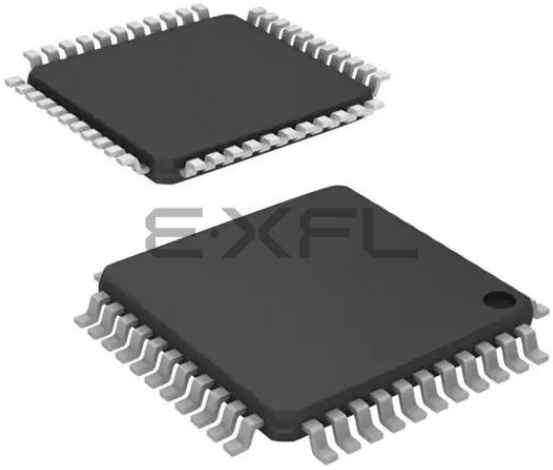

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Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1.5K x 8
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Oscillator Type	External
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Supplier Device Package	44-TQFP (10x10)
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TABLE 4-1: SPECIAL FUNCTION REGISTER MAP

Address	Name	Address	Name	Address	Name	Address	Name
FFFh	TOSU	FDFh	INDF2(2)	FBFh	CCPR1H	F9Fh	IPR1
FFEh	TOSH	FDEh	POSTINC2(2)	FBEh	CCPR1L	F9Eh	PIR1
FFDh	TOSL	FDDh	POSTDEC2(2)	FBDh	CCP1CON	F9Dh	PIE1
FFCh	STKPTR	FDCh	PREINC2(2)	FBCh	ECCPR1H(5)	F9Ch	—
FFBh	PCLATU	FDBh	PLUSW2(2)	FBBh	ECCPR1L(5)	F9Bh	—
FFAh	PCLATH	FDAh	FSR2H	FBAh	ECCP1CON(5)	F9Ah	—
FF9h	PCL	FD9h	FSR2L	FB9h	—	F99h	—
FF8h	TBLPTRU	FD8h	STATUS	FB8h	—	F98h	—
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	ECCP1DEL(5)	F97h	—
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	ECCPAS(5)	F96h	TRISE(5)
FF5h	TABLAT	FD5h	T0CON	FB5h	CVRCON(5)	F95h	TRISD(5)
FF4h	PRODH	FD4h	—	FB4h	CMCON(5)	F94h	TRISC
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB
FF2h	INTCON	FD2h	LVDCON	FB2h	TMR3L	F92h	TRISA
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	—
FF0h	INTCON3	FD0h	RCON	FB0h	—	F90h	—
FEFh	INDF0(2)	FCFh	TMR1H	FAFh	SPBRG	F8Fh	—
FEEh	POSTINC0(2)	FCEh	TMR1L	FAEh	RCREG	F8Eh	—
FEDh	POSTDEC0(2)	FCDh	T1CON	FADh	TXREG	F8Dh	LATE(5)
FECh	PREINC0(2)	FCCh	TMR2	FACh	TXSTA	F8Ch	LATD(5)
FEBh	PLUSW0(2)	FCBh	PR2	FABh	RCSTA	F8Bh	LATC
FEAh	FSR0H	FCAh	T2CON	FAAh	—	F8Ah	LATB
FE9h	FSR0L	FC9h	SSPBUF	FA9h	EEADR	F89h	LATA
FE8h	WREG	FC8h	SSPADD	FA8h	EEDATA	F88h	—
FE7h	INDF1(2)	FC7h	SSPSTAT	FA7h	EECON2	F87h	—
FE6h	POSTINC1(2)	FC6h	SSPCON1	FA6h	EECON1	F86h	—
FE5h	POSTDEC1(2)	FC5h	SSPCON2	FA5h	IPR3	F85h	—
FE4h	PREINC1(2)	FC4h	ADRESH	FA4h	PIR3	F84h	PORTE(5)
FE3h	PLUSW1(2)	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD(5)
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB
FE0h	BSR	FC0h	—	FA0h	PIE2	F80h	PORTA

Note 1: Unimplemented registers are read as '0'.

2: This is not a physical register.

3: Contents of register are dependent on WIN2:WIN0 bits in the CANCON register.

4: CANSTAT register is repeated in these locations to simplify application firmware. Unique names are given for each instance of the CANSTAT register due to the Microchip header file requirement.

5: These registers are not implemented on the PIC18F248 and PIC18F258.

9.2 PORTB, TRISB and LATB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding Data Direction register is TRISB. Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., put the contents of the output latch on the selected pin).

Read-modify-write operations on the LATB register, read and write the latched output value for PORTB.

EXAMPLE 9-2: INITIALIZING PORTB

```
CLRF    PORTB       ; Initialize PORTB by
                    ; clearing output
                    ; data latches
CLRF    LATB        ; Alternate method
                    ; to clear output
                    ; data latches
MOVLW   0CFh        ; Value used to
                    ; initialize data
                    ; direction
MOVWF   TRISB       ; Set RB3:RB0 as inputs
                    ; RB5:RB4 as outputs
                    ; RB7:RB6 as inputs
```

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit $\overline{\text{RBPU}}$ (INTCON2 register). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of the PORTB pins (RB7:RB4) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB7:RB4) are compared with the old value latched on the last read of PORTB. The “mismatch” outputs of RB7:RB4 are ORed together to generate the RB Port Change Interrupt with Flag bit RBIF (INTCON register).

This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

a) Any read or write of PORTB (except with the MOVFF instruction). This will end the mismatch condition.

b) Clear flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition and allow flag bit RBIF to be cleared.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

Note 1: While in Low-Voltage ICSP mode, the RB5 pin can no longer be used as a general purpose I/O pin and should not be held low during normal operation to protect against inadvertent ICSP mode entry.

2: When using Low-Voltage ICSP Programming (LVP), the pull-up on RB5 becomes disabled. If TRISB bit 5 is cleared, thereby setting RB5 as an output, LATB bit 5 must also be cleared for proper operation.

17.3.8 SLEEP OPERATION

In Master mode, all module clocks are halted and the transmission/reception will remain in that state until the device wakes from Sleep. After the device returns to normal mode, the module will continue to transmit/ receive data.

In Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device from Sleep.

17.3.9 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

17.3.10 BUS MODE COMPATIBILITY

Table 17-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

TABLE 17-1: SPI™ BUS MODES

Standard SPI Mode Terminology	Control Bits State	
	CKP	CKE
0, 0	0	1
0, 1	0	0
1, 0	1	1
1, 1	1	0

There is also an SMP bit which controls when the data is sampled.

TABLE 17-2: REGISTERS ASSOCIATED WITH SPI™ OPERATION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF(1)	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE(1)	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP(1)	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	1111 1111	1111 1111
TRISC	PORTC Data Direction Register								1111 1111	1111 1111
TRISA	—	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	-111 1111	-111 1111
SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
SSPSTAT	SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the MSSP in SPI™ mode.

