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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	32MHz
Connectivity	I ² C, IrDA, LINbus, SPI, UART/USART, USB
Peripherals	Brown-out Detect/Reset, Cap Sense, DMA, I ² S, POR, PWM, WDT
Number of I/O	83
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	4K x 8
RAM Size	16K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 24x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32l151vbt6tr

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Table 5. Working mode-dependent functionalities (from Run/active down to standby)

Ips	Run/Active	Sleep	Low-power Run	Low-power Sleep	Stop		Standby	
						Wakeup capability		Wakeup capability
CPU	Y	-	Y	-	-	-	-	-
Flash	Y	Y	Y	Y	-	-	-	-
RAM	Y	Y	Y	Y	Y	-	-	-
Backup Registers	Y	Y	Y	Y	Y	-	Y	-
EEPROM	Y	-	Y	Y	Y	-	-	-
Brown-out rest (BOR)	Y	Y	Y	Y	Y	Y	Y	-
DMA	Y	Y	Y	Y	-	-	-	-
Programmable Voltage Detector (PVD)	Y	Y	Y	Y	Y	Y	Y	-
Power On Reset (POR)	Y	Y	Y	Y	Y	Y	Y	-
Power Down Rest (PDR)	Y	Y	Y	Y	Y	-	Y	-
High Speed Internal (HSI)	Y	Y	-	-	-	-	-	-
High Speed External (HSE)	Y	Y	-	-	-	-	-	-
Low Speed Internal (LSI)	Y	Y	Y	Y	Y	-	Y	-
Low Speed External (LSE)	Y	Y	Y	Y	Y	-	Y	-
Multi-Speed Internal (MSI)	Y	Y	Y	Y	-	-	-	-
Inter-Connect Controller	Y	Y	Y	Y	-	-	-	-
RTC	Y	Y	Y	Y	Y	Y	Y	-
RTC Tamper	Y	Y	Y	Y	Y	Y	Y	Y
Auto Wakeup (AWU)	Y	Y	Y	Y	Y	Y	Y	Y
LCD	Y	Y	Y	Y	Y	-	-	-
USB	Y	Y	-	-	-	Y	-	-
USART	Y	Y	Y	Y	Y	(1)	-	-
SPI	Y	Y	Y	Y	-	-	-	-
I2C	Y	Y	Y	Y	-	(1)	-	-
ADC	Y	Y	-	-	-	-	-	-

Nested vectored interrupt controller (NVIC)

The ultra-low-power STM32L151x6/8/B and STM32L152x6/8/B devices embed a nested vectored interrupt controller able to handle up to 45 maskable interrupt channels (not including the 16 interrupt lines of Cortex®-M3) and 16 priority levels.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of *late arriving*, higher-priority interrupts
- Support for tail-chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

3.3 Reset and supply management

3.3.1 Power supply schemes

- $V_{DD} = 1.65$ to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through V_{DD} pins.
- V_{SSA} , $V_{DDA} = 1.65$ to 3.6 V: external analog power supplies for ADC, reset blocks, RCs and PLL (minimum voltage to be applied to V_{DDA} is 1.8 V when the ADC is used). V_{DDA} and V_{SSA} must be connected to V_{DD} and V_{SS} , respectively.

3.3.2 Power supply supervisor

The device has an integrated ZEROPOWER power-on reset (POR)/power-down reset (PDR) that can be coupled with a brownout reset (BOR) circuitry.

