—	8.00	—	6.06	—	ns
B4	CLKOUT rise time ³	—	4.00	—	4.00	—	4.00	—	4.00	ns
B5 ³³	CLKOUT fall time ³	—	4.00	—	4.00	—	4.00	—	4.00	ns
B7	CLKOUT to A(0:31), BADDR(28:30), RD/WR, BURST, D(0:31), DP(0:3) invalid	7.58	—	6.25	—	5.00	—	3.80	—	ns
B7a	CLKOUT to TSIZ(0:1), REG, RSV, AT(0:3), BDIP, PTR invalid	7.58	—	6.25	—	5.00	—	3.80	—	ns
B7b	CLKOUT to BR, BG, FRZ, VFLS(0:1), VF(0:2) IWP(0:2), LWP(0:1), STS invalid ⁴	7.58	—	6.25	—	5.00	—	3.80	—	ns
B8	CLKOUT to A(0:31), BADDR(28:30) RD/WR, BURST, D(0:31), DP(0:3) valid	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns
B8a	CLKOUT to TSIZ(0:1), REG, RSV, AT(0:3) BDIP, PTR valid	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns
B8b	CLKOUT to BR, BG, VFLS(0:1), VF(0:2), IWP(0:2), FRZ, LWP(0:1), STS valid ⁴	7.58	14.33	6.25	13.00	5.00	11.75	3.80	10.04	ns

Table 7. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B29d	$\overline{WE}(0:3)$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 1, CSNT = 1, EBDF = 0	43.45	—	35.5	—	28.00	—	20.73	—	ns
B29e	$\overline{CS}$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 1, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 0	43.45	—	35.5	—	28.00	—	29.73	—	ns
B29f	$\overline{WE}(0:3)$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 0, CSNT = 1, EBDF = 1	8.86	—	6.88	—	5.00	—	3.18	—	ns
B29g	$\overline{CS}$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 0, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 1	8.86	—	6.88	—	5.00	—	3.18	—	ns
B29h	$\overline{WE}(0:3)$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 1, CSNT = 1, EBDF = 1	38.67	—	31.38	—	24.50	—	17.83	—	ns
B29i	$\overline{CS}$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 1, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 1	38.67	—	31.38	—	24.50	—	17.83	—	ns
B30	$\overline{CS}$, $\overline{WE}(0:3)$ negated to A(0:31), BADDR(28:30) invalid GPCM write access ⁸	5.58	—	4.25	—	3.00	—	1.79	—	ns
B30a	$\overline{WE}(0:3)$ negated to A(0:31), BADDR(28:30) invalid GPCM, write access, TRLX = 0, CSNT = 1, $\overline{CS}$ negated to A(0:31) invalid GPCM write access, TRLX = 0, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 0	13.15	—	10.50	—	8.00	—	5.58	—	ns
B30b	$\overline{WE}(0:3)$ negated to A(0:31), invalid GPCM BADDR(28:30) invalid GPCM write access, TRLX = 1, CSNT = 1. $\overline{CS}$ negated to A(0:31), Invalid GPCM, write access, TRLX = 1, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 0	43.45	—	35.50	—	28.00	—	20.73	—	ns
B30c	$\overline{WE}(0:3)$ negated to A(0:31), BADDR(28:30) invalid GPCM write access, TRLX = 0, CSNT = 1. $\overline{CS}$ negated to A(0:31) invalid GPCM write access, TRLX = 0, CSNT = 1, ACS = 10, ACS = 11, EBDF = 1	8.36	—	6.38	—	4.50	—	2.68	—	ns
B30d	$\overline{WE}(0:3)$ negated to A(0:31), BADDR(28:30) invalid GPCM write access, TRLX = 1, CSNT = 1. $\overline{CS}$ negated to A(0:31) invalid GPCM write access TRLX = 1, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 1	38.67	—	31.38	—	24.50	—	17.83	—	ns
B31	CLKOUT falling edge to $\overline{CS}$ valid—as requested by control bit CST4 in the corresponding word in UPM	1.50	6.00	1.50	6.00	1.50	6.00	1.50	6.00	ns

Figure 17 provides the timing for the external bus controlled by the UPM.

