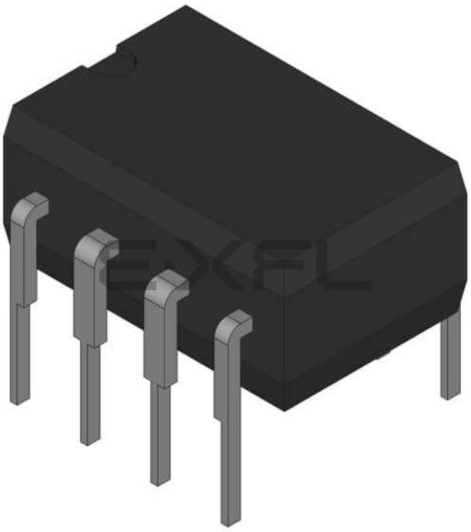


What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	HC08
Core Size	8-Bit
Speed	8MHz
Connectivity	-
Peripherals	LVD, POR, PWM
Number of I/O	5
Program Memory Size	1.5KB (1.5K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	8-DIP (0.300", 7.62mm)
Supplier Device Package	8-PDIP
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mc68hc908qt2cpe

MC68HC908QY4
MC68HC908QT4
MC68HC908QY2
MC68HC908QT2
MC68HC908QY1
MC68HC908QT1

Data Sheet

To provide the most up-to-date information, the revision of our documents on the World Wide Web will be the most current. Your printed copy may be an earlier revision. To verify you have the latest information available, refer to:

http://freescale.com/

- On-chip in-application programmable FLASH memory (with internal program/erase voltage generation)
  - MC68HC908QY4 and MC68HC908QT4 — 4096 bytes
  - MC68HC908QY2, MC68HC908QY1, MC68HC908QT2, and MC68HC908QT1 — 1536 bytes
- 128 bytes of on-chip random-access memory (RAM)
- 2-channel, 16-bit timer interface module (TIM)
- 4-channel, 8-bit analog-to-digital converter (ADC) on MC68HC908QY2, MC68HC908QY4, MC68HC908QT2, and MC68HC908QT4
- 5 or 13 bidirectional input/output (I/O) lines and one input only:
  - Six shared with keyboard interrupt function and ADC
  - Two shared with timer channels
  - One shared with external interrupt (IRQ)
  - Eight extra I/O lines on 16-pin package only
  - High current sink/source capability on all port pins
  - Selectable pullups on all ports, selectable on an individual bit basis
  - Three-state ability on all port pins
- 6-bit keyboard interrupt with wakeup feature (KBI)
- Low-voltage inhibit (LVI) module features:
  - Software selectable trip point in CONFIG register
- System protection features:
  - Computer operating properly (COP) watchdog
  - Low-voltage detection with reset
  - Illegal opcode detection with reset
  - Illegal address detection with reset
- External asynchronous interrupt pin with internal pullup ($\overline{\text{IRQ}}$) shared with general-purpose input pin
- Master asynchronous reset pin ($\overline{\text{RST}}$) shared with general-purpose input/output (I/O) pin
- Power-on reset
- Internal pullups on $\overline{\text{IRQ}}$ and $\overline{\text{RST}}$ to reduce external components
- Memory mapped I/O registers
- Power saving stop and wait modes
- MC68HC908QY4, MC68HC908QY2, and MC68HC908QY1 are available in these packages:
  - 16-pin plastic dual in-line package (PDIP)
  - 16-pin small outline integrated circuit (SOIC) package
  - 16-pin thin shrink small outline package (TSSOP)
- MC68HC908QT4, MC68HC908QT2, and MC68HC908QT1 are available in these packages:
  - 8-pin PDIP
  - 8-pin SOIC
  - 8-pin dual flat no lead (DFN) package

ERASE — Erase Control Bit

This read/write bit configures the memory for erase operation. ERASE is interlocked with the PGM bit such that both bits cannot be equal to 1 or set to 1 at the same time.

1 = Erase operation selected
0 = Erase operation unselected

PGM — Program Control Bit

This read/write bit configures the memory for program operation. PGM is interlocked with the ERASE bit such that both bits cannot be equal to 1 or set to 1 at the same time.

1 = Program operation selected
0 = Program operation unselected

2.6.2 FLASH Page Erase Operation

Use the following procedure to erase a page of FLASH memory. A page consists of 64 consecutive bytes starting from addresses $XX00, $XX40, $XX80, or $XXC0. The 48-byte user interrupt vectors area also forms a page. Any FLASH memory page can be erased alone.

1. Set the ERASE bit and clear the MASS bit in the FLASH control register.
2. Read the FLASH block protect register.
3. Write any data to any FLASH location within the address range of the block to be erased.
4. Wait for a time, t_{NVS} (minimum 10 μs).
5. Set the HVEN bit.
6. Wait for a time, t_{Erase} (minimum 1 ms or 4 ms).
7. Clear the ERASE bit.
8. Wait for a time, t_{NVH} (minimum 5 μs).
9. Clear the HVEN bit.
10. After time, t_{RCV} (typical 1 μs), the memory can be accessed in read mode again.