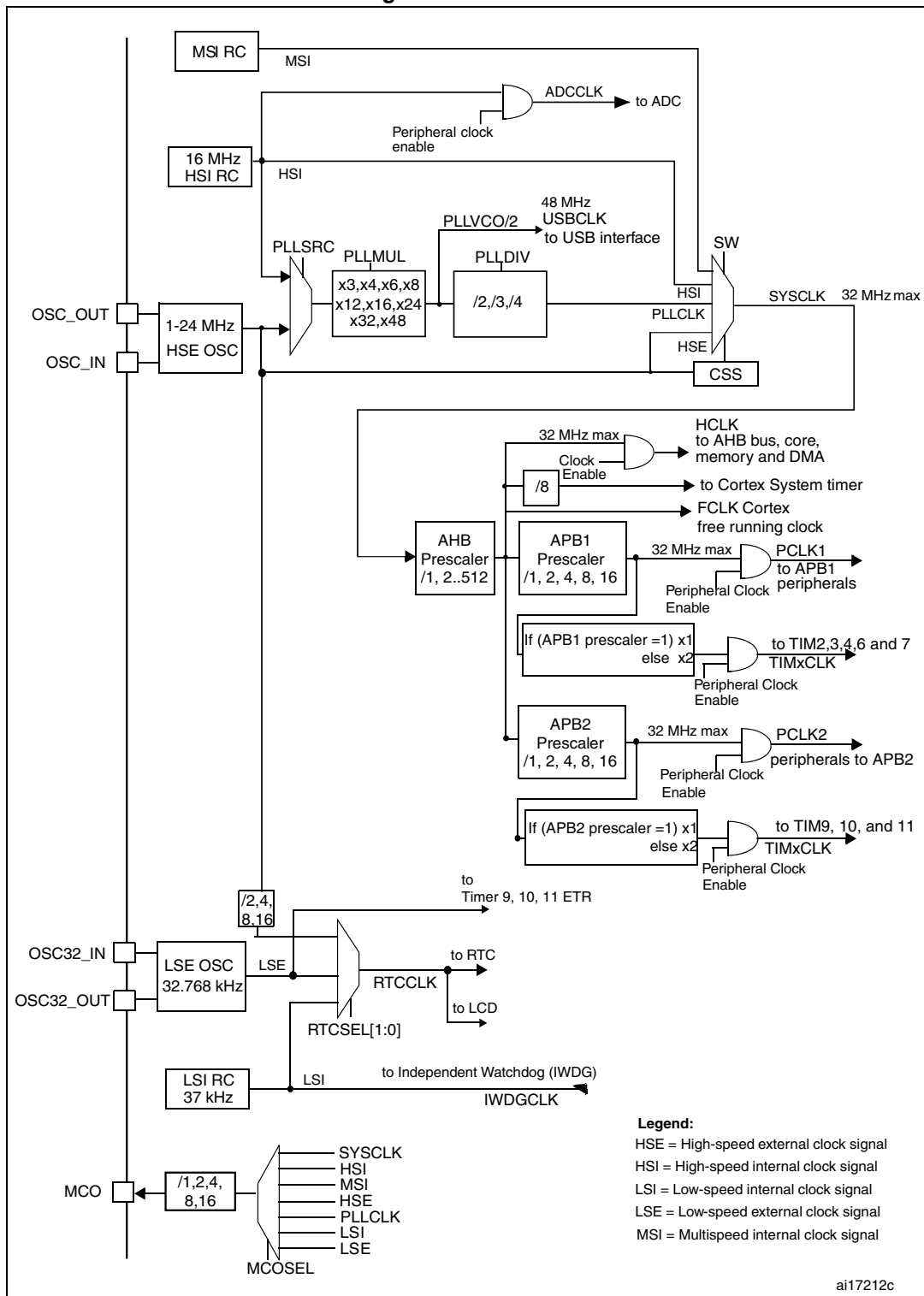
The device exists in two versions:

- The version with BOR activated at power-on operates between 1.8 V and 3.6 V.
- The other version without BOR operates between 1.65 V and 3.6 V.

After the V_{DD} threshold is reached (1.65 V or 1.8 V depending on the BOR which is active or not at power-on), the option byte loading process starts, either to confirm or modify default thresholds, or to disable the BOR permanently: in this case, the V_{DD} min value becomes 1.65 V (whatever the version, BOR active or not, at power-on).

When BOR is active at power-on, it ensures proper operation starting from 1.8 V whatever the power ramp-up phase before it reaches 1.8 V. When BOR is not active at power-up, the power ramp-up should guarantee that 1.65 V is reached on V_{DD} at least 1 ms after it exits the POR area.

Figure 2. Clock tree



3.10 ADC (analog-to-digital converter)

A 12-bit analog-to-digital converters is embedded into STM32L151x6/8/B and STM32L152x6/8/B devices with up to 24 external channels, performing conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start trigger and injection trigger, to allow the application to synchronize A/D conversions and timers. An injection mode allows high priority conversions to be done by interrupting a scan mode which runs in as a background task.

The ADC includes a specific low power mode. The converter is able to operate at maximum speed even if the CPU is operating at a very low frequency and has an auto-shutdown function. The ADC's runtime and analog front-end current consumption are thus minimized whatever the MCU operating mode.

3.10.1 Temperature sensor

The temperature sensor (TS) generates a voltage V_{SENSE} that varies linearly with temperature.

The temperature sensor is internally connected to the ADC_IN16 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode, see [Table 58: Temperature sensor calibration values](#).

3.10.2 Internal voltage reference (V_{REFINT})

The internal voltage reference (V_{REFINT}) provides a stable (bandgap) voltage output for the ADC and Comparators. V_{REFINT} is internally connected to the ADC_IN17 input channel. It enables accurate monitoring of the V_{DD} value (when no external voltage, $V_{\text{REF+}}$, is available for ADC). The precise voltage of V_{REFINT} is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode see [Table 16: Embedded internal reference voltage](#).

3.11 DAC (digital-to-analog converter)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in non-inverting configuration.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channels' independent or simultaneous conversions
- DMA capability for each channel (including the underrun interrupt)
- external triggers for conversion
- input reference voltage V_{REF+}

Eight DAC trigger inputs are used in the STM32L151x6/8/B and STM32L152x6/8/B devices. The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

3.12 Ultra-low-power comparators and reference voltage

The STM32L151x6/8/B and STM32L152x6/8/B devices embed two comparators sharing the same current bias and reference voltage. The reference voltage can be internal or external (coming from an I/O).

- one comparator with fixed threshold
- one comparator with rail-to-rail inputs, fast or slow mode. The threshold can be one of the following:
 - DAC output
 - External I/O
 - Internal reference voltage (V_{REFINT}) or V_{REFINT} submultiple (1/4, 1/2, 3/4)

Both comparators can wake up from Stop mode, and be combined into a window comparator.

The internal reference voltage is available externally via a low power / low current output buffer (driving current capability of 1 μ A typical).

