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### **Understanding Embedded - FPGAs (Field Programmable Gate Array)**

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

#### **Details**

Product Status	Discontinued at Digi-Key
Number of LABs/CLBs	32256
Number of Logic Elements/Cells	-
Total RAM Bits	294912
Number of I/O	684
Number of Gates	2000000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	1152-BGA
Supplier Device Package	1152-FPBGA (35x35)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microsemi/ax2000-2fgg1152i">https://www.e-xfl.com/product-detail/microsemi/ax2000-2fgg1152i</a>

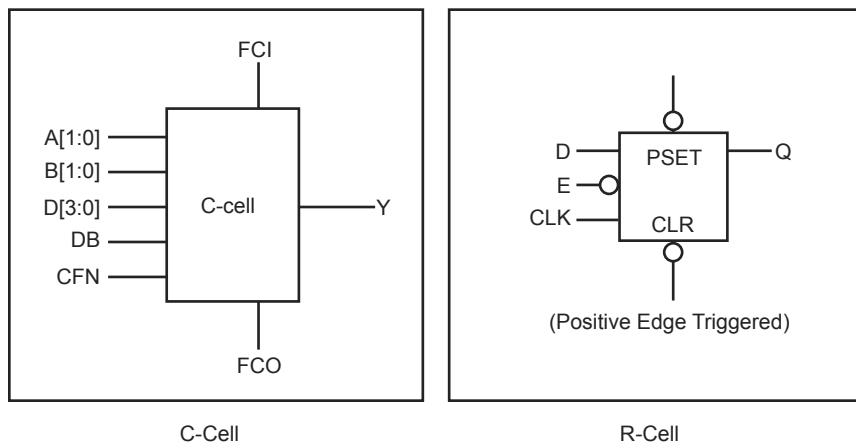
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**Figure 1-2 • Axcelerator Family Interconnect Elements**

## Logic Modules

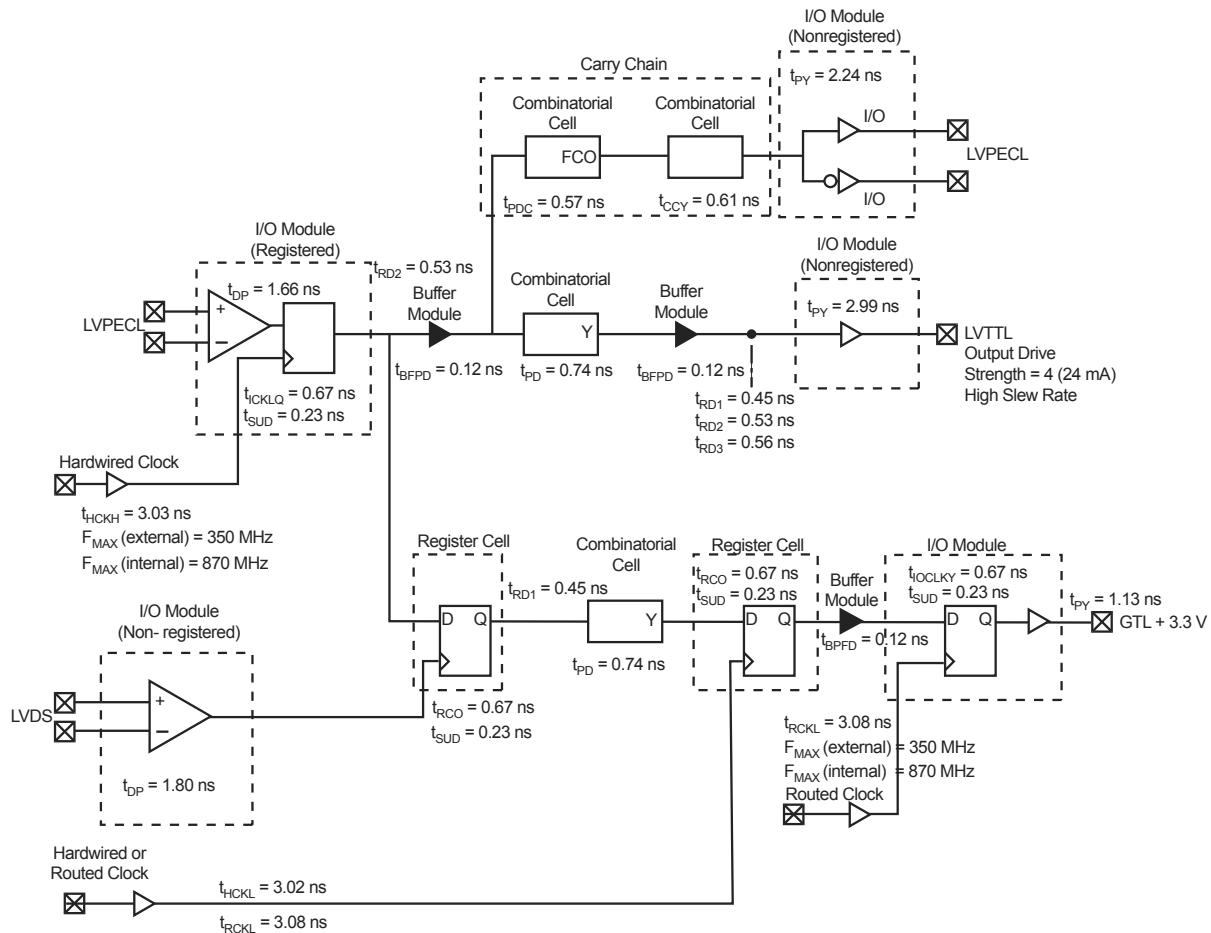
Microsemi's Axcelerator family provides two types of logic modules: the register cell (R-cell) and the combinatorial cell (C-cell). The Axcelerator device can implement more than 4,000 combinatorial functions of up to five inputs (Figure 1-3).

