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Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	27648
Number of I/O	37
Number of Gates	90000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	108-WFQFN
Supplier Device Package	108-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/afs090-2qng108

With Fusion, Microsemi also introduces the Analog Quad I/O structure (Figure 1-1). Each quad consists of three analog inputs and one gate driver. Each quad can be configured in various built-in circuit combinations, such as three prescaler circuits, three digital input circuits, a current monitor circuit, or a temperature monitor circuit. Each prescaler has multiple scaling factors programmed by FPGA signals to support a large range of analog inputs with positive or negative polarity. When the current monitor circuit is selected, two adjacent analog inputs measure the voltage drop across a small external sense resistor. For more information, refer to the "Analog System Characteristics" section on page 2-117. Built-in operational amplifiers amplify small voltage signals for accurate current measurement. One analog input in each quad can be connected to an external temperature monitor diode. In addition to the external temperature monitor diode(s), a Fusion device can monitor an internal temperature diode using dedicated channel 31 of the ADCMUX.

Figure 1-1 on page 1-5 illustrates a typical use of the Analog Quad I/O structure. The Analog Quad shown is configured to monitor and control an external power supply. The AV pad measures the source of the power supply. The AC pad measures the voltage drop across an external sense resistor to calculate current. The AG MOSFET gate driver pad turns the external MOSFET on and off. The AT pad measures the load-side voltage level.

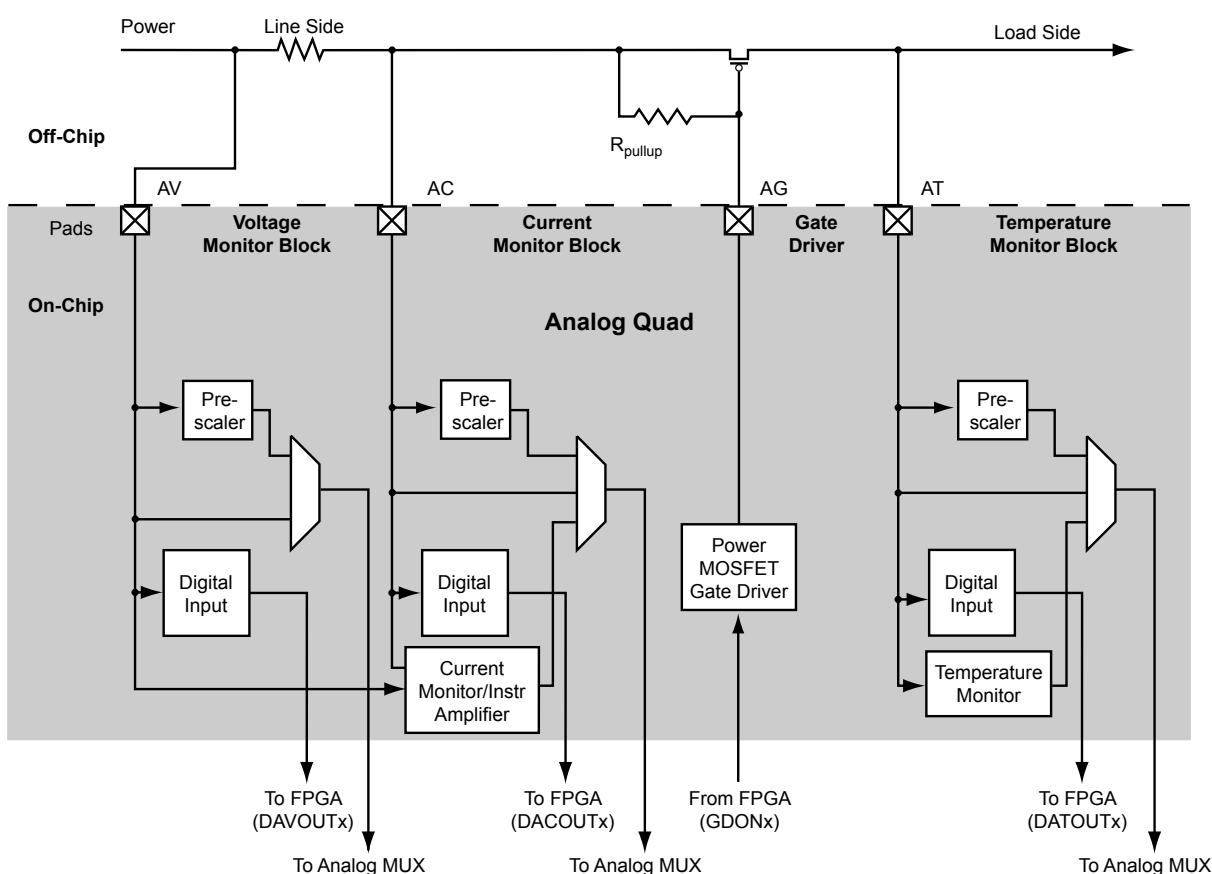


Figure 1-1 • Analog Quad

RAM512X18 exhibits slightly different behavior from RAM4K9, as it has dedicated read and write ports.

WW and RW

These signals enable the RAM to be configured in one of the two allowable aspect ratios (Table 2-30).

Table 2-30 • Aspect Ratio Settings for WW[1:0]

WW[1:0]	RW[1:0]	D×W
01	01	512×9
10	10	256×18
00, 11	00, 11	Reserved

WD and RD

These are the input and output data signals, and they are 18 bits wide. When a 512×9 aspect ratio is used for write, WD[17:9] are unused and must be grounded. If this aspect ratio is used for read, then RD[17:9] are undefined.

WADDR and RADDR

These are read and write addresses, and they are nine bits wide. When the 256×18 aspect ratio is used for write or read, WADDR[8] or RADDR[8] are unused and must be grounded.

WCLK and RCLK

These signals are the write and read clocks, respectively. They are both active high.

WEN and REN

These signals are the write and read enables, respectively. They are both active low by default. These signals can be configured as active high.

RESET

This active low signal resets the output to zero, disables reads and/or writes from the SRAM block, and clears the data hold registers when asserted. It does not reset the contents of the memory.

PIPE

This signal is used to specify pipelined read on the output. A Low on PIPE indicates a nonpipelined read, and the data appears on the output in the same clock cycle. A High indicates a pipelined read, and data appears on the output in the next clock cycle.

Clocking

The dual-port SRAM blocks are only clocked on the rising edge. SmartGen allows falling-edge-triggered clocks by adding inverters to the netlist, hence achieving dual-port SRAM blocks that are clocked on either edge (rising or falling). For dual-port SRAM, each port can be clocked on either edge or by separate clocks, by port.

Fusion devices support inversion (bubble pushing) throughout the FPGA architecture, including the clock input to the SRAM modules. Inversions added to the SRAM clock pin on the design schematic or in the HDL code will be automatically accounted for during design compile without incurring additional delay in the clock path.

The two-port SRAM can be clocked on the rising edge or falling edge of WCLK and RCLK.

If negative-edge RAM and FIFO clocking is selected for memory macros, clock edge inversion management (bubble pushing) is automatically used within the Fusion development tools, without performance penalty.