Note 1: These registers or register bits are not implemented on the PIC18F248 and PIC18F258 and read as '0's.

17.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I^2C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I^2C protocol. It consists of all '0's with R/$\overline{W}$ = 0.

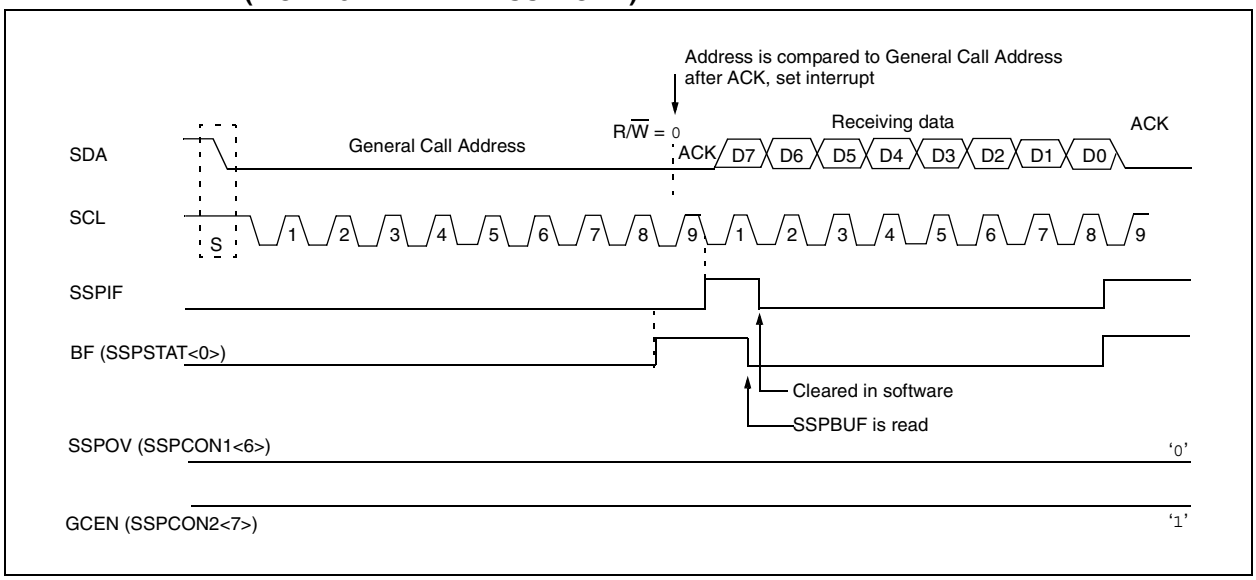
The general call address is recognized when the General Call Enable bit (GCEN) is enabled (SSPCON2<7> set). Following a Start bit detect, 8 bits are shifted into the SSPSR and the address is compared against the SSPADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF flag bit is set (eighth bit) and on the falling edge of the ninth bit ($\overline{ACK}$ bit), the SSPIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF. The value can be used to determine if the address was device specific or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match and the UA bit is set (SSPSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-bit Address mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 17-15).

FIGURE 17-15: SLAVE MODE GENERAL CALL ADDRESS SEQUENCE (7 OR 10-BIT ADDRESS MODE)



17.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 17-26).

b) SCL is sampled low before SDA is asserted low (Figure 17-27).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCLIF flag is set and
- the MSSP module is reset to its Idle state (Figure 17-26).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 17-28). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0 and during this time, if the SCL pins are sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note:	The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

FIGURE 17-26: BUS COLLISION DURING START CONDITION (SDA ONLY)

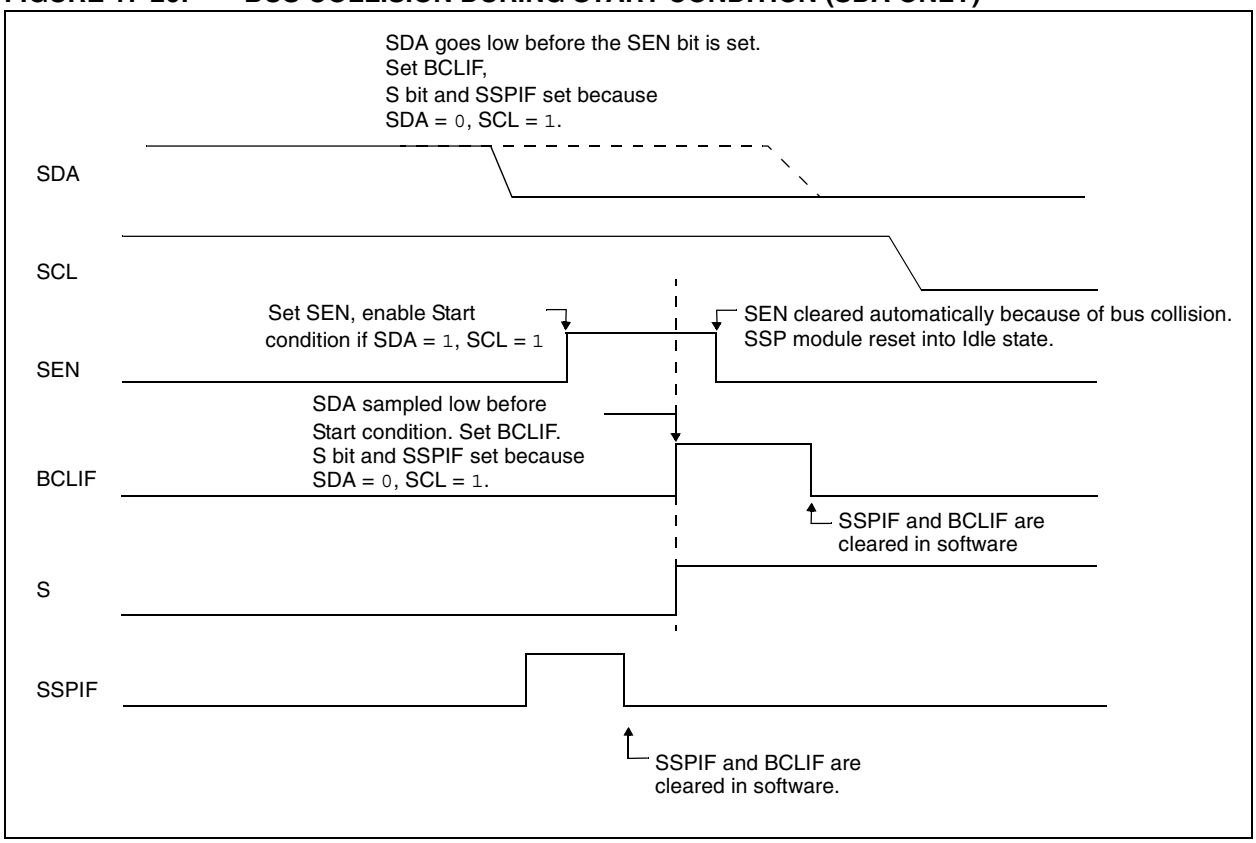


TABLE 18-4: BAUD RATES FOR ASYNCHRONOUS MODE (BRGH = 0)

