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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	93
Number of Gates	250000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/m1afs250-2pqg208

VersaTile Characteristics

Sample VersaTile Specifications—Combinatorial Module

The Fusion library offers all combinations of LUT-3 combinatorial functions. In this section, timing characteristics are presented for a sample of the library (Figure 2-3). For more details, refer to the *IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide*.

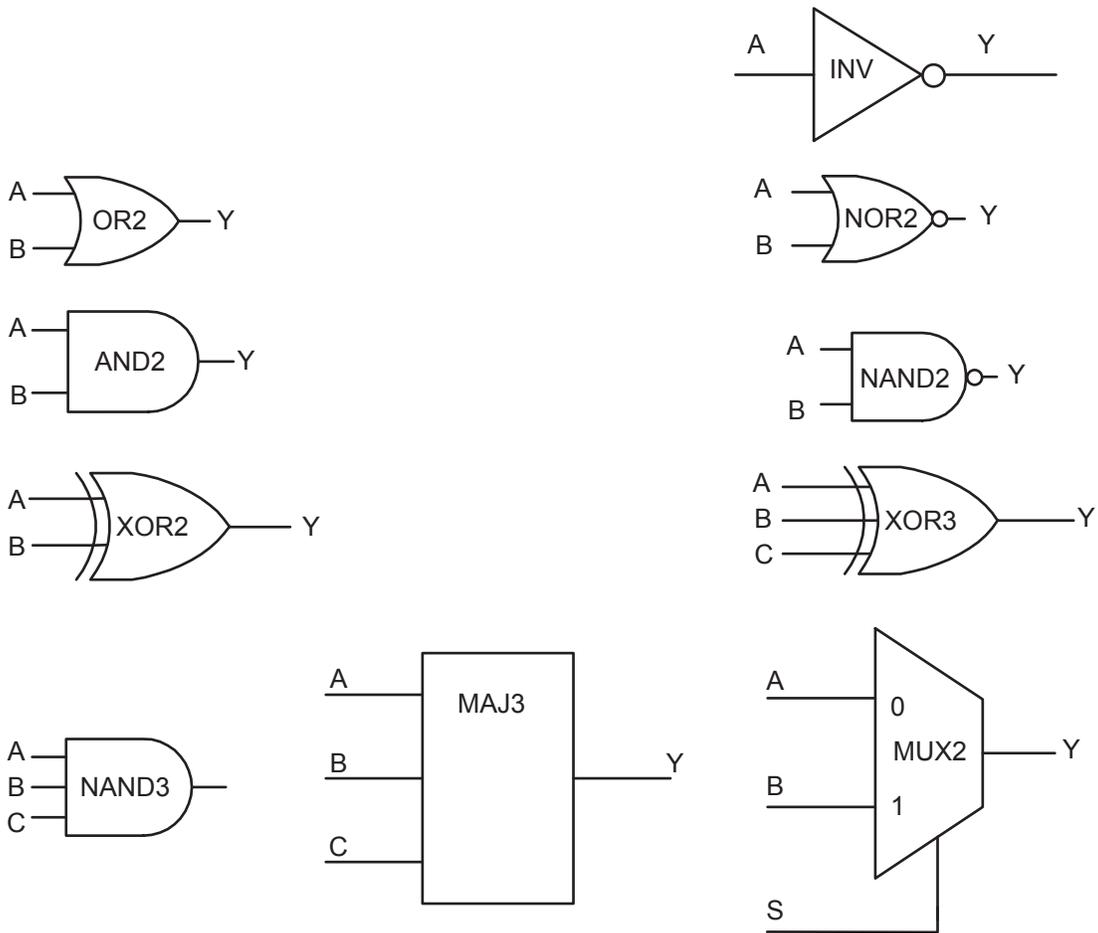


Figure 2-3 • Sample of Combinatorial Cells

VersaNet Global Networks and Spine Access

The Fusion architecture contains a total of 18 segmented global networks that can access the VersaTiles, SRAM, and I/O tiles on the Fusion device. There are 6 chip (main) global networks that access the entire device and 12 quadrant networks (3 in each quadrant). Each device has a total of 18 globals. These VersaNet global networks offer fast, low-skew routing resources for high-fanout nets, including clock signals. In addition, these highly segmented global networks offer users the flexibility to create low-skew local networks using spines for up to 180 internal/external clocks (in an AFS1500 device) or other high-fanout nets in Fusion devices. Optimal usage of these low-skew networks can result in significant improvement in design performance on Fusion devices.

The nine spines available in a vertical column reside in global networks with two separate regions of scope: the quadrant global network, which has three spines, and the chip (main) global network, which has six spines. Note that there are three quadrant spines in each quadrant of the device. There are four quadrant global network regions per device (Figure 2-12 on page 2-12).

The spines are the vertical branches of the global network tree, shown in Figure 2-11 on page 2-11. Each spine in a vertical column of a chip (main) global network is further divided into two equal-length spine segments: one in the top and one in the bottom half of the die.

Each spine and its associated ribs cover a certain area of the Fusion device (the "scope" of the spine; see Figure 2-11 on page 2-11). Each spine is accessed by the dedicated global network MUX tree architecture, which defines how a particular spine is driven—either by the signal on the global network from a CCC, for example, or another net defined by the user (Figure 2-13). Quadrant spines can be driven from user I/Os on the north and south sides of the die, via analog I/Os configured as direct digital inputs. The ability to drive spines in the quadrant global networks can have a significant effect on system performance for high-fanout inputs to a design.

Details of the chip (main) global network spine-selection MUX are presented in Figure 2-13. The spine drivers for each spine are located in the middle of the die.

Quadrant spines are driven from a north or south rib. Access to the top and bottom ribs is from the corner CCC or from the I/Os on the north and south sides of the device. For details on using spines in Fusion devices, see the application note [Using Global Resources in Actel Fusion Devices](#).

