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Details

Product Status	Obsolete
Core Processor	CPU32
Core Size	32-Bit Single-Core
Speed	20MHz
Connectivity	CANbus, EBI/EMI, SCI, SPI
Peripherals	POR, PWM, WDT
Number of I/O	18
Program Memory Size	-
Program Memory Type	ROMless
EEPROM Size	-
RAM Size	7.5K x 8
Voltage - Supply (Vcc/Vdd)	4.75V ~ 5.25V
Data Converters	A/D 16x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	160-BQFP
Supplier Device Package	160-QFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc68376bgcab20

SECTION 1 INTRODUCTION

The MC68336 and the MC68376 are highly-integrated 32-bit microcontrollers, combining high-performance data manipulation capabilities with powerful peripheral subsystems.

MC68300 microcontrollers are built up from standard modules that interface through a common intermodule bus (IMB). Standardization facilitates rapid development of devices tailored for specific applications.

The MC68336 incorporates a 32-bit CPU (CPU32), a system integration module (SIM), a time processor unit (TPU), a configurable timer module (CTM4), a queued serial module (QSM), a 10-bit queued analog-to-digital converter module (QADC), a 3.5-Kbyte TPU emulation RAM module (TPURAM), and a 4-Kbyte standby RAM module (SRAM).

The MC68376 includes all of the aforementioned modules, plus a CAN 2.0B protocol controller module (TouCAN™) and an 8-Kbyte masked ROM (MRM).

The MC68336/376 can either synthesize the system clock signal from a fast reference or use an external clock input directly. Operation with a 4.194 MHz reference frequency is standard. The maximum system clock speed is 20.97 MHz. System hardware and software allow changes in clock rate during operation. Because MCU operation is fully static, register and memory contents are not affected by clock rate changes.

High-density complementary metal-oxide semiconductor (HCMOS) architecture makes the basic power consumption of the MCU low. Power consumption can be minimized by stopping the system clock. The CPU32 instruction set includes a low-power stop (LPSTOP) instruction that efficiently implements this capability.

Documentation for the Modular Microcontroller Family follows the modular construction of the devices in the product line. Each microcontroller has a comprehensive user's manual that provides sufficient information for normal operation of the device. The user's manual is supplemented by module reference manuals that provide detailed information about module operation and applications. Refer to Motorola publication *Advanced Microcontroller Unit (AMCU) Literature* (BR1116/D) for a complete listing of documentation.



2.2 CPU32 Registers

- A6–A0 — Address registers (index registers)
- A7 (SSP) — Supervisor stack pointer
- A7 (USP) — User stack pointer
- CCR — Condition code register (user portion of SR)
- D7–D0 — Data registers (index registers)
- DFC — Alternate function code register
- PC — Program counter
- SFC — Alternate function code register
- SR — Status register
- VBR — Vector base register
- X — Extend indicator
- N — Negative indicator
- Z — Zero indicator
- V — Two's complement overflow indicator
- C — Carry/borrow indicator

2.3 Pin and Signal Mnemonics

- ADDR[23:0] — Address Bus
- AN[59:48]/[3:0] — QADC Analog Input
- AN[w, x, y, z] — QADC Analog Input
- \overline{AS} — Address Strobe
- \overline{AVEC} — Autovector
- \overline{BERR} — Bus Error
- \overline{BG} — Bus Grant
- \overline{BGACK} — Bus Grant Acknowledge
- \overline{BKPT} — Breakpoint
- \overline{BR} — Bus Request
- CANRX0 — TouCAN Receive Data
- CANTX0 — TouCAN Transmit Data
- CLKOUT — System Clock
- $\overline{CS}[10:0]$ — Chip Selects
- \overline{CSBOOT} — Boot ROM Chip Select
- CPWM[8:5] — CTM Pulse Width Modulation Channel
- CTD[10:9]/[4:3] — CTM Double Action Channel
- CTM2C — CTM Modulus Clock
- DATA[15:0] — Data Bus
- \overline{DS} — Data Strobe

$\overline{\text{DSACK}}[1:0]$	— Data and Size Acknowledge
DSCLK	— Development Serial Clock
DSI	— Development Serial Input
DSO	— Development Serial Output
ECLK	— MC6800 Devices and Peripherals Bus Clock
ETRIG[2:1]	— QADC External Trigger
EXTAL	— Crystal Oscillator Input
FC[2:0]	— Function Codes
FREEZE	— Freeze
HALT	— Halt
$\overline{\text{IFETCH}}$	— Instruction Fetch
$\overline{\text{IPIPE}}$	— Instruction Pipeline
$\overline{\text{IRQ}}[7:1]$	— Interrupt Request
MA[2:0]	— QADC Multiplexed Address
MISO	— QSM Master In Slave Out
MODCLK	— Clock Mode Select
MOSI	— QSM Master Out Slave In
PCS[3:0]	— QSM Peripheral Chip-Selects
PQA[7:0]	— QADC Port A
PQB[7:0]	— QADC Port B
PC[6:0]	— SIM Port C
PE[7:0]	— SIM Port E
PF[7:0]	— SIM Port F
QUOT	— Quotient Out
R/W	— Read/Write
$\overline{\text{RESET}}$	— Reset
$\overline{\text{RMC}}$	— Read-Modify-Write Cycle
RXD	— SCI Receive Data
SCK	— QSPI Serial Clock
SIZ[1:0]	— Size
$\overline{\text{SS}}$	— Slave Select
T2CLK	— TPU Clock In
TPUCH[15:0]	— TPU Channel Signals
TSC	— Three-State Control
$\overline{\text{TSTME}}$	— Test Mode Enable
V_{RH}	— QADC High Reference Voltage
V_{RL}	— QADC Low Reference Voltage
XFC	— External Filter Capacitor
XTAL	— Crystal Oscillator Output