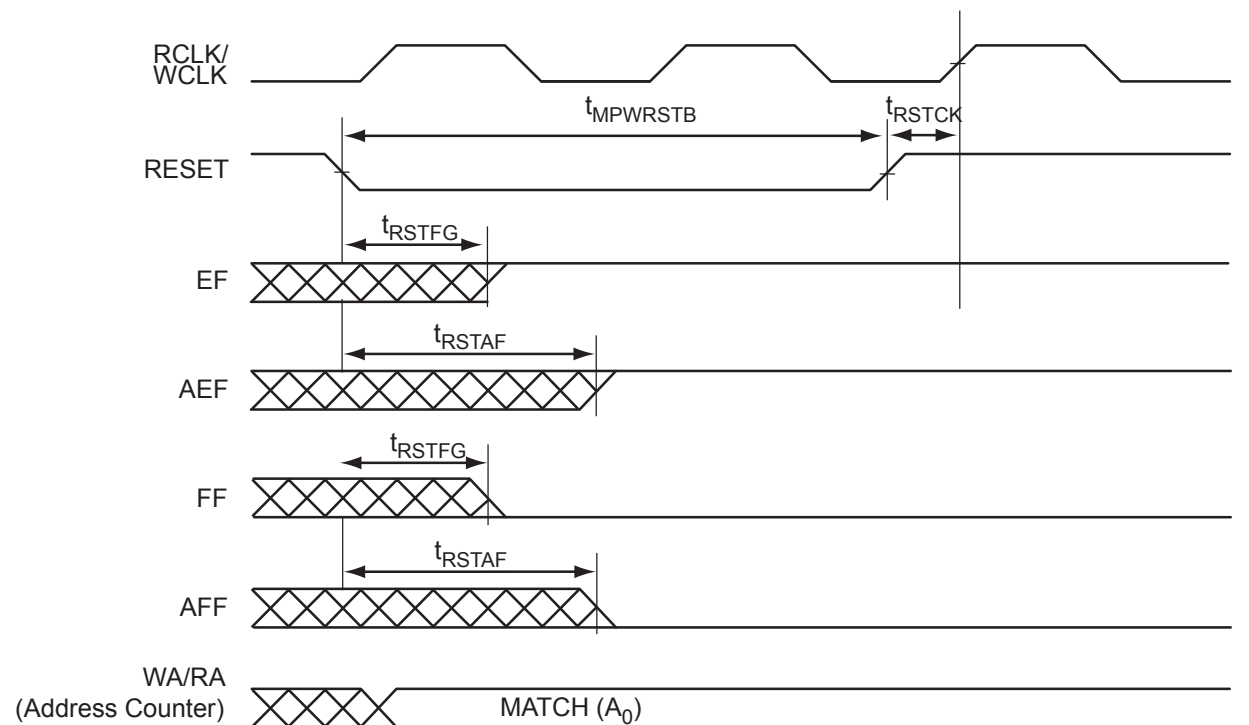


Figure 2-59 • FIFO Reset

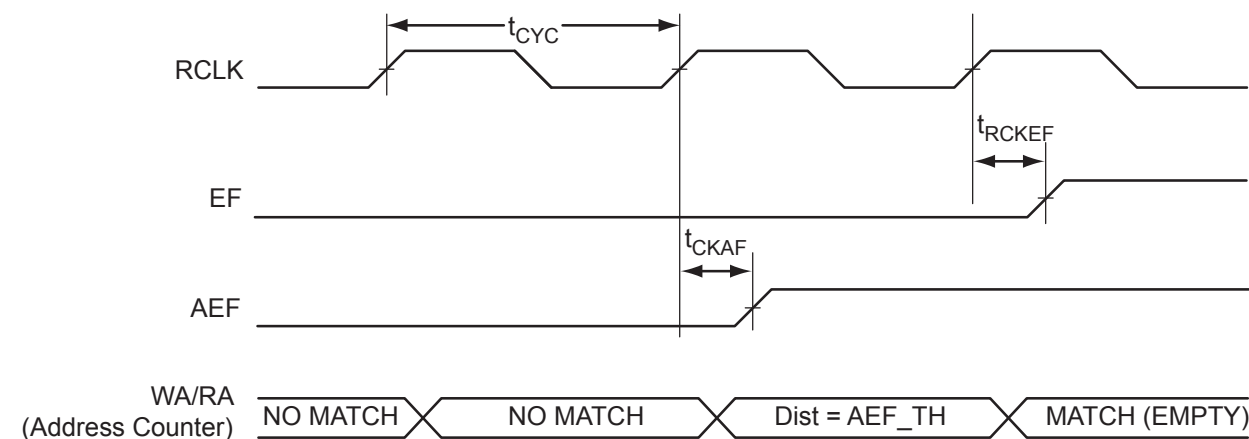


Figure 2-60 • FIFO EMPTY Flag and AEMPTY Flag Assertion

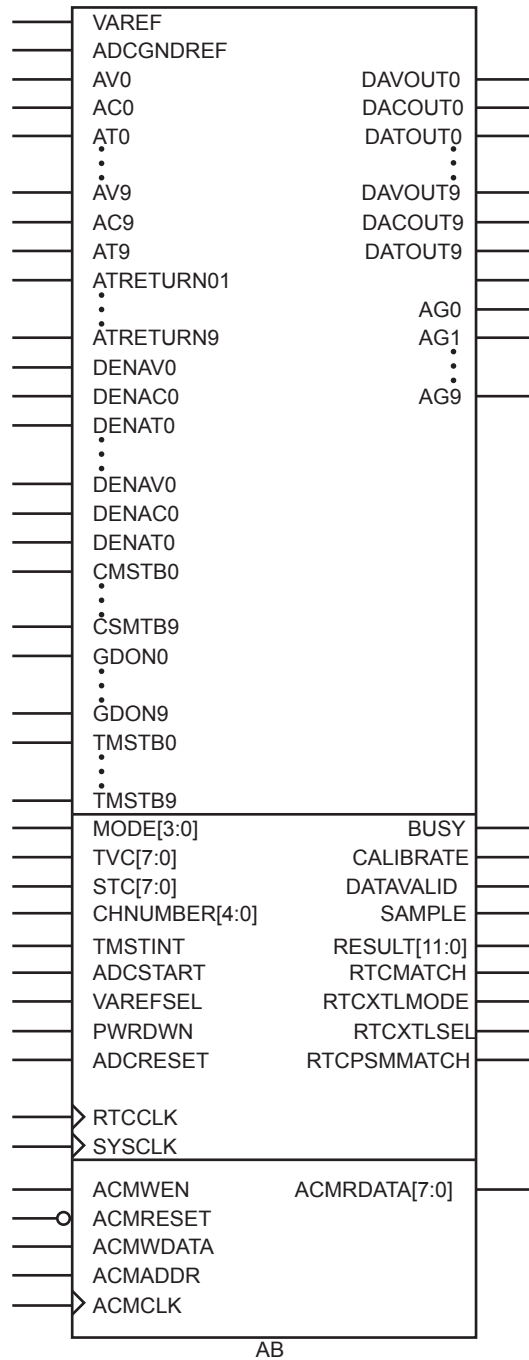


Figure 2-64 • Analog Block Macro

Table 2-36 • Analog Block Pin Description (continued)

