



Welcome to E-XFL.COM

Understanding **Embedded - FPGAs (Field Programmable Gate Array)**

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details

Product Status	Active
Number of LABs/CLBs	963
Number of Logic Elements/Cells	15408
Total RAM Bits	516096
Number of I/O	343
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	484-BGA
Supplier Device Package	484-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep4ce15f23i7

This section provides a complete overview of all features relating to the Cyclone[®] IV device family, which is the most architecturally advanced, high-performance, low-power FPGA in the marketplace. This section includes the following chapters:

- Chapter 1, Cyclone IV FPGA Device Family Overview
- Chapter 2, Logic Elements and Logic Array Blocks in Cyclone IV Devices
- Chapter 3, Memory Blocks in Cyclone IV Devices
- Chapter 4, Embedded Multipliers in Cyclone IV Devices
- Chapter 5, Clock Networks and PLLs in Cyclone IV Devices

Revision History

Refer to each chapter for its own specific revision history. For information about when each chapter was updated, refer to the Chapter Revision Dates section, which appears in the complete handbook.

 For more information, refer to the *External Memory Interfaces in Cyclone IV Devices* chapter.

Configuration

Cyclone IV devices use SRAM cells to store configuration data. Configuration data is downloaded to the Cyclone IV device each time the device powers up. Low-cost configuration options include the Altera EPCS family serial flash devices and commodity parallel flash configuration options. These options provide the flexibility for general-purpose applications and the ability to meet specific configuration and wake-up time requirements of the applications.

Table 1-9 lists which configuration schemes are supported by Cyclone IV devices.

Table 1-9. Configuration Schemes for Cyclone IV Device Family

Devices	Supported Configuration Scheme
Cyclone IV GX	AS, PS, JTAG, and FPP ⁽¹⁾
Cyclone IV E	AS, AP, PS, FPP, and JTAG

Note to Table 1-9:

(1) The FPP configuration scheme is only supported by the EP4CGX30F484 and EP4CGX50/75/110/150 devices.

IEEE 1149.6 (AC JTAG) is supported on all transceiver I/O pins. All other pins support IEEE 1149.1 (JTAG) for boundary scan testing.

 For more information, refer to the *JTAG Boundary-Scan Testing for Cyclone IV Devices* chapter.

For Cyclone IV GX devices to meet the PCIe 100 ms wake-up time requirement, you must use passive serial (PS) configuration mode for the EP4CGX15/22/30 devices and use fast passive parallel (FPP) configuration mode for the EP4CGX30F484 and EP4CGX50/75/110/150 devices.

 For more information, refer to the *Configuration and Remote System Upgrades in Cyclone IV Devices* chapter.

The cyclical redundancy check (CRC) error detection feature during user mode is supported in all Cyclone IV GX devices. For Cyclone IV E devices, this feature is only supported for the devices with the core voltage of 1.2 V.

 For more information about CRC error detection, refer to the *SEU Mitigation in Cyclone IV Devices* chapter.

High-Speed Transceivers (Cyclone IV GX Devices Only)

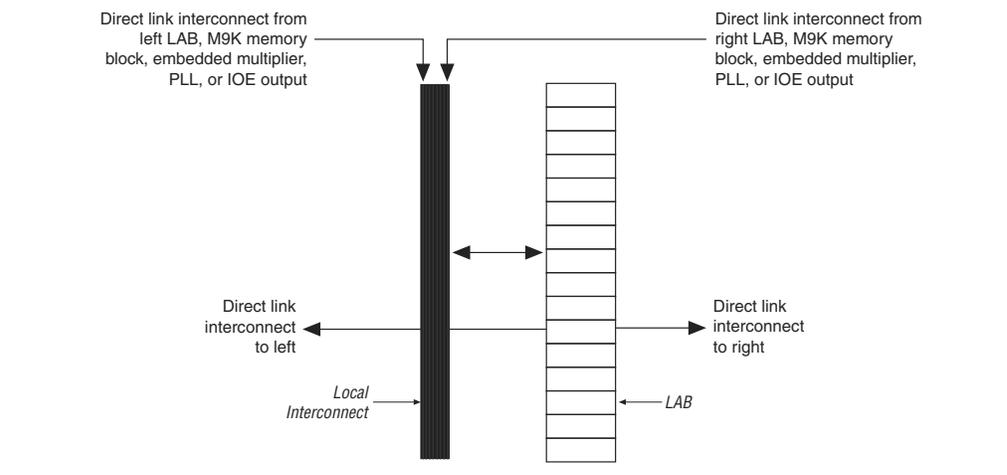
Cyclone IV GX devices contain up to eight full duplex high-speed transceivers that can operate independently. These blocks support multiple industry-standard communication protocols, as well as Basic mode, which you can use to implement your own proprietary protocols. Each transceiver channel has its own pre-emphasis and equalization circuitry, which you can set at compile time to optimize signal integrity and reduce bit error rates. Transceiver blocks also support dynamic reconfiguration, allowing you to change data rates and protocols on-the-fly.

LAB Interconnects

The LAB local interconnect is driven by column and row interconnects and LE outputs in the same LAB. Neighboring LABs, phase-locked loops (PLLs), M9K RAM blocks, and embedded multipliers from the left and right can also drive the local interconnect of a LAB through the direct link connection. The direct link connection feature minimizes the use of row and column interconnects, providing higher performance and flexibility. Each LE can drive up to 48 LEs through fast local and direct link interconnects.

Figure 2-5 shows the direct link connection.

Figure 2-5. Cyclone IV Device Direct Link Connection



LAB Control Signals

Each LAB contains dedicated logic for driving control signals to its LEs. The control signals include:

- Two clocks
- Two clock enables
- Two asynchronous clears
- One synchronous clear
- One synchronous load

You can use up to eight control signals at a time. Register packing and synchronous load cannot be used simultaneously.

Each LAB can have up to four non-global control signals. You can use additional LAB control signals as long as they are global signals.