3.5 Signal Descriptions

The following tables define the MC68336/376 signals. **Table 3-4** shows signal origin, type, and active state. **Table 3-5** describes signal functions. Both tables are sorted alphabetically by mnemonic. MCU pins often have multiple functions. More than one description can apply to a pin.

Table 3-4 MC68336/376 Signal Characteristics

Signal Name	MCU Module	Signal Type	Active State
ADDR[23:0]	SIM	Bus	—
AN[59:48]/[3:0]	QADC	Input	—
AN[w, x, y, z]	QADC	Input	—
AS	SIM	Output	0
AVEC	SIM	Input	0
BERR	SIM	Input	0
BG	SIM	Output	0
BGACK	SIM	Input	0
BKPT	CPU32	Input	0
BR	SIM	Input	0
CLKOUT	SIM	Output	—
CANRX0 (MC68376 Only)	TouCAN	Input	—
CANTX0 (MC68376 Only)	TouCAN	Output	—
CS[10:0]	SIM	Output	0
CSBOOT	SIM	Output	0
CPWM[8:5]	CTM4	Output	—
CTD[10:9]/[4:3]	CTM4	Input/Output	—
CTM2C	CTM4	Input	—
DATA[15:0]	SIM	Bus	—
DS	SIM	Output	0
DSACK[1:0]	SIM	Input	0
DSCLK	CPU32	Input	Serial Clock
DSI	CPU32	Input	Serial Data
DSO	CPU32	Output	Serial Data
ECLK	SIM	Output	—
ETRIG[2:1]	QADC	Input	—
EXTAL	SIM	Input	—
FC[2:0]	SIM	Output	—
FREEZE	SIM	Output	1
HALT	SIM	Input/Output	0
IFETCH	CPU32	Output	0
IPIPE	CPU32	Output	0
IRQ[7:1]	SIM	Input	0
MA[2:0]	QADC	Output	1
MISO	QSM	Input/Output	—
MODCLK	SIM	Input	—
MOSI	QSM	Input/Output	—
PC[6:0]	SIM	Output	—
PCS[3:0]	QSM	Input/Output	—
PE[7:0]	SIM	Input/Output	—
PF[7:0]	SIM	Input/Output	—

4.2.4.2 Alternate Function Code Registers

Alternate function code registers (SFC and DFC) contain 3-bit function codes. Function codes can be considered extensions of the 24-bit linear address that optionally provide as many as eight 16-Mbyte address spaces. The processor automatically generates function codes to select address spaces for data and programs at the user and supervisor privilege levels and to select a CPU address space used for processor functions (such as breakpoint and interrupt acknowledge cycles).

Registers SFC and DFC are used by the MOVES instruction to specify explicitly the function codes of the memory address. The MOVEC instruction is used to transfer values to and from the alternate function code registers. This is a long-word transfer; the upper 29 bits are read as zeros and are ignored when written.

4.2.5 Vector Base Register (VBR)

The VBR contains the base address of the 1024-byte exception vector table, consisting of 256 exception vectors. Exception vectors contain the memory addresses of routines that begin execution at the completion of exception processing. More information on the VBR and exception processing can be found in **4.9 Exception Processing**.

4.3 Memory Organization

Memory is organized on a byte-addressable basis in which lower addresses correspond to higher order bytes. For example, the address N of a long-word data item corresponds to the address of the most significant byte of the highest order word. The address of the most significant byte of the low-order word is $N + 2$, and the address of the least significant byte of the long word is $N + 3$. The CPU32 requires long-word and word data and all instructions to be aligned on word boundaries. Refer to **Figure 4-6**. If this does not happen, an exception will occur when the CPU32 accesses the misaligned instruction or data. Data misalignment is not supported.