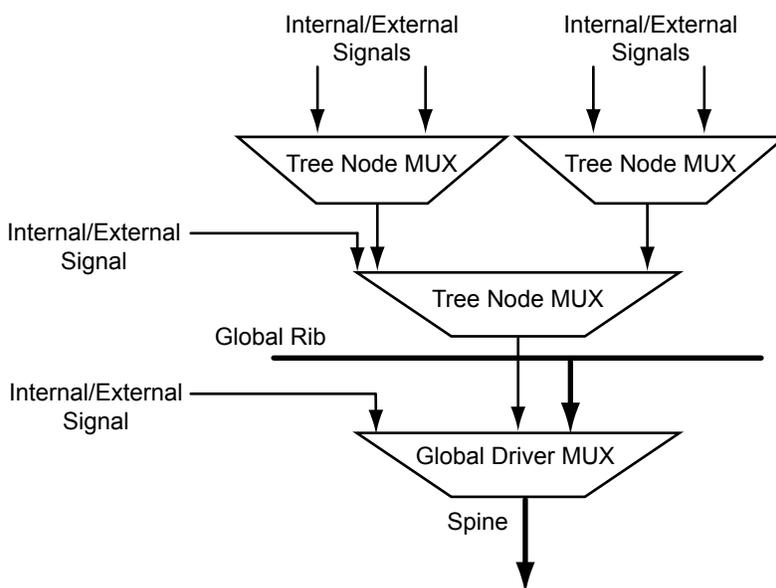


Figure 2-13 • Spine-Selection MUX of Global Tree

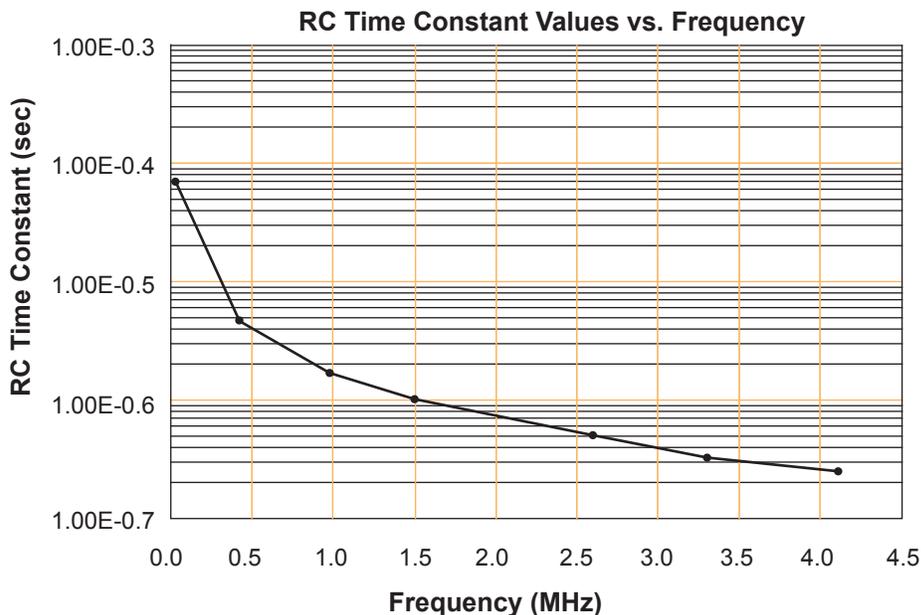


Figure 2-18 • Crystal Oscillator: RC Time Constant Values vs. Frequency (typical)

Table 2-10 • XTLOSC Signals Descriptions

Signal Name	Width	Direction	Function		
XTL_EN*	1		Enables the crystal. Active high.		
XTL_MODE*	2		Settings for the crystal clock for different frequency.		
			Value	Modes	Frequency Range
			b'00	RC network	32 KHz to 4 MHz
			b'01	Low gain	32 to 200 KHz
			b'10	Medium gain	0.20 to 2.0 MHz
			b'11	High gain	2.0 to 20.0 MHz
SELMODE	1	IN	Selects the source of XTL_MODE and also enables the XTL_EN. Connect from RTCXTLSEL from AB.		
			0	For normal operation or sleep mode, XTL_EN depends on FPGA_EN, XTL_MODE depends on MODE	
			1	For Standby mode, XTL_EN is enabled, XTL_MODE depends on RTC_MODE	
RTC_MODE[1:0]	2	IN	Settings for the crystal clock for different frequency ranges. XTL_MODE uses RTC_MODE when SELMODE is '1'.		
MODE[1:0]	2	IN	Settings for the crystal clock for different frequency ranges. XTL_MODE uses MODE when SELMODE is '0'. In Standby, MODE inputs will be 0's.		
FPGA_EN*	1	IN	0 when 1.5 V is not present for VCC 1 when 1.5 V is present for VCC		
XTL	1	IN	Crystal Clock source		
CLKOUT	1	OUT	Crystal Clock output		

Note: *Internal signal—does not exist in macro.

RAM4K9 Description

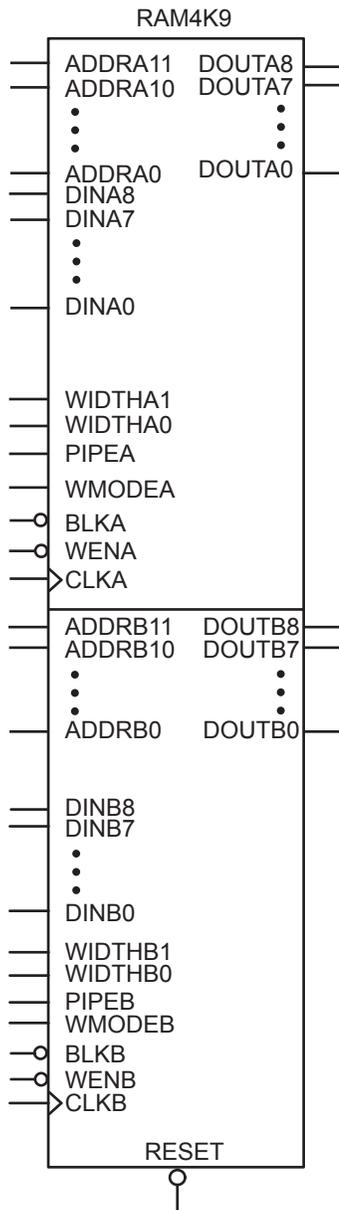


Figure 2-48 • RAM4K9

The following signals are used to configure the FIFO4K18 memory element.

WW and RW

These signals enable the FIFO to be configured in one of the five allowable aspect ratios (Table 2-33).

