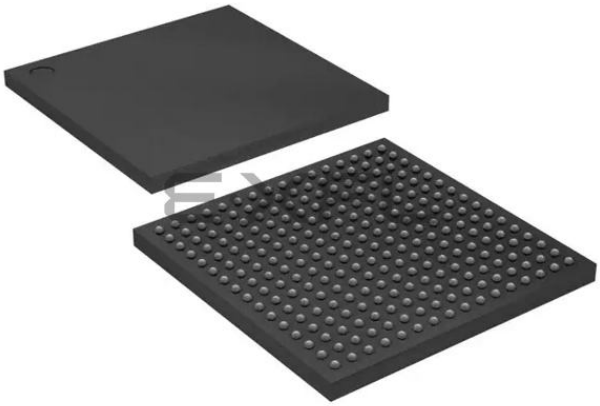

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

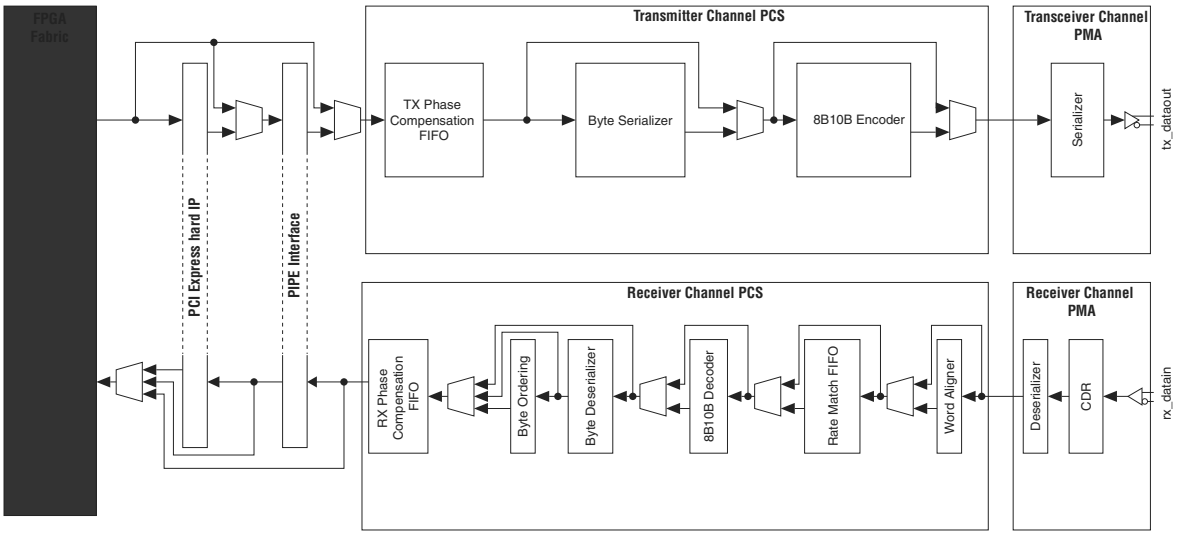
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	963
Number of Logic Elements/Cells	15408
Total RAM Bits	516096
Number of I/O	165
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep4ce15f17c7n

Figure 1–1 shows the structure of the Cyclone IV GX transceiver.

Figure 1–1. Transceiver Channel for the Cyclone IV GX Device



For more information, refer to the *Cyclone IV Transceivers Architecture* chapter.

Hard IP for PCI Express (Cyclone IV GX Devices Only)

Cyclone IV GX devices incorporate a single hard IP block for ×1, ×2, or ×4 PCIe (PIPE) in each device. This hard IP block is a complete PCIe (PIPE) protocol solution that implements the PHY-MAC layer, Data Link Layer, and Transaction Layer functionality. The hard IP for the PCIe (PIPE) block supports root-port and end-point configurations. This pre-verified hard IP block reduces risk, design time, timing closure, and verification. You can configure the block with the Quartus II software's PCI Express Compiler, which guides you through the process step by step.

For more information, refer to the *PCI Express Compiler User Guide*.

Document Revision History

Table 1–10 lists the revision history for this chapter.

Table 1–10. Document Revision History

Date	Version	Changes
March 2016	2.0	■ Updated Table 1–4 and Table 1–5 to remove support for the N148 package. ■ Updated Figure 1–2 to remove support for the N148 package.
April 2014	1.9	Updated "Packaging Ordering Information for the Cyclone IV E Device".
May 2013	1.8	Updated Table 1–3, Table 1–6 and Figure 1–3 to add new device options and packages.
February 2013	1.7	Updated Table 1–3, Table 1–6 and Figure 1–3 to add new device options and packages.
October 2012	1.6	Updated Table 1–3 and Table 1–4.
November 2011	1.5	■ Updated "Cyclone IV Device Family Features" section. ■ Updated Figure 1–2 and Figure 1–3.
December 2010	1.4	■ Updated for the Quartus II software version 10.1 release. ■ Added Cyclone IV E new device package information. ■ Updated Table 1–1, Table 1–2, Table 1–3, Table 1–5, and Table 1–6. ■ Updated Figure 1–3. ■ Minor text edits.
July 2010	1.3	Updated Table 1–2 to include F484 package information.
March 2010	1.2	■ Updated Table 1–3 and Table 1–6. ■ Updated Figure 1–3. ■ Minor text edits.
February 2010	1.1	■ Added Cyclone IV E devices in Table 1–1, Table 1–3, and Table 1–6 for the Quartus II software version 9.1 SP1 release. ■ Added the "Cyclone IV Device Family Speed Grades" and "Configuration" sections. ■ Added Figure 1–3 to include Cyclone IV E Device Packaging Ordering Information. ■ Updated Table 1–2, Table 1–4, and Table 1–5 for Cyclone IV GX devices. ■ Minor text edits.
November 2009	1.0	Initial release.

In addition to the three general routing outputs, LEs in an LAB have register chain outputs, which allows registers in the same LAB to cascade together. The register chain output allows the LUTs to be used for combinational functions and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources.

LE Operating Modes

Cyclone IV LEs operate in the following modes:

- Normal mode
- Arithmetic mode

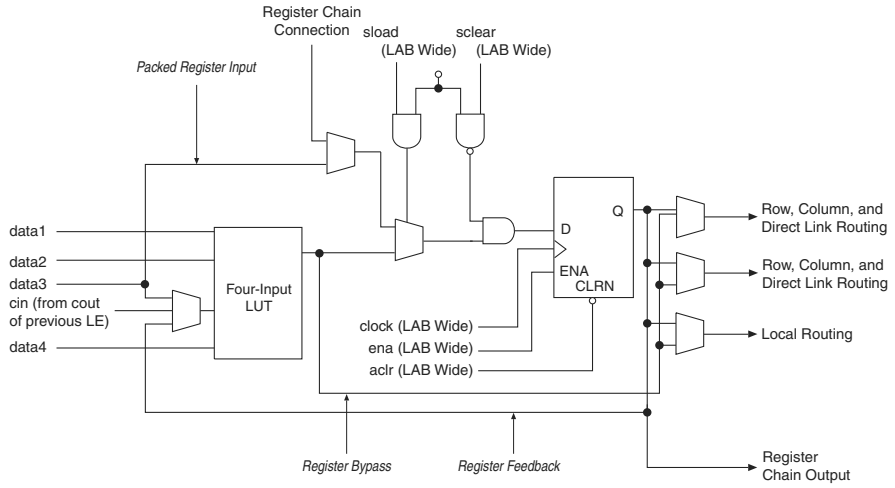
The Quartus® II software automatically chooses the appropriate mode for common functions, such as counters, adders, subtractors, and arithmetic functions, in conjunction with parameterized functions such as the library of parameterized modules (LPM) functions. You can also create special-purpose functions that specify which LE operating mode to use for optimal performance, if required.