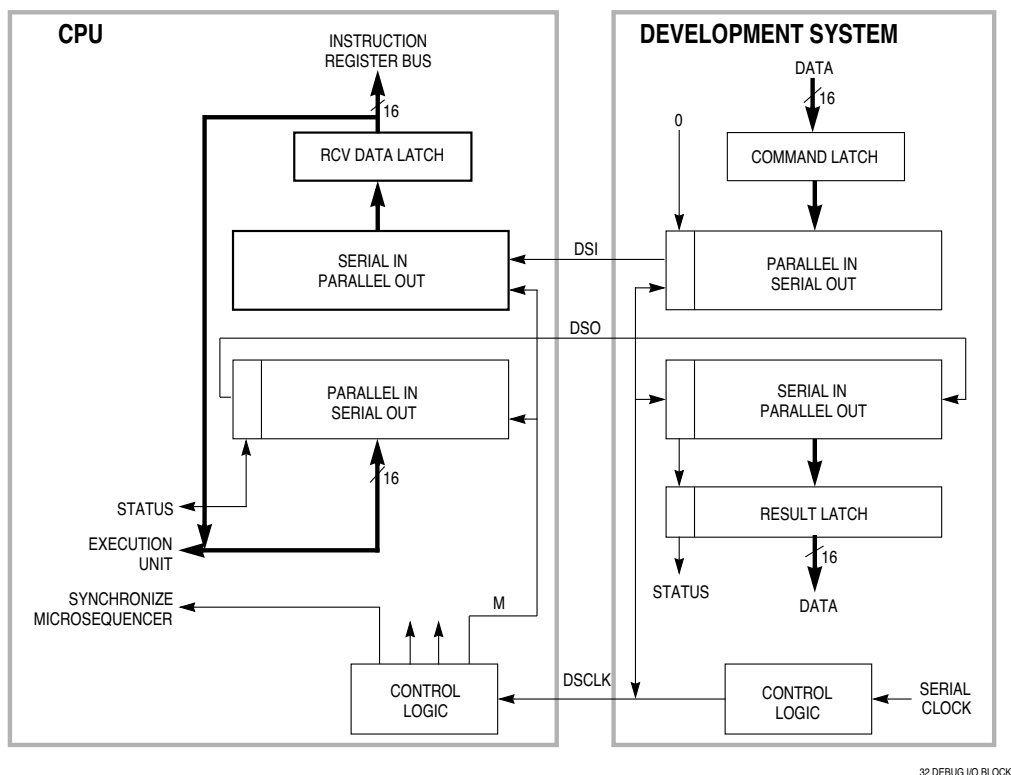


Figure 4-10 Debug Serial I/O Block Diagram

The serial interface uses a full-duplex synchronous protocol similar to the serial peripheral interface (SPI) protocol. The development system serves as the master of the serial link since it is responsible for the generation of DSCLK. If DSCLK is derived from the CPU32 system clock, development system serial logic is unhindered by the operating frequency of the target processor. Operable frequency range of the serial clock is from DC to one-half the processor system clock frequency.

The serial interface operates in full-duplex mode — data is transmitted and received simultaneously by both master and slave devices. In general, data transitions occur on the falling edge of DSCLK and are stable by the following rising edge of DSCLK. Data is transmitted MSB first, and is latched on the rising edge of DSCLK.

The serial data word is 17 bits wide, including 16 data bits and a status/control bit (refer to **Figure 4-11**). Bit 16 indicates the status of CPU-generated messages. **Table 4-7** shows the CPU-generated message types.

Table 5-2 Clock Control Multipliers

Modulus	Prescalers			
Y	[W:X] = 00	[W:X] = 01	[W:X] = 10	[W:X] = 11
000000	.03125	.625	.125	.25
000001	.0625	.125	.25	.5
000010	.09375	.1875	.375	.75
000011	.125	.25	.5	1
000100	.15625	.3125	.625	1.25
000101	.1875	.375	.75	1.5
000110	.21875	.4375	.875	1.75
000111	.25	.5	1	2
001000	.21825	.5625	1.125	2.25
001001	.3125	.625	1.25	2.5
001010	.34375	.6875	1.375	2.75
001011	.375	.75	1.5	3
001100	.40625	.8125	1.625	3.25
001101	.4375	.875	1.75	3.5
001110	.46875	.9375	1.875	3.75
001111	.5	1	2	4
010000	.53125	1.0625	2.125	4.25
010001	.5625	1.125	2.25	4.5
010010	.59375	1.1875	2.375	4.75
010011	.625	1.25	2.5	5
010100	.65625	1.3125	2.625	5.25
010101	.6875	1.375	2.75	5.5
010110	.71875	1.4375	2.875	5.75
010111	.75	1.5	3	6
011000	.78125	1.5625	3.125	6.25
011001	.8125	1.625	3.25	6.5
011010	.84375	1.6875	3.375	6.75
011011	.875	1.75	3.5	7
011100	.90625	1.8125	3.625	7.25
011101	.9375	1.875	3.75	7.5
011110	.96875	1.9375	3.875	7.75
011111	1	2	4	8

If bus termination signals remain unasserted, the MCU will continue to insert wait states, and the bus cycle will never end. If no peripheral responds to an access, or if an access is invalid, external logic should assert the $\overline{\text{BERR}}$ or $\overline{\text{HALT}}$ signals to abort the bus cycle (when $\overline{\text{BERR}}$ and $\overline{\text{HALT}}$ are asserted simultaneously, the CPU32 acts as though only $\overline{\text{BERR}}$ is asserted). When enabled, the SIM bus monitor asserts $\overline{\text{BERR}}$ when $\overline{\text{DSACK}}$ response time exceeds a predetermined limit. The bus monitor timeout period is determined by the BMT[1:0] field in SYPCR. The maximum bus monitor timeout period is 64 system clock cycles.