3.13 Routing interface

This interface controls the internal routing of I/Os to TIM2, TIM3, TIM4 and to the comparator and reference voltage output.

3.14 Touch sensing

The STM32L151x6/8/B and STM32L152x6/8/B devices provide a simple solution for adding capacitive sensing functionality to any application. These devices offer up to 20 capacitive sensing channels distributed over 10 analog I/O groups. Only software capacitive sensing acquisition mode is supported.

Capacitive sensing technology is able to detect the presence of a finger near a sensor which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven

Table 8. STM32L151x6/8/B and STM32L152x6/8/B pin definitions (continued)

Pins					Pin name	Pin type ⁽¹⁾	I/O structure	Main function ⁽²⁾ (after reset)	Pins functions	Additional functions
LQFP100	LQFP64	TFBGA64	UFBGA100	LQFP48 or UFQFPN48					Alternate functions	
71	45	B8	A12	33	PA12	I/O	FT	PA12	USART1_RTS/ SPI1_MOSI	USB_DP
72	46	A8	A11	34	PA13	I/O	FT	JTMS- SWDIO	JTMS-SWDIO	-
73	-	-	C11	-	PH2	I/O	FT	PH2	-	-
74	47	D5	F11	35	V _{SS_2}	S	-	V _{SS_2}	-	-
75	48	E5	G11	36	V _{DD_2}	S	-	V _{DD_2}	-	-
76	49	A7	A10	37	PA14	I/O	FT	JTCK- SWCLK	JTCK-SWCLK	-
77	50	A6	A9	38	PA15	I/O	FT	JTDI	TIM2_CH1_ETR/PA15/ SPI1_NSS/ LCD_SEG17	-
78	51	B7	B11	-	PC10	I/O	FT	PC10	USART3_TX/LCD_SEG28 /LCD_SEG40/LCD_COM4	-
79	52	B6	C10	-	PC11	I/O	FT	PC11	USART3_RX/LCD_SEG29 /LCD_SEG41/LCD_COM5	-
80	53	C5	B10	-	PC12	I/O	FT	PC12	USART3_CK/LCD_SEG30 /LCD_SEG42/LCD_COM6	-
81	-	-	C9	-	PD0	I/O	FT	PD0	SPI2_NSS/TIM9_CH1	-
82	-	-	B9	-	PD1	I/O	FT	PD1	SPI2_SCK	-
83	54	B5	C8	-	PD2	I/O	FT	PD2	TIM3_ETR/LCD_SEG31/ LCD_SEG43/LCD_COM7	-
84	-	-	B8	-	PD3	I/O	FT	PD3	USART2_CTS/ SPI2_MISO	-
85	-	-	B7	-	PD4	I/O	FT	PD4	USART2_RTS/ SPI2_MOSI	-
86	-	-	A6	-	PD5	I/O	FT	PD5	USART2_TX	-
87	-	-	B6	-	PD6	I/O	FT	PD6	USART2_RX	-
88	-	-	A5	-	PD7	I/O	FT	PD7	USART2_CK/TIM9_CH2	-
89	55	A5	A8	39	PB3	I/O	FT	JTDO	TIM2_CH2/PB3/ SPI1_SCK/LCD_SEG7/ JTDO	COMP2_INM

6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in [Table 10: Voltage characteristics](#), [Table 11: Current characteristics](#), and [Table 12: Thermal characteristics](#) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 10. Voltage characteristics

Symbol	Ratings	Min	Max	Unit
$V_{DD}-V_{SS}$	External main supply voltage (including V_{DDA} and V_{DD}) ⁽¹⁾	-0.3	4.0	V
V_{IN} ⁽²⁾	Input voltage on five-volt tolerant pin	$V_{SS}-0.3$	$V_{DD}+4.0$	
	Input voltage on any other pin	$V_{SS}-0.3$	4.0	
$ \Delta V_{DDx} $	Variations between different V_{DD} power pins	-	50	mV
$ V_{SSx}-V_{SS} $	Variations between all different ground pins ⁽³⁾	-	50	
$V_{REF+}-V_{DDA}$	Allowed voltage difference for $V_{REF+} > V_{DDA}$	-	0.4	V
$V_{ESD(HBM)}$	Electrostatic discharge voltage (human body model)	see Section 6.3.11		-

1. All main power (V_{DD} , V_{DDA}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. V_{IN} maximum must always be respected. Refer to [Table 11](#) for maximum allowed injected current values.
3. Include V_{REF-} pin.

Table 11. Current characteristics

Symbol	Ratings	Max.	Unit
$I_{VDD\Sigma}$	Total current into V_{DD}/V_{DDA} power lines (source) ⁽¹⁾	80	mA
$I_{VSS\Sigma}$	Total current out of V_{SS} ground lines (sink) ⁽¹⁾	80	
I_{IO}	Output current sunk by any I/O and control pin	25	
	Output current sourced by any I/O and control pin	- 25	
$I_{INJ(PIN)}$ ⁽²⁾	Injected current on five-volt tolerant I/O ⁽³⁾	-5/+0	
	Injected current on any other pin ⁽⁴⁾	± 5	
$\Sigma I_{INJ(PIN)}$	Total injected current (sum of all I/O and control pins) ⁽⁵⁾	± 25	

