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Details

Product Status	Obsolete
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	39
Program Memory Size	8KB (8K x 8)
Program Memory Type	FLASH
EEPROM Size	· · · · · · · · · · · · · · · · · · ·
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN-EP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08gt8acfder

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makes the MCU secure. During development, whenever the FLASH is erased, it is good practice to immediately program the SEC00 bit to 0 in NVOPT so SEC01:SEC00 = 1:0. This would allow the MCU to remain unsecured after a subsequent reset.

The on-chip debug module cannot be enabled while the MCU is secure. The separate background debug controller can still be used for background memory access commands, but the MCU cannot enter active background mode except by holding BKGD/MS low at the rising edge of reset.

A user can choose to allow or disallow a security unlocking mechanism through an 8-byte backdoor security key. If the nonvolatile KEYEN bit in NVOPT/FOPT is 0, the backdoor key is disabled and there is no way to disengage security without completely erasing all FLASH locations. If KEYEN is 1, a secure user program can temporarily disengage security by:

- 1. Writing 1 to KEYACC in the FCNFG register. This makes the FLASH module interpret writes to the backdoor comparison key locations (NVBACKKEY through NVBACKKEY+7) as values to be compared against the key rather than as the first step in a FLASH program or erase command.
- 2. Writing the user-entered key values to the NVBACKKEY through NVBACKKEY+7 locations. These writes must be done in order, starting with the value for NVBACKKEY and ending with NVBACKKEY+7. STHX should not be used for these writes because these writes cannot be done on adjacent bus cycles. User software normally would get the key codes from outside the MCU system through a communication interface such as a serial I/O.
- 3. Writing 0 to KEYACC in the FCNFG register. If the 8-byte key that was just written matches the key stored in the FLASH locations, SEC01:SEC00 are automatically changed to 1:0 and security will be disengaged until the next reset.

The security key can be written only from RAM, so it cannot be entered through background commands without the cooperation of a secure user program.

The backdoor comparison key (NVBACKKEY through NVBACKKEY+7) is located in FLASH memory locations in the nonvolatile register space so users can program these locations exactly as they would program any other FLASH memory location. The nonvolatile registers are in the same 512-byte block of FLASH as the reset and interrupt vectors, so block protecting that space also block protects the backdoor comparison key. Block protects cannot be changed from user application programs, so if the vector space is block protected, the backdoor security key mechanism cannot permanently change the block protect, security settings, or the backdoor key.

Security can always be disengaged through the background debug interface by performing these steps:

- 1. Disable any block protections by writing FPROT. FPROT can be written only with background debug commands, not from application software.
- 2. Mass erase FLASH, if necessary.
- 3. Blank check FLASH. Provided FLASH is completely erased, security is disengaged until the next reset.

To avoid returning to secure mode after the next reset, program NVOPT so SEC01:SEC00 = 1:0.

	7	6	5	4	3	2	1	0
R	FPS7	FPS6	FPS5	FPS4	FPS3	FPS2	FPS1	FPDIS
w	Note ⁽¹⁾							

Reset

This register is loaded from nonvolatile location NVPROT during reset.

Figure 4-8. FLASH Protection Register (FPROT)

¹ Background commands can be used to change the contents of these bits in FPROT.

Table 4-10. FPROT Field Descriptions

Field	Description
7:1 FPS[7:1]	FLASH Protect Select Bits — When FPDIS = 0, this 7-bit field determines the ending address of unprotected FLASH locations at the high address end of the FLASH. Protected FLASH locations cannot be erased or programmed.
0 FPDIS	 FLASH Protection Disable 0 FLASH block specified by FPS2:FPS0 is block protected (program and erase not allowed). 1 No FLASH block is protected.

4.6.5 FLASH Status Register (FSTAT)

Bits 3, 1, and 0 always read 0 and writes have no meaning or effect. The remaining five bits are status bits that can be read at any time. Writes to these bits have special meanings that are discussed in the bit descriptions.