BAUD RATE (Kbps)	Fosc = 40 MHz		SPBRG value (decimal)	33 MHz		SPBRG value (decimal)	25 MHz		SPBRG value (decimal)	20 MHz		SPBRG value (decimal)
	KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR	
0.3	NA	-	-	NA	-	-	NA	-	-	NA	-	-
1.2	NA	-	-	NA	-	-	NA	-	-	NA	-	-
2.4	NA	-	-	2.40	-0.07	214	2.40	-0.15	162	2.40	+0.16	129
9.6	9.62	+0.16	64	9.55	-0.54	53	9.53	-0.76	40	9.47	-1.36	32
19.2	18.94	-1.36	32	19.10	-0.54	26	19.53	+1.73	19	19.53	+1.73	15
76.8	78.13	+1.73	7	73.66	-4.09	6	78.13	+1.73	4	78.13	+1.73	3
96	89.29	-6.99	6	103.13	+7.42	4	97.66	+1.73	3	104.17	+8.51	2
300	312.50	+4.17	1	257.81	-14.06	1	NA	-	-	312.50	+4.17	0
500	625	+25.00	0	NA	-	-	NA	-	-	NA	-	-
HIGH	625	-	0	515.63	-	0	390.63	-	0	312.50	-	0
LOW	2.44	-	255	2.01	-	255	1.53	-	255	1.22	-	255

BAUD RATE (Kbps)	Fosc = 16 MHz		SPBRG value (decimal)	10 MHz		SPBRG value (decimal)	7.15909 MHz		SPBRG value (decimal)	5.0688 MHz		SPBRG value (decimal)
	KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR	
0.3	NA	-	-	NA	-	-	NA	-	-	NA	-	-
1.2	1.20	+0.16	207	1.20	+0.16	129	1.20	+0.23	92	1.20	0	65
2.4	2.40	+0.16	103	2.40	+0.16	64	2.38	-0.83	46	2.40	0	32
9.6	9.62	+0.16	25	9.77	+1.73	15	9.32	-2.90	11	9.90	+3.13	7
19.2	19.23	+0.16	12	19.53	+1.73	7	18.64	-2.90	5	19.80	+3.13	3
76.8	83.33	+8.51	2	78.13	+1.73	1	111.86	+45.65	0	79.20	+3.13	0
96	83.33	-13.19	2	78.13	-18.62	1	NA	-	-	NA	-	-
300	250	-16.67	0	156.25	-47.92	0	NA	-	-	NA	-	-
500	NA	-	-	NA	-	-	NA	-	-	NA	-	-
HIGH	250	-	0	156.25	-	0	111.86	-	0	79.20	-	0
LOW	0.98	-	255	0.61	-	255	0.44	-	255	0.31	-	255

BAUD RATE (Kbps)	Fosc = 4 MHz		SPBRG value (decimal)	3.579545 MHz		SPBRG value (decimal)	1 MHz		SPBRG value (decimal)	32.768 kHz		SPBRG value (decimal)
	KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR		KBAUD	% ERROR	
0.3	0.30	-0.16	207	0.30	+0.23	185	0.30	+0.16	51	0.26	-14.67	1
1.2	1.20	+1.67	51	1.19	-0.83	46	1.20	+0.16	12	NA	-	-
2.4	2.40	+1.67	25	2.43	+1.32	22	2.23	-6.99	6	NA	-	-
9.6	8.93	-6.99	6	9.32	-2.90	5	7.81	-18.62	1	NA	-	-
19.2	20.83	+8.51	2	18.64	-2.90	2	15.63	-18.62	0	NA	-	-
76.8	62.50	-18.62	0	55.93	-27.17	0	NA	-	-	NA	-	-
96	NA	-	-	NA	-	-	NA	-	-	NA	-	-
300	NA	-	-	NA	-	-	NA	-	-	NA	-	-
500	NA	-	-	NA	-	-	NA	-	-	NA	-	-
HIGH	62.50	-	0	55.93	-	0	15.63	-	0	0.51	-	0
LOW	0.24	-	255	0.22	-	255	0.06	-	255	0.002	-	255

18.3 USART Synchronous Master Mode

In Synchronous Master mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA register). In addition, enable bit SPEN (RCSTA register) is set in order to configure the RC6/TX/CK and RC7/RX/DT I/O pins to CK (clock) and DT (data) lines, respectively. The Master mode indicates that the processor transmits the master clock on the CK line. The Master mode is entered by setting bit CSRC (TXSTA register).

18.3.1 USART SYNCHRONOUS MASTER TRANSMISSION

The USART transmitter block diagram is shown in Figure 18-1. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The shift register obtains its data from the Read/Write Transmit Buffer register (TXREG). The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once the TXREG register transfers the data to the TSR register (occurs in one TCY), the TXREG is empty and interrupt bit TXIF (PIR1 register) is set. The interrupt can be enabled/disabled by setting/clearing enable bit TXIE (PIE1 register). Flag bit TXIF will be set regardless of the state of enable bit TXIE and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register. While flag bit, TXIF, indicates the status of the TXREG register, another bit, TRMT (TXSTA register), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory, so it is not available to the user.

Steps to follow when setting up a Synchronous Master Transmission:

1. Initialize the SPBRG register for the appropriate baud rate (**Section 18.1 "USART Baud Rate Generator (BRG)"**).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set bit TX9.
5. Enable the transmission by setting bit TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREG register.

Note:	TXIF is not cleared immediately upon loading data into the transmit buffer TXREG. The flag bit becomes valid in the second instruction cycle following the load instruction.

TABLE 18-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF(1)	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE(1)	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP(1)	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	1111 1111	1111 1111
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000u
TXREG	USART Transmit Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

Note 1: These registers or register bits are not implemented on the PIC18F248 and PIC18F258 and read as '0's.

19.0 CAN MODULE

19.1 Overview

The Controller Area Network (CAN) module is a serial interface, useful for communicating with other peripherals or microcontroller devices. This interface/protocol was designed to allow communications within noisy environments.

The CAN module is a communication controller, implementing the CAN 2.0 A/B protocol as defined in the BOSCH specification. The module will support CAN 1.2, CAN 2.0A, CAN 2.0B Passive and CAN 2.0B Active versions of the protocol. The module implementation is a full CAN system. The CAN specification is not covered within this data sheet. The reader may refer to the BOSCH CAN specification for further details.

