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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Communications; QUICC Engine
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	668-BBGA Exposed Pad
Supplier Device Package	668-PBGA-PGE (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8358cvragdga

- Data bus widths:
  - Single 32-bit data PCI interface that operates at up to 66 MHz
- PCI 3.3-V compatible (not 5-V compatible)
- PCI host bridge capabilities on both interfaces
- PCI agent mode supported on PCI interface
- Support for PCI-to-memory and memory-to-PCI streaming
- Memory prefetching of PCI read accesses and support for delayed read transactions
- Support for posting of processor-to-PCI and PCI-to-memory writes
- On-chip arbitration, supporting five masters on PCI
- Support for accesses to all PCI address spaces
- Parity support
- Selectable hardware-enforced coherency
- Address translation units for address mapping between host and peripheral
- Dual address cycle supported when the device is the target
- Internal configuration registers accessible from PCI

- Local bus controller (LBC)
  - Multiplexed 32-bit address and data operating at up to 133 MHz
  - Eight chip selects support eight external slaves
  - Up to eight-beat burst transfers
  - 32-, 16-, and 8-bit port sizes are controlled by an on-chip memory controller
  - Three protocol engines available on a per chip select basis:
    - General-purpose chip select machine (GPCM)
    - Three user programmable machines (UPMs)
    - Dedicated single data rate SDRAM controller
  - Parity support
  - Default boot ROM chip select with configurable bus width (8-, 16-, or 32-bit)
- Programmable interrupt controller (PIC)
  - Functional and programming compatibility with the MPC8260 interrupt controller
  - Support for 8 external and 35 internal discrete interrupt sources
  - Support for one external (optional) and seven internal machine checkstop interrupt sources
  - Programmable highest priority request
  - Four groups of interrupts with programmable priority
  - External and internal interrupts directed to communication processor
  - Redirects interrupts to external $\overline{\text{INTA}}$ pin when in core disable mode
  - Unique vector number for each interrupt source
- Dual industry-standard I^2C interfaces
  - Two-wire interface

2.1.3 Output Driver Characteristics

Table 3 provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

Table 3. Output Drive Capability

Driver Type	Output Impedance (Ω)	Supply Voltage
Local bus interface utilities signals	42	OV_{DD} = 3.3 V
PCI signals	25	
PCI output clocks (including PCI_SYNC_OUT)	42	
DDR signal	20 36 (half-strength mode)[1]	GV_{DD} = 2.5 V
DDR2 signal	18 36 (half-strength mode)[1]	GV_{DD} = 1.8 V
10/100/1000 Ethernet signals	42	LV_{DD} = 2.5/3.3 V
DUART, system control, I^2C, SPI, JTAG	42	OV_{DD} = 3.3 V
GPIO signals	42	OV_{DD} = 3.3 V LV_{DD} = 2.5/3.3 V

[1] DDR output impedance values for half strength mode are verified by design and not tested.

2.2 Power Sequencing

This section details the power sequencing considerations for the MPC8358E.

2.2.1 Power-Up Sequencing

MPC8358E does not require the core supply voltage (V_{DD} and AV_{DD}) and I/O supply voltages (GV_{DD}, LV_{DD}, and OV_{DD}) to be applied in any particular order. During the power ramp up, before the power supplies are stable and if the I/O voltages are supplied before the core voltage, there may be a period of time that all input and output pins will actively be driven and cause contention and excessive current. In order to avoid actively driving the I/O pins and to eliminate excessive current draw, apply the core voltage (V_{DD}) before the I/O voltage (GV_{DD}, LV_{DD}, and OV_{DD}) and assert $\overline{\text{PORESET}}$ before the power supplies fully ramp up. In the case where the core voltage is applied first, the core voltage supply must rise to 90% of its nominal value before the I/O supplies reach 0.7 V, see Figure 4.

Table 9. RESET Pins DC Electrical Characteristics (continued)

Characteristic	Symbol	Condition	Min	Max	Unit
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Notes:
1. This table applies for pins $\overline{\text{PORESET}}$, $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, and $\overline{\text{QUIESCE}}$.
2. $\overline{\text{HRESET}}$ and $\overline{\text{SRESET}}$ are open drain pins, thus V_{OH} is not relevant for those pins.

5.2 RESET AC Electrical Characteristics

This section describes the AC electrical specifications for the reset initialization timing requirements of the device. Table 10 provides the reset initialization AC timing specifications for the DDR SDRAM component(s).