Figure 4-9. FLASH Status Register (FSTAT)

Table 4-11. FSTAT Field Descriptions

Field	Description
7 FCBEF	 FLASH Command Buffer Empty Flag — The FCBEF bit is used to launch commands. It also indicates that the command buffer is empty so that a new command sequence can be executed when performing burst programming. The FCBEF bit is cleared by writing a 1 to it or when a burst program command is transferred to the array for programming. Only burst program commands can be buffered. 0 Command buffer is full (not ready for additional commands). 1 A new burst program command may be written to the command buffer.
6 FCCF	 FLASH Command Complete Flag — FCCF is set automatically when the command buffer is empty and no command is being processed. FCCF is cleared automatically when a new command is started (by writing 1 to FCBEF to register a command). Writing to FCCF has no meaning or effect. 0 Command in progress 1 All commands complete



Memory



Resets, Interrupts, and System Configuration

5.7.4 System Options Register (SOPT)

This register may be read at any time. Bits 3 and 2 are unimplemented and always read 0. This is a write-once register so only the first write after reset is honored. Any subsequent attempt to write to SOPT (intentionally or unintentionally) is ignored to avoid accidental changes to these sensitive settings. SOPT should be written during the user's reset initialization program to set the desired controls even if the desired settings are the same as the reset settings.



Figure 5-5. System Options Register (SOPT)

Table 5-5. SOPT Field Descriptions

Field	Description
7 COPE	 COP Watchdog Enable — This write-once bit defaults to 1 after reset. 0 COP watchdog timer disabled. 1 COP watchdog timer enabled (force reset on timeout).
6 COPT	 COP Watchdog Timeout — This write-once bit defaults to 1 after reset. 0 Short timeout period selected (2¹³ cycles of BUSCLK). 1 Long timeout period selected (2¹⁸ cycles of BUSCLK).
5 STOPE	 Stop Mode Enable — This write-once bit defaults to 0 after reset, which disables stop mode. If stop mode is disabled and a user program attempts to execute a STOP instruction, an illegal opcode reset is forced. 0 Stop mode disabled. 1 Stop mode enabled.
1 BKGDPE	 Background Debug Mode Pin Enable — The BKGDPE bit enables the PTG0/BKGD/MS pin to function as BKGD/MS. When the bit is clear, the pin will function as PTG0, which is an output-only general-purpose I/O. This pin always defaults to BKGD/MS function after any reset. 0 BKGD pin disabled. 1 BKGD pin enabled.



1

Resets, Interrupts, and System Configuration

5.7.8 System Power Management Status and Control 2 Register (SPMSC2)

This register is used to report the status of the low voltage warning function, and to configure the stop mode behavior of the MCU.

	7	6	5	4	3	2	1	0
R	LVWF	0			PPDF	0	PDC	
W		LVWACK	LVDV			PPDACK	1 DC	FFDC
Power-on reset:	0 Note ⁽¹⁾	0	0	0	0	0	0	0
LVD reset:	0 Note ⁽¹⁾	0	U	U	0	0	0	0
Any other reset:	0 Note ⁽¹⁾	0	U	U	0	0	0	0
		= Unimplemented or Reserved		ed	U = Unaffected	d by reset		

LVWF will be set in the case when V_{Supply} transitions below the trip point or after reset and V_{Supply} is already below V_{LVW}.