Normal Mode

Normal mode is suitable for general logic applications and combinational functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (Figure 2–2). The Quartus II Compiler automatically selects the carry-in (`cin`) or the `data3` signal as one of the inputs to the LUT. LEs in normal mode support packed registers and register feedback.

Figure 2–2 shows LEs in normal mode.

Figure 2–2. Cyclone IV Device LEs in Normal Mode



Control Signals

The clock-enable control signal controls the clock entering the input and output registers and the entire M9K memory block. This signal disables the clock so that the M9K memory block does not see any clock edges and does not perform any operations.

The `rden` and `wren` control signals control the read and write operations for each port of M9K memory blocks. You can disable the `rden` or `wren` signals independently to save power whenever the operation is not required.

Parity Bit Support

Parity checking for error detection is possible with the parity bit along with internal logic resources. Cyclone IV devices M9K memory blocks support a parity bit for each storage byte. You can use this bit as either a parity bit or as an additional data bit. No parity function is actually performed on this bit.

Byte Enable Support

Cyclone IV devices M9K memory blocks support byte enables that mask the input data so that only specific bytes of data are written. The unwritten bytes retain the previous written value. The `wren` signals, along with the byte-enable (`byteena`) signals, control the write operations of the RAM block. The default value of the `byteena` signals is high (enabled), in which case writing is controlled only by the `wren` signals. There is no clear port to the `byteena` registers. M9K blocks support byte enables when the write port has a data width of ×16, ×18, ×32, or ×36 bits.

Byte enables operate in one-hot manner, with the LSB of the `byteena` signal corresponding to the least significant byte of the data bus. For example, if `byteena = 01` and you are using a RAM block in ×18 mode, `data[8..0]` is enabled and `data[17..9]` is disabled. Similarly, if `byteena = 11`, both `data[8..0]` and `data[17..9]` are enabled. Byte enables are active high.

Table 3–2 lists the byte selection.

Table 3–2. byteena for Cyclone IV Devices M9K Blocks *(1)*

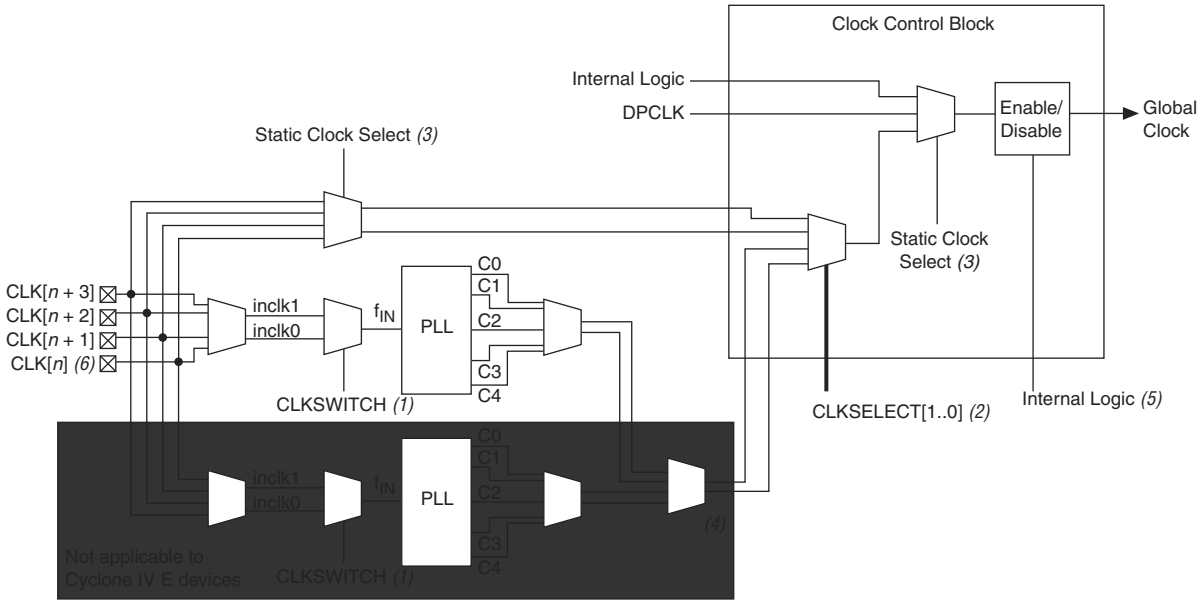
byteena[3..0]	Affected Bytes			
	datain ×16	datain ×18	datain ×32	datain ×36
[0] = 1	[7..0]	[8..0]	[7..0]	[8..0]
[1] = 1	[15..8]	[17..9]	[15..8]	[17..9]
[2] = 1	—	—	[23..16]	[26..18]
[3] = 1	—	—	[31..24]	[35..27]

Note to Table 3–2:

(1) Any combination of byte enables is possible.

Figure 5–1 shows the clock control block.

Figure 5–1. Clock Control Block



Notes to Figure 5–1:

(1) The `clkswitch` signal can either be set through the configuration file or dynamically set when using the manual PLL switchover feature. The output of the multiplexer is the input clock (f_{IN}) for the PLL.

(2) The `clkselect[1..0]` signals are fed by internal logic and are used to dynamically select the clock source for the GCLK when the device is in user mode.

(3) The static clock select signals are set in the configuration file. Therefore, dynamic control when the device is in user mode is not feasible.

(4) Two out of four PLL clock outputs are selected from adjacent PLLs to drive into the clock control block.

(5) You can use internal logic to enable or disable the GCLK in user mode.

