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### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Active
Number of LABs/CLBs	392
Number of Logic Elements/Cells	6272
Total RAM Bits	276480
Number of I/O	179
Number of Gates	-
Voltage - Supply	0.97V ~ 1.03V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/ep4ce6f17c9ln">https://www.e-xfl.com/product-detail/intel/ep4ce6f17c9ln</a>

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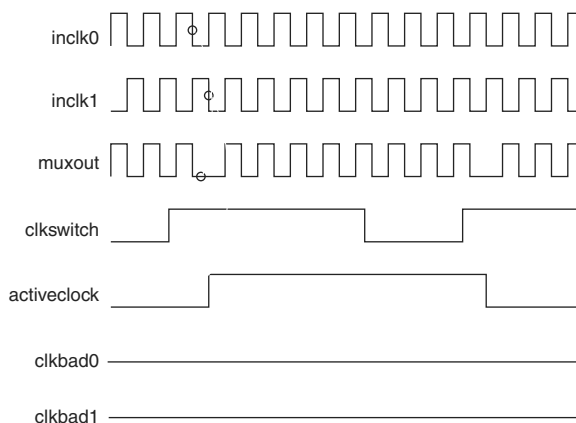
- Cyclone IV GX devices offer up to eight high-speed transceivers that provide:
  - Data rates up to 3.125 Gbps
  - 8B/10B encoder/decoder
  - 8-bit or 10-bit physical media attachment (PMA) to physical coding sublayer (PCS) interface
  - Byte serializer/deserializer (SERDES)
  - Word aligner
  - Rate matching FIFO
  - TX bit slipper for Common Public Radio Interface (CPRI)
  - Electrical idle
  - Dynamic channel reconfiguration allowing you to change data rates and protocols on-the-fly
  - Static equalization and pre-emphasis for superior signal integrity
  - 150 mW per channel power consumption
  - Flexible clocking structure to support multiple protocols in a single transceiver block
- Cyclone IV GX devices offer dedicated hard IP for PCI Express (PIPE) (PCIe) Gen 1:
  - ×1, ×2, and ×4 lane configurations
  - End-point and root-port configurations
  - Up to 256-byte payload
  - One virtual channel
  - 2 KB retry buffer
  - 4 KB receiver (Rx) buffer
- Cyclone IV GX devices offer a wide range of protocol support:
  - PCIe (PIPE) Gen 1 ×1, ×2, and ×4 (2.5 Gbps)
  - Gigabit Ethernet (1.25 Gbps)
  - CPRI (up to 3.072 Gbps)
  - XAUI (3.125 Gbps)
  - Triple rate serial digital interface (SDI) (up to 2.97 Gbps)
  - Serial RapidIO (3.125 Gbps)
  - Basic mode (up to 3.125 Gbps)
  - V-by-One (up to 3.0 Gbps)
  - DisplayPort (2.7 Gbps)
  - Serial Advanced Technology Attachment (SATA) (up to 3.0 Gbps)
  - OBSAI (up to 3.072 Gbps)

In this mode, the `activeclock` signal mirrors the `clkswitch` signal. As both blocks are still functional during the manual switch, neither `clkbad` signals go high. Because the switchover circuit is positive edge-sensitive, the falling edge of the `clkswitch` signal does not cause the circuit to switch back from `inclk1` to `inclk0`. When the `clkswitch` signal goes high again, the process repeats. The `clkswitch` signal and the automatic switch only works depending on the availability of the clock that is switched to. If the clock is unavailable, the state machine waits until the clock is available.



When `CLKSWITCH = 1`, it overrides the automatic switch-over function. As long as `clkswitch` signal is high, further switch-over action is blocked.

**Figure 5–19. Clock Switchover Using the `clkswitch` Control <sup>(1)</sup>**



**Note to Figure 5–19:**

(1) Both `inclk0` and `inclk1` must be running when the `clkswitch` signal goes high to start a manual clock switchover event.

## Manual Clock Switchover

PLLs of Cyclone IV devices support manual switchover, in which the `clkswitch` signal controls whether `inclk0` or `inclk1` is the input clock to the PLL. The characteristics of a manual switchover are similar to the manual override feature in an automatic clock switchover, in which the switchover circuit is edge-sensitive. When the `clkswitch` signal goes high, the switchover sequence starts. The falling edge of the `clkswitch` signal does not cause the circuit to switch back to the previous input clock.



For more information about PLL software support in the Quartus II software, refer to the *ALTPLL Megafunction User Guide*.

## Guidelines

Use the following guidelines to design with clock switchover in PLLs:

- Clock loss detection and automatic clock switchover require the `inclk0` and `inclk1` frequencies be within 20% of each other. Failing to meet this requirement causes the `clkbad0` and `clkbad1` signals to function improperly.



Table 6-6 and Table 6-7 summarize which I/O banks support these I/O standards in the Cyclone IV device family.

**Table 6-6. Differential I/O Standards Supported in Cyclone IV E I/O Banks**

Differential I/O Standards	I/O Bank Location	External Resistor Network at Transmitter	Transmitter (TX)	Receiver (RX)
LVDS	1,2,5,6	Not Required	✓	✓
	All	Three Resistors		
RSDS	1,2,5,6	Not Required	✓	—
	3,4,7,8	Three Resistors		
	All	Single Resistor		
mini-LVDS	1,2,5,6	Not Required	✓	—
	All	Three Resistors		
PPDS	1,2,5,6	Not Required	✓	—
	All	Three Resistors		
BLVDS <sup>(1)</sup>	All	Single Resistor	✓	✓
LVPECL <sup>(2)</sup>	All	—	—	✓
Differential SSTL-2 <sup>(3)</sup>	All	—	✓	✓
Differential SSTL-18 <sup>(3)</sup>	All	—	✓	✓
Differential HSTL-18 <sup>(3)</sup>	All	—	✓	✓
Differential HSTL-15 <sup>(3)</sup>	All	—	✓	✓
Differential HSTL-12 <sup>(3), (4)</sup>	All	—	✓	✓

**Notes to Table 6-6:**

- (1) Transmitter and Receiver  $f_{MAX}$  depend on system topology and performance requirement.
- (2) The LVPECL I/O standard is only supported on dedicated clock input pins.
- (3) The differential SSTL-2, SSTL-18, HSTL-18, HSTL-15, and HSTL-12 I/O standards are only supported on clock input pins and PLL output clock pins. PLL output clock pins do not support Class II interface type of differential SSTL-18, HSTL-18, HSTL-15, and HSTL-12 I/O standards.
- (4) Differential HSTL-12 Class II is supported only in column I/O banks.