Figure 5-10. System Power Management Status and Control 2 Register (SPMSC2)

Field	Description		
7 LVWF	 Low-Voltage Warning Flag — The LVWF bit indicates the low voltage warning status. 0 Low voltage warning not present. 1 Low voltage warning is present or was present. 		
6 LVWACK	Low-Voltage Warning Acknowledge — The LVWACK bit is the low-voltage warning acknowledge. Writing a 1 to LVWACK clears LVWF to 0 if a low voltage warning is not present.		
5 LVDV	Low-Voltage Detect Voltage Select — The LVDV bit selects the LVD trip point voltage (V _{LVD}). 0 Low trip point selected (V _{LVD} = V _{LVDL}). 1 High trip point selected (V _{LVD} = V _{LVDH}).		
4 LVWV	 Low-Voltage Warning Voltage Select — The LVWV bit selects the LVW trip point voltage (V_{LVW}). 0 Low trip point selected (V_{LVW} = V_{LVWL}). 1 High trip point selected (V_{LVW} = V_{LVWH}). 		
3 PPDF	 Partial Power Down Flag — The PPDF bit indicates that the MCU has exited the stop2 mode. 0 Not stop2 mode recovery. 1 Stop2 mode recovery. 		
2 PPDACK	Partial Power Down Acknowledge — Writing a 1 to PPDACK clears the PPDF bit.		
1 PDC	 Power Down Control — The write-once PDC bit controls entry into the power down (stop2 and stop1) modes. 0 Power down modes are disabled. 1 Power down modes are enabled. 		
0 PPDC	Partial Power Down Control — The write-once PPDC bit controls which power down mode, stop1 or stop2, is selected. 0 Stop1, full power down, mode enabled if PDC set. 1 Stop2, partial power down, mode enabled if PDC set.		

Table 5-11. SPMSC2 Field Descriptions



6.1.2 Block Diagram



NOTES:

- 1. Port pins are software configurable with pullup device if input port.
- 2. Pin contains pullup/pulldown device if IRQ enabled (IRQPE = 1).
- 3. IRQ does not have a clamp diode to VDD. IRQ should not be driven above VDD.
- 4. Pin contains integrated pullup device.
- 5. High current drive
- 6. Pins PTA[7:4] contain both pullup and pulldown devices. Pulldown available when KBI enabled (KBIPn = 1).

Figure 6-1. Block Diagram Highlighting Parallel Input/Output Pins

MC9S08GT16A/GT8A Data Sheet, Rev. 1



Central Processor Unit (S08CPUV2)

8.2 Programmer's Model and CPU Registers

Figure 8-1 shows the five CPU registers. CPU registers are not part of the memory map.



Figure 8-1. CPU Registers

8.2.1 Accumulator (A)

The A accumulator is a general-purpose 8-bit register. One operand input to the arithmetic logic unit (ALU) is connected to the accumulator and the ALU results are often stored into the A accumulator after arithmetic and logical operations. The accumulator can be loaded from memory using various addressing modes to specify the address where the loaded data comes from, or the contents of A can be stored to memory using various addressing modes to specify the addressing modes to specify the addressing modes to specify the address where data from A will be stored.

Reset has no effect on the contents of the A accumulator.

8.2.2 Index Register (H:X)

This 16-bit register is actually two separate 8-bit registers (H and X), which often work together as a 16-bit address pointer where H holds the upper byte of an address and X holds the lower byte of the address. All indexed addressing mode instructions use the full 16-bit value in H:X as an index reference pointer; however, for compatibility with the earlier M68HC05 Family, some instructions operate only on the low-order 8-bit half (X).

Many instructions treat X as a second general-purpose 8-bit register that can be used to hold 8-bit data values. X can be cleared, incremented, decremented, complemented, negated, shifted, or rotated. Transfer instructions allow data to be transferred from A or transferred to A where arithmetic and logical operations can then be performed.

For compatibility with the earlier M68HC05 Family, H is forced to 0x00 during reset. Reset has no effect on the contents of X.



9.3 Register Definition

Refer to the direct-page register summary in the Memory chapter of this data sheet for the absolute address assignments for all ICG registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

9.3.1 ICG Control Register 1 (ICGC1)



Figure 9-6. ICG Control Register 1 (ICGC1)

¹ This bit can be written only once after reset. Additional writes are ignored.

