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The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	93
Number of Gates	250000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/m1afs250-pq208

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Fusion Device Family Overview

With Fusion, Microsemi also introduces the Analog Quad I/O structure (Figure 1-1). Each quad consists of three analog inputs and one gate driver. Each quad can be configured in various built-in circuit combinations, such as three prescaler circuits, three digital input circuits, a current monitor circuit, or a temperature monitor circuit. Each prescaler has multiple scaling factors programmed by FPGA signals to support a large range of analog inputs with positive or negative polarity. When the current monitor circuit is selected, two adjacent analog inputs measure the voltage drop across a small external sense resistor. For more information, refer to the "Analog System Characteristics" section on page 2-117. Built-in operational amplifiers amplify small voltage signals for accurate current measurement. One analog input in each quad can be connected to an external temperature monitor diode. In addition to the external temperature monitor diode(s), a Fusion device can monitor an internal temperature diode using dedicated channel 31 of the ADCMUX.

Figure 1-1 on page 1-5 illustrates a typical use of the Analog Quad I/O structure. The Analog Quad shown is configured to monitor and control an external power supply. The AV pad measures the source of the power supply. The AC pad measures the voltage drop across an external sense resistor to calculate current. The AG MOSFET gate driver pad turns the external MOSFET on and off. The AT pad measures the load-side voltage level.



Figure 1-1 • Analog Quad



RC Oscillator

The RC oscillator is an on-chip free-running clock source generating a 100 MHz clock. It can be used as a source clock for both on-chip and off-chip resources. When used in conjunction with the Fusion PLL and CCC circuits, the RC oscillator clock source can be used to generate clocks of varying frequency and phase.

The Fusion RC oscillator is very accurate at $\pm 1\%$ over commercial temperature ranges and and $\pm 3\%$ over industrial temperature ranges. It is an automated clock, requiring no setup or configuration by the user. It requires only that the power and GNDOSC pins be connected; no external components are required. The RC oscillator can be used to drive either a PLL or another internal signal.

RC Oscillator Characteristics

Parameter	Description	Conditions	Min.	Тур.	Max.	Units
	Operating Frequency			100		MHz
	Accuracy	Temperature: 0°C to 85°C Voltage: 3.3 V ± 5%		1		%
		Temperature: -40° C to 125° C Voltage: 3.3 V ± 5%		3		%
F _{RC} OL	Output Jitter	Period Jitter (at 5 k cycles)		100		ps
NO		Cycle–Cycle Jitter (at 5 k cycles)		100		ps
		Period Jitter (at 5 k cycles) with 1 KHz / 300 mV peak-to-peak noise on power supply		150		ps
		Cycle–Cycle Jitter (at 5 k cycles) with 1 KHz / 300 mV peak-to-peak noise on power supply		150		ps
	Output Duty Cycle			50		%
IDYNRC	Operating Current			1		mA

Table 2-9 • Electrical Characteristics of RC Oscillator





Figure 2-21 • Fusion CCC Options: Global Buffers with Programmable Delay

Global Input Selections

Each global buffer, as well as the PLL reference clock, can be driven from one of the following (Figure 2-22):

- 3 dedicated single-ended I/Os using a hardwired connection
- 2 dedicated differential I/Os using a hardwired connection
- The FPGA core



GAA[0:2]: GA represents global in the northwest corner of the device. A[0:2]: designates specific A clock source.

Notes:

- 1. Represents the global input pins. Globals have direct access to the clock conditioning block and are not routed via the FPGA fabric. Refer to the "User I/O Naming Convention" section on page 2-158 for more information.
- 2. Instantiate the routed clock source input as follows:
 - a) Connect the output of a logic element to the clock input of the PLL, CLKDLY, or CLKINT macro. b) Do not place a clock source I/O (INBUF or INBUF_LVPECL/LVDS) in a relevant global pin location.
- 3. LVDS-based clock sources are available in the east and west banks on all Fusion devices.

Figure 2-22 • Clock Input Sources Including CLKBUF, CLKBUF_LVDS/LVPECL, and CLKINT



Device Architecture

Table 2-19 • Flash Memory Block Pin Names (continued)

Interface Name	Width	Direction	Description
STATUS[1:0]	2	Out	Status of the last operation completed:
			00: Successful completion
			01: Read-/Unprotect-Page: single error detected and corrected
			Write: operation addressed a write-protected page Erase-Page: protection violation Program: Page Buffer is unmodified Protection violation
			10: Read-/Unprotect-Page: two or more errors detected
			11: Write: attempt to write to another page before programming current page
			Erase-Page/Program: page write count has exceeded the 10-year retention threshold
UNPROTECTPAGE	1	In	When asserted, the page addressed is copied into the Page Buffer and the Page Buffer is made writable.
WD[31:0]	32	In	Write data
WEN	1	In	When asserted, stores WD in the page buffer.

