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Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	PowerPC e300c2
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	333MHz
Co-Processors/DSP	Communications; QUICC Engine, Security; SEC 2.2
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100Mbps (3)
SATA	-
USB	USB 2.0 (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography
Package / Case	516-BBGA
Supplier Device Package	516-PBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8321eczqafdc

1 Overview

The MPC8323E incorporates the e300c2 (MPC603e-based) core built on Power Architecture® technology, which includes 16 Kbytes of L1 instruction and data caches, dual integer units, and on-chip memory management units (MMUs). The e300c2 core does not contain a floating point unit (FPU). The MPC8323E also includes a 32-bit PCI controller, four DMA channels, a security engine, and a 32-bit DDR1/DDR2 memory controller.

A new communications complex based on QUICC Engine technology forms the heart of the networking capability of the MPC8323E. The QUICC Engine block contains several peripheral controllers and a 32-bit RISC controller. Protocol support is provided by the main workhorses of the device—the unified communication controllers (UCCs). Note that the MPC8321 and MPC8321E do not support UTOPIA. A block diagram of the MPC8323E is shown in Figure 1.

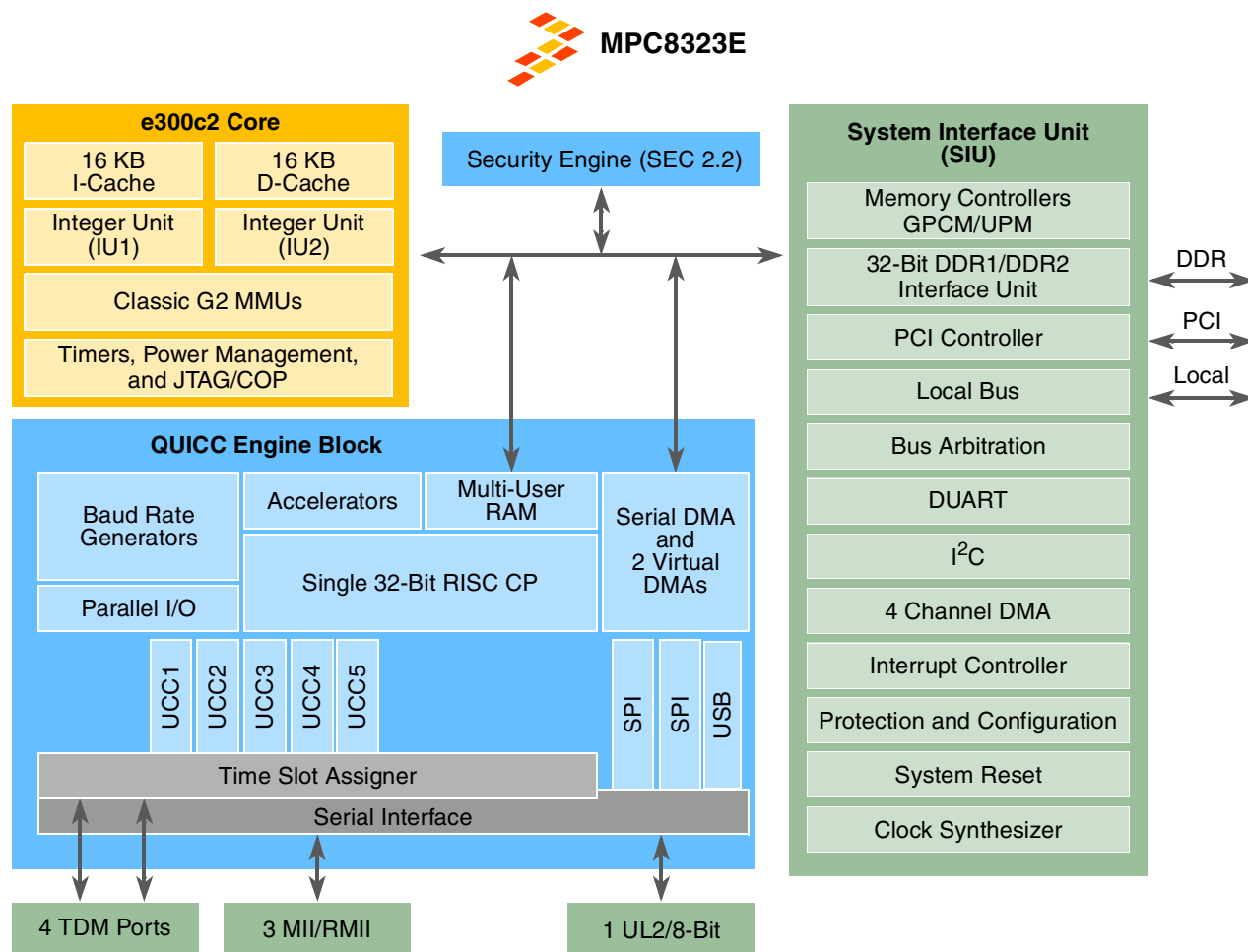


Figure 1. MPC8323E Block Diagram

Each of the five UCCs can support a variety of communication protocols: 10/100 Mbps Ethernet, serial ATM, HDLC, UART, and BISYNC—and, in the MPC8323E and MPC8323, multi-PHY ATM and ATM support for up to OC-3 speeds.

NOTE

The QUICC Engine block can also support a UTOPIA level 2 capable of supporting 31 multi-PHY (MPC8323E- and MPC8323-specific).

The MPC8323E security engine (SEC 2.2) allows CPU-intensive cryptographic operations to be offloaded from the main CPU core. The security-processing accelerator provides hardware acceleration for the DES, 3DES, AES, SHA-1, and MD-5 algorithms.

In summary, the MPC8323E family provides users with a highly integrated, fully programmable communications processor. This helps ensure that a low-cost system solution can be quickly developed and offers flexibility to accommodate new standards and evolving system requirements.