You can begin reconfiguration by pulling the `nCONFIG` pin low. The `nCONFIG` pin must be low for at least 500 ns. When `nCONFIG` is pulled low, the Cyclone IV device is reset. The Cyclone IV device also pulls `nSTATUS` and `CONF_DONE` low and all I/O pins are tri-stated. When `nCONFIG` returns to a logic-high level and `nSTATUS` is released by the Cyclone IV device, reconfiguration begins.

## Configuration Error

If an error occurs during configuration, Cyclone IV devices assert the `nSTATUS` signal low, indicating a data frame error and the `CONF_DONE` signal stays low. If the **Auto-restart configuration after error** option (available in the Quartus II software in the **General** tab of the **Device and Pin Options** dialog box) is turned on, the Cyclone IV device releases `nSTATUS` after a reset time-out period (a maximum of 230  $\mu$ s), and retries configuration. If this option is turned off, the system must monitor `nSTATUS` for errors and then pulse `nCONFIG` low for at least 500 ns to restart configuration.

## Initialization

In Cyclone IV devices, the initialization clock source is either the internal oscillator or the optional `CLKUSR` pin. By default, the internal oscillator is the clock source for initialization. If you use the internal oscillator, the device provides itself with enough clock cycles for proper initialization. When using the internal oscillator, you do not have to send additional clock cycles from an external source to the `CLKUSR` pin during the initialization stage. Additionally, you can use the `CLKUSR` pin as a user I/O pin.

You also have the flexibility to synchronize initialization of multiple devices or to delay initialization with the **CLKUSR** option. The `CLKUSR` pin allows you to control when your device enters user mode for an indefinite amount of time. You can turn on the **Enable user-supplied start-up clock (CLKUSR)** option in the Quartus II software in the **General** tab of the **Device and Pin Options** dialog box. When you turn on the **Enable user supplied start-up clock option (CLKUSR)** option, the `CLKUSR` pin is the initialization clock source. Supplying a clock on the `CLKUSR` pin does not affect the configuration process. After the configuration data is accepted and `CONF_DONE` goes high, Cyclone IV devices require 3,192 clock cycles to initialize properly and enter user mode.



If you use the optional `CLKUSR` pin and the `nCONFIG` pin is pulled low to restart configuration during device initialization, ensure that the `CLKUSR` pin continues to toggle when `nSTATUS` is low (a maximum of 230  $\mu$ s).

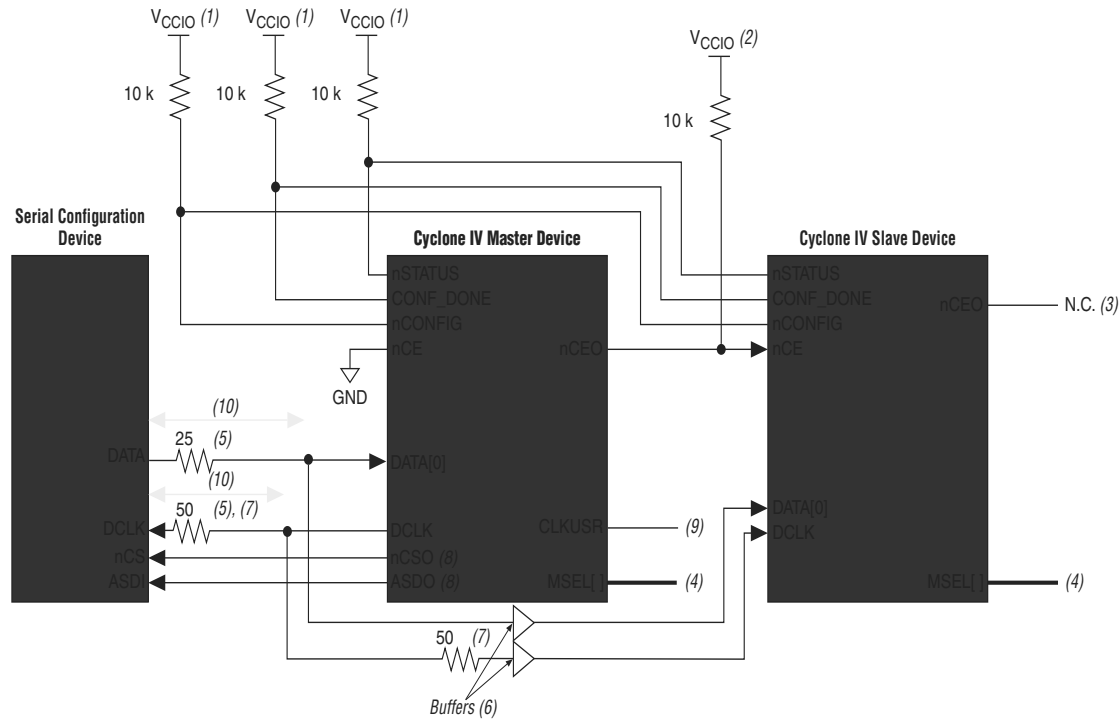
## User Mode

An optional `INIT_DONE` pin is available, which signals the end of initialization and the start of user mode with a low-to-high transition. The **Enable INIT\_DONE Output** option is available in the Quartus II software in the **General** tab of the **Device and Pin Options** dialog box. If you use the `INIT_DONE` pin, it is high due to an external 10-k $\Omega$  pull-up resistor when `nCONFIG` is low and during the beginning of configuration. After the option bit to enable `INIT_DONE` is programmed into the device (during the first frame of configuration data), the `INIT_DONE` pin goes low. When initialization is complete, the `INIT_DONE` pin is released and pulled high. This low-to-high transition signals that the device has entered user mode. In user mode, the user I/O pins function as assigned in your design and no longer have weak pull-up resistors.

## Multi-Device AS Configuration

You can configure multiple Cyclone IV devices with a single serial configuration device. When the first device captures all its configuration data from the bitstream, it drives the `nCEO` pin low, enabling the next device in the chain. If the last device in the chain is a Cyclone IV device, you can leave the `nCEO` pin of the last device unconnected or use it as a user I/O pin after configuration. The `nCONFIG`, `nSTATUS`, `CONF_DONE`, `DCLK`, and `DATA[0]` pins of each device in the chain are connected together (Figure 8-3).

