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"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "<u>Embedded - Microcontrollers</u>"

Supplier Device Package	48-LQFP (7x7)
Package / Case	48-LQFP
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 105°C (TA)
Oscillator Type	Internal
Data Converters	A/D 24x10b
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
RAM Size	1.5K x 8
EEPROM Size	-
Program Memory Type	FLASH
Program Memory Size	32KB (32K x 8)
Number of I/O	37
Peripherals	LVD, POR, PWM
Connectivity	SCI, SPI
Speed	8MHz
Core Size	8-Bit
Core Processor	HC08
Product Status	Obsolete
Details	

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1.4 Pin Assignments

Figure 1-2, Figure 1-3, and Figure 1-4 illustrate the pin assignments for the 32-pin LQFP, 48-pin LQFP, and 64-pin QFP respectively.

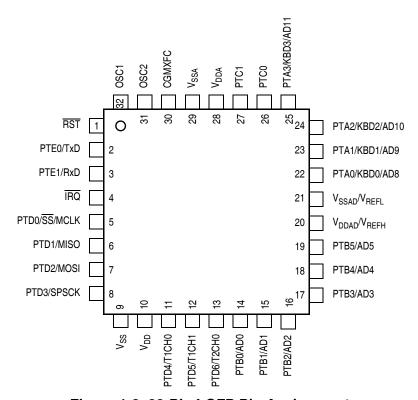


Figure 1-2. 32-Pin LQFP Pin Assignments



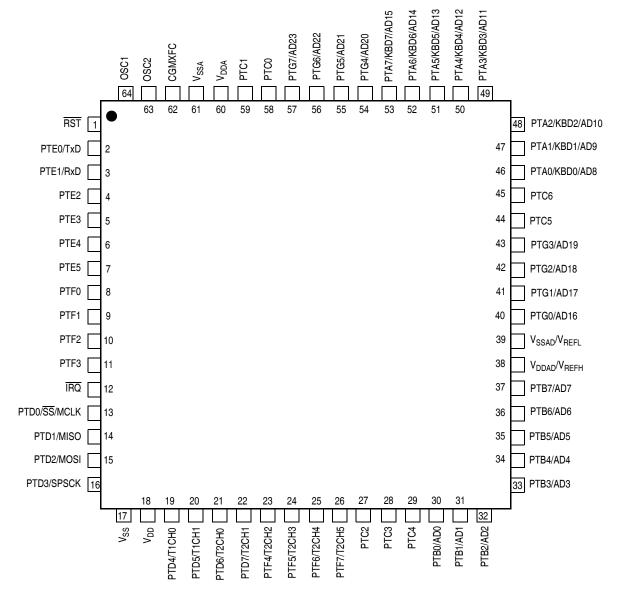


Figure 1-4. 64-Pin QFP Pin Assignments

1.5 Pin Functions

Descriptions of the pin functions are provided here.

1.5.1 Power Supply Pins (V_{DD} and V_{SS})

 V_{DD} and V_{SS} are the power supply and ground pins. The MCU operates from a single power supply.

Fast signal transitions on MCU pins place high, short-duration current demands on the power supply. To prevent noise problems, take special care to provide power supply bypassing at the MCU as Figure 1-5 shows. Place the C1 bypass capacitor as close to the MCU as possible. Use a high-frequency-response ceramic capacitor for C1. C2 is an optional bulk current bypass capacitor for use in applications that require the port pins to source high current levels.