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**Figure 1-3 • AX C-Cell and R-Cell**

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and active-low enable control signals (Figure 1-3). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility (e.g., easy mapping of dual-data-rate functions into the FPGA) while conserving valuable clock resources. The clock source for the R-cell can be chosen from the hardwired clocks, routed clocks, or internal logic.

## Timing Model



Note: Worst case timing data for the AX1000, -2 speed grade

Figure 2-1 • Worst Case Timing Data

### Hardwired Clock – Using LVTTL 24 mA High Slew Clock I/O

#### External Setup

$$\begin{aligned} &= (t_{DP} + t_{RD2} + t_{SUD}) - t_{HCKL} \\ &= (1.72 + 0.53 + 0.23) - 3.02 = -0.54\text{ ns} \end{aligned}$$

#### Clock-to-Out (Pad-to-Pad)

$$\begin{aligned} &= t_{HCKL} + t_{RCO} + t_{RD1} + t_{PY} \\ &= 3.02 + 0.67 + 0.45 + 2.99 = 7.13\text{ ns} \end{aligned}$$

### Routed Clock – Using LVTTL 24 mA High Slew Clock I/O

#### External Setup

$$\begin{aligned} &= (t_{DP} + t_{RD2} + t_{SUD}) - t_{RCKH} \\ &= (1.72 + 0.53 + 0.23) - 3.13 = -0.65\text{ ns} \end{aligned}$$

#### Clock-to-Out (Pad-to-Pad)

$$\begin{aligned} &= t_{RCKH} + t_{RCO} + t_{RD1} + t_{PY} \\ &= 3.13 + 0.67 + 0.45 + 3.03 = 7.24\text{ ns} \end{aligned}$$

# I/O Specifications

## Pin Descriptions

### Supply Pins

**GND**                      **Ground**

Low supply voltage.

**VCCA**                      **Supply Voltage**

Supply voltage for array (1.5V). See "Operating Conditions" on page 2-1 for more information.

**VCCIBx**                      **Supply Voltage**

Supply voltage for I/Os. Bx is the I/O Bank ID – 0 to 7. See "Operating Conditions" on page 2-1 for more information.

**VCCDA**                      **Supply Voltage**

Supply voltage for the I/O differential amplifier and JTAG and probe interfaces. See "Operating Conditions" on page 2-1 for more information. VCCDA should be tied to 3.3V.

**VCCPLA/B/C/D/E/F/G/H**    **Supply Voltage**

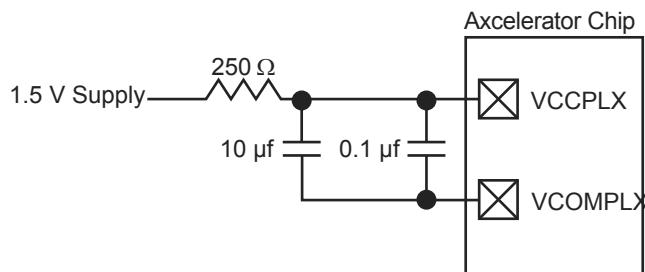
PLL analog power supply (1.5V) for internal PLL. There are eight in each device. VCCPLA supports the PLL associated with global resource HCLKA, VCCPLB supports the PLL associated with global resource HCLKB, etc. The PLL analog power supply pins should be connected to 1.5V whether PLL is used or not.

**VCOMPLA/B/C/D/E/F/G/H**    **Supply Voltage**

Compensation reference signals for internal PLL. There are eight in each device. **VCOMPLA** supports the PLL associated with global resource HCLKA, VCOMPLE supports the PLL associated with global resource CLKE, etc. (see Figure 2-2 on page 2-9 for correct external connection to the supply). The VCOMPLX pins should be left floating if PLL is not used.

**VPUMP**                      **Supply Voltage (External Pump)**

In the low power mode, VPUMP will be used to access an external charge pump (if the user desires to bypass the internal charge pump to further reduce power). The device starts using the external charge pump when the voltage level on VPUMP reaches  $V_{IH}$ <sup>1</sup>. In normal device operation, when using the internal charge pump, VPUMP should be tied to GND.



**Figure 2-2 • VCCPLX and VCOMPLX Power Supply Connect**

1. When  $V_{PUMP} = V_{IH}$ , it shuts off the internal charge pump. See "Low Power Mode" on page 2-106.

**Table 2-8 • I/O Standards Supported by the Axcelerator Family**

I/O Standard	Input/Output Supply Voltage (VCCI)	Input Reference Voltage (VREF)	Board Termination Voltage (VTT)
LVTTL	3.3	N/A	N/A
LVCMOS 2.5 V	2.5	N/A	N/A
LVCMOS 1.8 V	1.8	N/A	N/A
LVCMOS 1.5 V (JDEC8-11)	1.5	N/A	N/A
3.3V PCI/PCI-X	3.3	N/A	N/A
GTL+ 3.3 V	3.3	1.0	1.2
GTL+ 2.5 V*	2.5	1.0	1.2
HSTL Class 1	1.5	0.75	0.75
SSTL3 Class 1 and II	3.3	1.5	1.5
SSTL2 Class1 and II	2.5	1.25	1.25
LVDS	2.5	N/A	N/A
LVPECL	3.3	N/A	N/A

Note: \*2.5 V GTL+ is not supported across the full military temperature range.

