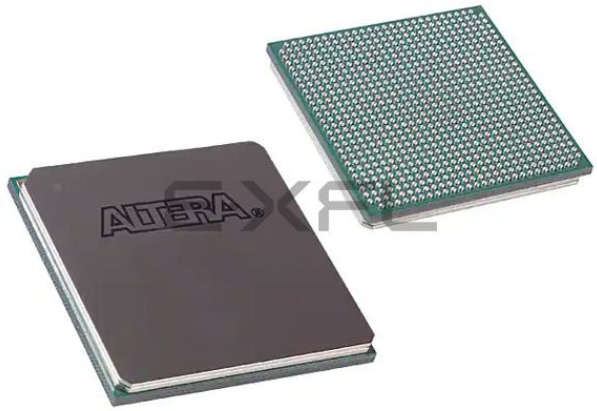


Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	3491
Number of Logic Elements/Cells	55856
Total RAM Bits	2396160
Number of I/O	374
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	780-BGA
Supplier Device Package	780-FBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep4ce55f29c7n

Document Revision History

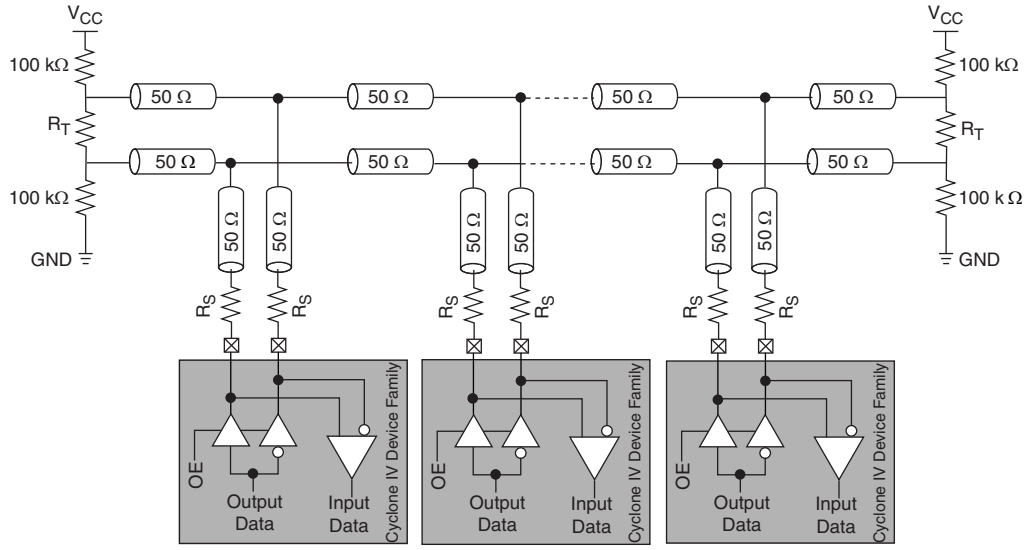
Table 4–3 lists the revision history for this chapter.

Table 4–3. Document Revision History

Date	Version	Changes
February 2010	1.1	Added Cyclone IV E devices in Table 4–1 for the Quartus II software version 9.1 SP1 release.
November 2009	1.0	Initial release.

Figure 6–14 shows a typical BLVDS topology with multiple transmitter and receiver pairs.

Figure 6–14. BLVDS Topology with Cyclone IV Devices Transmitters and Receivers



The BLVDS I/O standard is supported on the top, bottom, and right I/O banks of Cyclone IV devices. The BLVDS transmitter uses two single-ended output buffers with the second output buffer programmed as inverted, while the BLVDS receiver uses a true LVDS input buffer. The transmitter and receiver share the same pins. An output-enabled (OE) signal is required to tristate the output buffers when the LVDS input buffer receives a signal.

For more information, refer to the *Cyclone IV Device Datasheet* chapter.

Designing with BLVDS

The BLVDS bidirectional communication requires termination at both ends of the bus in BLVDS. The termination resistor (R_T) must match the bus differential impedance, which in turn depends on the loading on the bus. Increasing the load decreases the bus differential impedance. With termination at both ends of the bus, termination is not required between the two signals at the input buffer. A single series resistor (R_S) is required at the output buffer to match the output buffer impedance to the transmission line impedance. However, this series resistor affects the voltage swing at the input buffer. The maximum data rate achievable depends on many factors.

Altera recommends that you perform simulation using the IBIS model while considering factors such as bus loading, termination values, and output and input buffer location on the bus to ensure that the required performance is achieved.

For more information about BLVDS interface support in Altera devices, refer to *AN 522: Implementing Bus LVDS Interface in Supported Altera Device Families*.

Table 7–1 lists the number of DQS or DQ groups supported on each side of the Cyclone IV GX device.

Table 7–1. Cyclone IV GX Device DQS and DQ Bus Mode Support for Each Side of the Device

Device	Package	Side	Number ×8 Groups	Number ×9 Groups	Number ×16 Groups	Number ×18 Groups	Number ×32 Groups	Number ×36 Groups
EP4CGX15	169-pin FBGA	Right	1	0	0	0	—	—
		Top (1)	1	0	0	0	—	—
		Bottom (2)	1	0	0	0	—	—
EP4CGX22 EP4CGX30	169-pin FBGA	Right	1	0	0	0	—	—
		Top (1)	1	0	0	0	—	—
		Bottom (2)	1	0	0	0	—	—
	324-pin FBGA	Right	2	2	1	1	—	—
		Top	2	2	1	1	—	—
		Bottom	2	2	1	1	—	—
	484-pin FBGA (3)	Right	4	2	2	2	1	1
		Top	4	2	2	2	1	1
		Bottom	4	2	2	2	1	1
EP4CGX50 EP4CGX75	484-pin FBGA	Right	4	2	2	2	1	1
		Top	4	2	2	2	1	1
		Bottom	4	2	2	2	1	1
	672-pin FBGA	Right	4	2	2	2	1	1
		Top	4	2	2	2	1	1
		Bottom	4	2	2	2	1	1
EP4CGX110 EP4CGX150	484-pin FBGA	Right	4	2	2	2	1	1
		Top	4	2	2	2	1	1
		Bottom	4	2	2	2	1	1
	672-pin FBGA	Right	4	2	2	2	1	1
		Top	4	2	2	2	1	1
		Bottom	4	2	2	2	1	1
	896-pin FBGA	Right	6	3	2	2	1	1
		Top	6	3	3	3	1	1
		Bottom	6	3	3	3	1	1

Notes to Table 7–1:

(1) Some of the DQ pins can be used as RUP and RDN pins. You cannot use these groups if you are using these pins as RUP and RDN pins for OCT calibration.