Table 9-1. ICGC1 Register Field Descriptions

Field	Description
7 HGO	 High Gain Oscillator Select — The HGO bit is used to select between low power operation and high gain operation for improved noise immunity. This bit is write-once after reset. O Oscillator configured for low power operation. 1 Oscillator configured for high gain operation.
6 RANGE	 Frequency Range Select — The RANGE bit controls the oscillator, reference divider, and FLL loop prescaler multiplication factor (P). It selects one of two reference frequency ranges for the ICG. The RANGE bit is write-once after a reset. The RANGE bit only has an effect in FLL engaged external and FLL bypassed external modes. O Oscillator configured for low frequency range. FLL loop prescale factor P is 64. 1 Oscillator configured for high frequency range. FLL loop prescale factor P is 1.
5 REFS	 External Reference Select — The REFS bit controls the external reference clock source for ICGERCLK. The REFS bit is write-once after a reset. 0 External clock requested. 1 Oscillator using crystal or resonator requested.
4:3 CLKS	Clock Mode Select — The CLKS bits control the clock mode as described below. If FLL bypassed external is requested, it will not be selected until ERCS = 1. If the ICG enters off mode, the CLKS bits will remain unchanged. Writes to the CLKS bits will not take effect if a previous write is not complete. 00 Self-clocked 01 FLL engaged, internal reference 10 FLL bypassed, external reference 11 FLL engaged, external reference The CLKS bits are writable at any time, unless the first write after a reset was CLKS = 0X, the CLKS bits cannot be written to 1X until after the next reset (because the EXTAL pin was not reserved).

	Clock Reference Source = Internal	Clock Reference Source = External
FLL Engaged	FEI 4 MHz < f _{Bus} < 20 MHz. Medium power (will be less than FEE if oscillator range = high) Good clock accuracy (After IRG is trimmed) <u>Lowest system cost</u> (no external components required) IRG is on. DCO is on. ¹	FEE 4 MHz < f _{Bus} < 20 MHz Medium power (will be less than FEI if oscillator range = low) High clock accuracy Medium/High system cost (crystal, resonator or external clock source required) IRG is off. DCO is on.
FLL Bypassed	SCM This mode is mainly provided for quick and reliable system startup. 3 MHz < f _{Bus} < 5 MHz (default). 3 MHz < f _{Bus} < 20 MHz (via filter bits). Medium power Poor accuracy. IRG is off. DCO is on and open loop.	FBE f _{Bus} range ≤ 8 MHz when crystal or resonator is used. Lowest power Highest clock accuracy Medium/High system cost (Crystal, resonator or external clock source required) IRG is off. DCO is off.

Table 9-10. ICG Configuration Consideration

 1 The IRG typically consumes 100 $\mu A.$ The FLL and DCO typically consumes 0.5 to 2.5 mA, depending upon output frequency. For minimum power consumption and minimum jitter, choose N and R to be as small as possible.

The following sections contain initialization examples for various configurations.

NOTE

Hexadecimal values designated by a preceding \$, binary values designated by a preceding %, and decimal values have no preceding character.

Important configuration information is repeated here for reference.

Table 9-11. ICGOUT Frequency Calculation Options

Clock Scheme	ficgour ¹	Р	Note
SCM — self-clocked mode (FLL bypassed internal)	f _{ICGDCLK} / R	NA	Typical f _{ICGOUT} = 8 MHz immediately after reset
FBE — FLL bypassed external	f _{ext} / R	NA	
FEI — FLL engaged internal	(f _{IRG} / 7)* 64 * N / R	64	Typical f _{IRG} = 243 kHz
FEE — FLL engaged external	f _{ext} * P * N / R	Range = 0 ; P = 64 Range = 1; P = 1	

¹ Ensure that f_{ICGDCLK}, which is equal to f_{ICGOUT} * R, does not exceed f_{ICGDCLKmax}.