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Addr.	Register Name		Bit 7	6	5	4	3	2	1	Bit 0		
\$0030 TIM2 Channel 0 Status and Control Register (T2SC0) See page 237.	Control Register (T2SC0)	Read:	CH0F	- CH0IE	MS0B	MS0A	ELS0B	ELS0A	TOV0	CH0MAX		
		Write:	0	CHOIL	WISOD	WISOA	LLSOD	LLOUA	1000	CHOWAX		
	Reset:	0	0	0	0	0	0	0	0			
\$0031	TIM2 Channel 0 \$0031 Register High (T2CH0H)	Read: Write:	Bit 15	14	13	12	11	10	9	Bit 8		
	See page 240.	Reset:		Indeterminate after reset								
\$0032	TIM2 Channel 0 Register Low (T2CH0L)	Read: Write:	Bit 7	6	5	4	3	2	1	Bit 0		
	See page 240.	Reset:		Indeterminate after reset								
	TIM2 Channel 1 Status and	Read:	CH1F	CH1IE	0	MS1A	ELS1B	ELS1A	TOV1	CH1MAX		
\$0033	Control Register (T2SC1) See page 237.	Write:	0									
	See page 237.	Reset:	0	0	0	0	0	0	0	0		
\$0034	TIM2 Channel 1 Register High (T2CH1H)	Read: Write:	Bit 15	14	13	12	11	10	9	Bit 8		
	See page 240.	Reset:				Indetermina	te after reset					
\$0035	TIM2 Channel 1 Register Low (T2CH1L)	Read: Write:	Bit 7	6	5	4	3	2	1	Bit 0		
	See page 240.	Reset:	Reset: Indeterminate after reset									
\$0036	PLL Control Register (PCTL)	Read: Write:	PLLIE	PLLF	PLLON	BCS	R	R	VPR1	VPR0		
	See page 81.	Reset:	0	0	1	0	0	0	0	0		
	PLL Bandwidth Control	Read:	AUTO	LOCK	ACQ	0	0	0	0	R		
\$0037	Register (PBWC)	Write:	AOTO		nou					11		
	See page 82.	Reset:	0	0	0	0	0	0	0	0		
	PLL Multiplier Select High	Read:	0	0	0	0	MUL11	MUL10	MUL9	MUL8		
\$0038	Register (PMSH) See page 83.	Write:										
	Gee page 65.	Reset:	0	0	0	0	0	0	0	0		
\$0039	PLL Multiplier Select Low Register (PMSL)	Read: Write:	MUL7	MUL6	MUL5	MUL4	MUL3	MUL2	MUL1	MUL0		
	See page 84.	Reset:	0	1	0	0	U	U	U	U		
	PLL VCO Select Range Register (PMRS)	Read: Write:	VRS7	VRS6	VRS5	VRS4	VRS3	VRS2	VRS1	VRS0		
	See page 84.	Reset:	0	1	0	0	0	0	0	0		
		Read:	0	0	0	0	R	R	R	R		
\$003B	Reserved	Write:					11	11	11	11		
		Reset:	0	0	0	0	0	0	0	1		
				= Unimplemented R = Reserved				U = Unaffect	ed			

Figure 2-2. Control, Status, and Data Registers (Sheet 5 of 9)

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Memory

Addr.	Register Name		Bit 7	6	5	4	3	2	1	Bit 0
\$FE03	Break Flag Control Register (BFCR)	Read: Write:	BCFE	R	R	R	R	R	R	R
	See page 200.	Reset:	0	0	0	0	0	0	0	0
	Interrupt Status Register 1	Read:	IF6	IF5	IF4	IF3	IF2	IF1	0	0
\$FE04	(INT1) See page 195.	Write:	R	R	R	R	R	R	R	R
		Reset:	0	0	0	0	0	0	0	0
	Interrupt Status Register 2	Read:	IF14	IF13	IF12	IF11	IF10	IF9	IF8	IF7
\$FE05	(INT2)	Write:	R	R	R	R	R	R	R	R
	See page 195.	Reset:	0	0	0	0	0	0	0	0
	Interrupt Status Register 3	Read:	IF22	IF21	IF20	IF19	IF18	IF17	IF16	IF15
\$FE06	(INT3)	Write:	R	R	R	R	R	R	R	R
	See page 195.	Reset:	0	0	0	0	0	0	0	0
	Interrupt Status Register 4	Read:	0	0	0	0	0	0	IF24	IF23
\$FE07	(INT4)	Write:	R	R	R	R	R	R	R	R
	See page 196.	Reset:	0	0	0	0	0	0	0	0
\$FE08	FLASH-2 Control Register (FL2CR) See page 50.	Read:	0	0	0	0	LIVEN	MACC	ERASE	DOM
		Write:					HVEN	MASS	ERASE	PGM
		Reset:	0	0	0	0	0	0	0	0
\$FE09	Break Address Register High (BRKH)	Read: Write:	Bit 15	14	13	12	11	10	9	Bit 8
	See page 265.	Reset:	0	0	0	0	0	0	0	0
\$FE0A	Break Address Register Low (BRKL)	Read: Write:	Bit 7	6	5	4	3	2	1	Bit 0
	See page 265.	Reset:	0	0	0	0	0	0	0	0
	Break Status and Control	Read:	BRKE	BRKA	0	0	0	0	0	0
\$FE0B	Register (BRKSCR)	Write:	Dince	Brillot						
	See page 265.	Reset:	0	0	0	0	0	0	0	0
	LVI Status Register	Read:	LVIOUT	0	0	0	0	0	0	0
\$FE0C	(LVISR)	Write:								
	See page 129.	Reset:	0	0	0	0	0	0	0	0
\$FE0D	FLASH-2 Test Control Register (FLTCR2)	Read: Write:	R	R	R	R	R	R	R	R
	negister (FLTGR2)	Reset:	0	0	0	0	0	0	0	0
\$FE0E	FLASH-1 Test Control Register (FLTCR1)	Read: Write:	R	R	R	R	R	R	R	R
		Reset:	0	0	0	0	0	0	0	0
				= Unimplemented R = Reserved		ed	U = Unaffect	ted		

Figure 2-2. Control, Status, and Data Registers (Sheet 8 of 9)

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Chapter 3 Analog-to-Digital Converter (ADC)

3.1 Introduction

This section describes the 10-bit analog-to-digital converter (ADC).

