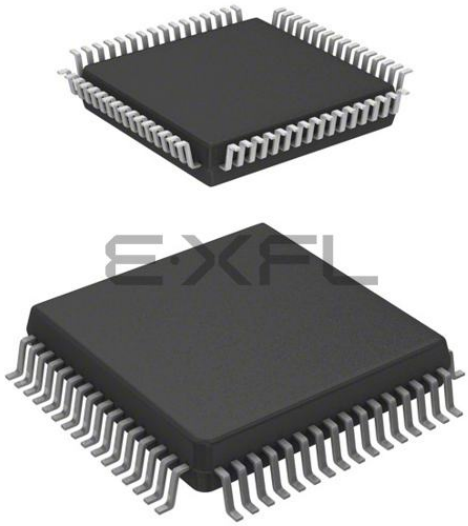

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Details	
Product Status	Obsolete
Core Processor	HC08
Core Size	8-Bit
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Connectivity	SCI, SPI
Peripherals	LVD, POR, PWM
Number of I/O	53
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 24x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	64-QFP
Supplier Device Package	64-QFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc908gr32avfue

MC68HC908GR60A
MC68HC908GR48A
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Data Sheet

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Chapter 15
Serial Peripheral Interface (SPI) Module

Chapter 16
Timebase Module (TBM)

Addr.	Register Name		Bit 7	6	5	4	3	2	1	Bit 0
$0459	TIM2 Channel 3 Status and Control Register (T2SC3) See page 255.	Read:	CH3F	CH3IE	0	MS3A	ELS3B	ELS3A	TOV3	CH3MAX
		Write:	0							
		Reset:	0	0	0	0	0	0	0	0
$045A	TIM2 Channel 3 Register High (T2CH3H) See page 258.	Read:	Bit 15	14	13	12	11	10	9	Bit 8
		Write:								
		Reset:	Indeterminate after reset							
$045B	TIM2 Channel 3 Register Low (T2CH3L) See page 258.	Read:	Bit 7	6	5	4	3	2	1	Bit 0
		Write:								
		Reset:	Indeterminate after reset							
$045C	TIM2 Channel 4 Status and Control Register (T2SC4) See page 255.	Read:	CH4F	CH4IE	MS4B	MS4A	ELS4B	ELS4A	TOV4	CH4MAX
		Write:	0							
		Reset:	0	0	0	0	0	0	0	0
$045D	TIM2 Channel 4 Register High (T2CH4H) See page 258.	Read:	Bit 15	14	13	12	11	10	9	Bit 8
		Write:								
		Reset:	Indeterminate after reset							
$045E	TIM2 Channel 4 Register Low (T2CH4L) See page 258.	Read:	Bit 7	6	5	4	3	2	1	Bit 0
		Write:								
		Reset:	Indeterminate after reset							
$045F	TIM2 Channel 5 Status and Control Register (T2SC5) See page 255.	Read:	CH5F	CH5IE	0	MS5A	ELS5B	ELS5A	TOV 5	CH5MAX
		Write:	0							
		Reset:	0	0	0	0	0	0	0	0
$0460	TIM2 Channel 5 Register High (T2CH5H) See page 258.	Read:	Bit 15	14	13	12	11	10	9	Bit 8
		Write:								
		Reset:	Indeterminate after reset							
$0461	TIM2 Channel 5 Register Low (T2CH5L) See page 258.	Read:	Bit 7	6	5	4	3	2	1	Bit 0
		Write:								
		Reset:	Indeterminate after reset							
$FE00	Break Status Register (BSR) See page 199.	Read:	R	R	R	R	R	R	SBSW	R
		Write:							NOTE 1	
		Reset:	0	0	0	0	0	0	0	0

1. Writing a 0 clears SBSW.

$FE01	SIM Reset Status Register (SRSR) See page 199.	Read:	POR	PIN	COP	ILOP	ILAD	MODRST	LVI	0
		Write:								
		POR:	1	0	0	0	0	0	0	0
$FE02	Reserved	Read:	R	R	R	R	R	R	R	R
		Write:								
		Reset:	0	0	0	0	0	0	0	0

= Unimplemented R = Reserved U = Unaffected

Figure 2-2. Control, Status, and Data Registers (Sheet 7 of 9)

Addr.	Register Name		Bit 7	6	5	4	3	2	1	Bit 0
$FF80	FLASH-1 Block Protect Register (FL1BPR)(1) See page 43.	Read: Write:	BPR7	BPR6	BPR5	BPR4	BPR3	BPR2	BPR1	BPR0
		Reset:	Unaffected by reset							
$FF81	FLASH-2 Block Protect Register (FL2BPR)(1) See page 51.	Read: Write:	BPR7	BPR6	BPR5	BPR4	BPR3	BPR2	BPR1	BPR0
		Reset:	Unaffected by reset							
$FF88	FLASH-1 Control Register (FL1CR) See page 42.	Read:	0	0	0	0	HVEN	MASS	ERASE	PGM
		Write:								
		Reset:	0	0	0	0	0	0	0	0
$FFFF	COP Control Register (COPCTL) See page 95.	Read:	Low byte of reset vector							
		Write:	Writing clears COP counter (any value)							
		Reset:	Unaffected by reset							

1. Non-volatile FLASH register

= Unimplemented R = Reserved U = Unaffected

Figure 2-2. Control, Status, and Data Registers (Sheet 9 of 9)

Table 2-1. Vector Addresses

Vector Priority	Vector	Address	Vector
Lowest ↓	IF24	$FFCC	TIM2 Channel 5 Vector (High)
		$FFCD	TIM2 Channel 5 Vector (Low)
	IF23	$FFCE	TIM2 Channel 4 Vector (High)
		$FFCF	TIM2 Channel 4 Vector (Low)
	IF22	$FFD0	TIM2 Channel 3 Vector (High)
		$FFD1	TIM2 Channel 3 Vector (Low)
	IF21	$FFD2	TIM2 Channel 2 Vector (High)
		$FFD3	TIM2 Channel 2 Vector (Low)
	IF20 ↓ IF17	$FFD4 ↓ $FFDB	Reserved
	IF16	$FFDC	Timebase Vector (High)
		$FFDD	Timebase Vector (Low)
	IF15	$FFDE	ADC Conversion Complete Vector (High)
		$FFDF	ADC Conversion Complete Vector (Low)