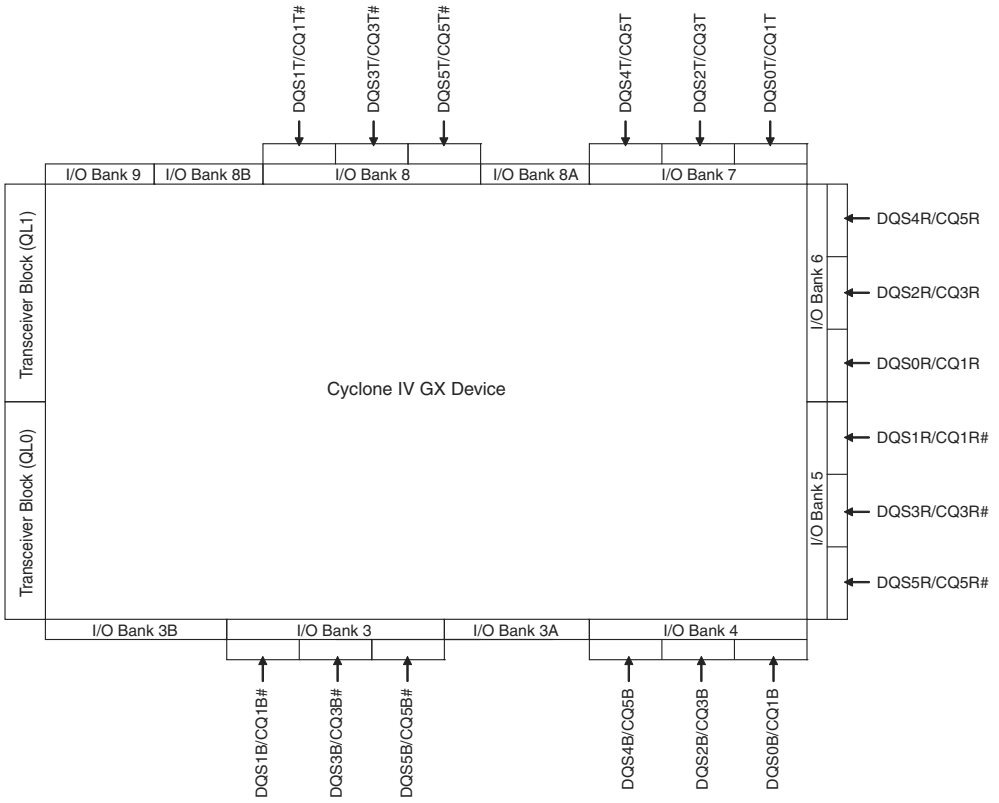
(6) `CLK[n]` is not available on the left side of Cyclone IV E devices.

Each PLL generates five clock outputs through the `c[4..0]` counters. Two of these clocks can drive the GCLK through a clock control block, as shown in Figure 5–1.

For more information about how to use the clock control block in the Quartus II software, refer to the *ALTCLKCTRL Megafunction User Guide*.

Figure 7–2 shows the location and numbering of the DQS, DQ, or CQ# pins in the Cyclone IV GX I/O banks.

Figure 7–2. DQS, CQ, or CQ# Pins in Cyclone IV GX I/O Banks *(1)*



Note to Figure 7–2:

(1) The DQS, CQ, or CQ# pin locations in this diagram apply to all packages in Cyclone IV GX devices except devices in 169-pin FBGA and 324-pin FBGA.

Table 8–8 provides the configuration time for AS configuration.

Table 8–8. AS Configuration Time for Cyclone IV Devices (1)

Symbol	Parameter	Cyclone IV E	Cyclone IV GX	Unit
t_{SU}	Setup time	10	8	ns
t_H	Hold time	0	0	ns
t_{CO}	Clock-to-output time	4	4	ns

Note to Table 8–8:

(1) For the AS configuration timing diagram, refer to the *Serial Configuration (EPCS) Devices Datasheet*.

Enabling compression reduces the amount of configuration data that is sent to the Cyclone IV device, which also reduces configuration time. On average, compression reduces configuration time by 50%.

Programming Serial Configuration Devices

Serial configuration devices are non-volatile, flash memory-based devices. You can program these devices in-system with the USB-Blaster™ or ByteBlaster™ II download cables. Alternatively, you can program them with the Altera Programming Unit (APU), supported third-party programmers, or a microprocessor with the SRunner software driver.

You can perform in-system programming of serial configuration devices through the AS programming interface. During in-system programming, the download cable disables device access to the AS interface by driving the nCE pin high. Cyclone IV devices are also held in reset by a low level on nCONFIG. After programming is complete, the download cable releases nCE and nCONFIG, allowing the pull-down and pull-up resistors to drive V_{CC} and GND, respectively.

To perform in-system programming of a serial configuration device through the AS programming interface, you must place the diodes and capacitors as close as possible to the Cyclone IV device. You must ensure that the diodes and capacitors maintain a maximum AC voltage of 4.1 V (Figure 8–6).

If you want to use the setup shown in Figure 8–6 to perform in-system programming of a serial configuration device and single- or multi-device AS configuration, you do not require a series resistor on the DATA line at the near end of the serial configuration device. The existing diodes and capacitors are sufficient.

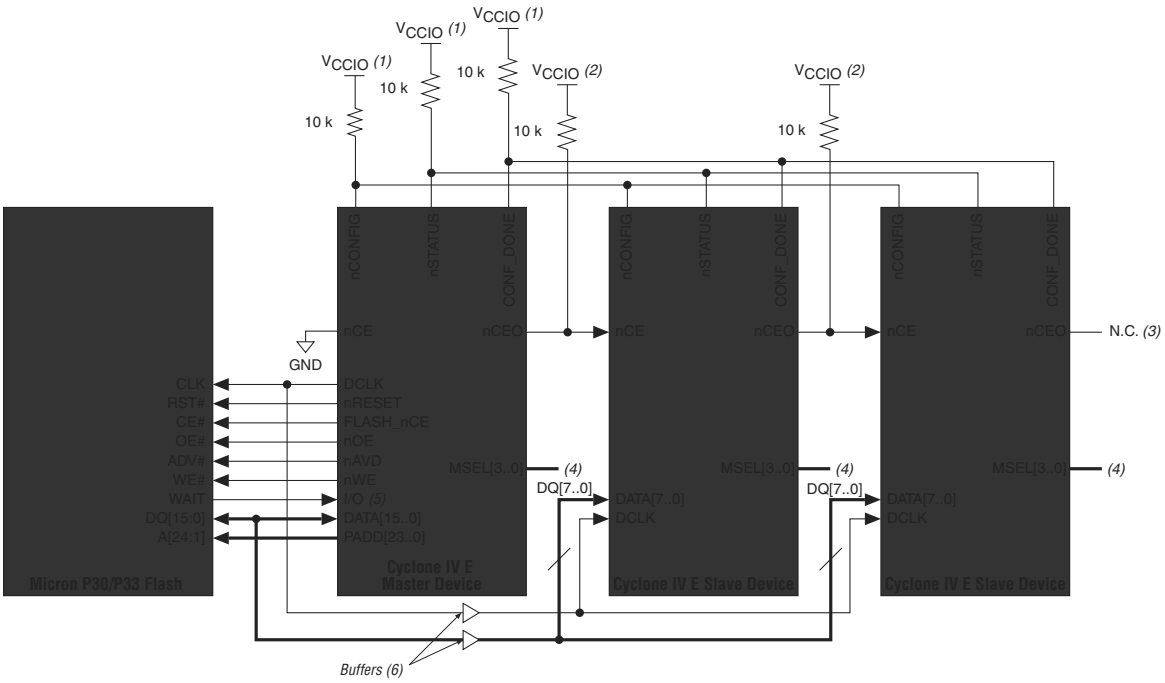
Altera has developed the Serial FlashLoader (SFL), a JTAG-based in-system programming solution for Altera serial configuration devices. The SFL is a bridge design for the Cyclone IV device that uses its JTAG interface to access the EPCS JIC (JTAG Indirect Configuration Device Programming) file and then uses the AS interface to program the EPCS device. Both the JTAG interface and AS interface are bridged together inside the SFL design.