Signal Name	Number of Bits	Direction	Function	Location of Details
AG6	1	Output		Analog Quad
AT6	1	Input		Analog Quad
ATRETURN67	1	Input	Temperature monitor return shared by Analog Quads 6 and 7	Analog Quad
AV7	1	Input	Analog Quad 7	Analog Quad
AC7	1	Input		Analog Quad
AG7	1	Output		Analog Quad
AT7	1	Input		Analog Quad
AV8	1	Input	Analog Quad 8	Analog Quad
AC8	1	Input		Analog Quad
AG8	1	Output		Analog Quad
AT8	1	Input		Analog Quad
ATRETURN89	1	Input	Temperature monitor return shared by Analog Quads 8 and 9	Analog Quad
AV9	1	Input	Analog Quad 9	Analog Quad
AC9	1	Input		Analog Quad
AG9	1	Output		Analog Quad
AT9	1	Input		Analog Quad
RTCMATCH	1	Output	MATCH	RTC
RTCPSMMATCH	1	Output	MATCH connected to VRPSM	RTC
RTCXTLMODE[1:0]	2	Output	Drives XTLOSC RTCMODE[1:0] pins	RTC
RTCXTLSEL	1	Output	Drives XTLOSC MODESEL pin	RTC
RTCCLK	1	Input	RTC clock input	RTC

Analog Quad

With the Fusion family, Microsemi introduces the Analog Quad, shown in [Figure 2-65 on page 2-81](#), as the basic analog I/O structure. The Analog Quad is a four-channel system used to precondition a set of analog signals before sending it to the ADC for conversion into a digital signal. To maximize the usefulness of the Analog Quad, the analog input signals can also be configured as LVTTTL digital input signals. The Analog Quad is divided into four sections.

The first section is called the Voltage Monitor Block, and its input pin is named AV. It contains a two-channel analog multiplexer that allows an incoming analog signal to be routed directly to the ADC or allows the signal to be routed to a prescaler circuit before being sent to the ADC. The prescaler can be configured to accept analog signals between -12 V and 0 or between 0 and $+12\text{ V}$. The prescaler circuit scales the voltage applied to the ADC input pad such that it is compatible with the ADC input voltage range. The AV pin can also be used as a digital input pin.

The second section of the Analog Quad is called the Current Monitor Block. Its input pin is named AC. The Current Monitor Block contains all the same functions as the Voltage Monitor Block with one addition, which is a current monitoring function. A small external current sensing resistor (typically less than $1\ \Omega$) is connected between the AV and AC pins and is in series with a power source. The Current Monitor Block contains a current monitor circuit that converts the current through the external resistor to a voltage that can then be read using the ADC.

Channel Input Offset Error

Channel Offset error is measured as the input voltage that causes the transition from zero to a count of one. An Ideal Prescaler will have offset equal to $\frac{1}{2}$ of LSB voltage. Offset error is a positive or negative when the first transition point is higher or lower than ideal. Offset error is expressed in LSB or input voltage.

Total Channel Error

Total Channel Error is defined as the total error measured compared to the ideal value. Total Channel Error is the sum of gain error and offset error combined. Figure 2-68 shows how Total Channel Error is measured.

Total Channel Error is defined as the difference between the actual ADC output and ideal ADC output. In the example shown in Figure 2-68, the Total Channel Error would be a negative number.

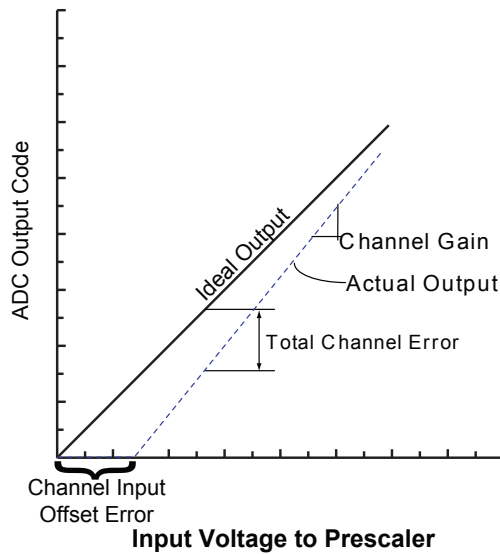


Figure 2-68 • Total Channel Error Example

Fusion uses a remote diode as a temperature sensor. The Fusion Temperature Monitor uses a differential input; the AT pin and ATRTN (AT Return) pin are the differential inputs to the Temperature Monitor. There is one Temperature Monitor in each Quad. A simplified block diagram is shown in [Figure 2-77](#).

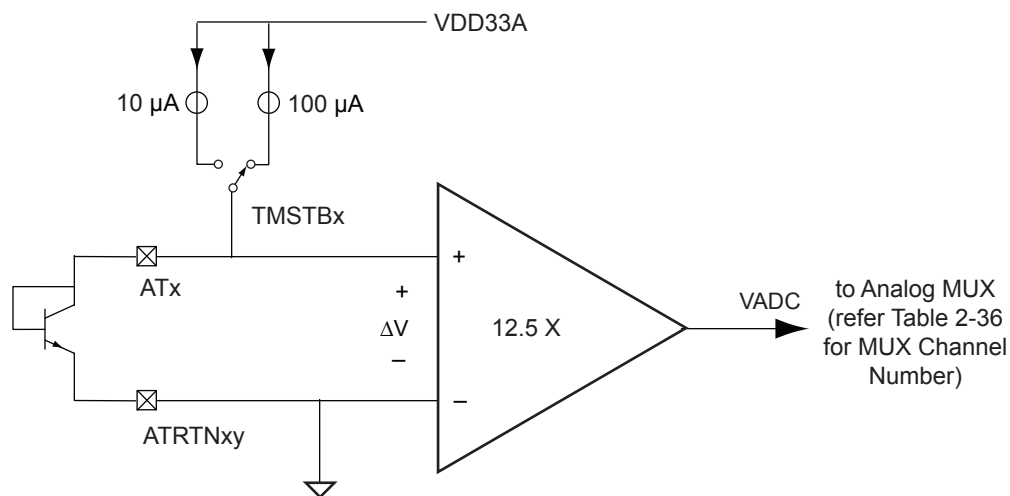


Figure 2-77 • Block Diagram for Temperature Monitor Circuit

The Fusion approach to measuring temperature is forcing two different currents through the diode with a ratio of 10:1. The switch that controls the different currents is controlled by the Temperature Monitor Strobe signal, TMSTB. Setting TMSTB to '1' will initiate a Temperature reading. The TMSTB should remain '1' until the ADC finishes sampling the voltage from the Temperature Monitor. The minimum sample time for the Temperature Monitor cannot be less than the minimum strobe high time minus the setup time. [Figure 2-78](#) shows the timing diagram.