Continued on next page

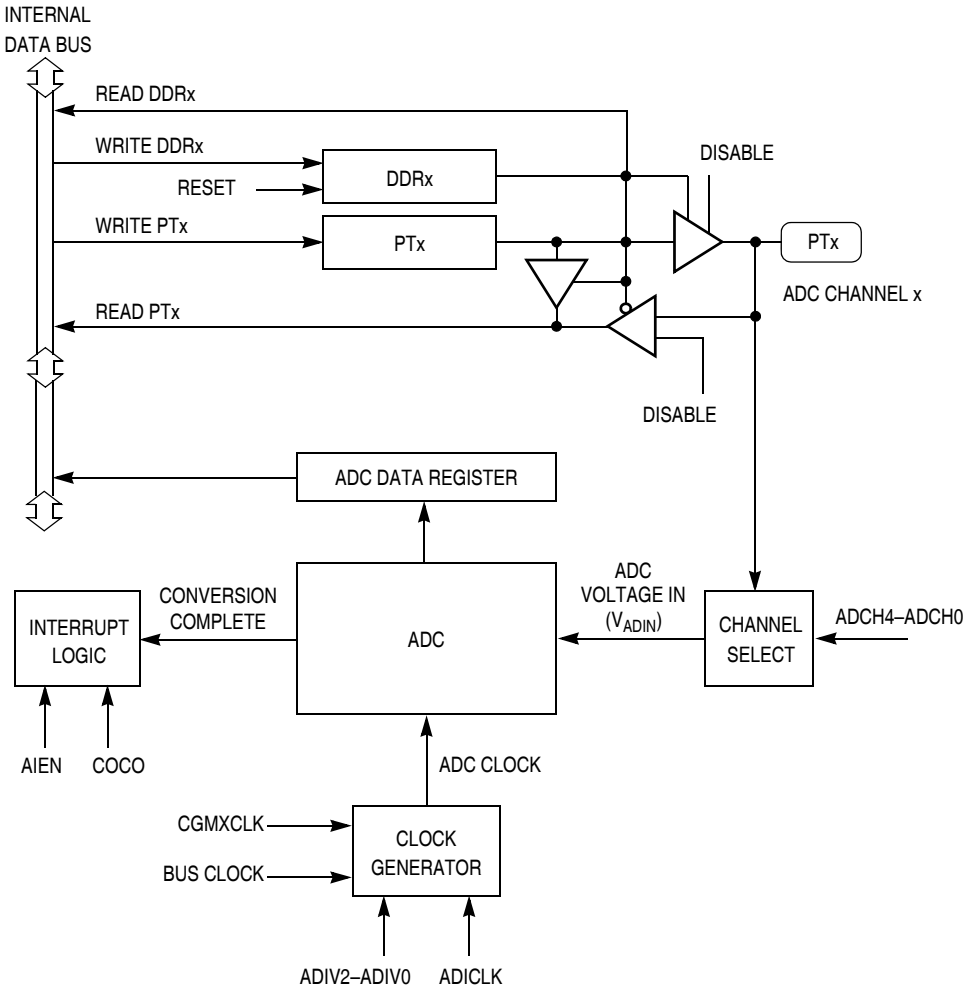


Figure 3-2. ADC Block Diagram

3.3.2 Voltage Conversion

When the input voltage to the ADC equals V_{REFH}, the ADC converts the signal to $3FF (full scale). If the input voltage equals V_{REFL}, the ADC converts it to $000. Input voltages between V_{REFH} and V_{REFL} are a straight-line linear conversion.

NOTE

The ADC input voltage must always be greater than V_{SSAD} and less than V_{DDAD}.

Connect the V_{DDAD} pin to the same voltage potential as the V_{DD} pin, and connect the V_{SSAD} pin to the same voltage potential as the V_{SS} pin.

The V_{DDAD} pin should be routed carefully for maximum noise immunity.

4.3.1 Crystal Oscillator Circuit

The crystal oscillator circuit consists of an inverting amplifier and an external crystal. The OSC1 pin is the input to the amplifier and the OSC2 pin is the output. The SIMOSCEN signal from the system integration module (SIM) or the OSCENINSTOP bit in the CONFIG register enable the crystal oscillator circuit.

The CGMXCLK signal is the output of the crystal oscillator circuit and runs at a rate equal to the crystal frequency. CGMXCLK is then buffered to produce CGMRCLK, the PLL reference clock.

CGMXCLK can be used by other modules which require precise timing for operation. The duty cycle of CGMXCLK is not guaranteed to be 50% and depends on external factors, including the crystal and related external components. An externally generated clock also can feed the OSC1 pin of the crystal oscillator circuit. Connect the external clock to the OSC1 pin and let the OSC2 pin float.

4.3.2 Phase-Locked Loop Circuit (PLL)

The PLL is a frequency generator that can operate in either acquisition mode or tracking mode, depending on the accuracy of the output frequency. The PLL can change between acquisition and tracking modes either automatically or manually.

4.3.3 PLL Circuits

The PLL consists of these circuits:

- Voltage-controlled oscillator (VCO)
- Modulo VCO frequency divider
- Phase detector
- Loop filter
- Lock detector

The operating range of the VCO is programmable for a wide range of frequencies and for maximum immunity to external noise, including supply and CGMXFC noise. The VCO frequency is bound to a range from roughly one-half to twice the center-of-range frequency, f_{VRS}. Modulating the voltage on the CGMXFC pin changes the frequency within this range. By design, f_{VRS} is equal to the nominal center-of-range frequency, f_{NOM}, (71.4 kHz) times a linear factor, L, and a power-of-two factor, E, or $(L \times 2^E)f_{NOM}$.

CGMRCLK is the PLL reference clock, a buffered version of CGMXCLK. CGMRCLK runs at a frequency, f_{RCLK}. The VCO's output clock, CGMVCLK, running at a frequency, f_{VCLK}, is fed back through a programmable modulo divider. The modulo divider reduces the VCO clock by a factor, N. The dividers output is the VCO feedback clock, CGMVDV, running at a frequency, $f_{VDV} = f_{VCLK}/(N)$. (For more information, see 4.3.6 Programming the PLL.)