**Figure 8-3. Multi-Device AS Configuration**



### Notes to Figure 8-3:

- (1) Connect the pull-up resistors to the `VCCIO` supply of the bank in which the pin resides.
- (2) Connect the pull-up resistor to the `VCCIO` supply voltage of I/O bank in which the `nCEO` pin resides.
- (3) You can leave the `nCEO` pin unconnected or use it as a user I/O pin when it does not feed the `nCEO` pin of another device.
- (4) The `MSEL` pin settings vary for different configuration voltage standards and POR time. You must set the master device of the Cyclone IV device in AS mode and the slave devices in PS mode. To connect the `MSEL` pins for the master device in AS mode and slave devices in PS mode, refer to Table 8-3 on page 8-8, Table 8-4 on page 8-8, and Table 8-5 on page 8-9. Connect the `MSEL` pins directly to `VCCA` or GND.
- (5) Connect the series resistor at the near end of the serial configuration device.
- (6) Connect the repeater buffers between the master and slave devices of the Cyclone IV device for `DATA[0]` and `DCLK`. All I/O inputs must maintain a maximum AC voltage of 4.1 V. The output resistance of the repeater buffers must fit the maximum overshoot equation outlined in "Configuration and JTAG Pin I/O Requirements" on page 8-5.
- (7) The 50-Ω series resistors are optional if the 3.3-V configuration voltage standard is applied. For optimal signal integrity, connect these 50-Ω series resistors if the 2.5- or 3.0-V configuration voltage standard is applied.
- (8) These pins are dual-purpose I/O pins. The `nCSO` pin functions as `FLASH_nCEO` pin in AP mode. The `ASDO` pin functions as `DATA[1]` pin in AP and FPP modes.
- (9) Only Cyclone IV GX devices have an option to select `CLKUSR` (40 MHz maximum) as the external clock source for `DCLK`.
- (10) For multi-devices AS configuration using Cyclone IV E with 1.0 V core voltage, the maximum board trace-length from the serial configuration device to the junction-split on both `DCLK` and `Data0` line is 3.5 inches.



Altera recommends putting a buffer before the DATA and DCLK output from the master device to avoid signal strength and signal integrity issues. The buffer must not significantly change the DATA-to-DCLK relationships or delay them with respect to other AS signals (ASDI and nCS). Also, the buffer must only drive the slave devices to ensure that the timing between the master device and the serial configuration device is unaffected.

This configuration method supports both compressed and uncompressed .sof. Therefore, if the configuration bitstream size exceeds the capacity of a serial configuration device, you can enable the compression feature in the .sof or you can select a larger serial configuration device.

### Guidelines for Connecting a Serial Configuration Device to Cyclone IV Devices for an AS Interface

For single- and multi-device AS configurations, the board trace length and loading between the supported serial configuration device and Cyclone IV device must follow the recommendations listed in Table 8-7.

**Table 8-7. Maximum Trace Length and Loading for AS Configuration**

Cyclone IV Device AS Pins	Maximum Board Trace Length from a Cyclone IV Device to a Serial Configuration Device (Inches)		Maximum Board Load (pF)
	Cyclone IV E	Cyclone IV GX	
DCLK	10	6	15
DATA [0]	10	6	30
nCSO	10	6	30
ASDO	10	6	30

**Note to Table 8-7:**

- (1) For multi-devices AS configuration using Cyclone IV E with 1.0 V core voltage, the maximum board trace-length from the serial configuration device to the junction-split on both DCLK and Data0 line is 3.5 inches.

### Estimating AS Configuration Time

AS configuration time is dominated by the time it takes to transfer data from the serial configuration device to the Cyclone IV device. This serial interface is clocked by the Cyclone IV device DCLK output (generated from a 40-MHz internal oscillator for Cyclone IV E devices, a 20- or 40-MHz internal oscillator, or an external CLKUSR of up to 40 MHz for Cyclone IV GX devices).



Equation 8-2 and Equation 8-3 show the configuration time calculations.

**Equation 8-2.**

$$\text{Size} \times \left( \frac{\text{maximum DCLK period}}{1 \text{ bit}} \right) = \text{estimated maximum configuration time}$$

**Equation 8-3.**

$$9,600,000 \text{ bits} \times \left( \frac{50 \text{ ns}}{1 \text{ bit}} \right) = 480 \text{ ms}$$

-  The **.rbf** used by the JRunner software driver cannot be a compressed **.rbf** because the JRunner software driver uses JTAG-based configuration. During JTAG-based configuration, the real-time decompression feature is not available.
-  For more information about the JRunner software driver, refer to *AN 414: JRunner Software Driver: An Embedded Solution for PLD JTAG Configuration* and the source files on the Altera website at ([www.altera.com](http://www.altera.com)).