1.1 MPC8323E Features

Major features of the MPC8323E are as follows:

- High-performance, low-power, and cost-effective single-chip data-plane/control-plane solution for ATM or IP/Ethernet packet processing (or both).
- MPC8323E QUICC Engine block offers a future-proof solution for next generation designs by supporting programmable protocol termination and network interface termination to meet evolving protocol standards.
- Single platform architecture supports the convergence of IP packet networks and ATM networks.
- DDR1/DDR2 memory controller—one 32-bit interface at up to 266 MHz supporting both DDR1 and DDR2.
- An e300c2 core built on Power Architecture technology with 16-Kbyte instruction and data caches, and dual integer units.
- Peripheral interfaces such as 32-bit PCI (2.2) interface up to 66-MHz operation, 16-bit local bus interface up to 66-MHz operation, and USB 2.0 (full-/low-speed).
- Security engine provides acceleration for control and data plane security protocols.
- High degree of software compatibility with previous-generation PowerQUICC processor-based designs for backward compatibility and easier software migration.

1.1.1 Protocols

The protocols are as follows:

- ATM SAR up to 155 Mbps (OC-3) full duplex, with ATM traffic shaping (ATF TM4.1)
- Support for ATM AAL1 structured and unstructured circuit emulation service (CES 2.0)
- Support for IMA and ATM transmission convergence sub-layer
- ATM OAM handling features compatible with ITU-T I.610
- IP termination support for IPv4 and IPv6 packets including TOS, TTL, and header checksum processing
- Extensive support for ATM statistics and Ethernet RMON/MIB statistics
- Support for 64 channels of HDLC/transparent

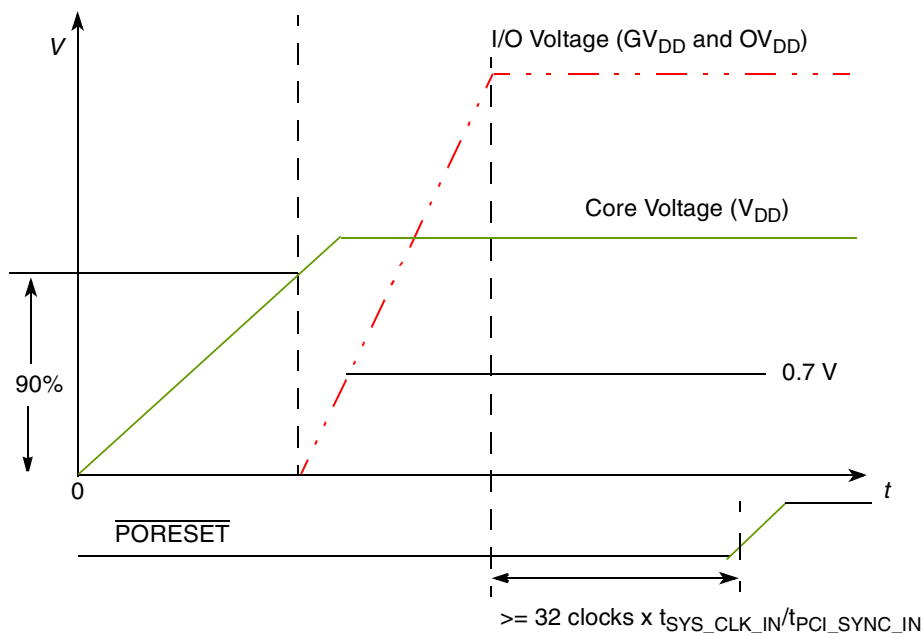


Figure 3. MPC8323E Power-Up Sequencing Example

3 Power Characteristics

The estimated typical power dissipation for this family of MPC8323E devices is shown in [Table 5](#).

Table 5. MPC8323E Power Dissipation

CSB Frequency (MHz)	QUICC Engine Frequency (MHz)	Core Frequency (MHz)	Typical	Maximum	Unit	Notes
133	200	266	0.74	1.48	W	1, 2, 3
133	200	333	0.78	1.62	W	1, 2, 3

Notes:

1. The values do not include I/O supply power (OV_{DD} and GV_{DD}) or AV_{DD} . For I/O power values, see [Table 6](#).
2. Typical power is based on a nominal voltage of $V_{DD} = 1.0$ V, ambient temperature, and the core running a Dhrystone benchmark application. The measurements were taken on the MPC8323MDS evaluation board using WC process silicon.
3. Maximum power is based on a voltage of $V_{DD} = 1.07$ V, WC process, a junction $T_J = 110^{\circ}\text{C}$, and an artificial smoke test.

[Table 6](#) shows the estimated typical I/O power dissipation for the device.

Table 6. Estimated Typical I/O Power Dissipation

Interface	Parameter	GV_{DD} (1.8 V)	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	Unit	Comments
DDR I/O 65% utilization 2.5 V $R_s = 20\ \Omega$ $R_t = 50\ \Omega$ 1 pair of clocks	266 MHz, 1×32 bits	0.212	0.367	—	W	—

Table 7. CLKIN DC Electrical Characteristics (continued)

CLKIN input current	$0\text{ V} \leq V_{\text{IN}} \leq \text{OV}_{\text{DD}}$	I_{IN}	—	± 5	μA
PCI_SYNC_IN input current	$0\text{ V} \leq V_{\text{IN}} \leq 0.5\text{ V}$ or $\text{OV}_{\text{DD}} - 0.5\text{ V} \leq V_{\text{IN}} \leq \text{OV}_{\text{DD}}$	I_{IN}	—	± 5	μA
PCI_SYNC_IN input current	$0.5\text{ V} \leq V_{\text{IN}} \leq \text{OV}_{\text{DD}} - 0.5\text{ V}$	I_{IN}	—	± 50	μA

4.2 AC Electrical Characteristics

The primary clock source for the MPC8323E can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. Table 8 provides the clock input (CLKIN/PCI_CLK) AC timing specifications for the MPC8323E.