For more information about implementing the SFL with Cyclone IV devices, refer to *AN 370: Using the Serial FlashLoader with the Quartus II Software*.

Byte-Wide Multi-Device AP Configuration

The simpler method for multi-device AP configuration is the byte-wide multi-device AP configuration. In the byte-wide multi-device AP configuration, the LSB of the DATA[7..0] pin from the flash and master device (set to the AP configuration scheme) is connected to the slave devices set to the FPP configuration scheme, as shown in Figure 8–8.

Figure 8–8. Byte-Wide Multi-Device AP Configuration



Notes to Figure 8–8:

(1) Connect the pull-up resistors to the V_{CCIO} supply of the bank in which the pin resides.

(2) Connect the pull-up resistor to the V_{CCIO} supply voltage of the I/O bank in which the nCE pin resides.

(3) The nCEO pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.

(4) The MSEL pin settings vary for different configuration voltage standards and POR time. You must set the master device in AP mode and the slave devices in FPP mode. To connect MSEL[3..0] for the master device in AP mode and the slave devices in FPP mode, refer to Table 8–5 on page 8–9. Connect the MSEL pins directly to V_{CCA} or GND.

(5) The AP configuration ignores the WAIT signal during configuration mode. However, if you are accessing flash during user mode with user logic, you can optionally use the normal I/O to monitor the WAIT signal from the Micron P30 or P33 flash.

(6) Connect the repeater buffers between the Cyclone IV E master device and slave devices for DATA[15..0] and DCLK. All I/O inputs must maintain a maximum AC voltage of 4.1 V. The output resistance of the repeater buffers must fit the maximum overshoot equation outlined in "Configuration and JTAG Pin I/O Requirements" on page 8–5.

Word-Wide Multi-Device AP Configuration

The more efficient setup is one in which some of the slave devices are connected to the LSB of the DATA[7..0] and the remaining slave devices are connected to the MSB of the DATA[15..8]. In the word-wide multi-device AP configuration, the nCEO pin of the master device enables two separate daisy chains of slave devices, allowing both chains to be programmed concurrently, as shown in Figure 8–9.

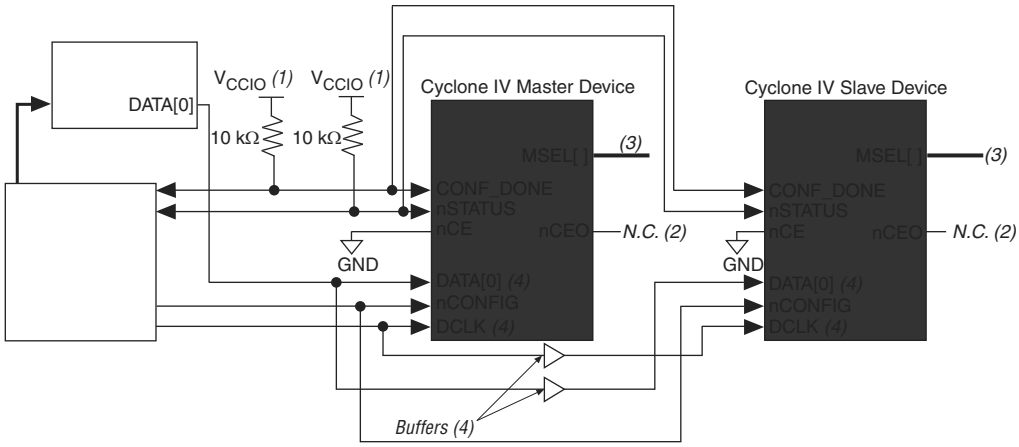
After the first device completes configuration in a multi-device configuration chain, its nCEO pin drives low to activate the nCE pin of the second device, which prompts the second device to begin configuration. The second device in the chain begins configuration in one clock cycle. Therefore, the transfer of data destinations is transparent to the external host device. nCONFIG, nSTATUS, DCLK, DATA[0], and CONF_DONE configuration pins are connected to every device in the chain. To ensure signal integrity and prevent clock skew problems, configuration signals may require buffering. Ensure that DCLK and DATA lines are buffered. All devices initialize and enter user mode at the same time because all CONF_DONE pins are tied together.

If any device detects an error, configuration stops for the entire chain and you must reconfigure the entire chain because all nSTATUS and CONF_DONE pins are tied together. For example, if the first device flags an error on nSTATUS, it resets the chain by pulling its nSTATUS pin low. This behavior is similar to a single device detecting an error.

You can have multiple devices that contain the same configuration data in your system. To support this configuration scheme, all device nCE inputs are tied to GND, while the nCEO pins are left floating. nCONFIG, nSTATUS, DCLK, DATA[0], and CONF_DONE configuration pins are connected to every device in the chain. To ensure signal integrity and prevent clock skew problems, configuration signals may require buffering. Ensure that the DCLK and DATA lines are buffered. Devices must be of the same density and package. All devices start and complete configuration at the same time.

Figure 8–15 shows a multi-device PS configuration when both Cyclone IV devices are receiving the same configuration data.

Figure 8–15. Multi-Device PS Configuration When Both Devices Receive the Same Data



Notes to Figure 8–15:

(1) You must connect the pull-up resistor to a supply that provides an acceptable input signal for all devices in the chain. V_{CC} must be high enough to meet the V_{IH} specification of the I/O on the device and the external host.

(2) The nCEO pins of both devices are left unconnected or used as user I/O pins when configuring the same configuration data into multiple devices.

(3) The MSEL pin settings vary for different configuration voltage standards and POR time. To connect the MSEL pins, refer to Table 8–3 on page 8–8, Table 8–4 on page 8–8, and Table 8–5 on page 8–9. Connect the MSEL pins directly to V_{CCA} or GND.