3.2 Features

Features of the ADC module include:

- 24 channels with multiplexed input
- Linear successive approximation with monotonicity
- 10-bit resolution
- Single or continuous conversion
- Conversion complete flag or conversion complete interrupt
- Selectable ADC clock
- Left or right justified result
- Left justified sign data mode

3.3 Functional Description

The ADC provides 24 pins for sampling external sources at pins PTG7/AD23–PTG0/AD16, PTA7/KBD7/AD15–PTA0/KBD0/AD8, and PTB7/AD7–PTB0/AD0. An analog multiplexer allows the single ADC converter to select one of 24 ADC channels as ADC voltage in (V_{ADIN}). V_{ADIN} is converted by the successive approximation register-based analog-to-digital converter. When the conversion is completed, ADC places the result in the ADC data register and sets a flag or generates an interrupt. See Figure 3-2.

3.3.1 ADC Port I/O Pins

PTG7/AD23—PTG0/AD16, PTA7/KBD7/AD15—PTA0/KBD0/AD8, and PTB7/AD7—PTB0/AD0 are general-purpose I/O (input/output) pins that share with the ADC channels. The channel select bits define which ADC channel/port pin will be used as the input signal. The ADC overrides the port I/O logic by forcing that pin as input to the ADC. The remaining ADC channels/port pins are controlled by the port I/O logic and can be used as general-purpose I/O. Writes to the port register or data direction register (DDR) will not have any affect on the port pin that is selected by the ADC. A read of a port pin in use by the ADC will return a 0.

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3.7.4 ADC Voltage Reference Low Pin (V_{REFL})

The ADC analog portion uses V_{REFL} as its lower voltage reference pin. By default, connect the V_{REFL} pin to the same voltage potential as V_{SS} . External filtering is often necessary to ensure a clean V_{REFL} for good results. Any noise present on this pin will be reflected and possibly magnified in A/D conversion values.

NOTE

For maximum noise immunity, route V_{REFL} carefully and, if not connected to V_{SS} , place bypass capacitors as close as possible to the package. Routing V_{REFH} close and parallel to V_{REFL} may improve common mode noise rejection.

V_{SSAD} and V_{REFL} are bonded internally.

3.7.5 ADC Voltage In (V_{ADIN})

V_{ADIN} is the input voltage signal from one of the 24 ADC channels to the ADC module.

3.8 I/O Registers

These I/O registers control and monitor ADC operation:

- ADC status and control register (ADSCR)
- ADC data register (ADRH and ADRL)
- ADC clock register (ADCLK)

3.8.1 ADC Status and Control Register

Function of the ADC status and control register (ADSCR) is described here.

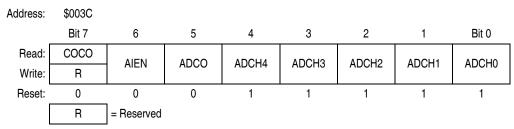


Figure 3-4. ADC Status and Control Register (ADSCR)

COCO — Conversions Complete Bit

In non-interrupt mode (AIEN = 0), COCO is a read-only bit that is set at the end of each conversion. COCO will stay set until cleared by a read of the ADC data register. Reset clears this bit.

In interrupt mode (AIEN = 1), COCO is a read-only bit that is not set at the end of a conversion. It always reads as a 0.

- 1 = Conversion completed (AIEN = 0)
- 0 = Conversion not completed (AIEN = 0) or CPU interrupt enabled (AIEN = 1)

NOTE

The write function of the COCO bit is reserved. When writing to the ADSCR register, always have a 0 in the COCO bit position.

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Clock Generator Module (CGM)