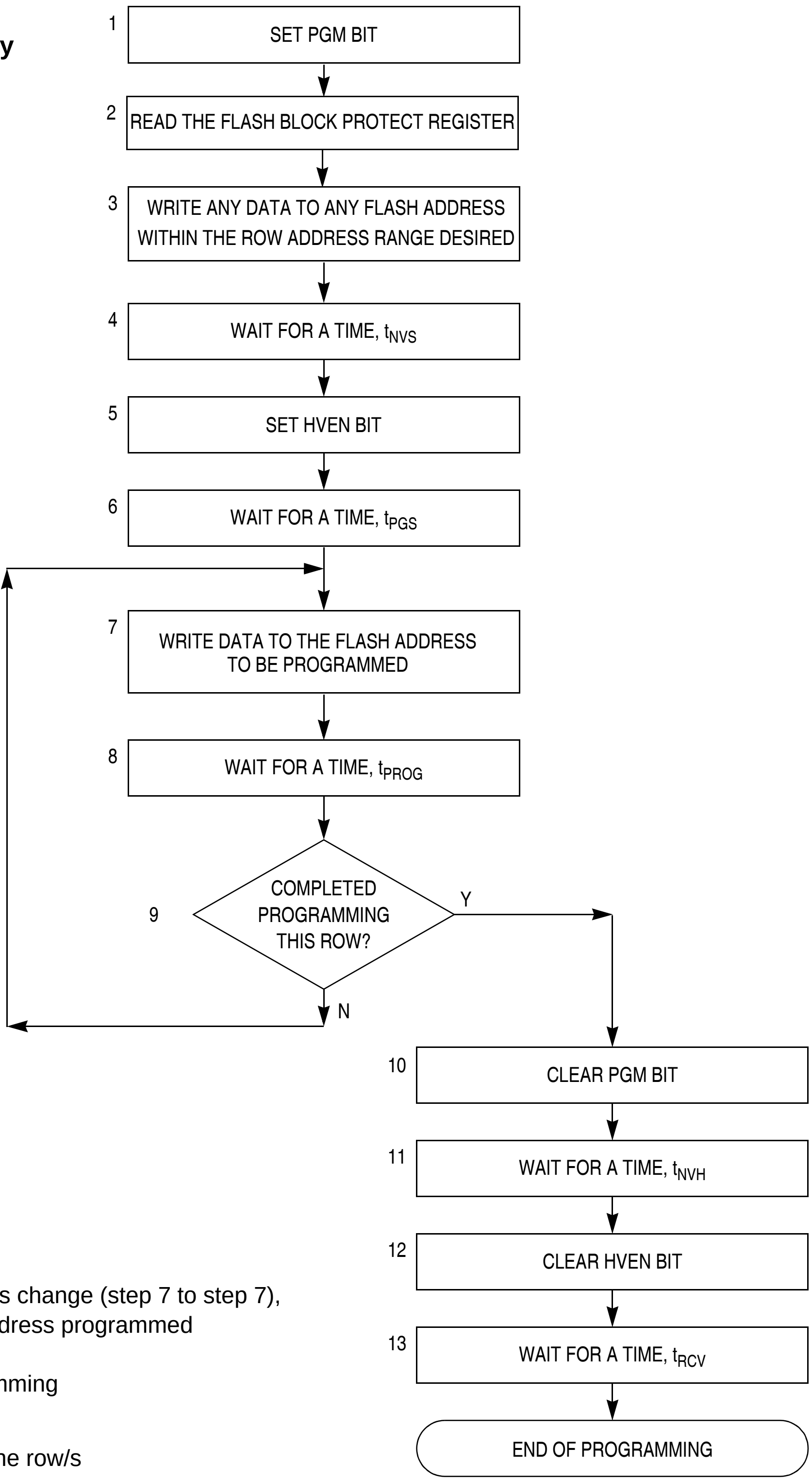
NOTE

Programming and erasing of FLASH locations cannot be performed by code being executed from the FLASH memory. While these operations must be performed in the order as shown, but other unrelated operations may occur between the steps.

CAUTION

A page erase of the vector page will erase the internal oscillator trim values at $FFC0 and $FFC1.

In applications that require more than 1000 program/erase cycles, use the 4 ms page erase specification to get improved long-term reliability. Any application can use this 4 ms page erase specification. However, in applications where a FLASH location will be erased and reprogrammed less than 1000 times, and speed is important, use the 1 ms page erase specification to get a shorter cycle time.



NOTES:

The time between each FLASH address change (step 7 to step 7), or the time between the last FLASH address programmed to clearing PGM bit (step 7 to step 10) must not exceed the maximum programming time, t_{PROG} max.

This row program algorithm assumes the row/s to be programmed are initially erased.

Figure 2-4. FLASH Programming Flowchart

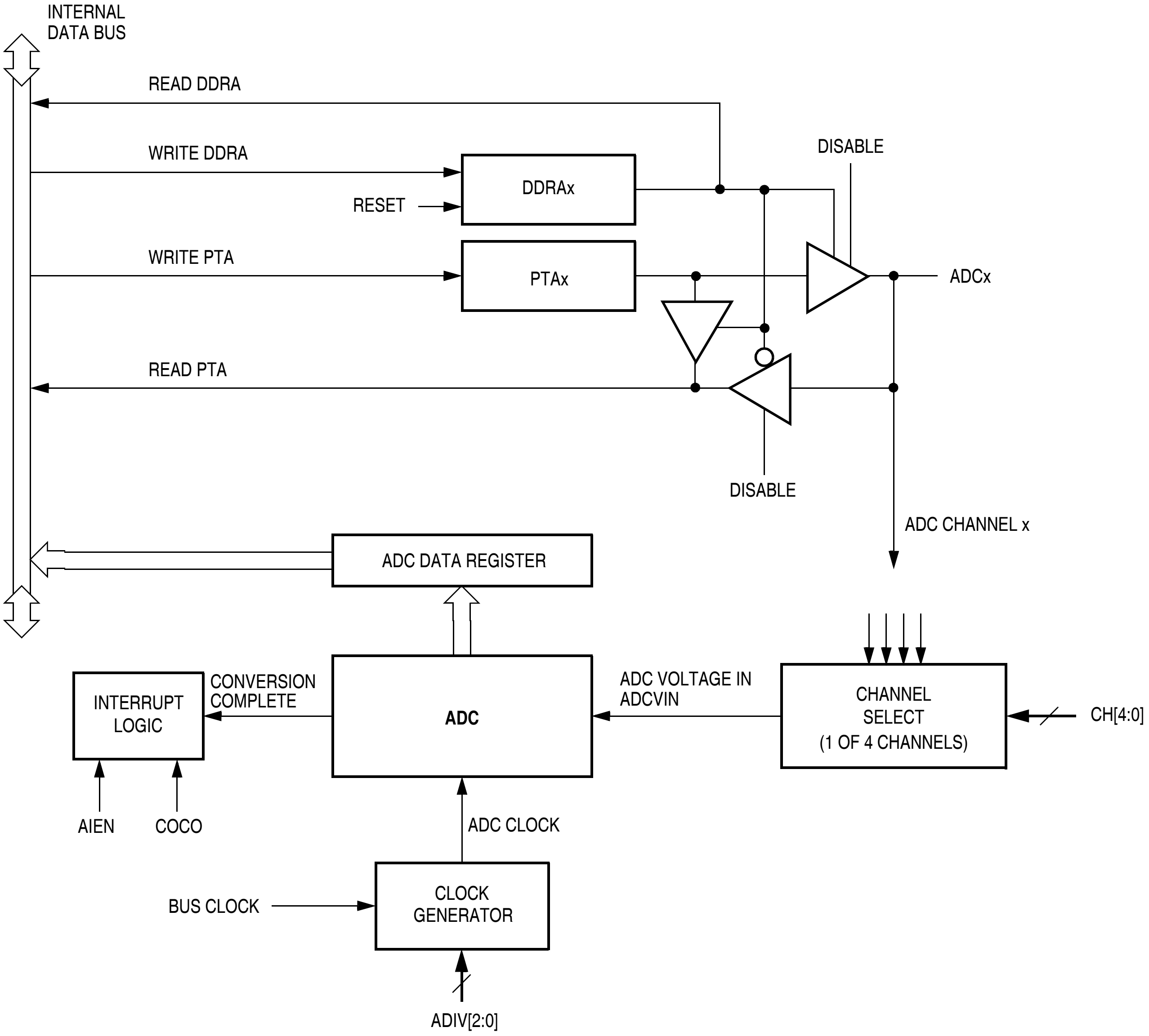
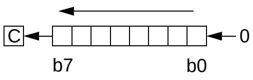
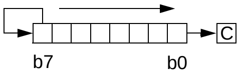


Figure 3-2. ADC Block Diagram

7.7 Instruction Set Summary

Table 7-1 provides a summary of the M68HC08 instruction set.