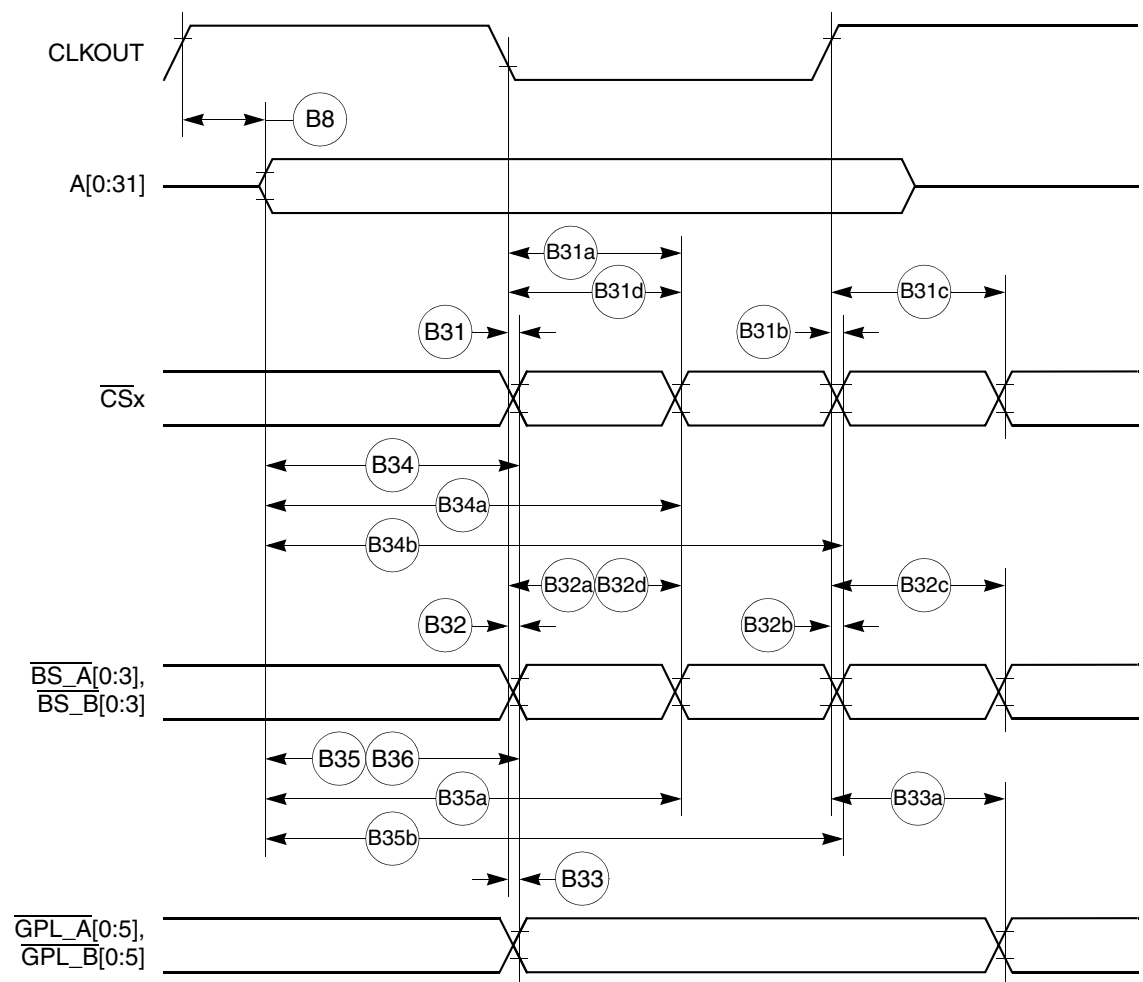


Figure 17. External Bus Timing (UPM Controlled Signals)

Figure 26 provides the PCMCIA access cycle timing for the external bus write.