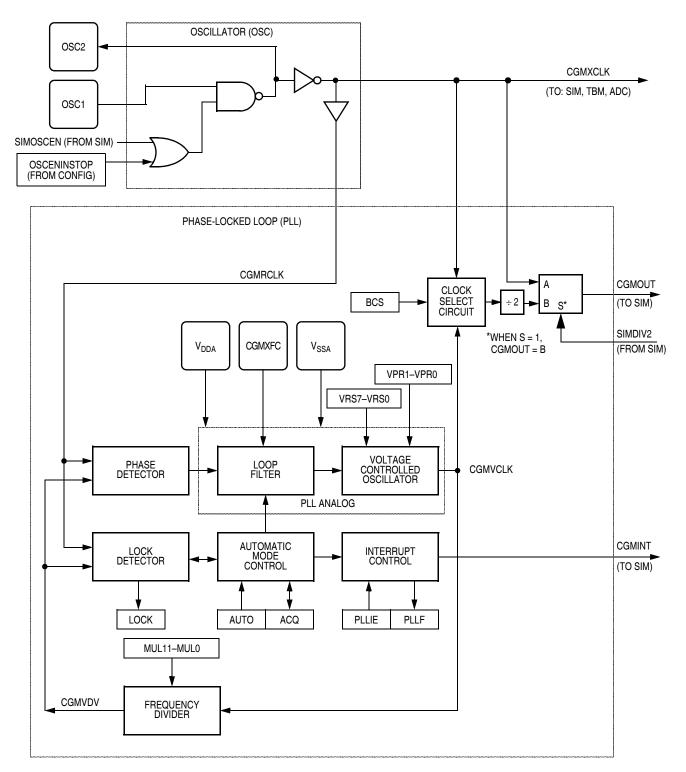


Figure 4-1. CGM Block Diagram



The most critical parameter which affects the reaction times of the PLL is the reference frequency, f_{RCLK} . This frequency is the input to the phase detector and controls how often the PLL makes corrections. For stability, the corrections must be small compared to the desired frequency, so several corrections are required to reduce the frequency error. Therefore, the slower the reference the longer it takes to make these corrections. This parameter is under user control via the choice of crystal frequency f_{XCLK} . (See 4.3.3 PLL Circuits and 4.3.6 Programming the PLL.)

Another critical parameter is the external filter network. The PLL modifies the voltage on the VCO by adding or subtracting charge from capacitors in this network. Therefore, the rate at which the voltage changes for a given frequency error (thus change in charge) is proportional to the capacitance. The size of the capacitor also is related to the stability of the PLL. If the capacitor is too small, the PLL cannot make small enough adjustments to the voltage and the system cannot lock. If the capacitor is too large, the PLL may not be able to adjust the voltage in a reasonable time. (See 4.8.3 Choosing a Filter.)

Also important is the operating voltage potential applied to V_{DDA} . The power supply potential alters the characteristics of the PLL. A fixed value is best. Variable supplies, such as batteries, are acceptable if they vary within a known range at very slow speeds. Noise on the power supply is not acceptable, because it causes small frequency errors which continually change the acquisition time of the PLL.

Temperature and processing also can affect acquisition time because the electrical characteristics of the PLL change. The part operates as specified as long as these influences stay within the specified limits. External factors, however, can cause drastic changes in the operation of the PLL. These factors include noise injected into the PLL through the filter capacitor, filter capacitor leakage, stray impedances on the circuit board, and even humidity or circuit board contamination.

4.8.3 Choosing a Filter

As described in 4.8.2 Parametric Influences on Reaction Time, the external filter network is critical to the stability and reaction time of the PLL. The PLL is also dependent on reference frequency and supply voltage.

Figure 4-9 shows two types of filter circuits. In low-cost applications, where stability and reaction time of the PLL are not critical, the three component filter network shown in Figure 4-9 (B) can be replaced by a single capacitor, C_F, as shown in shown in Figure 4-9 (A). Refer to Table 4-5 for recommended filter components at various reference frequencies. For reference frequencies between the values listed in the table, extrapolate to the nearest common capacitor value. In general, a slightly larger capacitor provides more stability at the expense of increased lock time.

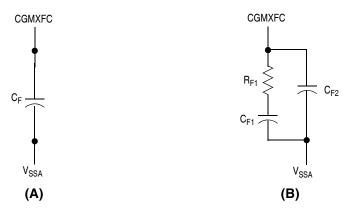


Figure 4-9. PLL Filter

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Clock Generator Module (CGM)

Table 4-5. Example Filter Component Values

f _{RCLK}	C _{F1}	C _{F2}	R _{F1}	C _F
1 MHz	8.2 nF	820 pF	2k	18 nF
2 MHz	4.7 nF	470 pF	2k	6.8 nF
3 MHz	3.3 nF	330 pF	2k	5.6 nF
4 MHz	2.2 nF	220 pF	2k	4.7 nF
5 MHz	1.8 nF	180 pF	2k	3.9 nF
6 MHz	1.5 nF	150 pF	2k	3.3 nF
7 MHz	1.2 nF	120 pF	2k	2.7 nF
8 MHz	1 nF	100 pF	2k	2.2 nF



Low-Voltage Inhibit (LVI)

LVISTOP, LVIPWRD, LVI5OR3, and LVIRSTD are in the configuration register (CONFIG1). See Figure 5-2. Configuration Register 1 (CONFIG1) for details of the LVI's configuration bits. Once an LVI reset occurs, the MCU remains in reset until V_{DD} rises above a voltage, V_{TRIPR} , which causes the MCU to exit reset. See 14.3.2.5 Low-Voltage Inhibit (LVI) Reset for details of the interaction between the SIM and the LVI. The output of the comparator controls the state of the LVIOUT flag in the LVI status register (LVISR).

An LVI reset also drives the RST pin low to provide low-voltage protection to external peripheral devices.