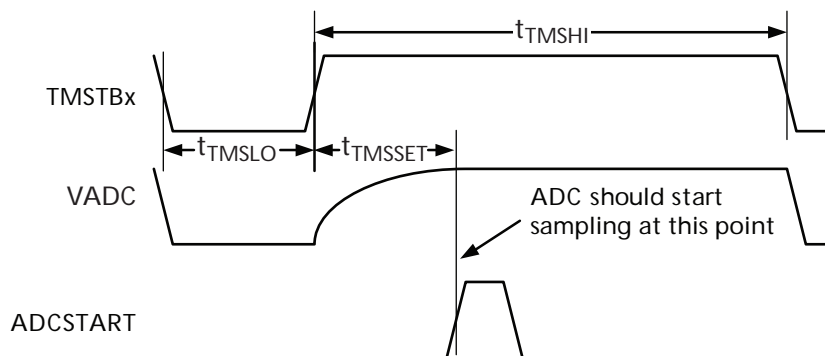


Figure 2-78 • Timing Diagram for the Temperature Monitor Strobe Signal

Note: When the IEEE 1149.1 Boundary Scan EXTEST instruction is executed, the AG pad drive strength ceases and becomes a 1 µA sink into the Fusion device.

This process results in a binary approximation of VIN. Generally, there is a fixed interval T, the sampling period, between the samples. The inverse of the sampling period is often referred to as the sampling frequency $f_s = 1 / T$. The combined effect is illustrated in Figure 2-82.

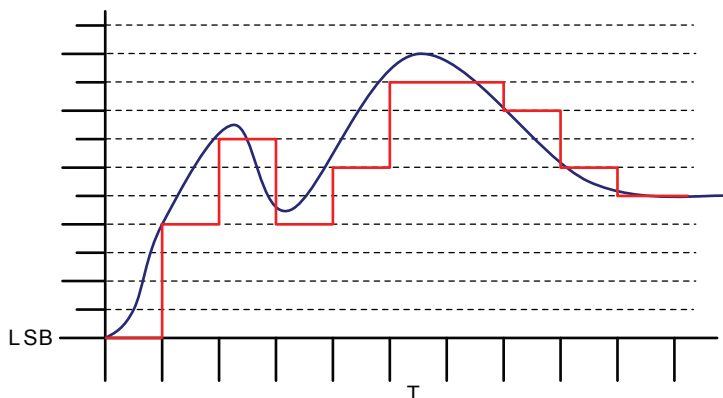


Figure 2-82 • Conversion Example

Figure 2-82 demonstrates that if the signal changes faster than the sampling rate can accommodate, or if the actual value of VIN falls between counts in the result, this information is lost during the conversion. There are several techniques that can be used to address these issues.

First, the sampling rate must be chosen to provide enough samples to adequately represent the input signal. Based on the Nyquist-Shannon Sampling Theorem, the minimum sampling rate must be at least twice the frequency of the highest frequency component in the target signal (Nyquist Frequency). For example, to recreate the frequency content of an audio signal with up to 22 KHz bandwidth, the user must sample it at a minimum of 44 ksp/s. However, as shown in Figure 2-82, significant post-processing of the data is required to interpolate the value of the waveform during the time between each sample.

Similarly, to re-create the amplitude variation of a signal, the signal must be sampled with adequate resolution. Continuing with the audio example, the dynamic range of the human ear (the ratio of the amplitude of the threshold of hearing to the threshold of pain) is generally accepted to be 135 dB, and the dynamic range of a typical symphony orchestra performance is around 85 dB. Most commercial recording media provide about 96 dB of dynamic range using 16-bit sample resolution. But 16-bit fidelity does not necessarily mean that you need a 16-bit ADC. As long as the input is sampled at or above the Nyquist Frequency, post-processing techniques can be used to interpolate intermediate values and reconstruct the original input signal to within desired tolerances.

If sophisticated digital signal processing (DSP) capabilities are available, the best results are obtained by implementing a reconstruction filter, which is used to interpolate many intermediate values with higher resolution than the original data. Interpolating many intermediate values increases the effective number of samples, and higher resolution increases the effective number of bits in the sample. In many cases, however, it is not cost-effective or necessary to implement such a sophisticated reconstruction algorithm. For applications that do not require extremely fine reproduction of the input signal, alternative methods can enhance digital sampling results with relatively simple post-processing. The details of such techniques are out of the scope of this chapter; refer to the [Improving ADC Results through Oversampling and Post-Processing of Data](#) white paper for more information.

Table 2-50 • ADC Characteristics in Direct Input Mode
Commercial Temperature Range Conditions, $T_J = 85^\circ\text{C}$ (unless noted otherwise),
Typical: $V_{CC33A} = 3.3\text{ V}$, $V_{CC} = 1.5\text{ V}$

Parameter	Description	Condition	Min.	Typ.	Max.	Units
Direct Input using Analog Pad AV, AC, AT						
VINADC	Input Voltage (Direct Input)	Refer to Table 3-2 on page 3-3				
CINADC	Input Capacitance	Channel not selected		7		pF
		Channel selected but not sampling		8		pF
		Channel selected and sampling		18		pF
ZINADC	Input Impedance	8-bit mode		2		kΩ
		10-bit mode		2		kΩ
		12-bit mode		2		kΩ
Analog Reference Voltage VAREF						
VAREF	Accuracy	T _J = 25°C	2.537	2.56	2.583	V
	Temperature Drift of Internal Reference			65		ppm / °C
	External Reference		2.527		VCC33A + 0.05	V
ADC Accuracy (using external reference) ^{1,2}						
DC Accuracy						
TUE	Total Unadjusted Error	8-bit mode	0.29			LSB
		10-bit mode	0.72			LSB
		12-bit mode	1.8			LSB
INL	Integral Non-Linearity	8-bit mode		0.20	0.25	LSB
		10-bit mode		0.32	0.43	LSB
		12-bit mode		1.71	1.80	LSB
DNL	Differential Non-Linearity (no missing code)	8-bit mode		0.20	0.24	LSB
		10-bit mode		0.60	0.65	LSB
		12-bit mode		2.40	2.48	LSB
	Offset Error	8-bit mode		0.01	0.17	LSB
		10-bit mode		0.05	0.20	LSB
		12-bit mode		0.20	0.40	LSB
	Gain Error	8-bit mode		0.0004	0.003	LSB
		10-bit mode		0.002	0.011	LSB
		12-bit mode		0.007	0.044	LSB
	Gain Error (with internal reference)	All modes		2		% FSR

Notes:

1. Accuracy of the external reference is $2.56\text{ V} \pm 4.6\text{ mV}$.
2. Data is based on characterization.
3. The sample rate is time-shared among active analog inputs.

Table 2-57 details the settings available to control the prescaler values of the AV, AC, and AT pins. Note that the AT pin has a reduced number of available prescaler values.