(2) Some of the DQ pins can be used as RUP pins while the DM pins can be used as RDN pins. You cannot use these groups if you are using the RUP and RDN pins for OCT calibration.

(3) Only available for EP4CGX30 device.

8. Configuration and Remote System Upgrades in Cyclone IV Devices

CYIV-51008-1.7

This chapter describes the configuration and remote system upgrades in Cyclone® IV devices. Cyclone IV (Cyclone IV GX and Cyclone IV E) devices use SRAM cells to store configuration data. You must download the configuration data to Cyclone IV devices each time the device powers up because SRAM memory is volatile.

Cyclone IV devices are configured using one of the following configuration schemes:

- Active serial (AS)
- Active parallel (AP) (supported in Cyclone IV E devices only)
- Passive serial (PS)
- Fast passive parallel (FPP) (not supported in EP4CGX15, EP4CGX22, and EP4CGX30 [except for the F484 package] devices)
- JTAG

Cyclone IV devices offer the following configuration features:

- Configuration data decompression ("Configuration Data Decompression" on page 8–2)
- Remote system upgrade ("Remote System Upgrade" on page 8–69)

System designers face difficult challenges, such as shortened design cycles, evolving standards, and system deployments in remote locations. Cyclone IV devices help overcome these challenges with inherent re-programmability and dedicated circuitry to perform remote system upgrades. Remote system upgrades help deliver feature enhancements and bug fixes without costly recalls, reduced time-to-market, and extended product life.

Configuration

This section describes Cyclone IV device configuration and includes the following topics:

Table 8–2. Uncompressed Raw Binary File (.rbf) Sizes for Cyclone IV Devices (Part 2 of 2)

Device		Data Size (bits)
Cyclone IV GX	EP4CGX15	3,805,568
	EP4CGX22	7,600,040
	EP4CGX30	7,600,040
		22,010,888 (1)
	EP4CGX50	22,010,888
	EP4CGX75	22,010,888
	EP4CGX110	39,425,016
	EP4CGX150	39,425,016

Note to Table 8–2:

(1) Only for the F484 package.

Use the data in Table 8–2 to estimate the file size before design compilation. Different configuration file formats, such as Hexadecimal (**.hex**) or Tabular Text File (**.ttf**) formats, have different file sizes. However, for any specific version of the Quartus II software, any design targeted for the same device has the same uncompressed configuration file size. If you use compression, the file size varies after each compilation, because the compression ratio depends on the design.

For more information about setting device configuration options or creating configuration files, refer to the *Software Settings* section in volume 2 of the *Configuration Handbook*.

Configuration and JTAG Pin I/O Requirements

Cyclone IV devices are manufactured using the TSMC 60-nm low-k dielectric process. Although Cyclone IV devices use TSMC 2.5-V transistor technology in the I/O buffers, the devices are compatible and able to interface with 2.5, 3.0, and 3.3-V configuration voltage standards by following specific requirements.

All I/O inputs must maintain a maximum AC voltage of 4.1 V. When using a serial configuration device in an AS configuration scheme, you must connect a 25-Ω series resistor for the DATA[0] pin. When cascading the Cyclone IV device family in a multi-device configuration for AS, AP, FPP, and PS configuration schemes, you must connect the repeater buffers between the master and slave devices for the DATA and DCLK pins. When using the JTAG configuration scheme in a multi-device configuration, connect 25-Ω resistors on both ends of the TDO-TDI path if the TDO output driver is a non-Cyclone IV device.

The output resistance of the repeater buffers and the TDO path for all cases must fit the maximum overshoot equation shown in Equation 8–1.

Equation 8–1. (1)

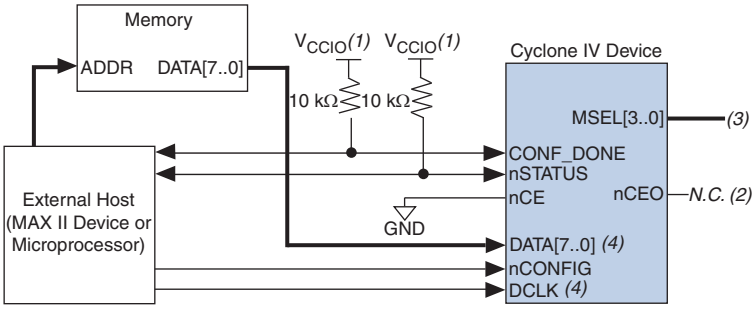
$$0.8Z_O \leq R_E \leq 1.8Z_O$$

Note to Equation 8–1:

(1) Z_O is the transmission line impedance and R_E is the equivalent resistance of the output buffer.

the device, must be stored in the external host device. Figure 8–19 shows the configuration interface connections between the Cyclone IV devices and an external device for single-device configuration.

Figure 8–19. Single-Device FPP Configuration Using an External Host



Notes to Figure 8–19:

(1) Connect the pull-up resistor to a supply that provides an acceptable input signal for the device. V_{CC} must be high enough to meet the V_{IH} specification of the I/O on the device and the external host.

(2) The nCEO pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.

(3) The MSEL pin settings vary for different configuration voltage standards and POR time. To connect the MSEL pins, refer to Table 8–4 on page 8–8 and Table 8–5 on page 8–9. Connect the MSEL pins directly to V_{CCA} or GND.