Table 10. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{\text{HRESET}}$ or $\overline{\text{SRESET}}$ (input) to activate reset flow	32	—	$t_{PCI\_SYNC\_IN}$	1
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to CLKIN when the device is in PCI host mode	32	—	t_{CLKIN}	2
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to PCI_SYNC_IN when the device is in PCI agent mode	32	—	$t_{PCI\_SYNC\_IN}$	1
$\overline{\text{HRESET}}$/$\overline{\text{SRESET}}$ assertion (output)	512	—	$t_{PCI\_SYNC\_IN}$	1
$\overline{\text{HRESET}}$ negation to $\overline{\text{SRESET}}$ negation (output)	16	—	$t_{PCI\_SYNC\_IN}$	1
Input setup time for POR config signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI host mode	4	—	t_{CLKIN}	2
Input setup time for POR config signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI agent mode	4	—	$t_{PCI\_SYNC\_IN}$	1
Input hold time for POR config signals with respect to negation of $\overline{\text{HRESET}}$	0	—	ns	
Time for the device to turn off POR config signals with respect to the assertion of $\overline{\text{HRESET}}$	—	4	ns	3
Time for the device to turn on POR config signals with respect to the negation of $\overline{\text{HRESET}}$	1	—	$t_{PCI\_SYNC\_IN}$	1, 3

Notes:
1. $t_{PCI\_SYNC\_IN}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the device is In PCI host mode the primary clock is applied to the CLKIN input, and PCI_SYNC_IN period depends on the value of CFG_CLKIN_DIV. See the *MPC8360E PowerQUICC II Pro Integrated Communications Processor Family Reference Manual* for more details.
2. t_{CLKIN} is the clock period of the input clock applied to CLKIN. It is only valid when the device is in PCI host mode. See the *MPC8360E PowerQUICC II Pro Integrated Communications Processor Family Reference Manual* for more details.
3. POR config signals consists of CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV.

Table 20. DDR and DDR2 SDRAM Output AC Timing Specifications for Source Synchronous Mode (continued)

At recommended operating conditions with GV_{DD} of (1.8 V or 2.5 V) ± 5%.

Parameter[8]	Symbol[1]	Min	Max	Unit	Notes
ADDR/CMD output setup with respect to MCK 266 MHz 200 MHz	t_{DDKHAS}	 2.8 3.5	—	ns	4
ADDR/CMD output hold with respect to MCK 266 MHz—DDR1 266 MHz—DDR2 200 MHz	t_{DDKHAX}	 2.6 2.8 3.5	—	ns	4
$\overline{MCS}$(n) output setup with respect to MCK 266 MHz 200 MHz	t_{DDKHCS}	 2.8 3.5	—	ns	4
$\overline{MCS}$(n) output hold with respect to MCK 266 MHz 200 MHz	t_{DDKHCX}	 2.7 3.5	—	ns	4
MCK to MDQS	t_{DDKHMH}	–0.75	0.6	ns	5
MDQ/MECC/MDM output setup with respect to MDQS 266 MHz 200 MHz	t_{DDKHDS}, t_{DDKLDS}	 1.0 1.2	—	ns	6
MDQ/MECC/MDM output hold with respect to MDQS 266 MHz 200 MHz	t_{DDKHDX}, t_{DDKLDX}	 1.0 1.2	—	ns	6
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.6$	$-0.5 \times t_{MCK} + 0.6$	ns	7

Figure 7 provides the AC test load for the DDR bus.

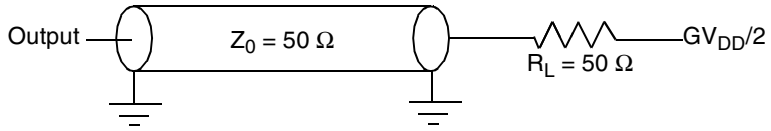

Figure 7. DDR AC Test Load

Table 21. DDR and DDR2 SDRAM Measurement Conditions

Symbol	DDR	DDR2	Unit	Notes
V_{TH}	MV_{REF} ± 0.31 V	MV_{REF} ± 0.25 V	V	1
V_{OUT}	0.5 × GV_{DD}	0.5 × GV_{DD}	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

Figure 8 shows the DDR SDRAM output timing diagram for source synchronous mode.