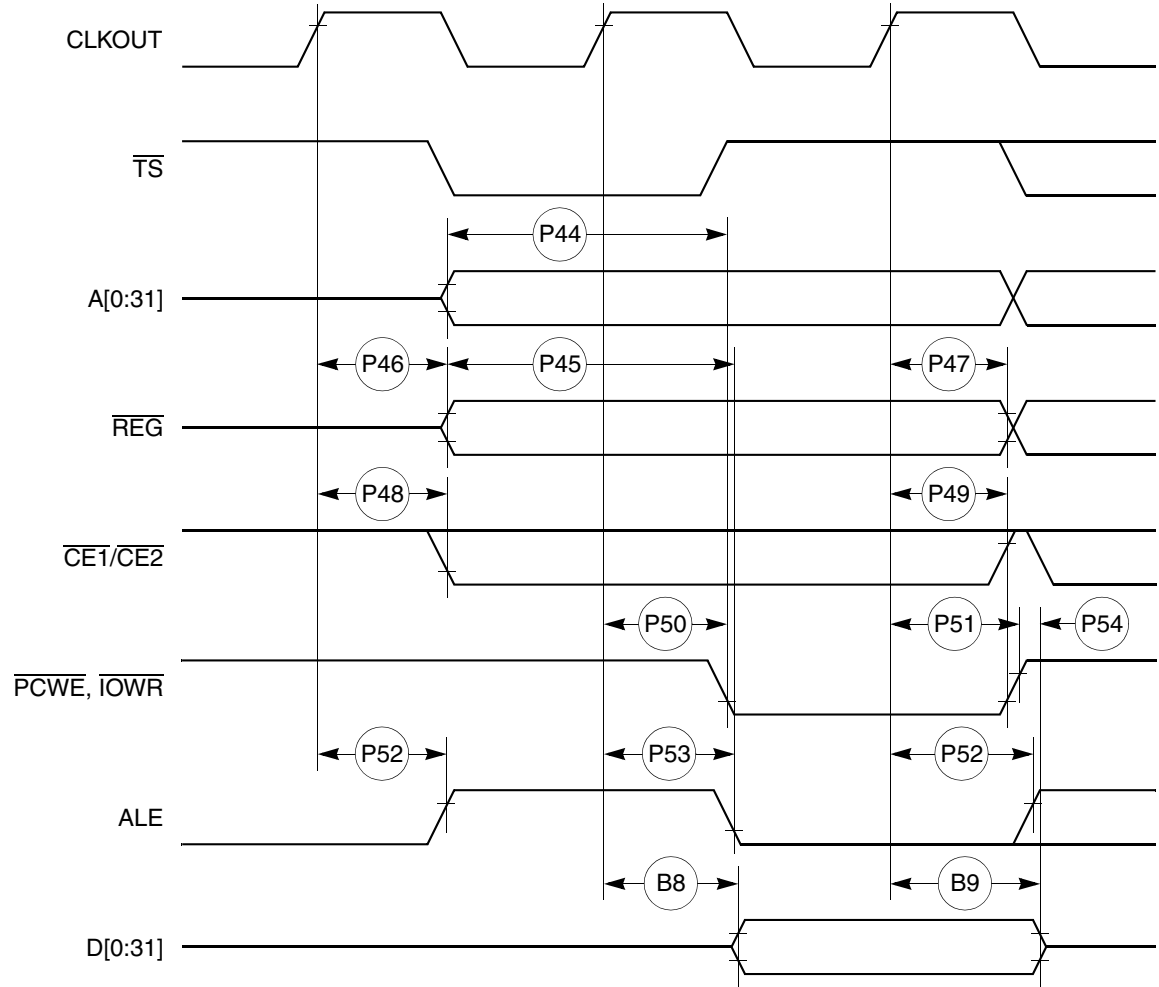


Figure 26. PCMCIA Access Cycle Timing External Bus Write

Figure 27 provides the PCMCIA $\overline{\text{WAIT}}$ signal detection timing.

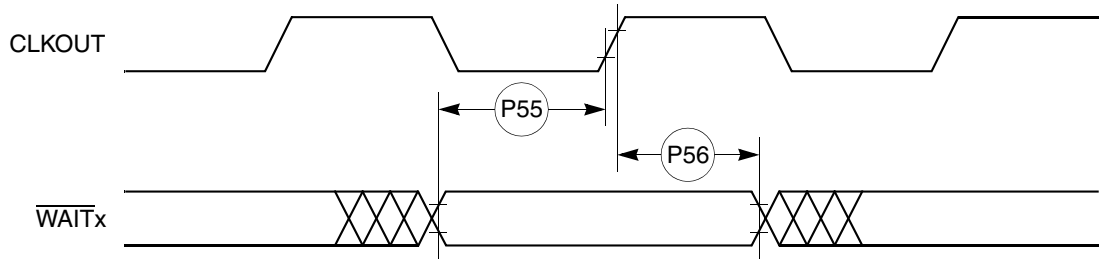


Figure 27. PCMCIA $\overline{\text{WAIT}}$ Signal Detection Timing

Table 10 shows the PCMCIA port timing for the MPC860.

Table 10. PCMCIA Port Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P57	CLKOUT to OPx valid	—	19.00	—	19.00	—	19.00	—	19.00	ns
P58	$\overline{\text{HRESET}}$ negated to OPx drive ¹	25.73	—	21.75	—	18.00	—	14.36	—	ns
P59	IP_Xx valid to CLKOUT rising edge	5.00	—	5.00	—	5.00	—	5.00	—	ns
P60	CLKOUT rising edge to IP_Xx invalid	1.00	—	1.00	—	1.00	—	1.00	—	ns

¹ OP2 and OP3 only.

Figure 28 provides the PCMCIA output port timing for the MPC860.

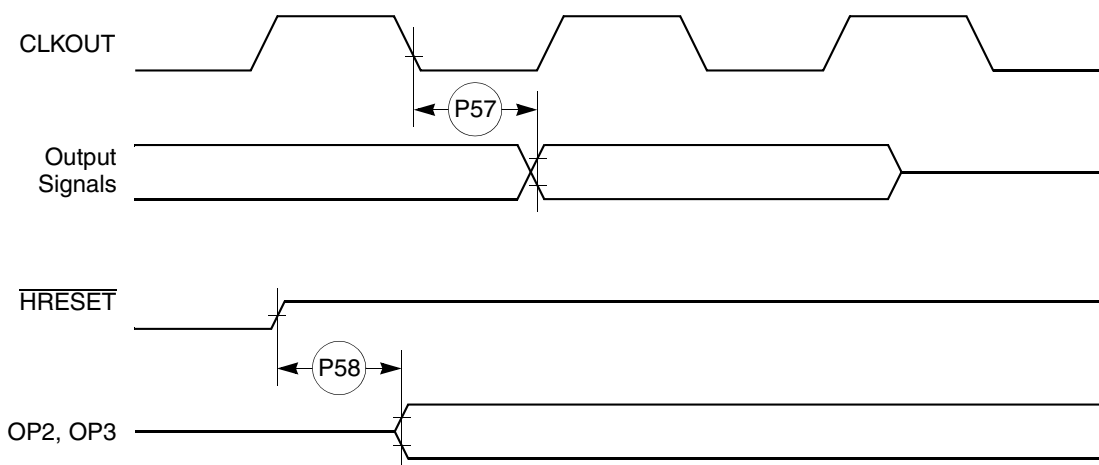


Figure 28. PCMCIA Output Port Timing

Figure 29 provides the PCMCIA output port timing for the MPC860.

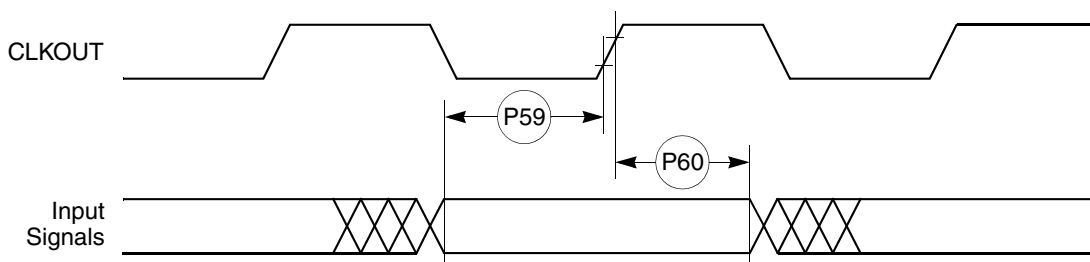


Figure 29. PCMCIA Input Port Timing

Table 12 shows the reset timing for the MPC860.