Table 7-1. Instruction Set Summary (Sheet 1 of 6)

Source Form	Operation	Description	Effect on CCR V	H	I	N	Z	C	Address Mode	Opcode	Operand	Cycles
ADC #*opr* ADC *opr* ADC *opr* ADC *opr*,X ADC *opr*,X ADC ,X ADC *opr*,SP ADC *opr*,SP	Add with Carry	A ← (A) + (M) + (C)	↕	↕	–	↕	↕	↕	IMM DIR EXT IX2 IX1 IX SP1 SP2	A9 B9 C9 D9 E9 F9 9EE9 9ED9	ii dd hh ll ee ff ff ff ee ff	2 3 4 4 3 2 4 5
ADD #*opr* ADD *opr* ADD *opr* ADD *opr*,X ADD *opr*,X ADD ,X ADD *opr*,SP ADD *opr*,SP	Add without Carry	A ← (A) + (M)	↕	↕	–	↕	↕	↕	IMM DIR EXT IX2 IX1 IX SP1 SP2	AB BB CB DB EB FB 9EEB 9EDB	ii dd hh ll ee ff ff ff ee ff	2 3 4 4 3 2 4 5
AIS #*opr*	Add Immediate Value (Signed) to SP	SP ← (SP) + (16 « M)	–	–	–	–	–	–	IMM	A7	ii	2
AIX #*opr*	Add Immediate Value (Signed) to H:X	H:X ← (H:X) + (16 « M)	–	–	–	–	–	–	IMM	AF	ii	2
AND #*opr* AND *opr* AND *opr* AND *opr*,X AND *opr*,X AND ,X AND *opr*,SP AND *opr*,SP	Logical AND	A ← (A) & (M)	0	–	–	↕	↕	–	IMM DIR EXT IX2 IX1 IX SP1 SP2	A4 B4 C4 D4 E4 F4 9EE4 9ED4	ii dd hh ll ee ff ff ff ee ff	2 3 4 4 3 2 4 5
ASL *opr* ASLA ASLX ASL *opr*,X ASL ,X ASL *opr*,SP	Arithmetic Shift Left (Same as LSL)		↕	–	–	↕	↕	↕	DIR INH INH IX1 IX SP1	38 48 58 68 78 9E68	dd ff ff	4 1 1 4 3 5
ASR *opr* ASRA ASRX ASR *opr*,X ASR *opr*,X ASR *opr*,SP	Arithmetic Shift Right		↕	–	–	↕	↕	↕	DIR INH INH IX1 IX SP1	37 47 57 67 77 9E67	dd ff ff	4 1 1 4 3 5
BCC *rel*	Branch if Carry Bit Clear	PC ← (PC) + 2 + rel ? (C) = 0	–	–	–	–	–	–	REL	24	rr	3
BCLR *n*, *opr*	Clear Bit n in M	Mn ← 0	–	–	–	–	–	–	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	11 13 15 17 19 1B 1D 1F	dd dd dd dd dd dd dd dd	4 4 4 4 4 4 4 4
BCS *rel*	Branch if Carry Bit Set (Same as BLO)	PC ← (PC) + 2 + *rel* ? (C) = 1	–	–	–	–	–	–	REL	25	rr	3
BEQ *rel*	Branch if Equal	PC ← (PC) + 2 + *rel* ? (Z) = 1	–	–	–	–	–	–	REL	27	rr	3
BGE *opr*	Branch if Greater Than or Equal To (Signed Operands)	PC ← (PC) + 2 + *rel* ? (N ⊕ V) = 0	–	–	–	–	–	–	REL	90	rr	3
BGT *opr*	Branch if Greater Than (Signed Operands)	PC ← (PC) + 2 + *rel* ? (Z) | (N ⊕ V) = 0	–	–	–	–	–	–	REL	92	rr	3
BHCC *rel*	Branch if Half Carry Bit Clear	PC ← (PC) + 2 + *rel* ? (H) = 0	–	–	–	–	–	–	REL	28	rr	3
BHCS *rel*	Branch if Half Carry Bit Set	PC ← (PC) + 2 + *rel* ? (H) = 1	–	–	–	–	–	–	REL	29	rr	3
BHI *rel*	Branch if Higher	PC ← (PC) + 2 + *rel* ? (C) | (Z) = 0	–	–	–	–	–	–	REL	22	rr	3

Chapter 8
External Interrupt (IRQ)

8.1 Introduction

The $\overline{\text{IRQ}}$ pin (external interrupt), shared with PTA2 (general purpose input) and keyboard interrupt (KBI), provides a maskable interrupt input

8.2 Features

Features of the IRQ module include the following:

- External interrupt pin, $\overline{\text{IRQ}}$
- $\overline{\text{IRQ}}$ interrupt control bits
- Programmable edge-only or edge and level interrupt sensitivity
- Automatic interrupt acknowledge
- Selectable internal pullup resistor