1. All main power (V_{DD} , V_{DDA}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. Negative injection disturbs the analog performance of the device. See note in [Section 6.3.17](#).
3. Positive current injection is not possible on these I/Os. A negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer to [Table 10](#) for maximum allowed input voltage values.
4. A positive injection is induced by $V_{IN} > V_{DD}$ while a negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer to [Table 10: Voltage characteristics](#) for the maximum allowed input voltage values.
5. When several inputs are submitted to a current injection, the maximum $\Sigma I_{INJ(PIN)}$ is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 18. Current consumption in Run mode, code with data processing running from RAM

Symbol	Parameter	Conditions		f _{HCLK}	Typ	Max ⁽¹⁾			Unit
						55 °C	85 °C	105 °C	
I _{DD} (Run from RAM)	Supply current in Run mode, code executed from RAM, Flash switched off	f _{HSE} = f _{HCLK} up to 16 MHz, included f _{HSE} = f _{HCLK} /2 above 16 MHz (PLL ON) ⁽²⁾	Range 3, V _{CORE} =1.2 V VOS[1:0] = 11	1 MHz	200	300	300	300	μA
				2 MHz	380	500	500	500	
				4 MHz	720	860	860	860 ⁽³⁾	
			Range 2, V _{CORE} =1.5 V VOS[1:0] = 10	4 MHz	0.9	1	1	1	mA
				8 MHz	1.65	2	2	2	
				16 MHz	3.2	3.7	3.7	3.7	
			Range 1, V _{CORE} =1.8 V VOS[1:0] = 01	8 MHz	2	2.5	2.5	2.5	
				16 MHz	4	4.5	4.5	4.5	
				32 MHz	7.7	8.5	8.5	8.5	
		HSI clock source (16 MHz)	Range 2, V _{CORE} =1.5 V VOS[1:0] = 10	16 MHz	3.3	3.8	3.8	3.8	
			Range 1, V _{CORE} =1.8 V VOS[1:0] = 01	32 MHz	7.8	9.2	9.2	9.2	
		MSI clock, 65 kHz	Range 3, V _{CORE} =1.2 V VOS[1:0] = 11	65 kHz	40	60	60	80	μA
		MSI clock, 524 kHz		524 kHz	110	140	140	160	
		MSI clock, 4.2 MHz		4.2 MHz	700	800	800	820	

1. Guaranteed by characterization results, unless otherwise specified.

2. Oscillator bypassed (HSEBYP = 1 in RCC_CR register).

3. Tested in production.

Table 24. Peripheral current consumption⁽¹⁾ (continued)

Peripheral		Typical consumption, V _{DD} = 3.0 V, T _A = 25 °C				Unit
		Range 1, V _{CORE} =1.8 V VOS[1:0] = 01	Range 2, V _{CORE} =1.5 V VOS[1:0] = 10	Range 3, V _{CORE} =1.2 V VOS[1:0] = 11	Low power sleep and run	
AHB	GPIOA	5	4.5	3.5	4	μA/MHz (f _{HCLK})
	GPIOB	5	4.5	3.5	4.5	
	GPIOC	5	4.5	3.5	4.5	
	GPIOD	5	4.5	3.5	4.5	
	GPIOE	5	4.5	3.5	4.5	
	GPIOH	4	4	3	3.5	
	CRC	1	0.5	0.5	0.5	
	FLASH	13	11.5	9	18.5	
	DMA1	12	10	8	10.5	
All enabled		166	138	106	130	
I _{DD} (RTC)		0.47				μA
I _{DD} (LCD)		3.1				
I _{DD} (ADC) ⁽³⁾		1450				
I _{DD} (DAC) ⁽⁴⁾		340				
I _{DD} (COMP1)		0.16				
I _{DD} (COMP2)	Slow mode	2				
	Fast mode	5				
I _{DD} (PVD / BOR) ⁽⁵⁾		2.6				
I _{DD} (IWDG)		0.25				

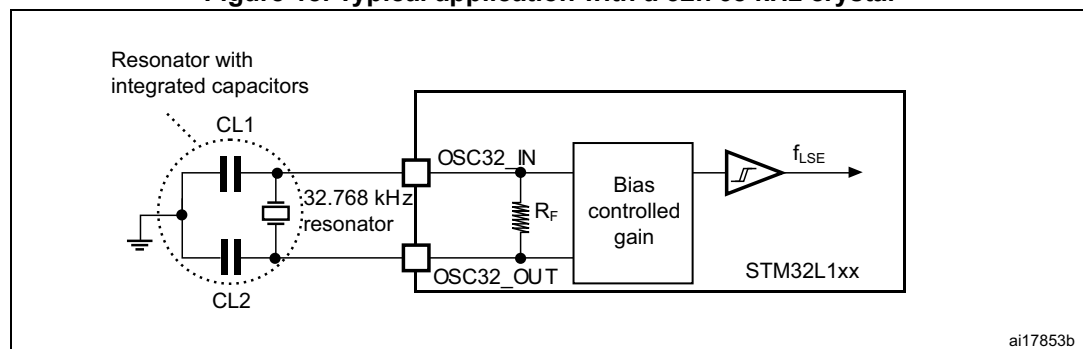
1. Data based on differential I_{DD} measurement between all peripherals OFF and one peripheral with clock enabled, in the following conditions: f_{HCLK} = 32 MHz (Range 1), f_{HCLK} = 16 MHz (Range 2), f_{HCLK} = 4 MHz (Range 3), f_{HCLK} = 64kHz (Low power run/sleep), f_{APB1} = f_{HCLK}, f_{APB2} = f_{HCLK}, default prescaler value for each peripheral. The CPU is in Sleep mode in both cases. No I/O pins toggling.
2. HSI oscillator is OFF for this measure.
3. Data based on a differential I_{DD} measurement between ADC in reset configuration and continuous ADC conversion (HSI consumption not included).
4. Data based on a differential I_{DD} measurement between DAC in reset configuration and continuous DAC conversion of V_{DD}/2. DAC is in buffered mode, output is left floating.
5. Including supply current of internal reference voltage.