### Combining JTAG and AS Configuration Schemes

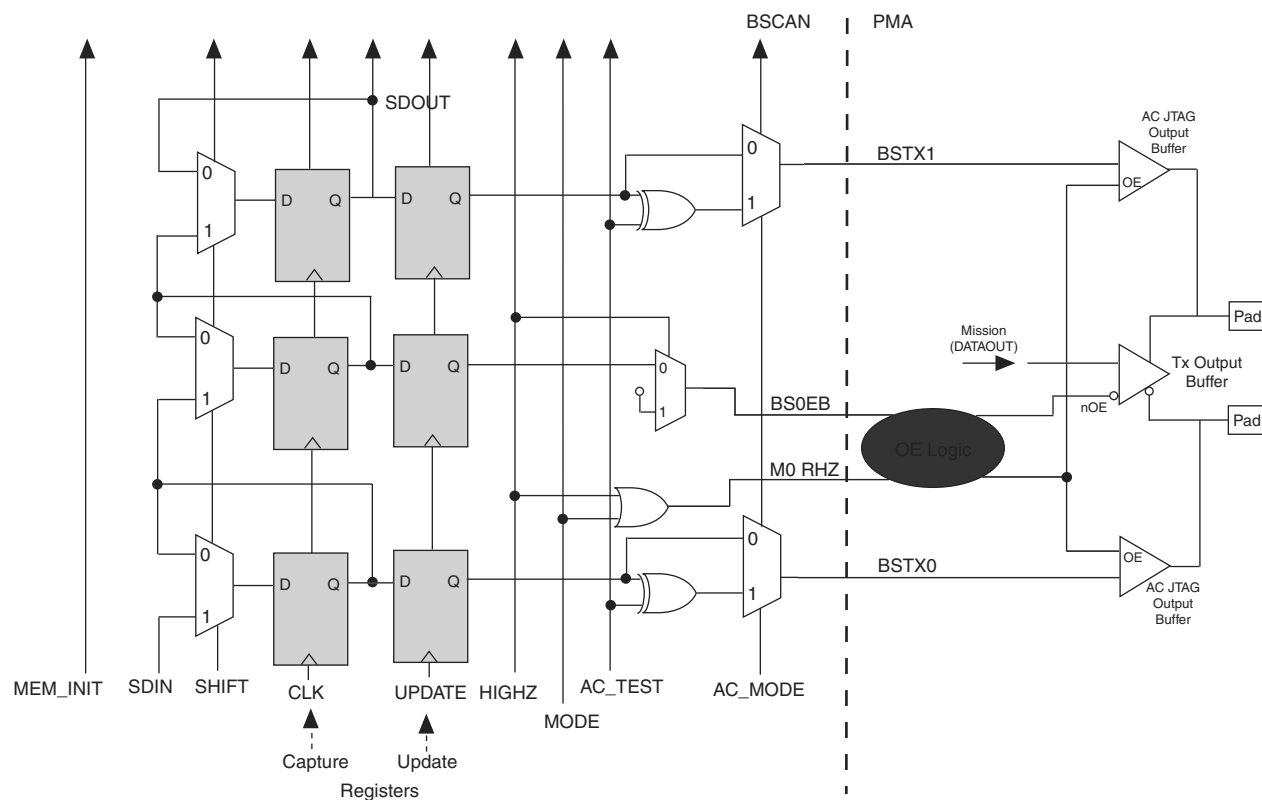
You can combine the AS configuration scheme with the JTAG-based configuration (Figure 8-28). This setup uses two 10-pin download cable headers on the board. One download cable is used in JTAG mode to configure the Cyclone IV device directly through the JTAG interface. The other download cable is used in AS mode to program the serial configuration device in-system through the AS programming interface. If you try configuring the device using both schemes simultaneously, JTAG configuration takes precedence and AS configuration terminates.

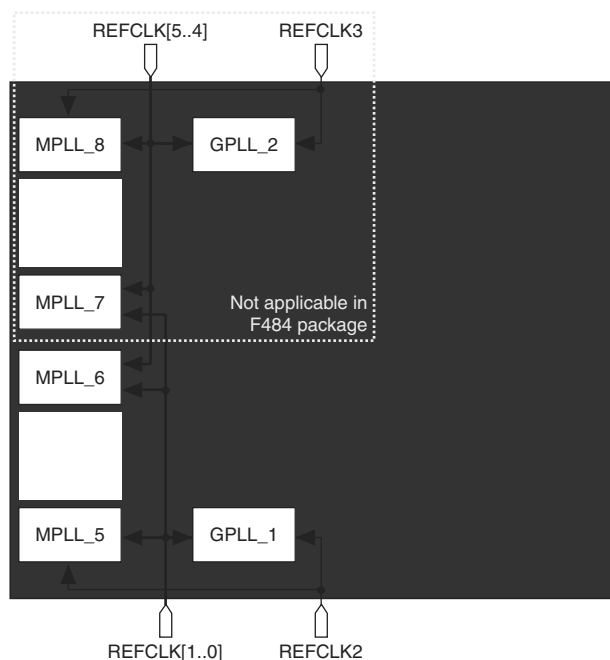
## IEEE Std. 1149.6 Boundary-Scan Register

The boundary-scan cell (BSC) for HSSI transmitters (GXB\_TX[p,n]) and receivers (GXB\_RX[p,n]) in Cyclone IV GX devices are different from the BSCs for I/O pins.

Figure 10-1 shows the Cyclone IV GX HSSI transmitter boundary-scan cell.

**Figure 10-1. HSSI Transmitter BSC with IEEE Std. 1149.6 BST Circuitry for Cyclone IV GX Devices**



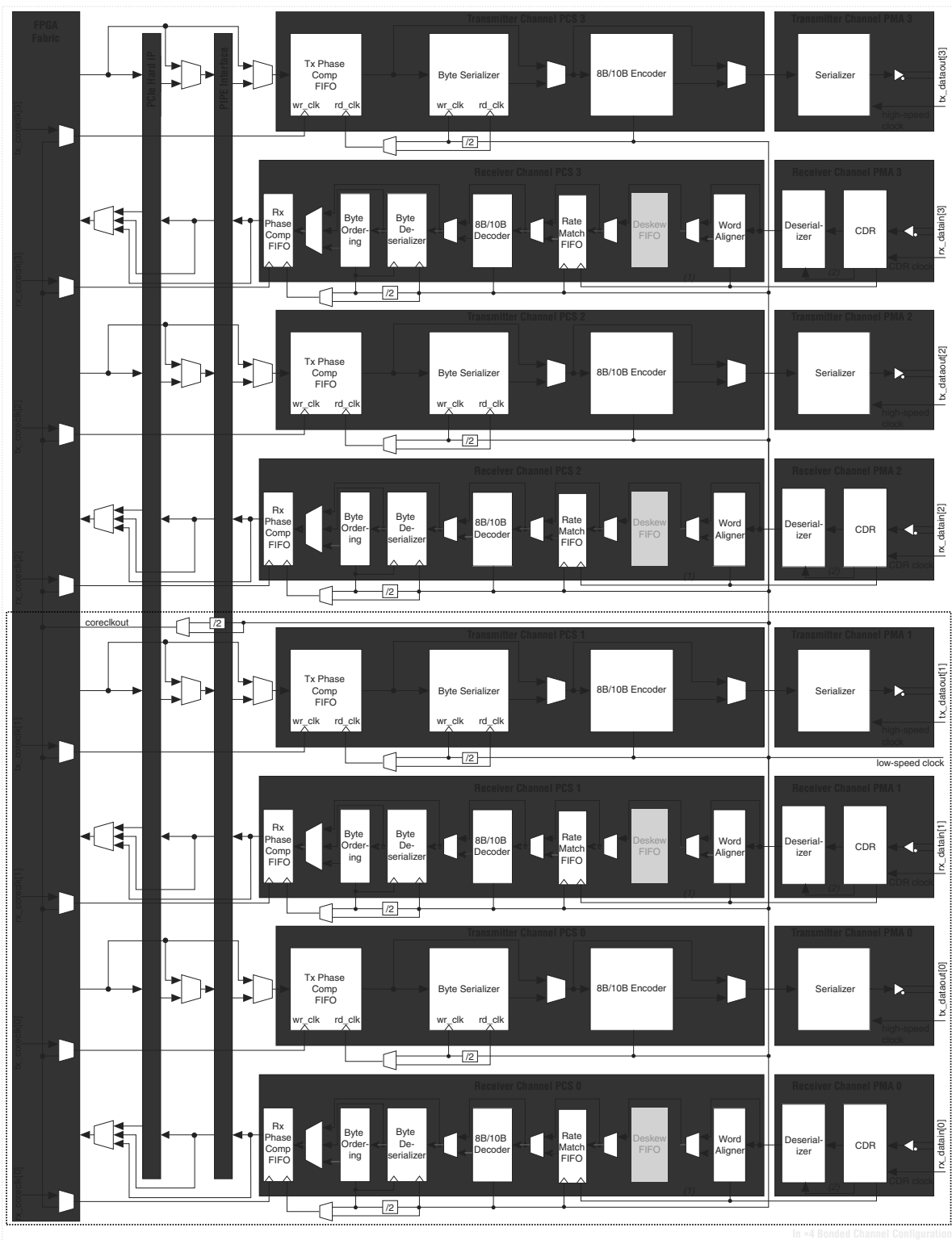
**Figure 1–26. PLL Input Reference Clocks in Transceiver Operation for F484 and Larger Packages**  
(1), (2), (3)**Notes to Figure 1–26:**