8.3 Functional Description

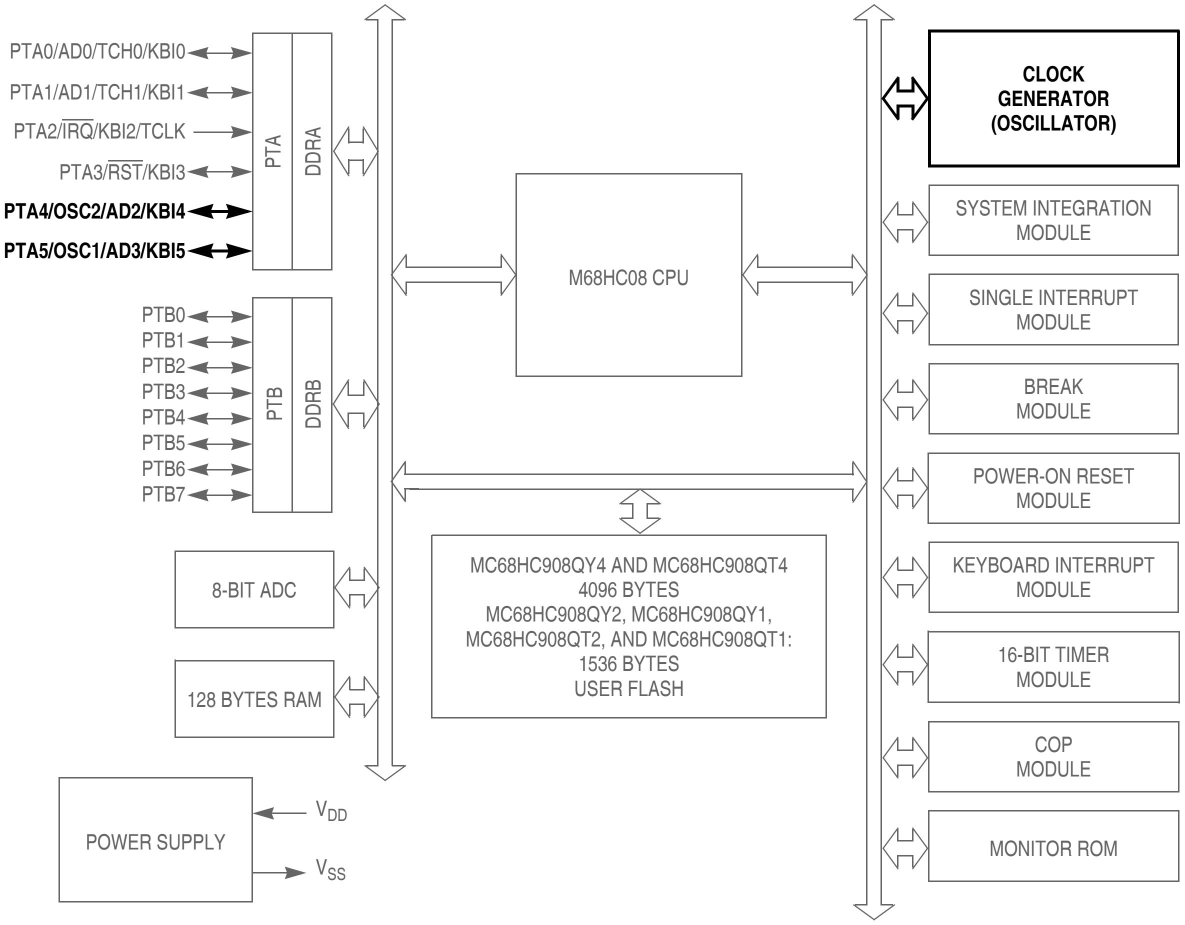
$\overline{\text{IRQ}}$ pin functionality is enabled by setting configuration register 2 (CONFIG2) IRQEN bit accordingly. A zero disables the IRQ function and PTA2 will assume the other shared functionalities. A one enables the IRQ function.

A low level applied to the external interrupt request ($\overline{\text{IRQ}}$) pin can latch a CPU interrupt request. Figure 8-2 shows the structure of the IRQ module.

Interrupt signals on the $\overline{\text{IRQ}}$ pin are latched into the IRQ latch. The IRQ latch remains set until one of the following actions occurs:

- IRQ vector fetch — An IRQ vector fetch automatically generates an interrupt acknowledge signal that clears the IRQ latch.
- Software clear — Software can clear the IRQ latch by writing a 1 to the ACK bit in the interrupt status and control register (INTSCR).
- Reset — A reset automatically clears the IRQ latch.

The external interrupt pin is falling-edge-triggered out of reset and is software-configurable to be either falling-edge or falling-edge and low-level triggered. The MODE bit in INTSCR controls the triggering sensitivity of the $\overline{\text{IRQ}}$ pin.



$\overline{\text{RST}}$, $\overline{\text{IRQ}}$: Pins have internal (about 30K Ohms) pull up

PTA[0:5]: High current sink and source capability

PTA[0:5]: Pins have programmable keyboard interrupt and pull up

PTB[0:7]: Not available on 8-pin devices – MC68HC908QT1, MC68HC908QT2, and MC68HC908QT4 (see note in 12.1 Introduction)

ADC: Not available on the MC68HC908QY1 and MC68HC908QT1

Figure 11-1. Block Diagram Highlighting OSC Block and Pins

11.3.1 Internal Oscillator

The internal oscillator circuit is designed for use with no external components to provide a clock source with tolerance less than ±25% untrimmed.An 8-bit trimming register allows adjustment to a tolerance of less than ±5%.

The internal oscillator will generate a clock of 12.8 MHz typical (INTCLK) resulting in a bus speed (internal clock ÷ 4) of 3.2 MHz. 3.2 MHz came from the maximum bus speed guaranteed at 3 V which is 4 MHz.Since the internal oscillator will have a ±25% tolerance (pre-trim), then the +25% case should not allow a frequency higher than 4 MHz:

3.2 MHz + 25% = 4 MHz

Figure 11-3 shows how BUSCLKX4 is derived from INTCLK and, like the RC oscillator, OSC2 can output BUSCLKX4 by setting OSC2EN in PTAPUE register. See Chapter 12 Input/Output Ports (PORTS)

11.3.1.1 Internal Oscillator Trimming

The 8-bit trimming register, OSCTRIM, allows a clock period adjust of +127 and –128 steps. Increasing OSCTRIM value increases the clock period. Trimming allows the internal clock frequency to be set to 12.8 MHz ± 5%.

All devices are factory programmed with trim values in reserved FLASH memory locations $FFC0 and $FFC1. The trim value is not automatically loaded into the OSCTRIM register. User software must copy the trim value from $FFC0 or $FFC1 into OSCTRIM if needed. The factory trim value provides the accuracy required for communication using forced monitor mode. Some production programmers erase the factory trim values, so confirm with your programmer vendor that the trim values at $FFC0 and $FFC1 are preserved, or are re-trimmed. Trimming the device in the user application board will provide the most accurate trim value.

11.3.1.2 Internal to External Clock Switching

When external clock source (external OSC, RC, or XTAL) is desired, the user must perform the following steps:

1. For external crystal circuits only, OSCOPT[1:0] = 1:1: To help precharge an external crystal oscillator, set PTA4 (OSC2) as an output and drive high for several cycles. This may help the crystal circuit start more robustly.
2. Set CONFIG2 bits OSCOPT[1:0] according to . The oscillator module control logic will then set OSC1 as an external clock input and, if the external crystal option is selected, OSC2 will also be set as the clock output.
3. Create a software delay to wait the stabilization time needed for the selected clock source (crystal, resonator, RC) as recommended by the component manufacturer. A good rule of thumb for crystal oscillators is to wait 4096 cycles of the crystal frequency, i.e., for a 4-MHz crystal, wait approximately 1 msec.
4. After the manufacturer’s recommended delay has elapsed, the ECGON bit in the OSC status register (OSCSTAT) needs to be set by the user software.
5. After ECGON set is detected, the OSC module checks for oscillator activity by waiting two external clock rising edges.
6. The OSC module then switches to the external clock. Logic provides a glitch free transition.
7. The OSC module first sets the ECGST bit in the OSCSTAT register and then stops the internal oscillator.

NOTE

Once transition to the external clock is done, the internal oscillator will only be reactivated with reset. No post-switch clock monitor feature is implemented (clock does not switch back to internal if external clock dies).