4. $t_{SU(LSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

Note: For CL1 and CL2, it is recommended to use high-quality ceramic capacitors in the 5 pF to 15 pF range selected to match the requirements of the crystal or resonator (see [Figure 18](#)). CL1 and CL2, are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of CL1 and CL2. Load capacitance CL has the following formula: $CL = CL1 \times CL2 / (CL1 + CL2) + C_{stray}$ where C_{stray} is the pin capacitance and board or trace PCB-related capacitance. Typically, it is between 2 pF and 7 pF.

Caution: To avoid exceeding the maximum value of CL1 and CL2 (15 pF) it is strongly recommended to use a resonator with a load capacitance $CL \leq 7$ pF. Never use a resonator with a load capacitance of 12.5 pF. Example: if a resonator is chosen with a load capacitance of $CL = 6$ pF and $C_{stray} = 2$ pF, then $CL1 = CL2 = 8$ pF.

Figure 18. Typical application with a 32.768 kHz crystal



Multi-speed internal (MSI) RC oscillator

Table 32. MSI oscillator characteristics

Symbol	Parameter	Condition	Typ	Max	Unit
f_{MSI}	Frequency after factory calibration, done at $V_{\text{DD}} = 3.3 \text{ V}$ and $T_{\text{A}} = 25 \text{ }^{\circ}\text{C}$	MSI range 0	65.5	-	kHz
		MSI range 1	131	-	
		MSI range 2	262	-	
		MSI range 3	524	-	
		MSI range 4	1.05	-	MHz
		MSI range 5	2.1	-	
		MSI range 6	4.2	-	
ACC_{MSI}	Frequency error after factory calibration	-	± 0.5	-	%
$D_{\text{TEMP}(\text{MSI})}^{(1)}$	MSI oscillator frequency drift $0 \text{ }^{\circ}\text{C} \leq T_{\text{A}} \leq 85 \text{ }^{\circ}\text{C}$	-	± 3	-	%
$D_{\text{VOLT}(\text{MSI})}^{(1)}$	MSI oscillator frequency drift $1.65 \text{ V} \leq V_{\text{DD}} \leq 3.6 \text{ V}$, $T_{\text{A}} = 25 \text{ }^{\circ}\text{C}$	-	-	2.5	%/V
$I_{\text{DD}(\text{MSI})}^{(2)}$	MSI oscillator power consumption	MSI range 0	0.75	-	μA
		MSI range 1	1	-	
		MSI range 2	1.5	-	
		MSI range 3	2.5	-	
		MSI range 4	4.5	-	
		MSI range 5	8	-	
		MSI range 6	15	-	
$t_{\text{SU}(\text{MSI})}$	MSI oscillator startup time	MSI range 0	30	-	μs
		MSI range 1	20	-	
		MSI range 2	15	-	
		MSI range 3	10	-	
		MSI range 4	6	-	
		MSI range 5	5	-	
		MSI range 6, Voltage range 1 and 2	3.5	-	
		MSI range 6, Voltage range 3	5	-	

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 38. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. frequency range			Unit
				4 MHz voltage Range 3	16 MHz voltage Range 2	32 MHz voltage Range 1	
S _{EMI}	Peak level	V _{DD} = 3.3 V, T _A = 25 °C, LQFP100 package compliant with IEC 61967-2	0.1 to 30 MHz	3	-6	-5	dBμV
			30 to 130 MHz	18	4	-7	
			130 MHz to 1GHz	15	5	-7	
			SAE EMI Level	2.5	2	1	-

6.3.11 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

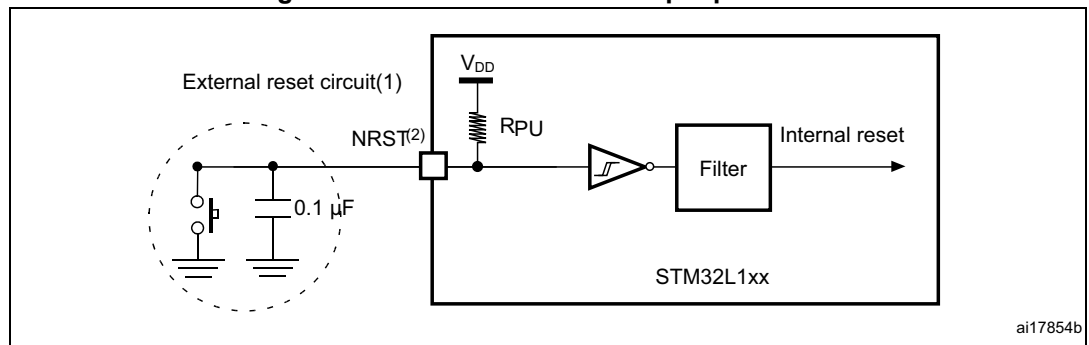
Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 39. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Packages	Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T _A = +25 °C, conforming to JESD22-A114	All	2	2000	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C, conforming to JESD22-C101	All	III	500	

1. Guaranteed by characterization results.

Figure 20. Recommended NRST pin protection

1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)}$ max level specified in [Table 45](#). Otherwise the reset will not be taken into account by the device.