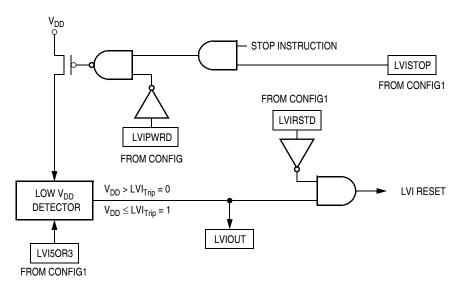


Figure 11-1. LVI Module Block Diagram

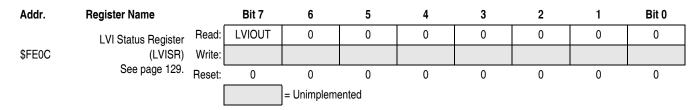


Figure 11-2. LVI I/O Register Summary

11.3.1 Polled LVI Operation

In applications that can operate at V_{DD} levels below the V_{TRIPF} level, software can monitor V_{DD} by polling the LVIOUT bit. In the configuration register, the LVIPWRD bit must be 0 to enable the LVI module, and the LVIRSTD bit must be 1 to disable LVI resets.

11.3.2 Forced Reset Operation

In applications that require V_{DD} to remain above the V_{TRIPF} level, enabling LVI resets allows the LVI module to reset the MCU when V_{DD} falls below the V_{TRIPF} level. In the configuration register, the LVIPWRD and LVIRSTD bits must be cleared to enable the LVI module and to enable LVI resets.

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Input/Output (I/O) Ports

12.5.3 Port C Input Pullup Enable Register

The port C input pullup enable register (PTCPUE) contains a software configurable pullup device for each of the seven port C pins. Each bit is individually configurable and requires that the data direction register, DDRC, bit be configured as an input. Each pullup is automatically and dynamically disabled when a port bit's DDRC is configured for output mode.

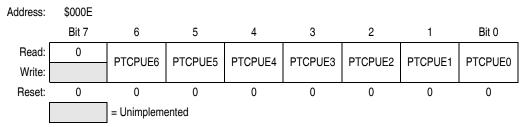


Figure 12-12. Port C Input Pullup Enable Register (PTCPUE)

PTCPUE6-PTCPUE0 — Port C Input Pullup Enable Bits

These writable bits are software programmable to enable pullup devices on an input port bit.

- 1 = Corresponding port C pin configured to have internal pullup
- 0 = Corresponding port C pin internal pullup disconnected

12.6 Port D

Port D is an 8-bit special-function port that shares four of its pins with the serial peripheral interface (SPI) module and four of its pins with two timer interface (TIM1 and TIM2) modules. Port D also has software configurable pullup devices if configured as an input port. PTD0 is shared with the MCLK output.

12.6.1 Port D Data Register

The port D data register (PTD) contains a data latch for each of the eight port D pins.

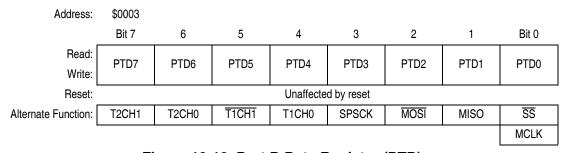


Figure 12-13. Port D Data Register (PTD)

PTD7-PTD0 — Port D Data Bits

These read/write bits are software-programmable. Data direction of each port D pin is under the control of the corresponding bit in data direction register D. Reset has no effect on port D data.

T2CH1 and T2CH0 — Timer 2 Channel I/O Bits

The PTD5/T2CH1-PTD4/T2CH0 pins are the TIM2 input capture/output compare pins. The edge/level select bits, ELSxB:ELSxA, determine whether the PTD7/T2CH1-PTD6/T2CH0 pins are timer channel I/O pins or general-purpose I/O pins. See Chapter 17 Timer Interface Module (TIM1) and Chapter 18 Timer Interface Module (TIM2).

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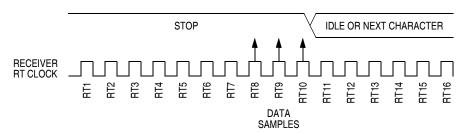


Figure 13-9. Fast Data

For an 8-bit character, data sampling of the stop bit takes the receiver 9 bit times \times 16 RT cycles + 10 RT cycles = 154 RT cycles.

With the misaligned character shown in Figure 13-9, the receiver counts 154 RT cycles at the point when the count of the transmitting device is 10 bit times \times 16 RT cycles = 160 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a fast 8-bit character with no errors is

$$\left| \frac{154 - 160}{154} \right| \times 100 = 3.90\%.$$

For a 9-bit character, data sampling of the stop bit takes the receiver 10 bit times \times 16 RT cycles + 10 RT cycles = 170 RT cycles.