Table 8. CLKIN AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
CLKIN/PCI_CLK frequency	f_{CLKIN}	25	—	66.67	MHz	1
CLKIN/PCI_CLK cycle time	t_{CLKIN}	15	—	—	ns	—
CLKIN rise and fall time	$t_{\text{KH}}, t_{\text{KL}}$	0.6	0.8	4	ns	2
PCI_CLK rise and fall time	$t_{\text{PCH}}, t_{\text{PCL}}$	0.6	0.8	1.2	ns	2
CLKIN/PCI_CLK duty cycle	$t_{\text{KHK}}/t_{\text{CLKIN}}$	40	—	60	%	3
CLKIN/PCI_CLK jitter	—	—	—	± 150	ps	4, 5

Notes:

- Caution:** The system, core, security, and QUICC Engine block must not exceed their respective maximum or minimum operating frequencies.
- Rise and fall times for CLKIN/PCI_CLK are measured at 0.4 and 2.7 V.
- Timing is guaranteed by design and characterization.
- This represents the total input jitter—short term and long term—and is guaranteed by design.
- The CLKIN/PCI_CLK driver's closed loop jitter bandwidth should be < 500 kHz at –20 dB. The bandwidth must be set low to allow cascade-connected PLL-based devices to track CLKIN drivers with the specified jitter.

5 RESET Initialization

This section describes the AC electrical specifications for the reset initialization timing requirements of the MPC8323E. Table 9 provides the reset initialization AC timing specifications for the reset component(s).

Table 9. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{\text{HRESET}}$ or $\overline{\text{SRESET}}$ (input) to activate reset flow	32	—	$t_{\text{PCI_SYNC_IN}}$	1
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to CLKIN when the MPC8323E is in PCI host mode	32	—	t_{CLKIN}	2
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to PCI_SYNC_IN when the MPC8323E is in PCI agent mode	32	—	$t_{\text{PCI_SYNC_IN}}$	1

Table 9. RESET Initialization Timing Specifications (continued)

Parameter/Condition	Min	Max	Unit	Notes
$\overline{\text{HRESET}}/\overline{\text{SRESET}}$ assertion (output)	512	—	$t_{\text{PCI_SYNC_IN}}$	1
$\overline{\text{HRESET}}$ negation to $\overline{\text{SRESET}}$ negation (output)	16	—	$t_{\text{PCI_SYNC_IN}}$	1
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the MPC8323E is in PCI host mode	4	—	t_{CLKIN}	2
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the MPC8323E is in PCI agent mode	4	—	$t_{\text{PCI_SYNC_IN}}$	1
Input hold time for POR config signals with respect to negation of $\overline{\text{HRESET}}$	0	—	ns	—
Time for the MPC8323E to turn off POR configuration signals with respect to the assertion of $\overline{\text{HRESET}}$	—	4	ns	3
Time for the MPC8323E to turn on POR configuration signals with respect to the negation of $\overline{\text{HRESET}}$	1	—	$t_{\text{PCI_SYNC_IN}}$	1, 3

Notes:

1. $t_{\text{PCI_SYNC_IN}}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the MPC8323E is in PCI host mode the primary clock is applied to the CLKIN input, and PCI_SYNC_IN period depends on the value of CFG_CLKIN_DIV. See the *MPC8323E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more details.
2. t_{CLKIN} is the clock period of the input clock applied to CLKIN. It is only valid when the MPC8323E is in PCI host mode. See the *MPC8323E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more details.
3. POR configuration signals consists of CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV.

Table 10 provides the PLL lock times.

Table 10. PLL Lock Times

Parameter/Condition	Min	Max	Unit	Notes
PLL lock times	—	100	μs	—

5.1 Reset Signals DC Electrical Characteristics

Table 11 provides the DC electrical characteristics for the MPC8323E reset signals mentioned in Table 9.

Table 11. Reset Signals DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit	Notes
Output high voltage	V_{OH}	$I_{\text{OH}} = -6.0 \text{ mA}$	2.4	—	V	1
Output low voltage	V_{OL}	$I_{\text{OL}} = 6.0 \text{ mA}$	—	0.5	V	1
Output low voltage	V_{OL}	$I_{\text{OL}} = 3.2 \text{ mA}$	—	0.4	V	1
Input high voltage	V_{IH}	—	2.0	$\text{OV}_{\text{DD}} + 0.3$	V	1
Input low voltage	V_{IL}	—	-0.3	0.8	V	—

Table 19. DDR1 and DDR2 SDRAM Output AC Timing Specifications (continued)

At recommended operating conditions with $D_n_GV_{DD}$ of (1.8 or 2.5 V) $\pm$ 5%.

Parameter	Symbol ¹	Min	Max	Unit	Notes
MDQ/MDM output setup with respect to MDQS 266 MHz 200 MHz	t_{DDKHDS} , t_{DDKLDS}	0.9 1.0	— —	ns	5
MDQ/MDM output hold with respect to MDQS 266 MHz 200 MHz	t_{DDKHDX} , t_{DDKLDX}	1100 1200	— —	ps	5
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.6$	$-0.5 \times t_{MCK} + 0.6$	ns	6
MDQS epilogue end	t_{DDKHME}	-0.6	0.6	ns	6

Notes:

- The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
- All $MCK/\overline{MCK}$ referenced measurements are made from the crossing of the two signals ± 0.1 V.
- ADDR/CMD includes all DDR SDRAM output signals except $MCK/\overline{MCK}$, $\overline{MCS}$, and MDQ/MDM/MDQS. For the ADDR/CMD setup and hold specifications, it is assumed that the Clock Control register is set to adjust the memory clocks by 1/2 applied cycle.
- Note that t_{DDKMHM} follows the symbol conventions described in note 1. For example, t_{DDKMHM} describes the DDR timing (DD) from the rising edge of the $MCK(n)$ clock (KH) until the MDQS signal is valid (MH). t_{DDKMHM} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. This is typically set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these 2 parameters have been set to the same adjustment value. See the *MPC8323E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
- Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the microprocessor.
- All outputs are referenced to the rising edge of $MCK(n)$ at the pins of the microprocessor. Note that t_{DDKHMP} follows the symbol conventions described in note 1.

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface of the MPC8323E.

7.1 DUART DC Electrical Characteristics

Table 20 provides the DC electrical characteristics for the DUART interface of the MPC8323E.