- (1) The REFCLK2 and REFCLK3 pins are dual-purpose CLKIO, REFCLK, or DIFFCLK pins that reside in banks 3A and 8A respectively.
- (2) The REFCLK[1..0] and REFCLK[5..4] pins are dual-purpose differential REFCLK or DIFFCLK pins that reside in banks 3B and 8B respectively. These clock input pins do not have access to the clock control blocks and GCLK networks. For more details, refer to the *Clock Networks and PLLs in Cyclone IV Devices* chapter.
- (3) Using any clock input pins other than the designated REFCLK pins as shown here to drive the MPLLs and GPLLs may have reduced jitter performance.

The input reference clocks reside in banks 3A, 3B, 8A, and 8B have dedicated  $V_{CC\_CLKIN3A}$ ,  $V_{CC\_CLKIN3B}$ ,  $V_{CC\_CLKIN8A}$ , and  $V_{CC\_CLKIN8B}$  power supplies separately in their respective I/O banks to avoid the different power level requirements in the same bank for general purpose I/Os (GPIOs). Table 1–6 lists the supported I/O standard for the REFCLK pins.

**Table 1–6. REFCLK I/O Standard Support**

I/O Standard	HSSI Protocol	Coupling	Termination	VCC_CLKIN Level		I/O Pin Type		
				Input	Output	Column I/O	Row I/O	Supported Banks
LVDS	ALL	Differential AC (Needs off-chip resistor to restore $V_{CM}$ )	Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B
LVPECL	ALL		Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B
1.2 V, 1.5 V, 3.3 V PCML	ALL		Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B
	ALL		Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B
	ALL		Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B
HCSL	PCIe	Differential DC	Off-chip	2.5 V	Not Supported	Yes	No	3A, 3B, 8A, 8B

**Figure 1–39. Transmitter and Receiver Datapath Clocking with Rate Match FIFO in Bonded Channel Configuration****Notes to Figure 1–39:**

- (1) Low-speed recovered clock.
- (2) High-speed recovered clock.

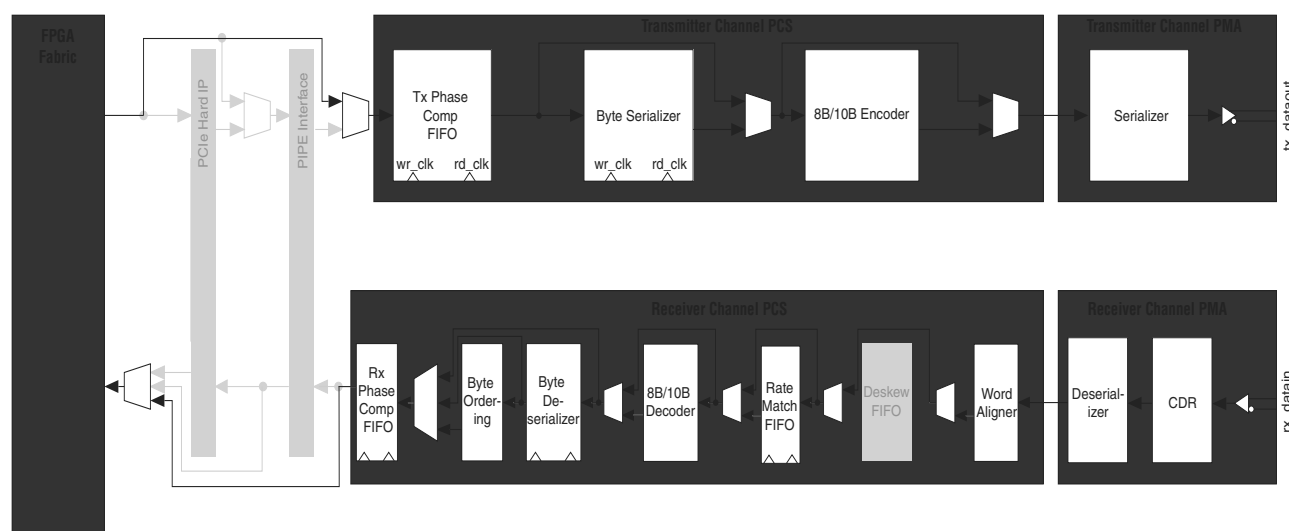
**Table 1–14. Transceiver Functional Modes for Protocol Implementation (Part 2 of 2)**

Functional Mode	Protocol	Key Feature	Reference
Deterministic Latency	Proprietary, CPRI, OBSAI	TX PLL phase frequency detector (PFD) feedback, registered mode FIFO, TX bit-slip control	“Deterministic Latency Mode” on page 1–73
SDI	SDI	High-speed SERDES, CDR	“SDI Mode” on page 1–76

## Basic Mode

The Cyclone IV GX transceiver channel datapath is highly flexible in Basic mode to implement proprietary protocols. SATA, V-by-One, and Display Port protocol implementations in Cyclone IV GX transceiver are supported with Basic mode.

Figure 1–44 shows the transceiver channel datapath supported in Basic mode.