(4) All I/O inputs must maintain a maximum AC voltage of 4.1 V. DATA[7..0] and DCLK must fit the maximum overshoot outlined in Equation 8–1 on page 8–5.

After nSTATUS is released, the device is ready to receive configuration data and the configuration stage begins. When nSTATUS is pulled high, the external host device places the configuration data one byte at a time on the DATA[7..0] pins.

Cyclone IV devices receive configuration data on the DATA[7..0] pins and the clock is received on the DCLK pin. Data is latched into the device on the rising edge of DCLK. Data is continuously clocked into the target device until CONF_DONE goes high. The CONF_DONE pin goes high one byte early in FPP configuration mode. The last byte is required for serial configuration (AS and PS) modes.

Two DCLK falling edges are required after CONF_DONE goes high to begin initialization of the device.

Supplying a clock on CLKUSR does not affect the configuration process. After the CONF_DONE pin goes high, CLKUSR is enabled after the time specified as t_{CD2CU}. After this time period elapses, Cyclone IV devices require 3,192 clock cycles to initialize properly and enter user mode. For more information about the supported CLKUSR f_{MAX} value for Cyclone IV devices, refer to Table 8–13 on page 8–44.

The INIT_DONE pin is released and pulled high when initialization is complete. The external host device must be able to detect this low-to-high transition, which signals the device has entered user mode. When initialization is complete, the device enters user mode. In user mode, the user I/O pins no longer have weak pull-up resistors and function as assigned in your design.

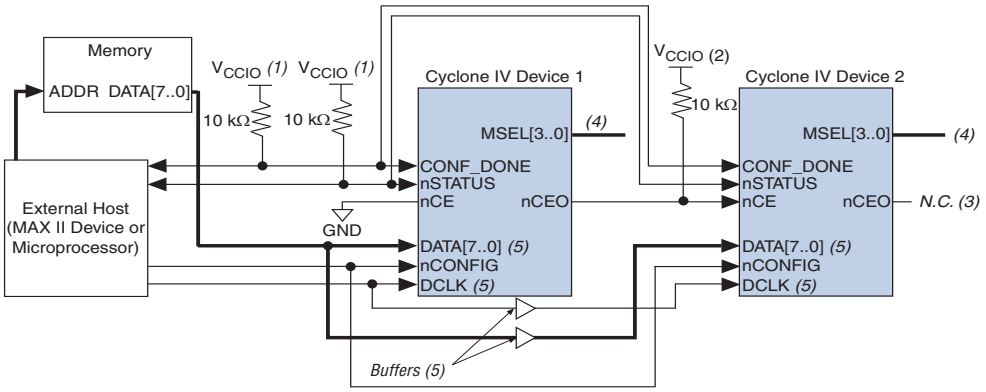
To ensure that DCLK and DATA[0] are not left floating at the end of the configuration, the MAX II device must drive them either high or low, whichever is convenient on your board. The DATA[0] pin is available as a user I/O pin after configuration. When you choose the FPP scheme in the Quartus II software, the DATA[0] pin is tri-stated by default in user mode and must be driven by the external host device. To change this default option in the Quartus II software, select the **Dual-Purpose Pins** tab of the **Device and Pin Options** dialog box.

The DCLK speed must be below the specified system frequency to ensure correct configuration. No maximum DCLK period exists, which means you can pause configuration by halting DCLK for an indefinite amount of time.

The external host device can also monitor the CONF_DONE and INIT_DONE pins to ensure successful configuration. The CONF_DONE pin must be monitored by the external device to detect errors and to determine when programming is complete. If all configuration data is sent, but CONF_DONE or INIT_DONE has not gone high, the external device must reconfigure the target device.

Figure 8–20 shows how to configure multiple devices with a MAX II device. This circuit is similar to the FPP configuration circuit for a single device, except the Cyclone IV devices are cascaded for multi-device configuration.

Figure 8–20. Multi-Device FPP Configuration Using an External Host



Notes to Figure 8–20:

(1) The pull-up resistor must be connected to a supply that provides an acceptable input signal for all devices in the chain. V_{CC} must be high enough to meet the V_{IH} specification of the I/O on the device and the external host.

(2) Connect the pull-up resistor to the V_{CCIO} supply voltage of the I/O bank in which the nCE pin resides.

(3) The nCEO pin is left unconnected or used as a user I/O pin when it does not feed the nCE pin of another device.

(4) The MSEL pin settings vary for different configuration voltage standards and POR time. To connect the MSEL pins, refer to Table 8–4 on page 8–8 and Table 8–5 on page 8–9. Connect the MSEL pins directly to V_{CCA} or GND.

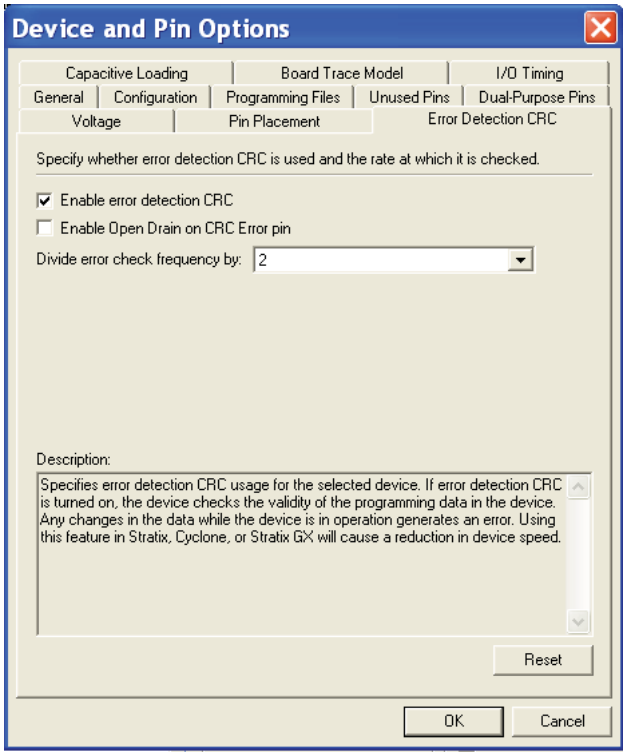
(5) All I/O inputs must maintain a maximum AC voltage of 4.1 V. DATA[7..0] and DCLK must fit the maximum overshoot outlined in Equation 8–1 on page 8–5.