With the misaligned character shown in Figure 13-9, the receiver counts 170 RT cycles at the point when the count of the transmitting device is 11 bit times \times 16 RT cycles = 176 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a fast 9-bit character with no errors is:

$$\left| \frac{170 - 176}{170} \right| \times 100 = 3.53\%.$$

13.4.3.6 Receiver Wakeup

So that the MCU can ignore transmissions intended only for other receivers in multiple-receiver systems, the receiver can be put into a standby state. Setting the receiver wakeup bit, RWU, in SCC2 puts the receiver into a standby state during which receiver interrupts are disabled.

Depending on the state of the WAKE bit in SCC1, either of two conditions on the RxD pin can bring the receiver out of the standby state:

- 1. Address mark An address mark is a 1 in the MSB position of a received character. When the WAKE bit is set, an address mark wakes the receiver from the standby state by clearing the RWU bit. The address mark also sets the ESCI receiver full bit, SCRF. Software can then compare the character containing the address mark to the user-defined address of the receiver. If they are the same, the receiver remains awake and processes the characters that follow. If they are not the same, software can set the RWU bit and put the receiver back into the standby state.
- Idle input line condition When the WAKE bit is clear, an idle character on the RxD pin wakes the
 receiver from the standby state by clearing the RWU bit. The idle character that wakes the receiver
 does not set the receiver idle bit, IDLE, or the ESCI receiver full bit, SCRF. The idle line type bit,
 ILTY, determines whether the receiver begins counting 1s as idle character bits after the start bit
 or after the stop bit.

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System Integration Module (SIM)

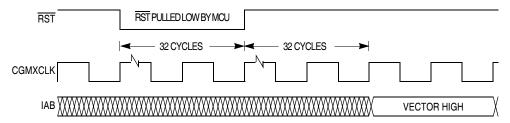


Figure 14-5. Internal Reset Timing

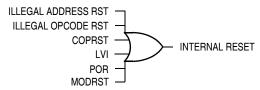


Figure 14-6. Sources of Internal Reset

Table 14-2. Reset Recovery

Reset Recovery Type	Actual Number of Cycles
POR/LVI	4163 (4096 + 64 + 3)
All others	67 (64 + 3)

14.3.2.1 Power-On Reset

When power is first applied to the MCU, the power-on reset module (POR) generates a pulse to indicate that power-on has occurred. The external reset pin (RST) is held low while the SIM counter counts out 4096 + 32 CGMXCLK cycles. Thirty-two CGMXCLK cycles later, the CPU and memories are released from reset to allow the reset vector sequence to occur.

At power-on, these events occur:

- A POR pulse is generated.
- The internal reset signal is asserted.
- The SIM enables CGMOUT.
- Internal clocks to the CPU and modules are held inactive for 4096 CGMXCLK cycles to allow stabilization of the oscillator.
- The RST pin is driven low during the oscillator stabilization time.
- The POR bit of the SIM reset status register (SRSR) is set.

14.3.2.2 Computer Operating Properly (COP) Reset

An input to the SIM is reserved for the COP reset signal. The overflow of the COP counter causes an internal reset and sets the COP bit in the SIM reset status register (SRSR) if the COPD bit in the CONFIG1 register is cleared. The SIM actively pulls down the RST pin for all internal reset sources.

The COP module is disabled if the \overline{RST} pin or the \overline{IRQ} pin is held at V_{TST} while the MCU is in monitor mode. During a break state, V_{TST} on the \overline{RST} pin disables the COP module.

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If more than one interrupt is pending at the end of an instruction execution, the highest priority interrupt is serviced first. Figure 14-11 demonstrates what happens when two interrupts are pending. If an interrupt is pending upon exit from the original interrupt service routine, the pending interrupt is serviced before the LDA instruction is executed.

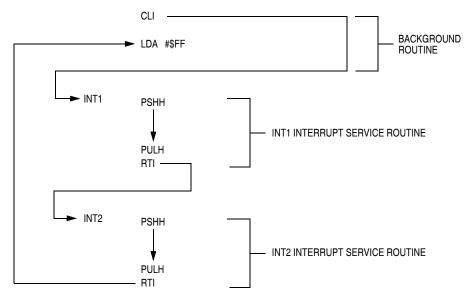


Figure 14-11. Interrupt Recognition Example

The LDA opcode is prefetched by both the INT1 and INT2 RTI instructions. However, in the case of the INT1 RTI prefetch, this is a redundant operation.

NOTE

To maintain compatibility with the M6805 Family, the H register is not pushed on the stack during interrupt entry. If the interrupt service routine modifies the H register or uses the indexed addressing mode, software should save the H register and then restore it prior to exiting the routine.

14.5.1.2 SWI Instruction

The SWI instruction is a non-maskable instruction that causes an interrupt regardless of the state of the interrupt mask (I bit) in the condition code register.

NOTE

A software interrupt pushes PC onto the stack. A software interrupt does not push PC – 1, as a hardware interrupt does.