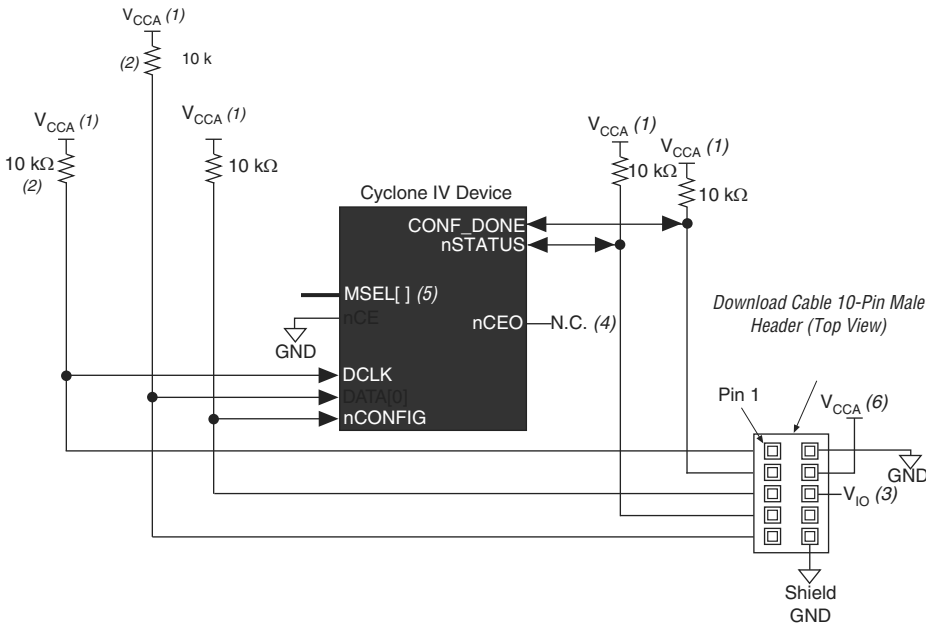
(4) All I/O inputs must maintain a maximum AC voltage of 4.1 V. DATA[0] and DCLK must fit the maximum overshoot outlined in Equation 8–1 on page 8–5.

The programming hardware or download cable then places the configuration data one bit at a time on the DATA[0] pin of the device. The configuration data is clocked into the target device until CONF_DONE goes high. The CONF_DONE pin must have an external 10-kΩ pull-up resistor for the device to initialize.

When you use a download cable, setting the **Auto-restart configuration after error** option does not affect the configuration cycle because you must manually restart configuration in the Quartus II software if an error occurs. Additionally, the **Enable user-supplied start-up clock (CLKUSR)** option has no effect on device initialization, because this option is disabled in the **.sof** when programming the device with the Quartus II Programmer and download cable. Therefore, if you turn on the **CLKUSR** option, you do not have to provide a clock on CLKUSR when you configure the device with the Quartus II Programmer and a download cable.

Figure 8–17 shows PS configuration for Cyclone IV devices with a download cable.

Figure 8–17. PS Configuration Using a Download Cable



Notes to Figure 8–17:

(1) You must connect the pull-up resistor to the same supply voltage as the V_{CCA} supply.

(2) The pull-up resistors on DATA[0] and DCLK are only required if the download cable is the only configuration scheme used on your board. This is to ensure that DATA[0] and DCLK are not left floating after configuration. For example, if you also use a configuration device, the pull-up resistors on DATA[0] and DCLK are not required.

(3) Pin 6 of the header is a V_{IO} reference voltage for the MasterBlaster output driver. V_{IO} must match the V_{CCA} of the device. For this value, refer to the *MasterBlaster Serial/USB Communications Cable User Guide*. With the USB-Blaster, ByteBlaster II, ByteBlaster MV, and EthernetBlaster, this pin is a no connect.

(4) The nCEO pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.

(5) The MSEL pin settings vary for different configuration voltage standards and POR time. To connect the MSEL pins, refer to Table 8–3 on page 8–8, Table 8–4 on page 8–8, and Table 8–5 on page 8–9 for PS configuration schemes. Connect the MSEL pins directly to V_{CCA} or GND.

(6) Power up the V_{CC} of the ByteBlaster II, USB-Blaster, or ByteBlasterMV cable with a 2.5-V supply from V_{CCA}. Third-party programmers must switch to 2.5 V. Pin 4 of the header is a V_{CC} power supply for the MasterBlaster cable. The MasterBlaster cable can receive power from either 5.0- or 3.3-V circuit boards, DC power supply, or 5.0 V from the USB cable. For this value, refer to the *MasterBlaster Serial/USB Communications Cable User Guide*.

☞ WYSIWYG is an optimization technique that performs optimization on a VQM (Verilog Quartus Mapping) netlist in the Quartus II software.

Error Detection Block

Table 9–3 lists the types of CRC detection to check the configuration bits.

Table 9–3. Types of CRC Detection to Check the Configuration Bits

First Type of CRC Detection	Second Type of CRC Detection
■ CRAM error checking ability (32-bit CRC) during user mode, for use by the CRC_ERROR pin. ■ There is only one 32-bit CRC value. This value covers all the CRAM data.	■ 16-bit CRC embedded in every configuration data frame. ■ During configuration, after a frame of data is loaded into the device, the pre-computed CRC is shifted into the CRC circuitry. ■ Simultaneously, the CRC value for the data frame shifted-in is calculated. If the pre-computed CRC and calculated CRC values do not match, nSTATUS is set low. ■ Every data frame has a 16-bit CRC. Therefore, there are many 16-bit CRC values for the whole configuration bit stream. ■ Every device has a different length of configuration data frame.

This section focuses on the first type—the 32-bit CRC when the device is in user mode.

Error Detection Registers

There are two sets of 32-bit registers in the error detection circuitry that store the computed CRC signature and pre-calculated CRC value. A non-zero value on the signature register causes the CRC_ERROR pin to set high.

Figure 9–1 shows the block diagram of the error detection block and the two related 32-bit registers: the signature register and the storage register.