The phase detector then compares the VCO feedback clock, CGMVDV, with the final reference clock, CGMRDV. A correction pulse is generated based on the phase difference between the two signals. The loop filter then slightly alters the DC voltage on the external capacitor connected to CGMXFC based on the width and direction of the correction pulse. The filter can make fast or slow corrections depending on its mode, described in 4.3.4 Acquisition and Tracking Modes. The value of the external capacitor and the reference frequency determines the speed of the corrections and the stability of the PLL.

The lock detector compares the frequencies of the VCO feedback clock, CGMVDV, and the reference clock, CGMRCLK. Therefore, the speed of the lock detector is directly proportional to the reference frequency, f_{RCLK}. The circuit determines the mode of the PLL and the lock condition based on this comparison.

The following conditions apply when in manual mode:

- $\overline{ACQ}$ is a writable control bit that controls the mode of the filter. Before turning on the PLL in manual mode, the $\overline{ACQ}$ bit must be clear.
- Before entering tracking mode ($\overline{ACQ}$ = 1), software must wait a given time, t_{ACQ} (See 4.8 Acquisition/Lock Time Specifications.), after turning on the PLL by setting PLLON in the PLL control register (PCTL).
- Software must wait a given time, t_{AL}, after entering tracking mode before selecting the PLL as the clock source to CGMOUT (BCS = 1).
- The LOCK bit is disabled.
- CPU interrupts from the CGM are disabled.

4.3.6 Programming the PLL

Use the following procedure to program the PLL. For reference, the variables used and their meaning are shown in Table 4-1.

Table 4-1. Variable Definitions

Variable	Definition
f_{BUSDES}	Desired bus clock frequency
$f_{VCLKDES}$	Desired VCO clock frequency
f_{RCLK}	Chosen reference crystal frequency
f_{VCLK}	Calculated VCO clock frequency
f_{BUS}	Calculated bus clock frequency
f_{NOM}	Nominal VCO center frequency
f_{VRS}	Programmed VCO center frequency

NOTE

The round function in the following equations means that the real number should be rounded to the nearest integer number.

1. Choose the desired bus frequency, f_{BUSDES}.
2. Calculate the desired VCO frequency (four times the desired bus frequency).

$$f_{VCLKDES} = 4 \times f_{BUSDES}$$

3. Choose a practical PLL (crystal) reference frequency, f_{RCLK}. Typically, the reference crystal is 1–8 MHz.

   Frequency errors to the PLL are corrected at a rate of f_{RCLK}.
   For stability and lock time reduction, this rate must be as fast as possible. The VCO frequency must be an integer multiple of this rate. The relationship between the VCO frequency, f_{VCLK}, and the reference frequency, f_{RCLK}, is:

$$f_{VCLK} = (N)\ (f_{RCLK})$$

   N, the range multiplier, must be an integer.

In cases where desired bus frequency has some tolerance, choose f_{RCLK} to a value determined either by other module requirements (such as modules which are clocked by CGMXCLK), cost requirements, or ideally, as high as the specified range allows. See Chapter 20 Electrical Specifications. After choosing N, the actual bus frequency can be determined using equation in 2 above.

4. Select a VCO frequency multiplier, N.

$$N = \text{round}\left(\frac{f_{VCLKDES}}{f_{RCLK}}\right)$$

5. Calculate and verify the adequacy of the VCO and bus frequencies f_{VCLK} and f_{BUS}.

$$f_{VCLK} = (N) \times f_{RCLK}$$
$$f_{BUS} = (f_{VCLK})/4$$

6. Select the VCO's power-of-two range multiplier E, according to Table 4-2.

Table 4-2. Power-of-Two Range Selectors

Frequency Range	E
$0 < f_{VCLK} \leq 8$ MHz	0
8 MHz$< f_{VCLK} \leq 16$ MHz	1
16 MHz$< f_{VCLK} \leq 32$ MHz	$2^{(1)}$

1. Do not program E to a value of 3.

7. Select a VCO linear range multiplier, L, where f_{NOM} = 71.4 kHz

$$L = \text{Round}\left(\frac{f_{VCLK}}{2^E \times f_{NOM}}\right)$$

8. Calculate and verify the adequacy of the VCO programmed center-of-range frequency, f_{VRS}. The center-of-range frequency is the midpoint between the minimum and maximum frequencies attainable by the PLL.

$$f_{VRS} = (L \times 2^E)\, f_{NOM}$$

9. For proper operation,

$$|f_{VRS} - f_{VCLK}| \leq \frac{f_{NOM} \times 2^E}{2}$$

10. Verify the choice of N, E, and L by comparing f_{VCLK} to f_{VRS} and $f_{VCLKDES}$. For proper operation, f_{VCLK} must be within the application's tolerance of $f_{VCLKDES}$, and f_{VRS} must be as close as possible to f_{VCLK}.

NOTE

Exceeding the recommended maximum bus frequency or VCO frequency can crash the MCU.

4.7.2 Stop Mode

If the OSCENINSTOP bit in the CONFIG2 register is cleared (default), then the STOP instruction disables the CGM (oscillator and phase locked loop) and holds low all CGM outputs (CGMXCLK, CGMOUT, and CGMINT).

If the OSCENINSTOP bit in the CONFIG2 register is set, then the phase locked loop is shut off but the oscillator will continue to operate in stop mode.

4.7.3 CGM During Break Interrupts

The system integration module (SIM) controls whether status bits in other modules can be cleared during the break state. The BCFE bit in the SIM break flag control register (SBFCR) enables software to clear status bits during the break state. (See 14.7.3 Break Flag Control Register.)

To allow software to clear status bits during a break interrupt, write a 1 to the BCFE bit. If a status bit is cleared during the break state, it remains cleared when the MCU exits the break state.

To protect the PLLF bit during the break state, write a 0 to the BCFE bit. With BCFE at 0 (its default state), software can read and write the PLL control register during the break state without affecting the PLLF bit.

4.8 Acquisition/Lock Time Specifications

The acquisition and lock times of the PLL are, in many applications, the most critical PLL design parameters. Proper design and use of the PLL ensures the highest stability and lowest acquisition/lock times.