Table 2-33 • Aspect Ratio Settings for WW[2:0]

WW2, WW1, WW0	RW2, RW1, RW0	D×W
000	000	4k×1
001	001	2k×2
010	010	1k×4
011	011	512×9
100	100	256×18
101, 110, 111	101, 110, 111	Reserved

WBLK and RBLK

These signals are active low and will enable the respective ports when Low. When the RBLK signal is High, the corresponding port's outputs hold the previous value.

WEN and REN

Read and write enables. WEN is active low and REN is active high by default. These signals can be configured as active high or low.

WCLK and RCLK

These are the clock signals for the synchronous read and write operations. These can be driven independently or with the same driver.

RPIPE

This signal is used to specify pipelined read on the output. A Low on RPIPE indicates a nonpipelined read, and the data appears on the output in the same clock cycle. A High indicates a pipelined read, and data appears on the output in the next clock cycle.

RESET

This active low signal resets the output to zero when asserted. It resets the FIFO counters. It also sets all the RD pins Low, the FULL and AFULL pins Low, and the EMPTY and AEMPTY pins High (Table 2-34).

Table 2-34 • Input Data Signal Usage for Different Aspect Ratios

D×W	WD/RD Unused
4k×1	WD[17:1], RD[17:1]
2k×2	WD[17:2], RD[17:2]
1k×4	WD[17:4], RD[17:4]
512×9	WD[17:9], RD[17:9]
256×18	—

WD

This is the input data bus and is 18 bits wide. Not all 18 bits are valid in all configurations. When a data width less than 18 is specified, unused higher-order signals must be grounded (Table 2-34).

RD

This is the output data bus and is 18 bits wide. Not all 18 bits are valid in all configurations. Like the WD bus, high-order bits become unusable if the data width is less than 18. The output data on unused pins is undefined (Table 2-34).

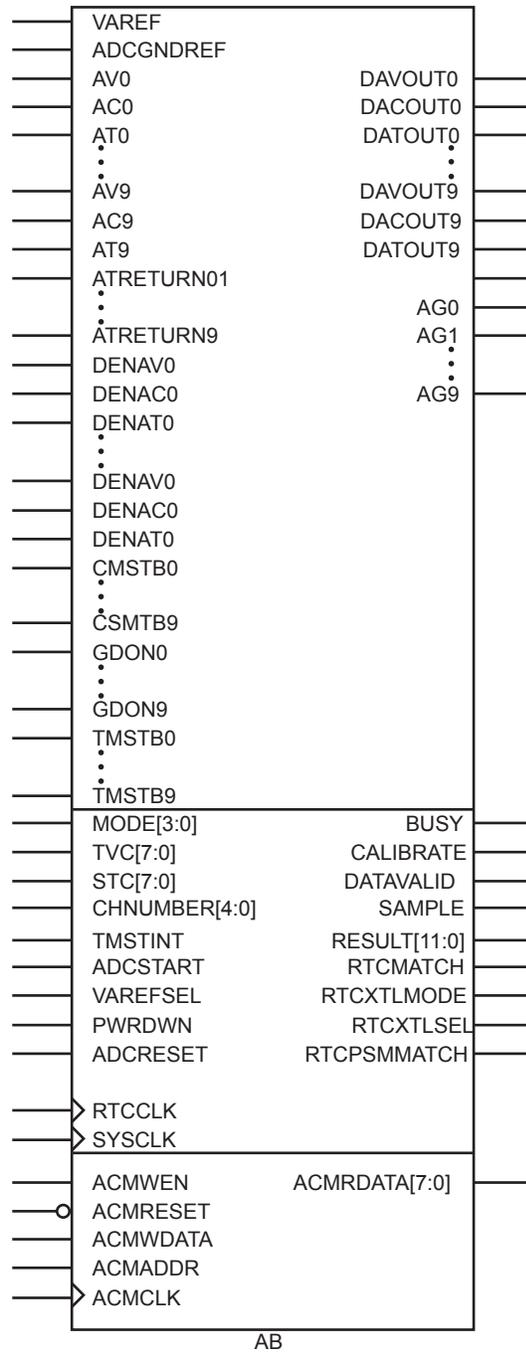


Figure 2-64 • Analog Block Macro

Terminology

Resolution

Resolution defines the smallest temperature change Fusion Temperature Monitor can resolve. For ADC configured as 8-bit mode, each LSB represents 4°C, and 1°C per LSB for 10-bit mode. With 12-bit mode, the Temperature Monitor can still only resolve 1°C due to Temperature Monitor design.

Offset

The Fusion Temperature Monitor has a systematic offset ([Table 2-49 on page 2-117](#)), excluding error due to board resistance and ideality factor of the external diode. Microsemi provides an IP block (CalibIP) that is required in order to mitigate the systematic temperature offset. For further details on CalibIP, refer to the “*Temperature, Voltage, and Current Calibration in Fusion FPGAs*” chapter of the *Fusion FPGA Fabric User Guide*.

Table 2-49 • Analog Channel Specifications (continued)
 Commercial Temperature Range Conditions, $T_J = 85^\circ\text{C}$ (unless noted otherwise),
 Typical: $V_{CC33A} = 3.3\text{ V}$, $V_{CC} = 1.5\text{ V}$

Parameter	Description	Condition	Min.	Typ.	Max.	Units
Temperature Monitor Using Analog Pad AT						
External Temperature Monitor (external diode 2N3904, $T_J = 25^\circ\text{C}$) ⁴	Resolution	8-bit ADC		4		$^\circ\text{C}$
		10-bit ADC		1		$^\circ\text{C}$
		12-bit ADC		0.25		$^\circ\text{C}$
	Systematic Offset ⁵	AFS090, AFS250, AFS600, AFS1500, uncalibrated ⁷		5		$^\circ\text{C}$
		AFS090, AFS250, AFS600, AFS1500, calibrated ⁷		± 5		$^\circ\text{C}$
	Accuracy			± 3	± 5	$^\circ\text{C}$
	External Sensor Source Current	High level, TMSTBx = 0			10	
Low level, TMSTBx = 1				100		μA
Max Capacitance on AT pad					1.3	nF
Internal Temperature Monitor	Resolution	8-bit ADC	4			$^\circ\text{C}$
		10-bit ADC	1			$^\circ\text{C}$
		12-bit ADC	0.25			$^\circ\text{C}$
	Systematic Offset ⁵	AFS090 ⁷		5		$^\circ\text{C}$
		AFS250, AFS600, AFS1500 ⁷		11		$^\circ\text{C}$
Accuracy			± 3	± 5	$^\circ\text{C}$	
t_{TMSHI}	Strobe High time		10		105	μs
t_{TMSLO}	Strobe Low time		5			μs
t_{TMSSET}	Settling time		5			μs