All flash memory block input signals are active high, except for RESET.



Direct Digital Input

The AV, AC, and AT pads can also be configured as high-voltage digital inputs (Figure 2-69). As these pads are 12 V–tolerant, the digital input can also be up to 12 V. However, the frequency at which these pads can operate is limited to 10 MHz.

To enable one of these analog input pads to operate as a digital input, its corresponding Digital Input Enable (DENAxy) pin on the Analog Block must be pulled High, where x is either V, C, or T (for AV, AC, or AT pads, respectively) and y is in the range 0 to 9, corresponding to the appropriate Analog Quad.

When the pad is configured as a digital input, the signal will come out of the Analog Block macro on the appropriate DAxOUTy pin, where x represents the pad type (V for AV pad, C for AC pad, or T for AT pad) and y represents the appropriate Analog Quad number. Example: If the AT pad in Analog Quad 5 is configured as a digital input, it will come out on the DATOUT5 pin of the Analog Block macro.



Figure 2-69 • Analog Quad Direct Digital Input Configuration



Terminology

Resolution

Resolution defines the smallest temperature change Fusion Temperature Monitor can resolve. For ADC configured as 8-bit mode, each LSB represents 4°C, and 1°C per LSB for 10-bit mode. With 12-bit mode, the Temperature Monitor can still only resolve 1°C due to Temperature Monitor design.

Offset

The Fusion Temperature Monitor has a systematic offset (Table 2-49 on page 2-117), excluding error due to board resistance and ideality factor of the external diode. Microsemi provides an IP block (CalibIP) that is required in order to mitigate the systematic temperature offset. For further details on CalibIP, refer to the "Temperature, Voltage, and Current Calibration in Fusion FPGAs" chapter of the Fusion FPGA Fabric User Guide.

There are several popular ADC architectures, each with advantages and limitations.

The analog-to-digital converter in Fusion devices is a switched-capacitor Successive Approximation Register (SAR) ADC. It supports 8-, 10-, and 12-bit modes of operation with a cumulative sample rate up to 600 k samples per second (ksps). Built-in bandgap circuitry offers 1% internal voltage reference accuracy or an external reference voltage can be used.

As shown in Figure 2-81, a SAR ADC contains N capacitors with binary-weighted values.



Figure 2-81 • Example SAR ADC Architecture

To begin a conversion, all of the capacitors are quickly discharged. Then VIN is applied to all the capacitors for a period of time (acquisition time) during which the capacitors are charged to a value very close to VIN. Then all of the capacitors are switched to ground, and thus –VIN is applied across the comparator. Now the conversion process begins. First, C is switched to VREF Because of the binary weighting of the capacitors, the voltage at the input of the comparator is then shown by EQ 11.

Voltage at input of comparator = -VIN + VREF / 2

EQ 11

If VIN is greater than VREF / 2, the output of the comparator is 1; otherwise, the comparator output is 0. A register is clocked to retain this value as the MSB of the result. Next, if the MSB is 0, C is switched back to ground; otherwise, it remains connected to VREF, and C / 2 is connected to VREF. The result at the comparator input is now either –VIN + VREF / 4 or –VIN + 3 VREF / 4 (depending on the state of the MSB), and the comparator output now indicates the value of the next most significant bit. This bit is likewise registered, and the process continues for each subsequent bit until a conversion is complete. The conversion process requires some acquisition time plus N + 1 ADC clock cycles to complete.

INL – Integral Non-Linearity

INL is the deviation of an actual transfer function from a straight line. After nullifying offset and gain errors, the straight line is either a best-fit straight line or a line drawn between the end points of the transfer function (Figure 2-85).



Figure 2-85 • Integral Non-Linearity (INL)

LSB – Least Significant Bit

In a binary number, the LSB is the least weighted bit in the group. Typically, the LSB is the furthest right bit. For an ADC, the weight of an LSB equals the full-scale voltage range of the converter divided by 2^N , where N is the converter's resolution.

EQ 13 shows the calculation for a 10-bit ADC with a unipolar full-scale voltage of 2.56 V:

EQ 13

No Missing Codes

An ADC has no missing codes if it produces all possible digital codes in response to a ramp signal applied to the analog input.



Device Architecture

Similarly,

Min. Output Voltage = (Max. Negative input offset) + (Input Voltage x Max. Negative Channel Gain) = $(-88 \text{ mV}) + (5 \text{ V} \times 0.96) = 4.712 \text{ V}$

Calculating Accuracy for a Calibrated Analog Channel

Formula

For a given prescaler range, EQ 31 gives the output voltage.

Output Voltage = Channel Error in V + Input Voltage

EQ 31

where

Channel Error in V = Total Channel Error in LSBs x Equivalent voltage per LSB

Example

Input Voltage = 5 VChosen Prescaler range = 8 V range Refer to Table 2-52 on page 2-123.