5.6.2.1 Read Cycle

During a read cycle, the MCU transfers data from an external memory or peripheral device. If the instruction specifies a long-word or word operation, the MCU attempts to read two bytes at once. For a byte operation, the MCU reads one byte. The portion of the data bus from which each byte is read depends on operand size, peripheral address, and peripheral port size. **Figure 5-10** is a flowchart of a word read cycle. Refer to **5.5.2 Dynamic Bus Sizing**, **5.5.4 Misaligned Operands**, and the *SIM Reference Manual* (SIMRM/AD) for more information.

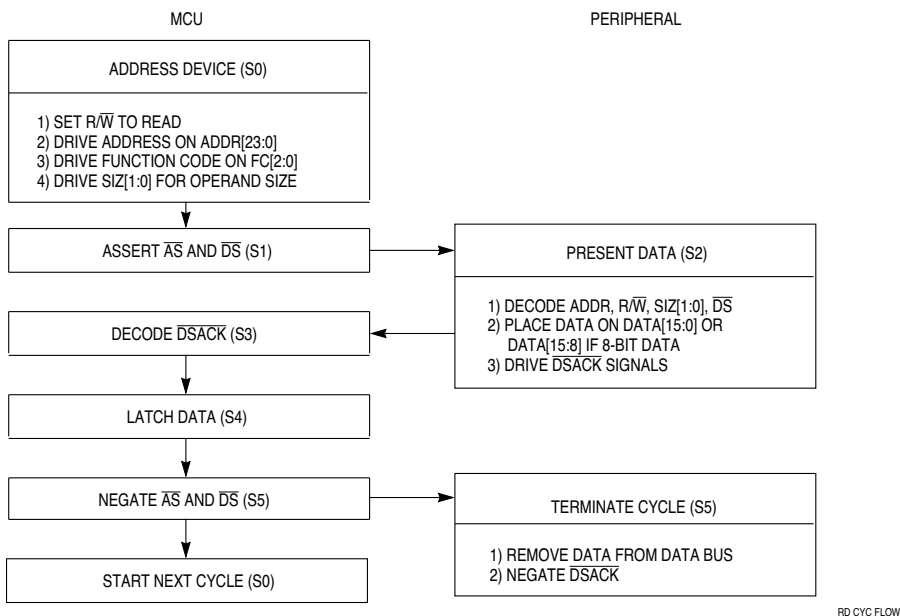


Figure 5-10 Word Read Cycle Flowchart

5.6.3 Fast Termination Cycles

When an external device has a fast access time, the chip-select circuit fast termination option can provide a two-cycle external bus transfer. Because the chip-select circuits are driven from the system clock, the bus cycle termination is inherently synchronized with the system clock.

If multiple chip-selects are to be used to provide control signals to a single device and match conditions occur simultaneously, all $\overline{\text{MODE}}$, STRB , and associated $\overline{\text{DSACK}}$ fields must be programmed to the same value. This prevents a conflict on the internal bus when the wait states are loaded into the $\overline{\text{DSACK}}$ counter shared by all chip-selects.

Fast termination cycles use internal handshaking signals generated by the chip-select logic. To initiate a transfer, the MCU asserts an address and the $\text{SIZ}[1:0]$ signals. When $\overline{\text{AS}}$, $\overline{\text{DS}}$, and R/W are valid, a peripheral device either places data on the bus (read cycle) or latches data from the bus (write cycle). At the appropriate time, chip-select logic asserts data and size acknowledge signals.

The $\overline{\text{DSACK}}$ option fields in the chip-select option registers determine whether internally generated $\overline{\text{DSACK}}$ or externally generated $\overline{\text{DSACK}}$ is used. The external $\overline{\text{DSACK}}$ lines are always active, regardless of the setting of the $\overline{\text{DSACK}}$ field in the chip-select option registers. Thus, an external $\overline{\text{DSACK}}$ can always terminate a bus cycle. Holding a $\overline{\text{DSACK}}$ line low will cause all external bus cycles to be three-cycle (zero wait states) accesses unless the chip-select option register specifies fast accesses.

For fast termination cycles, the fast termination encoding (%1110) must be used. Refer to **5.9.1 Chip-Select Registers** for information about fast termination setup.

To use fast termination, an external device must be fast enough to have data ready within the specified setup time (for example, by the falling edge of S4). Refer to **Table A-6** and **Figures A-6** and **A-7** for information about fast termination timing.

When fast termination is in use, $\overline{\text{DS}}$ is asserted during read cycles but not during write cycles. The STRB field in the chip-select option register used must be programmed with the address strobe encoding to assert the chip-select signal for a fast termination write.