Notes:

1. V_{RSM} is the maximum voltage drop across the current sense resistor.
2. Analog inputs used as digital inputs can tolerate the same voltage limits as the corresponding analog pad. There is no reliability concern on digital inputs as long as V_{IND} does not exceed these limits.
3. V_{IND} is limited to $V_{\text{CC33A}} + 0.2$ to allow reaching 10 MHz input frequency.
4. An averaging of 1,024 samples (LPF setting in Analog System Builder) is required and the maximum capacitance allowed across the AT pins is 500 pF.
5. The temperature offset is a fixed positive value.
6. The high current mode has a maximum power limit of 20 mW. Appropriate current limit resistors must be used, based on voltage on the pad.
7. When using SmartGen Analog System Builder, CalibIP is required to obtain specified offset. For further details on CalibIP, refer to the "Temperature, Voltage, and Current Calibration in Fusion FPGAs" chapter of the [Fusion FPGA Fabric User Guide](#).

**Table 2-73 • Maximum I/O Frequency for Single-Ended, Voltage-Referenced, and Differential I/Os;
All I/O Bank Types (maximum drive strength and high slew selected)**

Specification	Performance Up To
LVTTTL/LVCMOS 3.3 V	200 MHz
LVCMOS 2.5 V	250 MHz
LVCMOS 1.8 V	200 MHz
LVCMOS 1.5 V	130 MHz
PCI	200 MHz
PCI-X	200 MHz
HSTL-I	300 MHz
HSTL-II	300 MHz
SSTL2-I	300 MHz
SSTL2-II	300 MHz
SSTL3-I	300 MHz
SSTL3-II	300 MHz
GTL+ 3.3 V	300 MHz
GTL+ 2.5 V	300 MHz
GTL 3.3 V	300 MHz
GTL 2.5 V	300 MHz
LVDS	350 MHz
LVPECL	300 MHz

Table 2-78 • Fusion Standard I/O Standards—OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)					
	2	4	6	8	Slew	
LVTTTL/LVCMOS 3.3 V	3	3	3	3	High	Low
LVCMOS 2.5 V	3	3	3	3	High	Low
LVCMOS 1.8 V	3	3	–	–	High	Low
LVCMOS 1.5 V	3	–	–	–	High	Low

Table 2-79 • Fusion Advanced I/O Standards—SLEW and OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)							Slew	
	2	4	6	8	12	16	High	Low	
LVTTTL/LVCMOS 3.3 V	3	3	3	3	3	3	High	Low	
LVCMOS 2.5 V	3	3	3	3	3	–	High	Low	
LVCMOS 1.8 V	3	3	3	3	–	–	High	Low	
LVCMOS 1.5 V	3	3	–	–	–	–	High	Low	

Table 2-80 • Fusion Pro I/O Standards—SLEW and OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)							Slew	
	2	4	6	8	12	16	24	High	Low
LVTTTL/LVCMOS 3.3 V	3	3	3	3	3	3	3	High	Low
LVCMOS 2.5 V	3	3	3	3	3	3	3	High	Low
LVCMOS 2.5 V/5.0 V	3	3	3	3	3	3	3	High	Low
LVCMOS 1.8 V	3	3	3	3	3	3	–	High	Low
LVCMOS 1.5 V	3	3	3	3	3	–	–	High	Low