Synchronous clear and load signals are useful for implementing counters and other functions. The synchronous clear and synchronous load signals are LAB-wide signals that affect all registers in the LAB.

Figure 3-3 and Figure 3-4 show the address clock enable waveform during read and write cycles, respectively.

Figure 3-3. Cyclone IV Devices Address Clock Enable During Read Cycle Waveform

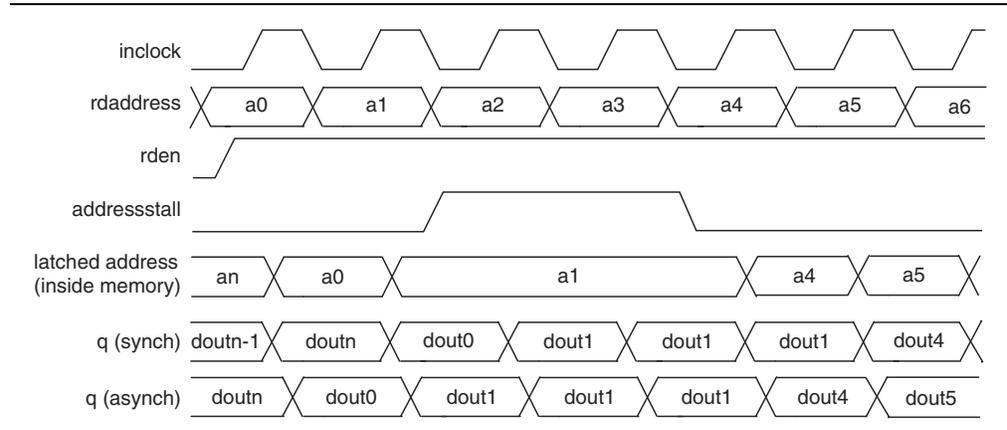
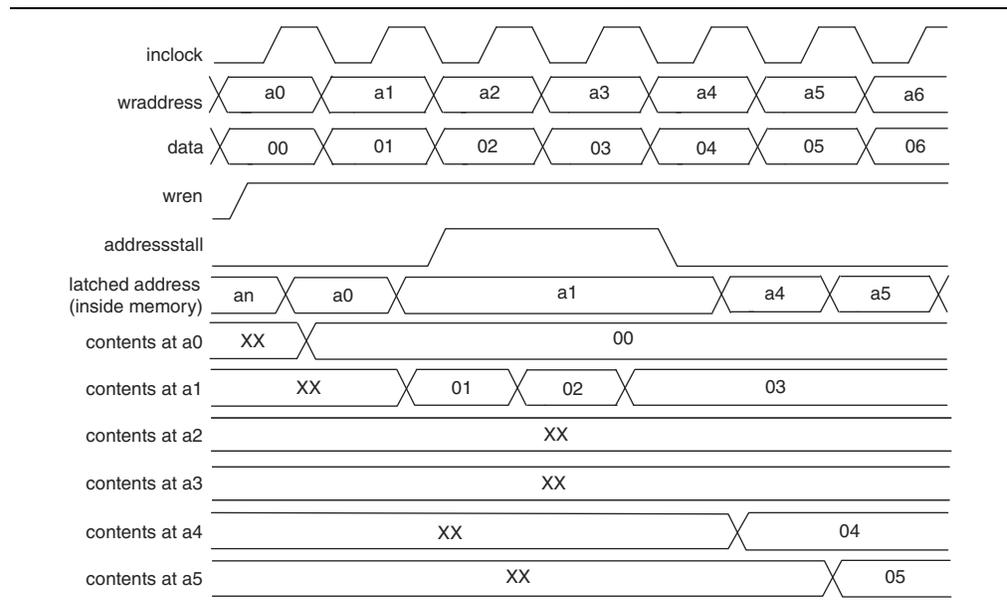


Figure 3-4. Cyclone IV Devices Address Clock Enable During Write Cycle Waveform

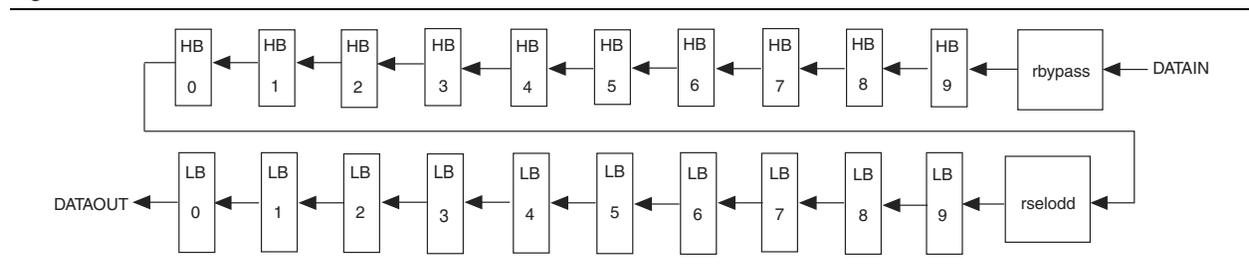


Mixed-Width Support

M9K memory blocks support mixed data widths. When using simple dual-port, true dual-port, or FIFO modes, mixed width support allows you to read and write different data widths to an M9K memory block. For more information about the different widths supported per memory mode, refer to “Memory Modes” on page 3-7.

Figure 5–25 shows the scan chain bit order sequence for one PLL post-scale counter in PLLs of Cyclone IV devices.

Figure 5–25. Scan Chain Bit Order



Charge Pump and Loop Filter

You can reconfigure the charge pump and loop filter settings to update the PLL bandwidth in real time. Table 5–8 through Table 5–10 list the possible settings for charge pump current (I_{CP}), loop filter resistor (R), and capacitor (C) values for PLLs of Cyclone IV devices.