6.3.15 TIM timer characteristics

The parameters given in [Table 46](#) are guaranteed by design.

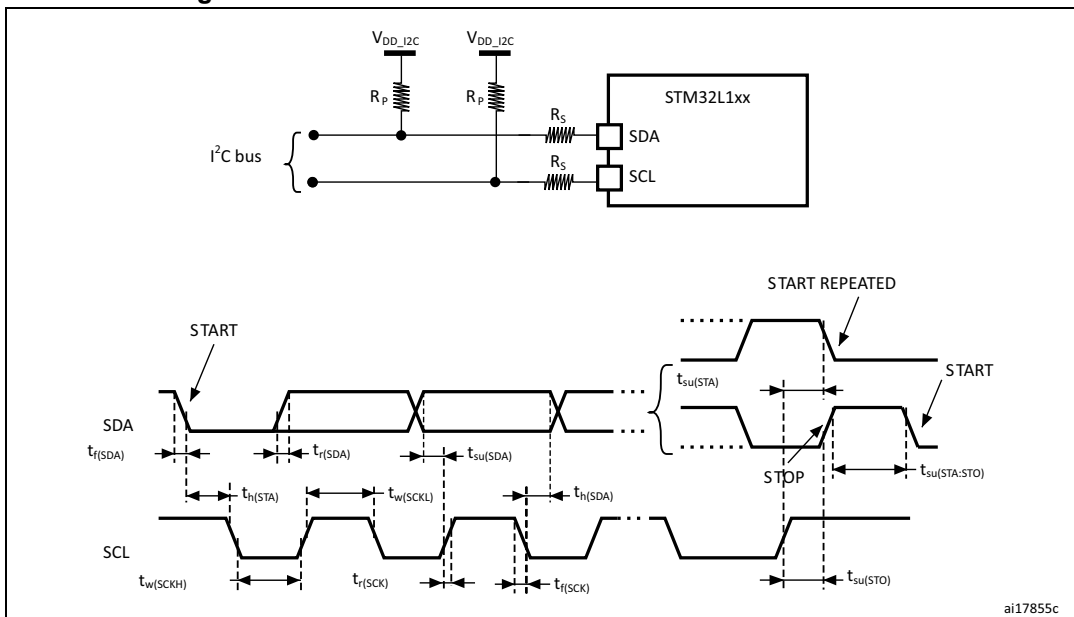
Refer to [Section 6.3.13: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 46. TIMx⁽¹⁾ characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{res(TIM)}$	Timer resolution time	-	1	-	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 32 \text{ MHz}$	31.25	-	ns
f_{EXT}	Timer external clock frequency on CH1 to CH4	-	0	$f_{TIMxCLK}/2$	MHz
		$f_{TIMxCLK} = 32 \text{ MHz}$	0	16	MHz
Res_{TIM}	Timer resolution	-	-	16	bit
$t_{COUNTER}$	16-bit counter clock period when internal clock is selected (timer's prescaler disabled)	-	1	65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 32 \text{ MHz}$	0.0312	2048	μs
t_{MAX_COUNT}	Maximum possible count	-	-	65536×65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 32 \text{ MHz}$	-	134.2	s

1. TIMx is used as a general term to refer to the TIM2, TIM3 and TIM4 timers.

Figure 21. I²C bus AC waveforms and measurement circuit



1. R_S = series protection resistors
2. R_P = pull-up resistors
3. V_{DD_I2C} = I2C bus supply
4. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

Table 48. SCL frequency ($f_{PCLK1} = 32 \text{ MHz}$, $V_{DD} = V_{DD \text{ I2C}} = 3.3 \text{ V}$)⁽¹⁾⁽²⁾

f _{SCL} (kHz)	I2C_CCR value
	R _P = 4.7 kΩ
400	0x801B
300	0x8024
200	0x8035
100	0x00A0
50	0x0140
20	0x0320

1. R_P = External pull-up resistance, f_{SCL} = I^2C speed.
2. For speeds around 200 kHz, the tolerance on the achieved speed is of $\pm 5\%$. For other speed ranges, the tolerance on the achieved speed is $\pm 2\%$. These variations depend on the accuracy of the external components used to design the application.