Max. Output Voltage = Max. Positive Channel Error in V + Input Voltage Max. Positive Channel Error in V = (6 LSB) × (8 mV per LSB in 10-bit mode) = 48 mV Max. Output Voltage = 48 mV + 5 V = **5.048 V**

Similarly,

Min. Output Voltage = Max. Negative Channel Error in V + Input Voltage = (-48 mV) + 5 V = 4.952 V

Calculating LSBs from a Given Error Budget

Formula

For a given prescaler range, LSB count = ± (Input Voltage × Required % error) / (Equivalent voltage per LSB)

Example

Input Voltage = $3.3 \vee$ Required error margin= 1% Refer to Table 2-52 on page 2-123. Equivalent voltage per LSB = 16 mV for a 16V prescaler, with ADC in 10-bit mode LSB Count = $\pm (5.0 \vee \times 1\%) / (0.016)$ LSB Count = ± 3.125 Equivalent voltage per LSB = 8 mV for an $8 \vee$ prescaler, with ADC in 10-bit mode LSB Count = $\pm (5.0 \vee \times 1\%) / (0.008)$ LSB Count = $\pm (5.0 \vee \times 1\%) / (0.008)$ LSB Count = ± 6.25 The $8 \vee$ prescaler satisfies the calculated LSB count accuracy requirement (see Table 2-52 on page 2-123).





Figure 2-116 • Input Buffer Timing Model and Delays (example)

Table 2-99 • Short Current Event Duration before Failure

Temperature	Time Before Failure
-40°C	>20 years
0°C	>20 years
25°C	>20 years
70°C	5 years
85°C	2 years
100°C	6 months

Table 2-100 • Schmitt Trigger Input Hysteresis Hysteresis Voltage Value (typ.) for Schmitt Mode Input Buffers

Input Buffer Configuration	Hysteresis Value (typ.)
3.3 V LVTTL/LVCMOS/PCI/PCI-X (Schmitt trigger mode)	240 mV
2.5 V LVCMOS (Schmitt trigger mode)	140 mV
1.8 V LVCMOS (Schmitt trigger mode)	80 mV
1.5 V LVCMOS (Schmitt trigger mode)	60 mV

Table 2-101 • I/O Input Rise Time, Fall Time, and Related I/O Reliability

Input Buffer	Input Rise/Fall Time (min.)	Input Rise/Fall Time (max.)	Reliability
LVTTL/LVCMOS (Schmitt trigger disabled)	No requirement	10 ns*	20 years (100°C)
LVTTL/LVCMOS (Schmitt trigger enabled)	No requirement	No requirement, but input noise voltage cannot exceed Schmitt hysteresis	20 years (100°C)
HSTL/SSTL/GTL	No requirement	10 ns*	10 years (100°C)
LVDS/BLVDS/M-LVDS/LVPECL	No requirement	10 ns*	10 years (100°C)

Note: * The maximum input rise/fall time is related only to the noise induced into the input buffer trace. If the noise is low, the rise time and fall time of input buffers, when Schmitt trigger is disabled, can be increased beyond the maximum value. The longer the rise/fall times, the more susceptible the input signal is to the board noise. Microsemi recommends signal integrity evaluation/characterization of the system to ensure there is no excessive noise coupling into input signals.

Timing Characteristics

Table 2-112 • 2.5 V LVCMOS Low Slew

Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V Applicable to Pro I/Os

Drive	Speed													
Strength	Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{PYS}	t _{EOUT}	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
4 mA	Std.	0.60	12.00	0.04	1.51	1.66	0.43	12.23	11.61	2.72	2.20	14.46	13.85	ns
	-1	0.51	10.21	0.04	1.29	1.41	0.36	10.40	9.88	2.31	1.87	12.30	11.78	ns
	-2	0.45	8.96	0.03	1.13	1.24	0.32	9.13	8.67	2.03	1.64	10.80	10.34	ns
8 mA	Std.	0.60	8.73	0.04	1.51	1.66	0.43	8.89	8.01	3.10	2.93	11.13	10.25	ns
	-1	0.51	7.43	0.04	1.29	1.41	0.36	7.57	6.82	2.64	2.49	9.47	8.72	ns
	-2	0.45	6.52	0.03	1.13	1.24	0.32	6.64	5.98	2.32	2.19	8.31	7.65	ns
12 mA	Std.	0.66	6.77	0.04	1.51	1.66	0.43	6.90	6.11	3.37	3.39	9.14	8.34	ns
	-1	0.56	5.76	0.04	1.29	1.41	0.36	5.87	5.20	2.86	2.89	7.77	7.10	ns
	-2	0.49	5.06	0.03	1.13	1.24	0.32	5.15	4.56	2.51	2.53	6.82	6.23	ns
16 mA	Std.	0.66	6.31	0.04	1.51	1.66	0.43	6.42	5.73	3.42	3.52	8.66	7.96	ns
	-1	0.56	5.37	0.04	1.29	1.41	0.36	5.46	4.87	2.91	3.00	7.37	6.77	ns
	-2	0.49	4.71	0.03	1.13	1.24	0.32	4.80	4.28	2.56	2.63	6.47	5.95	ns
24 mA	Std.	0.66	5.93	0.04	1.51	1.66	0.43	6.04	5.70	3.49	4.00	8.28	7.94	ns
	-1	0.56	5.05	0.04	1.29	1.41	0.36	5.14	4.85	2.97	3.40	7.04	6.75	ns
	-2	0.49	4.43	0.03	1.13	1.24	0.32	4.51	4.26	2.61	2.99	6.18	5.93	ns