Table 20. DUART DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	$OV_{DD} + 0.3$	V
Low-level input voltage OV_{DD}	V_{IL}	-0.3	0.8	V
High-level output voltage, $I_{OH} = -100 \mu A$	V_{OH}	$OV_{DD} - 0.2$	—	V
Low-level output voltage, $I_{OL} = 100 \mu A$	V_{OL}	—	0.2	V
Input current ($0 V \leq V_{IN} \leq OV_{DD}$) ¹	I_{IN}	—	± 5	μA

Note:

- Note that the symbol V_{IN} , in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

7.2 DUART AC Electrical Specifications

Table 21 provides the AC timing parameters for the DUART interface of the MPC8323E.

Table 21. DUART AC Timing Specifications

Parameter	Value	Unit	Notes
Minimum baud rate	256	baud	
Maximum baud rate	> 1,000,000	baud	1
Oversample rate	16	—	2

Notes:

- Actual attainable baud rate is limited by the latency of interrupt processing.
- The middle of a start bit is detected as the 8th sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16th sample.

8 Ethernet and MII Management

This section provides the AC and DC electrical characteristics for Ethernet and MII management.

8.1 Ethernet Controller (10/100 Mbps)—MII/RMII Electrical Characteristics

The electrical characteristics specified here apply to all MII (media independent interface) and RMII (reduced media independent interface), except MDIO (management data input/output) and MDC

8.3.1 MII Management DC Electrical Characteristics

MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in [Table 27](#).

Table 27. MII Management DC Electrical Characteristics When Powered at 3.3 V

Parameter	Symbol	Conditions	Min	Max	Unit
Supply voltage (3.3 V)	OV_{DD}	—	2.97	3.63	V
Output high voltage	V_{OH}	$I_{OH} = -1.0 \text{ mA}$ $OV_{DD} = \text{Min}$	2.10	$OV_{DD} + 0.3$	V
Output low voltage	V_{OL}	$I_{OL} = 1.0 \text{ mA}$ $OV_{DD} = \text{Min}$	GND	0.50	V
Input high voltage	V_{IH}	—	2.00	—	V
Input low voltage	V_{IL}	—	—	0.80	V
Input current	I_{IN}	$0 \text{ V} \leq V_{IN} \leq OV_{DD}$	—	± 5	μA

8.3.2 MII Management AC Electrical Specifications

[Table 28](#) provides the MII management AC timing specifications.

Table 28. MII Management AC Timing Specifications

At recommended operating conditions with OV_{DD} is 3.3 V $\pm 10\%$.

Parameter/Condition	Symbol ¹	Min	Typical	Max	Unit	Notes
MDC frequency	f_{MDC}	—	2.5	—	MHz	—
MDC period	t_{MDC}	—	400	—	ns	—
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	—
MDC to MDIO delay	t_{MDKHDX}	10	—	70	ns	—
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	—
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	—
MDC rise time	t_{MDCR}	—	—	10	ns	—
MDC fall time	t_{MDHF}	—	—	10	ns	—

Note:

- The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 13 shows the MII management AC timing diagram.

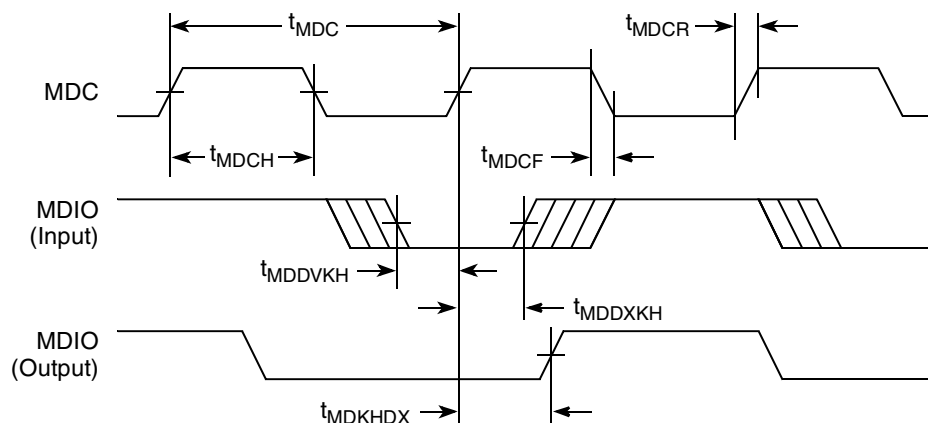


Figure 13. MII Management Interface Timing Diagram

9 Local Bus

This section describes the DC and AC electrical specifications for the local bus interface of the MPC8323E.

9.1 Local Bus DC Electrical Characteristics

Table 29 provides the DC electrical characteristics for the local bus interface.

Table 29. Local Bus DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	$OV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}	-0.3	0.8	V
High-level output voltage, $I_{OH} = -100 \mu A$	V_{OH}	$OV_{DD} - 0.2$	—	V
Low-level output voltage, $I_{OL} = 100 \mu A$	V_{OL}	—	0.2	V
Input current	I_{IN}	—	± 5	μA

9.2 Local Bus AC Electrical Specifications

Table 30 describes the general timing parameters of the local bus interface of the MPC8323E.

Table 30. Local Bus General Timing Parameters

Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	15	—	ns	2
Input setup to local bus clock (LCLK _n)	t_{LBIVKH}	7	—	ns	3, 4
Input hold from local bus clock (LCLK _n)	t_{LBIXKH}	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5

Table 31. JTAG Interface DC Electrical Characteristics (continued)

Characteristic	Symbol	Condition	Min	Max	Unit
Input low voltage	V_{IL}	—	−0.3	0.8	V
Input current	I_{IN}	$0\text{ V} \leq V_{IN} \leq OV_{DD}$	—	±5	μA

10.2 JTAG AC Electrical Characteristics

This section describes the AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8323E. [Table 32](#) provides the JTAG AC timing specifications as defined in [Figure 19](#) through [Figure 22](#).