11.3.2 External Oscillator

The external clock option is designed for use when a clock signal is available in the application to provide a clock source to the microcontroller. The OSC1 pin is enabled as an input by the oscillator module. The clock signal is used directly to create BUSCLKX4 and also divided by two to create BUSCLKX2.

In this configuration, the OSC2 pin cannot output BUSCLKX4.So the OSC2EN bit in the port A pullup enable register will be clear to enable PTA4 I/O functions on the pin

13.5.2 SIM Counter During Stop Mode Recovery

The SIM counter also is used for stop mode recovery. The STOP instruction clears the SIM counter. After an interrupt, break, or reset, the SIM senses the state of the short stop recovery bit, SSREC, in the configuration register 1 (CONFIG1). If the SSREC bit is a 1, then the stop recovery is reduced from the normal delay of 4096 BUSCLKX4 cycles down to 32 BUSCLKX4 cycles. This is ideal for applications using canned oscillators that do not require long start-up times from stop mode. External crystal applications should use the full stop recovery time, that is, with SSREC cleared in the configuration register 1 (CONFIG1).

13.5.3 SIM Counter and Reset States

External reset has no effect on the SIM counter (see 13.7.2 Stop Mode for details.) The SIM counter is free-running after all reset states. See 13.4.2 Active Resets from Internal Sources for counter control and internal reset recovery sequences.

13.6 Exception Control

Normal sequential program execution can be changed in three different ways:

1. Interrupts
    a. Maskable hardware CPU interrupts
    b. Non-maskable software interrupt instruction (SWI)
2. Reset
3. Break interrupts

13.6.1 Interrupts

An interrupt temporarily changes the sequence of program execution to respond to a particular event. Figure 13-7 flow charts the handling of system interrupts.

Interrupts are latched, and arbitration is performed in the SIM at the start of interrupt processing. The arbitration result is a constant that the CPU uses to determine which vector to fetch. Once an interrupt is latched by the SIM, no other interrupt can take precedence, regardless of priority, until the latched interrupt is serviced (or the I bit is cleared).

At the beginning of an interrupt, the CPU saves the CPU register contents on the stack and sets the interrupt mask (I bit) to prevent additional interrupts. At the end of an interrupt, the RTI instruction recovers the CPU register contents from the stack so that normal processing can resume. Figure 13-8 shows interrupt entry timing. Figure 13-9 shows interrupt recovery timing.

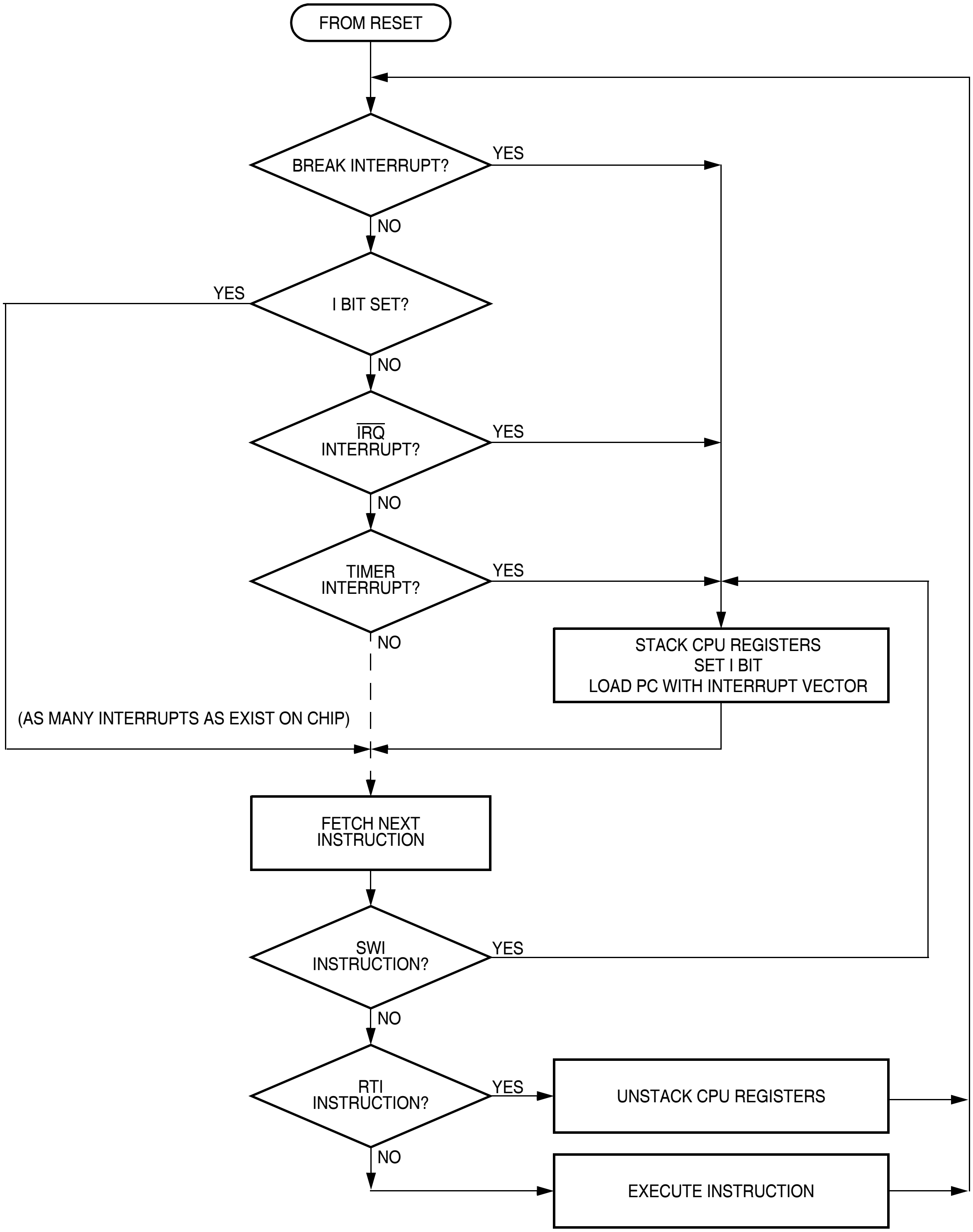


Figure 13-7. Interrupt Processing

control the output are the ones written to last. TSC0 controls and monitors the buffered output compare function, and TIM channel 1 status and control register (TSC1) is unused. While the MS0B bit is set, the channel 1 pin, TCH1, is available as a general-purpose I/O pin.

NOTE

In buffered output compare operation, do not write new output compare values to the currently active channel registers. User software should track the currently active channel to prevent writing a new value to the active channel. Writing to the active channel registers is the same as generating unbuffered output compares.