Table 2-122 • 1.8 V LVCMOS Low Slew

Commercial Temperature Range Conditions: $T_J = 70^{\circ}$ C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.7 V Applicable to Advanced I/Os

Drive Strength	Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
2 mA	Std.	0.66	15.53	0.04	1.31	0.43	14.11	15.53	2.78	1.60	16.35	17.77	ns
	-1	0.56	13.21	0.04	1.11	0.36	12.01	13.21	2.36	1.36	13.91	15.11	ns
	-2 ²	0.49	11.60	0.03	0.98	0.32	10.54	11.60	2.07	1.19	12.21	13.27	ns
4 mA	Std.	0.66	10.48	0.04	1.31	0.43	10.41	10.48	3.23	2.73	12.65	12.71	ns
	-1	0.56	8.91	0.04	1.11	0.36	8.86	8.91	2.75	2.33	10.76	10.81	ns
	-2	0.49	7.82	0.03	0.98	0.32	7.77	7.82	2.41	2.04	9.44	9.49	ns
8 mA	Std.	0.66	8.05	0.04	1.31	0.43	8.20	7.84	3.54	3.27	10.43	10.08	ns
	-1	0.56	6.85	0.04	1.11	0.36	6.97	6.67	3.01	2.78	8.88	8.57	ns
	-2	0.49	6.01	0.03	0.98	0.32	6.12	5.86	2.64	2.44	7.79	7.53	ns
12 mA	Std.	0.66	7.50	0.04	1.31	0.43	7.64	7.30	3.61	3.41	9.88	9.53	ns
	-1	0.56	6.38	0.04	1.11	0.36	6.50	6.21	3.07	2.90	8.40	8.11	ns
	-2	0.49	5.60	0.03	0.98	0.32	5.71	5.45	2.69	2.55	7.38	7.12	ns
16 mA	Std.	0.66	7.29	0.04	1.31	0.43	7.23	7.29	3.71	3.95	9.47	9.53	ns
	-1	0.56	6.20	0.04	1.11	0.36	6.15	6.20	3.15	3.36	8.06	8.11	ns
	-2	0.49	5.45	0.03	0.98	0.32	5.40	5.45	2.77	2.95	7.07	7.12	ns

Timing Characteristics

Table 2-128 • 1.5 V LVCMOS Low Slew

Commercial Temperature Range Conditions: $T_J = 70^{\circ}$ C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.4 V Applicable to Pro I/Os

Drive Strength	Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{PYS}	t _{EOUT}	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
2 mA	Std.	0.66	14.11	0.04	1.70	2.14	0.43	14.37	13.14	3.40	2.68	16.61	15.37	ns
	-1	0.56	12.00	0.04	1.44	1.82	0.36	12.22	11.17	2.90	2.28	14.13	13.08	ns
	-2	0.49	10.54	0.03	1.27	1.60	0.32	10.73	9.81	2.54	2.00	12.40	11.48	ns
4 mA	Std.	0.66	11.23	0.04	1.70	2.14	0.43	11.44	9.87	3.77	3.36	13.68	12.10	ns
	-1	0.56	9.55	0.04	1.44	1.82	0.36	9.73	8.39	3.21	2.86	11.63	10.29	ns
	-2	0.49	8.39	0.03	1.27	1.60	0.32	8.54	7.37	2.81	2.51	10.21	9.04	ns
8 mA	Std.	0.66	10.45	0.04	1.70	2.14	0.43	10.65	9.24	3.84	3.55	12.88	11.48	ns
	-1	0.56	8.89	0.04	1.44	1.82	0.36	9.06	7.86	3.27	3.02	10.96	9.76	ns
	-2	0.49	7.81	0.03	1.27	1.60	0.32	7.95	6.90	2.87	2.65	9.62	8.57	ns
12 mA	Std.	0.66	10.02	0.04	1.70	2.14	0.43	10.20	9.23	3.97	4.22	12.44	11.47	ns
	-1	0.56	8.52	0.04	1.44	1.82	0.36	8.68	7.85	3.38	3.59	10.58	9.75	ns
	-2	0.49	7.48	0.03	1.27	1.60	0.32	7.62	6.89	2.97	3.15	9.29	8.56	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-129 • 1.5 V LVCMOS High Slew