Table 12. Reset Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to $\overline{\text{HRESET}}$ high impedance	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to $\overline{\text{SRESET}}$ high impedance	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	$\overline{\text{RSTCONF}}$ pulse width	515.15	—	425.00	—	340.00	—	257.58	—	ns
R72	—	—	—	—	—	—	—	—	—	
R73	Configuration data to $\overline{\text{HRESET}}$ rising edge setup time	504.55	—	425.00	—	350.00	—	277.27	—	ns
R74	Configuration data to $\overline{\text{RSTCONF}}$ rising edge setup time	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after $\overline{\text{RSTCONF}}$ negation	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after $\overline{\text{HRESET}}$ negation	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	$\overline{\text{HRESET}}$ and $\overline{\text{RSTCONF}}$ asserted to data out drive	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	$\overline{\text{RSTCONF}}$ negated to data out high impedance	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-state $\overline{\text{HRESET}}$ to data out high impedance	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK setup	90.91	—	75.00	—	60.00	—	45.45	—	ns
R81	DSDI, DSCK hold time	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	$\overline{\text{SRESET}}$ negated to CLKOUT rising edge for DSDI and DSCK sample	242.42	—	200.00	—	160.00	—	121.21	—	ns

11 CPM Electrical Characteristics

This section provides the AC and DC electrical specifications for the communications processor module (CPM) of the MPC860.

11.1 PIP/PIO AC Electrical Specifications

Table 14 provides the PIP/PIO AC timings as shown in Figure 39 through Figure 43.

Table 14. PIP/PIO Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
21	Data-in setup time to STBI low	0	—	ns
22	Data-in hold time to STBI high	$2.5 - t_3^1$	—	CLK
23	STBI pulse width	1.5	—	CLK
24	STBO pulse width	1 CLK – 5 ns	—	ns
25	Data-out setup time to STBO low	2	—	CLK
26	Data-out hold time from STBO high	5	—	CLK
27	STBI low to STBO low (Rx interlock)	—	2	CLK
28	STBI low to STBO high (Tx interlock)	2	—	CLK
29	Data-in setup time to clock high	15	—	ns
30	Data-in hold time from clock high	7.5	—	ns
31	Clock low to data-out valid (CPU writes data, control, or direction)	—	25	ns

¹ t_3 = Specification 23.

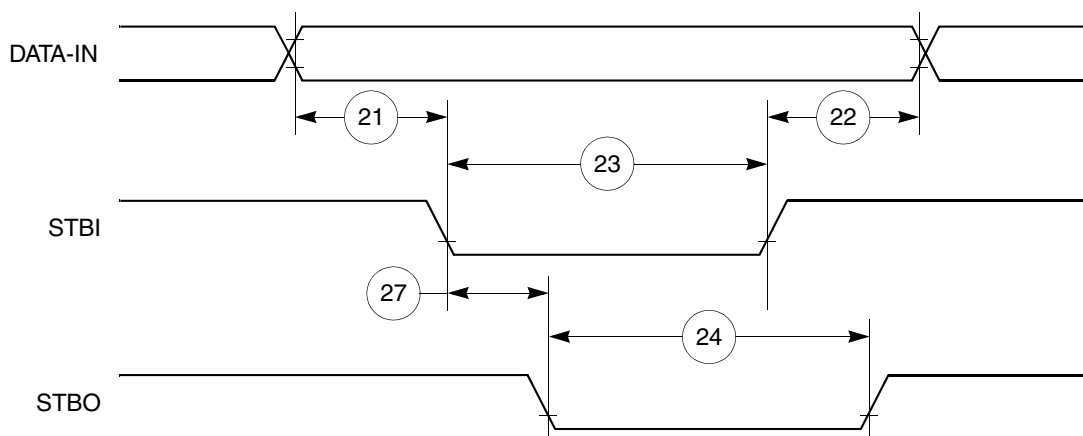


Figure 39. PIP Rx (Interlock Mode) Timing Diagram

Table 16. IDMA Controller Timing (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
42	$\overline{\text{SDACK}}$ assertion delay from clock high	—	12	ns
43	$\overline{\text{SDACK}}$ negation delay from clock low	—	12	ns
44	$\overline{\text{SDACK}}$ negation delay from $\overline{\text{TA}}$ low	—	20	ns
45	$\overline{\text{SDACK}}$ negation delay from clock high	—	15	ns
46	$\overline{\text{TA}}$ assertion to rising edge of the clock setup time (applies to external $\overline{\text{TA}}$)	7	—	ns