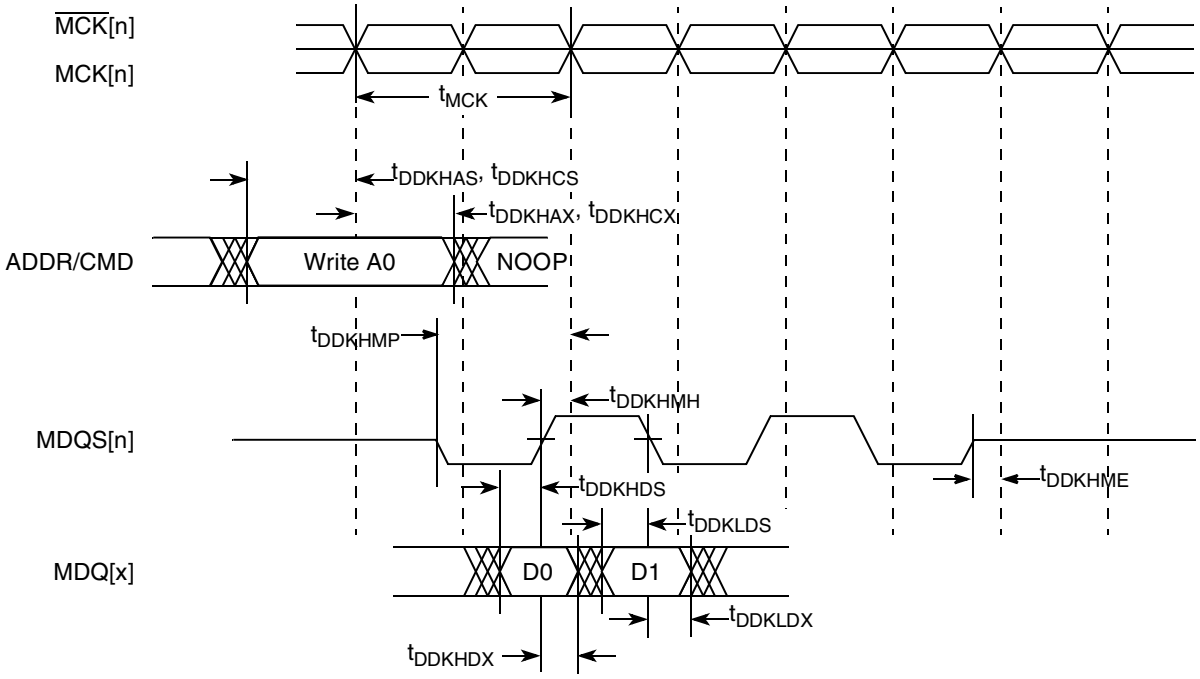

Figure 8. DDR SDRAM Output Timing Diagram for Source Synchronous Mode

Table 39. Local Bus General Timing Parameters—DLL Enabled (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus clock to LALE rise	t_{LBKHLR}	—	4.5	ns	—
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	4.5	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	4.5	ns	3
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	4.5	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	1.0	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	1.0	—	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	3.8	ns	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to rising edge of LSYNC_IN.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on LALE output pin equals to the load on LAD output pins.
8. For purposes of active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.

Table 40 describes the general timing parameters of the local bus interface of the device.

Table 40. Local Bus General Timing Parameters—DLL Bypass Mode

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	15	—	ns	2
Input setup to local bus clock	t_{LBIVKH}	7	—	ns	3, 4
Input hold from local bus clock	t_{LBIXKH}	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7
Local bus clock to output valid	t_{LBKHOV}	—	3	ns	3

Figure 22 through Figure 27 show the local bus signals.