Table 5–8. Charge Pump Bit Control

CP[2]	CP[1]	CP[0]	Setting (Decimal)
0	0	0	0
1	0	0	1
1	1	0	3
1	1	1	7

Table 5–9. Loop Filter Resistor Value Control

LFR[4]	LFR[3]	LFR[2]	LFR[1]	LFR[0]	Setting (Decimal)
0	0	0	0	0	0
0	0	0	1	1	3
0	0	1	0	0	4
0	1	0	0	0	8
1	0	0	0	0	16
1	0	0	1	1	19
1	0	1	0	0	20
1	1	0	0	0	24
1	1	0	1	1	27
1	1	1	0	0	28
1	1	1	1	0	30

Table 6–4. Number of VREF Pins Per I/O Bank for Cyclone IV E Devices (Part 2 of 2)

Device	EP4CE6			EP4CE10			EP4CE15						EP4CE22			EP4CE30			EP4CE40				EP4CE55			EP4CE75			EP4CE115	
	I/O Bank (1)	144-EQPF	256-UBGA	256-FBGA	144-EQPF	256-UBGA	256-FBGA	144-EQPF	164-MBGA	256-MBGA	256-UBGA	256-FBGA	484-FBGA	144-EQPF	256-UBGA	256-FBGA	324-FBGA	484-FBGA	780-FBGA	324-FBGA	484-UBGA	484-FBGA	780-FBGA	484-UBGA	484-FBGA	780-FBGA	484-UBGA	484-FBGA	780-FBGA	484-FBGA
8	1	1	1	1	1	1	2	2	2	2	2	2	1	1	1	4	4	4	4	4	4	4	2	2	2	3	3	3	3	3

Note to Table 6–4:

- (1) User I/O pins are used as inputs or outputs; clock input pins are used as inputs only; clock output pins are used as output only.

Table 6–5. Number of VREF Pins Per I/O Bank for Cyclone IV GX Devices

Device	4CGX15	4CGX22		4CGX30			4CGX50		4CGX75		4CGX110			4CGX150		
	I/O Bank (1)	169-FBGA	169-FBGA	324-FBGA	169-FBGA	324-FBGA	484-FBGA	484-FBGA	672-FBGA	484-FBGA	672-FBGA	484-FBGA	672-FBGA	896-FBGA	484-FBGA	672-FBGA
3	1	1			1	3		3		3		3				3
4	1	1			1	3		3		3		3				3
5	1	1			1	3		3		3		3				3
6	1	1			1	3		3		3		3				3
7	1	1			1	3		3		3		3				3
8 (2)	1	1			1	3		3		3		3				3

Notes to Table 6–5:

- (1) User I/O pins are used as inputs or outputs; clock input pins are used as inputs only; clock output pins are used as output only.
- (2) Bank 9 does not have VREF pin. If input pins with VREF I/O standards are used in bank 9 during user mode, it shares the VREF pin in bank 8.

Each Cyclone IV I/O bank has its own VCCIO pins. Each I/O bank can support only one VCCIO setting from among 1.2, 1.5, 1.8, 2.5, 3.0, or 3.3 V. Any number of supported single-ended or differential standards can be simultaneously supported in a single I/O bank, as long as they use the same VCCIO levels for input and output pins.

When designing LVTTTL/LVCMOS inputs with Cyclone IV devices, refer to the following guidelines:

- All pins accept input voltage (V_I) up to a maximum limit (3.6 V), as stated in the recommended operating conditions provided in the *Cyclone IV Device Datasheet* chapter.
- Whenever the input level is higher than the bank V_{CCIO} , expect higher leakage current.
- The LVTTTL/LVCMOS I/O standard input pins can only meet the V_{IH} and V_{IL} levels according to bank voltage level.

Voltage-referenced standards are supported in an I/O bank using any number of single-ended or differential standards, as long as they use the same V_{REF} and V_{CCIO} values. For example, if you choose to implement both SSTL-2 and SSTL-18 in your Cyclone IV devices, I/O pins using these standards—because they require different V_{REF} values—must be in different banks from each other. However, the same I/O bank can support SSTL-2 and 2.5-V LVCMOS with the V_{CCIO} set to 2.5 V and the V_{REF} set to 1.25 V.

 When using Cyclone IV devices as a receiver in 3.3-, 3.0-, or 2.5-V LVTTTL/LVCMOS systems, you are responsible for managing overshoot or undershoot to stay in the absolute maximum ratings and the recommended operating conditions, provided in the *Cyclone IV Device Datasheet* chapter.

 The PCI clamping diode is enabled by default in the Quartus II software for input signals with bank V_{CCIO} at 2.5, 3.0, or 3.3 V.

High-Speed Differential Interfaces

Cyclone IV devices can send and receive data through LVDS signals. For the LVDS transmitter and receiver, the input and output pins of Cyclone IV devices support serialization and deserialization through internal logic.

The BLVDS extends the benefits of LVDS to multipoint applications such as bidirectional backplanes. The loading effect and the need to terminate the bus at both ends for multipoint applications require BLVDS to drive out a higher current than LVDS to produce a comparable voltage swing. All the I/O banks of Cyclone IV devices support BLVDS for user I/O pins.

The RSDS and mini-LVDS standards are derivatives of the LVDS standard. The RSDS and mini-LVDS I/O standards are similar in electrical characteristics to LVDS, but have a smaller voltage swing and therefore provide increased power benefits and reduced electromagnetic interference (EMI).

The PPDS standard is the next generation of the RSDS standard introduced by National Semiconductor Corporation. Cyclone IV devices meet the National Semiconductor Corporation PPDS Interface Specification and support the PPDS standard for outputs only. All the I/O banks of Cyclone IV devices support the PPDS standard for output pins only.

The LVDS standard does not require an input reference voltage, but it does require a 100- Ω termination resistor between the two signals at the input buffer. An external resistor network is required on the transmitter side for the top and bottom I/O banks.