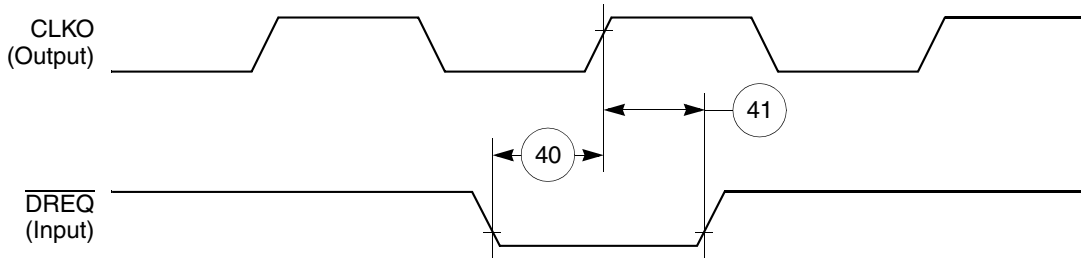


Figure 45. IDMA External Requests Timing Diagram

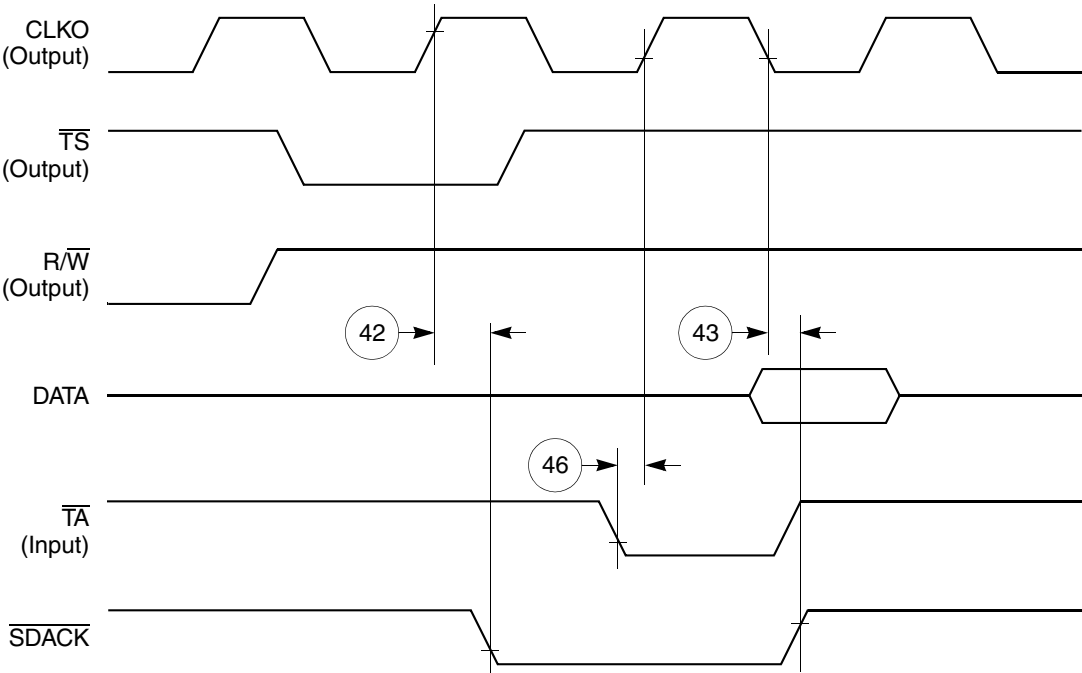


Figure 46. $\overline{\text{SDACK}}$ Timing Diagram—Peripheral Write, Externally-Generated $\overline{\text{TA}}$