**Table 2-9 • Supply Voltages**

VCCA	VCCI	Input Tolerance	Output Drive Level
1.5 V	1.5 V	3.3 V	1.5 V
1.5 V	1.8 V	3.3 V	1.8 V
1.5 V	2.5 V	3.3 V	2.5 V
1.5 V	3.3 V	3.3 V	3.3 V

**Table 2-10 • I/O Features Comparison**

I/O Assignment	Clamp Diode	Hot Insertion	5 V Tolerance	Input Buffer	Output Buffer
LVTTL	No	Yes	Yes <sup>1</sup>	Enabled/Disabled	
3.3 V PCI, 3.3 V PCI-X	Yes	No	Yes <sup>1, 2</sup>	Enabled/Disabled	
LVCMOS 2.5 V	No	Yes	No	Enabled/Disabled	
LVCMOS 1.8 V	No	Yes	No	Enabled/Disabled	
LVCMOS 1.5 V (JESD8-11)	No	Yes	No	Enabled/Disabled	
Voltage-Referenced Input Buffer	No	Yes	No	Enabled/Disabled	
Differential, LVDS/LVPECL, Input	No	Yes	No	Enabled	Disabled <sup>3</sup>
Differential, LVDS/LVPECL, Output	No	Yes	No	Disabled	Enabled <sup>4</sup>

Notes:

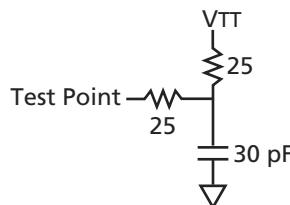
1. Can be implemented with an IDT bus switch.
2. Can be implemented with an external resistor.
3. The OE input of the output buffer must be deasserted permanently (handled by software).
4. The OE input of the output buffer must be asserted permanently (handled by software).

## Class II

**Table 2-53 • DC Input and Output Levels**

VIL		VIH		VOL	VOH	IOL	IOH
Min., V	Max., V	Min., V	Max., V	Max., V	Min., V	mA	mA
-0.3	VREF - 0.2	VREF + 0.2	3.6	VREF - 0.8	VREF + 0.8	16	-16

## AC Loadings



**Figure 2-24 • AC Test Loads**

**Table 2-54 • AC Waveforms, Measuring Points, and Capacitive Loads**

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ) (V)	C <sub>load</sub> (pF)
VREF - 1.0	VREF + 1.0	VREF	1.50	30

Note: \* Measuring Point = VTRIP

## Timing Characteristics

**Table 2-55 • 3.3 V SSTL3 Class II I/O Module**

Worst-Case Commercial Conditions VCCA = 1.425 V, VCCI = 3.0V, T<sub>J</sub> = 70°C

Parameter	Description	-2 Speed		-1 Speed		Std Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	
<b>3.3 V SSTL3 Class II I/O Module Timing</b>								
t <sub>DP</sub>	Input Buffer			1.85	2.10	2.47		ns
t <sub>PY</sub>	Output Buffer			2.17	2.47	2.91		ns
t <sub>ICLKQ</sub>	Clock-to-Q for the I/O input register			0.67	0.77	0.90		ns
t <sub>OCLKQ</sub>	Clock-to-Q for the I/O output register and the I/O enable register			0.67	0.77	0.90		ns
t <sub>SUD</sub>	Data Input Set-Up			0.23	0.27	0.31		ns
t <sub>SUE</sub>	Enable Input Set-Up			0.26	0.30	0.35		ns
t <sub>HD</sub>	Data Input Hold			0.00	0.00	0.00		ns
t <sub>HE</sub>	Enable Input Hold			0.00	0.00	0.00		ns
t <sub>CPWHL</sub>	Clock Pulse Width High to Low	0.39		0.39		0.39		ns
t <sub>CPWLH</sub>	Clock Pulse Width Low to High	0.39		0.39		0.39		ns
t <sub>WASYN</sub>	Asynchronous Pulse Width	0.37		0.37		0.37		ns
t <sub>REASYN</sub>	Asynchronous Recovery Time			0.13	0.15	0.17		ns
t <sub>HASYN</sub>	Asynchronous Removal Time			0.00	0.00	0.00		ns
t <sub>CLR</sub>	Asynchronous Clear-to-Q			0.23	0.27	0.31		ns
t <sub>PRESET</sub>	Asynchronous Preset-to-Q			0.23	0.27	0.31		ns

# Routing Specifications

## Routing Resources

The routing structure found in Axcelerator devices enables any logic module to be connected to any other logic module while retaining high performance. There are multiple paths and routing resources that can be used to route one logic module to another, both within a SuperCluster and elsewhere on the chip.

There are four primary types of routing within the AX architecture: DirectConnect, CarryConnect, FastConnect, and Vertical and Horizontal Routing.