Table 32. JTAG AC Timing Specifications (Independent of CLKIN)¹

At recommended operating conditions (see [Table 2](#)).

Parameter	Symbol ²	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	11	—	ns	—
JTAG external clock rise and fall times	t_{JTGR}, t_{JTGF}	0	2	ns	—
$\overline{TRST}$ assert time	t_{TRST}	25	—	ns	3
Input setup times:				ns	
Boundary-scan data TMS, TDI	t_{JTDVKH} t_{JTIVKH}	4 4	— —		4
Input hold times:				ns	
Boundary-scan data TMS, TDI	t_{JTDXKH} t_{JTIXKH}	10 10	— —		4
Valid times:				ns	
Boundary-scan data TDO	t_{JTKLDV} t_{JTKLOV}	2 2	15 15		5
Output hold times:				ns	
Boundary-scan data TDO	t_{JTKLDX} t_{JTKLOX}	2 2	— —		5

16 SPI

This section describes the DC and AC electrical specifications for the SPI of the MPC8323E.

16.1 SPI DC Electrical Characteristics

Table 44 provides the DC electrical characteristics for the MPC8323E SPI.

Table 44. SPI DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	$I_{OH} = -6.0 \text{ mA}$	2.4	—	V
Output low voltage	V_{OL}	$I_{OL} = 6.0 \text{ mA}$	—	0.5	V
Output low voltage	V_{OL}	$I_{OL} = 3.2 \text{ mA}$	—	0.4	V
Input high voltage	V_{IH}	—	2.0	$OV_{DD} + 0.3$	V
Input low voltage	V_{IL}	—	-0.3	0.8	V
Input current	I_{IN}	$0 \text{ V} \leq V_{IN} \leq OV_{DD}$	—	± 5	μA

16.2 SPI AC Timing Specifications

Table 45 and provide the SPI input and output AC timing specifications.

Table 45. SPI AC Timing Specifications¹

Characteristic	Symbol ²	Min	Max	Unit
SPI outputs—Master mode (internal clock) delay	t_{NIKHOV}	0.5	6	ns
SPI outputs—Slave mode (external clock) delay	t_{NEKHOV}	2	8	ns
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	6	—	ns
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns

Notes:

- Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
- The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, t_{NIKHOV} symbolizes the NMSI outputs internal timing (NI) for the time t_{SPI} memory clock reference (K) goes from the high state (H) until outputs (O) are valid (V).

Figure 30 provides the AC test load for the SPI.

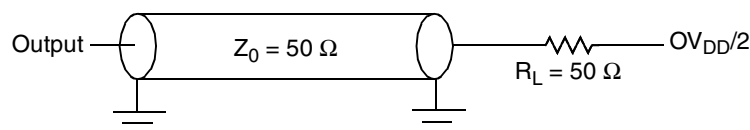


Figure 30. SPI AC Test Load

Table 55. MPC8323E PBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
MEMC_MDQ29	AD20	IO	GV _{DD}	—
MEMC_MDQ30	AF23	IO	GV _{DD}	—
MEMC_MDQ31	AD22	IO	GV _{DD}	—
MEMC_MDM0	AC9	O	GV _{DD}	—
MEMC_MDM1	AD5	O	GV _{DD}	—
MEMC_MDM2	AE20	O	GV _{DD}	—
MEMC_MDM3	AE22	O	GV _{DD}	—
MEMC_MDQS0	AE8	IO	GV _{DD}	—
MEMC_MDQS1	AE5	IO	GV _{DD}	—
MEMC_MDQS2	AC19	IO	GV _{DD}	—
MEMC_MDQS3	AE23	IO	GV _{DD}	—
MEMC_MBA0	AD16	O	GV _{DD}	—
MEMC_MBA1	AD17	O	GV _{DD}	—
MEMC_MBA2	AE17	O	GV _{DD}	—
MEMC_MA0	AD12	O	GV _{DD}	—
MEMC_MA1	AE12	O	GV _{DD}	—
MEMC_MA2	AF12	O	GV _{DD}	—
MEMC_MA3	AC13	O	GV _{DD}	—
MEMC_MA4	AD13	O	GV _{DD}	—
MEMC_MA5	AE13	O	GV _{DD}	—
MEMC_MA6	AF13	O	GV _{DD}	—
MEMC_MA7	AC15	O	GV _{DD}	—
MEMC_MA8	AD15	O	GV _{DD}	—
MEMC_MA9	AE15	O	GV _{DD}	—
MEMC_MA10	AF15	O	GV _{DD}	—
MEMC_MA11	AE16	O	GV _{DD}	—
MEMC_MA12	AF16	O	GV _{DD}	—
MEMC_MA13	AB16	O	GV _{DD}	—
MEMC_MWE	AC17	O	GV _{DD}	—
MEMC_MRAS	AE11	O	GV _{DD}	—
MEMC_MCAS	AD11	O	GV _{DD}	—
MEMC_MCS	AC11	O	GV _{DD}	—