Table 2-57 • Prescaler Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3)

Control Lines Bx[2:0]	Scaling Factor, Pad to ADC Input	LSB for an 8-Bit Conversion ¹ (mV)	LSB for a 10-Bit Conversion ¹ (mV)	LSB for a 12-Bit Conversion ¹ (mV)	Full-Scale Voltage in 10-Bit Mode ²	Range Name
000 ³	0.15625	64	16	4	16.368 V	16 V
001	0.3125	32	8	2	8.184 V	8 V
010 ³	0.625	16	4	1	4.092 V	4 V
011	1.25	8	2	0.5	2.046 V	2 V
100	2.5	4	1	0.25	1.023 V	1 V
101	5.0	2	0.5	0.125	0.5115 V	0.5 V
110	10.0	1	0.25	0.0625	0.25575 V	0.25 V
111	20.0	0.5	0.125	0.03125	0.127875 V	0.125 V

Notes:

1. LSB voltage equivalences assume $V_{AREF} = 2.56$ V.
2. Full Scale voltage for n-bit mode: $((2^n) - 1) \times (\text{LSB for a n-bit Conversion})$
3. These are the only valid ranges for the Temperature Monitor Block Prescaler.

Table 2-58 details the settings available to control the MUX within each of the AV, AC, and AT circuits. This MUX determines whether the signal routed to the ADC is the direct analog input, prescaled signal, or output of either the Current Monitor Block or the Temperature Monitor Block.

Table 2-58 • Analog Multiplexer Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3)

Control Lines Bx[4]	Control Lines Bx[3]	ADC Connected To
0	0	Prescaler
0	1	Direct input
1	0	Current amplifier temperature monitor
1	1	Not valid

Table 2-59 details the settings available to control the Direct Analog Input switch for the AV, AC, and AT pins.

Table 2-59 • Direct Analog Input Switch Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3)

Control Lines Bx[5]	Direct Input Switch
0	Off
1	On

Table 2-60 details the settings available to control the polarity of the signals coming to the AV, AC, and AT pins. Note that the only valid setting for the AT pin is logic 0 to support positive voltages.

Table 2-60 • Voltage Polarity Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3)*

Control Lines Bx[6]	Input Signal Polarity
0	Positive
1	Negative

Note: *The B3[6] signal for the AT pad should be kept at logic 0 to accept only positive voltages.

Table 2-83 • Fusion Pro I/O Supported Standards and Corresponding VREF and VTT Voltages

I/O Standard	Input/Output Supply Voltage (VCCI_TYP)	Input Reference Voltage (VREF_TYP)	Board Termination Voltage (VTT_TYP)
LVTTTL/LVCMOS 3.3 V	3.30 V	–	–
LVCMOS 2.5 V	2.50 V	–	–
LVCMOS 2.5 V / 5.0 V Input	2.50 V	–	–
LVCMOS 1.8 V	1.80 V	–	–
LVCMOS 1.5 V	1.50 V	–	–
PCI 3.3 V	3.30 V	–	–
PCI-X 3.3 V	3.30 V	–	–
GTL+ 3.3 V	3.30 V	1.00 V	1.50 V
GTL+ 2.5 V	2.50 V	1.00 V	1.50 V
GTL 3.3 V	3.30 V	0.80 V	1.20 V
GTL 2.5 V	2.50 V	0.80 V	1.20 V
HSTL Class I	1.50 V	0.75 V	0.75 V
HSTL Class II	1.50 V	0.75 V	0.75 V
SSTL3 Class I	3.30 V	1.50 V	1.50 V
SSTL3 Class II	3.30 V	1.50 V	1.50 V
SSTL2 Class I	2.50 V	1.25 V	1.25 V
SSTL2 Class II	2.50 V	1.25 V	1.25 V
LVDS, BLVDS, M-LVDS	2.50 V	–	–
LVPECL	3.30 V	–	–

Table 2-93 • Summary of I/O Timing Characteristics – Software Default Settings
Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = \text{I/O Standard Dependent}$
Applicable to Advanced I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
3.3 V LVTTTL/ 3.3 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.90	0.32	2.69	2.11	2.40	2.68	4.36	3.78	ns
2.5 V LVCMOS	12 mA	High	35 pF	–	0.49	2.66	0.03	0.98	0.32	2.71	2.56	2.47	2.57	4.38	4.23	ns
1.8 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.91	0.32	2.69	2.27	2.76	3.05	4.36	3.94	ns
1.5 V LVCMOS	12 mA	High	35 pF	–	0.49	3.05	0.03	1.07	0.32	3.10	2.67	2.95	3.14	4.77	4.34	ns
3.3 V PCI	Per PCI spec	High	10 pF	25^2	0.49	2.00	0.03	0.65	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
3.3 V PCI-X	Per PCI-X spec	High	10 pF	25^2	0.49	2.00	0.03	0.62	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
LVDS	24 mA	High	–	–	0.49	1.37	0.03	1.20	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns
LVPECL	24 mA	High	–	–	0.49	1.34	0.03	1.05	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns

Notes:

1. For specific junction temperature and voltage-supply levels, refer to [Table 3-6 on page 3-7](#) for derating values.
2. Resistance is used to measure I/O propagation delays as defined in PCI specifications. See [Figure 2-123 on page 2-197](#) for connectivity. This resistor is not required during normal operation.

Table 2-94 • Summary of I/O Timing Characteristics – Software Default Settings
Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = \text{I/O Standard Dependent}$
Applicable to Standard I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	Units
3.3 V LVTTTL/ 3.3 V LVCMOS	8 mA	High	35 pF	–	0.49	3.29	0.03	0.75	0.32	3.36	2.80	1.79	2.01	ns
2.5 V LVCMOS	8 mA	High	35pF	–	0.49	3.56	0.03	0.96	0.32	3.40	3.56	1.78	1.91	ns
1.8 V LVCMOS	4 mA	High	35pF	–	0.49	4.74	0.03	0.90	0.32	4.02	4.74	1.80	1.85	ns
1.5 V LVCMOS	2 mA	High	35pF	–	0.49	5.71	0.03	1.06	0.32	4.71	5.71	1.83	1.83	ns

Note: For specific junction temperature and voltage-supply levels, refer to [Table 3-6 on page 3-7](#) for derating values.

2.5 V GTL+

Gunning Transceiver Logic Plus is a high-speed bus standard (JESD8-3). It provides a differential amplifier input buffer and an open-drain output buffer. The VCCI pin should be connected to 2.5 V.

Table 2-147 • Minimum and Maximum DC Input and Output Levels

2.5 V GTL+	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL ¹	IIH ²
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA ³	Max. mA ³	μA ⁴	μA ⁴
33 mA	−0.3	VREF − 0.1	VREF + 0.1	3.6	0.6	−	33	33	124	169	10	10

Notes:

1. IIL is the input leakage current per I/O pin over recommended operation conditions where $-0.3\text{ V} < V_{IN} < V_{IL}$.
2. IIH is the input leakage current per I/O pin over recommended operating conditions $V_{IH} < V_{IN} < V_{CCI}$. Input current is larger when operating outside recommended ranges.
3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.
4. Currents are measured at 85°C junction temperature.