Figure 9–1. Error Detection Block Diagram

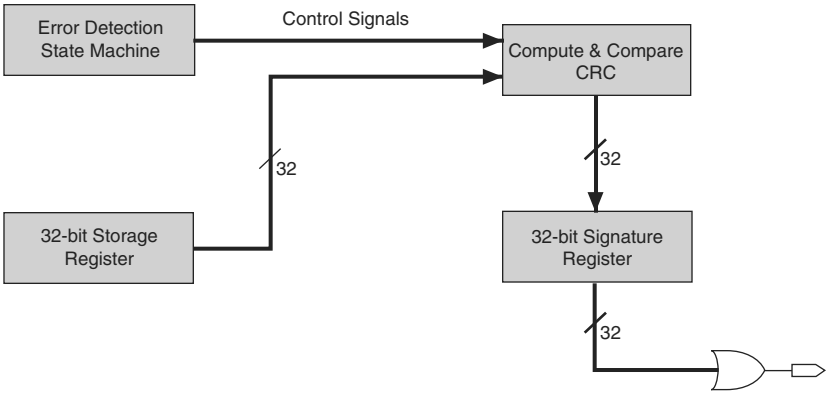
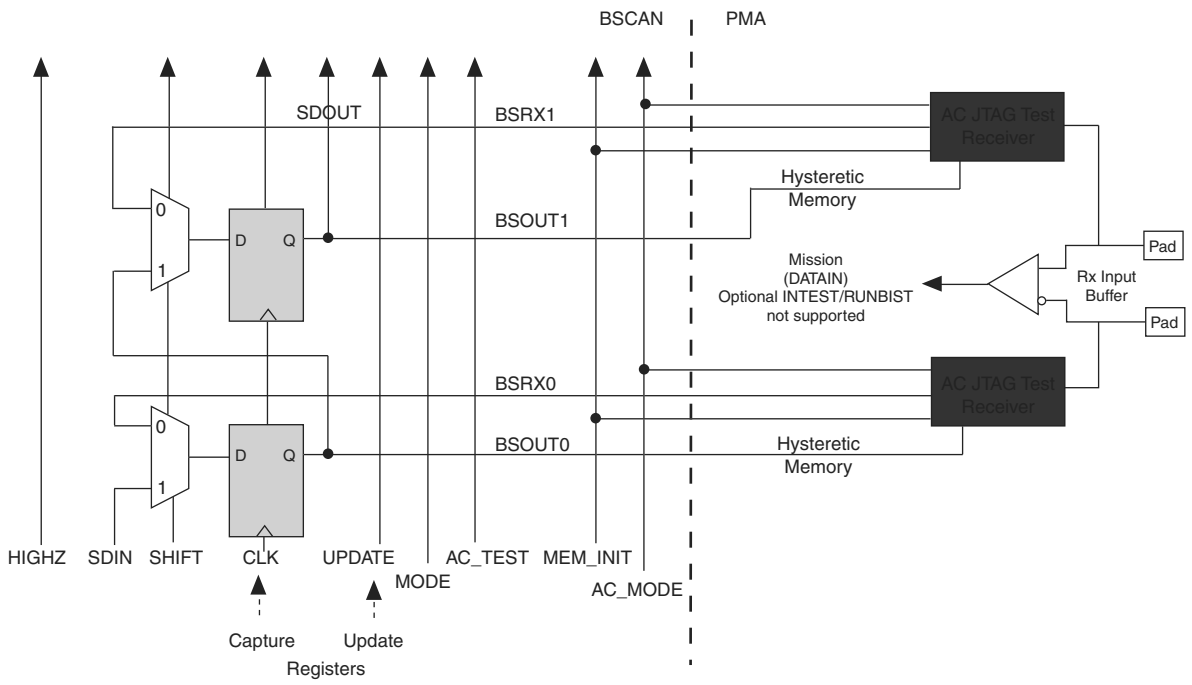


Figure 10–2 shows the Cyclone IV GX HSSI receiver BSC.

Figure 10–2. HSSI Receiver BSC with IEEE Std. 1149.6 BST Circuitry for the Cyclone IV GX Devices



For more information about Cyclone IV devices user I/O boundary-scan cells, refer to the *IEEE 1149.1 (JTAG) Boundary-Scan Testing for Cyclone III Devices* chapter.

BST Operation Control

Table 10–1 lists the boundary-scan register length for Cyclone IV devices.

Table 10–1. Boundary-Scan Register Length for Cyclone IV Devices (Part 1 of 2)

Device	Boundary-Scan Register Length
EP4CE6	603
EP4CE10	603
EP4CE15	1080
EP4CE22	732
EP4CE30	1632
EP4CE40	1632
EP4CE55	1164
EP4CE75	1314
EP4CE115	1620
EP4CGX15	260
EP4CGX22	494
EP4CGX30 (1)	494
EP4CGX50	1006

Fast Recovery from P0s State

The PCIe protocol defines fast training sequences for bit and byte synchronization to transition from L0s to L0 (PIPE P0s to P0) power states. The PHY must acquire bit and byte synchronization when transitioning from L0s to L0 state between 16 ns to 4 µs. Each Cyclone IV GX receiver channel has built-in fast recovery circuit that allows the receiver to meet the requirement when enabled.

Electrical Idle Inference

In PIPE mode, the Cyclone IV GX transceiver supports inferring the electrical idle condition at each receiver instead of detecting the electrical idle condition using analog circuitry, as defined in the version 2.0 of PCIe Base Specification. The inference is supported using `rx_elecidleinfersel[2..0]` port, with valid driven values as listed in Table 1–17 in each link training and status state machine substate.

Table 1–17. Electrical Idle Inference Conditions

rx_elecidleinfersel [2..0]	Link Training and Status State Machine State	Description
3'b100	L0	Absence of `update_FC` or alternatively skip ordered set in 128 µs window
3'b101	Recovery.RcvrCfg	Absence of `TS1` or `TS2` ordered set in 1280 UI interval
3'b101	Recovery.Speed when successful speed negotiation = 1'b1	Absence of `TS1` or `TS2` ordered set in 1280 UI interval
3'b110	Recovery.Speed when successful speed negotiation = 1'b0	Absence of an exit from electrical idle in 2000 UI interval
3'b111	Loopback.Active (as slave)	Absence of an exit from electrical idle in 128 µs window

The electrical idle inference module drives the `pipeelecidle` signal high in each receiver channel when an electrical idle condition is inferred. The electrical idle inference module cannot detect electrical idle exit condition based on the reception of the electrical idle exit ordered set, as specified in the PCI Express (PIPE) Base Specification.

When enabled, the electrical idle inference block uses electrical idle ordered set detection from the fast recovery circuitry to drive the `pipeelecidle` signal.

Compliance Pattern Transmission

In PIPE mode, the Cyclone IV GX transceiver supports compliance pattern transmission which requires the first /K28.5/ code group of the compliance pattern to be encoded with negative current disparity. This requirement is supported using a `tx_forcedispcompliance` port that when driven with logic high, the transmitter data on the `tx_datain` port is transmitted with negative current running disparity.

User Reset and Power-Down Signals

Each transceiver channel in the Cyclone IV GX device has individual reset signals to reset its physical coding sublayer (PCS) and physical medium attachment (PMA). The transceiver block also has a power-down signal that affects the multipurpose phase-locked loops (PLLs), general purpose PLLs, and all the channels in the transceiver block.

☞ All reset and power-down signals are asynchronous.

Table 2–1 lists the reset signals available for each transceiver channel.

Table 2–1. Transceiver Channel Reset Signals

Signal	ALTGX MegaWizard Plug-In Manager Configurations	Description
tx_digitalreset (1)	■ Transmitter Only ■ Receiver and Transmitter	Provides asynchronous reset to all digital logic in the transmitter PCS, including the XAUI transmit state machine. The minimum pulse width for this signal is two parallel clock cycles.
rx_digitalreset (1)	■ Receiver Only ■ Receiver and Transmitter	Resets all digital logic in the receiver PCS, including: ■ XAUI receiver state machines ■ GIGE receiver state machines ■ XAUI channel alignment state machine ■ BIST-PRBS verifier ■ BIST-incremental verifier The minimum pulse width for this signal is two parallel clock cycles.
rx_analogreset	■ Receiver Only ■ Receiver and Transmitter	Resets the receiver CDR present in the receiver channel. The minimum pulse width is two parallel clock cycles.