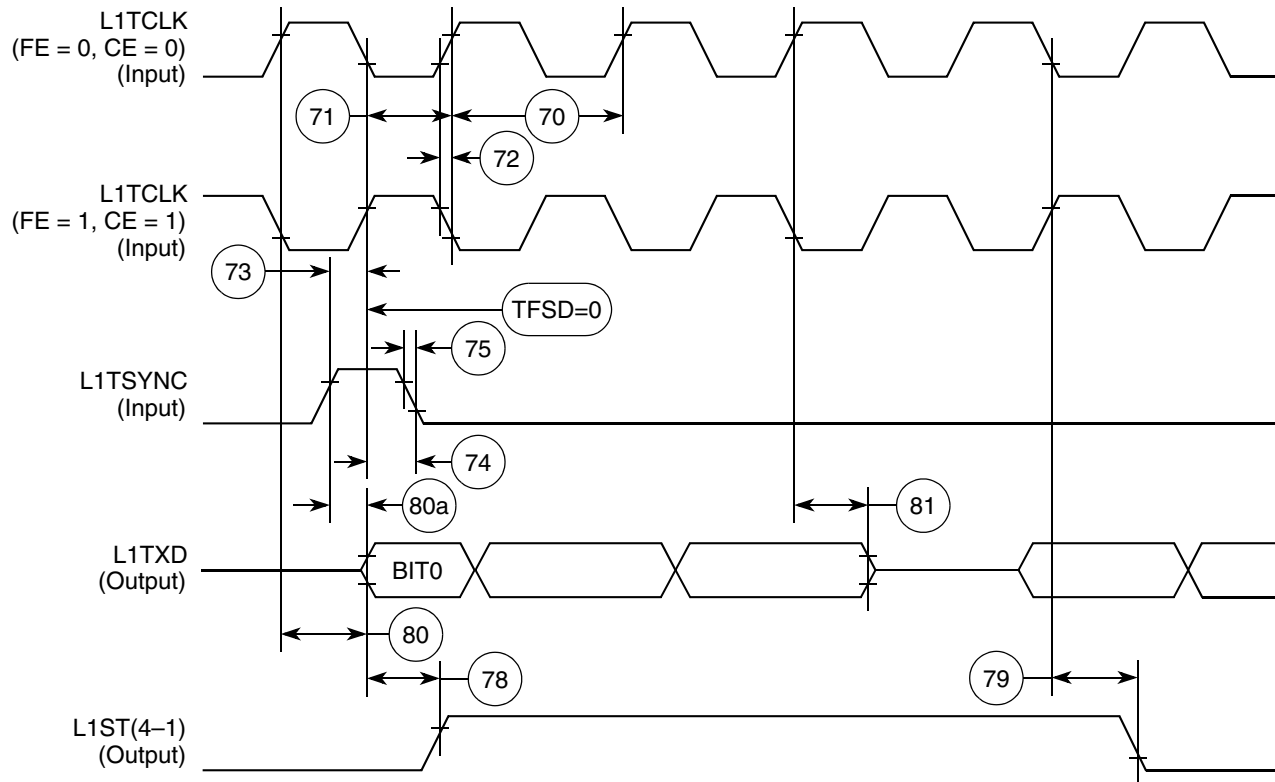
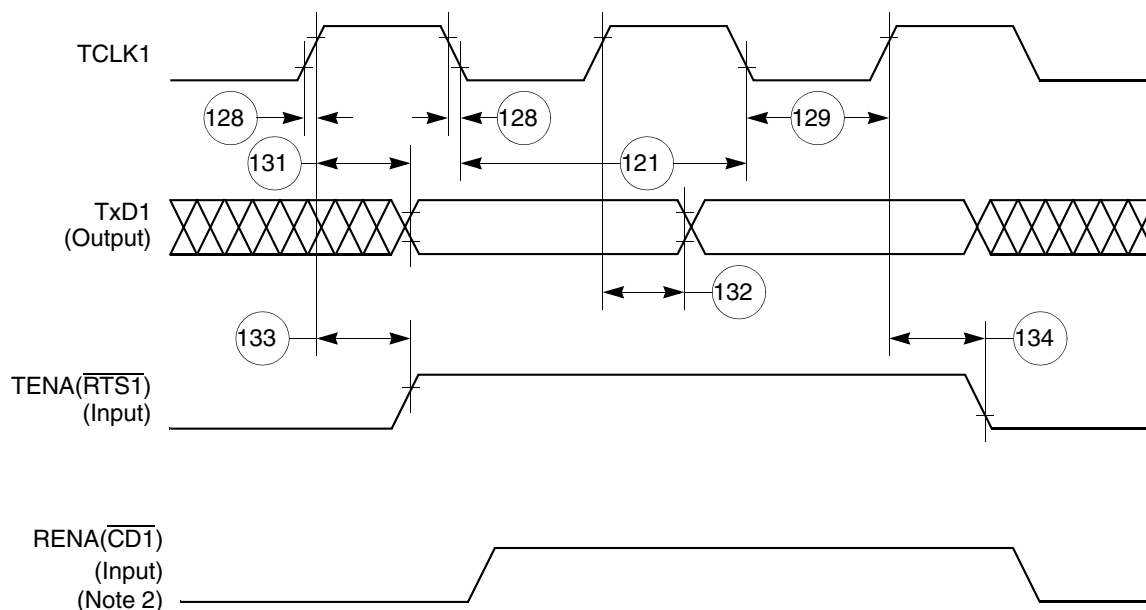


Figure 53. SI Transmit Timing Diagram (DSC = 0)



Notes:

1. Transmit clock invert (TCI) bit in GSMR is set.
2. If RENA is deasserted before TENA, or RENA is not asserted at all during transmit, then the CSL bit is set in the buffer descriptor at the end of the frame transmission.

Figure 61. Ethernet Transmit Timing Diagram

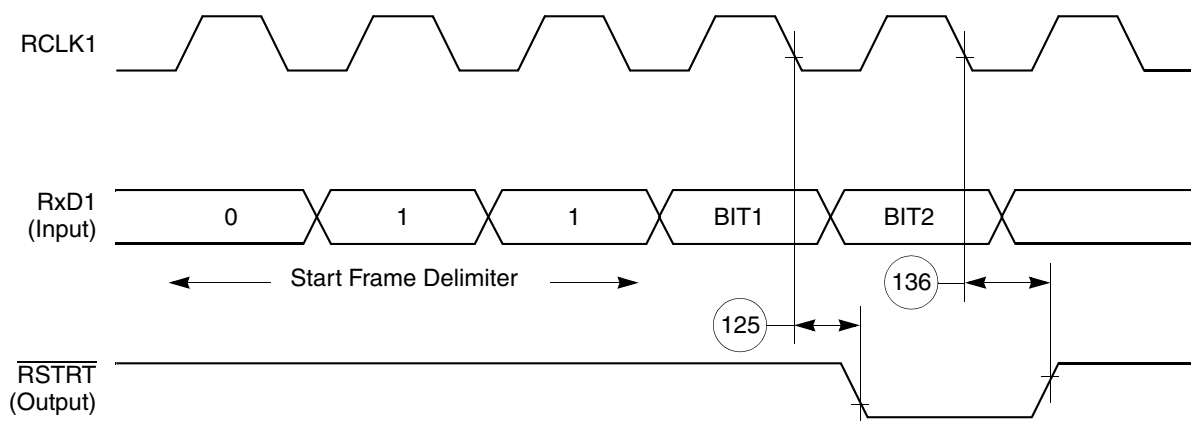


Figure 62. CAM Interface Receive Start Timing Diagram



Figure 63. CAM Interface $\overline{\text{REJECT}}$ Timing Diagram

11.10 SPI Master AC Electrical Specifications

Table 24 provides the SPI master timings as shown in Figure 65 and Figure 66.