**Figure 1–44. Transceiver Channel Datapath in Basic Mode**

**Table 1–25. PRBS, High and Low Frequency Patterns, and Channel Settings (Part 2 of 2)**

Patterns	Polynomial	8-bit Channel Width				10-bit Channel Width			
		Channel Width of 8 bits <sup>(1)</sup>	Word Alignment Pattern	Maximum Data Rate (Gbps) for F324 and Smaller Packages	Maximum Data Rate (Gbps) for F484 and Larger Packages	Channel Width of 10-bits <sup>(1)</sup>	Word Alignment Pattern	Maximum Data Rate (Gbps) for F324 and Smaller Packages	Maximum Data Rate (Gbps) for F484 and Larger Packages
Low Frequency <sup>(2)</sup>	1111100000	N	—	—	—	Y	—	2.5	3.125

**Notes to Table 1–25:**

- (1) Channel width refers to the **What is the channel width?** option in the **General** screen of the ALTGX MegaWizard Plug-In Manager. Based on the selection, an 8 or 10 bits wide pattern is generated as indicated by a **Yes (Y)** or **No (N)**.
- (2) A verifier and associated rx\_bistdone and rx\_bisterr signals are not available for the specified patterns.

You can enable the serial loopback option to loop the generated PRBS patterns to the receiver channel for verifier to check the PRBS patterns. When the PRBS pattern is received, the rx\_bisterr and rx\_bistdone signals indicate the status of the verifier. After the word aligner restores the word boundary, the rx\_bistdone signal is driven high when the verifier receives a complete pattern cycle and remains asserted until it is reset using the rx\_digitalreset port. After the assertion of rx\_bistdone, the rx\_bisterr signal is asserted for a minimum of three rx\_clkout cycles when errors are detected in the data and deasserts if the following PRBS sequence contains no error. You can reset the PRBS pattern generator and verifier by asserting the tx\_digitalreset and rx\_digitalreset ports, respectively.

## Transceiver Top-Level Port Lists

Table 1–26 through Table 1–29 provide descriptions of the ports available when instantiating a transceiver using the ALTGX megafunction. The ALTGX megafunction requires a relatively small number of signals. There are also a large number of optional signals that facilitate debugging by providing information about the state of the transceiver.

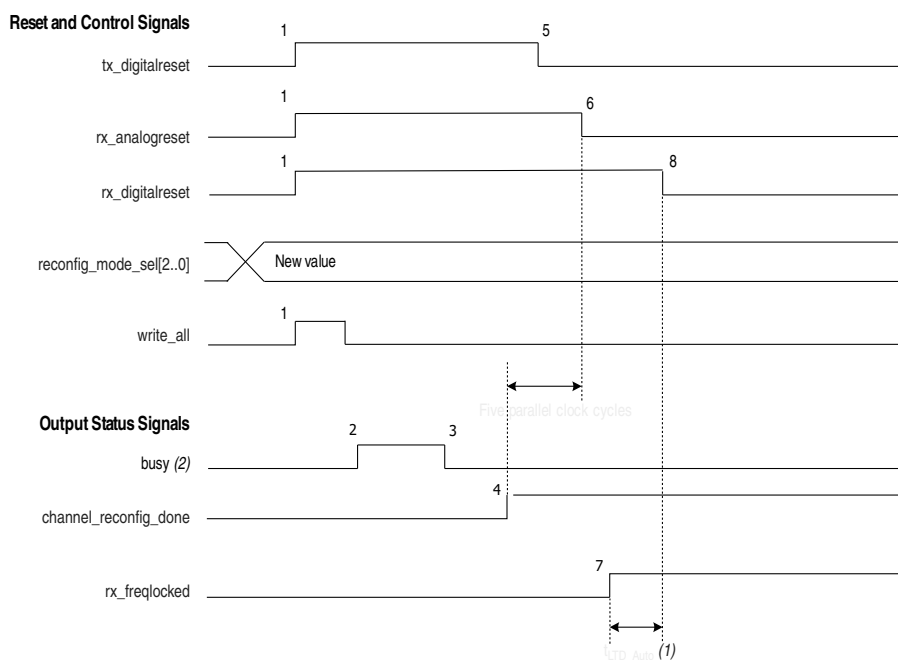


2. After the PLL is reset, wait for the `p11_locked` signal to go high (marker 4) indicating that the PLL is locked to the input reference clock. After the assertion of the `p11_locked` signal, deassert the `tx_digitalreset` signal (marker 5).
3. Wait at least five parallel clock cycles after the `p11_locked` signal is asserted to deassert the `rx_analogreset` signal (marker 6).
4. When the `rx_freqlocked` signal goes high (marker 7), from that point onwards, wait for at least  $t_{LTD\_Auto}$  time, then deassert the `rx_digitalreset` signal (marker 8). At this point, the receiver is ready for data traffic.

## Reset Sequence in Channel Reconfiguration Mode

Use the example reset sequence shown in Figure 2-12 when you are using the dynamic reconfiguration controller to change the PCS settings of the transceiver channel. In this example, the dynamic reconfiguration is used to dynamically reconfigure the transceiver channel configured in Basic  $\times 1$  mode with receiver CDR in automatic lock mode.

**Figure 2-12. Reset Sequence When Using the Dynamic Reconfiguration Controller to Change the PCS Settings of the Transceiver Channel**



### Notes to Figure 2-12:

- (1) For  $t_{LTD\_Auto}$  duration, refer to the *Cyclone IV Device Datasheet* chapter.
- (2) The `busy` signal is asserted and deasserted only during initial power up when offset cancellation occurs. In subsequent reset sequences, the `busy` signal is asserted and deasserted only if there is a read or write operation to the `ALTGX_RECONFIG` megafunction.

## Dynamic Reconfiguration Controller Port List

Table 3–2 lists the input control ports and output status ports of the dynamic reconfiguration controller.