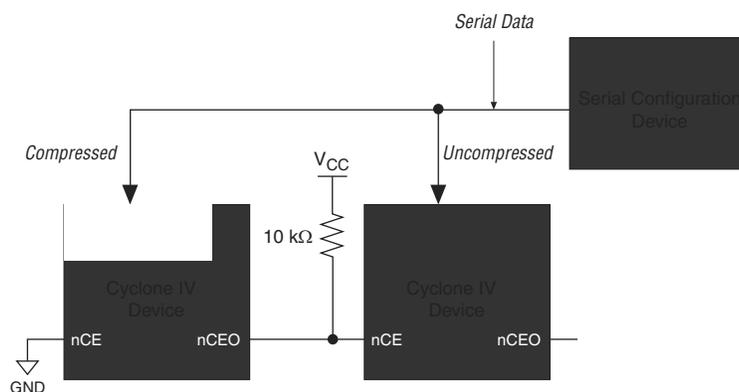
3. Click the **Configuration** tab.
4. Turn on **Generate compressed bitstreams**.
5. Click **OK**.
6. In the **Settings** dialog box, click **OK**.

You can enable compression when creating programming files from the **Convert Programming Files** dialog box. To enable compression, perform the following steps:

1. On the File menu, click **Convert Programming Files**.
2. Under **Output programming file**, select your desired file type from the **Programming file type** list.
3. If you select **Programmer Object File (.pof)**, you must specify the configuration device in the **Configuration device** list.
4. Under **Input files to convert**, select **SOE Data**.
5. Click **Add File** to browse to the Cyclone IV device SRAM object files (.sof).
6. In the **Convert Programming Files** dialog box, select the .pof you added to **SOE Data** and click **Properties**.
7. In the **SOE File Properties** dialog box, turn on the **Compression** option.

When multiple Cyclone IV devices are cascaded, you can selectively enable the compression feature for each device in the chain. Figure 8–1 shows a chain of two Cyclone IV devices. The first device has compression enabled and receives compressed bitstream from the configuration device. The second device has the compression feature disabled and receives uncompressed data. You can generate programming files for this setup in the **Convert Programming Files** dialog box.

Figure 8–1. Compressed and Uncompressed Configuration Data in the Same Configuration File



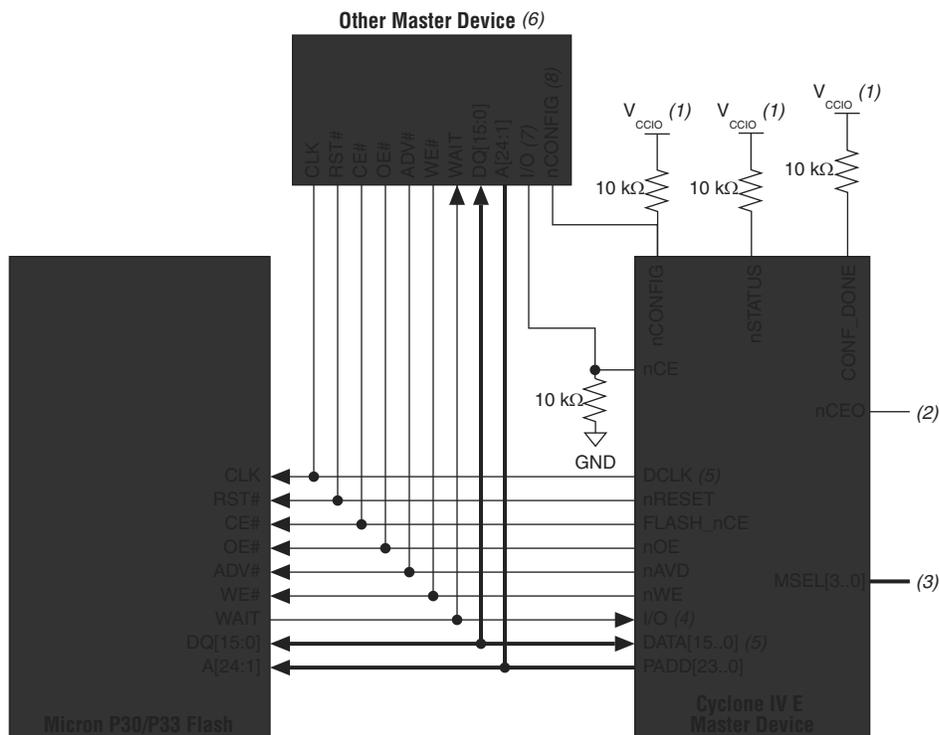
Configuration Requirement

This section describes Cyclone IV device configuration requirement and includes the following topics:

- “Power-On Reset (POR) Circuit” on page 8–4
- “Configuration File Size” on page 8–4
- “Power Up” on page 8–6

Figure 8-10 shows the AP configuration with multiple bus masters.

Figure 8-10. AP Configuration with Multiple Bus Masters



Notes to Figure 8-10:

- (1) Connect the pull-up resistors to the V_{CCIO} supply of the bank in which the pin resides.
- (2) The $nCEO$ pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.
- (3) The $MSEL$ pin settings vary for different configuration voltage standards and POR time. To connect $MSEL[3:0]$, refer to Table 8-5 on page 8-9. Connect the $MSEL$ pins directly to V_{CCA} or GND.
- (4) The AP configuration ignores the $WAIT$ signal during configuration mode. However, if you are accessing flash during user mode with user logic, you can optionally use the normal I/O to monitor the $WAIT$ signal from the Micron P30 or P33 flash.
- (5) When cascading Cyclone IV E devices in a multi-device AP configuration, connect the repeater buffers between the master device and slave devices for $DATA[15:0]$ and $DCLK$. All I/O inputs must maintain a maximum AC voltage of 4.1 V. The output resistance of the repeater buffers must fit the maximum overshoot equation outlined in “Configuration and JTAG Pin I/O Requirements” on page 8-5.
- (6) The other master device must fit the maximum overshoot equation outlined in “Configuration and JTAG Pin I/O Requirements” on page 8-5.
- (7) The other master device can control the AP configuration bus by driving the nCE to high with an output high on the I/O pin.
- (8) The other master device can pulse $nCONFIG$ if it is under system control and not tied to V_{CCIO} .