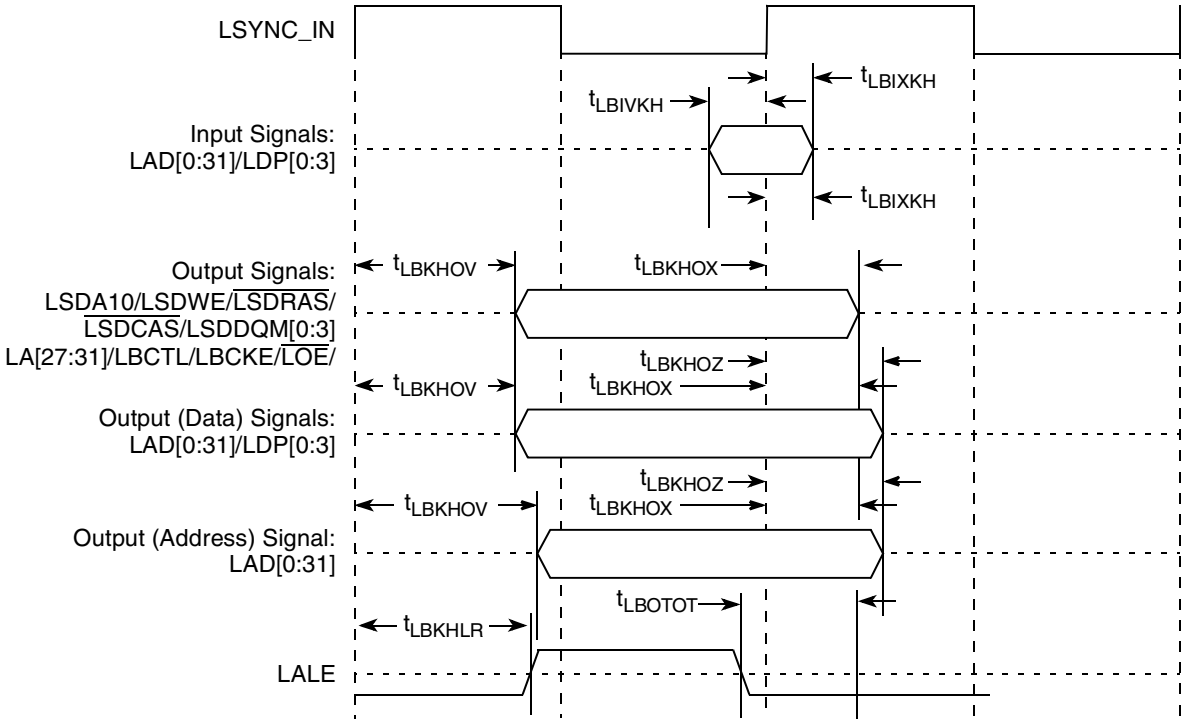

Figure 22. Local Bus Signals, Nonspecial Signals Only (DLL Enabled)