14.4.4 Pulse Width Modulation (PWM)

By using the toggle-on-overflow feature with an output compare channel, the TIM can generate a PWM signal. The value in the TIM counter modulo registers determines the period of the PWM signal. The channel pin toggles when the counter reaches the value in the TIM counter modulo registers. The time between overflows is the period of the PWM signal

As Figure 14-3 shows, the output compare value in the TIM channel registers determines the pulse width of the PWM signal. The time between overflow and output compare is the pulse width. Program the TIM to clear the channel pin on output compare if the state of the PWM pulse is logic 1 (ELSxA = 0). Program the TIM to set the pin if the state of the PWM pulse is logic 0 (ELSxA = 1).

The value in the TIM counter modulo registers and the selected prescaler output determines the frequency of the PWM output The frequency of an 8-bit PWM signal is variable in 256 increments. Writing $00FF (255) to the TIM counter modulo registers produces a PWM period of 256 times the internal bus clock period if the prescaler select value is 000. See 14.9.1 TIM Status and Control Register.

The value in the TIM channel registers determines the pulse width of the PWM output. The pulse width of an 8-bit PWM signal is variable in 256 increments. Writing $0080 (128) to the TIM channel registers produces a duty cycle of 128/256 or 50%.

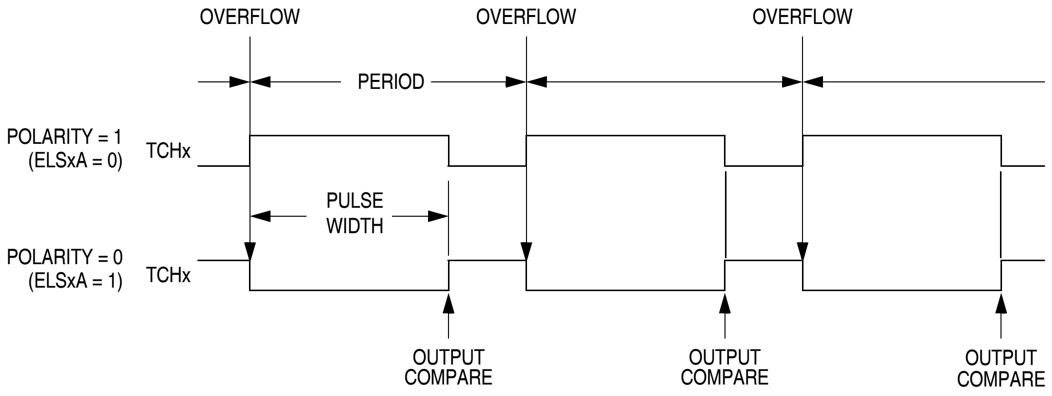


Figure 14-3. PWM Period and Pulse Width

15.3.1 Functional Description

Figure 15-9 shows a simplified diagram of monitor mode entry.

The monitor module receives and executes commands from a host computer. Figure 15-10, Figure 15-11, and Figure 15-12 show example circuits used to enter monitor mode and communicate with a host computer via a standard RS-232 interface.