Table 55. MPC8323E PBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{LCS1}}$	AB25	O	OV_{DD}	4
$\overline{\text{LCS2}}$	AA23	O	OV_{DD}	4
$\overline{\text{LCS3}}$	AA24	O	OV_{DD}	4
$\overline{\text{LWE0}}$	Y23	O	OV_{DD}	4
$\overline{\text{LWE1}}$	W25	O	OV_{DD}	4
LBCTL	V25	O	OV_{DD}	4
LALE	V24	O	OV_{DD}	7
CFG_RESET_SOURCE[0]/LSDA10/LGPL0	L23	IO	OV_{DD}	—
CFG_RESET_SOURCE[1]/ $\overline{\text{LSDWE}}$ /LGPL1	K23	IO	OV_{DD}	—
$\overline{\text{LSDRAS}}$ /LGPL2/ $\overline{\text{LOE}}$	J23	O	OV_{DD}	4
CFG_RESET_SOURCE[2]/ $\overline{\text{LSDCAS}}$ /LGPL3	H23	IO	OV_{DD}	—
LGPL4/ $\overline{\text{LGTA}}$ /LUPWAIT/LPBSE	G23	IO	OV_{DD}	4, 8
LGPL5	AC22	O	OV_{DD}	4
LCLK0	Y24	O	OV_{DD}	7
LCLK1	Y25	O	OV_{DD}	7
DUART				
UART_SOUT1/MSRCID0 (DDR ID)/LSRCID0	G1	IO	OV_{DD}	—
UART_SIN1/MSRCID1 (DDR ID)/LSRCID1	G2	IO	OV_{DD}	—
$\overline{\text{UART_CTS1}}$ /MSRCID2 (DDR ID)/LSRCID2	H3	IO	OV_{DD}	—
$\overline{\text{UART_RTS1}}$ /MSRCID3 (DDR ID)/LSRCID3	K3	IO	OV_{DD}	—
UART_SOUT2/MSRCID4 (DDR ID)/LSRCID4	H2	IO	OV_{DD}	—
UART_SIN2/MDVAL (DDR ID)/LDVAL	H1	IO	OV_{DD}	—
$\overline{\text{UART_CTS2}}$	J3	IO	OV_{DD}	—
$\overline{\text{UART_RTS2}}$	K4	IO	OV_{DD}	—
I²C interface				
IIC_SDA/ $\overline{\text{CKSTOP_OUT}}$	AE24	IO	OV_{DD}	2
IIC_SCL/ $\overline{\text{CKSTOP_IN}}$	AF24	IO	OV_{DD}	2
Programmable Interrupt Controller				
$\overline{\text{MCP_OUT}}$	AD25	O	OV_{DD}	—
$\overline{\text{IRQ0/MCP_IN}}$	AD26	I	OV_{DD}	—
$\overline{\text{IRQ1}}$	K1	IO	OV_{DD}	—
$\overline{\text{IRQ2}}$	K2	I	OV_{DD}	—

Table 55. MPC8323E PBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
GPIO_PA26/Enet2_RX_ER/SER2_CD/TDMB_REQ/LA10 (LBIU)	E26	IO	OV _{DD}	—
GPIO_PA27/Enet2_TX_ER/TDMB_CLKO/LA11 (LBIU)	F25	IO	OV _{DD}	—
GPIO_PA28/Enet2_RX_DV/SER2_CTS/TDMB_RSYNC/LA12 (LBIU)	E25	IO	OV _{DD}	—
GPIO_PA29/Enet2_COL/RXD[4]/SER2_RXD[4]/TDMB_STROBE/LA13 (LBIU)	J25	IO	OV _{DD}	—
GPIO_PA30/Enet2_TX_EN/SER2_RTS/TDMB_TSYNC/LA14 (LBIU)	F26	IO	OV _{DD}	—
GPIO_PA31/Enet2_CRS/SDET LA15 (LBIU)	J26	IO	OV _{DD}	—
GPIO_PB0/Enet3_TXD[0]/SER3_TXD[0]/TDMC_TXD[0]	A13	IO	OV _{DD}	—
GPIO_PB1/Enet3_TXD[1]/SER3_TXD[1]/TDMC_TXD[1]	B13	IO	OV _{DD}	—
GPIO_PB2/Enet3_TXD[2]/SER3_TXD[2]/TDMC_TXD[2]	A14	IO	OV _{DD}	—
GPIO_PB3/Enet3_TXD[3]/SER3_TXD[3]/TDMC_TXD[3]	B14	IO	OV _{DD}	—
GPIO_PB4/Enet3_RXD[0]/SER3_RXD[0]/TDMC_RXD[0]	B8	IO	OV _{DD}	—
GPIO_PB5/Enet3_RXD[1]/SER3_RXD[1]/TDMC_RXD[1]	A8	IO	OV _{DD}	—
GPIO_PB6/Enet3_RXD[2]/SER3_RXD[2]/TDMC_RXD[2]	A9	IO	OV _{DD}	—
GPIO_PB7/Enet3_RXD[3]/SER3_RXD[3]/TDMC_RXD[3]	B9	IO	OV _{DD}	—
GPIO_PB8/Enet3_RX_ER/SER3_CD/TDMC_REQ	A11	IO	OV _{DD}	—
GPIO_PB9/Enet3_TX_ER/TDMC_CLKO	B11	IO	OV _{DD}	—
GPIO_PB10/Enet3_RX_DV/SER3_CTS/TDMC_RSYNC	A10	IO	OV _{DD}	—
GPIO_PB11/Enet3_COL/RXD[4]/SER3_RXD[4]/TDMC_STROBE	A15	IO	OV _{DD}	—
GPIO_PB12/Enet3_TX_EN/SER3_RTS/TDMC_TSYNC	B12	IO	OV _{DD}	—
GPIO_PB13/Enet3_CRS/SDET	B15	IO	OV _{DD}	—
GPIO_PB14/CLK12	D9	IO	OV _{DD}	—
GPIO_PB15 UPC1_TxADDR[4]	D14	IO	OV _{DD}	—
GPIO_PB16 UPC1_RxADDR[4]	B16	IO	OV _{DD}	—