Note to Table 2–1:

(1) Assert this signal until the clocks coming out of the multipurpose PLL and receiver CDR are stabilized. Stable parallel clocks are essential for proper operation of transmitter and receiver phase-compensation FIFOs in the PCS.

Table 3–3. Cyclone IV GX Supported Dynamic Reconfiguration Mode (Part 2 of 2)

Dynamic Reconfiguration Supported Mode	Operational Mode			Quartus II Instances			.mif Requirements
	Transmitter Only	Receiver Only	Transmitter and Receiver Only	ALTGX	ALTGX_ RECONFIG	ALTPLL_ RECONFIG	
Channel Reconfiguration							
Channel Interface	✓	✓	✓	✓	✓	—	✓
Data Rate Division in Receiver Channel	—	✓	✓	✓	✓	—	✓
PLL Reconfiguration	✓	✓	✓	✓	—	✓	✓

The following modes are available for dynamically reconfiguring the Cyclone IV transceivers:

- "PMA Controls Reconfiguration Mode" on page 3–13
- "Transceiver Channel Reconfiguration Mode" on page 3–21
  - Channel interface (**.mif** based)
  - Data rate division in receiver channel (**.mif** based)

The following sections describe each of these modes in detail.

The following modes are unsupported for dynamic reconfiguration:

- Dynamically enable/disable PRBS or BIST
- Switch between a receiver-only channel and a transmitter-only channel
- Switch between a ×1 mode to a bonded ×4 mode

PMA Controls Reconfiguration Mode

You can dynamically reconfigure the following PMA controls for all supported transceiver configurations channels as configured in the ALTGX instances:

- Pre-emphasis settings
- Equalization settings (channel reconfiguration mode does not support equalization settings)
- DC gain settings
- V_{OD} settings

You can use the analog reconfiguration feature to dynamically reconfigure the transceivers channels setting in either the transmitter or the receivers in the PMA blocks. You can update the PMA controls on-the-fly based on the desired input. You can perform both read and write transaction separately for this analog reconfiguration mode.

PMA Control Ports Used in a Read Transaction

- `tx_vodctrl_out` is 3 bits per channel
- `tx_preemp_out` is 5 bits per channel
- `rx_eqdcgain_out` is 2 bits per channel
- `rx_eqctrl_out` is 4 bits per channel

For example, assume the number of channels controlled by the dynamic reconfiguration controller is two, `tx_vodctrl_out` is 6 bits wide.

Write Transaction

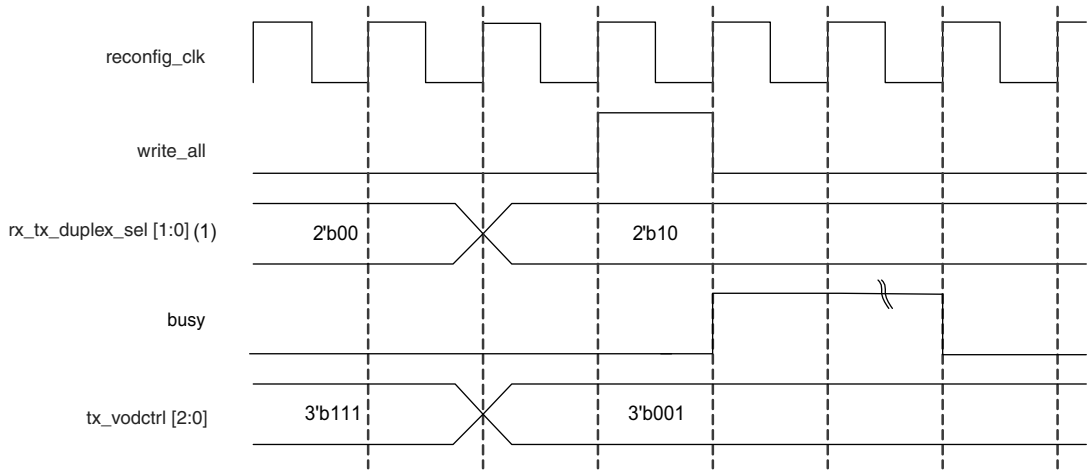
The value you set at the selected PMA control ports is written to all the transceiver channels connected to the ALTGX_RECONFIG instance.

For example, assume you have enabled `tx_vodctrl` in the ALTGX_RECONFIG MegaWizard Plug-In Manager to reconfigure the V_{OD} of the transceiver channels. To complete a write transaction to reconfigure the V_{OD}, perform the following steps:

1. Before you initiate a write transaction, set the selected PMA control ports to the desired settings (for example, `tx_vodctrl` = **3'b001**).
2. Set the `rx_tx_duplex_sel` port to **2'b10** so that only the transmit PMA controls are written to the transceiver channel.
3. Ensure that the `busy` signal is low before you start a write transaction.
4. Assert the `write_all` signal for one `reconfig_clk` clock cycle. This initiates the write transaction.
5. The `busy` output status signal is asserted high to indicate that the dynamic reconfiguration controller is busy writing the PMA control values. When the write transaction has completed, the `busy` signal goes low.

Figure 3–6 shows the write transaction for Method 2.

Figure 3–6. Write Transaction Waveform—Use the same control signal for all the channels Option



Note to Figure 3–6:

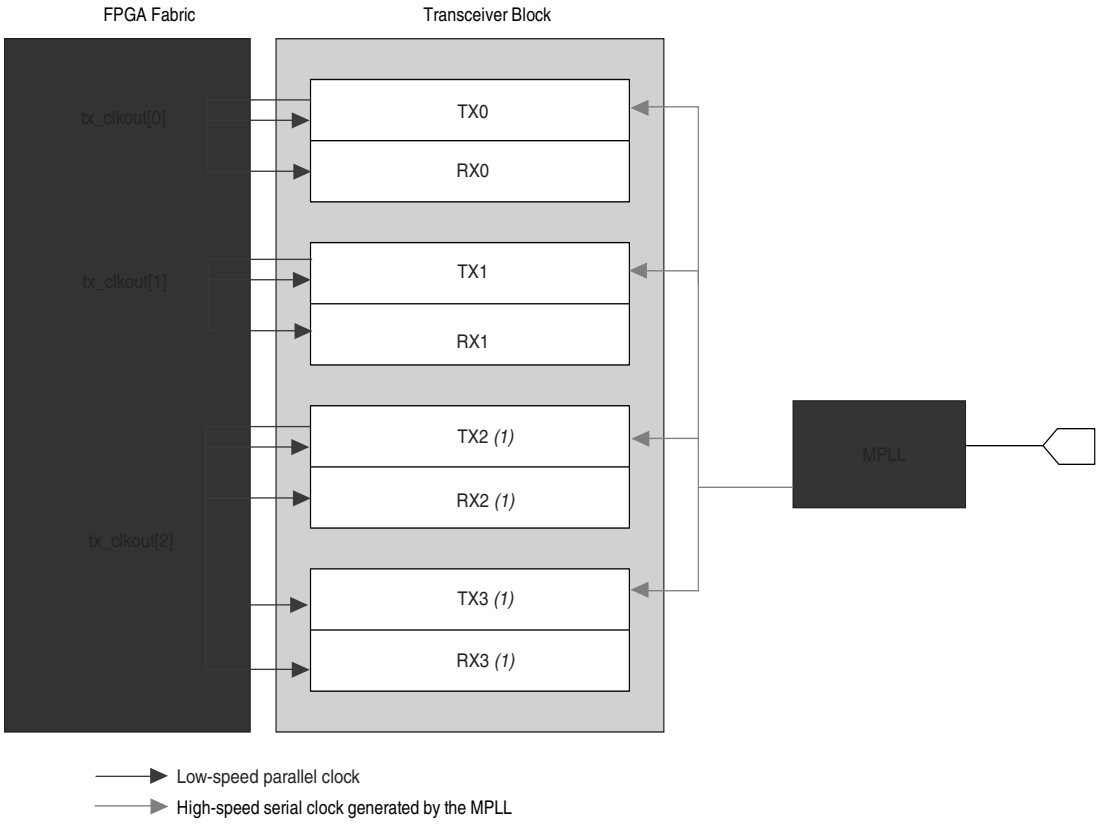
(1) In this waveform example, you want to write to only the transmitter portion of the channel.