5.6.4 CPU Space Cycles

Function code signals $\text{FC}[2:0]$ designate which of eight external address spaces is accessed during a bus cycle. Address space 7 is designated CPU space. CPU space is used for control information not normally associated with read or write bus cycles. Function codes are valid only while $\overline{\text{AS}}$ is asserted. Refer to **5.5.1.7 Function Codes** for more information on codes and encoding.

During a CPU space access, $\text{ADDR}[19:16]$ are encoded to reflect the type of access being made. **Figure 5-12** shows the three encodings used by 68300 family microcontrollers. These encodings represent breakpoint acknowledge (Type \$0) cycles, low power stop broadcast (Type \$3) cycles, and interrupt acknowledge (Type \$F) cycles. Refer to **5.8 Interrupts** for information about interrupt acknowledge bus cycles.

Table 5-16 Module Pin Functions During Reset

Module	Pin Mnemonic	Function
CPU32	DSI/IFETCH	DSI/IFETCH
	DSO/IPIPE	DSO/IPIPE
	BKPT/DSCLK	BKPT/DSCLK
CTM4	CPWM[8:5]	Discrete output
	CTD[10:9]/[4:3]	Discrete input
	CTM4C	Discrete input
QADC	PQA[7:5]/AN[59:57]	Discrete input
	PQA[4:3]/AN[56:55]/ETRIG[2:1]	Discrete input
	PQA[2:0]/AN[54:52]/MA[2:0]	Discrete input
	PQB[7:4]/AN[51:48]	Discrete input
	PQB[3:0]/AN[z, y, x, w]/AN[3:0]	Discrete input
QSM	PQS0/MISO	Discrete input
	PQS1/MOSI	Discrete input
	PQS2/SCK	Discrete input
	PQS3/PCS0/ \overline{SS}	Discrete input
	PQS[6:4]/PCS[3:1]	RXD
	PQS7/TXD	Discrete input
	RXD	Discrete input
TouCAN (MC68376 only)	CANRX0	TouCAN receive
	CANTX0	TouCAN transmit
TPU	TPUCH[15:0]	TPU input
	T2CLK	TCR2 clock

5.7.5 Pin States During Reset

It is important to keep the distinction between pin function and pin electrical state clear. Although control register values and mode select inputs determine pin function, a pin driver can be active, inactive or in high-impedance state while reset occurs. During power-on reset, pin state is subject to the constraints discussed in **5.7.7 Power-On Reset**.

NOTE

Pins that are not used should either be configured as outputs, or (if configured as inputs) pulled to the appropriate inactive state. This decreases additional I_{DD} caused by digital inputs floating near mid-supply level.

5.7.5.1 Reset States of SIM Pins

Generally, while \overline{RESET} is asserted, SIM pins either go to an inactive high-impedance state or are driven to their inactive states. After \overline{RESET} is released, mode selection occurs and reset exception processing begins. Pins configured as inputs must be driven to the desired active state. Pull-up or pull-down circuitry may be necessary. Pins configured as outputs begin to function after \overline{RESET} is released. **Table 5-17** is a summary of SIM pin states during reset.

9.4.3.6 Receiver Operation

The RE bit in SCCR1 enables (RE = 1) and disables (RE = 0) the receiver. The receiver contains a receive serial shifter and a parallel receive data register (RDR) located in the SCI data register (SCDR). The serial shifter cannot be directly accessed by the CPU32. The receiver is double-buffered, allowing data to be held in the RDR while other data is shifted in.

Receiver bit processor logic drives a state machine that determines the logic level for each bit-time. This state machine controls when the bit processor logic is to sample the RXD pin and also controls when data is to be passed to the receive serial shifter. A receive time clock is used to control sampling and synchronization. Data is shifted into the receive serial shifter according to the most recent synchronization of the receive time clock with the incoming data stream. From this point on, data movement is synchronized with the MCU system clock. Operation of the receiver state machine is detailed in the *QSM Reference Manual* (QSMRM/AD).

The number of bits shifted in by the receiver depends on the serial format. However, all frames must end with at least one stop bit. When the stop bit is received, the frame is considered to be complete, and the received data in the serial shifter is transferred to the RDR. The receiver data register flag (RDRF) is set when the data is transferred.

Noise errors, parity errors, and framing errors can be detected while a data stream is being received. Although error conditions are detected as bits are received, the noise flag (NF), the parity flag (PF), and the framing error (FE) flag in SCSR are not set until data is transferred from the serial shifter to the RDR.

RDRF must be cleared before the next transfer from the shifter can take place. If RDRF is set when the shifter is full, transfers are inhibited and the overrun error (OR) flag in SCSR is set. OR indicates that the RDR needs to be serviced faster. When OR is set, the data in the RDR is preserved, but the data in the serial shifter is lost. Because framing, noise, and parity errors are detected while data is in the serial shifter, FE, NF, and PF cannot occur at the same time as OR.

When the CPU32 reads SCSR and SCDR in sequence, it acquires status and data, and also clears the status flags. Reading SCSR acquires status and arms the clearing mechanism. Reading SCDR acquires data and clears SCSR.