After the first device completes configuration in a multi-device configuration chain, its nCEO pin drives low to activate the nCE pin of the second device, which prompts the second device to begin configuration. The second device in the chain begins configuration in one clock cycle; therefore, the transfer of data destinations is transparent to the MAX II device. All other configuration pins (nCONFIG, nSTATUS,

The divisor value divides the frequency of the configuration oscillator output clock. This output clock is used as the clock source for the error detection process.

8. Click **OK**.

Figure 9–2. Enabling the Error Detection CRC Feature in the Quartus II Software



Accessing Error Detection Block Through User Logic

The error detection circuit stores the computed 32-bit CRC signature in a 32-bit register, which is read out by user logic from the core. The `cycloneiv_crcblock` primitive is a WYSIWYG component used to establish the interface from the user logic to the error detection circuit. The `cycloneiv_crcblock` primitive atom contains the input and output ports that must be included in the atom. To access the logic array, the `cycloneiv_crcblock` WYSIWYG atom must be inserted into your design.

Chapter 3. Cyclone IV Dynamic Reconfiguration

Word Aligner

Figure 1–16 shows the word aligner block diagram. The word aligner receives parallel data from the deserializer and restores the word boundary based on a pre-defined alignment pattern that must be received during link synchronization. The word aligner supports three operational modes as listed in Table 1–3.

Figure 1–16. Word Aligner Block Diagram

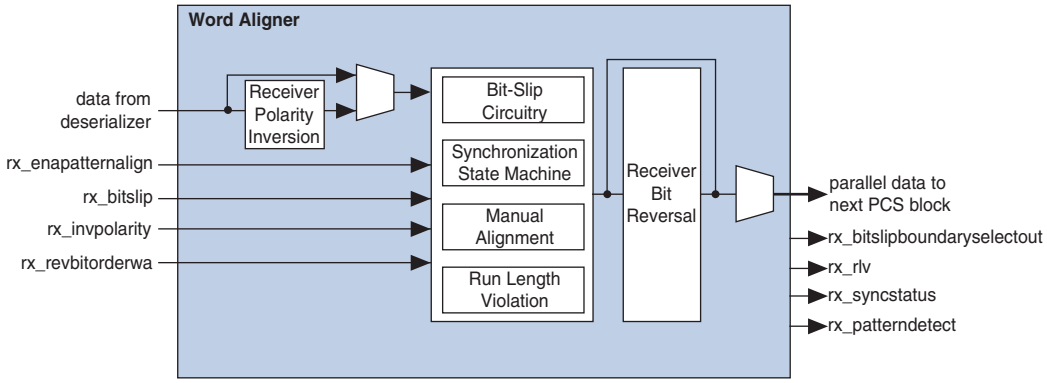


Table 1–3. Word Aligner Modes

Modes	PMA-PCS Interface Widths	Allowed Word Alignment Pattern Lengths
Manual Alignment	8-bit	16 bits
	10-bit	7 or 10 bits
Bit-Slip	8-bit	16 bits
	10-bit	7 or 10 bits
Automatic Synchronization State Machine	10-bit	7 or 10 bits

Manual Alignment Mode

In manual alignment mode, the `rx_enapatternalign` port controls the word aligner with either an 8- or 10-bit data width setting.

The 8-bit word aligner is edge-sensitive to the `rx_enapatternalign` signal. A rising edge on `rx_enapatternalign` signal after deassertion of the `rx_digitalreset` signal triggers the word aligner to look for the word alignment pattern in the received data stream. It updates the word boundary if it finds the word alignment pattern in a new word boundary. Any word alignment pattern received thereafter in a different word boundary causes the word aligner to re-align to the new word boundary only if there is a rising edge in the `rx_enapatternalign` signal.

The 10-bit word aligner is level-sensitive to the `rx_enapatternalign` signal. The word aligner looks for the programmed 7-bit or 10-bit word alignment pattern or its complement in the received data stream, if the `rx_enapatternalign` signal is held high. It updates the word boundary if it finds the word alignment pattern in a new word boundary. If the `rx_enapatternalign` signal is deasserted, the word aligner maintains the current word boundary even when it receives the word alignment pattern in a new word boundary.

Table 1–4 lists the synchronization state machine parameters for the word aligner in this mode.

Table 1–4. Synchronization State Machine Parameters

Parameter	Allowed Values
Number of erroneous code groups received to lose synchronization	1–64
Number of continuous good code groups received to reduce the error count by one	1–256

After deassertion of the `rx_digitalreset` signal in automatic synchronization state machine mode, the word aligner starts looking for the synchronization code groups, word alignment pattern or its complement in the received data stream. When the programmed number of valid synchronization code groups or ordered sets are received, the `rx_syncstatus` signal is driven high to indicate that synchronization is acquired. The `rx_syncstatus` signal is constantly driven high until the programmed number of erroneous code groups are received without receiving intermediate good groups; after which the `rx_syncstatus` signal is driven low. The word aligner indicates loss of synchronization (`rx_syncstatus` signal remains low) until the programmed number of valid synchronization code groups are received again.