The module features are as follows:

- Complies with ISO CAN Conformance Test
- Implementation of the CAN protocol CAN 1.2, CAN 2.0A and CAN 2.0B
- Standard and extended data frames
- 0-8 bytes data length
- Programmable bit rate up to 1 Mbit/sec
- Support for remote frames
- Double-buffered receiver with two prioritized received message storage buffers
- 6 full (standard/extended identifier) acceptance filters, 2 associated with the high priority receive buffer and 4 associated with the low priority receive buffer
- 2 full acceptance filter masks, one each associated with the high and low priority receive buffers
- Three transmit buffers with application specified prioritization and abort capability
- Programmable wake-up functionality with integrated low-pass filter
- Programmable Loopback mode supports self-test operation
- Signaling via interrupt capabilities for all CAN receiver and transmitter error states
- Programmable clock source
- Programmable link to timer module for time-stamping and network synchronization
- Low-power Sleep mode

19.1.1 OVERVIEW OF THE MODULE

The CAN bus module consists of a protocol engine and message buffering and control. The CAN protocol engine handles all functions for receiving and transmitting messages on the CAN bus. Messages are transmitted by first loading the appropriate data registers. Status and errors can be checked by reading the appropriate registers. Any message detected on the CAN bus is checked for errors and then matched against filters to see if it should be received and stored in one of the 2 receive registers.

The CAN module supports the following frame types:

- Standard Data Frame
- Extended Data Frame
- Remote Frame
- Error Frame
- Overload Frame Reception
- Interframe Space

CAN module uses RB3/CANRX and RB2/CANTX/INT2 pins to interface with CAN bus. In order to configure CANRX and CANTX as CAN interface:

- bit TRISB<3> must be set;
- bit TRISB<2> must be cleared.

19.1.2 TRANSMIT/RECEIVE BUFFERS

The PIC18FXX8 has three transmit and two receive buffers, two acceptance masks (one for each receive buffer) and a total of six acceptance filters. Figure 19-1 is a block diagram of these buffers and their connection to the protocol engine.

19.2.3.1 Message Acceptance Filters and Masks

This subsection describes the message acceptance filters and masks for the CAN receive buffers.

REGISTER 19-21: RXFnSIDH: RECEIVE ACCEPTANCE FILTER n STANDARD IDENTIFIER FILTER, HIGH BYTE REGISTERS

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
SID10	SID9	SID8	SID7	SID6	SID5	SID4	SID3
bit 7							bit 0

bit 7-0 **SID10:SID3:** Standard Identifier Filter bits if EXIDEN = 0 or
Extended Identifier Filter bits EID28:EID21 if EXIDEN = 1

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

REGISTER 19-22: RXFnSIDL: RECEIVE ACCEPTANCE FILTER n STANDARD IDENTIFIER FILTER, LOW BYTE REGISTERS

R/W-x	R/W-x	R/W-x	U-0	R/W-x	U-0	R/W-x	R/W-x
SID2	SID1	SID0	—	EXIDEN	—	EID17	EID16
bit 7							bit 0

bit 7-5 **SID2:SID0:** Standard Identifier Filter bits if EXIDEN = 0 or
Extended Identifier Filter bits EID20:EID18 if EXIDEN = 1

bit 4 **Unimplemented:** Read as '0'

bit 3 **EXIDEN:** Extended Identifier Filter Enable bit
1 = Filter will only accept extended ID messages
0 = Filter will only accept standard ID messages

bit 2 **Unimplemented:** Read as '0'

bit 1-0 **EID17:EID16:** Extended Identifier Filter bits

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

19.6 Message Acceptance Filters and Masks

The message acceptance filters and masks are used to determine if a message in the message assembly buffer should be loaded into either of the receive buffers. Once a valid message has been received into the MAB, the identifier fields of the message are compared to the filter values. If there is a match, that message will be loaded into the appropriate receive buffer. The filter masks are used to determine which bits in the identifier are examined with the filters. A truth table is shown below in Table 19-2 that indicates how each bit in the identifier is compared to the masks and filters to determine if a message should be loaded into a receive buffer. The mask essentially determines which bits to apply the acceptance filters to. If any mask bit is set to a zero, then that bit will automatically be accepted regardless of the filter bit.

TABLE 19-2: FILTER/MASK TRUTH TABLE

Mask bit n	Filter bit n	Message Identifier bit n001	Accept or Reject bit n
0	x	x	Accept
1	0	0	Accept
1	0	1	Reject
1	1	0	Reject
1	1	1	Accept

Legend: x = don't care

As shown in the receive buffer block diagram (Figure 19-4), acceptance filters RXF0 and RXF1 and filter mask RXM0 are associated with RXB0. Filters RXF2, RXF3, RXF4 and RXF5 and mask RXM1 are associated with RXB1. When a filter matches and a message is loaded into the receive buffer, the filter number that enabled the message reception is loaded into the FILHIT bit(s).

For RXB1, the RXB1CON register contains the FILHIT<2:0> bits. They are coded as follows:

- 101 = Acceptance Filter 5 (RXF5)
- 100 = Acceptance Filter 4 (RXF4)
- 011 = Acceptance Filter 3 (RXF3)
- 010 = Acceptance Filter 2 (RXF2)
- 001 = Acceptance Filter 1 (RXF1)
- 000 = Acceptance Filter 0 (RXF0)

Note: '000' and '001' can only occur if the RXB0DBEN bit is set in the RXB0CON register allowing RXB0 messages to rollover into RXB1.

The coding of the RXB0DBEN bit enables these three bits to be used similarly to the FILHIT bits and to distinguish a hit on filter RXF0 and RXF1, in either RXB0, or after a rollover into RXB1.

- 111 = Acceptance Filter 1 (RXF1)
- 110 = Acceptance Filter 0 (RXF0)
- 001 = Acceptance Filter 1 (RXF1)
- 000 = Acceptance Filter 0

If the RXB0DBEN bit is clear, there are six codes corresponding to the six filters. If the RXB0DBEN bit is set, there are six codes corresponding to the six filters plus two additional codes corresponding to RXF0 and RXF1 filters that rollover into RXB1.

If more than one acceptance filter matches, the FILHIT bits will encode the binary value of the lowest numbered filter that matched. In other words, if filter RXF2 and filter RXF4 match, FILHIT will be loaded with the value for RXF2. This essentially prioritizes the acceptance filters with a lower number filter having higher priority. Messages are compared to filters in ascending order of filter number.

The mask and filter registers can only be modified when the PIC18FXX8 is in Configuration mode. The mask and filter registers cannot be read outside of Configuration mode. When outside of Configuration mode, all mask and filter registers will be read as '0'.