User I/O Characteristics

Timing Model

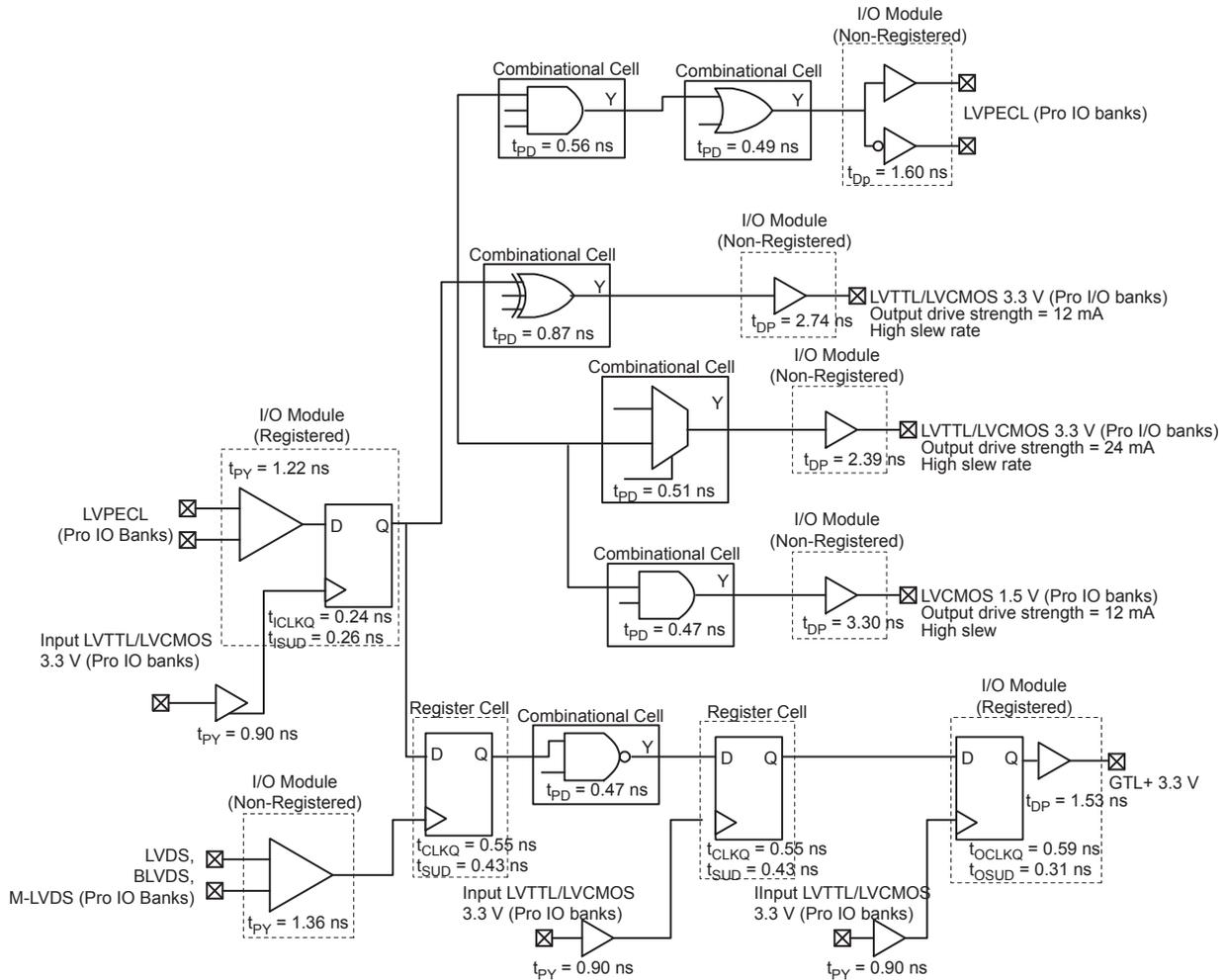


Figure 2-115 • Timing Model
Operating Conditions: -2 Speed, Commercial Temperature Range ($T_J = 70^\circ\text{C}$),
Worst-Case VCC = 1.425 V

Table 2-92 • Summary of I/O Timing Characteristics – Software Default Settings
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = I/O Standard Dependent
Applicable to Pro I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{PYS}	t _{EOUT}	t _{ZL}	t _{ZH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{ZHS}	Units
3.3 V LVTTTL/ 3.3 V LVCMOS	12 mA	High	35	–	0.49	2.74	0.03	0.90	1.17	0.32	2.79	2.14	2.45	2.70	4.46	3.81	ns
2.5 V LVCMOS	12 mA	High	35	–	0.49	2.80	0.03	1.13	1.24	0.32	2.85	2.61	2.51	2.61	4.52	4.28	ns
1.8 V LVCMOS	12 mA	High	35	–	0.49	2.83	0.03	1.08	1.42	0.32	2.89	2.31	2.79	3.16	4.56	3.98	ns
1.5 V LVCMOS	12 mA	High	35	–	0.49	3.30	0.03	1.27	1.60	0.32	3.36	2.70	2.96	3.27	5.03	4.37	ns
3.3 V PCI	Per PCI spec	High	10	25 ²	0.49	2.09	0.03	0.78	1.25	0.32	2.13	1.49	2.45	2.70	3.80	3.16	ns
3.3 V PCI-X	Per PCI-X spec	High	10	25 ²	0.49	2.09	0.03	0.77	1.17	0.32	2.13	1.49	2.45	2.70	3.80	3.16	ns
3.3 V GTL	20 mA	High	10	25	0.49	1.55	0.03	2.19	–	0.32	1.52	1.55	0.00	0.00	3.19	3.22	ns
2.5 V GTL	20 mA	High	10	25	0.49	1.59	0.03	1.83	–	0.32	1.61	1.59	0.00	0.00	3.28	3.26	ns
3.3 V GTL+	35 mA	High	10	25	0.49	1.53	0.03	1.19	–	0.32	1.56	1.53	0.00	0.00	3.23	3.20	ns
2.5 V GTL+	33 mA	High	10	25	0.49	1.65	0.03	1.13	–	0.32	1.68	1.57	0.00	0.00	3.35	3.24	ns
HSTL (I)	8 mA	High	20	50	0.49	2.37	0.03	1.59	–	0.32	2.42	2.35	0.00	0.00	4.09	4.02	ns
HSTL (II)	15 mA	High	20	25	0.49	2.26	0.03	1.59	–	0.32	2.30	2.03	0.00	0.00	3.97	3.70	ns
SSTL2 (I)	17 mA	High	30	50	0.49	1.59	0.03	1.00	–	0.32	1.62	1.38	0.00	0.00	3.29	3.05	ns
SSTL2 (II)	21 mA	High	30	25	0.49	1.62	0.03	1.00	–	0.32	1.65	1.32	0.00	0.00	3.32	2.99	ns
SSTL3 (I)	16 mA	High	30	50	0.49	1.72	0.03	0.93	–	0.32	1.75	1.37	0.00	0.00	3.42	3.04	ns
SSTL3 (II)	24 mA	High	30	25	0.49	1.54	0.03	0.93	–	0.32	1.57	1.25	0.00	0.00	3.24	2.92	ns
LVDS	24 mA	High	–	–	0.49	1.57	0.03	1.36	–	–	–	–	–	–	–	–	ns
LVPECL	24 mA	High	–	–	0.49	1.60	0.03	1.22	–	–	–	–	–	–	–	–	ns

Notes:

1. For specific junction temperature and voltage-supply levels, refer to [Table 3-6 on page 3-7](#) for derating values.
2. Resistance is used to measure I/O propagation delays as defined in PCI specifications. See [Figure 2-123 on page 2-197](#) for connectivity. This resistor is not required during normal operation.

SSTL2 Class I

Stub-Speed Terminated Logic for 2.5 V memory bus standard (JESD8-9). Fusion devices support Class I. This provides a differential amplifier input buffer and a push-pull output buffer.

Table 2-156 • Minimum and Maximum DC Input and Output Levels

SSTL2 Class I	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL ¹	IIH ²
	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA ³	Max. mA ³	μA ⁴	μA ⁴
15 mA	-0.3	VREF - 0.2	VREF + 0.2	3.6	0.54	VCCI - 0.62	15	15	87	83	10	10

Notes:

1. IIL is the input leakage current per I/O pin over recommended operation conditions where $-0.3\text{ V} < V_{IN} < V_{IL}$.
2. IIH is the input leakage current per I/O pin over recommended operating conditions $V_{IH} < V_{IN} < V_{CCI}$. Input current is larger when operating outside recommended ranges.
3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.
4. Currents are measured at 85°C junction temperature.

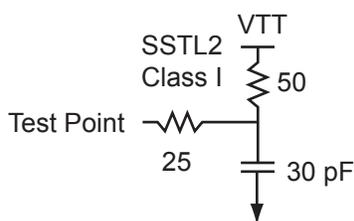


Figure 2-130 • AC Loading

Table 2-157 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C _{LOAD} (pF)
VREF - 0.2	VREF + 0.2	1.25	1.25	1.25	30

Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.