MFD Value	Multiplication Factor (N)
000	4
001	6
010	8
011	10
100	12

Table 9-12. MFD and RFD Decode Table

RFD	Division Factor (R)					
000	÷1					
001	÷2					
010	÷4					
011	÷8					
100	÷16					



CLKSB:CLKSA	TPM Clock Source to Prescaler Input
0:0	No clock selected (TPMx disabled)
0:1	Bus rate clock (BUSCLK)
1:0	Fixed system clock (XCLK)
1:1	External source (TPMxCLK) ^{1,2}

Table 10-2. TPM Clock Source Selection

¹ The maximum frequency that is allowed as an external clock is one-fourth of the bus frequency.

² If the external clock input is shared with channel n and is selected as the TPM clock source, the corresponding ELSnB:ELSnA control bits should be set to 0:0 so channel n does not try to use the same pin for a conflicting function.

PS2:PS1:PS0	TPM Clock Source Divided-By
0:0:0	1
0:0:1	2
0:1:0	4
0:1:1	8
1:0:0	16
1:0:1	32
1:1:0	64
1:1:1	128

Table 10-3. Prescale Divisor Selection

10.3.2 Timer x Counter Registers (TPMxCNTH:TPMxCNTL)

The two read-only TPM counter registers contain the high and low bytes of the value in the TPM counter. Reading either byte (TPMxCNTH or TPMxCNTL) latches the contents of both bytes into a buffer where they remain latched until the other byte is read. This allows coherent 16-bit reads in either order. The coherency mechanism is automatically restarted by an MCU reset, a write of any value to TPMxCNTH or TPMxCNTL, or any write to the timer status/control register (TPMxSC).

Reset clears the TPM counter registers.





Chapter 13 Inter-Integrated Circuit (S08IICV1)

13.1 Introduction

The MC9S08GT16A/GT8A series of microcontrollers provides one inter-integrated circuit (IIC) module for communication with other integrated circuits. The two pins associated with this module, SDA and SCL share port C pins 2 and 3, respectively. All functionality as described in this section is available on MC9S08GT16A/GT8A. When the IIC is enabled, the direction of pins is controlled by module configuration. If the IIC is disabled, both pins can be used as general-purpose I/O.





13.4 Functional Description

This section provides a complete functional description of the IIC module.

13.4.1 IIC Protocol

The IIC bus system uses a serial data line (SDA) and a serial clock line (SCL) for data transfer. All devices connected to it must have open drain or open collector outputs. A logic AND function is exercised on both lines with external pull-up resistors. The value of these resistors is system dependent.

Normally, a standard communication is composed of four parts:

- START signal
- Slave address transmission
- Data transfer
- STOP signal

The STOP signal should not be confused with the CPU STOP instruction. The IIC bus system communication is described briefly in the following sections and illustrated in Figure 13-8.



Figure 13-8. IIC Bus Transmission Signals



13.6.1 Byte Transfer Interrupt

The TCF (transfer complete flag) bit is set at the falling edge of the 9th clock to indicate the completion of byte transfer.

13.6.2 Address Detect Interrupt

When the calling address matches the programmed slave address (IIC address register), the IAAS bit in the status register is set. The CPU is interrupted provided the IICIE is set. The CPU must check the SRW bit and set its Tx mode accordingly.

13.6.3 Arbitration Lost Interrupt

The IIC is a true multi-master bus that allows more than one master to be connected on it. If two or more masters try to control the bus at the same time, the relative priority of the contending masters is determined by a data arbitration procedure. The IIC module asserts this interrupt when it loses the data arbitration process and the ARBL bit in the status register is set.

Arbitration is lost in the following circumstances:

- SDA sampled as a low when the master drives a high during an address or data transmit cycle.
- SDA sampled as a low when the master drives a high during the acknowledge bit of a data receive cycle.
- A START cycle is attempted when the bus is busy.
- A repeated START cycle is requested in slave mode.
- A STOP condition is detected when the master did not request it.