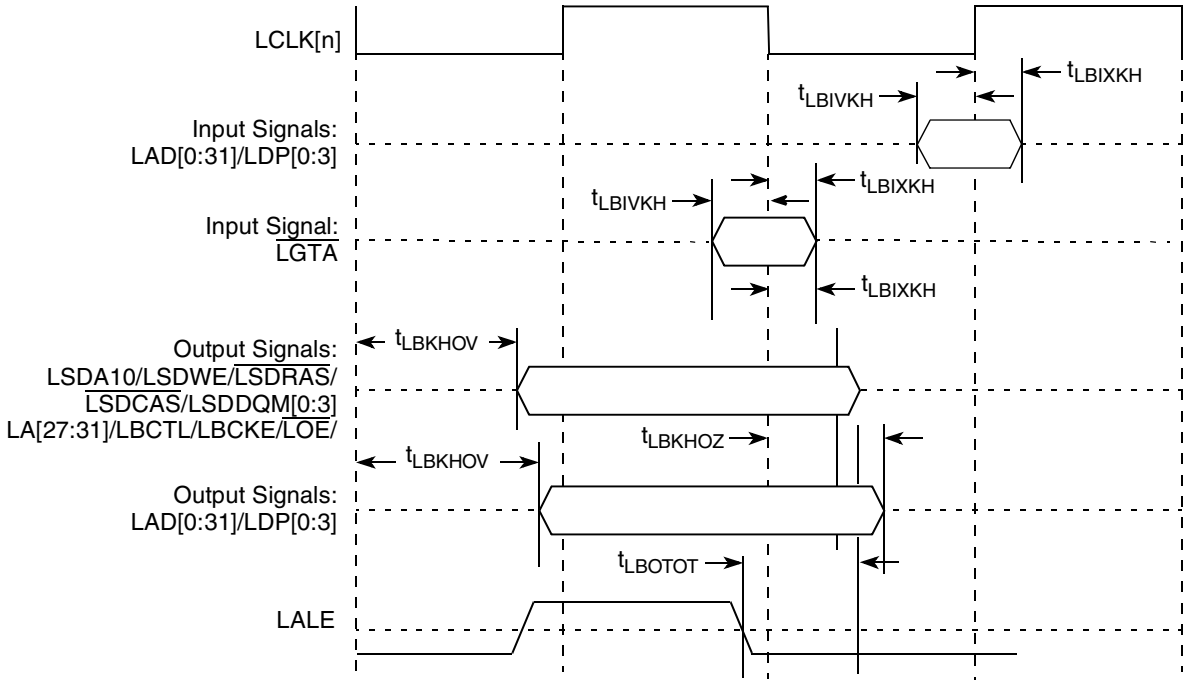

Figure 23. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

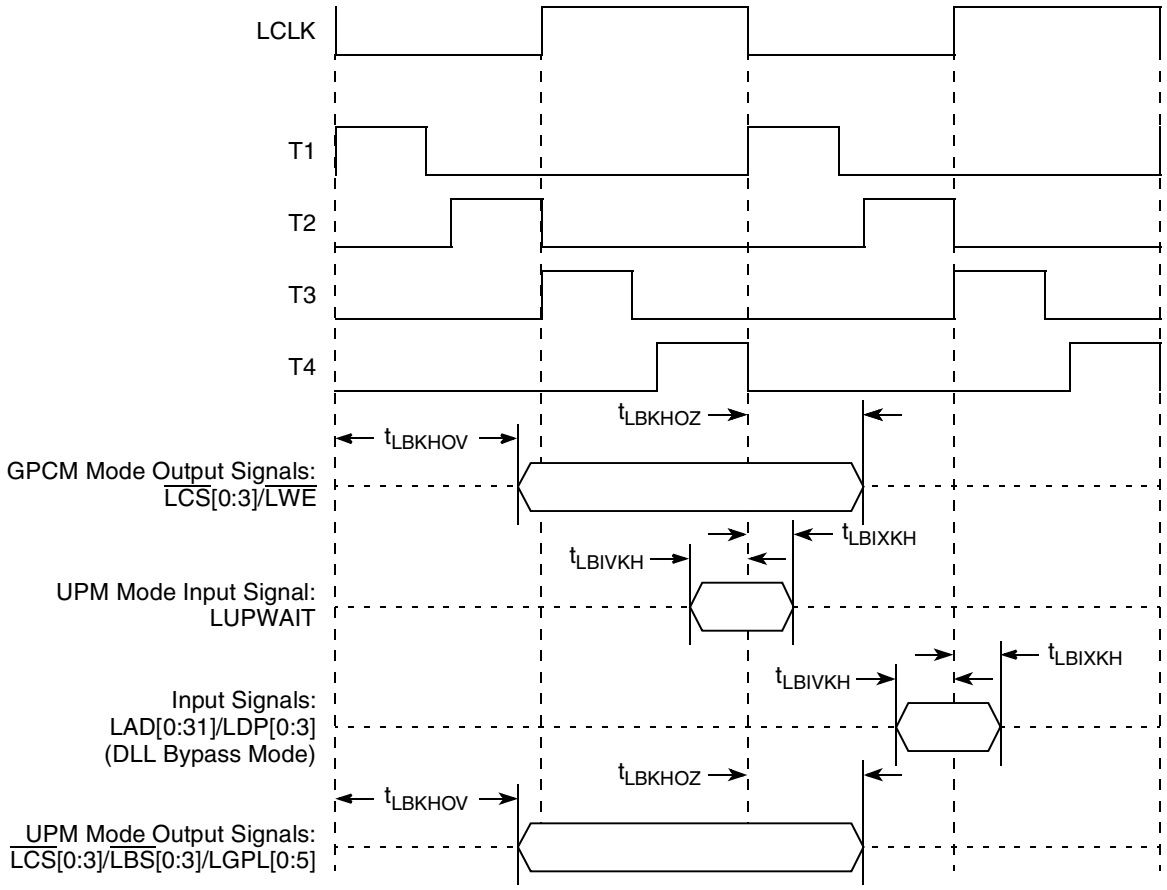

Figure 26. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Bypass Mode)

Figure 28 provides the AC test load for TDO and the boundary-scan outputs of the device.

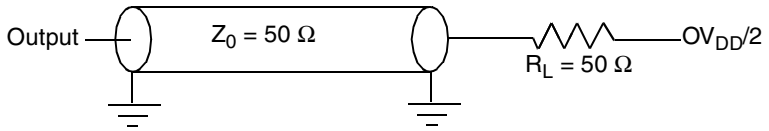

Figure 28. AC Test Load for the JTAG Interface

Figure 29 provides the JTAG clock input timing diagram.