Table 24. SPI Master Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
160	MASTER cycle time	4	1024	t _{cyc}
161	MASTER clock (SCK) high or low time	2	512	t _{cyc}
162	MASTER data setup time (inputs)	50	—	ns
163	Master data hold time (inputs)	0	—	ns
164	Master data valid (after SCK edge)	—	20	ns
165	Master data hold time (outputs)	0	—	ns
166	Rise time output	—	15	ns
167	Fall time output	—	15	ns

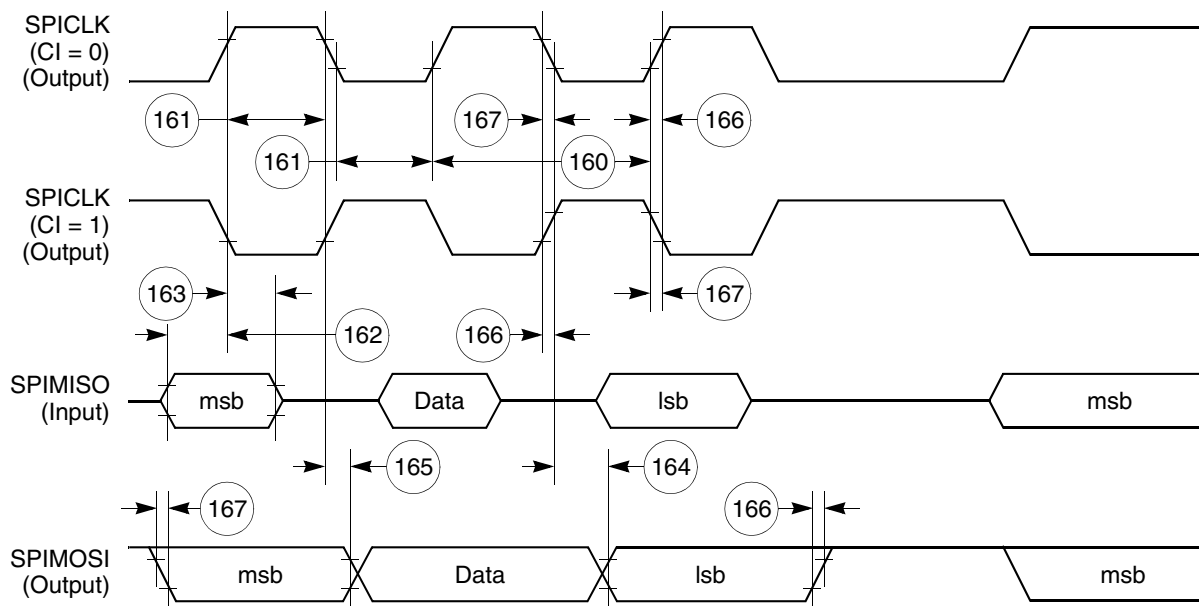


Figure 65. SPI Master (CP = 0) Timing Diagram

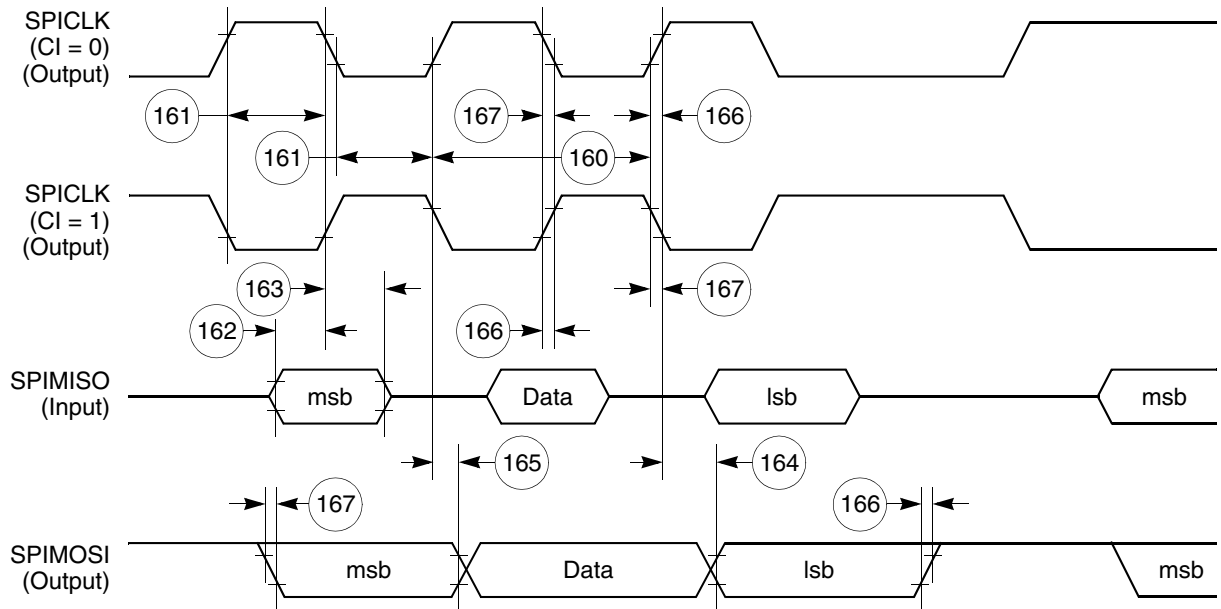


Figure 66. SPI Master (CP = 1) Timing Diagram

11.11 SPI Slave AC Electrical Specifications

Table 25 provides the SPI slave timings as shown in Figure 67 and Figure 68.