Timing Characteristics

Table 2-158 • SSTL 2 Class I

Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V, VREF = 1.25 V

Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{ZH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{ZHS}	Units
Std.	0.66	2.13	0.04	1.33	0.43	2.17	1.85			4.40	4.08	ns
-1	0.56	1.81	0.04	1.14	0.36	1.84	1.57			3.74	3.47	ns
-2	0.49	1.59	0.03	1.00	0.32	1.62	1.38			3.29	3.05	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Pin Descriptions

Supply Pins

GND **Ground**

Ground supply voltage to the core, I/O outputs, and I/O logic.

GNDQ **Ground (quiet)**

Quiet ground supply voltage to input buffers of I/O banks. Within the package, the GNDQ plane is decoupled from the simultaneous switching noise originated from the output buffer ground domain. This minimizes the noise transfer within the package and improves input signal integrity. GNDQ needs to always be connected on the board to GND. Note: In FG256, FG484, and FG676 packages, GNDQ and GND pins are connected within the package and are labeled as GND pins in the respective package pin assignment tables.

ADCGNDREF **Analog Reference Ground**

Analog ground reference used by the ADC. This pad should be connected to a quiet analog ground.

GND A **Ground (analog)**

Quiet ground supply voltage to the Analog Block of Fusion devices. The use of a separate analog ground helps isolate the analog functionality of the Fusion device from any digital switching noise. A 0.2 V maximum differential voltage between GND and GND A/GNDQ should apply to system implementation.

GND AQ **Ground (analog quiet)**

Quiet ground supply voltage to the analog I/O of Fusion devices. The use of a separate analog ground helps isolate the analog functionality of the Fusion device from any digital switching noise. A 0.2 V maximum differential voltage between GND and GND A/GNDQ should apply to system implementation. Note: In FG256, FG484, and FG676 packages, GND AQ and GND A pins are connected within the package and are labeled as GND A pins in the respective package pin assignment tables.

GND NVM **Flash Memory Ground**

Ground supply used by the Fusion device's flash memory block module(s).

GND OSC **Oscillator Ground**

Ground supply for both integrated RC oscillator and crystal oscillator circuit.

VCC15 A **Analog Power Supply (1.5 V)**

1.5 V clean analog power supply input for use by the 1.5 V portion of the analog circuitry.

VCC33 A **Analog Power Supply (3.3 V)**

3.3 V clean analog power supply input for use by the 3.3 V portion of the analog circuitry.

VCC33 N **Negative 3.3 V Output**

This is the -3.3 V output from the voltage converter. A 2.2 μ F capacitor must be connected from this pin to ground.

VCC33 PMP **Analog Power Supply (3.3 V)**

3.3 V clean analog power supply input for use by the analog charge pump. To avoid high current draw, VCC33 PMP should be powered up simultaneously with or after VCC33 A.

VCC NVM **Flash Memory Block Power Supply (1.5 V)**

1.5 V power supply input used by the Fusion device's flash memory block module(s). To avoid high current draw, VCC should be powered up before or simultaneously with VCC NVM.

VCC OSC **Oscillator Power Supply (3.3 V)**

Power supply for both integrated RC oscillator and crystal oscillator circuit. The internal 100 MHz oscillator, powered by the VCC OSC pin, is needed for device programming, operation of the VDD N33 pump, and eNVM operation. VCC OSC is off only when VCCA is off. VCC OSC must be powered whenever the Fusion device needs to function.

Table 3-5 • FPGA Programming, Storage, and Operating Limits

Product Grade	Storage Temperature	Element	Grade Programming Cycles	Retention
Commercial	Min. $T_J = 0^\circ\text{C}$ Max. $T_J = 85^\circ\text{C}$	FPGA/FlashROM	500	20 years
		Embedded Flash	< 1,000	20 years
			< 10,000	10 years
			< 15,000	5 years
Industrial	Min. $T_J = -40^\circ\text{C}$ Max. $T_J = 100^\circ\text{C}$	FPGA/FlashROM	500	20 years
		Embedded Flash	< 1,000	20 years
			< 10,000	10 years
			< 15,000	5 years

I/O Power-Up and Supply Voltage Thresholds for Power-On Reset (Commercial and Industrial)

Sophisticated power-up management circuitry is designed into every Fusion device. These circuits ensure easy transition from the powered off state to the powered up state of the device. The many different supplies can power up in any sequence with minimized current spikes or surges. In addition, the I/O will be in a known state through the power-up sequence. The basic principle is shown in [Figure 3-1 on page 3-6](#).

There are five regions to consider during power-up.

Fusion I/Os are activated only if ALL of the following three conditions are met:

1. VCC and VCCI are above the minimum specified trip points ([Figure 3-1](#)).
2. $V_{CCI} > V_{CC} - 0.75\text{ V}$ (typical).
3. Chip is in the operating mode.

VCCI Trip Point:

Ramping up: $0.6\text{ V} < \text{trip_point_up} < 1.2\text{ V}$

Ramping down: $0.5\text{ V} < \text{trip_point_down} < 1.1\text{ V}$

VCC Trip Point:

Ramping up: $0.6\text{ V} < \text{trip_point_up} < 1.1\text{ V}$

Ramping down: $0.5\text{ V} < \text{trip_point_down} < 1\text{ V}$

VCC and VCCI ramp-up trip points are about 100 mV higher than ramp-down trip points. This specifically built-in hysteresis prevents undesirable power-up oscillations and current surges. Note the following:

- During programming, I/Os become tristated and weakly pulled up to VCCI.
- JTAG supply, PLL power supplies, and charge pump VPUMP supply have no influence on I/O behavior.

Internal Power-Up Activation Sequence

1. Core
2. Input buffers
3. Output buffers, after 200 ns delay from input buffer activation

PLL Behavior at Brownout Condition

Microsemi recommends using monotonic power supplies or voltage regulators to ensure proper power-up behavior. Power ramp-up should be monotonic at least until VCC and VCCPLX exceed brownout activation levels. The V_{CC} activation level is specified as 1.1 V worst-case (see [Figure 3-1 on page 3-6](#) for more details).

When PLL power supply voltage and/or VCC levels drop below the VCC brownout levels ($0.75\text{ V} \pm 0.25\text{ V}$), the PLL output lock signal goes low and/or the output clock is lost.