Commercial Temperature Range Conditions: $T_J = 70^{\circ}$ C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.4 V Applicable to Pro I/Os

Drive Strength	Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{PYS}	t _{EOU} T	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
2 mA	Std.	0.66	8.53	0.04	1.70	2.14	0.43	7.26	8.53	3.39	2.79	9.50	10.77	ns
	-1	0.56	7.26	0.04	1.44	1.82	0.36	6.18	7.26	2.89	2.37	8.08	9.16	ns
	-2	0.49	6.37	0.03	1.27	1.60	0.32	5.42	6.37	2.53	2.08	7.09	8.04	ns
4 mA	Std.	0.66	5.41	0.04	1.70	2.14	0.43	5.22	5.41	3.75	3.48	7.45	7.65	ns
	-1	0.56	4.60	0.04	1.44	1.82	0.36	4.44	4.60	3.19	2.96	6.34	6.50	ns
	-2	0.49	4.04	0.03	1.27	1.60	0.32	3.89	4.04	2.80	2.60	5.56	5.71	ns
8 mA	Std.	0.66	4.80	0.04	1.70	2.14	0.43	4.89	4.75	3.83	3.67	7.13	6.98	ns
	-1	0.56	4.09	0.04	1.44	1.82	0.36	4.16	4.04	3.26	3.12	6.06	5.94	ns
	-2	0.49	3.59	0.03	1.27	1.60	0.32	3.65	3.54	2.86	2.74	5.32	5.21	ns
12 mA	Std.	0.66	4.42	0.04	1.70	2.14	0.43	4.50	3.62	3.96	4.37	6.74	5.86	ns
	-1	0.56	3.76	0.04	1.44	1.82	0.36	3.83	3.08	3.37	3.72	5.73	4.98	ns
	-2	0.49	3.30	0.03	1.27	1.60	0.32	3.36	2.70	2.96	3.27	5.03	4.37	ns



Voltage Referenced I/O Characteristics

3.3 V GTL

Gunning Transceiver Logic is a high-speed bus standard (JESD8-3). It provides a differential amplifier input buffer and an open-drain output buffer. The VCCI pin should be connected to 3.3 V.

3.3 V GTL		VIL	VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL ¹	IIH ²
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA ³	Max. mA ³	μA ⁴	μA ⁴
20 mA ³	-0.3	VREF – 0.05	VREF + 0.05	3.6	0.4	_	20	20	181	268	10	10

Table 2-138 • Minimum and Maximum DC Input and Output Levels

Notes:

1. IIL is the input leakage current per I/O pin over recommended operation conditions where –0.3 V < VIN < VIL.

2. IIH is the input leakage current per I/O pin over recommended operating conditions VIH < VIN < VCCI. Input current is larger when operating outside recommended ranges.

3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.

4. Currents are measured at 85°C junction temperature.



Figure 2-124 • AC Loading

Table 2-139 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C _{LOAD} (pF)
VREF – 0.05	VREF + 0.05	0.8	0.8	1.2	10

Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.

Timing Characteristics

Table 2-140 • 3.3 V GTL

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Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 3.0 V, VREF = 0.8 V
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Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
Std.	0.66	2.08	0.04	2.93	0.43	2.04	2.08			4.27	4.31	ns
-1	0.56	1.77	0.04	2.50	0.36	1.73	1.77			3.63	3.67	ns
-2	0.49	1.55	0.03	2.19	0.32	1.52	1.55			3.19	3.22	ns





Figure	2-145 •	Output DDR	Timing	Diagram

Timing Characteristics

Table 2-182 • Output DDR Propagation Delays	
Commercial Temperature Range Conditions: T ₁ = 70°C, Worst-Case VCC = 1.425 V	

Parameter	Description	-2	-1	Std.	Units
t _{DDROCLKQ}	Clock-to-Out of DDR for Output DDR	0.70	0.80	0.94	ns
t _{DDROSUD1}	Data_F Data Setup for Output DDR	0.38	0.43	0.51	ns
t _{DDROSUD2}	Data_R Data Setup for Output DDR	0.38	0.43	0.51	ns
t _{DDROHD1}	Data_F Data Hold for Output DDR	0.00	0.00	0.00	ns
t _{DDROHD2}	Data_R Data Hold for Output DDR	0.00	0.00	0.00	ns
t _{DDROCLR2Q}	Asynchronous Clear-to-Out for Output DDR	0.80	0.91	1.07	ns
t _{DDROREMCLR}	Asynchronous Clear Removal Time for Output DDR	0.00	0.00	0.00	ns
t _{DDRORECCLR}	Asynchronous Clear Recovery Time for Output DDR	0.22	0.25	0.30	ns
t _{DDROWCLR1}	Asynchronous Clear Minimum Pulse Width for Output DDR	0.22	0.25	0.30	ns
t _{DDROCKMPWH}	Clock Minimum Pulse Width High for the Output DDR	0.36	0.41	0.48	ns
t _{DDROCKMPWL}	Clock Minimum Pulse Width Low for the Output DDR	0.32	0.37	0.43	ns
F _{DDOMAX}	Maximum Frequency for the Output DDR	1404	1232	1048	MHz

Methodology

Total Power Consumption—PTOTAL

Operating Mode, Standby Mode, and Sleep Mode

 $P_{TOTAL} = P_{STAT} + P_{DYN}$

P_{STAT} is the total static power consumption.