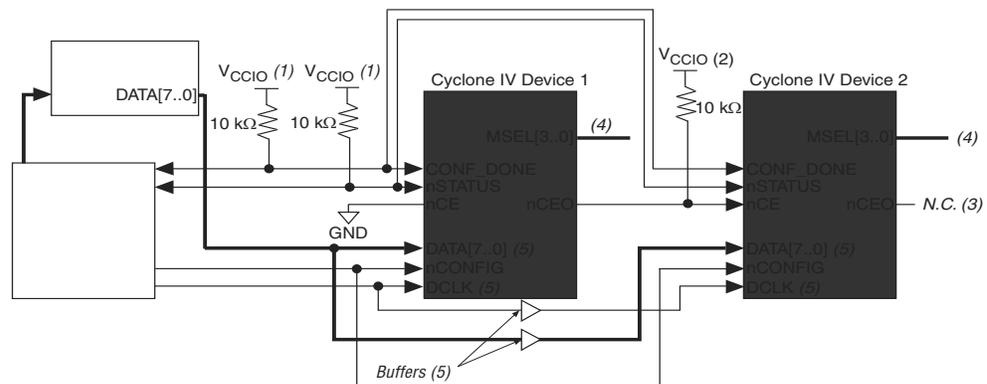
To ensure that DCLK and DATA[0] are not left floating at the end of the configuration, the MAX II device must drive them either high or low, whichever is convenient on your board. The DATA[0] pin is available as a user I/O pin after configuration. When you choose the FPP scheme in the Quartus II software, the DATA[0] pin is tri-stated by default in user mode and must be driven by the external host device. To change this default option in the Quartus II software, select the **Dual-Purpose Pins** tab of the **Device and Pin Options** dialog box.

The DCLK speed must be below the specified system frequency to ensure correct configuration. No maximum DCLK period exists, which means you can pause configuration by halting DCLK for an indefinite amount of time.

The external host device can also monitor the CONF_DONE and INIT_DONE pins to ensure successful configuration. The CONF_DONE pin must be monitored by the external device to detect errors and to determine when programming is complete. If all configuration data is sent, but CONF_DONE or INIT_DONE has not gone high, the external device must reconfigure the target device.

Figure 8-20 shows how to configure multiple devices with a MAX II device. This circuit is similar to the FPP configuration circuit for a single device, except the Cyclone IV devices are cascaded for multi-device configuration.

Figure 8-20. Multi-Device FPP Configuration Using an External Host



Notes to Figure 8-20:

- (1) The pull-up resistor must be connected to a supply that provides an acceptable input signal for all devices in the chain. V_{CC} must be high enough to meet the V_{IH} specification of the I/O on the device and the external host.
- (2) Connect the pull-up resistor to the V_{CCIO} supply voltage of the I/O bank in which the nCE pin resides.
- (3) The nCEO pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.
- (4) The MSEL pin settings vary for different configuration voltage standards and POR time. To connect the MSEL pins, refer to Table 8-4 on page 8-8 and Table 8-5 on page 8-9. Connect the MSEL pins directly to V_{CCA} or GND.
- (5) All I/O inputs must maintain a maximum AC voltage of 4.1 V. DATA[7..0] and DCLK must fit the maximum overshoot outlined in Equation 8-1 on page 8-5.

After the first device completes configuration in a multi-device configuration chain, its nCEO pin drives low to activate the nCE pin of the second device, which prompts the second device to begin configuration. The second device in the chain begins configuration in one clock cycle; therefore, the transfer of data destinations is transparent to the MAX II device. All other configuration pins (nCONFIG, nSTATUS,

Figure 8-34 shows the control register bit positions. Table 8-23 defines the control register bit contents. The numbers in Figure 8-34 show the bit position of a setting in a register. For example, bit number 35 is the enable bit for the watchdog timer.

Figure 8-34. Remote System Upgrade Control Register

38	37	36	35	34	33	12	11	0
Rsv2	Cd_early	Osc_int	Wd_en	Rsv1	Ru_address[21..0]	Wd_timer[11..0]		

Table 8-23. Remote System Upgrade Control Register Contents

Control Register Bit	Value	Definition
Wd_timer[11..0]	12'b000000000000	User watchdog time-out value (most significant 12 bits of 29-bit count value: {Wd_timer[11..0], 17'b1000})
Ru_address[21..0]	22'b00000000000000000000	Configuration address (most significant 22 bits of 24-bit boot address value: boot_address[23:0] = {Ru_address[21..0], 2'b0})
Rsv1	1'b0	Reserved bit
Wd_en	1'b1	User watchdog timer enable bit
Osc_int ⁽¹⁾	1'b1	Internal oscillator as startup state machine clock enable bit
Cd_early ⁽¹⁾	1'b1	Early CONF_DONE check
Rsv2	1'b1	Reserved bit

Note to Table 8-23:

(1) Option bit for the application configuration.

When enabled, the early CONF_DONE check (Cd_early) option bit ensures that there is a valid configuration at the boot address specified by the factory configuration and that it is of the proper size. If an invalid configuration is detected or the CONF_DONE pin is asserted too early, the device resets and then reconfigures the factory configuration image. The internal oscillator (as the startup state machine clock [Osc_int] option bit) ensures a functional startup clock to eliminate the hanging of startup. When all option bits are turned on, they provide complete coverage for the programming and startup portions of the application configuration. Altera recommends turning on both the Cd_early and Osc_int option bits.



The Cd_early and Osc_int option bits for the application configuration must be turned on by the factory configuration.