**Table 3–2. Dynamic Reconfiguration Controller Port List (ALTGX\_RECONFIG Instance) (Part 1 of 7)**

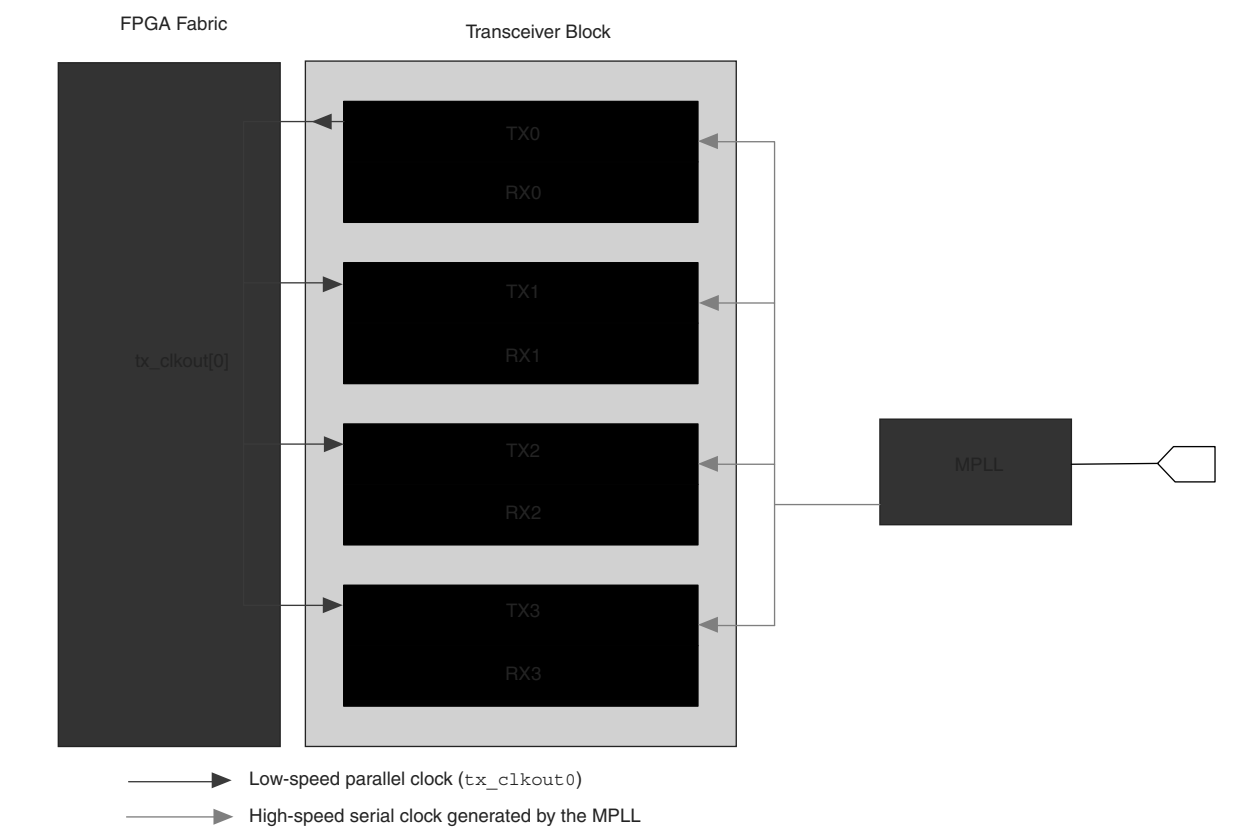
Port Name	Input/ Output	Description
<b>Clock Inputs to ALTGX_RECONFIG Instance</b>		
reconfig_clk	Input	<p>The frequency range of this clock depends on the following transceiver channel configuration modes:</p> <ul style="list-style-type: none"> <li>■ <b>Receiver only</b> (37.5 MHz to 50 MHz)</li> <li>■ <b>Receiver and Transmitter</b> (37.5 MHz to 50 MHz)</li> <li>■ <b>Transmitter only</b> (2.5 MHz to 50 MHz)</li> </ul> <p>By default, the Quartus® II software assigns a global clock resource to this port. This clock must be a free-running clock sourced from an I/O clock pin. Do not use dedicated transceiver REFCLK pins or any clocks generated by transceivers.</p>
<b>ALTGX and ALTGX_RECONFIG Interface Signals</b>		
reconfig_fromgxb [n..0]	Input	<p>An output port in the ALTGX instance and an input port in the ALTGX_RECONFIG instance. This signal is transceiver-block based. Therefore, the width of this signal increases in steps of 5 bits per transceiver block.</p> <p>In the ALTGX MegaWizard Plug-In Manager, the width of this signal depends on the number of channels you select in the <b>What is the number of channels?</b> option in the <b>General</b> screen.</p> <p>For example, if you select the number of channels in the ALTGX instance as follows:</p> <p>1 ≤ Channels ≤ 4, then the output port reconfig_fromgxb[4..0] = 5 bits</p> <p>5 ≤ Channels ≤ 8, then the output port reconfig_fromgxb[9..0] = 10 bits</p> <p>9 ≤ Channels ≤ 12, then the output port reconfig_fromgxb[14..0] = 15 bits</p> <p>13 ≤ Channels ≤ 16, then the output port reconfig_fromgxb[19..0] = 20 bits</p> <p>To connect the reconfig_fromgxb port between the ALTGX_RECONFIG instance and multiple ALTGX instances, follow these rules:</p> <ul style="list-style-type: none"> <li>■ Connect the reconfig_fromgxb[4..0] of ALTGX Instance 1 to the reconfig_fromgxb[4..0] of the ALTGX_RECONFIG instance. Connect the reconfig_fromgxb[] port of the next ALTGX instance to the next available bits of the ALTGX_RECONFIG instance, and so on.</li> <li>■ Connect the reconfig_fromgxb port of the ALTGX instance, which has the highest <b>What is the starting channel number?</b> option, to the MSB of the reconfig_fromgxb port of the ALTGX_RECONFIG instance.</li> </ul> <p>The Quartus II Fitter produces a warning if the dynamic reconfiguration option is enabled in the ALTGX instance but the reconfig_fromgxb and reconfig_togxb ports are not connected to the ALTGX_RECONFIG instance.</p>
reconfig_togxb [3..0]	Output	<p>An input port of the ALTGX instance and an output port of the ALTGX_RECONFIG instance. You must connect the reconfig_togxb[3..0] input port of every ALTGX instance controlled by the dynamic reconfiguration controller to the reconfig_togxb[3..0] output port of the ALTGX_RECONFIG instance.</p> <p>The width of this port is always fixed to 4 bits.</p>

### Option 1: Share a Single Transmitter Core Clock Between Transmitters

- Enable this option if you want tx\_clkout of the first channel (channel 0) of the transceiver block to provide the write clock to the Transmitter Phase Compensation FIFOs of the remaining channels in the transceiver block.
- This option is typically enabled when all the channels of a transceiver block have the same functional mode and data rate and are reconfigured to the identical functional mode and data rate.

Figure 3–11 shows the sharing of channel 0's tx\_clkout between all four regular channels of a transceiver block.