### **DirectConnect**

DirectConnects provide a high-speed connection between an R-cell and its adjacent C-cell (Figure 2-35). This connection can be made from DCOUT of the C-cell to DCIN of the R-cell by configuring of the S1 line of the R-cell. This provides a connection that does not require an antifuse and has a delay of less than 0.1 ns.

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**Figure 2-35 • DirectConnect and CarryConnect**

### **CarryConnect**

CarryConnects are used to build carry chains for arithmetic functions (Figure 2-35). The FCO output of the right C-cell of a two-C-cell Cluster drives the FCI input of the left C-cell in the two-C-cell Cluster immediately below it. This pattern continues down both sides of each SuperCluster column.

Similar to the DirectConnects, CarryConnects can be built without an antifuse connection. This connection has a delay of less than 0.1 ns from the FCO of one two-C-cell cluster to the FCI of the two-C-cell cluster immediately below it (see the "Carry-Chain Logic" section on page 2-56 for more information).

### **FastConnect**

For high-speed routing of logic signals, FastConnects can be used to build a short distance connection using a single antifuse (Figure 2-36 on page 2-62). FastConnects provide a maximum delay of 0.3 ns. The outputs of each logic module connect directly to the Output Tracks within a SuperCluster. Signals on the Output Tracks can then be routed through a single antifuse connection to drive the inputs of logic modules either within one SuperCluster or in the SuperCluster immediately below it.

**Table 2-67 • AX500 Predicted Routing Delays**

Worst-Case Commercial Conditions VCCA = 1.425 V, TJ = 70°C

Parameter	Description	-2 Speed	-1 Speed	Std Speed	
<b>Predicted Routing Delays</b>					
t <sub>DC</sub>	DirectConnect Routing Delay, FO1	0.11	0.12	0.15	ns
t <sub>FC</sub>	FastConnect Routing Delay, FO1	0.35	0.39	0.46	ns
t <sub>RD1</sub>	Routing delay for FO1	0.39	0.45	0.53	ns
t <sub>RD2</sub>	Routing delay for FO2	0.41	0.46	0.54	ns
t <sub>RD3</sub>	Routing delay for FO3	0.48	0.55	0.64	ns
t <sub>RD4</sub>	Routing delay for FO4	0.56	0.63	0.75	ns
t <sub>RD5</sub>	Routing delay for FO5	0.60	0.68	0.80	ns
t <sub>RD6</sub>	Routing delay for FO6	0.84	0.96	1.13	ns
t <sub>RD7</sub>	Routing delay for FO7	0.90	1.02	1.20	ns
t <sub>RD8</sub>	Routing delay for FO8	1.00	1.13	1.33	ns
t <sub>RD16</sub>	Routing delay for FO16	2.17	2.46	2.89	ns
t <sub>RD32</sub>	Routing delay for FO32	3.55	4.03	4.74	ns

**Table 2-68 • AX1000 Predicted Routing Delays**

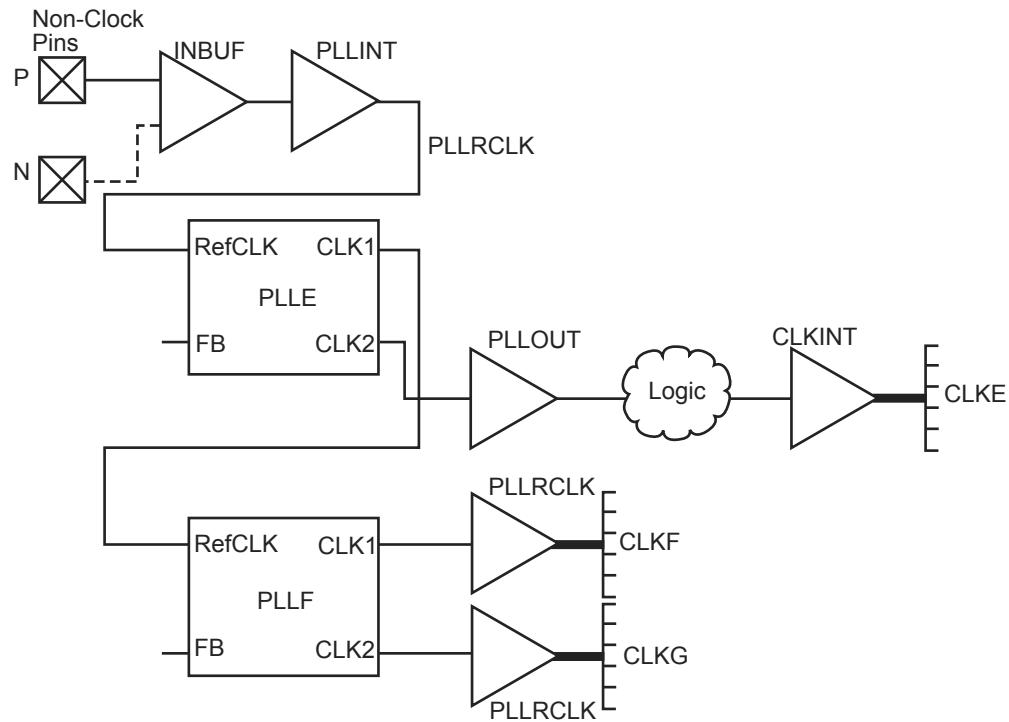
Worst-Case Commercial Conditions VCCA = 1.425 V, TJ = 70°C