This bit must be cleared by software by writing a one to it.



of how straight the line is (how far it deviates from a straight line). The adjusted ideal transition voltage is:

Adjusted Ideal Trans. V =
$$\frac{(\text{Current Code - 1/2})}{2^{N}} * ((V_{\text{REFH}} + E_{\text{FS}}) - (V_{\text{REFL}} + E_{\text{ZS}}))$$

• Zero scale error (E_{ZS}) — This is the difference between the transition voltage to the first valid code and the ideal transition to that code. Normally, it is defined as the difference between the actual and ideal transition to code \$001, but in some cases the first transition may be to a higher code. The ideal transition to any code is:

Eqn. 14-7

Ideal Transition V =
$$\frac{(Current Code - 1/2)}{2^{N}} * (V_{REFH} - V_{REFL})$$

• Full scale error (E_{FS}) — This is the difference between the transition voltage to the last valid code and the ideal transition to that code. Normally, it is defined as the difference between the actual and ideal transition to code \$3FF, but in some cases the last transition may be to a lower code. The ideal transition to any code is:

Eqn. 14-8

Ideal Transition V =
$$\frac{(\text{Current Code - 1/2})}{2^{\text{N}}} * (\text{V}_{\text{REFH}} - \text{V}_{\text{REFL}})$$

- Total unadjusted error (E_{TU}) This is the difference between the transition voltage to a given code and the ideal straight-line transfer function. An alternate definition (with the same result) is the difference between the actual transfer function and the ideal straight-line transfer function. This measure of error includes inherent quantization error and all forms of circuit error (INL, DNL, zero-scale, and full-scale) except input leakage error, which is not due to the ATD.
- Input leakage error (E_{IL}) This is the error between the transition voltage to the current code and the ideal transition to that code that is the result of input leakage across the real portion of the impedance of the network that drives the analog input. This error is a system-observable error which is not inherent to the ATD, so it is not added to total error. This error is:

E_{IL} (in V) = input leakage * R_{AS} Eqn. 14-9

There are two other forms of error which are not specified which can also affect ATD accuracy. These are:

- Sampling error (E_S) The error due to inadequate time to charge the ATD circuitry
- Noise error (E_N) The error due to noise on V_{AIN}, V_{REFH}, or V_{REFL} due to either direct coupling (noise source capacitively coupled directly on the signal) or power supply (V_{DDAD}, V_{SSAD}, V_{DD}, and V_{SS}) noise interfering with the ATD's ability to resolve the input accurately. The error due to internal sources can be reduced (and specified operation achieved) by operating the ATD conversion in wait mode and ceasing all IO activity. Reducing the error due to external sources is dependent on system activity and board layout.



14.5 Resets

The ATD module is reset on system reset. If the system reset signal is activated, the ATD registers are initialized back to their reset state and the ATD module is powered down. This occurs as a function of the register file initialization; the reset definition of the ATDPU bit (power down bit) is zero or disabled.

The MCU places the module back into an initialized state. If the module is performing a conversion, the current conversion is terminated, the conversion complete flag is cleared, and the SAR register bits are cleared. Any pending interrupts are also cancelled. Note that the control, test, and status registers are initialized on reset; the initialized register state is defined in the register description section of this specification.

Enabling the module (using the ATDPU bit) does not cause the module to reset since the register file is not initialized. Finally, writing to control register ATDC does not cause the module to reset; the current conversion will be terminated.

14.6 Interrupts

The ATD module originates interrupt requests and the MCU handles or services these requests. Details on how the ATD interrupt requests are handled can be found in the resets and interrupts chapter of this data sheet.

The ATD interrupt function is enabled by setting the ATDIE bit in the ATDSC register. When the ATDIE bit is set, an interrupt is generated at the end of an ATD conversion and the ATD result registers (ATDRH and ATDRL) contain the result data generated by the conversion. If the interrupt function is disabled (ATDIE = 0), then the CCF flag must be polled to determine when a conversion is complete.