**Figure 3–11. Option 1 for Transmitter Core Clocking (Channel Reconfiguration Mode)**



**Table 1–4. Recommended Operating Conditions for Cyclone IV GX Devices (Part 2 of 2)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>CCA_GXB</sub>	Transceiver PMA and auxiliary power supply	—	2.375	2.5	2.625	V
V <sub>CCL_GXB</sub>	Transceiver PMA and auxiliary power supply	—	1.16	1.2	1.24	V
V <sub>I</sub>	DC input voltage	—	–0.5	—	3.6	V
V <sub>O</sub>	DC output voltage	—	0	—	V <sub>CCIO</sub>	V
T <sub>J</sub>	Operating junction temperature	For commercial use	0	—	85	°C
		For industrial use	–40	—	100	°C
t <sub>RAMP</sub>	Power supply ramp time	Standard power-on reset (POR) <sup>(7)</sup>	50 μs	—	50 ms	—
		Fast POR <sup>(8)</sup>	50 μs	—	3 ms	—
I <sub>Diode</sub>	Magnitude of DC current across PCI-clamp diode when enabled	—	—	—	10	mA

**Notes to Table 1–4:**

- (1) All V<sub>CCA</sub> pins must be powered to 2.5 V (even when PLLs are not used) and must be powered up and powered down at the same time.
- (2) You must connect V<sub>CCD\_PLL</sub> to V<sub>CCINT</sub> through a decoupling capacitor and ferrite bead.
- (3) Power supplies must rise monotonically.
- (4) V<sub>CCIO</sub> for all I/O banks must be powered up during device operation. Configurations pins are powered up by V<sub>CCIO</sub> of I/O Banks 3, 8, and 9 where I/O Banks 3 and 9 only support V<sub>CCIO</sub> of 1.5, 1.8, 2.5, 3.0, and 3.3 V. For fast passive parallel (FPP) configuration mode, the V<sub>CCIO</sub> level of I/O Bank 8 must be powered up to 1.5, 1.8, 2.5, 3.0, and 3.3 V.
- (5) You must set V<sub>CC\_CLKIN</sub> to 2.5 V if you use CLKIN as a high-speed serial interface (HSSI) refclk or as a DIFFCLK input.
- (6) The CLKIN pins in I/O Banks 3B and 8B can support single-ended I/O standard when the pins are used to clock left PLLs in non-transceiver applications.
- (7) The POR time for Standard POR ranges between 50 and 200 ms. V<sub>CCINT</sub>, V<sub>CCA</sub>, and V<sub>CCIO</sub> of I/O Banks 3, 8, and 9 must reach the recommended operating range within 50 ms.
- (8) The POR time for Fast POR ranges between 3 and 9 ms. V<sub>CCINT</sub>, V<sub>CCA</sub>, and V<sub>CCIO</sub> of I/O Banks 3, 8, and 9 must reach the recommended operating range within 3 ms.

## ESD Performance

This section lists the electrostatic discharge (ESD) voltages using the human body model (HBM) and charged device model (CDM) for Cyclone IV devices general purpose I/Os (GPIOs) and high-speed serial interface (HSSI) I/Os. Table 1–5 lists the ESD for Cyclone IV devices GPIOs and HSSI I/Os.

**Table 1–5. ESD for Cyclone IV Devices GPIOs and HSSI I/Os**

Symbol	Parameter	Passing Voltage	Unit
V <sub>ESDHBM</sub>	ESD voltage using the HBM (GPIOs) <sup>(1)</sup>	± 2000	V
	ESD using the HBM (HSSI I/Os) <sup>(2)</sup>	± 1000	V
V <sub>ESDCDM</sub>	ESD using the CDM (GPIOs)	± 500	V
	ESD using the CDM (HSSI I/Os) <sup>(2)</sup>	± 250	V

**Notes to Table 1–5:**

- (1) The passing voltage for EP4CGX15 and EP4CGX30 row I/Os is ±1000V.
- (2) This value is applicable only to Cyclone IV GX devices.

## Embedded Multiplier Specifications

Table 1–26 lists the embedded multiplier specifications for Cyclone IV devices.

**Table 1–26. Embedded Multiplier Specifications for Cyclone IV Devices**

Mode	Resources Used	Performance					Unit
	Number of Multipliers	C6	C7, I7, A7	C8	C8L, I8L	C9L	
9 × 9-bit multiplier	1	340	300	260	240	175	MHz
18 × 18-bit multiplier	1	287	250	200	185	135	MHz

## Memory Block Specifications

Table 1–27 lists the M9K memory block specifications for Cyclone IV devices.

**Table 1–27. Memory Block Performance Specifications for Cyclone IV Devices**

Memory	Mode	Resources Used		Performance					Unit
		LEs	M9K Memory	C6	C7, I7, A7	C8	C8L, I8L	C9L	
M9K Block	FIFO 256 × 36	47	1	315	274	238	200	157	MHz
	Single-port 256 × 36	0	1	315	274	238	200	157	MHz
	Simple dual-port 256 × 36 CLK	0	1	315	274	238	200	157	MHz
	True dual port 512 × 18 single CLK	0	1	315	274	238	200	157	MHz

## Configuration and JTAG Specifications

Table 1–28 lists the configuration mode specifications for Cyclone IV devices.

**Table 1–28. Passive Configuration Mode Specifications for Cyclone IV Devices <sup>(1)</sup>**

Programming Mode	V <sub>CCINT</sub> Voltage Level (V)	DCLK f <sub>MAX</sub>	Unit
Passive Serial (PS)	1.0 <sup>(3)</sup>	66	MHz
	1.2	133	MHz
Fast Passive Parallel (FPP) <sup>(2)</sup>	1.0 <sup>(3)</sup>	66	MHz
	1.2 <sup>(4)</sup>	100	MHz

**Notes to Table 1–28:**

- (1) For more information about PS and FPP configuration timing parameters, refer to the *Configuration and Remote System Upgrades in Cyclone IV Devices* chapter.
- (2) FPP configuration mode supports all Cyclone IV E devices (except for E144 package devices) and EP4CGX50, EP4CGX75, EP4CGX110, and EP4CGX150 only.
- (3) V<sub>CCINT</sub> = 1.0 V is only supported for Cyclone IV E 1.0 V core voltage devices.
- (4) Cyclone IV E devices support 1.2 V V<sub>CCINT</sub>. Cyclone IV E 1.2 V core voltage devices support 133 MHz DCLK f<sub>MAX</sub> for EP4CE6, EP4CE10, EP4CE15, EP4CE22, EP4CE30, and EP4CE40 only.