Figure 26. ADC accuracy characteristics

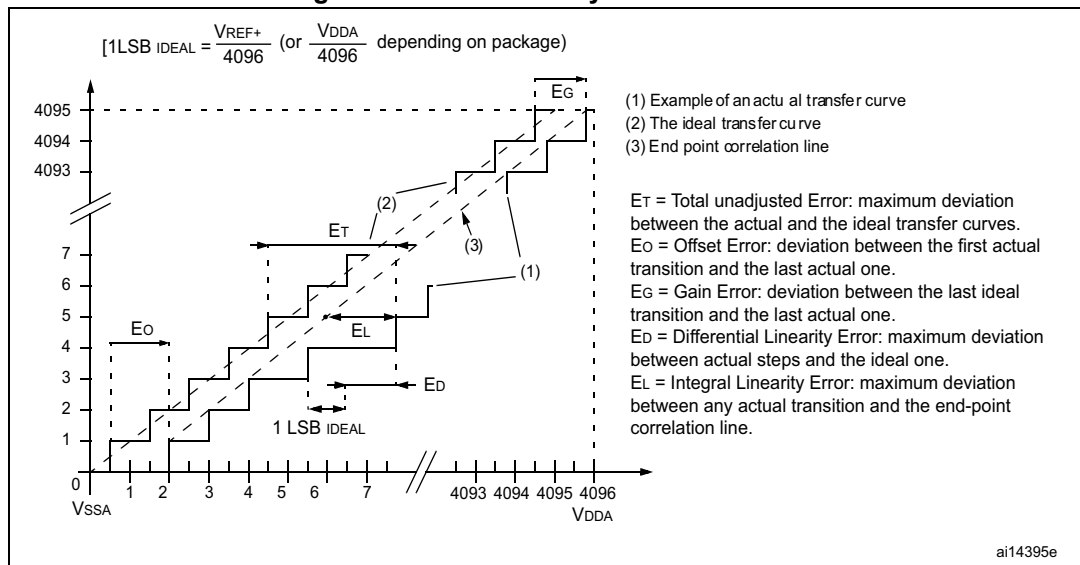
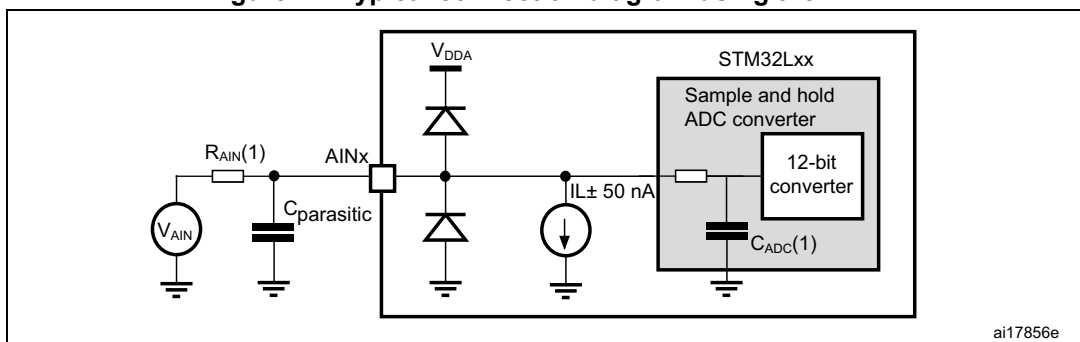


Figure 27. Typical connection diagram using the ADC



1. Refer to [Table 56: Maximum source impedance \$R_{\text{AIN max}}\$](#) for the value of R_{AIN} and [Table 54: ADC characteristics](#) for the value of C_{ADC}
2. $C_{\text{parasitic}}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high $C_{\text{parasitic}}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

6.3.18 DAC electrical specifications

Data guaranteed by design, unless otherwise specified.

Table 57. DAC characteristics

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V _{DDA}	Analog supply voltage	-		1.8	-	3.6	V
V _{REF+}	Reference supply voltage	V _{REF+} must always be below V _{DDA}		1.8	-	3.6	V
V _{REF-}	Lower reference voltage	-		V _{SSA}			V
I _{DDVREF+} ⁽¹⁾	Current consumption on V _{REF+} supply V _{REF+} = 3.3 V	No load, middle code (0x800)		-	130	220	μA
		No load, worst code (0x000)		-	220	350	μA
I _{DDA} ⁽¹⁾	Current consumption on V _{DDA} supply V _{DDA} = 3.3 V	No load, middle code (0x800)		-	210	320	μA
		No load, worst code (0xF1C)		-	320	520	μA
R _L	Resistive load	DAC output buffer ON	Connected to V _{SSA}	5	-	-	kΩ
			Connected to V _{DDA}	25	-	-	
C _L	Capacitive load	DAC output buffer ON		-	-	50	pF
R _O	Output impedance	DAC output buffer OFF		12	16	20	kΩ
V _{DAC_OUT}	Voltage on DAC_OUT output	DAC output buffer ON		0.2	-	V _{DDA} − 0.2	V
		DAC output buffer OFF		0.5	-	V _{REF+} − 1LSB	mV
DNL ⁽¹⁾	Differential non linearity ⁽²⁾	C _L ≤ 50 pF, R _L ≥ 5 kΩ DAC output buffer ON		-	1.5	3	LSB
		No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF		-	1.5	3	
INL ⁽¹⁾	Integral non linearity ⁽³⁾	C _L ≤ 50 pF, R _L ≥ 5 kΩ DAC output buffer ON		-	2	4	
		No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF		-	2	4	
Offset ⁽¹⁾	Offset error at code 0x800 ⁽⁴⁾	C _L ≤ 50 pF, R _L ≥ 5 kΩ DAC output buffer ON		-	±10	±25	
		No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF		-	±5	±8	
Offset1 ⁽¹⁾	Offset error at code 0x001 ⁽⁵⁾	No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF		-	±1.5	±5	