Parameter	Description	-2 Speed	-1 Speed	Std Speed	
<b>Predicted Routing Delays</b>					
t <sub>DC</sub>	DirectConnect Routing Delay, FO1	0.12	0.13	0.15	ns
t <sub>FC</sub>	FastConnect Routing Delay, FO1	0.35	0.39	0.46	ns
t <sub>RD1</sub>	Routing delay for FO1	0.45	0.51	0.60	ns
t <sub>RD2</sub>	Routing delay for FO2	0.53	0.60	0.71	ns
t <sub>RD3</sub>	Routing delay for FO3	0.56	0.63	0.74	ns
t <sub>RD4</sub>	Routing delay for FO4	0.63	0.71	0.84	ns
t <sub>RD5</sub>	Routing delay for FO5	0.73	0.82	0.97	ns
t <sub>RD6</sub>	Routing delay for FO6	0.99	1.13	1.32	ns
t <sub>RD7</sub>	Routing delay for FO7	1.02	1.15	1.36	ns
t <sub>RD8</sub>	Routing delay for FO8	1.48	1.68	1.97	ns
t <sub>RD16</sub>	Routing delay for FO16	2.57	2.91	3.42	ns
t <sub>RD32</sub>	Routing delay for FO32	4.24	4.81	5.65	ns



### **Implementation Example:**

Figure 2-47 shows a complex clock distribution example. The reference clock (RefCLK) of PLLE is being sourced from non-clock signal pins (INBUF to PLLINT). The CLK1 output of PLLE is being fed to the RefCLK input of PLLF. The CLK2 output of PLLE is driving logic (via PLLOUT). In turn, this logic is driving the global resource CLKE. PLLF is driving both CLKF and CLKG global resources.



**Figure 2-47 • Complex Clock Distribution Example**

single-ended, or voltage-referenced standard. The [H]CLKxN pad can only be used as a differential pair with [H]CLKxP.

The block marked “/i Delay Match” is a fixed delay equal to that of the i divider. The “/j Delay Match” block has the same function as its j divider counterpart.

## Functional Description

Figure 2-48 on page 2-75 illustrates a block diagram of the PLL. The PLL contains two dividers, i and j, that allow frequency scaling of the clock signal:

- The i divider in the feedback path allows multiplication of the input clock by integer factors ranging from 1 to 64, and the resultant frequency is available at the output of the PLL block.
- The j divider divides the PLL output by integer factors ranging from 1 to 64, and the divided clock is available at CLK1.
- The two dividers together can implement any combination of multiplication and division up to a maximum frequency of 1 GHz on CLK1. Both the CLK1 and CLK2 outputs have a fixed 50/50 duty cycle.
- The output frequencies of the two clocks are given by the following formulas ( $f_{REF}$  is the reference clock frequency):
 
$$f_{CLK1} = f_{REF} * (\text{DividerI}) / (\text{DividerJ}) \quad \text{EQ 4}$$

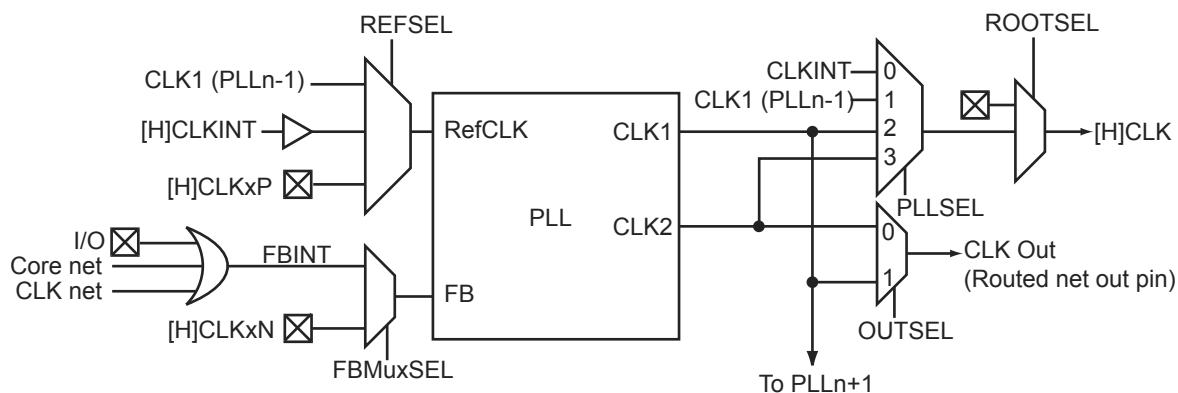
$$f_{CLK2} = f_{REF} * (\text{DividerI}) \quad \text{EQ 5}$$

- CLK2 provides the PLL output directly—without division

The input and output frequency ranges are selected by LowFreq and Osc(2:0), respectively. These functions and their possible values are detailed in Table 2-80 on page 2-77.

The delay lines shown in Figure 2-48 on page 2-75 are programmable. The feedback clock path can be delayed (using the five DelayLine bits) relative to the reference clock (or vice versa) by up to 3.75 ns in increments of 250 ps. Table 2-80 on page 2-77 describes the usage of these bits. The delay increments are independent of frequency, so this results in phase changes that vary with frequency. The delay value is highly dependent on  $V_{CC}$  and the speed grade.