Remote System Upgrade Status Register

The remote system upgrade status register specifies the reconfiguration trigger condition. The various trigger and error conditions include:

- Cyclical redundancy check (CRC) error during application configuration
- nSTATUS assertion by an external device due to an error
- Cyclone IV device logic array triggers a reconfiguration cycle, possibly after downloading a new application configuration image

Table 8–25 lists the contents of previous state register 1 and previous state register 2 in the status register. The status register bit in Table 8–25 shows the bit positions in a 3-bit register. The previous state register 1 and previous state register 2 have the same bit definitions. The previous state register 1 reflects the current application configuration and the previous state register 2 reflects the previous application configuration.

Table 8–25. Remote System Upgrade Previous State Register 1 and Previous State Register 2 Contents in Status Register

Status Register Bit	Definition	Description
30	nCONFIG source	One-hot, active-high field that describes the reconfiguration source that caused the Cyclone IV device to leave the previous application configuration. If there is a tie, the higher bit order indicates precedence. For example, if nCONFIG and remote system upgrade nCONFIG reach the reconfiguration state machine at the same time, the nCONFIG precedes the remote system upgrade nCONFIG.
29	CRC error source	
28	nSTATUS source	
27	User watchdog timer source	
26	Remote system upgrade nCONFIG source	
25 : 24	Master state machine current state	The state of the master state machine during reconfiguration causes the Cyclone IV device to leave the previous application configuration.
23 : 0	Boot address	The address used by the configuration scheme to load the previous application configuration.

If a capture is inappropriately done while capturing a previous state before the system has entered remote update application configuration for the first time, a value outputs from the shift register to indicate that the capture is incorrectly called.

Remote System Upgrade State Machine

The remote system upgrade control and update registers have identical bit definitions, but serve different roles (Table 8–22 on page 8–75). While both registers can only be updated when the device is loaded with a factory configuration image, the update register writes are controlled by the user logic, and the control register writes are controlled by the remote system upgrade state machine.

In factory configurations, the user logic should send the option bits (Cd_early and Osc_int), the configuration address, and watchdog timer settings for the next application configuration bit to the update register. When the logic array configuration reset (RU_nCONFIG) goes high, the remote system upgrade state machine updates the control register with the contents of the update register and starts system reconfiguration from the new application page.



To ensure the successful reconfiguration between the pages, assert the RU_nCONFIG signal for a minimum of 250 ns. This is equivalent to strobing the reconfig input of the ALTREMOTE_UPDATE megafunction high for a minimum of 250 ns.

If there is an error or reconfiguration trigger condition, the remote system upgrade state machine directs the system to load a factory or application configuration (based on mode and error condition) by setting the control register accordingly.

Table 8–26 lists the contents of the control register after such an event occurs for all possible error or trigger conditions.

In some applications, it is necessary for a device to wake up very quickly to begin operation. Cyclone IV devices offer the Fast-On feature to support fast wake-up time applications. The MSEL pin settings determine the POR time (t_{POR}) of the device.

- For more information about the MSEL pin settings, refer to the *Configuration and Remote System Upgrades in Cyclone IV Devices* chapter.
- For more information about the POR specifications, refer to the *Cyclone IV Device Datasheet* chapter.

Document Revision History

Table 11-3 lists the revision history for this chapter.

Table 11-3. Document Revision History

Date	Version	Changes
May 2013	1.3	Updated Note (4) in Table 11-1.
July 2010	1.2	<ul style="list-style-type: none"> ■ Updated for the Quartus II software version 10.0 release. ■ Updated “I/O Pins Remain Tri-stated During Power-Up” section. ■ Updated Table 11-1.
February 2010	1.1	Updated Table 11-1 and Table 11-2 for the Quartus II software version 9.1 SP1 release.
November 2009	1.0	Initial release.

Section I. Transceivers

This section provides a complete overview of all features relating to the Cyclone[®] IV device transceivers. This section includes the following chapters:

- Chapter 1, Cyclone IV Transceivers Architecture
- Chapter 2, Cyclone IV Reset Control and Power Down
- Chapter 3, Cyclone IV Dynamic Reconfiguration

Revision History

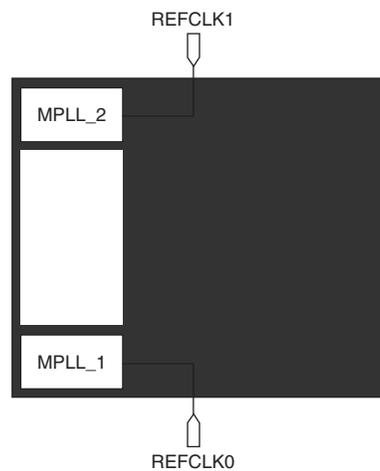
Refer to the chapter for its own specific revision history. For information about when the chapter was updated, refer to the Chapter Revision Dates section, which appears in the complete handbook.

Input Reference Clocking

When used for transceiver, the left PLLs synthesize the input reference clock to generate the required clocks for the transceiver channels. Figure 1–25 and Figure 1–26 show the sources of input reference clocks for PLLs used in the transceiver operation.

 Clock output from PLLs in the FPGA core cannot feed into PLLs used by the transceiver as input reference clock.

Figure 1–25. PLL Input Reference Clocks in Transceiver Operation for F324 and Smaller Packages ^{(1), (2)}



Notes to Figure 1–25:

- (1) The REFCLK0 and REFCLK1 pins are dual-purpose CLK, REFCLK, or DIFFCLK pins that reside in banks 3A and 8A respectively.
- (2) Using any clock input pins other than the designated REFCLK pins as shown here to drive the MPLLs may have reduced jitter performance.

User Reset and Power-Down Signals

Each transceiver channel in the Cyclone IV GX device has individual reset signals to reset its physical coding sublayer (PCS) and physical medium attachment (PMA). The transceiver block also has a power-down signal that affects the multipurpose phase-locked loops (PLLs), general purpose PLLs, and all the channels in the transceiver block.



All reset and power-down signals are asynchronous.

Table 2–1 lists the reset signals available for each transceiver channel.