4.8.1 Acquisition/Lock Time Definitions

Typical control systems refer to the acquisition time or lock time as the reaction time, within specified tolerances, of the system to a step input. In a PLL, the step input occurs when the PLL is turned on or when it suffers a noise hit. The tolerance is usually specified as a percent of the step input or when the output settles to the desired value plus or minus a percent of the frequency change. Therefore, the reaction time is constant in this definition, regardless of the size of the step input. For example, consider a system with a 5 percent acquisition time tolerance. If a command instructs the system to change from 0 Hz to 1 MHz, the acquisition time is the time taken for the frequency to reach 1 MHz ±50 kHz. Fifty kHz = 5% of the 1-MHz step input. If the system is operating at 1 MHz and suffers a –100-kHz noise hit, the acquisition time is the time taken to return from 900 kHz to 1 MHz ±5 kHz. Five kHz = 5% of the 100-kHz step input.

Other systems refer to acquisition and lock times as the time the system takes to reduce the error between the actual output and the desired output to within specified tolerances. Therefore, the acquisition or lock time varies according to the original error in the output. Minor errors may not even be registered. Typical PLL applications prefer to use this definition because the system requires the output frequency to be within a certain tolerance of the desired frequency regardless of the size of the initial error.

4.8.2 Parametric Influences on Reaction Time

Acquisition and lock times are designed to be as short as possible while still providing the highest possible stability. These reaction times are not constant, however. Many factors directly and indirectly affect the acquisition time.

Address: $001E

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	MCLKSEL	MCLK1	MCLK0	R	TMBCLKSEL	OSCENINSTOP	SCIBDSRC
Write:								
Reset:	0	0	0	0	0	0	0	1

= Unimplemented R = Reserved

Figure 5-1. Configuration Register 2 (CONFIG2)

MCLKSEL — MCLK Source Select Bit

1 = Crystal frequency
0 = Bus frequency

MCLK1 and MCLK0 — MCLK Output Select Bits

Setting the MCLK1 and MCLK0 bits enables the PTD0/$\overline{SS}$ pin to be used as a MCLK output clock. Once configured for MCLK, the PTD data direction register for PTD0 is used to enable and disable the MCLK output. See Table 5-1 for MCLK options.

Table 5-1. MCLK Output Select

MCLK1	MCLK0	MCLK Frequency
0	0	MCLK not enabled
0	1	Clock
1	0	Clock divided by 2
1	1	Clock divided by 4

TMBCLKSEL— Timebase Clock Select Bit

TMBCLKSEL enables an extra divide-by-128 prescaler in the timebase module. Setting this bit enables the extra prescaler and clearing this bit disables it. See Chapter 16 Timebase Module (TBM) for a more detailed description of the external clock operation.

1 = Enables extra divide-by-128 prescaler in timebase module
0 = Disables extra divide-by-128 prescaler in timebase module

OSCENINSTOP — Oscillator Enable In Stop Mode Bit

OSCENINSTOP, when set, will enable the oscillator to continue to generate clocks in stop mode. See Chapter 4 Clock Generator Module (CGM). This function is used to keep the timebase running while the rest of the MCU stops. See Chapter 16 Timebase Module (TBM). When clear, the oscillator will cease to generate clocks while in stop mode. The default state for this option is clear, disabling the oscillator in stop mode.

1 = Oscillator enabled during stop mode
0 = Oscillator disabled during stop mode (default)

SCIBDSRC — SCI Baud Rate Clock Source Bit

SCIBDSRC controls the clock source used for the serial communications interface (SCI). The setting of this bit affects the frequency at which the SCI operates.See Chapter 13 Enhanced Serial Communications Interface (ESCI) Module.

1 = Internal data bus clock used as clock source for SCI (default)
0 = External oscillator used as clock source for SCI

Z — Zero Flag

The CPU sets the zero flag when an arithmetic operation, logic operation, or data manipulation produces a result of $00.

1 = Zero result
0 = Non-zero result

C — Carry/Borrow Flag

The CPU sets the carry/borrow flag when an addition operation produces a carry out of bit 7 of the accumulator or when a subtraction operation requires a borrow. Some instructions — such as bit test and branch, shift, and rotate — also clear or set the carry/borrow flag.

1 = Carry out of bit 7
0 = No carry out of bit 7

7.4 Arithmetic/Logic Unit (ALU)

The ALU performs the arithmetic and logic operations defined by the instruction set.

Refer to the *CPU08 Reference Manual* (document order number CPU08RM/AD) for a description of the instructions and addressing modes and more detail about the architecture of the CPU.

7.5 Low-Power Modes

The WAIT and STOP instructions put the MCU in low power-consumption standby modes.

7.5.1 Wait Mode

The WAIT instruction:

- Clears the interrupt mask (I bit) in the condition code register, enabling interrupts. After exit from wait mode by interrupt, the I bit remains clear. After exit by reset, the I bit is set.
- Disables the CPU clock

7.5.2 Stop Mode

The STOP instruction:

- Clears the interrupt mask (I bit) in the condition code register, enabling external interrupts. After exit from stop mode by external interrupt, the I bit remains clear. After exit by reset, the I bit is set.
- Disables the CPU clock

After exiting stop mode, the CPU clock begins running after the oscillator stabilization delay.

7.6 CPU During Break Interrupts

If a break module is present on the MCU, the CPU starts a break interrupt by:

- Loading the instruction register with the SWI instruction
- Loading the program counter with $FFFC:$FFFD or with $FEFC:$FEFD in monitor mode

The break interrupt begins after completion of the CPU instruction in progress. If the break address register match occurs on the last cycle of a CPU instruction, the break interrupt begins immediately.

A return-from-interrupt instruction (RTI) in the break routine ends the break interrupt and returns the MCU to normal operation if the break interrupt has been deasserted.

Chapter 9
Keyboard Interrupt Module (KBI)

9.1 Introduction

The keyboard interrupt module (KBI) provides eight independently maskable external interrupts which are accessible via PTA0–PTA7. When a port pin is enabled for keyboard interrupt function, an internal pullup/pulldown device is also enabled on the pin.