P_{DYN} is the total dynamic power consumption.

Total Static Power Consumption—P_{STAT}

Operating Mode

 $\label{eq:pstat} \begin{array}{l} \mathsf{P}_{\mathsf{STAT}} = \mathsf{PDC1} + (\mathsf{N}_{\mathsf{NVM-BLOCKS}} * \mathsf{PDC4}) + \mathsf{PDC5} + (\mathsf{N}_{\mathsf{QUADS}} * \mathsf{PDC6}) + (\mathsf{N}_{\mathsf{INPUTS}} * \mathsf{PDC7}) + (\mathsf{N}_{\mathsf{OUTPUTS}} * \mathsf{PDC8}) + (\mathsf{N}_{\mathsf{PLLS}} * \mathsf{PDC9}) \end{array}$

 $N_{\ensuremath{\mathsf{NVM}}\xspace-BLOCKS}$ is the number of NVM blocks available in the device.

 N_{QUADS} is the number of Analog Quads used in the design.

N_{INPUTS} is the number of I/O input buffers used in the design.

N_{OUTPUTS} is the number of I/O output buffers used in the design.

N_{PLLS} is the number of PLLs available in the device.

Standby Mode

P_{STAT} = PDC2

Sleep Mode

P_{STAT} = PDC3

Total Dynamic Power Consumption—P_{DYN}

Operating Mode

P_{DYN} = P_{CLOCK} + P_{S-CELL} + P_{C-CELL} + P_{NET} + P_{INPUTS} + P_{OUTPUTS} + P_{MEMORY} + P_{PLL} + P_{NVM}+ P_{XTL-OSC} + P_{RC-OSC} + P_{AB}

Standby Mode

 $P_{DYN} = P_{XTL-OSC}$

Sleep Mode

 $P_{DYN} = 0 W$

Global Clock Dynamic Contribution—P_{CLOCK}

Operating Mode

 $P_{CLOCK} = (PAC1 + N_{SPINE} * PAC2 + N_{ROW} * PAC3 + N_{S-CELL} * PAC4) * F_{CLK}$

N_{SPINE} is the number of global spines used in the user design—guidelines are provided in the "Spine Architecture" section of the Global Resources chapter in the *Fusion and Extended Temperature Fusion FPGA Fabric User's Guide*.

N_{ROW} is the number of VersaTile rows used in the design—guidelines are provided in the "Spine Architecture" section of the Global Resources chapter in the *Fusion and Extended Temperature Fusion FPGA Fabric User's Guide*.

 $\mathsf{F}_{\mathsf{CLK}}$ is the global clock signal frequency.

N_{S-CELL} is the number of VersaTiles used as sequential modules in the design.