FIGURE 19-6: MESSAGE ACCEPTANCE MASK AND FILTER OPERATION

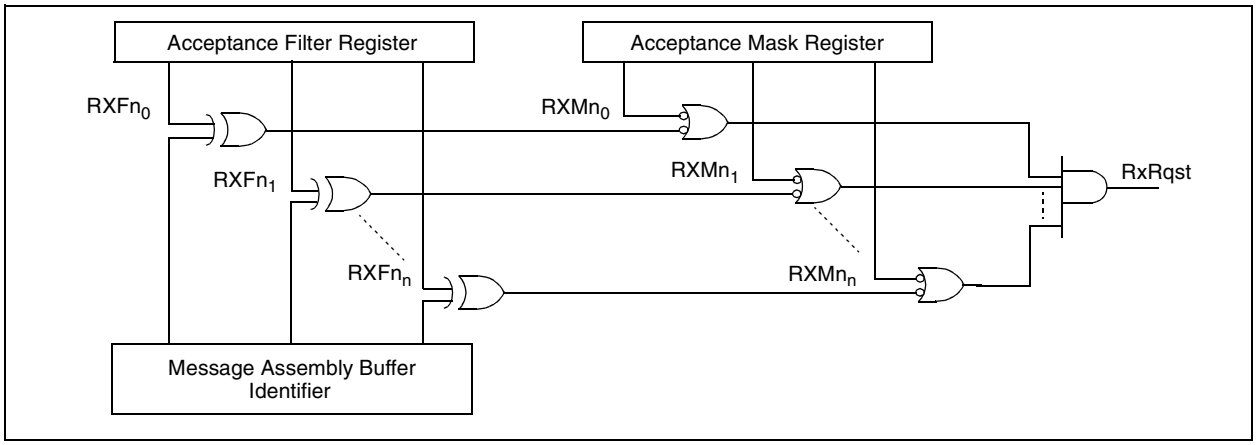
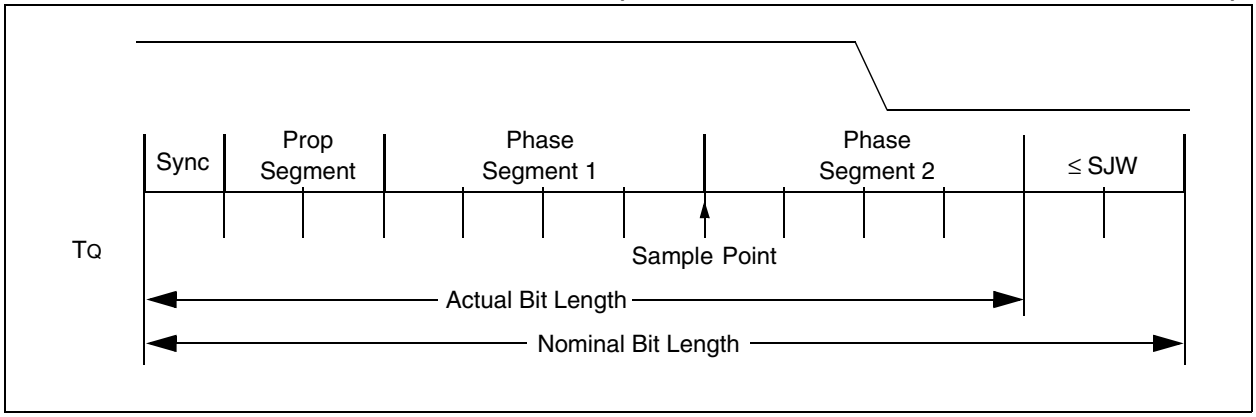


FIGURE 19-9: SHORTENING A BIT PERIOD (SUBTRACTING SJW FROM PHASE SEGMENT 2)



19.9 Programming Time Segments

Some requirements for programming of the time segments:

- Prop Seg + Phase Seg 1 ≥ Phase Seg 2
- Phase Seg 2 ≥ Sync Jump Width

For example, assume that a 125 kHz CAN baud rate is desired using 20 MHz for FOSC. With a TOSC of 50 ns, a baud rate prescaler value of 04h gives a TQ of 500 ns. To obtain a nominal bit rate of 125 kHz, the nominal bit time must be 8 µs or 16 TQ.

Using 1 TQ for the Sync Segment, 2 TQ for the Propagation Segment and 7 TQ for Phase Segment 1 would place the sample point at 10 TQ after the transition. This leaves 6 TQ for Phase Segment 2.

By the rules above, the Sync Jump Width could be the maximum of 4 TQ. However, normally a large SJW is only necessary when the clock generation of the different nodes is inaccurate or unstable, such as using ceramic resonators. Typically, an SJW of 1 is enough.

19.10 Oscillator Tolerance

As a rule of thumb, the bit timing requirements allow ceramic resonators to be used in applications with transmission rates of up to 125 Kbit/sec. For the full bus speed range of the CAN protocol, a quartz oscillator is required. A maximum node-to-node oscillator variation of 1.7% is allowed.

19.11 Bit Timing Configuration Registers

The Configuration registers (BRGCON1, BRGCON2, BRGCON3) control the bit timing for the CAN bus interface. These registers can only be modified when the PIC18FXX8 is in Configuration mode.

19.11.1 BRGCON1

The BRP bits control the baud rate prescaler. The SJW<1:0> bits select the synchronization jump width in terms of multiples of TQ.

19.11.2 BRGCON2

The PRSEG bits set the length of the Propagation Segment in terms of TQ. The SEG1PH bits set the length of Phase Segment 1 in TQ. The SAM bit controls how many times the RXCAN pin is sampled. Setting this bit to a '1' causes the bus to be sampled three times; twice at TQ/2 before the sample point and once at the normal sample point (which is at the end of Phase Segment 1). The value of the bus is determined to be the value read during at least two of the samples. If the SAM bit is set to a '0', then the RXCAN pin is sampled only once at the sample point. The SEG2PHTS bit controls how the length of Phase Segment 2 is determined. If this bit is set to a '1', then the length of Phase Segment 2 is determined by the SEG2PH bits of BRGCON3. If the SEG2PHTS bit is set to a '0', then the length of Phase Segment 2 is the greater of Phase Segment 1 and the information processing time (which is fixed at 2 TQ for the PIC18FXX8).

19.11.3 BRGCON3

The PHSEG2<2:0> bits set the length (in TQ) of Phase Segment 2 if the SEG2PHTS bit is set to a '1'. If the SEG2PHTS bit is set to a '0', then the PHSEG2<2:0> bits have no effect.

19.12 Error Detection

The CAN protocol provides sophisticated error detection mechanisms. The following errors can be detected.

19.12.1 CRC ERROR

With the Cyclic Redundancy Check (CRC), the transmitter calculates special check bits for the bit sequence, from the start of a frame until the end of the data field. This CRC sequence is transmitted in the CRC field. The receiving node also calculates the CRC sequence using the same formula and performs a comparison to the received sequence. If a mismatch is detected, a CRC error has occurred and an error frame is generated. The message is repeated.

19.12.2 ACKNOWLEDGE ERROR

In the Acknowledge field of a message, the transmitter checks if the Acknowledge slot (which was sent out as a recessive bit) contains a dominant bit. If not, no other node has received the frame correctly. An Acknowledge Error has occurred; an error frame is generated and the message will have to be repeated.

19.12.3 FORM ERROR

If a node detects a dominant bit in one of the four segments, including end of frame, interframe space, Acknowledge delimiter or CRC delimiter, then a Form Error has occurred and an error frame is generated. The message is repeated.