Option 2: Use the Respective Channel Transmitter Core Clocks

- Enable this option if you want the individual transmitter channel's `tx_clkout` signal to provide the read clock to its respective Receive Phase Compensation FIFO.
- This option is typically enabled when all the transceiver channels have rate matching enabled with different data rates and are reconfigured to another Basic or Protocol functional mode with rate matching enabled.

Figure 3–14 shows the respective `tx_clkout` of each channel clocking the respective channels of a transceiver block.

Figure 3–14. Option 2 for Receiver Core Clocking (Channel Reconfiguration Mode)



Note to Figure 3–14:

(1) Assuming channel 2 and 3 are running at the same data rate with rate matcher enabled and are reconfigured to another Basic or Protocol functional mode with rate matching enabled.

Cyclone IV E industrial devices I7 are offered with extended operating temperature range.

Absolute Maximum Ratings

Absolute maximum ratings define the maximum operating conditions for Cyclone IV devices. The values are based on experiments conducted with the device and theoretical modeling of breakdown and damage mechanisms. The functional operation of the device is not implied at these conditions. Table 1–1 lists the absolute maximum ratings for Cyclone IV devices.

CAUTION

Conditions beyond those listed in Table 1–1 cause permanent damage to the device. Additionally, device operation at the absolute maximum ratings for extended periods of time have adverse effects on the device.

Table 1–1. Absolute Maximum Ratings for Cyclone IV Devices *(1)*

Symbol	Parameter	Min	Max	Unit
V_{CCINT}	Core voltage, PCI Express® (PCIe®) hard IP block, and transceiver physical coding sublayer (PCS) power supply	–0.5	1.8	V
V_{CCA}	Phase-locked loop (PLL) analog power supply	–0.5	3.75	V
V_{CCD_PLL}	PLL digital power supply	–0.5	1.8	V
V_{CCIO}	I/O banks power supply	–0.5	3.75	V
V_{CC_CLKIN}	Differential clock input pins power supply	–0.5	4.5	V
V_{CCH_GXB}	Transceiver output buffer power supply	–0.5	3.75	V
V_{CCA_GXB}	Transceiver physical medium attachment (PMA) and auxiliary power supply	–0.5	3.75	V
V_{CCL_GXB}	Transceiver PMA and auxiliary power supply	–0.5	1.8	V
V_I	DC input voltage	–0.5	4.2	V
I_{OUT}	DC output current, per pin	–25	40	mA
T_{STG}	Storage temperature	–65	150	°C
T_J	Operating junction temperature	–40	125	°C

Note to Table 1–1:

(1) Supply voltage specifications apply to voltage readings taken at the device pins with respect to ground, not at the power supply.

Maximum Allowed Overshoot or Undershoot Voltage

During transitions, input signals may overshoot to the voltage shown in Table 1–2 and undershoot to –2.0 V for a magnitude of currents less than 100 mA and for periods shorter than 20 ns. Table 1–2 lists the maximum allowed input overshoot voltage and the duration of the overshoot voltage as a percentage over the lifetime of the device. The maximum allowed overshoot duration is specified as a percentage of high-time over the lifetime of the device.

For more information about the supported maximum clock rate, device and pin planning, IP implementation, and device termination, refer to *Section III: System Performance Specifications* of the *External Memory Interface Handbook*.

Table 1–37 lists the memory output clock jitter specifications for Cyclone IV devices.

Table 1–37. Memory Output Clock Jitter Specifications for Cyclone IV Devices *(1), (2)*

Parameter	Symbol	Min	Max	Unit
Clock period jitter	$t_{JIT(per)}$	–125	125	ps
Cycle-to-cycle period jitter	$t_{JIT(cc)}$	–200	200	ps
Duty cycle jitter	$t_{JIT(duty)}$	–150	150	ps

Notes to Table 1–37:

(1) Memory output clock jitter measurements are for 200 consecutive clock cycles, as specified in the JEDEC DDR2 standard.

(2) The clock jitter specification applies to memory output clock pins generated using DDIO circuits clocked by a PLL output routed on a global clock (GCLK) network.

Duty Cycle Distortion Specifications

Table 1–38 lists the worst case duty cycle distortion for Cyclone IV devices.

Table 1–38. Duty Cycle Distortion on Cyclone IV Devices I/O Pins *(1), (2), (3)*

Symbol	C6		C7, I7		C8, I8L, A7		C9L		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
Output Duty Cycle	45	55	45	55	45	55	45	55	%

Notes to Table 1–38:

(1) The duty cycle distortion specification applies to clock outputs from the PLLs, global clock tree, and IOE driving the dedicated and general purpose I/O pins.

(2) Cyclone IV devices meet the specified duty cycle distortion at the maximum output toggle rate for each combination of I/O standard and current strength.

(3) Cyclone IV E 1.0 V core voltage devices only support C8L, C9L, and I8L speed grades. Cyclone IV E 1.2 V core voltage devices only support C6, C7, C8, I7, and A7 speed grades. Cyclone IV GX devices only support C6, C7, C8, and I7 speed grades.

OCT Calibration Timing Specification

Table 1–39 lists the duration of calibration for series OCT with calibration at device power-up for Cyclone IV devices.

Table 1–39. Timing Specification for Series OCT with Calibration at Device Power-Up for Cyclone IV Devices *(1)*

Symbol	Description	Maximum	Units
t_{OCTCAL}	Duration of series OCT with calibration at device power-up	20	µs

Note to Table 1–39:

(1) OCT calibration takes place after device configuration and before entering user mode.