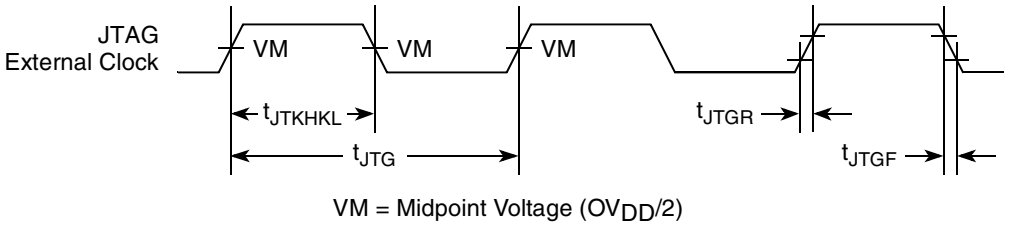

Figure 29. JTAG Clock Input Timing Diagram

Figure 30 provides the $\overline{\text{TRST}}$ timing diagram.

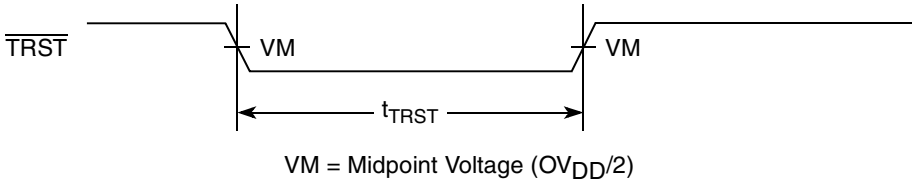

Figure 30. $\overline{\text{TRST}}$ Timing Diagram

Figure 31 provides the boundary-scan timing diagram.

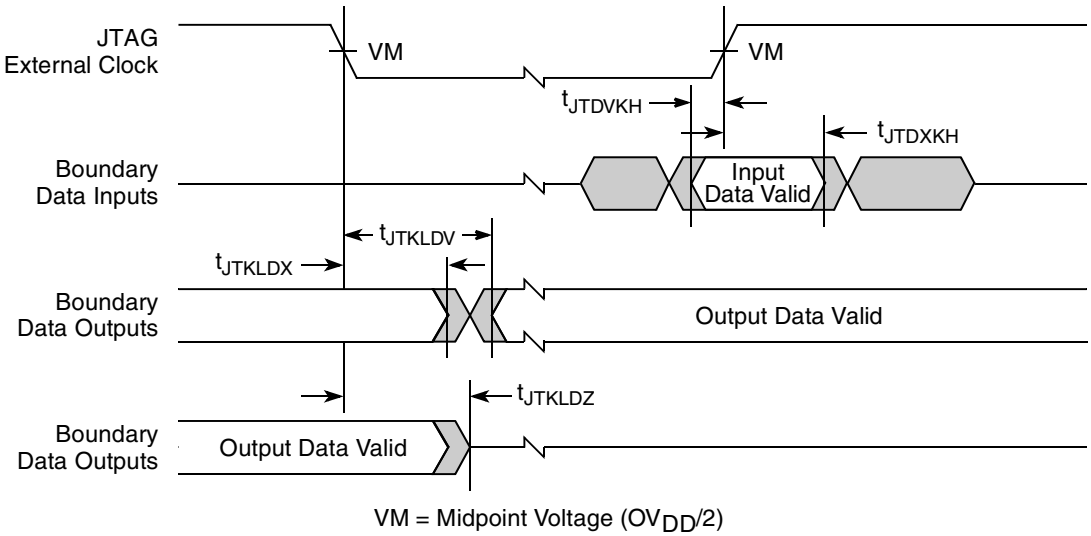

Figure 31. Boundary-Scan Timing Diagram

Figure 32 provides the test access port timing diagram.