When RIE in SCCR1 is set, an interrupt request is generated whenever RDRF is set. Because receiver status flags are set at the same time as RDRF, they do not have separate interrupt enables.

9.4.3.7 Idle-Line Detection

During a typical serial transmission, frames are transmitted isochronally and no idle time occurs between frames. Even when all the data bits in a frame are logic ones, the start bit provides one logic zero bit-time during the frame. An idle line is a sequence of contiguous ones equal to the current frame size. Frame size is determined by the state of the M bit in SCCR1.

When any of the end-of-queue conditions is recognized, a queue completion flag is set, and if enabled, an interrupt request is generated. The following situations prematurely terminate queue execution:

- Since queue 1 is higher in priority than queue 2, when a trigger event occurs on queue 1 during queue 2 execution, the execution of queue 2 is suspended by aborting execution of the CCW in progress, and queue 1 execution begins. When queue 1 execution is complete, queue 2 conversions restart with the first CCW entry in queue 2 or the first CCW of the queue 2 subqueue being executed when queue 2 was suspended. Alternately, conversions can restart with the aborted queue 2 CCW entry. The resume RES bit in QACR2 allows software to select where queue 2 begins after suspension. By choosing to re-execute all of the suspended queue 2 and subqueue CCWs, all of the samples are guaranteed to have been taken during the same scan pass. However, a high trigger event rate for queue 1 can prohibit the completion of queue 2. If this occurs, execution of queue 2 may begin with the aborted CCW entry.
- When a queue is disabled, any conversion taking place for that queue is aborted. Putting a queue into disabled mode does not power down the converter.
- When the operating mode of a queue is changed to another valid mode, any conversion taking place for that queue is aborted. The queue operating restarts at the beginning of the queue, once an appropriate trigger event occurs.
- When placed in low-power stop mode, the QADC aborts any conversion in progress.
- When the FRZ bit in the QADCMCR is set and the IMB FREEZE line is asserted, the QADC freezes at the end of the current conversion. When FREEZE is negated, the QADC resumes queue execution beginning with the next CCW entry.

8.12.8 Result Word Table

The result word table is a 40-word long, 10-bit wide RAM. The QADC writes a result word after completing an analog conversion specified by the corresponding CCW. The result word table can be read or written, but in normal operation, software reads the result word table to obtain analog conversions from the QADC. Unimplemented bits are read as zeros, and write operations have no effect. Refer to **D.5.9 Result Word Table** for register descriptions.

While there is only one result word table, the data can be accessed in three different alignment formats:

1. Right justified, with zeros in the higher order unused bits.
2. Left justified, with the most significant bit inverted to form a sign bit, and zeros in the unused lower order bits.
3. Left justified, with zeros in the unused lower order bits.

The left justified, signed format corresponds to a half-scale, offset binary, two's complement data format. The data is routed onto the IMB according to the selected format. The address used to access the table determines the data alignment format. All write operations to the result word table are right justified.



Table 13-7 Mask Examples for Normal/Extended Messages

Message Buffer (MB) /Mask	Base ID ID[28:18]	IDE	Extended ID ID[17:0]	Match
MB2	1 1 1 1 1 1 1 1 0 0 0	0	—	—
MB3	1 1 1 1 1 1 1 1 0 0 0	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	—
MB4	0 0 0 0 0 0 1 1 1 1 1	0	—	—
MB5	0 0 0 0 0 0 1 1 1 0 1	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	—
MB14	1 1 1 1 1 1 1 1 0 0 0	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	—
RX Global Mask	1 1 1 1 1 1 1 1 1 1 0		1 1 1 1 1 1 1 0 0 0 0 0 0 0 0 0 0 1	—
RX Message In	1 1 1 1 1 1 1 1 0 0 1	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	3 ¹
	1 1 1 1 1 1 1 1 0 0 1	0	—	2 ²
	1 1 1 1 1 1 1 1 0 0 1	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 0	— ³
	0 1 1 1 1 1 1 1 0 0 0	0	—	— ⁴
	0 1 1 1 1 1 1 1 0 0 0	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	— ⁵
RX 14 Mask	0 1 1 1 1 1 1 1 1 1 1		1 1 1 1 1 1 1 0 0 0 0 0 0 0 0 0 0 0	—
RX Message In	1 0 1 1 1 1 1 1 0 0 0	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	— ⁶
	0 1 1 1 1 1 1 1 0 0 0	1	0 1 0 1 0 1 0 1 0 1 0 1 0 1 0 1	14 ⁷

NOTES:

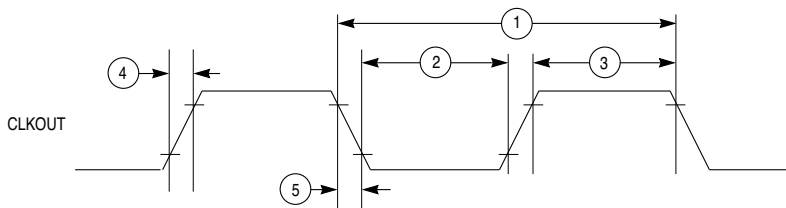
1. Match for extended format (MB3).
2. Match for standard format (MB2).
3. No match for MB3 because of ID0.
4. No match for MB2 because of ID28.
5. No match for MB3 because of ID28, match for MB14.
6. No match for MB14 because of ID27.
7. Match for MB14.