Table 3-11 • AFS090 Quiescent Supply Current Characteristics (continued)

Parameter	Description	Conditions	Temp.	Min	Typ	Max	Unit
ICCNVM	Embedded NVM current	Reset asserted, VCCNVM = 1.575 V	T _J = 25°C		10	40	μA
			T _J = 85°C		14	40	μA
			T _J = 100°C		14	40	μA
ICCPLL	1.5 V PLL quiescent current	Operational standby, VCCPLL = 1.575 V	T _J = 25°C		65	100	μA
			T _J = 85°C		65	100	μA
			T _J = 100°C		65	100	μA

Notes:

1. ICC is the 1.5 V power supplies, ICC, ICCPLL, ICC15A, ICCNVM.
2. ICC33A includes ICC33A, ICC33PMP, and ICCOSC.
3. ICCI includes all ICCI0, ICCI1, and ICCI2.
4. Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.
5. XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.
6. Sleep Mode, VCC = VJTAG = VPUMP = 0 V.

RAM Dynamic Contribution— P_{MEMORY}

Operating Mode

$$P_{MEMORY} = (N_{BLOCKS} * PAC11 * \beta_2 * F_{READ-CLOCK}) + (N_{BLOCKS} * PAC12 * \beta_3 * F_{WRITE-CLOCK})$$

N_{BLOCKS} is the number of RAM blocks used in the design.

$F_{READ-CLOCK}$ is the memory read clock frequency.

β_2 is the RAM enable rate for read operations—guidelines are provided in [Table 3-17 on page 3-27](#).

β_3 the RAM enable rate for write operations—guidelines are provided in [Table 3-17 on page 3-27](#).

$F_{WRITE-CLOCK}$ is the memory write clock frequency.

Standby Mode and Sleep Mode

$$P_{MEMORY} = 0 \text{ W}$$

PLL/CCC Dynamic Contribution— P_{PLL}

Operating Mode

$$P_{PLL} = PAC13 * F_{CLKOUT}$$

F_{CLKIN} is the input clock frequency.

F_{CLKOUT} is the output clock frequency.¹

Standby Mode and Sleep Mode

$$P_{PLL} = 0 \text{ W}$$

Nonvolatile Memory Dynamic Contribution— P_{NVM}

Operating Mode

The NVM dynamic power consumption is a piecewise linear function of frequency.

$$P_{NVM} = N_{NVM-BLOCKS} * \beta_4 * PAC15 * F_{READ-NVM} \text{ when } F_{READ-NVM} \leq 33 \text{ MHz,}$$

$$P_{NVM} = N_{NVM-BLOCKS} * \beta_4 * (PAC16 + PAC17 * F_{READ-NVM} \text{ when } F_{READ-NVM} > 33 \text{ MHz}$$

$N_{NVM-BLOCKS}$ is the number of NVM blocks used in the design (2 in AFS600).

β_4 is the NVM enable rate for read operations. Default is 0 (NVM mainly in idle state).

$F_{READ-NVM}$ is the NVM read clock frequency.

Standby Mode and Sleep Mode

$$P_{NVM} = 0 \text{ W}$$

Crystal Oscillator Dynamic Contribution— $P_{XTL-OSC}$

Operating Mode

$$P_{XTL-OSC} = PAC18$$

Standby Mode

$$P_{XTL-OSC} = PAC18$$

Sleep Mode

$$P_{XTL-OSC} = 0 \text{ W}$$

1. The PLL dynamic contribution depends on the input clock frequency, the number of output clock signals generated by the PLL, and the frequency of each output clock. If a PLL is used to generate more than one output clock, include each output clock in the formula output clock by adding its corresponding contribution ($P_{AC14} * F_{CLKOUT}$ product) to the total PLL contribution.

FG256				
Pin Number	AFS090 Function	AFS250 Function	AFS600 Function	AFS1500 Function
E13	VCCIB1	VCCIB1	VCCIB2	VCCIB2
E14	GCC2/IO33NDB1V0	IO42NDB1V0	IO32NDB2V0	IO46NDB2V0
E15	GCB2/IO33PDB1V0	GBC2/IO42PDB1V0	GBC2/IO32PDB2V0	GBC2/IO46PDB2V0
E16	GND	GND	GND	GND
F1	NC	NC	IO79NDB4V0	IO111NDB4V0
F2	NC	NC	IO79PDB4V0	IO111PDB4V0
F3	GFB1/IO48PPB3V0	IO72NDB3V0	IO76NDB4V0	IO112NDB4V0
F4	GFC0/IO49NDB3V0	IO72PDB3V0	IO76PDB4V0	IO112PDB4V0
F5	NC	NC	IO82PSB4V0	IO120PSB4V0
F6	GFC1/IO49PDB3V0	GAC2/IO74PPB3V0	GAC2/IO83PPB4V0	GAC2/IO123PPB4V0
F7	NC	IO09RSB0V0	IO04PPB0V0	IO05PPB0V1
F8	NC	IO19RSB0V0	IO08NDB0V1	IO11NDB0V1
F9	NC	NC	IO20PDB1V0	IO27PDB1V1
F10	NC	IO29RSB0V0	IO23NDB1V1	IO37NDB1V2
F11	NC	IO43NDB1V0	IO36NDB2V0	IO50NDB2V0
F12	NC	IO43PDB1V0	IO36PDB2V0	IO50PDB2V0
F13	NC	IO44NDB1V0	IO39NDB2V0	IO59NDB2V0
F14	NC	GCA2/IO44PDB1V0	GCA2/IO39PDB2V0	GCA2/IO59PDB2V0
F15	GCC1/IO34PDB1V0	GCB2/IO45PDB1V0	GCB2/IO40PDB2V0	GCB2/IO60PDB2V0
F16	GCC0/IO34NDB1V0	IO45NDB1V0	IO40NDB2V0	IO60NDB2V0
G1	GEC0/IO46NPB3V0	IO70NPB3V0	IO74NPB4V0	IO109NPB4V0
G2	VCCIB3	VCCIB3	VCCIB4	VCCIB4
G3	GEC1/IO46PPB3V0	GFB2/IO70PPB3V0	GFB2/IO74PPB4V0	GFB2/IO109PPB4V0
G4	GFA1/IO47PDB3V0	GFA2/IO71PDB3V0	GFA2/IO75PDB4V0	GFA2/IO110PDB4V0
G5	GND	GND	GND	GND
G6	GFA0/IO47NDB3V0	IO71NDB3V0	IO75NDB4V0	IO110NDB4V0
G7	GND	GND	GND	GND
G8	VCC	VCC	VCC	VCC
G9	GND	GND	GND	GND
G10	VCC	VCC	VCC	VCC
G11	GDA1/IO37NDB1V0	GCC0/IO47NDB1V0	GCC0/IO43NDB2V0	GCC0/IO62NDB2V0
G12	GND	GND	GND	GND
G13	IO37PDB1V0	GCC1/IO47PDB1V0	GCC1/IO43PDB2V0	GCC1/IO62PDB2V0
G14	GCB0/IO35NPB1V0	IO46NPB1V0	IO41NPB2V0	IO61NPB2V0
G15	VCCIB1	VCCIB1	VCCIB2	VCCIB2
G16	GCB1/IO35PPB1V0	GCC2/IO46PPB1V0	GCC2/IO41PPB2V0	GCC2/IO61PPB2V0
H1	GEB1/IO45PDB3V0	GFC2/IO69PDB3V0	GFC2/IO73PDB4V0	GFC2/IO108PDB4V0
H2	GEB0/IO45NDB3V0	IO69NDB3V0	IO73NDB4V0	IO108NDB4V0