The interrupt will remain pending as long as the CCF flag is set. The CCF bit is cleared whenever the ATD status and control (ATDSC) register is written. The CCF bit is also cleared whenever the ATD result registers (ATDRH or ATDRL) are read.

Interrupt	Local Enable	Description	
CCF	ATDIE	Conversion complete	

Table 14-8. Interrupt Summary



Analog-to-Digital Converter (S08ATDV3)



Figure 15-4 shows the host receiving a logic 0 from the target HCS08 MCU. Because the host is asynchronous to the target MCU, there is a 0-to-1 cycle delay from the host-generated falling edge on BKGD to the start of the bit time as perceived by the target MCU. The host initiates the bit time but the target HCS08 finishes it. Because the target wants the host to receive a logic 0, it drives the BKGD pin low for 13 BDC clock cycles, then briefly drives it high to speed up the rising edge. The host samples the bit level about 10 cycles after starting the bit time.



Figure 15-4. BDM Target-to-Host Serial Bit Timing (Logic 0)



15.4.3.8 Debug Trigger Register (DBGT)

This register can be read any time, but may be written only if ARM = 0, except bits 4 and 5 are hard-wired to 0s.



Figure 15-8. Debug Trigger Register (DBGT)

Table 15-5. DBGT Register Field Descriptions

Field	Description
7 TRGSEL	 Trigger Type — Controls whether the match outputs from comparators A and B are qualified with the opcode tracking logic in the debug module. If TRGSEL is set, a match signal from comparator A or B must propagate through the opcode tracking logic and a trigger event is only signalled to the FIFO logic if the opcode at the match address is actually executed. 0 Trigger on access to compare address (force) 1 Trigger if opcode at compare address is executed (tag)
6 BEGIN	 Begin/End Trigger Select — Controls whether the FIFO starts filling at a trigger or fills in a circular manner until a trigger ends the capture of information. In event-only trigger modes, this bit is ignored and all debug runs are assumed to be begin traces. Data stored in FIFO until trigger (end trace) Trigger initiates data storage (begin trace)
3:0 TRG[3:0]	Select Trigger Mode — Selects one of nine triggering modes, as described below. 0000 A-only 0001 A OR B 0010 A Then B 0011 Event-only B (store data) 0100 A then event-only B (store data) 0101 A AND B data (full mode) 0110 A AND NOT B data (full mode) 0111 Inside range: A \leq address \leq B 1000 Outside range: address < A or address > B $1001 - 1111$ (No trigger)



Electrical Characteristics

A.9 Internal Clock Generation Module Characteristics



Table A-10. ICG DC Electrical Specifications (Temperature Range = -40 to 125°C Ambient)

Characteristic	Symbol	Min	Typ ¹	Max	Unit
Load capacitors	C ₁ C ₂	See Note ²			
Feedback resistor Low range (32k to 100 kHz) High range (1M – 16 MHz)	R _F		10 1		ΜΩ ΜΩ
Series resistor Low range Low Gain (HGO = 0) High Gain (HGO = 1) High range Low Gain (HGO = 0) High Gain (HGO = 1) ≥ 8 MHz 4 MHz 1 MHz	R _S	 	0 100 0 10 20		kΩ

¹ Data in Typical column was characterized at 3.0 V, 25°C or is typical recommended value.

² See crystal or resonator manufacturer's recommendation.





DETAIL M PIN 1 BACKSIDE IDENTIFIER OPTION

DETAIL M PIN 1 BACKSIDE IDENTIFIER OPTION



DETAIL T

© FREESCALE SEMICONDUCTOR, INC. All RIGHTS RESERVED.	MECHANICA	LOUTLINE	PRINT VERSION NO	T TO SCALE
TITLE: THERMALLY ENHANCED	DOCUMENT NO): 98ARH99048A	REV: F	
FLAT NON-LEADED PACKA	CASE NUMBER	: 1314–05	05 DEC 2005	
48 TERMINAL, 0.5 PITCH (7	STANDARD: JE	DEC-MO-220 VKKD-2	2	