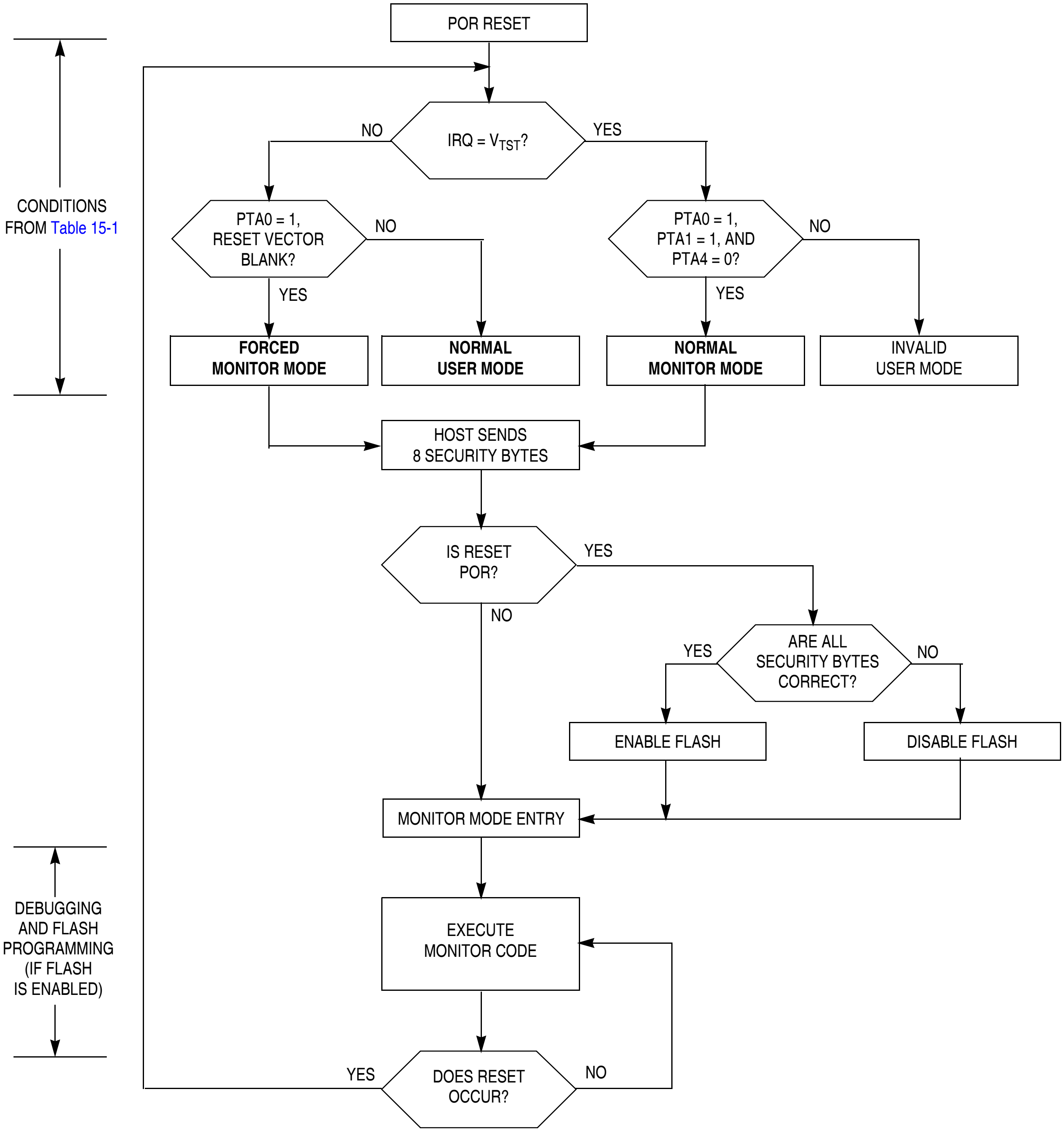


Figure 15-9. Simplified Monitor Mode Entry Flowchart

16.12 3-V Oscillator Characteristics

Characteristic	Symbol	Min	Typ	Max	Unit
Internal oscillator frequency[1]	f_{INTCLK}	—	12.8	—	MHz
Deviation from trimmed Internal oscillator [2][3] 12.8 MHz, fixed voltage, fixed temp 12.8 MHz, V_{DD} ± 10%, 0 to 70°C 12.8 MHz, V_{DD} ± 10%, –40 to 125°C	ACC_{INT}	 — — —	 ± 0.4 ± 2 —	 — — ± 5	%
Crystal frequency, XTALCLK[1]	$f_{OSCXCLK}$	1	—	16	MHz
External RC oscillator frequency, RCCLK [1]	f_{RCCLK}	2	—	10	MHz
External clock reference frequency[1] [4]	$f_{OSCXCLK}$	dc	—	16	MHz
Crystal load capacitance[5]	C_L	—	20	—	pF
Crystal fixed capacitance[3]	C_1	—	2 x C_L	—	—
Crystal tuning capacitance[3]	C_2	—	2 x C_L	—	—
Feedback bias resistor	R_B	0.5	1	10	MΩ
RC oscillator external resistor	R_{EXT}	See Figure 16-8			—
Crystal series damping resistor $f_{OSCXCLK}$ = 1 MHz $f_{OSCXCLK}$ = 4 MHz $f_{OSCXCLK}$ = > 8 MHz	R_S	 — — —	 10 5 0	 — — —	kΩ

1. Bus frequency, f_{OP}, is oscillator frequency divided by 4.
2. Deviation values assumes trimming @25•C and midpoint of voltage range.
3. Values are based on characterization results, not tested in production.
4. No more than 10% duty cycle deviation from 50%
5. Consult crystal vendor data sheet

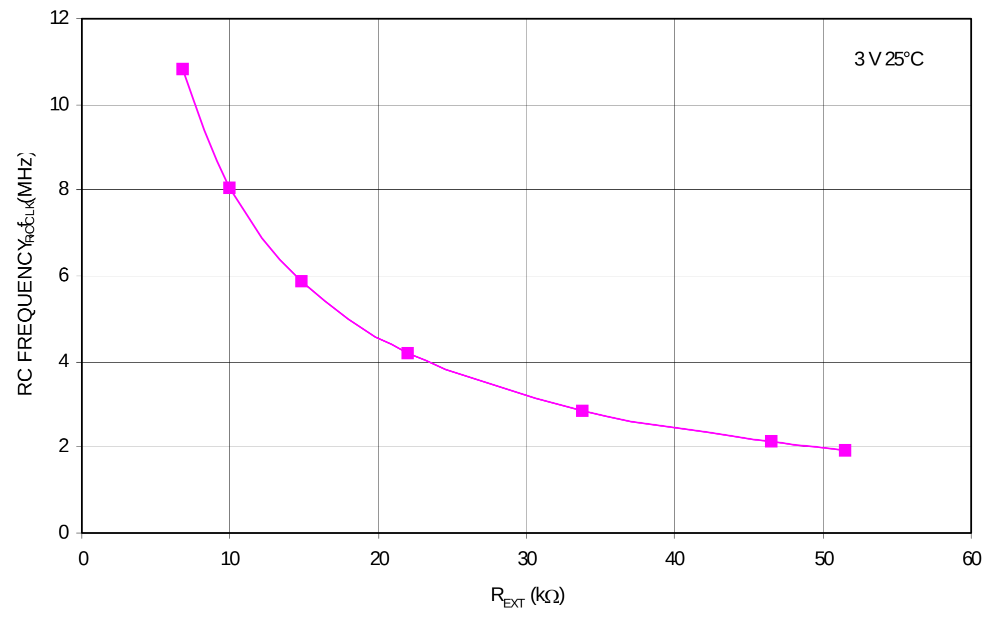


Figure 16-8. RC versus Frequency (3 Volts @ 25•C)

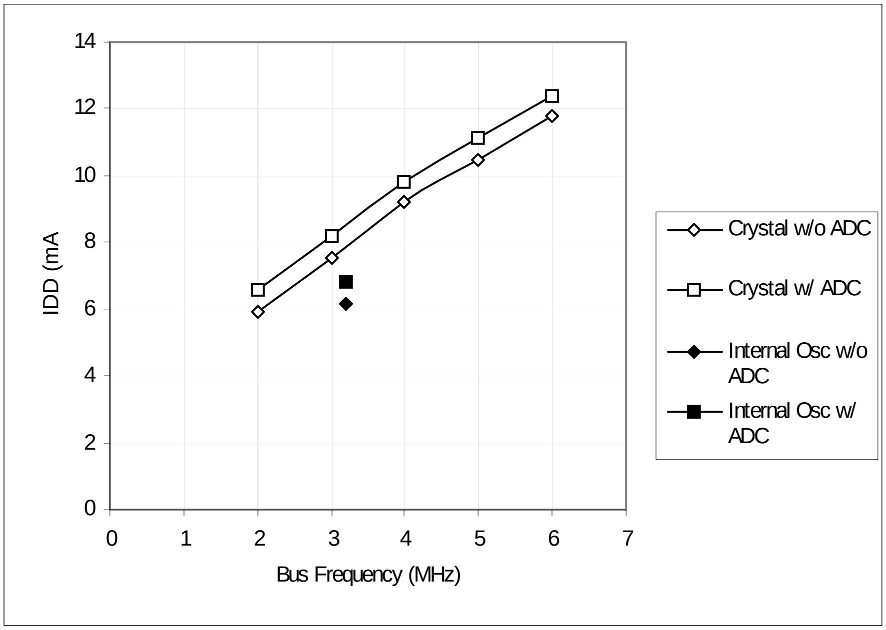


Figure 16-9. Typical 5-Volt Run Current versus Bus Frequency (25•C)