Revision	Changes	Page
Revision 3 (continued)	The "RC Oscillator" section was revised to correct a sentence that did not differentiate accuracy for commercial and industrial temperature ranges, which is given in Table 2-9 • Electrical Characteristics of RC Oscillator (SAR 33722).	2-19
	Figure 2-57 • FIFO Read and Figure 2-58 • FIFO Write are new (SAR 34840).	2-72
	The first paragraph of the "Offset" section was removed; it was intended to be replaced by the paragraph following it (SAR 22647).	2-95
	IOL and IOH values for 3.3 V GTL+ and 2.5 V GTL+ were corrected in Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions (SAR 39813).	2-164
	The drive strength, IOL, and IOH for 3.3 V GTL and 2.5 V GTL were changed from 25 mA to 20 mA in the following tables (SAR 37373): Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions, Table 2-92 • Summary of I/O Timing Characteristics – Software Default Settings Table 2-96 • I/O Output Buffer Maximum Resistances 1 Table 2-138 • Minimum and Maximum DC Input and Output Levels Table 2-141 • Minimum and Maximum DC Input and Output Levels	2-164 2-167 2-169 2-199 2-200
	The following sentence was deleted from the "2.5 V LVCMOS" section (SAR 34800): "It uses a 5 V–tolerant input buffer and push-pull output buffer."	2-181
	Corrected the inadvertent error in maximum values for LVPECL VIH and VIL and revised them to "3.6" in Table 2-171 • Minimum and Maximum DC Input and Output Levels, making these consistent with Table 3-1 • Absolute Maximum Ratings, and Table 3-4 • Overshoot and Undershoot Limits 1 (SAR 37687).	2-211
	The maximum frequency for global clock parameter was removed from Table 2-5 • AFS1500 Global Resource Timing through Table 2-8 • AFS090 Global Resource Timing because a frequency on the global is only an indication of what the global network can do. There are other limiters such as the SRAM, I/Os, and PLL. SmartTime software should be used to determine the design frequency (SAR 36955).	2-16 to 2-17
Revision 2 (March 2012)	The phrase "without debug" was removed from the "Soft ARM Cortex-M1 Fusion Devices (M1)" section (SAR 21390).	I
	The "In-System Programming (ISP) and Security" section, "Security" section, "Flash Advantages" section, and "Security" section were revised to clarify that although no existing security measures can give an absolute guarantee, Microsemi FPGAs implement the best security available in the industry (SAR 34679).	I, 1-2, 2-228
	The Y security option and Licensed DPA Logo was added to the "Product Ordering Codes" section. The trademarked Licensed DPA Logo identifies that a product is covered by a DPA counter-measures license from Cryptography Research (SAR 34721).	III
	The "Specifying I/O States During Programming" section is new (SAR 34693).	1-9
	The following information was added before Figure 2-17 • XTLOSC Macro: In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating (SAR 24119).	2-20
	Table 2-12 • Fusion CCC/PLL Specification was updated. A note was added indicating that when the CCC/PLL core is generated by Microsemi core generator software, not all delay values of the specified delay increments are available (SAR 34814).	2-28

Revision	Changes	Page
Advance v0.6 (continued)	The "Analog-to-Digital Converter Block" section was updated with the following statement: "All results are MSB justified in the ADC."	2-99
	The information about the ADCSTART signal was updated in the "ADC Description" section.	2-102
	Table 2-46 · Analog Channel Specifications was updated.	2-118
	Table 2-47 · ADC Characteristics in Direct Input Mode was updated.	2-121
	Table 2-51 · ACM Address Decode Table for Analog Quad was updated.	2-127
	In Table 2-53 · Analog Quad ACM Byte Assignment, the Function and Default Setting for Bit 6 in Byte 3 was updated.	2-130
	The "Introduction" section was updated to include information about digital inputs, outputs, and bibus.	2-133
	In Table 2-69 · Fusion Pro I/O Features, the programmable delay descriptions were updated for the following features: Single-ended receiver Voltage-referenced differential receiver LVDS/LVPECL differential receiver features	2-137
	The "User I/O Naming Convention" section was updated to include "V" and "z" descriptions	2-159
	The "VCC33PMP Analog Power Supply (3.3 V)" section was updated to include information about avoiding high current draw.	2-224
	The "VCCNVM Flash Memory Block Power Supply (1.5 V)" section was updated to include information about avoiding high current draw.	2-224
	The "VMVx I/O Supply Voltage (quiet)" section was updated to include this statement: VMV and VCCI must be connected to the same power supply and VCCI pins within a given I/O bank.	2-185
	The "PUB Push Button" section was updated to include information about leaving the pin floating if it is not used.	2-228
	The "PTBASE Pass Transistor Base" section was updated to include information about leaving the pin floating if it is not used.	2-228
The "PTM Pass Transistor Emitter" section was updated to include information about leaving the pin floating if it is not used.	2-228	
The heading was incorrect in the "208-Pin PQFP" table. It should be AFS250 and not AFS090.	3-8	