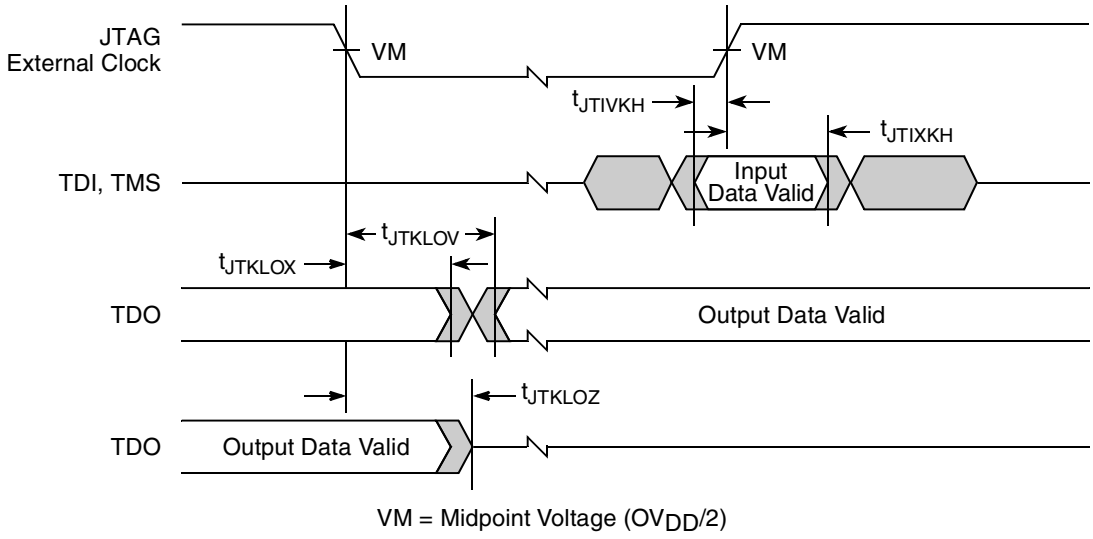

Figure 32. Test Access Port Timing Diagram

Figure 37 shows the PCI output AC timing conditions.

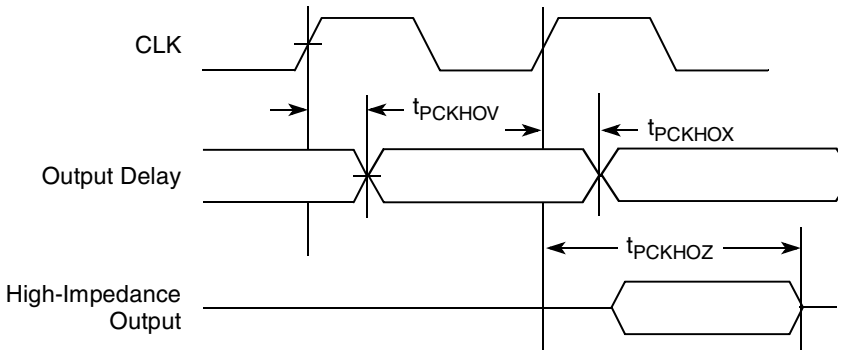

Figure 37. PCI Output AC Timing Measurement Condition

13 Timers

This section describes the DC and AC electrical specifications for the timers of the MPC8358E.

13.1 Timers DC Electrical Characteristics

Table 48 provides the DC electrical characteristics for the device timer pins, including TIN, $\overline{\text{TOUT}}$, $\overline{\text{TGATE}}$, and RTC_CLK.

Table 48. Timers DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V ≤ V_{IN} ≤ OV_{DD}	—	±10	μA

13.2 Timers AC Timing Specifications

Table 49 provides the timer input and output AC timing specifications.

Table 49. Timers Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Typ	Unit
Timers inputs—minimum pulse width	t_{TIWID}	20	ns

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. Timers inputs and outputs are asynchronous to any visible clock. Timers outputs should be synchronized before use by any external synchronous logic. Timers inputs are required to be valid for at least t_{TIWID} ns to ensure proper operation.

Table 56. TDM/SI DC Electrical Characteristics (continued)

Characteristic	Symbol	Condition	Min	Max	Unit
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	$0 V \le V_{IN} \le OV_{DD}$	—	±10	μA

17.2 TDM/SI AC Timing Specifications

Table 57 provides the TDM/SI input and output AC timing specifications.

Table 57. TDM/SI AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Max[3]	Unit
TDM/SI outputs—External clock delay	t_{SEKHOV}	2	10	ns
TDM/SI outputs—External clock high impedance	t_{SEKHOX}	2	10	ns
TDM/SI inputs—External clock input setup time	t_{SEIVKH}	5	—	ns
TDM/SI inputs—External clock input hold time	t_{SEIXKH}	2	—	ns

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{SEKHOX} symbolizes the TDM/SI outputs external timing (SE) for the time $t_{TDM/SI}$ memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).
3. Timings are measured from the positive or negative edge of the clock, according to SIxMR [CE] and SITXCEI[TXCEIx]. See the *MPC8360E Integrated Communications Processor Family Reference Manual* for more details.

Figure 43 provides the AC test load for the TDM/SI.