13.4.3 Bit Timing

The TouCAN module uses three 8-bit registers to set-up the bit timing parameters required by the CAN protocol. Control registers 1 and 2 (CANCTRL1, CANCTRL2) contain the PROPSEG, PSEG1, PSEG2, and the RJW fields which allow the user to configure the bit timing parameters. The prescaler divide register (PRESDIV) allows the user to select the ratio used to derive the S-clock from the system clock. The time quanta clock operates at the S-clock frequency. **Table 13-8** provides examples of system clock, CAN bit rate, and S-clock bit timing parameters. Refer to **APPENDIX D REGISTER SUMMARY** for more information on the bit timing registers.

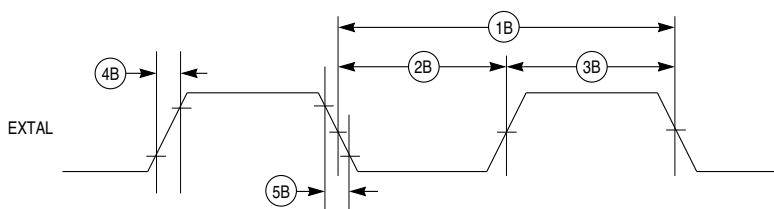
Table A-5 DC Characteristics
 $(V_{DD} \text{ and } V_{DDSYN} = 5.0 \text{ Vdc} \pm 5\%, V_{SS} = 0 \text{ Vdc}, T_A = T_L \text{ to } T_H)$

Num	Characteristic	Symbol	Min	Max	Unit
1	Input High Voltage	V_{IH}	$0.7 (V_{DD})$	$V_{DD} + 0.3$	V
2	Input Low Voltage	V_{IL}	$V_{SS} - 0.3$	$0.2 (V_{DD})$	V
3	Input Hysteresis ¹	V_{HYS}	0.5	—	V
4	Input Leakage Current ² $V_{in} = V_{DD} \text{ or } V_{SS}$ Input-only pins	I_{in}	-2.5	2.5	μA
5	High Impedance (Off-State) Leakage Current ² $V_{in} = V_{DD} \text{ or } V_{SS}$ All input/output and output pins	I_{OZ}	-2.5	2.5	μA
6	CMOS Output High Voltage ^{2, 3} $I_{OH} = -10.0 \mu\text{A}$ Group 1, 2, 4 input/output and all output pins	V_{OH}	$V_{DD} - 0.2$	—	V
7	CMOS Output Low Voltage ² $I_{OL} = 10.0 \mu\text{A}$ Group 1, 2, 4 input/output and all output pins	V_{OL}	—	0.2	V
8	Output High Voltage ^{2, 3} $I_{OH} = -0.8 \text{ mA}$ Group 1, 2, 4 input/output and all output pins	V_{OH}	$V_{DD} - 0.8$	—	V
9	Output Low Voltage ² $I_{OL} = 1.6 \text{ mA}$ Group 1 I/O Pins, CLKOUT, FREEZE/QUOT, $\overline{\text{IPIPE}}$ $I_{OL} = 5.3 \text{ mA}$ Group 2 and Group 4 I/O Pins, $\overline{\text{CSBOOT}}$, $\overline{\text{BG/CS}}$ $I_{OL} = 12 \text{ mA}$ Group 3	V_{OL}	— — —	0.4 0.4 0.4	V
10	Three State Control Input High Voltage	V_{IHTSC}	$1.6 (V_{DD})$	9.1	V
11	Data Bus Mode Select Pull-up Current ⁴ $V_{in} = V_{IL}$ DATA[15:0] $V_{in} = V_{IH}$ DATA[15:0]	I_{MSP}	— -15	-120 —	μA
12A	MC68336 V_{DD} Supply Current ⁵ RUN ⁶ RUN, TPU emulation mode LPSTOP, 4.194 MHz crystal, VCO Off (STSIM = 0) LPSTOP (External clock input frequency = maximum f_{sys})	I_{DD} I_{DD} S_{IDD} S_{IDD}	— — — —	140 150 3 7	mA mA mA mA
12B	MC68376 V_{DD} Supply Current ⁵ RUN ⁶ RUN, TPU emulation mode LPSTOP, 4.194 MHz crystal, VCO Off (STSIM = 0) LPSTOP (External clock input frequency = maximum f_{sys})	I_{DD} I_{DD} S_{IDD} S_{IDD}	— — — —	150 160 3 7	mA mA mA mA
13	Clock Synthesizer Operating Voltage	V_{DDSYN}	4.75	5.25	V
14	V_{DDSYN} Supply Current ⁵ 4.194 MHz crystal, VCO on, maximum f_{sys} External Clock, maximum f_{sys} LPSTOP, 4.194 MHz crystal, VCO off (STSIM = 0) 4.194 MHz crystal, V_{DD} powered down	I_{DDSYN} I_{DDSYN} S_{IDDSYN} I_{DDSYN}	— — — —	3 5 3 3	mA mA mA mA