9.2 Features

Features include:

- Eight keyboard interrupt pins with separate keyboard interrupt enable bits and one keyboard interrupt mask
- Hysteresis buffers
- Programmable edge-only or edge- and level- interrupt sensitivity
- Edge detect programmable for rising or falling edges
- Level detect programmable for high or low levels
- Exit from low-power modes
- Pullup/pulldown device automatically configured based on polarity of edge/level selection

9.3 Functional Description

Writing to the KBIE7–KBIE0 bits in the keyboard interrupt enable register independently enables or disables each port A pin as a keyboard interrupt pin. Enabling a keyboard interrupt pin also enables its internal pullup/pulldown device. On falling edge or low level selection a pullup device is configured. On rising edge or high level selection a pulldown device is configured.

- A falling edge is detected when an enabled keyboard input signal is seen as a 1 (the deasserted level) during one bus cycle and then a 0 (the asserted level) during the next cycle.
- A rising edge is detected when the input signal is seen as a 0 during one bus cycle and then a 1 during the next cycle.

A keyboard interrupt is latched when one or more keyboard pins are asserted. The MODEK bit in the keyboard status and control register controls the triggering mode of the keyboard interrupt.

13.3 Pin Name Conventions

The generic names of the ESCI input/output (I/O) pins are:

- RxD (receive data)
- TxD (transmit data)

ESCI I/O lines are implemented by sharing parallel I/O port pins. The full name of an ESCI input or output reflects the name of the shared port pin. Table 13-1 shows the full names and the generic names of the ESCI I/O pins. The generic pin names appear in the text of this section.

Table 13-1. Pin Name Conventions

Generic Pin Names	RxD	TxD
Full Pin Names	PTE1/RxD	PTE0/TxD

13.4 Functional Description

Figure 13-3 shows the structure of the ESCI module. The ESCI allows full-duplex, asynchronous, NRZ serial communication between the MCU and remote devices, including other MCUs. The transmitter and receiver of the ESCI operate independently, although they use the same baud rate generator. During normal operation, the CPU monitors the status of the ESCI, writes the data to be transmitted, and processes received data.

The baud rate clock source for the ESCI can be selected via the configuration bit, SCIBDSRC, of the CONFIG2 register ($001E)

For reference, a summary of the ESCI module input/output registers is provided in Figure 13-4.

13.4.1 Data Format

The SCI uses the standard non-return-to-zero mark/space data format illustrated in Figure 13-2.

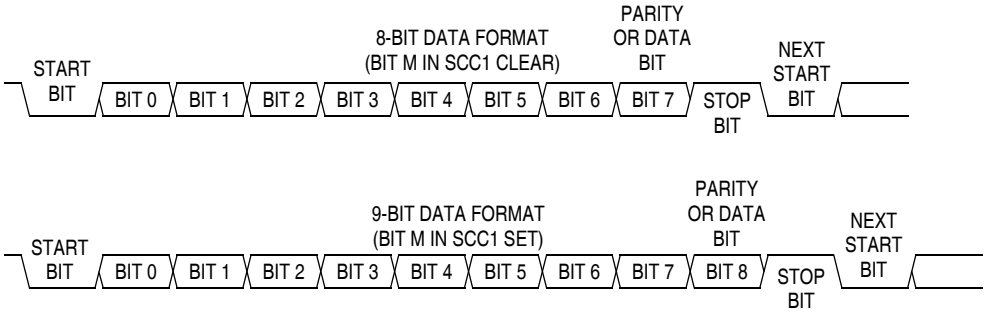


Figure 13-2. SCI Data Formats

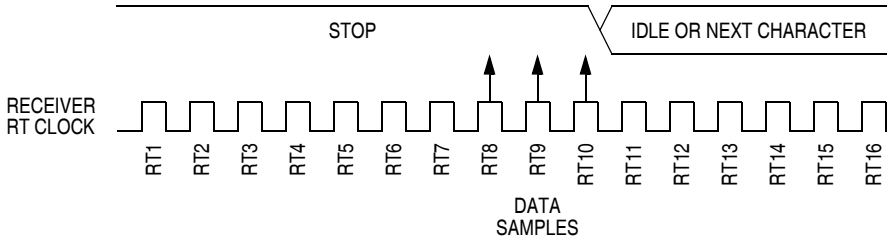


Figure 13-9. Fast Data

For an 8-bit character, data sampling of the stop bit takes the receiver 9 bit times × 16 RT cycles + 10 RT cycles = 154 RT cycles.

With the misaligned character shown in Figure 13-9, the receiver counts 154 RT cycles at the point when the count of the transmitting device is 10 bit times × 16 RT cycles = 160 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a fast 8-bit character with no errors is

$$\left|\frac{154 - 160}{154}\right| \times 100 = 3.90\%.$$

For a 9-bit character, data sampling of the stop bit takes the receiver 10 bit times × 16 RT cycles + 10 RT cycles = 170 RT cycles.

With the misaligned character shown in Figure 13-9, the receiver counts 170 RT cycles at the point when the count of the transmitting device is 11 bit times × 16 RT cycles = 176 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a fast 9-bit character with no errors is:

$$\left|\frac{170 - 176}{170}\right| \times 100 = 3.53\%.$$

13.4.3.6 Receiver Wakeup

So that the MCU can ignore transmissions intended only for other receivers in multiple-receiver systems, the receiver can be put into a standby state. Setting the receiver wakeup bit, RWU, in SCC2 puts the receiver into a standby state during which receiver interrupts are disabled.

Depending on the state of the WAKE bit in SCC1, either of two conditions on the RxD pin can bring the receiver out of the standby state:

1. Address mark — An address mark is a 1 in the MSB position of a received character. When the WAKE bit is set, an address mark wakes the receiver from the standby state by clearing the RWU bit. The address mark also sets the ESCI receiver full bit, SCRF. Software can then compare the character containing the address mark to the user-defined address of the receiver. If they are the same, the receiver remains awake and processes the characters that follow. If they are not the same, software can set the RWU bit and put the receiver back into the standby state.
2. Idle input line condition — When the WAKE bit is clear, an idle character on the RxD pin wakes the receiver from the standby state by clearing the RWU bit. The idle character that wakes the receiver does not set the receiver idle bit, IDLE, or the ESCI receiver full bit, SCRF. The idle line type bit, ILTY, determines whether the receiver begins counting 1s as idle character bits after the start bit or after the stop bit.