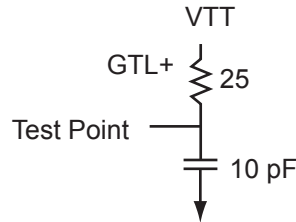


Figure 2-127 • AC Loading

Table 2-148 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C _{LOAD} (pF)
VREF − 0.1	VREF + 0.1	1.0	1.0	1.5	10

Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.

Timing Characteristics

Table 2-149 • 2.5 V GTL+

Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V, VREF = 1.0 V

Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{ZH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{ZHS}	Units
Std.	0.66	2.21	0.04	1.51	0.43	2.25	2.10			4.48	4.34	ns
−1	0.56	1.88	0.04	1.29	0.36	1.91	1.79			3.81	3.69	ns
−2	0.49	1.65	0.03	1.13	0.32	1.68	1.57			3.35	4.34	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

SSTL2 Class I

Stub-Speed Terminated Logic for 2.5 V memory bus standard (JESD8-9). Fusion devices support Class I. This provides a differential amplifier input buffer and a push-pull output buffer.

Table 2-156 • Minimum and Maximum DC Input and Output Levels

SSTL2 Class I	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL ¹	IIH ²
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA ³	Max. mA ³	μA ⁴	μA ⁴
15 mA	-0.3	VREF - 0.2	VREF + 0.2	3.6	0.54	VCCI - 0.62	15	15	87	83	10	10

Notes:

1. IIL is the input leakage current per I/O pin over recommended operation conditions where $-0.3\text{ V} < V_{IN} < V_{IL}$.
2. IIH is the input leakage current per I/O pin over recommended operating conditions $V_{IH} < V_{IN} < V_{CCI}$. Input current is larger when operating outside recommended ranges.
3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.
4. Currents are measured at 85°C junction temperature.

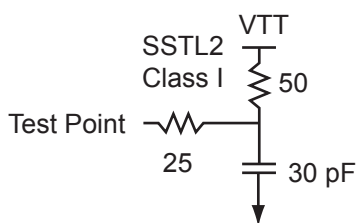


Figure 2-130 • AC Loading

Table 2-157 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C _{LOAD} (pF)
VREF - 0.2	VREF + 0.2	1.25	1.25	1.25	30

Note: *Measuring point = V_{trip} . See [Table 2-90 on page 2-166](#) for a complete table of trip points.

Timing Characteristics

Table 2-158 • SSTL 2 Class I

Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V, VREF = 1.25 V

Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{ZH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{ZHS}	Units
Std.	0.66	2.13	0.04	1.33	0.43	2.17	1.85			4.40	4.08	ns
-1	0.56	1.81	0.04	1.14	0.36	1.84	1.57			3.74	3.47	ns
-2	0.49	1.59	0.03	1.00	0.32	1.62	1.38			3.29	3.05	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

Output Register

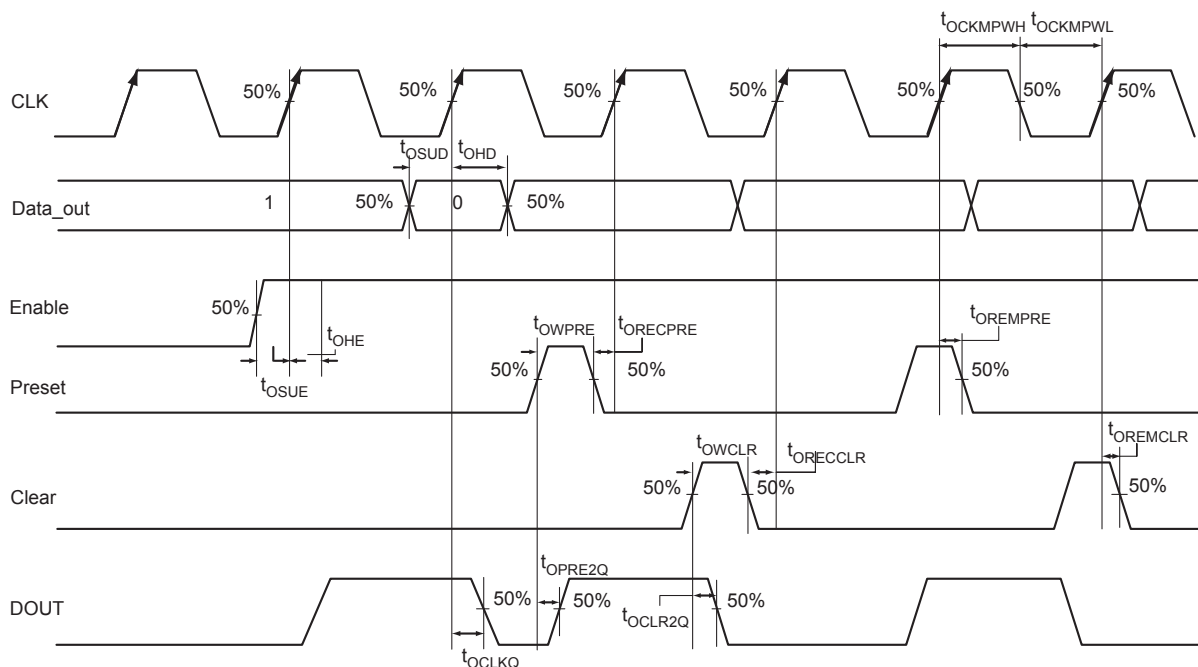


Figure 2-140 • Output Register Timing Diagram

Timing Characteristics

Table 2-177 • Output Data Register Propagation Delays
Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$

Parameter	Description	-2	-1	Std.	Units
t_{OCLKQ}	Clock-to-Q of the Output Data Register	0.59	0.67	0.79	ns
t_{OSUD}	Data Setup Time for the Output Data Register	0.31	0.36	0.42	ns
t_{OHD}	Data Hold Time for the Output Data Register	0.00	0.00	0.00	ns
t_{OSUE}	Enable Setup Time for the Output Data Register	0.44	0.50	0.59	ns
t_{OHE}	Enable Hold Time for the Output Data Register	0.00	0.00	0.00	ns
t_{OCLR2Q}	Asynchronous Clear-to-Q of the Output Data Register	0.80	0.91	1.07	ns
t_{OPRE2Q}	Asynchronous Preset-to-Q of the Output Data Register	0.80	0.91	1.07	ns
$t_{OREMCLR}$	Asynchronous Clear Removal Time for the Output Data Register	0.00	0.00	0.00	ns
$t_{ORECCLR}$	Asynchronous Clear Recovery Time for the Output Data Register	0.22	0.25	0.30	ns
$t_{OREMPRE}$	Asynchronous Preset Removal Time for the Output Data Register	0.00	0.00	0.00	ns
$t_{ORECPRE}$	Asynchronous Preset Recovery Time for the Output Data Register	0.22	0.25	0.30	ns
t_{OWCLR}	Asynchronous Clear Minimum Pulse Width for the Output Data Register	0.22	0.25	0.30	ns
t_{OWPRE}	Asynchronous Preset Minimum Pulse Width for the Output Data Register	0.22	0.25	0.30	ns
$t_{OCLKMPWH}$	Clock Minimum Pulse Width High for the Output Data Register	0.36	0.41	0.48	ns
$t_{OCLKMPWL}$	Clock Minimum Pulse Width Low for the Output Data Register	0.32	0.37	0.43	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

$$P_{S-CELL} = 0 \text{ W}$$

$$P_{C-CELL} = 0 \text{ W}$$

$$P_{NET} = 0 \text{ W}$$

$$P_{LOGIC} = 0 \text{ W}$$

I/O Input and Output Buffer Contribution— $P_{I/O}$

This example uses LVTTTL 3.3 V I/O cells. The output buffers are 12 mA-capable, configured with high output slew and driving a 35 pF output load.