Standby Mode and Sleep Mode

 $P_{CLOCK} = 0 W$

Sequential Cells Dynamic Contribution—P_{S-CELL}

Operating Mode



Package Pin Assignments

	PQ208		PQ208				
Pin Number	AFS250 Function	AFS600 Function	Pin Number	Pin Number AFS250 Function A			
1	VCCPLA	VCCPLA	38	IO60NDB3V0	GEB0/IO62NDB4V0		
2	VCOMPLA	VCOMPLA	39	GND	GEA1/IO61PDB4V0		
3	GNDQ	GAA2/IO85PDB4V0	40	VCCIB3	GEA0/IO61NDB4V0		
4	VCCIB3	IO85NDB4V0	41	GEB2/IO59PDB3V0	GEC2/IO60PDB4V0		
5	GAA2/IO76PDB3V0	GAB2/IO84PDB4V0	42	IO59NDB3V0	IO60NDB4V0		
6	IO76NDB3V0	IO84NDB4V0	43	GEA2/IO58PDB3V0	VCCIB4		
7	GAB2/IO75PDB3V0	GAC2/IO83PDB4V0	44	IO58NDB3V0	GNDQ		
8	IO75NDB3V0	IO83NDB4V0	45	VCC	VCC		
9	NC	IO77PDB4V0	45	VCC	VCC		
10	NC	IO77NDB4V0	46	VCCNVM	VCCNVM		
11	VCC	IO76PDB4V0	47	GNDNVM	GNDNVM		
12	GND	IO76NDB4V0	48	GND	GND		
13	VCCIB3	VCC	49	VCC15A	VCC15A		
14	IO72PDB3V0	GND	50	PCAP	PCAP		
15	IO72NDB3V0	VCCIB4	51	NCAP	NCAP		
16	GFA2/IO71PDB3V0	GFA2/IO75PDB4V0	52	VCC33PMP	VCC33PMP		
17	IO71NDB3V0	IO75NDB4V0	53	VCC33N	VCC33N		
18	GFB2/IO70PDB3V0	GFC2/IO73PDB4V0	54	GNDA	GNDA		
19	IO70NDB3V0	IO73NDB4V0	55	GNDAQ	GNDAQ		
20	GFC2/IO69PDB3V0	VCCOSC	56	NC	AV0		
21	IO69NDB3V0	XTAL1	57	NC	AC0		
22	VCC	XTAL2	58	NC	AG0		
23	GND	GNDOSC	59	NC	AT0		
24	VCCIB3	GFC1/IO72PDB4V0	60	NC	ATRTN0		
25	GFC1/IO68PDB3V0	GFC0/IO72NDB4V0	61	NC	AT1		
26	GFC0/IO68NDB3V0	GFB1/IO71PDB4V0	62	NC	AG1		
27	GFB1/IO67PDB3V0	GFB0/IO71NDB4V0	63	NC	AC1		
28	GFB0/IO67NDB3V0	GFA1/IO70PDB4V0	64	NC	AV1		
29	VCCOSC	GFA0/IO70NDB4V0	65	AV0	AV2		
30	XTAL1	IO69PDB4V0	66	AC0	AC2		
31	XTAL2	IO69NDB4V0	67	AG0	AG2		
32	GNDOSC	VCC	68	AT0	AT2		
33	GEB1/IO62PDB3V0	GND	69	ATRTN0	ATRTN1		
34	GEB0/IO62NDB3V0	VCCIB4	70	AT1	AT3		
35	GEA1/IO61PDB3V0	GEC1/IO63PDB4V0	71	AG1	AG3		
36	GEA0/IO61NDB3V0	GEC0/IO63NDB4V0	72	AC1	AC3		
37	GEC2/IO60PDB3V0	GEB1/IO62PDB4V0	73	AV1	AV3		



FG256							
Pin Number	AFS090 Function	AFS250 Function	AFS600 Function	AFS1500 Function			
C7	IO09RSB0V0	IO12RSB0V0	IO06NDB0V0	IO09NDB0V1			
C8	IO14RSB0V0	IO22RSB0V0	IO16PDB1V0	IO23PDB1V0			
C9	IO15RSB0V0	IO23RSB0V0	IO16NDB1V0	IO23NDB1V0			
C10	IO22RSB0V0	IO30RSB0V0	IO25NDB1V1	IO31NDB1V1			
C11	IO20RSB0V0	IO31RSB0V0	IO25PDB1V1	IO31PDB1V1			
C12	VCCIB0	VCCIB0	VCCIB1	VCCIB1			
C13	GBB1/IO28RSB0V0	GBC1/IO35RSB0V0	GBC1/IO26PPB1V1	GBC1/IO40PPB1V2			
C14	VCCIB1	VCCIB1	VCCIB2	VCCIB2			
C15	GND	GND	GND	GND			
C16	VCCIB1	VCCIB1	VCCIB2	VCCIB2			
D1	GFC2/IO50NPB3V0	IO75NDB3V0	IO84NDB4V0	IO124NDB4V0			
D2	GFA2/IO51NDB3V0	GAB2/IO75PDB3V0	GAB2/IO84PDB4V0	GAB2/IO124PDB4V0			
D3	GAC2/IO51PDB3V0	IO76NDB3V0	IO85NDB4V0	IO125NDB4V0			
D4	GAA2/IO52PDB3V0	GAA2/IO76PDB3V0	GAA2/IO85PDB4V0	GAA2/IO125PDB4V0			
D5	GAB2/IO52NDB3V0	GAB0/IO02RSB0V0	GAB0/IO02NPB0V0	GAB0/IO02NPB0V0			
D6	GAC0/IO04RSB0V0	GAC0/IO04RSB0V0	GAC0/IO03NDB0V0	GAC0/IO03NDB0V0			
D7	IO08RSB0V0	IO13RSB0V0	IO06PDB0V0	IO09PDB0V1			
D8	NC	IO20RSB0V0	IO14NDB0V1	IO15NDB0V2			
D9	NC	IO21RSB0V0	IO14PDB0V1	IO15PDB0V2			
D10	IO21RSB0V0	IO28RSB0V0	IO23PDB1V1	IO37PDB1V2			
D11	IO23RSB0V0	GBB0/IO36RSB0V0	GBB0/IO27NDB1V1	GBB0/IO41NDB1V2			
D12	NC	NC	VCCIB1	VCCIB1			
D13	GBA2/IO31PDB1V0	GBA2/IO40PDB1V0	GBA2/IO30PDB2V0	GBA2/IO44PDB2V0			
D14	GBB2/IO31NDB1V0	IO40NDB1V0	IO30NDB2V0	IO44NDB2V0			
D15	GBC2/IO32PDB1V0	GBB2/IO41PDB1V0	GBB2/IO31PDB2V0	GBB2/IO45PDB2V0			
D16	GCA2/IO32NDB1V0	IO41NDB1V0	IO31NDB2V0	IO45NDB2V0			
E1	GND	GND	GND	GND			
E2	GFB0/IO48NPB3V0	IO73NDB3V0	IO81NDB4V0	IO118NDB4V0			
E3	GFB2/IO50PPB3V0	IO73PDB3V0	IO81PDB4V0	IO118PDB4V0			
E4	VCCIB3	VCCIB3	VCCIB4	VCCIB4			
E5	NC	IO74NPB3V0	IO83NPB4V0	IO123NPB4V0			
E6	NC	IO08RSB0V0	IO04NPB0V0	IO05NPB0V1			
E7	GND	GND	GND	GND			
E8	NC	IO18RSB0V0	IO08PDB0V1	IO11PDB0V1			
E9	NC	NC	IO20NDB1V0	IO27NDB1V1			
E10	GND	GND	GND	GND			
E11	IO24RSB0V0	GBB1/IO37RSB0V0	GBB1/IO27PDB1V1	GBB1/IO41PDB1V2			
E12	NC	IO50PPB1V0	IO33PSB2V0	IO48PSB2V0			