AROVFL— Arbiter Counter Overflow Bit

This read-only bit indicates an arbiter counter overflow. Clear AROVFL by writing any value to SCIACTL. Writing 0s to AM1 and AM0 resets the counter keeps it in this idle state. Reset clears AROVFL.

1 = Arbiter counter overflow has occurred
0 = No arbiter counter overflow has occurred

ARD8— Arbiter Counter MSB

This read-only bit is the MSB of the 9-bit arbiter counter. Clear ARD8 by writing any value to SCIACTL. Reset clears ARD8.

13.9.2 ESCI Arbiter Data Register

Address: $000B

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	ARD7	ARD6	ARD5	ARD4	ARD3	ARD2	ARD1	ARD0
Write:								
Reset:	0	0	0	0	0	0	0	0

= Unimplemented

Figure 13-20. ESCI Arbiter Data Register (SCIADAT)

ARD7–ARD0 — Arbiter Least Significant Counter Bits

These read-only bits are the eight LSBs of the 9-bit arbiter counter. Clear ARD7–ARD0 by writing any value to SCIACTL. Writing 0s to AM1 and AM0 permanently resets the counter and keeps it in this idle state. Reset clears ARD7–ARD0.

13.9.3 Bit Time Measurement

Two bit time measurement modes, described here, are available according to the state of ACLK.

1. ACLK = 0 — The counter is clocked with the bus clock divided by four. The counter is started when a falling edge on the RxD pin is detected. The counter will be stopped on the next falling edge. ARUN is set while the counter is running, AFIN is set on the second falling edge on RxD (for instance, the counter is stopped). This mode is used to recover the received baud rate. See Figure 13-21.
2. ACLK = 1 — The counter is clocked with one half of the ESCI input clock generated by the ESCI prescaler. The counter is started when a 0 is detected on RxD (see Figure 13-22). A 0 on RxD on enabling the bit time measurement with ACLK = 1 leads to immediate start of the counter (see Figure 13-23). The counter will be stopped on the next rising edge of RxD. This mode is used to measure the length of a received break.

13.9.4 Arbitration Mode

If AM[1:0] is set to 10, the arbiter module operates in arbitration mode. On every rising edge of SCI_TxD (output of the ESCI module, internal chip signal), the counter is started. When the counter reaches $38 (ACLK = 0) or $08 (ACLK = 1), RxD is statically sensed. If in this case, RxD is sensed low (for example, another bus is driving the bus dominant) ALOST is set. As long as ALOST is set, the TxD pin is forced to 1, resulting in a seized transmission.

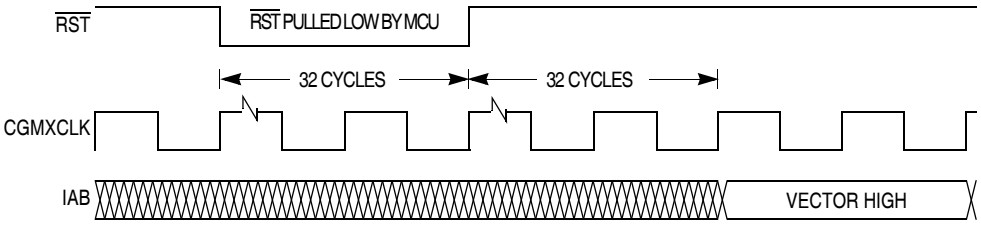


Figure 14-5. Internal Reset Timing

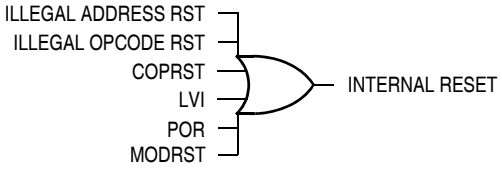


Figure 14-6. Sources of Internal Reset

Table 14-2. Reset Recovery

Reset Recovery Type	Actual Number of Cycles
POR/LVI	4163 (4096 + 64 + 3)
All others	67 (64 + 3)

14.3.2.1 Power-On Reset

When power is first applied to the MCU, the power-on reset module (POR) generates a pulse to indicate that power-on has occurred. The external reset pin ($\overline{RST}$) is held low while the SIM counter counts out 4096 + 32 CGMXCLK cycles. Thirty-two CGMXCLK cycles later, the CPU and memories are released from reset to allow the reset vector sequence to occur.

At power-on, these events occur:

- A POR pulse is generated.
- The internal reset signal is asserted.
- The SIM enables CGMOUT.
- Internal clocks to the CPU and modules are held inactive for 4096 CGMXCLK cycles to allow stabilization of the oscillator.
- The $\overline{RST}$ pin is driven low during the oscillator stabilization time.
- The POR bit of the SIM reset status register (SRSR) is set.

14.3.2.2 Computer Operating Properly (COP) Reset

An input to the SIM is reserved for the COP reset signal. The overflow of the COP counter causes an internal reset and sets the COP bit in the SIM reset status register (SRSR) if the COPD bit in the CONFIG1 register is cleared. The SIM actively pulls down the $\overline{RST}$ pin for all internal reset sources.

The COP module is disabled if the $\overline{RST}$ pin or the $\overline{IRQ}$ pin is held at V_{TST} while the MCU is in monitor mode. During a break state, V_{TST} on the $\overline{RST}$ pin disables the COP module.

an SPI configured as a slave does not have to correspond to any SPI baud rate. The baud rate only controls the speed of the SPSCK generated by an SPI configured as a master. Therefore, the frequency of the SPSCK for an SPI configured as a slave can be any frequency less than or equal to the bus speed.

When the master SPI starts a transmission, the data in the slave shift register begins shifting out on the MISO pin. The slave can load its shift register with a new byte for the next transmission by writing to its transmit data register. The slave must write to its transmit data register at least one bus cycle before the master starts the next transmission. Otherwise, the byte already in the slave shift register shifts out on the MISO pin. Data written to the slave shift register during a transmission remains in a buffer until the end of the transmission.

When the clock phase bit (CPHA) is set, the first edge of SPSCK starts a transmission. When CPHA is clear, the falling edge of $\overline{SS}$ starts a transmission. See 15.4 Transmission Formats.

NOTE

SPSCK must be in the proper idle state before the slave is enabled to prevent SPSCK from appearing as a clock edge.