$$F_{CLK} = 50 \text{ MHz}$$

$$\text{Number of input pins used: } N_{INPUTS} = 30$$

$$\text{Number of output pins used: } N_{OUTPUTS} = 40$$

$$\text{Estimated I/O buffer toggle rate: } \alpha_2 = 0.1 \text{ (10\%)}$$

$$\text{Estimated IO buffer enable rate: } \beta_1 = 1 \text{ (100\%)}$$

Operating Mode

$$P_{INPUTS} = N_{INPUTS} * (\alpha_2 / 2) * PAC9 * F_{CLK}$$

$$P_{INPUTS} = 30 * (0.1 / 2) * 0.01739 * 50$$

$$P_{INPUTS} = 1.30 \text{ mW}$$

$$P_{OUTPUTS} = N_{OUTPUTS} * (\alpha_2 / 2) * \beta_1 * PAC10 * F_{CLK}$$

$$P_{OUTPUTS} = 40 * (0.1 / 2) * 1 * 0.4747 * 50$$

$$P_{OUTPUTS} = 47.47 \text{ mW}$$

$$P_{I/O} = P_{INPUTS} + P_{OUTPUTS}$$

$$P_{I/O} = 1.30 \text{ mW} + 47.47 \text{ mW}$$

$$P_{I/O} = 48.77 \text{ mW}$$

Standby Mode and Sleep Mode

$$P_{INPUTS} = 0 \text{ W}$$

$$P_{OUTPUTS} = 0 \text{ W}$$

$$P_{I/O} = 0 \text{ W}$$

RAM Contribution— P_{MEMORY}

$$\text{Frequency of Read Clock: } F_{READ-CLOCK} = 10 \text{ MHz}$$

$$\text{Frequency of Write Clock: } F_{WRITE-CLOCK} = 10 \text{ MHz}$$

$$\text{Number of RAM blocks: } N_{BLOCKS} = 20$$

$$\text{Estimated RAM Read Enable Rate: } \beta_2 = 0.125 \text{ (12.5\%)}$$

$$\text{Estimated RAM Write Enable Rate: } \beta_3 = 0.125 \text{ (12.5\%)}$$

Operating Mode

$$P_{MEMORY} = (N_{BLOCKS} * PAC11 * \beta_2 * F_{READ-CLOCK}) + (N_{BLOCKS} * PAC12 * \beta_3 * F_{WRITE-CLOCK})$$

$$P_{MEMORY} = (20 * 0.025 * 0.125 * 10) + (20 * 0.030 * 0.125 * 10)$$

$$P_{MEMORY} = 1.38 \text{ mW}$$

Standby Mode and Sleep Mode

$$P_{MEMORY} = 0 \text{ W}$$

FG256				
Pin Number	AFS090 Function	AFS250 Function	AFS600 Function	AFS1500 Function
M15	TRST	TRST	TRST	TRST
M16	GND	GND	GND	GND
N1	GEB2/IO42PDB3V0	GEB2/IO59PDB3V0	GEB2/IO59PDB4V0	GEB2/IO86PDB4V0
N2	GEA2/IO42NDB3V0	IO59NDB3V0	IO59NDB4V0	IO86NDB4V0
N3	NC	GEA2/IO58PPB3V0	GEA2/IO58PPB4V0	GEA2/IO85PPB4V0
N4	VCC33PMP	VCC33PMP	VCC33PMP	VCC33PMP
N5	VCC15A	VCC15A	VCC15A	VCC15A
N6	NC	NC	AG0	AG0
N7	AC1	AC1	AC3	AC3
N8	AG3	AG3	AG5	AG5
N9	AV3	AV3	AV5	AV5
N10	AG4	AG4	AG6	AG6
N11	NC	NC	AC8	AC8
N12	GNDA	GNDA	GNDA	GNDA
N13	VCC33A	VCC33A	VCC33A	VCC33A
N14	VCCNVM	VCCNVM	VCCNVM	VCCNVM
N15	TCK	TCK	TCK	TCK
N16	TDI	TDI	TDI	TDI
P1	VCCNVM	VCCNVM	VCCNVM	VCCNVM
P2	GNDNVM	GNDNVM	GNDNVM	GNDNVM
P3	GNDA	GNDA	GNDA	GNDA
P4	NC	NC	AC0	AC0
P5	NC	NC	AG1	AG1
P6	NC	NC	AV1	AV1
P7	AG0	AG0	AG2	AG2
P8	AG2	AG2	AG4	AG4
P9	GNDA	GNDA	GNDA	GNDA
P10	NC	AC5	AC7	AC7
P11	NC	NC	AV8	AV8
P12	NC	NC	AG8	AG8
P13	NC	NC	AV9	AV9
P14	ADCGNDREF	ADCGNDREF	ADCGNDREF	ADCGNDREF
P15	PTBASE	PTBASE	PTBASE	PTBASE
P16	GNDNVM	GNDNVM	GNDNVM	GNDNVM
R1	VCCIB3	VCCIB3	VCCIB4	VCCIB4
R2	PCAP	PCAP	PCAP	PCAP
R3	NC	NC	AT1	AT1
R4	NC	NC	AT0	AT0

FG676	
Pin Number	AFS1500 Function
L17	VCCIB2
L18	GCB2/IO60PDB2V0
L19	IO58NDB2V0
L20	IO57NDB2V0
L21	IO59NDB2V0
L22	GCC2/IO61PDB2V0
L23	IO55PPB2V0
L24	IO56PDB2V0
L25	IO55NPB2V0
L26	GND
M1	NC
M2	VCCIB4
M3	GFC2/IO108PDB4V0
M4	GND
M5	IO109NDB4V0
M6	IO110NDB4V0
M7	GND
M8	IO104NDB4V0
M9	IO111NDB4V0
M10	GND
M11	VCC
M12	GND
M13	VCC
M14	GND
M15	VCC
M16	GND
M17	GND
M18	IO60NDB2V0
M19	IO58PDB2V0
M20	GND
M21	IO68NPB2V0
M22	IO61NDB2V0
M23	GND
M24	IO56NDB2V0
M25	VCCIB2
M26	IO65PDB2V0