Figure 2-49 is a logical diagram of the various control signals to the PLL and shows how the PLL interfaces with the global and routing networks of the FPGA. Note that not all signals are user-accessible. These non-user-accessible signals are used by the place-and-route tool to control the configuration of the PLL. The user gains access to these control signals either based upon the connections built in the user's design or through the special macros (Table 2-84 on page 2-81) inserted into the design. For example, connecting the macro PLLOUT to CLK2 will control the OUTSEL signal.



*Note: Not all signals are available to the user.*

Figure 2-49 • PLL Logical Interface

Note that the RAM blocks employ little-endian byte order for read and write operations.

**Table 2-88 • RAM Signal Description**

Signal	Direction	Description
WCLK	Input	Write clock (can be active on either edge).
WA[J:0]	Input	Write address bus. The value J is dependent on the RAM configuration and the number of cascaded memory blocks. The valid range for J is from 6 to 15.
WD[M-1:0]	Input	Write data bus. The value M is dependent on the RAM configuration and can be 1, 2, 4, 9, 18, or 36.
RCLK	Input	Read clock (can be active on either edge).
RA[K:0]	Input	Read address bus. The value K is dependent on the RAM configuration and the number of cascaded memory blocks. The valid range for K is from 6 to 15.
RD[N-1:0]	Output	Read data bus. The value N is dependent on the RAM configuration and can be 1, 2, 4, 9, 18, or 36.
REN	Input	Read enable. When this signal is valid on the active edge of the clock, data at location RA will be driven onto RD.
WEN	Input	Write enable. When this signal is valid on the active edge of the clock, WD data will be written at location WA.
RW[2:0]	Input	Width of the read operation dataword.
WW[2:0]	Input	Width of the write operation dataword.
Pipe	Input	Sets the pipe option to be on or off.

## Modes of Operation

There are two read modes and one write mode:

- Read Nonpipelined (synchronous – one clock edge)
- Read Pipelined (synchronous – two clock edges)
- Write (synchronous – one clock edge)

In the standard read mode, new data is driven onto the RD bus in the clock cycle immediately following RA and REN valid. The read address is registered on the read-port active-clock edge and data appears at read-data after the RAM access time. Setting the PIPE to OFF enables this mode.

The pipelined mode incurs an additional clock delay from address to data, but enables operation at a much higher frequency. The read-address is registered on the read-port active-clock edge, and the read data is registered and appears at RD after the second read clock edge. Setting the PIPE to ON enables this mode.

On the write active-clock edge, the write data are written into the SRAM at the write address when WEN is high. The setup time of the write address, write enables, and write data are minimal with respect to the write clock.

Write and read transfers are described with timing requirements beginning in the "Timing Characteristics" section on page 2-89.

## FIFO

Every memory block has its own embedded FIFO controller. Each FIFO block has one read port and one write port. This embedded FIFO controller uses no internal FPGA logic and features:

- Glitch-free FIFO Flags
- Gray-code address counters/pointers to prevent metastability problems
- Overflow and underflow control

Both ports are configurable in various sizes from 4k x 1 to 128 x 36, similar to the RAM block size. Each port is fully synchronous.

Read and write operations can be completely independent. Data on the appropriate WD pins are written to the FIFO on every active WCLK edge as long as WEN is high. Data is read from the FIFO and output on the appropriate RD pins on every active RCLK edge as long as REN is asserted.

The FIFO block offers programmable almost-empty (AEMPTY) and almost-full (AFULL) flags as well as EMPTY and FULL flags (Figure 2-61):

- The FULL flag is synchronous to WCLK. It allows the FIFO to inhibit writing when full.
- The EMPTY flag is synchronous to RCLK. It allows the FIFO to inhibit reading at the empty condition.

Gray code counters are used to prevent metastability problems associated with flag logic. The depth of the FIFO is dependent on the data width and the number of memory blocks used to create the FIFO. The write operations to the FIFO are synchronous with respect to the WCLK, and the read operations are synchronous with respect to the RCLK.

The FIFO block may be reset to the empty state.