14.5.1.3 Interrupt Status Registers

The flags in the interrupt status registers identify maskable interrupt sources. Table 14-3 summarizes the interrupt sources, hardware flag bits, hardware interrupt mask bits, interrupt status register flags, interrupt priority, and exception vectors. The interrupt status registers can be useful for debugging.

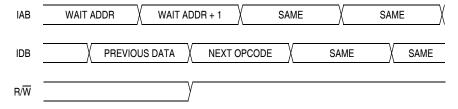


14.6.1 Wait Mode

In wait mode, the CPU clocks are inactive while the peripheral clocks continue to run. Figure 14-16 shows the timing for wait mode entry.

A module that is active during wait mode can wakeup the CPU with an interrupt if the interrupt is enabled. Stacking for the interrupt begins one cycle after the WAIT instruction during which the interrupt occurred. In wait mode, the CPU clocks are inactive. Refer to the wait mode subsection of each module to see if the module is active or inactive in wait mode. Some modules can be programmed to be active in wait mode.

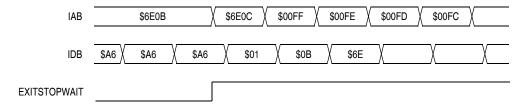
Wait mode also can be exited by a reset or break. A break interrupt during wait mode sets the SIM break stop/wait bit, SBSW, in the SIM break status register (BSR). If the COP disable bit, COPD, in the CONFIG1 register is 0, then the computer operating properly module (COP) is enabled and remains active in wait mode.



Note: Previous data can be operand data or the WAIT opcode, depending on the last instruction.

Figure 14-16. Wait Mode Entry Timing

Figure 14-17 and Figure 14-18 show the timing for WAIT recovery.



Note: EXITSTOPWAIT = RST pin, CPU interrupt, or break interrupt

Figure 14-17. Wait Recovery from Interrupt or Break

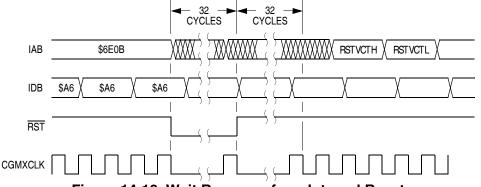


Figure 14-18. Wait Recovery from Internal Reset

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Timer Interface Module (TIM2)

18.7 I/O Signals

Port D shares two of its pins with the TIM2. Port F shares four of its pins with the TIM2. PTD6/T2CH0 is an external clock input to the TIM2 prescaler. The six TIM2 channel I/O pins are PTD6/T2CH0, PTD7/T2CH1, PTF4/T2CH2, PTF5/T2CH3, PTF6/T2CH4, and PTF7/T2CH5.

18.7.1 TIM2 Clock Pin (T2CH0)

T2CH0 is an external clock input that can be the clock source for the TIM2 counter instead of the prescaled internal bus clock. Select the T2CH0 input by writing 1s to the three prescaler select bits, PS[2:0]. (See 18.8.1 TIM2 Status and Control Register.) The minimum TCLK pulse width is specified in 20.14 Timer Interface Module Characteristics. The maximum TCLK frequency is the least: 4 MHz or bus frequency ÷ 2.

When the PTD6/T2CH0 pin is the TIM2 clock input, it is an input regardless of the state of the DDRD6 bit in data direction register D.

18.7.2 TIM2 Channel I/O Pins (T2CH5:T2CH2 and T2CH1:T2CH0)

Each channel I/O pin is programmable independently as an input capture pin or an output compare pin. T2CH0, T2CH2, and T2CH4 can be configured as buffered output compare or buffered PWM pins.

18.8 I/O Registers

These I/O registers control and monitor TIM2 operation:

- TIM2 status and control register (T2SC)
- TIM2 counter registers (T2CNTH:T2CNTL)
- TIM2 counter modulo registers (T2MODH:T2MODL)
- TIM2 channel status and control registers (T2SC0, T2SC1, T2SC2, T2SC3, T2SC4, and T2SC5)
- TIM2 channel registers (T2CH0H:T2CH0L, T2CH1H:T2CH1L, T2CH2H:T2CH2L, T2CH3H:T2CH3L, T2CH4H:T2CH4L, and T2CH5H:T2CH5L)

18.8.1 TIM2 Status and Control Register

The TIM2 status and control register:

- Enables TIM2 overflow interrupts
- Flags TIM2 overflows
- Stops the TIM2 counter
- Resets the TIM2 counter
- Prescales the TIM2 counter clock

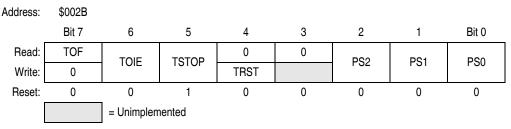


Figure 18-5. TIM2 Status and Control Register (T2SC)

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Timer Interface Module (TIM2)

CHxMAX — Channel x Maximum Duty Cycle Bit

When the TOVx bit is at a 1 and clear output on compare is selected, setting the CHxMAX bit forces the duty cycle of buffered and unbuffered PWM signals to 100%. As Figure 18-9 shows, the CHxMAX bit takes effect in the cycle after it is set or cleared. The output stays at 100% duty cycle level until the cycle after CHxMAX is cleared.