Table 2–1. Transceiver Channel Reset Signals

Signal	ALTGX MegaWizard Plug-In Manager Configurations	Description
tx_digitalreset ⁽¹⁾	<ul style="list-style-type: none"> ■ Transmitter Only ■ Receiver and Transmitter 	<p>Provides asynchronous reset to all digital logic in the transmitter PCS, including the XAUI transmit state machine.</p> <p>The minimum pulse width for this signal is two parallel clock cycles.</p>
rx_digitalreset ⁽¹⁾	<ul style="list-style-type: none"> ■ Receiver Only ■ Receiver and Transmitter 	<p>Resets all digital logic in the receiver PCS, including:</p> <ul style="list-style-type: none"> ■ XAUI receiver state machines ■ GIGE receiver state machines ■ XAUI channel alignment state machine ■ BIST-PRBS verifier ■ BIST-incremental verifier <p>The minimum pulse width for this signal is two parallel clock cycles.</p>
rx_analogreset	<ul style="list-style-type: none"> ■ Receiver Only ■ Receiver and Transmitter 	<p>Resets the receiver CDR present in the receiver channel.</p> <p>The minimum pulse width is two parallel clock cycles.</p>

Note to Table 2–1:

- (1) Assert this signal until the clocks coming out of the multipurpose PLL and receiver CDR are stabilized. Stable parallel clocks are essential for proper operation of transmitter and receiver phase-compensation FIFOs in the PCS.

Read Transaction

If you want to read the existing values from a specific channel connected to the ALTGX_RECONFIG instance, observe the corresponding byte positions of the PMA control output port after the read transaction is completed.

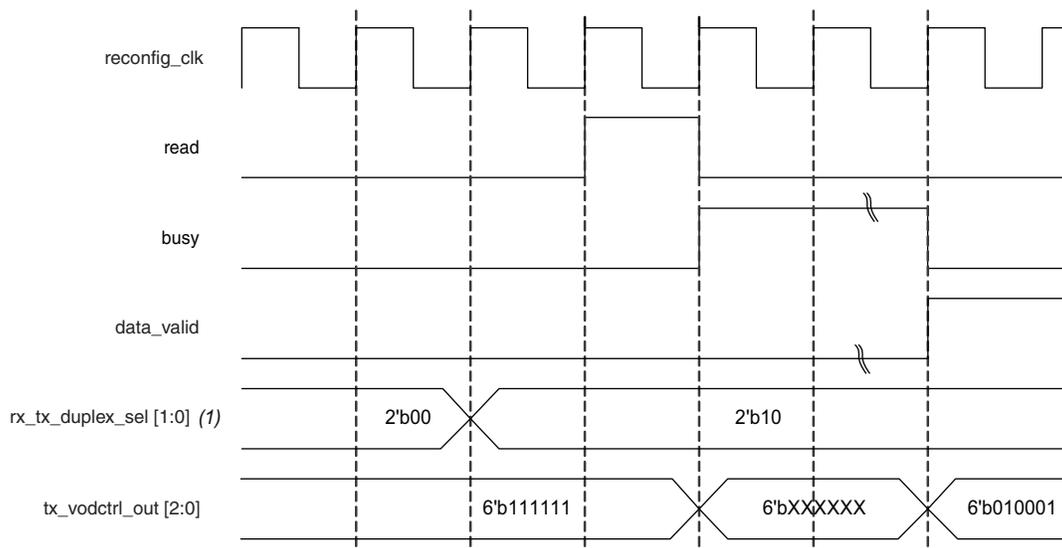
For example, if the number of channels controlled by the ALTGX_RECONFIG is two, the tx_vodctrl_out is 6 bits wide. The tx_vodctrl_out[2:0] signal corresponds to channel 1 and the tx_vodctrl_out[5:3] signal corresponds to channel 2.

To complete a read transaction to the V_{OD} values of the second channel, perform the following steps:

1. Before you initiate a read transaction, set the rx_tx_duplex_sel port to **2'b10** so that only the transmit PMA controls are read from the transceiver channel.
2. Ensure that the busy signal is low before you start a read transaction.
3. Assert the read signal for one reconfig_clk clock cycle. This initiates the read transaction.
4. The busy output status signal is asserted high to indicate that the dynamic reconfiguration controller is busy reading the PMA control settings.
5. When the read transaction has completed, the busy signal goes low. The data_valid signal is asserted, indicating that the data available at the read control signal is valid.
6. To read the current V_{OD} values in channel 2, observe the values in tx_vodctrl_out[5:3].

In the waveform example shown in Figure 3-7, the transmit V_{OD} settings written in channels 1 and 2 prior to the read transaction are 3'b001 and 3'b010, respectively.

Figure 3-7. Read Transaction Waveform—Use the same control signal for all the channels Option Enabled



Note to Figure 3-7:

- (1) In this waveform example, you want to read from only the transmitter portion of all the channels.