19.12.4 BIT ERROR

A Bit Error occurs if a transmitter sends a dominant bit and detects a recessive bit, or if it sends a recessive bit and detects a dominant bit, when monitoring the actual bus level and comparing it to the just transmitted bit. In the case where the transmitter sends a recessive bit and a dominant bit is detected during the arbitration field and the Acknowledge slot, no Bit Error is generated because normal arbitration is occurring.

19.12.5 STUFF BIT ERROR

If, between the start of frame and the CRC delimiter, six consecutive bits with the same polarity are detected, the bit stuffing rule has been violated. A Stuff Bit Error occurs and an error frame is generated. The message is repeated.

19.12.6 ERROR STATES

Detected errors are made public to all other nodes via error frames. The transmission of the erroneous message is aborted and the frame is repeated as soon as possible. Furthermore, each CAN node is in one of the three error states “error-active”, “error-passive” or “bus-off” according to the value of the internal error counters. The error-active state is the usual state, where the bus node can transmit messages and activate error frames (made of dominant bits) without any restrictions. In the error-passive state, messages and passive error frames (made of recessive bits) may be transmitted. The bus-off state makes it temporarily impossible for the station to participate in the bus communication. During this state, messages can neither be received nor transmitted.

19.12.7 ERROR MODES AND ERROR COUNTERS

The PIC18FXX8 contains two error counters: the Receive Error Counter (RXERRCNT) and the Transmit Error Counter (TXERRCNT). The values of both counters can be read by the MCU. These counters are incremented or decremented in accordance with the CAN bus specification.

The PIC18FXX8 is error-active if both error counters are below the error-passive limit of 128. It is error-passive if at least one of the error counters equals or exceeds 128. It goes to bus-off if the transmit error counter equals or exceeds the bus-off limit of 256. The device remains in this state until the bus-off recovery sequence is received. The bus-off recovery sequence consists of 128 occurrences of 11 consecutive recessive bits (see Figure 19-10). Note that the CAN module, after going bus-off, will recover back to error-active without any intervention by the MCU if the bus remains Idle for 128 x 11 bit times. If this is not desired, the error Interrupt Service Routine should address this. The current error mode of the CAN module can be read by the MCU via the COMSTAT register.

Additionally, there is an Error State Warning flag bit, EWARN, which is set if at least one of the error counters equals or exceeds the error warning limit of 96. EWARN is reset if both error counters are less than the error warning limit.

NOTES:

REGISTER 24-11: DEVID1: DEVICE ID REGISTER 1 FOR PIC18FXX8 DEVICES (BYTE ADDRESS 3FFFFEh)

R	R	R	R	R	R	R	R
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0
bit 7							bit 0

bit 7-5 **DEV2:DEV0:** Device ID bits

These bits are used with the DEV<10:3> bits in the Device ID Register 2 to identify the part number.

000 = PIC18F248
001 = PIC18F448
010 = PIC18F258
011 = PIC18F458

bit 4-0 **REV4:REV0:** Revision ID bits

These bits are used to indicate the device revision.

Legend:		
R = Readable bit	P = Programmable bit	U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed		u = Unchanged from programmed state

REGISTER 24-12: DEVID2: DEVICE ID REGISTER 2 FOR PIC18FXX8 DEVICES (BYTE ADDRESS 3FFFFFh)

R	R	R	R	R	R	R	R
DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3
bit 7							bit 0

bit 7-0 **DEV10:DEV3:** Device ID bits

These bits are used with the DEV<2:0> bits in the Device ID Register 1 to identify the part number.

00001000 = PIC18FXX8

Legend:		
R = Readable bit	P = Programmable bit	U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed		u = Unchanged from programmed state

DECFSZ **Decrement f, Skip if 0**

Syntax: [*label*] DECFSZ f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) – 1 → dest,
skip if result = 0

Status Affected: None

Encoding:

0010	11da	ffff	ffff

Description: The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If the result is '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    DECFSZ  CNT
        GOTO    LOOP
CONTINUE
```

Before Instruction
PC = Address (HERE)
After Instruction
CNT = CNT – 1
If CNT = 0;
PC = Address (CONTINUE)
If CNT ≠ 0;
PC = Address (HERE + 2)

DCFSNZ **Decrement f, Skip if not 0**

Syntax: [*label*] DCFSNZ f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) – 1 → dest,
skip if result ≠ 0

Status Affected: None

Encoding:

0100	11da	ffff	ffff

Description: The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    DCFSNZ TEMP
ZERO    :
NZERO   :
```

Before Instruction
TEMP = ?
After Instruction
TEMP = TEMP – 1,
If TEMP = 0;
PC = Address (ZERO)
If TEMP ≠ 0;
PC = Address (NZERO)

TBLRD **Table Read**

Syntax: [*label*] TBLRD (*; *+; *-; +*)

Operands: None

Operation: if TBLRD *,
(Prog Mem (TBLPTR)) → TABLAT;
TBLPTR – No Change;
if TBLRD *+,
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) + 1 → TBLPTR;
if TBLRD *-,
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) – 1 → TBLPTR;
if TBLRD +*,
(TBLPTR) + 1 → TBLPTR;
(Prog Mem (TBLPTR)) → TABLAT

Status Affected: None

Encoding:

0000	0000	0000	10nn nn=0 * =1 *+ =2 *- =3 +*

Description: This instruction is used to read the contents of Program Memory (P.M.). To address the program memory, a pointer called Table Pointer (TBLPTR) is used.
The TBLPTR (a 21-bit pointer) points to each byte in the program memory. TBLPTR has a 2-Mbyte address range.

TBLPTR[0] = 0: Least Significant Byte of Program Memory Word

TBLPTR[0] = 1: Most Significant Byte of Program Memory Word

The TBLRD instruction can modify the value of TBLPTR as follows:
- no change
- post-increment
- post-decrement
- pre-increment

Words: 1

Cycles: 2

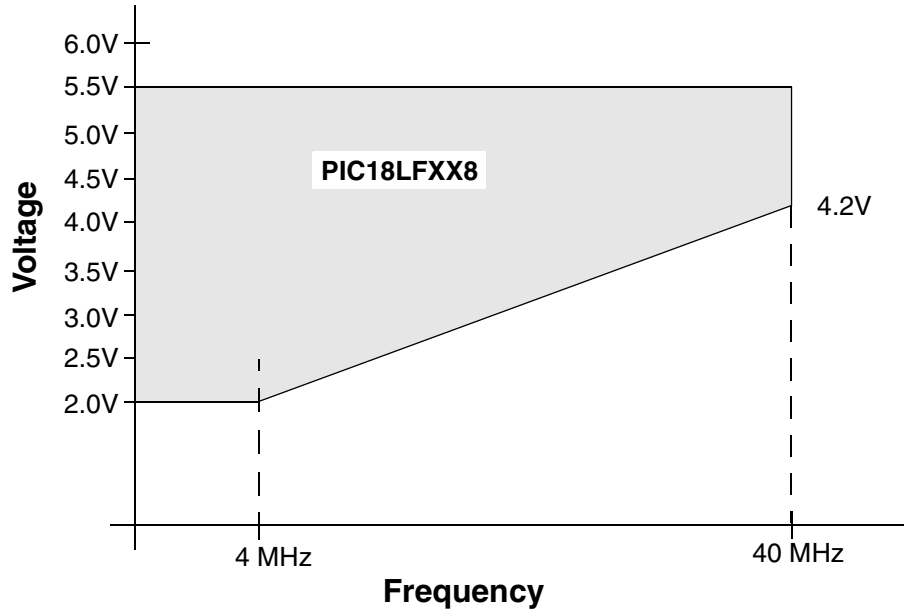
Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	No operation	No operation
No operation	No operation (Read Program Memory)	No operation	No operation (Write TABLAT)