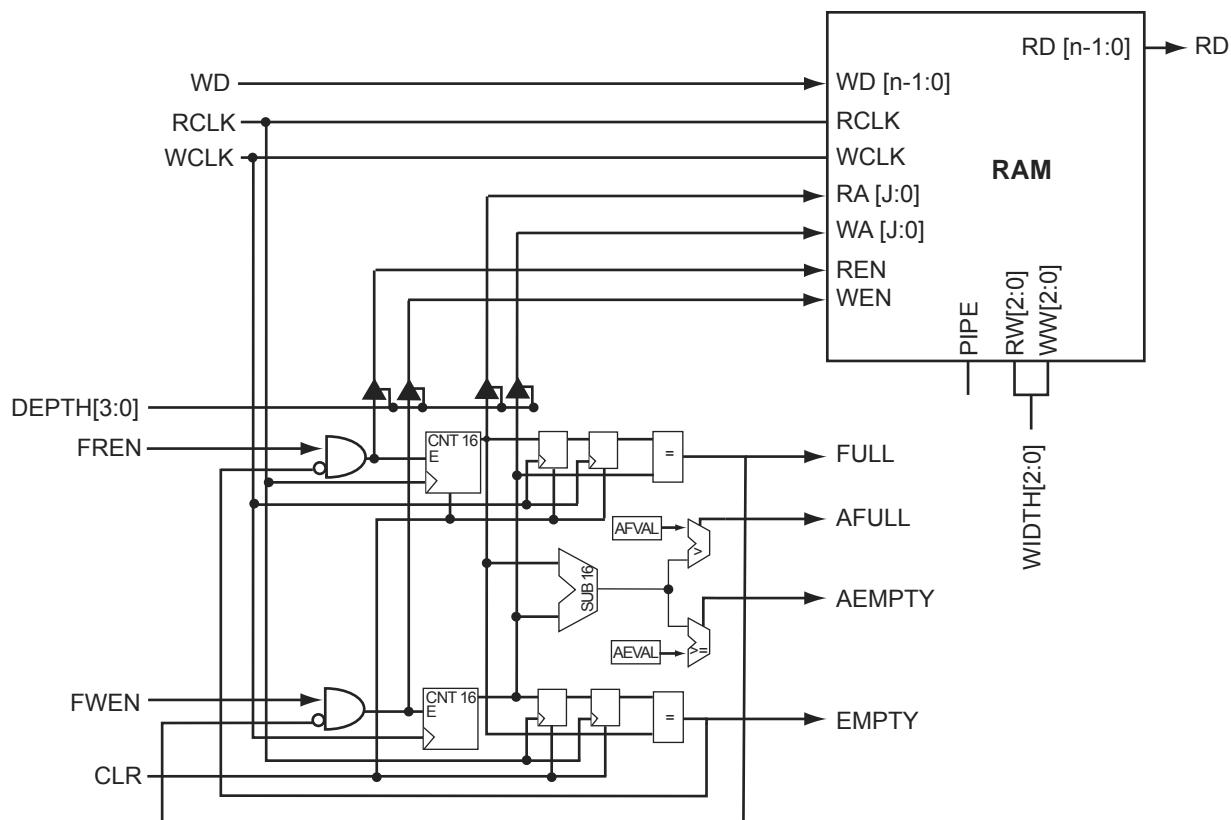


Figure 2-61 • Axcelerator RAM with Embedded FIFO Controller





<b>FG676</b>	
<b>AX500 Function</b>	<b>Pin Number</b>
GND	R10
GND	R11
GND	R12
GND	R13
GND	R14
GND	R15
GND	R16
GND	R17
GND	T10
GND	T11
GND	T12
GND	T13
GND	T14
GND	T15
GND	T16
GND	T17
GND	U10
GND	U11
GND	U12
GND	U13
GND	U14
GND	U15
GND	U16
GND	U17
GND	V18
GND	V9
GND	W1
GND	W19
GND	W26
GND	W8
GND	Y20
GND	Y7
GND/LP	C2
NC	A11
NC	A21

<b>FG676</b>	
<b>AX500 Function</b>	<b>Pin Number</b>
NC	A22
NC	A24
NC	A25
NC	AA11
NC	AA19
NC	AA20
NC	AA4
NC	AA5
NC	AA6
NC	AA7
NC	AA8
NC	AA9
NC	AB1
NC	AB11
NC	AB17
NC	AB18
NC	AB19
NC	AB20
NC	AB8
NC	AB9
NC	AC1
NC	AC13
NC	AC14
NC	AC25
NC	AD1
NC	AD11
NC	AD16
NC	AD25
NC	AE1
NC	AF2
NC	AF25
NC	B11
NC	B24
NC	B4
NC	C16

<b>FG676</b>	
<b>AX500 Function</b>	<b>Pin Number</b>
NC	C4
NC	D1
NC	D13
NC	D14
NC	D17
NC	D18
NC	D2
NC	D26
NC	D3
NC	D9
NC	E1
NC	E18
NC	E23
NC	E24
NC	E26
NC	E3
NC	E4
NC	E9
NC	F1
NC	F18
NC	F20
NC	F21
NC	F22
NC	F23
NC	F24
NC	F4
NC	F6
NC	F7
NC	G21
NC	G22
NC	H21
NC	H22
NC	H23
NC	H5
NC	H6

<b>FG676</b>	
<b>AX500 Function</b>	<b>Pin Number</b>
VCCIB3	T19
VCCIB3	U19
VCCIB3	U20
VCCIB3	V19
VCCIB3	V20
VCCIB3	W20
VCCIB4	W14
VCCIB4	W15
VCCIB4	W16
VCCIB4	W17
VCCIB4	W18
VCCIB4	Y17
VCCIB4	Y18
VCCIB4	Y19
VCCIB5	W10
VCCIB5	W11
VCCIB5	W12
VCCIB5	W13
VCCIB5	W9
VCCIB5	Y10
VCCIB5	Y8
VCCIB5	Y9
VCCIB6	P8
VCCIB6	R8
VCCIB6	T8
VCCIB6	U7
VCCIB6	U8
VCCIB6	V7
VCCIB6	V8
VCCIB6	W7
VCCIB7	H7
VCCIB7	J7
VCCIB7	J8
VCCIB7	K7
VCCIB7	K8

<b>FG676</b>	
<b>AX500 Function</b>	<b>Pin Number</b>
VCCIB7	L8
VCCIB7	M8
VCCIB7	N8
VCCPLA	E12
VCCPLB	F13
VCCPLC	E15
VCCPLD	G14
VCCPLE	AF15
VCCPLF	AA14
VCCPLG	AF12
VCCPLH	AB13
VCOMPLA	D12
VCOMPLB	G13
VCOMPLC	D15
VCOMPLD	F14
VCOMPLE	AD15
VCOMPLF	AB14
VCOMPLG	AD12
VCOMPLH	Y13
VPUMP	E22