15.4 Transmission Formats

During an SPI transmission, data is simultaneously transmitted (shifted out serially) and received (shifted in serially). A serial clock synchronizes shifting and sampling on the two serial data lines. A slave select line allows selection of an individual slave SPI device; slave devices that are not selected do not interfere with SPI bus activities. On a master SPI device, the slave select line can optionally be used to indicate multiple-master bus contention.

15.4.1 Clock Phase and Polarity Controls

Software can select any of four combinations of serial clock (SPSCK) phase and polarity using two bits in the SPI control register (SPCR). The clock polarity is specified by the CPOL control bit, which selects an active high or low clock and has no significant effect on the transmission format.

The clock phase (CPHA) control bit selects one of two fundamentally different transmission formats. The clock phase and polarity should be identical for the master SPI device and the communicating slave device. In some cases, the phase and polarity are changed between transmissions to allow a master device to communicate with peripheral slaves having different requirements.

NOTE

Before writing to the CPOL bit or the CPHA bit, disable the SPI by clearing the SPI enable bit (SPE).

15.4.2 Transmission Format When CPHA = 0

Figure 15-5 shows an SPI transmission in which CPHA = 0. The figure should not be used as a replacement for data sheet parametric information.

Two waveforms are shown for SPSCK: one for CPOL = 0 and another for CPOL = 1. The diagram may be interpreted as a master or slave timing diagram since the serial clock (SPSCK), master in/slave out (MISO), and master out/slave in (MOSI) pins are directly connected between the master and the slave. The MISO signal is the output from the slave, and the MOSI signal is the output from the master. The $\overline{SS}$ line is the slave select input to the slave. The slave SPI drives its MISO output only when its slave select input ($\overline{SS}$) is low, so that only the selected slave drives to the master. The $\overline{SS}$ pin of the master is not shown but is assumed to be inactive. The $\overline{SS}$ pin of the master must be high or must be reconfigured as

When enabled, the SPI controls data direction of the MOSI pin regardless of the state of the data direction register of the shared I/O port.

15.11.3 SPSCK (Serial Clock)

The serial clock synchronizes data transmission between master and slave devices. In a master MCU, the SPSCK pin is the clock output. In a slave MCU, the SPSCK pin is the clock input. In full-duplex operation, the master and slave MCUs exchange a byte of data in eight serial clock cycles.

When enabled, the SPI controls data direction of the SPSCK pin regardless of the state of the data direction register of the shared I/O port.

15.11.4 $\overline{SS}$ (Slave Select)

The $\overline{SS}$ pin has various functions depending on the current state of the SPI. For an SPI configured as a slave, the $\overline{SS}$ is used to select a slave. For CPHA = 0, the $\overline{SS}$ is used to define the start of a transmission. (See 15.4 Transmission Formats.) Since it is used to indicate the start of a transmission, $\overline{SS}$ must be toggled high and low between each byte transmitted for the CPHA = 0 format. However, it can remain low between transmissions for the CPHA = 1 format. See Figure 15-13.

When an SPI is configured as a slave, the $\overline{SS}$ pin is always configured as an input. It cannot be used as a general-purpose I/O regardless of the state of the MODFEN control bit. However, the MODFEN bit can still prevent the state of $\overline{SS}$ from creating a MODF error. See 15.12.2 SPI Status and Control Register.

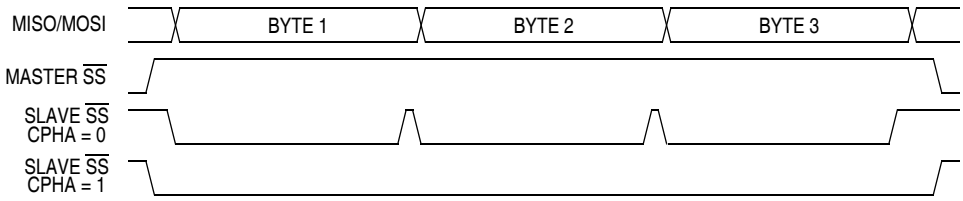


Figure 15-13. CPHA/$\overline{SS}$ Timing

NOTE

A high on the $\overline{SS}$ pin of a slave SPI puts the MISO pin in a high-impedance state. The slave SPI ignores all incoming SPSCK clocks, even if it was already in the middle of a transmission.

When an SPI is configured as a master, the $\overline{SS}$ input can be used in conjunction with the MODF flag to prevent multiple masters from driving MOSI and SPSCK. (See 15.6.2 Mode Fault Error.) For the state of the $\overline{SS}$ pin to set the MODF flag, the MODFEN bit in the SPSCK register must be set. If MODFEN is 0 for an SPI master, the $\overline{SS}$ pin can be used as a general-purpose I/O under the control of the data direction register of the shared I/O port. When MODFEN is 1, $\overline{SS}$ is an input-only pin to the SPI regardless of the state of the data direction register of the shared I/O port.

The CPU can always read the state of the $\overline{SS}$ pin by configuring the appropriate pin as an input and reading the port data register. See Table 15-2.

Table 18-2. Mode, Edge, and Level Selection

MSxB	MSxA	ELSxB	ELSxA	Mode	Configuration
X	0	0	0	Output preset	Pin under port control; initial output level high
X	1	0	0		Pin under port control; initial output level low
0	0	0	1	Input capture	Capture on rising edge only
0	0	1	0		Capture on falling edge only
0	0	1	1		Capture on rising or falling edge
0	1	0	0	Output compare or PWM	Software compare only
0	1	0	1		Toggle output on compare
0	1	1	0		Clear output on compare
0	1	1	1		Set output on compare
1	X	0	1	Buffered output compare or buffered PWM	Toggle output on compare
1	X	1	0		Clear output on compare
1	X	1	1		Set output on compare

ELSxB and ELSxA — Edge/Level Select Bits

When channel x is an input capture channel, these read/write bits control the active edge-sensing logic on channel x.

When channel x is an output compare channel, ELSxB and ELSxA control the channel x output behavior when an output compare occurs.

When ELSxB and ELSxA are both clear, channel x is not connected to port D or port F, and pin PTDx/T2CHx or pin PTFx/T2CHx is available as a general- purpose I/O pin. Table 18-2 shows how ELSxB and ELSxA work. Reset clears the ELSxB and ELSxA bits.