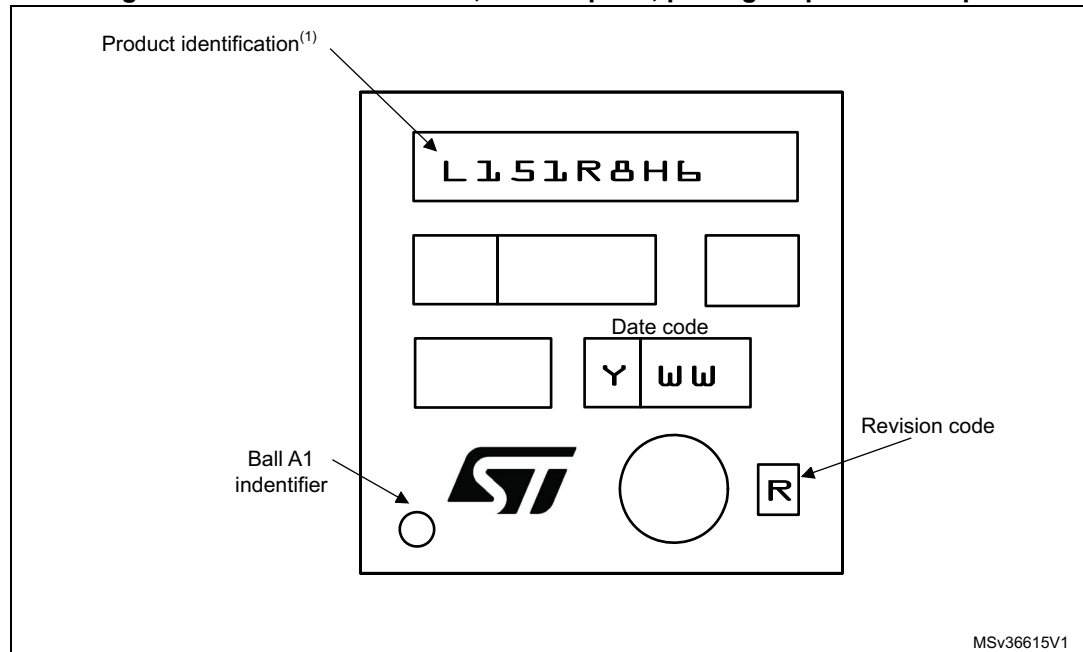
Table 57. DAC characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
dOffset/dT ⁽¹⁾	Offset error temperature coefficient (code 0x800)	V _{DDA} = 3.3V, T _A = 0 to 50 °C DAC output buffer OFF	-20	-10	0	μV/°C
		V _{DDA} = 3.3V, T _A = 0 to 50 °C DAC output buffer ON	0	20	50	
Gain ⁽¹⁾	Gain error ⁽⁶⁾	C _L ≤ 50 pF, R _L ≥ 5 kΩ DAC output buffer ON	-	+0.1 / -0.2%	+0.2 / - 0.5%	%
		No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF	-	+0 / -0.2%	+0 / -0.4%	
dGain/dT ⁽¹⁾	Gain error temperature coefficient	V _{DDA} = 3.3V, T _A = 0 to 50 °C DAC output buffer OFF	-10	-2	0	μV/°C
		V _{DDA} = 3.3V, T _A = 0 to 50 °C DAC output buffer ON	-40	-8	0	
TUE ⁽¹⁾	Total unadjusted error	C _L ≤ 50 pF, R _L ≥ 5 kΩ DAC output buffer ON	-	12	30	LSB
		No R _{LOAD} , C _L ≤ 50 pF DAC output buffer OFF	-	8	12	
t _{SETTLING}	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes till DAC_OUT reaches final value ±1LSB)	C _L ≤ 50 pF, R _L ≥ 5 kΩ	-	7	12	μs
Update rate	Max frequency for a correct DAC_OUT change (95% of final value) with 1 LSB variation in the input code	C _L ≤ 50 pF, R _L ≥ 5 kΩ	-	-	1	Msp/s
t _{WAKEUP}	Wakeup time from off state (setting the ENx bit in the DAC Control register) ⁽⁷⁾	C _L ≤ 50 pF, R _L ≥ 5 kΩ	-	9	15	μs
PSRR+	V _{DDA} supply rejection ratio (static DC measurement)	C _L ≤ 50 pF, R _L ≥ 5 kΩ	-	-60	-35	dB

1. Guaranteed by characterization results.
2. Difference between two consecutive codes - 1 LSB.
3. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
4. Difference between the value measured at Code (0x800) and the ideal value = V/2.
5. Difference between the value measured at Code (0x001) and the ideal value.
6. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFFF when buffer is OFF, and from code giving 0.2 V and (V_{DDA} - 0.2) V when buffer is ON.
7. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).

TFBGA64 device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Figure 49. TFBGA64 5 x 5 mm, 0.5 mm pitch, package top view example

1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.