68300 CLKOUT TIM

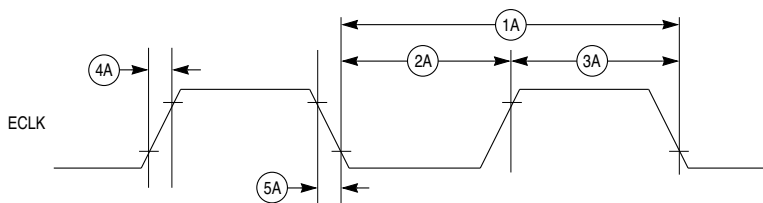
Figure A-1 CLKOUT Output Timing Diagram



NOTE: TIMING SHOWN WITH RESPECT TO 20% AND 70% V_{DD} .
PULSE WIDTH SHOWN WITH RESPECT TO 50% V_{DD} .

68300 EXT CLK INPUT TIM

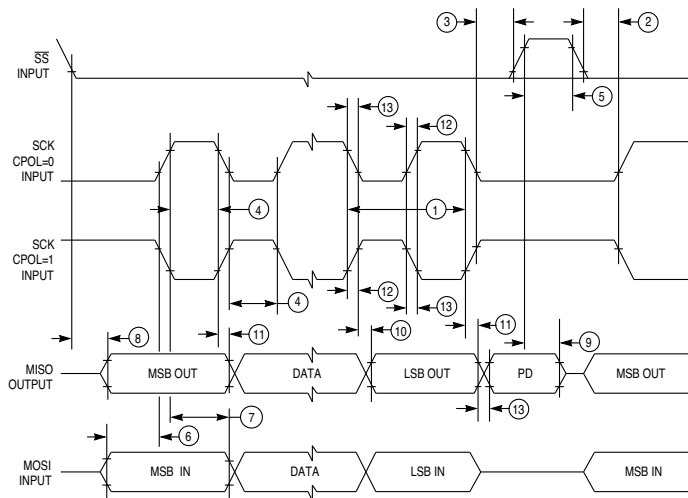
Figure A-2 External Clock Input Timing Diagram



NOTE: TIMING SHOWN WITH RESPECT TO 20% AND 70% V_{DD} .

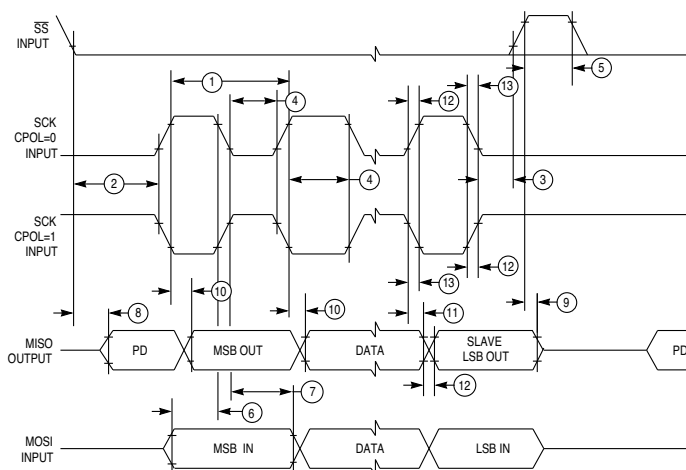
68300 ECLK OUTPUT TIM

Figure A-3 ECLK Output Timing Diagram



QSPI SLV CPHA0

Figure A-18 QSPI Timing — Slave, CPHA = 0



QSPI SLV CPHA1

Figure A-19 QSPI Timing — Slave, CPHA = 1



IVB[7:0] — Interrupt Vector Base

Only the upper six bits of IVB[7:0] can be initialized. During interrupt arbitration, the vector provided by the QADC is made up of IVB[7:2], plus two low-order bits that identify one of the four QADC interrupt sources. Once IVB is written, the two low-order bits always read as zeros.

D.5.4 Port A/B Data Register

PORTQA — Port QA Data Register

\$YFF206

PORTQB — Port QB Data Register

\$YFF207

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PQA7	PQA6	PQA5	PQA4	PQA3	PQA2	PQA1	PQA0	PQB7	PQB6	PQB5	PQB4	PQB3	PQB2	PQB1	PQB0
RESET:															
U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	U
ANALOG CHANNEL:															
AN59	AN58	AN57	AN56	AN55	AN54	AN53	AN52	AN51	AN50	AN49	AN48	AN3	AN2	AN1	AN0
EXTERNAL TRIGGER INPUTS:															
ETRIG2 ETRIG1															
MULTIPLEXED ADDRESS OUTPUTS:															



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