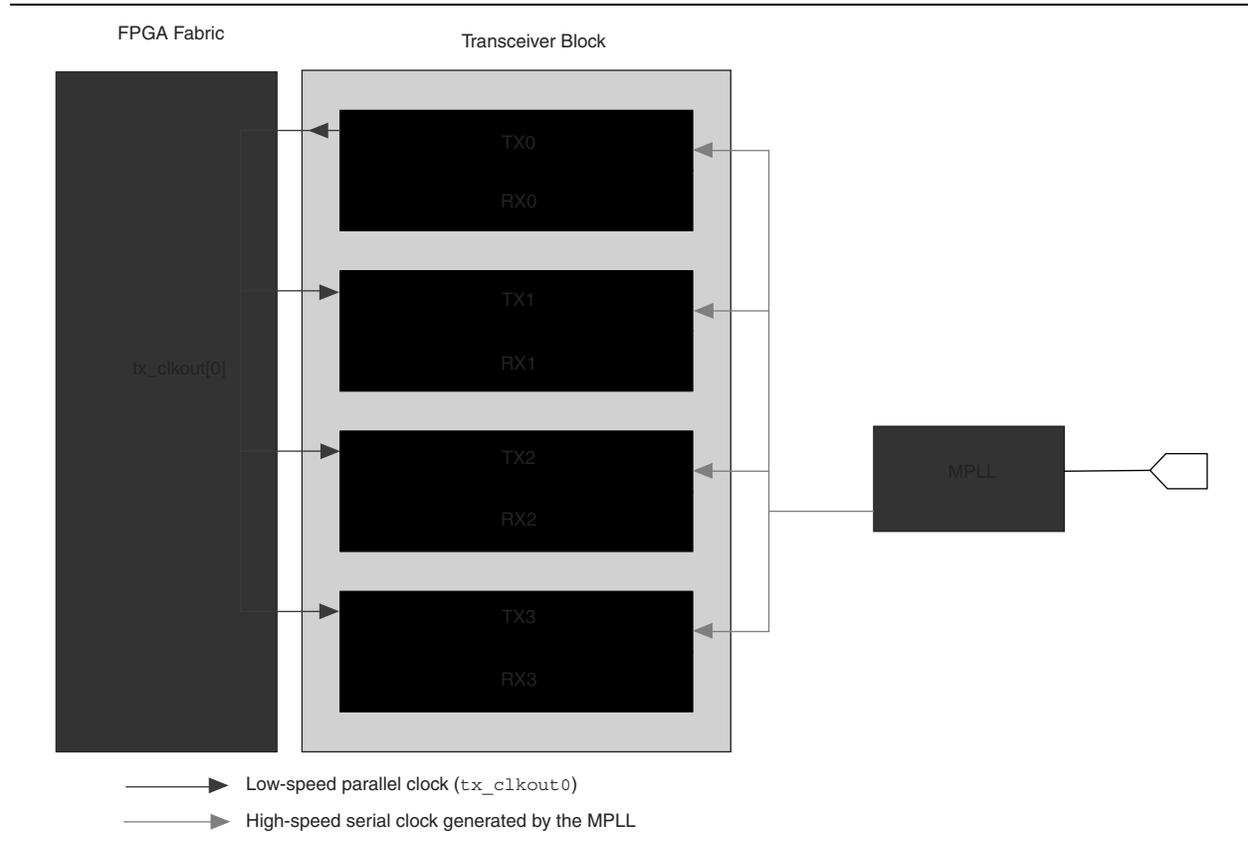
Simultaneous write and read transactions are not allowed.

Option 1: Share a Single Transmitter Core Clock Between Transmitters

- Enable this option if you want tx_clkout of the first channel (channel 0) of the transceiver block to provide the write clock to the Transmitter Phase Compensation FIFOs of the remaining channels in the transceiver block.
- This option is typically enabled when all the channels of a transceiver block have the same functional mode and data rate and are reconfigured to the identical functional mode and data rate.

Figure 3–11 shows the sharing of channel 0's tx_clkout between all four regular channels of a transceiver block.

Figure 3–11. Option 1 for Transmitter Core Clocking (Channel Reconfiguration Mode)



 For more information about the supported maximum clock rate, device and pin planning, IP implementation, and device termination, refer to *Section III: System Performance Specifications* of the *External Memory Interfaces Handbook*.

 Actual achievable frequency depends on design- and system-specific factors. Perform HSPICE/IBIS simulations based on your specific design and system setup to determine the maximum achievable frequency in your system.

High-Speed I/O Specifications

Table 1–31 through Table 1–36 list the high-speed I/O timing for Cyclone IV devices. For definitions of high-speed timing specifications, refer to “Glossary” on page 1–37.

Table 1–31. RSDS Transmitter Timing Specifications for Cyclone IV Devices ^{(1), (2), (4)} (Part 1 of 2)

Symbol	Modes	C6			C7, I7			C8, A7			C8L, I8L			C9L			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
f_{HSCLK} (input clock frequency)	×10	5	—	180	5	—	155.5	5	—	155.5	5	—	155.5	5	—	132.5	MHz
	×8	5	—	180	5	—	155.5	5	—	155.5	5	—	155.5	5	—	132.5	MHz
	×7	5	—	180	5	—	155.5	5	—	155.5	5	—	155.5	5	—	132.5	MHz
	×4	5	—	180	5	—	155.5	5	—	155.5	5	—	155.5	5	—	132.5	MHz
	×2	5	—	180	5	—	155.5	5	—	155.5	5	—	155.5	5	—	132.5	MHz
	×1	5	—	360	5	—	311	5	—	311	5	—	311	5	—	265	MHz
Device operation in Mbps	×10	100	—	360	100	—	311	100	—	311	100	—	311	100	—	265	Mbps
	×8	80	—	360	80	—	311	80	—	311	80	—	311	80	—	265	Mbps
	×7	70	—	360	70	—	311	70	—	311	70	—	311	70	—	265	Mbps
	×4	40	—	360	40	—	311	40	—	311	40	—	311	40	—	265	Mbps
	×2	20	—	360	20	—	311	20	—	311	20	—	311	20	—	265	Mbps
	×1	10	—	360	10	—	311	10	—	311	10	—	311	10	—	265	Mbps
t_{DUTY}	—	45	—	55	45	—	55	45	—	55	45	—	55	45	—	55	%
Transmitter channel-to-channel skew (TCCS)	—	—	—	200	—	—	200	—	—	200	—	—	200	—	—	200	ps
Output jitter (peak to peak)	—	—	—	500	—	—	500	—	—	550	—	—	600	—	—	700	ps
t_{RISE}	20 – 80%, $C_{\text{LOAD}} = 5 \text{ pF}$	—	500	—	—	500	—	—	500	—	—	500	—	—	500	—	ps
t_{FALL}	20 – 80%, $C_{\text{LOAD}} = 5 \text{ pF}$	—	500	—	—	500	—	—	500	—	—	500	—	—	500	—	ps