Fusion Family of Mixed Signal FPGAs

	FG676		FG676	FG676		
Pin Number	AFS1500 Function	Pin Number	AFS1500 Function	Pin Number	AFS1500 Function	
R21	IO72NDB2V0	U5	VCCIB4	V15	AC5	
R22	IO72PDB2V0	U6	IO91PDB4V0	V16	NC	
R23	GND	U7	IO91NDB4V0	V17	GNDA	
R24	IO71PDB2V0	U8	IO92PDB4V0	V18	IO77PPB2V0	
R25	VCCIB2	U9	GND	V19	IO74PDB2V0	
R26	IO67NDB2V0	U10	GND	V20	VCCIB2	
T1	GND	U11	VCC33A	V21	IO82NDB2V0	
T2	NC	U12	GNDA	V22	GDA2/IO82PDB2V0	
Т3	GFA1/IO105PDB4V0	U13	VCC33A	V23	GND	
T4	GFA0/IO105NDB4V0	U14	GNDA	V24	GDC1/IO79PDB2V0	
T5	IO101NDB4V0	U15	VCC33A	V25	VCCIB2	
Т6	IO96PDB4V0	U16	GNDA	V26	NC	
Τ7	IO96NDB4V0	U17	VCC	W1	GND	
Т8	IO99NDB4V0	U18	GND	W2	IO94PPB4V0	
Т9	IO97NDB4V0	U19	IO74NDB2V0	W3	IO98PDB4V0	
T10	VCCIB4	U20	GDA0/IO81NDB2V0	W4	IO98NDB4V0	
T11	VCC	U21	GDB0/IO80NDB2V0	W5	GEC1/IO90PDB4V0	
T12	GND	U22	VCCIB2	W6	GEC0/IO90NDB4V0	
T13	VCC	U23	IO75NDB2V0	W7	GND	
T14	GND	U24	IO75PDB2V0	W8	VCCNVM	
T15	VCC	U25	NC	W9	VCCIB4	
T16	GND	U26	NC	W10	VCC15A	
T17	VCCIB2	V1	NC	W11	GNDA	
T18	IO83NDB2V0	V2	VCCIB4	W12	AC4	
T19	IO78NDB2V0	V3	IO100PPB4V0	W13	VCC33A	
T20	GDA1/IO81PDB2V0	V4	GND	W14	GNDA	
T21	GDB1/IO80PDB2V0	V5	IO95PDB4V0	W15	AG5	
T22	IO73NDB2V0	V6	IO95NDB4V0	W16	GNDA	
T23	IO73PDB2V0	V7	VCCIB4	W17	PUB	
T24	IO71NDB2V0	V8	IO92NDB4V0	W18	VCCIB2	
T25	NC	V9	GNDNVM	W19	TDI	
T26	GND	V10	GNDA	W20	GND	
U1	NC	V11	NC	W21	IO84NDB2V0	
U2	NC	V12	AV4	W22	GDC2/IO84PDB2V0	
U3	IO102PDB4V0	V13	NC	W23	IO77NPB2V0	
U4	IO102NDB4V0	V14	AV5	W24	GDC0/IO79NDB2V0	