PQ208	
AX500 Function	Pin Number
IO150PB7F14	19
IO152NB7F14	16
IO152PB7F14	17
IO161NB7F15	12
IO161PB7F15	13
IO163NB7F15	10
IO163PB7F15	11
IO165PB7F15	7
IO166NB7F15	5
IO166PB7F15	6
IO167NB7F15	3
IO167PB7F15	4
<b>Dedicated I/O</b>	
V <sub>CCDA</sub>	1
V <sub>CCDA</sub>	26
V <sub>CCDA</sub>	53
V <sub>CCDA</sub>	63
V <sub>CCDA</sub>	78
V <sub>CCDA</sub>	95
V <sub>CCDA</sub>	105
V <sub>CCDA</sub>	130
V <sub>CCDA</sub>	157
V <sub>CCDA</sub>	167
V <sub>CCDA</sub>	182
V <sub>CCDA</sub>	202
GND	104
GND	9
GND	15
GND	21
GND	32
GND	39
GND	46
GND	51
GND	59
GND	65
GND	69
GND	90

PQ208	
AX500 Function	Pin Number
GND	94
GND	99
GND	113
GND	119
GND	125
GND	143
GND	136
GND	150
GND	155
GND	164
GND	169
GND	173
GND	194
GND	196
GND	201
GND/LP	208
PRA	184
PRB	183
PRC	80
PRD	79
TCK	205
TDI	204
TDO	203
TMS	206
TRST	207
VCCA	2
VCCA	14
VCCA	38
VCCA	52
VCCA	64
VCCA	93
VCCA	118
VCCA	142
VCCA	156
VCCA	168
VCCA	195
VCCPLA	189

PQ208	
AX500 Function	Pin Number
VCCPLB	187
VCCPLC	178
VCCPLD	176
VCCPLE	85
VCCPLF	83
VCCPLG	74
VCCPLH	72
VCCIB0	200
VCCIB0	193
VCCIB1	172
VCCIB1	163
VCCIB2	149
VCCIB2	135
VCCIB3	124
VCCIB3	112
VCCIB4	98
VCCIB4	89
VCCIB5	68
VCCIB5	58
VCCIB6	45
VCCIB6	31
VCCIB7	20
VCCIB7	8
VCOMPLA	190
VCOMPLB	188
VCOMPLC	179
VCOMPLD	177
VCOMPLE	86
VCOMPLF	84
VCOMPLG	75
VCOMPLH	73
VPUMP	158

<b>CG624</b>	
<b>AX1000 Function</b>	<b>Pin Number</b>
VCCA	U17
VCCA	U9
VCCA	Y4
VCCDA	A12
VCCDA	AA13
VCCDA	AA15
VCCDA	AA7
VCCDA	AC11
VCCDA	AD11
VCCDA	AE17
VCCDA	B15
VCCDA	C15
VCCDA	C6
VCCDA	D13
VCCDA	E13
VCCDA	E19
VCCDA	G5
VCCDA	N21
VCCDA	N5
VCCDA	W21
VCCIB0	A3
VCCIB0	B3
VCCIB0	C4
VCCIB0	D5
VCCIB0	J10
VCCIB0	J11
VCCIB0	K12
VCCIB1	A23
VCCIB1	B23
VCCIB1	C22
VCCIB1	D21
VCCIB1	J15
VCCIB1	J16
VCCIB1	K14
VCCIB2	C24
VCCIB2	C25

<b>CG624</b>	
<b>AX1000 Function</b>	<b>Pin Number</b>
VCCIB2	D23
VCCIB2	E22
VCCIB2	K17
VCCIB2	L17
VCCIB2	M16
VCCIB3	AA22
VCCIB3	AB23
VCCIB3	AC24
VCCIB3	AC25
VCCIB3	P16
VCCIB3	R17
VCCIB3	T17
VCCIB4	AB21
VCCIB4	AC22
VCCIB4	AD23
VCCIB4	AE23
VCCIB4	T14
VCCIB4	U15
VCCIB4	U16
VCCIB5	AB5
VCCIB5	AC4
VCCIB5	AD3
VCCIB5	AE3
VCCIB5	T12
VCCIB5	U10
VCCIB5	U11
VCCIB6	AA4
VCCIB6	AB3
VCCIB6	AC1
VCCIB6	AC2
VCCIB6	P10
VCCIB6	R9
VCCIB6	T9
VCCIB7	C1
VCCIB7	C2
VCCIB7	D3

<b>CG624</b>	
<b>AX1000 Function</b>	<b>Pin Number</b>
VCCIB7	E4
VCCIB7	K9
VCCIB7	L9
VCCIB7	M10
VCCPLA	E12
VCCPLB	J12
VCCPLC	E14
VCCPLD	H14
VCCPLE	Y14
VCCPLF	U14
VCCPLG	Y12
VCCPLH	U12
VCOMPLA	F12
VCOMPLB	H12
VCOMPLC	F14
VCOMPLD	J14
VCOMPLE	AA14
VCOMPLF	V14
VCOMPLG	AA12
VCOMPLH	V12
VPUMP	E20