Example 1:
```
TBLRD  *+ ;
```
Before Instruction
TABLAT = 0x55
TBLPTR = 0x00A356
MEMORY(0x00A356) = 0x34
After Instruction
TABLAT = 0x34
TBLPTR = 0x00A357

Example 2:
```
TBLRD  +* ;
```
Before Instruction
TABLAT = 0xAA
TBLPTR = 0x01A357
MEMORY(0x01A357) = 0x12
MEMORY(0x01A358) = 0x34
After Instruction
TABLAT = 0x34
TBLPTR = 0x01A358

FIGURE 27-3: PIC18LFXX8 VOLTAGE-FREQUENCY GRAPH (INDUSTRIAL)



FMAX = (16.36 MHz/V) (VDDAPPMIN – 2.0V) + 4 MHz, if VDDAPPMIN ≤ 4.2V = 40 MHz, if VDDAPPMIN > 4.2V

Note: VDDAPPMIN is the minimum voltage of the PICmicro® device in the application.

27.3.2 TIMING CONDITIONS

The temperature and voltages specified in Table 27-5 apply to all timing specifications unless otherwise noted. Figure 27-5 specifies the load conditions for the timing specifications.

TABLE 27-5: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

AC CHARACTERISTICS	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended Operating voltage VDD range as described in DC specification, **Section 27.1 "DC Characteristics"**. LF parts operate for industrial temperatures only.

FIGURE 27-5: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS

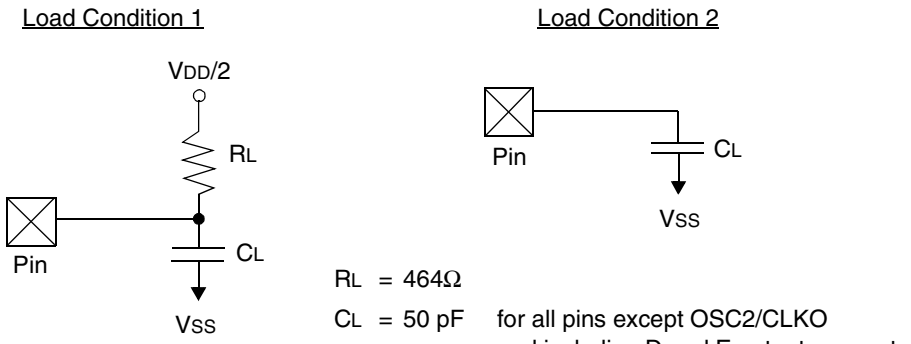


FIGURE 28-11: TYPICAL AND MAXIMUM IDD vs. VDD (TIMER1 AS MAIN OSCILLATOR 32.768 kHz, C1 AND C2 = 47 pF)

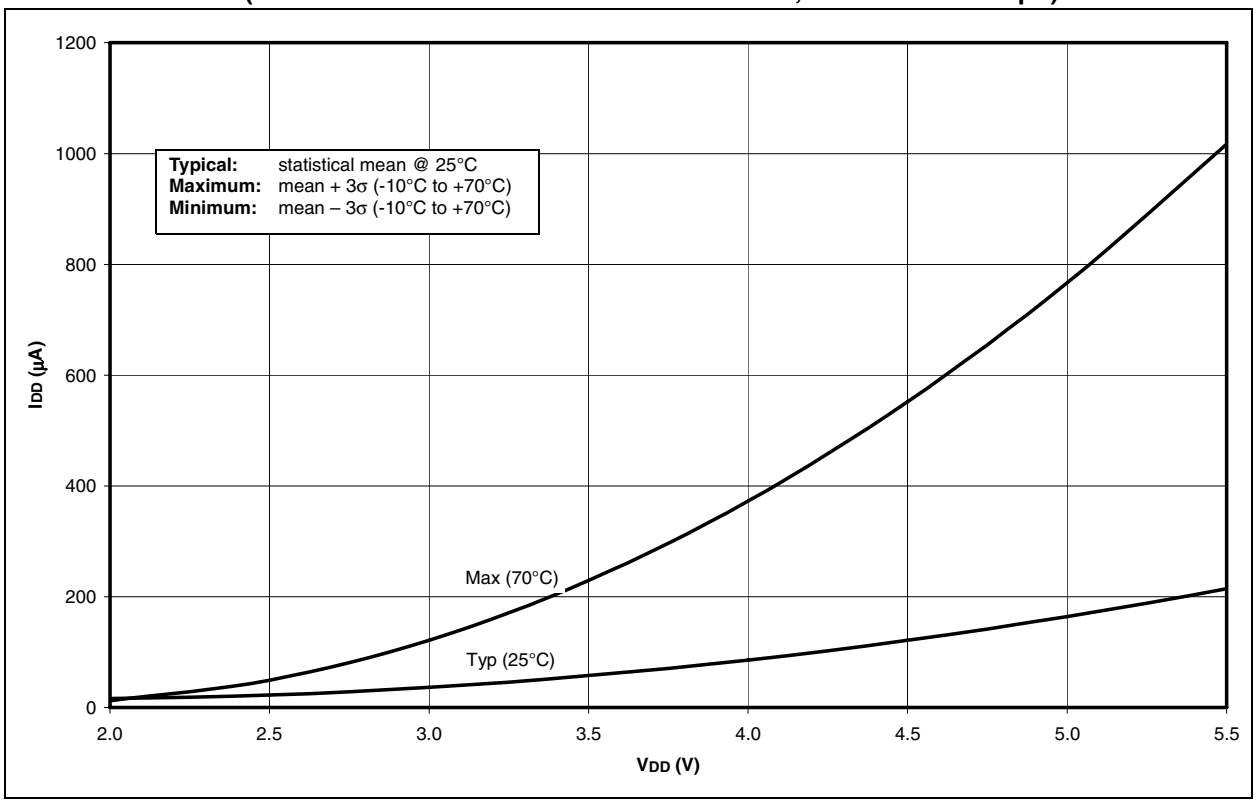


FIGURE 28-12: AVERAGE FOSC vs. VDD FOR VARIOUS VALUES OF R (RC MODE, C = 20 pF, +25°C)