NOTE

After initially enabling a TIM2 channel register for input capture operation and selecting the edge sensitivity, clear CHxF to ignore any erroneous edge detection flags.

TOVx — Toggle-On-Overflow Bit

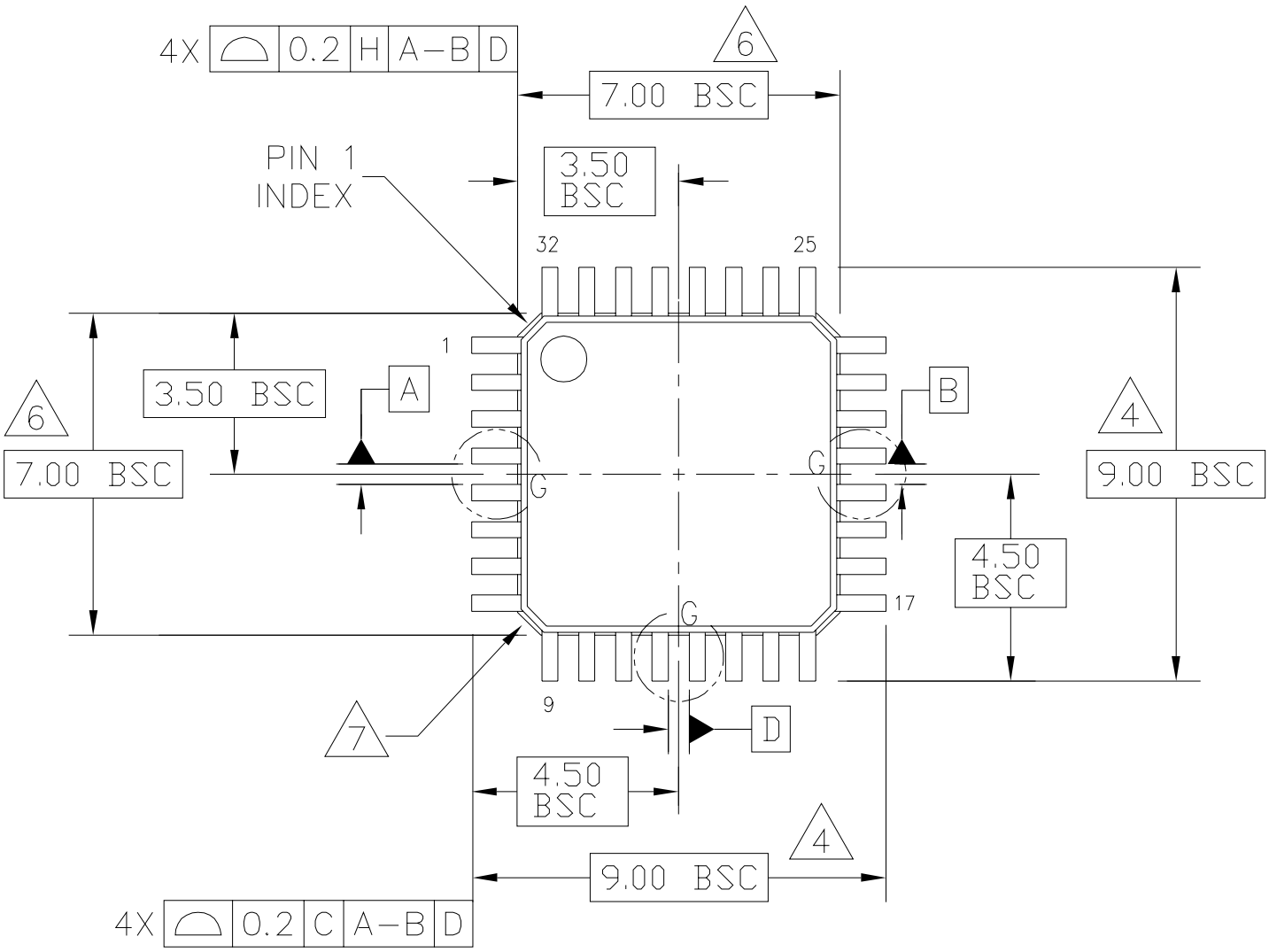
When channel x is an output compare channel, this read/write bit controls the behavior of the channel x output when the TIM2 counter overflows. When channel x is an input capture channel, TOVx has no effect. Reset clears the TOVx bit.

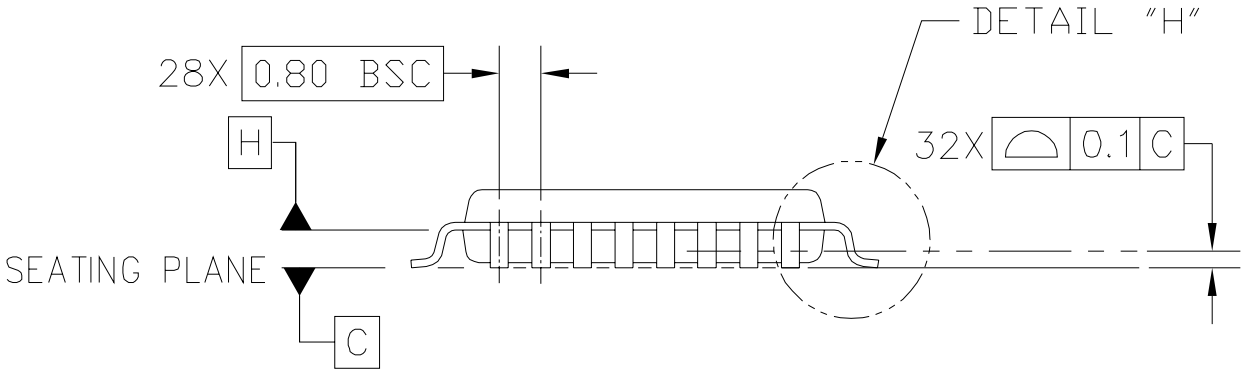
1 = Channel x pin toggles on TIM2 counter overflow.

0 = Channel x pin does not toggle on TIM2 counter overflow.

NOTE

When TOVx is set, a TIM2 counter overflow takes precedence over a channel x output compare if both occur at the same time.





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