Table 55. MPC8323E PBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
GPIO_PC10/UPC1_RxDATA[2]/SER5_RXD[2]	B21	IO	OV _{DD}	—
GPIO_PC11/UPC1_RxDATA[3]/SER5_RXD[3]	A20	IO	OV _{DD}	—
GPIO_PC12/UPC1_RxDATA[4]	D19	IO	OV _{DD}	—
GPIO_PC13/UPC1_RxDATA[5]/LSRCID0	C18	IO	OV _{DD}	—
GPIO_PC14/UPC1_RxDATA[6]/LSRCID1	D18	IO	OV _{DD}	—
GPIO_PC15/UPC1_RxDATA[7]/LSRCID2	A25	IO	OV _{DD}	—
GPIO_PC16/UPC1_TxADDR[0]	C21	IO	OV _{DD}	—
GPIO_PC17/UPC1_TxADDR[1]/LSRCID3	D22	IO	OV _{DD}	—
GPIO_PC18/UPC1_TxADDR[2]/LSRCID4	C23	IO	OV _{DD}	—
GPIO_PC19/UPC1_TxADDR[3]/LDVAL	D23	IO	OV _{DD}	—
GPIO_PC20/UPC1_RxADDR[0]	C17	IO	OV _{DD}	—
GPIO_PC21/UPC1_RxADDR[1]	D17	IO	OV _{DD}	—
GPIO_PC22/UPC1_RxADDR[2]	C16	IO	OV _{DD}	—
GPIO_PC23/UPC1_RxADDR[3]	D16	IO	OV _{DD}	—
GPIO_PC24/UPC1_RxSOC/SER5_CD	A16	IO	OV _{DD}	—
GPIO_PC25/UPC1_RxCLAV	D20	IO	OV _{DD}	—
GPIO_PC26/UPC1_RxPRTY/CE_EXT_REQ2	E23	IO	OV _{DD}	—
GPIO_PC27/UPC1_RxEN	B17	IO	OV _{DD}	—
GPIO_PC28/UPC1_TxSOC	B22	IO	OV _{DD}	—
GPIO_PC29/UPC1_TxCLAV/SER5_CTS	A17	IO	OV _{DD}	—
GPIO_PC30/UPC1_TxPRTY	A22	IO	OV _{DD}	—
GPIO_PC31/UPC1_TxEN/SER5_RTS	C20	IO	OV _{DD}	—
GPIO_PD0/SPIMOSI	A2	IO	OV _{DD}	—
GPIO_PD1/SPIMISO	B2	IO	OV _{DD}	—
GPIO_PD2/SPICLK	B3	IO	OV _{DD}	—
GPIO_PD3/SPISEL	A3	IO	OV _{DD}	—
GPIO_PD4/SPI_MDIO/CE_MUX_MDIO	A4	IO	OV _{DD}	—
GPIO_PD5/SPI_MDC/CE_MUX_MDC	B4	IO	OV _{DD}	—
GPIO_PD6/CLK8/BRGO16/CE_EXT_REQ3	F24	IO	OV _{DD}	—
GPIO_PD7/GTM1_TIN1/GTM2_TIN2/CLK5	G24	IO	OV _{DD}	—
GPIO_PD8/GTM1_TGATE1/GTM2_TGATE2/CLK6	H24	IO	OV _{DD}	—
GPIO_PD9/GTM1_TOUT1	D24	IO	OV _{DD}	—

22 Clocking

Figure 43 shows the internal distribution of clocks within the MPC8323E.

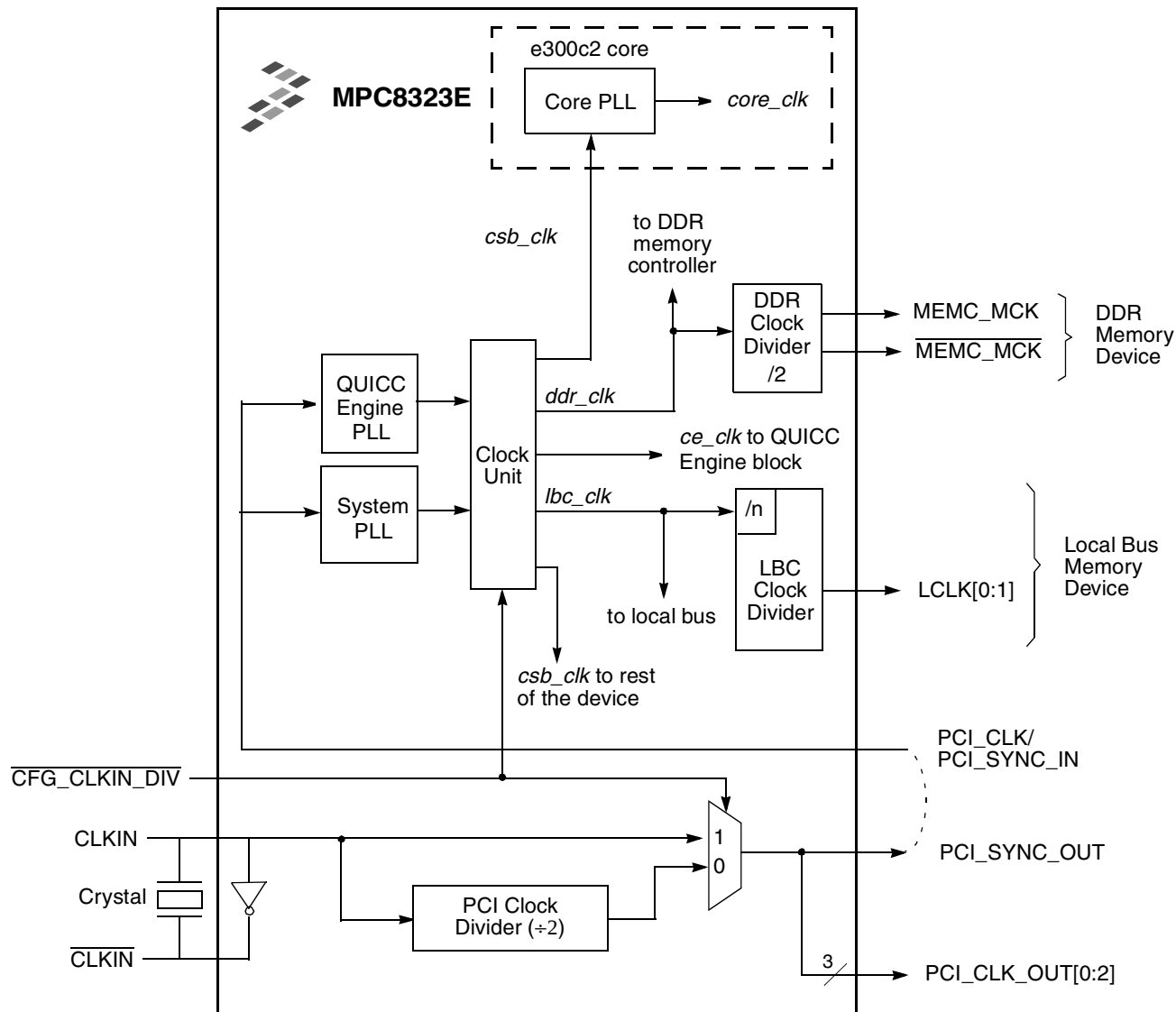


Figure 43. MPC8323E Clock Subsystem

The primary clock source for the MPC8323E can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode, respectively.