FG676	
Pin Number	AFS1500 Function
N1	NC
N2	NC
N3	IO108NDB4V0
N4	VCCOSC
N5	VCCIB4
N6	XTAL2
N7	GFC1/IO107PDB4V0
N8	VCCIB4
N9	GFB1/IO106PDB4V0
N10	VCCIB4
N11	GND
N12	VCC
N13	GND
N14	VCC
N15	GND
N16	VCC
N17	VCCIB2
N18	IO70PDB2V0
N19	VCCIB2
N20	IO69PDB2V0
N21	GCA1/IO64PDB2V0
N22	VCCIB2
N23	GCC0/IO62NDB2V0
N24	GCC1/IO62PDB2V0
N25	IO66PDB2V0
N26	IO65NDB2V0
P1	NC
P2	NC
P3	IO103PDB4V0
P4	XTAL1
P5	VCCIB4
P6	GNDOSC
P7	GFC0/IO107NDB4V0
P8	VCCIB4
P9	GFB0/IO106NDB4V0
P10	VCCIB4

FG676	
Pin Number	AFS1500 Function
P11	VCC
P12	GND
P13	VCC
P14	GND
P15	VCC
P16	GND
P17	VCCIB2
P18	IO70NDB2V0
P19	VCCIB2
P20	IO69NDB2V0
P21	GCA0/IO64NDB2V0
P22	VCCIB2
P23	GCB0/IO63NDB2V0
P24	GCB1/IO63PDB2V0
P25	IO66NDB2V0
P26	IO67PDB2V0
R1	NC
R2	VCCIB4
R3	IO103NDB4V0
R4	GND
R5	IO101PDB4V0
R6	IO100NPB4V0
R7	GND
R8	IO99PDB4V0
R9	IO97PDB4V0
R10	GND
R11	GND
R12	VCC
R13	GND
R14	VCC
R15	GND
R16	VCC
R17	GND
R18	GDB2/IO83PDB2V0
R19	IO78PDB2V0
R20	GND

Revision	Changes	Page
Advance v0.8 (continued)	This sentence was updated in the "No-Glitch MUX (NGMUX)" section to delete GLA: The GLMUXCFG[1:0] configuration bits determine the source of the CLK inputs (i.e., internal signal or GLC).	2-32
	In Table 2-13 • NGMUX Configuration and Selection Table, 10 and 11 were deleted.	2-32
	The method to enable sleep mode was updated for bit 0 in Table 2-16 • RTC Control/Status Register.	2-38
	S2 was changed to D2 in Figure 2-39 • Read Waveform (Pipe Mode, 32-bit access) for RD[31:0] was updated.	2-51
	The definitions for bits 2 and 3 were updated in Table 2-24 • Page Status Bit Definition.	2-52
	Figure 2-46 • FlashROM Timing Diagram was updated.	2-58
	Table 2-26 • FlashROM Access Time is new.	2-58
	Figure 2-55 • Write Access After Write onto Same Address, Figure 2-56 • Read Access After Write onto Same Address, and Figure 2-57 • Write Access After Read onto Same Address are new.	2-68– 2-70
	Table 2-31 • RAM4K9 and Table 2-32 • RAM512X18 were updated.	2-71, 2-72
	The VAREF and SAMPLE functions were updated in Table 2-36 • Analog Block Pin Description.	2-82
	The title of Figure 2-72 • Timing Diagram for Current Monitor Strobe was updated to add the word "positive."	2-91
	The "Gate Driver" section was updated to give information about the switching rate in High Current Drive mode.	2-94
	The "ADC Description" section was updated to include information about the SAMPLE and BUSY signals and the maximum frequencies for SYSCLK and ADCCLK. EQ 2 was updated to add parentheses around the entire expression in the denominator.	2-102
	Table 2-46 • Analog Channel Specifications and Table 2-47 • ADC Characteristics in Direct Input Mode were updated.	2-118, 2-121
	The note was removed from Table 2-55 • Analog Multiplexer Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3).	2-131
	Table 2-63 • Internal Temperature Monitor Control Truth Table is new.	2-132
	The "Cold-Sparing Support" section was updated to add information about cases where current draw can occur.	2-143
	Figure 2-104 • Solution 4 was updated.	2-147
	Table 2-75 • Fusion Standard I/O Standards—OUT_DRIVE Settings was updated.	2-153
	The "GNDA Ground (analog)" section and "GNDAQ Ground (analog quiet)" section were updated to add information about maximum differential voltage.	2-224
	The "VAREF Analog Reference Voltage" section and "VPUMP Programming Supply Voltage" section were updated.	2-226
	The "V _{CCPLA/B} PLL Supply Voltage" section was updated to include information about the east and west PLLs.	2-225
	The V _{COMPLF} pin description was deleted.	N/A
	The "Axy Analog Input/Output" section was updated with information about grounding and floating the pin.	2-226

Revision	Changes	Page
Advance v0.5 (June 2006)	The low power modes of operation were updated and clarified.	N/A
	The AFS1500 digital I/O count was updated in Table 1 • Fusion Family .	i
	The AFS1500 digital I/O count was updated in the "Package I/Os: Single-/Double-Ended (Analog)" table.	ii
	The "Voltage Regulator Power Supply Monitor (VRPSM)" was updated.	2-36
	Figure 2-45 • FlashROM Timing Diagram was updated.	2-53
	The "256-Pin FBGA" table for the AFS1500 is new.	3-12
Advance v0.4 (April 2006)	The G was moved in the "Product Ordering Codes" section.	III
Advance v0.3 (April 2006)	The "Features and Benefits" section was updated.	I
	The "Fusion Family" table was updated.	I
	The "Package I/Os: Single-/Double-Ended (Analog)" table was updated.	II
	The "Product Ordering Codes" table was updated.	III
	The "Temperature Grade Offerings" table was updated.	IV
	The "General Description" section was updated to include ARM information.	1-1
	Figure 2-46 • FlashROM Timing Diagram was updated.	2-58
	The "FlashROM" section was updated.	2-57
	The "RESET" section was updated.	2-61
	The "RESET" section was updated.	2-64
	Figure 2-27 • Real-Time Counter System was updated.	2-35
	Table 2-19 • Flash Memory Block Pin Names was updated.	2-43
	Figure 2-33 • Flash Memory Block Diagram was updated to include AUX block information.	2-45
	Figure 2-34 • Flash Memory Block Organization was updated to include AUX block information.	2-46
	The note in the "Program Operation" section was updated.	2-48
	Figure 2-76 • Gate Driver Example was updated.	2-95
	The "Analog Quad ACM Description" section was updated.	2-130
	Information about the maximum pad input frequency was added to the "Gate Driver" section.	2-94
	Figure 2-65 • Analog Block Macro was updated.	2-81
	Figure 2-65 • Analog Block Macro was updated.	2-81
	The "Analog Quad" section was updated.	2-84
	The "Voltage Monitor" section was updated.	2-86
	The "Direct Digital Input" section was updated.	2-89
	The "Current Monitor" section was updated.	2-90
	Information about the maximum pad input frequency was added to the "Gate Driver" section.	2-94