In addition to restoring word boundaries, the word aligner supports the following features:

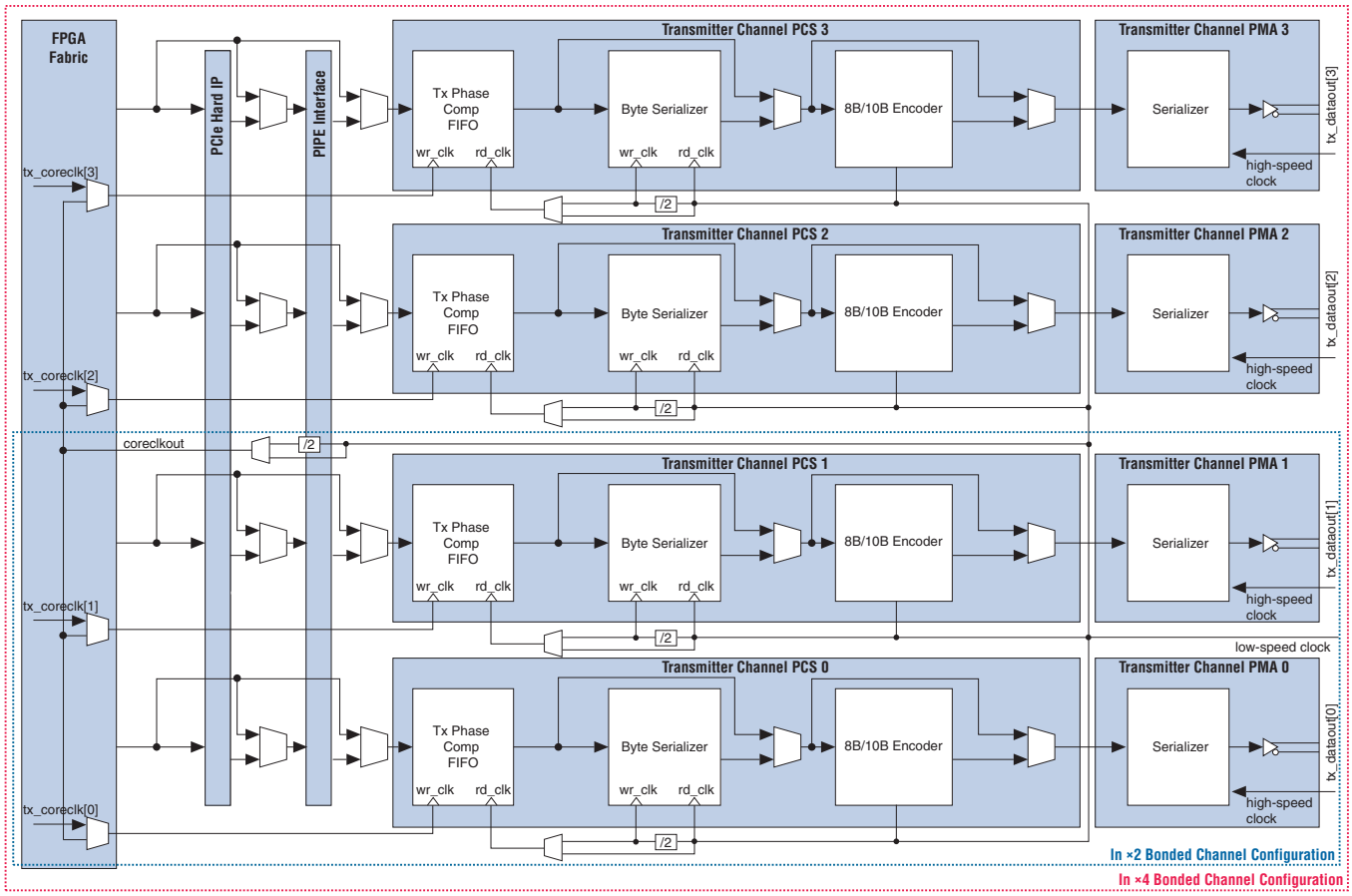
- Programmable run length violation detection—detects consecutive 1s or 0s in the data stream, and asserts run length violation signal (`rx_rlv`) when a preset run length threshold (maximum number of consecutive 1s or 0s) is detected. The `rx_rlv` signal in each channel is clocked by its parallel recovered clock and is asserted for a minimum of two recovered clock cycles to ensure that the FPGA fabric clock can latch the `rx_rlv` signal reliably because the FPGA fabric clock might have phase differences, ppm differences (in asynchronous systems), or both, with the recovered clock. Table 1–5 lists the run length violation circuit detection capabilities.

Table 1–5. Run Length Violation Circuit Detection Capabilities

Supported Data Width	Detector Range		Increment Step Settings
	Minimum	Maximum	
8-bit	4	128	4
10-bit	5	160	5

When the byte serializer is enabled, the common bonded low-speed clock frequency is halved before feeding to the read clock of TX phase compensation FIFO. The common bonded low-speed clock is available in FPGA fabric as `coreclkout` port, which can be used in FPGA fabric to send transmitter data and control signals to the bonded channels.

Figure 1–38. Transmitter Only Datapath Clocking in Bonded Channel Configuration



Bonded channel configuration is not available for Receiver Only channel operation because each of the channels are individually clocked by its recovered clock.

Clock Frequency Compensation

In Serial RapidIO mode, the rate match FIFO compensates up to ±100 ppm (200 ppm total) difference between the upstream transmitter and the local receiver reference clock.

Rate matcher is an optional block available for selection in Serial RapidIO mode. However, this block is not fully compliant to the SRIO specification. When enabled in the ALTGX MegaWizard Plug-In Manager, the default settings are:

- control pattern 1 = K28.5 with positive disparity
- skip pattern 1 = K29.7 with positive disparity
- control pattern 2 = K28.5 with negative disparity
- skip pattern 2 = K29.7 with negative disparity

When enabled, the rate match FIFO operation begins after the link is synchronized (indicated by assertion of `rx_syncstatus` from the word aligner). When the rate matcher receives either of the two 10-bit control patterns followed by the respective 10-bit skip pattern, it inserts or deletes the 10-bit skip pattern as necessary to avoid the rate match FIFO from overflowing or under-running. The rate match FIFO can delete/insert a maximum of one skip pattern from a cluster.

The rate match FIFO may perform multiple insertion or deletion if the ppm difference is more than the allowable 200 ppm range. Ensure that the ppm difference in your system is less than 200 ppm.