Table 25. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	t_{cyc}
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	t_{cyc}
174	Slave sequential transfer delay (does not require deselect)	1	—	t_{cyc}
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

Figure 69 shows the I²C bus timing.

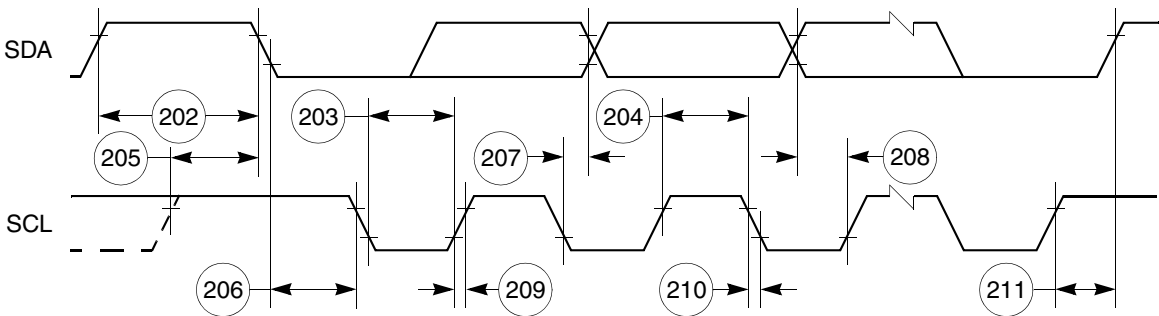


Figure 69. I²C Bus Timing Diagram

12 UTOPIA AC Electrical Specifications

Table 28 shows the AC electrical specifications for the UTOPIA interface.

Table 28. UTOPIA AC Electrical Specifications

Num	Signal Characteristic	Direction	Min	Max	Unit
U1	UtpClk rise/fall time (Internal clock option)	Output	—	3.5	ns
	Duty cycle		50	50	%
	Frequency		—	50	MHz
U1a	UtpClk rise/fall time (external clock option)	Input	—	3.5	ns
	Duty cycle		40	60	%
	Frequency		—	50	MHz
U2	$\overline{\text{RxEnb}}$ and $\overline{\text{TxEnb}}$ active delay	Output	2	16	ns
U3	UTPB, SOC, Rxclav and Txclav setup time	Input	8	—	ns
U4	UTPB, SOC, Rxclav and Txclav hold time	Input	1	—	ns
U5	UTPB, SOC active delay (and PHREQ and PHSEL active delay in MPHY mode)	Output	2	16	ns

13.2 MII Transmit Signal Timing (MII_TXD[3:0], MII_TX_EN, MII_TX_ER, MII_TX_CLK)

The transmitter functions correctly up to a MII_TX_CLK maximum frequency of 25 MHz +1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII_TX_CLK frequency – 1%.

Table 30 provides information on the MII transmit signal timing.

Table 30. MII Transmit Signal Timing

Num	Characteristic	Min	Max	Unit
M5	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER invalid	5	—	ns
M6	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER valid	—	25	
M7	MII_TX_CLK pulse width high	35	65%	MII_TX_CLK period
M8	MII_TX_CLK pulse width low	35%	65%	MII_TX_CLK period

Figure 73 shows the MII transmit signal timing diagram.

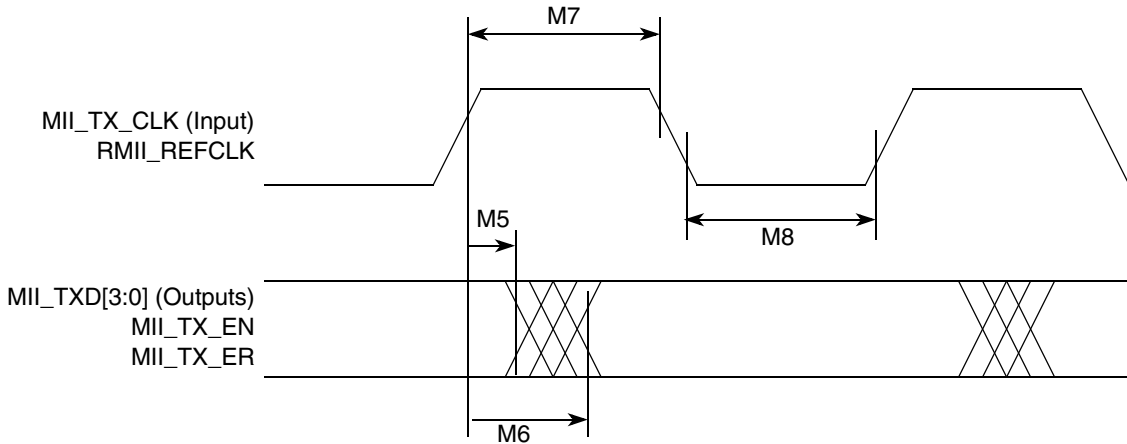


Figure 73. MII Transmit Signal Timing Diagram