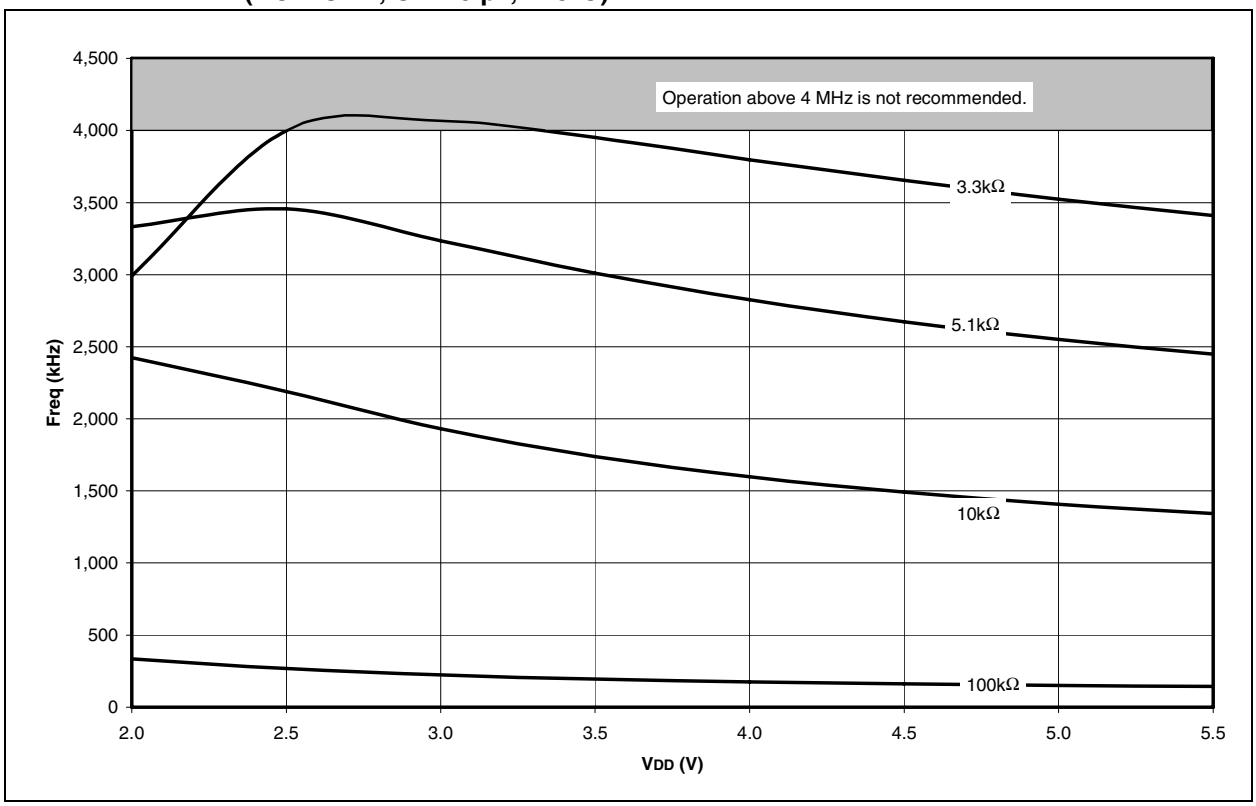


FIGURE 28-21: TYPICAL, MINIMUM AND MAXIMUM V_{OH} vs. I_{OH} (V_{DD} = 5V, -40°C TO +125°C)

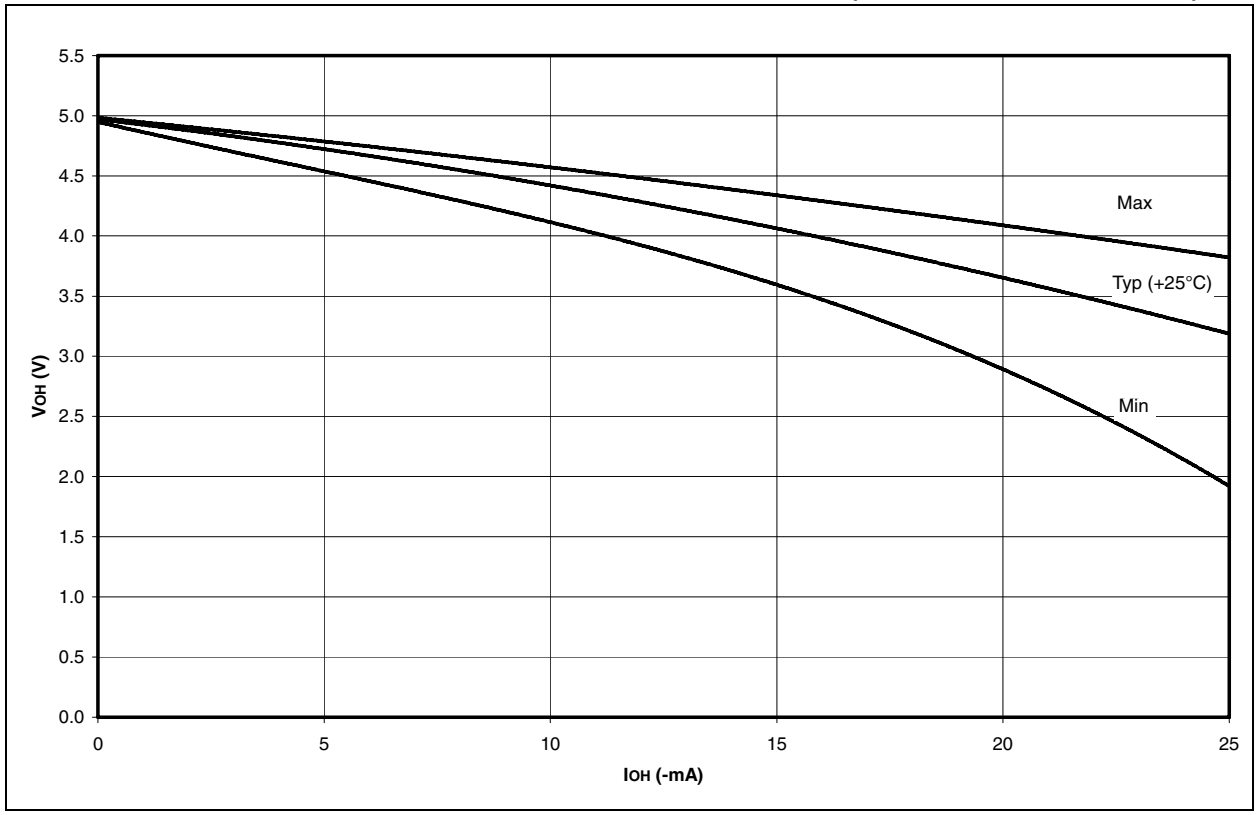


FIGURE 28-22: TYPICAL, MINIMUM AND MAXIMUM V_{OH} vs. I_{OH} (V_{DD} = 3V, -40°C TO +125°C)