XAUI Mode

XAUI mode provides the bonded (×4) transceiver channel datapath configuration for XAUI protocol implementation. The Cyclone IV GX transceivers configured in XAUI mode provides the following functions:

- XGMII-to-PCS code conversion at transmitter datapath
- PCS-to-XGMII code conversion at receiver datapath
- channel deskewing of four lanes
- 8B/10B encoding and decoding
- IEEE P802.3ae-compliant synchronization state machine
- clock rate compensation

The XAUI is a self-managed interface to transparently extend the physical reach of the XGMII between the reconciliation sublayer and the PHY layer in the 10 Gbps LAN as shown in Figure 1–62. The XAUI interface consists of four lanes, each running at 3.125 Gbps with 8B/10B encoded data for a total of actual 10 Gbps data throughput. At the transmit side of the XAUI interface, the data and control characters are

Table 1–25. PRBS, High and Low Frequency Patterns, and Channel Settings (Part 2 of 2)

Patterns	Polynomial	8-bit Channel Width				10-bit Channel Width			
		Channel Width of 8 bits (1)	Word Alignment Pattern	Maximum Data Rate (Gbps) for F324 and Smaller Packages	Maximum Data Rate (Gbps) for F484 and Larger Packages	Channel Width of 10-bits (1)	Word Alignment Pattern	Maximum Data Rate (Gbps) for F324 and Smaller Packages	Maximum Data Rate (Gbps) for F484 and Larger Packages
Low Frequency (2)	1111100000	N	—	—	—	Y	—	2.5	3.125

Notes to Table 1–25:

(1) Channel width refers to the **What is the channel width?** option in the **General** screen of the ALTGX MegaWizard Plug-In Manager. Based on the selection, an 8 or 10 bits wide pattern is generated as indicated by a **Yes (Y)** or **No (N)**.

(2) A verifier and associated `rx_bistdone` and `rx_bisterr` signals are not available for the specified patterns.

You can enable the serial loopback option to loop the generated PRBS patterns to the receiver channel for verifier to check the PRBS patterns. When the PRBS pattern is received, the `rx_bisterr` and `rx_bistdone` signals indicate the status of the verifier. After the word aligner restores the word boundary, the `rx_bistdone` signal is driven high when the verifier receives a complete pattern cycle and remains asserted until it is reset using the `rx_digitalreset` port. After the assertion of `rx_bistdone`, the `rx_bisterr` signal is asserted for a minimum of three `rx_clkout` cycles when errors are detected in the data and deasserts if the following PRBS sequence contains no error. You can reset the PRBS pattern generator and verifier by asserting the `tx_digitalreset` and `rx_digitalreset` ports, respectively.

As shown in Figure 2–5, perform the following reset procedure for the receiver CDR in manual lock mode configuration:

1. After power up, assert `pll_areset` for a minimum period of 1 µs (the time between markers 1 and 2).
2. Keep the `tx_digitalreset`, `rx_analogreset`, `rx_digitalreset`, and `rx_locktorefclk` signals asserted and the `rx_locktodata` signal deasserted during this time period. After you deassert the `pll_areset` signal, the multipurpose PLL starts locking to the input reference clock.
3. After the multipurpose PLL locks, as indicated by the `pll_locked` signal going high (marker 3), deassert the `tx_digitalreset` signal (marker 4). For the receiver operation, after deassertion of the `busy` signal, wait for **two parallel clock cycles** to deassert the `rx_analogreset` signal.
4. In a bonded channel group, wait for at least $t_{LTR\_LTD\_Manual}$, then deassert `rx_locktorefclk` and assert `rx_locktodata` (marker 7). At this point, the receiver CDR of all the channels enters into lock-to-data mode and starts locking to the received data.
5. After asserting the `rx_locktodata` signal, wait for at least t_{LTD_Manual} before deasserting `rx_digitalreset` (the time between markers 7 and 8). At this point, the transmitter and receiver are ready for data traffic.

Non-Bonded Channel Configuration

In non-bonded channels, each channel in the ALTGX MegaWizard Plug-In Manager instance contains its own `tx_digitalreset`, `rx_analogreset`, `rx_digitalreset`, and `rx_freqlocked` signals.

You can reset each channel independently. For example, if there are four non-bonded channels, the ALTGX MegaWizard Plug-In Manager provides four each of the following signals: `tx_digitalreset`, `rx_analogreset`, `rx_digitalreset`, and `rx_freqlocked`.

Table 2–6 lists the reset and power-down sequences for one channel in a non-bonded configuration under the stated functional modes.