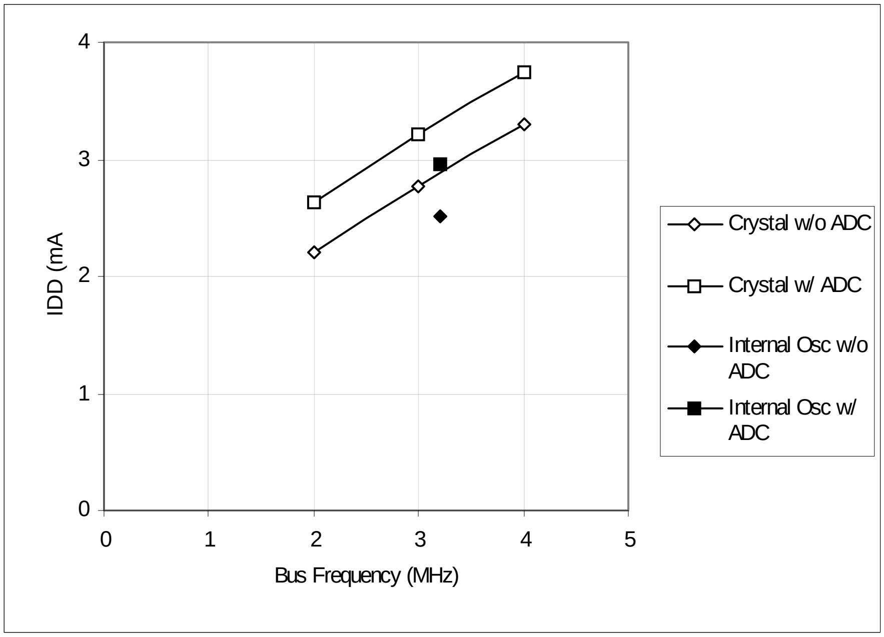
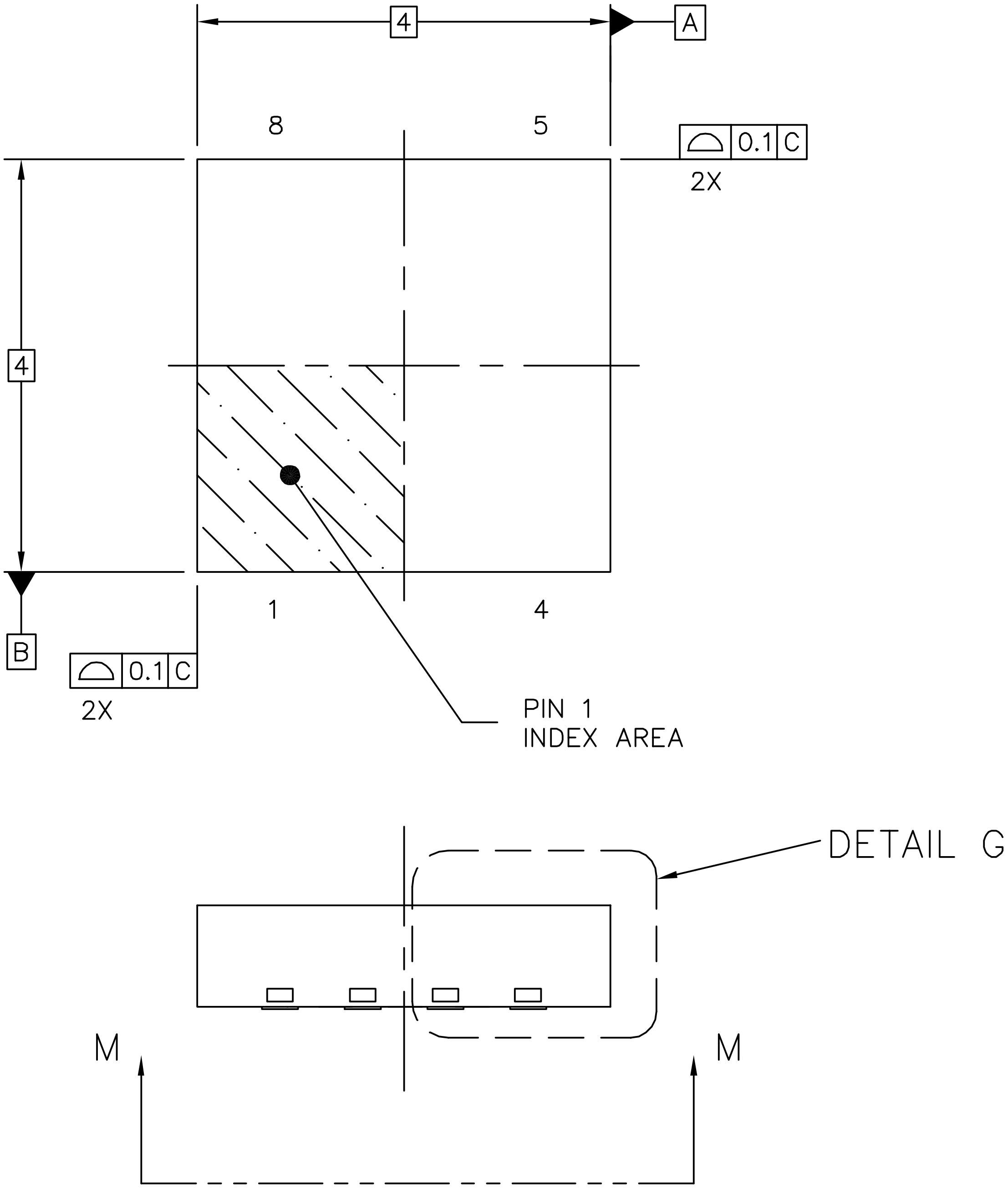


Figure 16-10. Typical 3-Volt Run Current versus Bus Frequency (25•C)

freescale semiconductor	MECHANICAL OUTLINES DICTIONARY	DOCUMENT NO: 98ARL10557D
		PAGE: 1452
	DO NOT SCALE THIS DRAWING	REV: A



TITLE: THERMALLY ENHANCED DUAL
FLAT NO LEAD PACKAGE (DFN)
8 TERMINAL, 0.8 PITCH (4 X 4 X 1)

CASE NUMBER: 1452–01
STANDARD: NON–JEDEC
PACKAGE CODE: 6165
SHEET: 1 OF 5

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MECHANICAL OUTLINES DICTIONARY

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DOCUMENT NO: 98ASB42567B
PAGE: 751G
REV: E

NOTES:

1. DIMENSIONS ARE IN MILLIMETERS.

2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M–1994.

3. DATUMS A AND B TO BE DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.

△4. THIS DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURRS. MOLD FLASH, PROTRUSION OR GATE BURRS SHALL NOT EXCEED 0.15 MM PER SIDE. THIS DIMENSION IS DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.

△5. THIS DIMENSION DOES NOT INCLUDE INTER–LEAD FLASH OR PROTRUSIONS. INTER–LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED 0.25 MM PER SIDE. THIS DIMENSION IS DETERMINED AT THE PLANE WHERE THE BOTTOM OF THE LEADS EXIT THE PLASTIC BODY.

△6. THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.62 mm.

TITLE:
16LD SOIC W/B, 1.27 PITCH,
CASE OUTLINE

CASE NUMBER: 751G–05
STANDARD: JEDEC MS–013AA
PACKAGE CODE: 2003
SHEET: 2 OF 3

NOTES:

1. CONTROLLING DIMENSION: MILLIMETER

2. DIMENSIONS AND TOLERANCES PER ANSI Y14.5M-1982.

△3 DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE.

△4 DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 PER SIDE.

△5 DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 TOTAL IN EXCESS OF THE DIMENSION AT MAXIMUM MATERIAL CONDITION.

6. TERMINAL NUMBERS ARE SHOWN FOR REFERENCE ONLY.

△7 DIMENSIONS ARE TO BE DETERMINED AT DATUM PLANE -W-.