NOTE

The 100% PWM duty cycle is defined as a continuous high level if the PWM polarity is 1 and a continuous low level if the PWM polarity is 0. Conversely, a 0% PWM duty cycle is defined as a continuous low level if the PWM polarity is 1 and a continuous high level if the PWM polarity is 0.

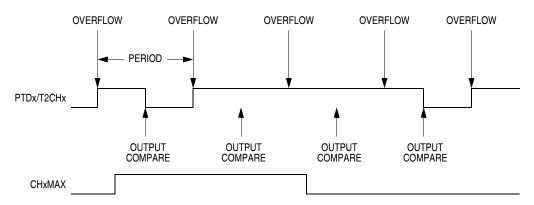


Figure 18-9. CHxMAX Latency

18.8.5 TIM2 Channel Registers

These read/write registers contain the captured TIM2 counter value of the input capture function or the output compare value of the output compare function. The state of the TIM2 channel registers after reset is unknown.

In input capture mode (MSxB:MSxA = 0:0), reading the high byte of the TIM2 channel x registers (T2CHxH) inhibits input captures until the low byte (T2CHxL) is read.

In output compare mode (MSxB:MSxA \neq 0:0), writing to the high byte of the TIM2 channel x registers (T2CHxH) inhibits output compares until the low byte (T2CHxL) is written.

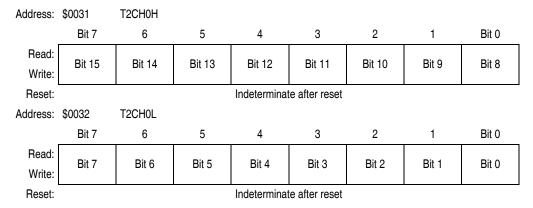


Figure 18-10. TIM2 Channel Registers (T2CH0H/L:T2CH5H/L) (Sheet 1 of 3)

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20.15 Memory Characteristics

Characteristic	Symbol	Min	Тур	Max	Unit
RAM data retention voltage	V _{RDR}	1.3	_	_	٧
FLASH program bus clock frequency	_	1	_	_	MHz
FLASH read bus clock frequency	f _{Read} ⁽¹⁾	0	_	8 M	Hz
FLASH page erase time <1 k cycles >1 k cycles	t _{Erase}	0.9 3.6	1 4	1.1 5.5	ms
FLASH mass erase time	t _{MErase}	4	_	_	ms
FLASH PGM/ERASE to HVEN setup time	t _{NVS}	10	_	_	μS
FLASH high-voltage hold time	t _{NVH}	5	_	_	μS
FLASH high-voltage hold time (mass erase)	t _{NVHL}	100	_	_	μS
FLASH program hold time	t _{PGS}	5	_	_	μS
FLASH program time	t _{PROG}	30	_	40	μS
FLASH return to read time	t _{RCV} ⁽²⁾	1	_	_	μS
FLASH cumulative program HV period	t _{HV} ⁽³⁾	_	_	4	ms
FLASH endurance ⁽⁴⁾	_	10 k	100 k	_	Cycles
FLASH data retention time ⁽⁵⁾	_	15	100	_	Years

- 1. $f_{\mbox{\scriptsize Read}}$ is defined as the frequency range for which the FLASH memory can be read.
- 2. t_{RCV} is defined as the time it needs before the FLASH can be read after turning off the high voltage charge pump, by clear-
- 3. t_{HV} is defined as the cumulative high voltage programming time to the same row before next erase.
- t_{HV} must satisfy this condition: t_{NVS} + t_{NVH} + t_{PGS} + (t_{PROG} x 32) ≤ t_{HV} maximum.

 4. Typical endurance was evaluated for this product family. For additional information on how Freescale Semiconductor defines Typical Endurance, please refer to Engineering Bulletin EB619.
- 5. Typical data retention values are based on intrinsic capability of the technology measured at high temperature and de-rated to 25°C using the Arrhenius equation. For additional information on how Freescale Semiconductor defines Typical Data Retention, please refer to Engineering Bulletin EB618.



NOTES:

- 1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M-1994.
- 2. CONTROLLING DIMENSION: MILLIMETER.
- 3. DATUM PLANE AB IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
- 4. DATUMS T, U, AND Z TO BE DETERMINED AT DATUM PLANE AB.



DIMENSIONS TO BE DETERMINED AT SEATING PLANE AC.



DIMENSIONS DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.250 PER SIDE. DIMENSIONS DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE AB.



THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.350.

8. MINIMUM SOLDER PLATE THICKNESS SHALL BE 0.0076.



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