Table 2–6. Reset and Power-Down Sequences for Non-Bonded Channel Configurations

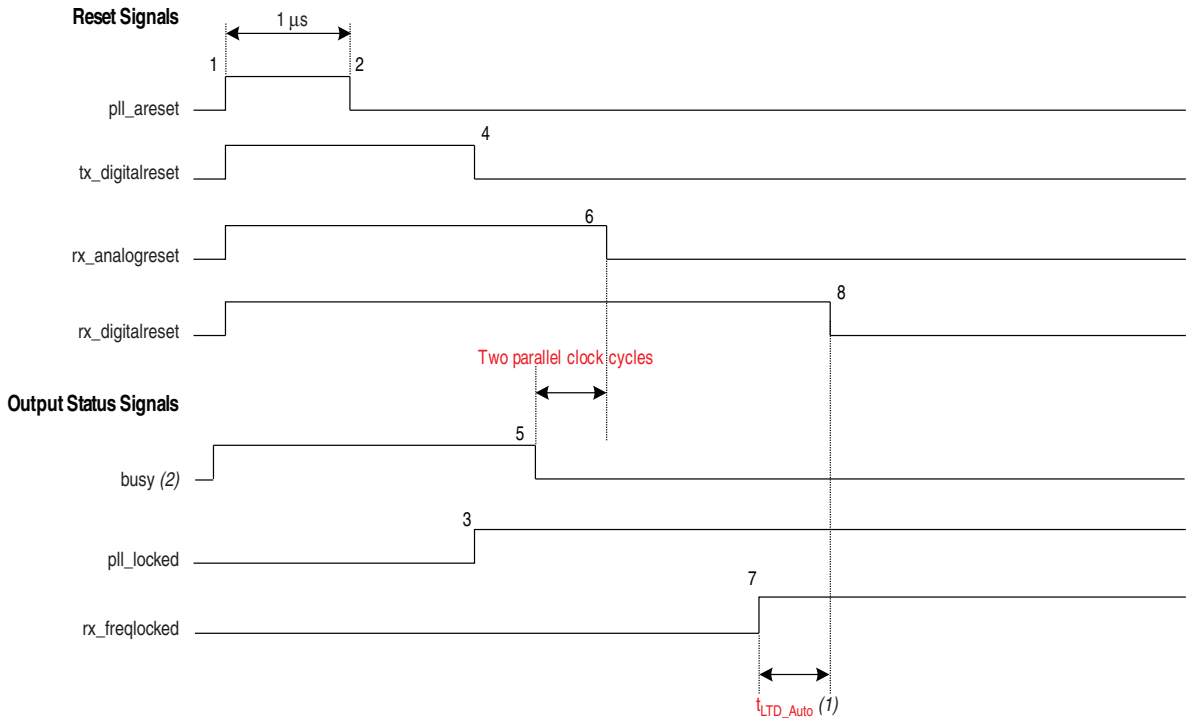
Channel Set Up	Receiver CDR Mode	Refer to
Transmitter Only	Basic ×1	"Transmitter Only Channel" on page 2–11
Receiver Only	Automatic lock mode	"Receiver Only Channel—Receiver CDR in Automatic Lock Mode" on page 2–11
Receiver Only	Manual lock mode	"Receiver Only Channel—Receiver CDR in Manual Lock Mode" on page 2–12
Receiver and Transmitter	Automatic lock mode	"Receiver and Transmitter Channel—Receiver CDR in Automatic Lock Mode" on page 2–13
Receiver and Transmitter	Manual lock mode	"Receiver and Transmitter Channel—Receiver CDR in Manual Lock Mode" on page 2–14

Follow the same reset sequence for all the other channels in the non-bonded configuration.

Receiver and Transmitter Channel—Receiver CDR in Automatic Lock Mode

This configuration contains both a transmitter and a receiver channel. If you create a **Receiver and Transmitter** instance in the ALTGX MegaWizard Plug-In Manager with the receiver CDR in automatic lock mode, use the reset sequence shown in Figure 2–8.

Figure 2–8. Sample Reset Sequence of Receiver and Transmitter Channel—Receiver CDR in Automatic Lock Mode



Notes to Figure 2–8:

(1) For t_{LTD_Auto} duration, refer to the *Cyclone IV Device Datasheet* chapter.

(2) The `busy` signal is asserted and deasserted only during initial power up when offset cancellation occurs. In subsequent reset sequences, the `busy` signal is asserted and deasserted only if there is a read or write operation to the ALTGX_RECONFIG megafunction.

As shown in Figure 2–8, perform the following reset procedure for the receiver in CDR automatic lock mode:

1. After power up, assert `pll_areset` for a minimum period of 1 µs (the time between markers 1 and 2).
2. Keep the `tx_digitalreset`, `rx_analogreset`, and `rx_digitalreset` signals asserted during this time period. After you deassert the `pll_areset` signal, the multipurpose PLL starts locking to the transmitter input reference clock.
3. After the multipurpose PLL locks, as indicated by the `pll_locked` signal going high (marker 3), deassert `tx_digitalreset`. For receiver operation, after deassertion of busy signal, wait for two parallel clock cycles to deassert the `rx_analogreset` signal.
4. Wait for the `rx_freqlocked` signal to go high (marker 7).
5. After the `rx_freqlocked` signal goes high, wait for at least t_{LTD_Auto}, then deassert the `rx_digitalreset` signal (marker 8). At this point, the transmitter and receiver are ready for data traffic.

There are three methods that you can use to dynamically reconfigure the PMA controls of a transceiver channel:

- "Method 1: Using logical_channel_address to Reconfigure Specific Transceiver Channels" on page 3–14
- "Method 2: Writing the Same Control Signals to Control All the Transceiver Channels" on page 3–16
- "Method 3: Writing Different Control Signals for all the Transceiver Channels at the Same Time" on page 3–19

Method 1: Using logical_channel_address to Reconfigure Specific Transceiver Channels

Enable the `logical_channel_address` port by selecting the **Use 'logical_channel_address' port** option on the **Analog controls** tab. This method is applicable only for a design where the dynamic reconfiguration controller controls more than one channel.

You can additionally reconfigure either the receiver portion, transmitter portion, or both the receiver and transmitter portions of the transceiver channel by setting the corresponding value on the `rx_tx_duplex_sel` input port. For more information, refer to Table 3–2 on page 3–4.

Connecting the PMA Control Ports

The selected PMA control ports remain fixed in width, regardless of the number of channels controlled by the ALTGX_RECONFIG instance:

- `tx_vodctrl` and `tx_vodctrl_out` are fixed to 3 bits
- `tx_preemp` and `tx_preemp_out` are fixed to 5 bits
- `rx_eqdcgain` and `rx_eqdcgain_out` are fixed to 2 bits
- `rx_eqctrl` and `rx_eqctrl_out` are fixed to 4 bits

Write Transaction

To complete a write transaction, perform the following steps:

1. Set the selected PMA control ports to the desired settings (for example, `tx_vodctrl` = **3'b001**).
2. Set the `logical_channel_address` input port to the logical channel address of the transceiver channel whose PMA controls you want to reconfigure.
3. Set the `rx_tx_duplex_sel` port to **2'b10** so that only the transmit PMA controls are written to the transceiver channel.
4. Ensure that the `busy` signal is low before you start a write transaction.
5. Assert the `write_all` signal for one `reconfig_clk` clock cycle.