22.1 Clocking in PCI Host Mode

When the MPC8323E is configured as a PCI host device ($RCWH[PCIHOST] = 1$), CLKIN is its primary input clock. CLKIN feeds the PCI clock divider ($\div 2$) and the PCI_SYNC_OUT and PCI_CLK_OUT multiplexors. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system.

22.1.1 PCI Clock Outputs (PCI_CLK_OUT[0:2])

When the MPC8323E is configured as a PCI host, it provides three separate clock output signals, PCI_CLK_OUT[0:2], for external PCI agents.

When the device comes out of reset, the PCI clock outputs are disabled and are actively driven to a steady low state. Each of the individual clock outputs can be enabled (enable toggling of the clock) by setting its corresponding OCCR[PCICOEn] bit. All output clocks are phase-aligned to each other.

22.2 Clocking in PCI Agent Mode

When the MPC8323E is configured as a PCI agent device, PCI_CLK is the primary input clock. In agent mode, the CLKIN signal should be tied to GND, and the clock output signals, PCI_CLK_OUTn and PCI_SYNC_OUT, are not used.

22.3 System Clock Domains

As shown in [Figure 43](#), the primary clock input (frequency) is multiplied up by the system phase-locked loop (PLL) and the clock unit to create three major clock domains:

- The coherent system bus clock (*csb_clk*)
- The QUICC Engine clock (*ce_clk*)
- The internal clock for the DDR controller (*ddr_clk*)
- The internal clock for the local bus controller (*lb_clk*)

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$csb_clk = [PCI_SYNC_IN \times (1 + \sim \overline{CFG_CLKIN_DIV})] \times SPMF$$

In PCI host mode, $PCI_SYNC_IN \times (1 + \sim \overline{CFG_CLKIN_DIV})$ is the CLKIN frequency.

The *csb_clk* serves as the clock input to the e300c2 core. A second PLL inside the core multiplies up the *csb_clk* frequency to create the internal clock for the core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL) which is loaded at power-on reset or by one of the hard-coded reset options. See the “Reset Configuration” section in the *MPC8323E PowerQUICC II Pro Communications Processor Reference Manual* for more information.

shows the expected frequency values for the CSB frequency for select *csb_clk* to CLKIN/PCI_SYNC_IN ratios.

Table 59. CSB Frequency Options

CFG_CLKIN_DIV_B at Reset ¹	SPMF	<i>csb_clk</i> : Input Clock Ratio ²	Input Clock Frequency (MHz) ²		
			25	33.33	66.67
			<i>csb_clk</i> Frequency (MHz)		
High	0010	2 : 1			133
High	0011	3 : 1		100	
High	0100	4 : 1	100	133	
High	0101	5 : 1	125		
High	0110	6 : 1			
High	0111	7 : 1			
High	1000	8 : 1			
High	1001	9 : 1			
High	1010	10 : 1			
High	1011	11 : 1			
High	1100	12 : 1			
High	1101	13 : 1			
High	1110	14 : 1			
High	1111	15 : 1			
High	0000	16 : 1			
Low	0010	2 : 1			133
Low	0011	3 : 1		100	
Low	0100	4 : 1		133	
Low	0101	5 : 1			
Low	0110	6 : 1			
Low	0111	7 : 1			
Low	1000	8 : 1			
Low	1001	9 : 1			
Low	1010	10 : 1			
Low	1011	11 : 1			
Low	1100	12 : 1			
Low	1101	13 : 1			
Low	1110	14 : 1			
Low	1111	15 : 1			
Low	0000	16 : 1			

¹ CFG_CLKIN_DIV_B is only used for host mode; CLKIN must be tied low and CFG_CLKIN_DIV_B must be pulled up (high) in agent mode.

² CLKIN is the input clock in host mode; PCI_CLK is the input clock in agent mode.

(edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature (°C)

T_B = board temperature at the package perimeter (°C)

$R_{\theta JB}$ = junction-to-board thermal resistance (°C/W) per JESD51-8

P_D = power dissipation in package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

23.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)

T_T = thermocouple temperature on top of package (°C)

Ψ_{JT} = thermal characterization parameter (°C/W)

P_D = power dissipation in package (W)

The thermal characterization parameter is measured per JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

23.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

In some application environments, a heat sink is required to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction-to-case thermal resistance and a case to ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

Figure 44 shows the PLL power supply filter circuit.

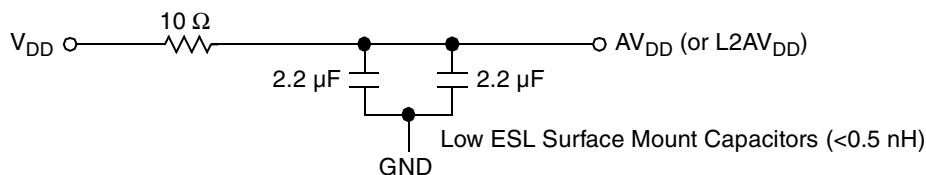


Figure 44. PLL Power Supply Filter Circuit

24.3 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, the MPC8323E can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC8323E system, and the MPC8323E itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD} , OV_{DD} , and GV_{DD} pins of the MPC8323E. These decoupling capacitors should receive their power from separate V_{DD} , OV_{DD} , GV_{DD} , and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 μF . Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD} , OV_{DD} , and GV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 μF (AVX TPS tantalum or Sanyo OSCON).

24.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD} , or GV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD} , GV_{DD} , OV_{DD} , and GND pins of the MPC8323E.

24.5 Output Buffer DC Impedance

The MPC8323E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 45). The