The `busy` output status signal is asserted high to indicate that the dynamic reconfiguration controller is busy writing the PMA control values. When the write transaction has completed, the `busy` signal goes low.

Transceiver Performance Specifications

Table 1–21 lists the Cyclone IV GX transceiver specifications.

Table 1–21. Transceiver Specification for Cyclone IV GX Devices (Part 1 of 4)

Symbol/ Description	Conditions	C6			C7, I7			C8			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
Reference Clock											
Supported I/O Standards	1.2 V PCML, 1.5 V PCML, 3.3 V PCML, Differential LVPECL, LVDS, HCSL										
Input frequency from REFCLK input pins	—	50	—	156.25	50	—	156.25	50	—	156.25	MHz
Spread-spectrum modulating clock frequency	Physical interface for PCI Express (PIPE) mode	30	—	33	30	—	33	30	—	33	kHz
Spread-spectrum downspread	PIPE mode	—	0 to –0.5%	—	—	0 to –0.5%	—	—	0 to –0.5%	—	—
Peak-to-peak differential input voltage	—	0.1	—	1.6	0.1	—	1.6	0.1	—	1.6	V
V_{ICM} (AC coupled)	—	1100 ± 5%			1100 ± 5%			1100 ± 5%			mV
V_{ICM} (DC coupled)	HCSL I/O standard for PCIe reference clock	250	—	550	250	—	550	250	—	550	mV
Transmitter REFCLK Phase Noise (1)	Frequency offset = 1 MHz – 8 MHZ	—	—	–123	—	—	–123	—	—	–123	dBc/Hz
Transmitter REFCLK Total Jitter (1)		—	—	42.3	—	—	42.3	—	—	42.3	ps
R_{ref}	—	—	2000 ± 1%	—	—	2000 ± 1%	—	—	2000 ± 1%	—	Ω
Transceiver Clock											
cal_blk_clk clock frequency	—	10	—	125	10	—	125	10	—	125	MHz
fixedclk clock frequency	PCIe Receiver Detect	—	125	—	—	125	—	—	125	—	MHz
reconfig_clk clock frequency	Dynamic reconfiguration clock frequency	2.5/ 37.5 (2)	—	50	2.5/ 37.5 (2)	—	50	2.5/ 37.5 (2)	—	50	MHz
Delta time between reconfig_clk	—	—	—	2	—	—	2	—	—	2	ms
Transceiver block minimum power-down pulse width	—	—	1	—	—	1	—	—	1	—	µs

Table 1–31. RSDS Transmitter Timing Specifications for Cyclone IV Devices (1), (2), (4) (Part 2 of 2)

Symbol	Modes	C6			C7, I7			C8, A7			C8L, I8L			C9L			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
t_{LOCK} (3)	—	—	—	1	—	—	1	—	—	1	—	—	1	—	—	1	ms

Notes to Table 1–31:

(1) Applicable for true RSDS and emulated RSDS_E_3R transmitter.

(2) Cyclone IV E devices—true RSDS transmitter is only supported at the output pin of Row I/O Banks 1, 2, 5, and 6. Emulated RSDS transmitter is supported at the output pin of all I/O Banks.
Cyclone IV GX devices—true RSDS transmitter is only supported at the output pin of Row I/O Banks 5 and 6. Emulated RSDS transmitter is supported at the output pin of I/O Banks 3, 4, 5, 6, 7, 8, and 9.

(3) t_{LOCK} is the time required for the PLL to lock from the end-of-device configuration.

(4) Cyclone IV E 1.0 V core voltage devices only support C8L, C9L, and I8L speed grades. Cyclone IV E 1.2 V core voltage devices only support C6, C7, C8, I7, and A7 speed grades. Cyclone IV GX devices only support C6, C7, C8, and I7 speed grades.

Table 1–32. Emulated RSDS_E_1R Transmitter Timing Specifications for Cyclone IV Devices (1), (3) (Part 1 of 2)

Symbol	Modes	C6			C7, I7			C8, A7			C8L, I8L			C9L			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
f_{HSCLK} (input clock frequency)	×10	5	—	85	5	—	85	5	—	85	5	—	85	5	—	72.5	MHz
	×8	5	—	85	5	—	85	5	—	85	5	—	85	5	—	72.5	MHz
	×7	5	—	85	5	—	85	5	—	85	5	—	85	5	—	72.5	MHz
	×4	5	—	85	5	—	85	5	—	85	5	—	85	5	—	72.5	MHz
	×2	5	—	85	5	—	85	5	—	85	5	—	85	5	—	72.5	MHz
	×1	5	—	170	5	—	170	5	—	170	5	—	170	5	—	145	MHz
Device operation in Mbps	×10	100	—	170	100	—	170	100	—	170	100	—	170	100	—	145	Mbps
	×8	80	—	170	80	—	170	80	—	170	80	—	170	80	—	145	Mbps
	×7	70	—	170	70	—	170	70	—	170	70	—	170	70	—	145	Mbps
	×4	40	—	170	40	—	170	40	—	170	40	—	170	40	—	145	Mbps
	×2	20	—	170	20	—	170	20	—	170	20	—	170	20	—	145	Mbps
	×1	10	—	170	10	—	170	10	—	170	10	—	170	10	—	145	Mbps
t_{DUTY}	—	45	—	55	45	—	55	45	—	55	45	—	55	45	—	55	%
TCCS	—	—	—	200	—	—	200	—	—	200	—	—	200	—	—	200	ps
Output jitter (peak to peak)	—	—	—	500	—	—	500	—	—	550	—	—	600	—	—	700	ps
t_{RISE}	20 – 80%, C_{LOAD} = 5 pF	—	500	—	—	500	—	—	500	—	—	500	—	—	500	—	ps
t_{FALL}	20 – 80%, C_{LOAD} = 5 pF	—	500	—	—	500	—	—	500	—	—	500	—	—	500	—	ps