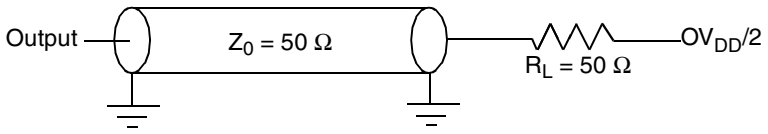

Figure 43. TDM/SI AC Test Load

Figure 44 represents the AC timing from Table 55. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

20 USB

This section provides the AC and DC electrical specifications for the USB interface of the MPC8358E.

20.1 USB DC Electrical Characteristics

Table 63 provides the DC electrical characteristics for the USB interface.

Table 63. USB DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
High-level output voltage, I_{OH} = –100 µA	V_{OH}	OV_{DD} – 0.4	—	V
Low-level output voltage, I_{OL} = 100 µA	V_{OL}	—	0.2	V
Input current	I_{IN}	—	±10	µA

20.2 USB AC Electrical Specifications

Table 64 describes the general timing parameters of the USB interface of the device.

Table 64. USB General Timing Parameters

Parameter	Symbol[1]	Min	Max	Unit	Notes
USB clock cycle time	t_{USCK}	20.83	—	ns	Full speed 48 MHz
USB clock cycle time	t_{USCK}	166.67	—	ns	Low speed 6 MHz
Skew between TXP and TXN	t_{USTSPN}	—	5	ns	—
Skew among RXP, RXN, and RXD	$t_{USRSPND}$	—	10	ns	Full speed transitions
Skew among RXP, RXN, and RXD	t_{USRPND}	—	100	ns	Low speed transitions

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(state)(signal)}}$ for receive signals and $t_{\text{(first two letters of functional block)(state)(signal)}}$ for transmit signals. For example, $t_{USRSPND}$ symbolizes USB timing (US) for the USB receive signals skew (RS) among RXP, RXN, and RXD (PND). Also, t_{USTSPN} symbolizes USB timing (US) for the USB transmit signals skew (TS) between TXP and TXN (PN).
2. Skew measurements are done at $OV_{DD}/2$ of the rising or falling edge of the signals.

Figure 51 provide the AC test load for the USB.

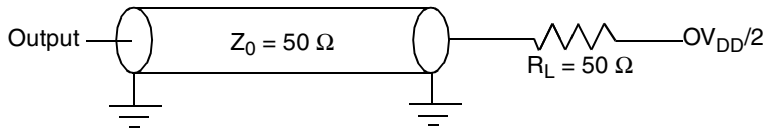

Figure 51. USB AC Test Load

21 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions. The MPC8358E is available in a plastic ball grid array (PBGA), see Section 21.1, “Package Parameters for the PBGA Package,” and Section 21.2, “Mechanical Dimensions of the PBGA Package,” for information on the package.

21.1 Package Parameters for the PBGA Package

The package parameters for rev 2.0 silicon are as provided in the following list. The package type is **2**9 mm x 29 mm, 668 plastic ball grid array (PBGA).

Package outline	**29 mm x 29 mm**
Interconnects	**668**
Pitch	**1.00 mm**
Module height (typical)	1.46 mm
Solder Balls	62 Sn/36 Pb/2 Ag (ZQ package)
	95.5 Sn/0.5 Cu/4Ag (VR package)
Ball diameter (typical)	0.64 mm

Table 65. MPC8358E PBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
TDI	AE8	I	OV_{DD}	4
TDO	AG7	O	OV_{DD}	3
TMS	AH7	I	OV_{DD}	4
$\overline{\text{TRST}}$	AG8	I	OV_{DD}	4
Test				
TEST	AF9	I	OV_{DD}	7
$\overline{\text{TEST\_SEL}}$	AE27	I	GV_{DD}	9
PMC				
$\overline{\text{QUIESCE}}$	AF4	O	OV_{DD}	—
System Control				
$\overline{\text{PORESET}}$	AE9	I	OV_{DD}	—
$\overline{\text{HRESET}}$	AG9	I/O	OV_{DD}	1
$\overline{\text{SRESET}}$	AH10	I/O	OV_{DD}	2
Thermal Management				
THERM0	K25	I	GV_{DD}	—
THERM1	AA26	I	GV_{DD}	—
Power and Ground Signals				
$AV_{DD}1$	AF8	Power for LBIU DLL (1.2 V)	$AV_{DD}1$	—
$AV_{DD}2$	AH8	Power for CE PLL (1.2 V)	$AV_{DD}2$	—
$AV_{DD}5$	AB26	Power for e300 PLL (1.2 V)	$AV_{DD}5$	—
$AV_{DD}6$	AH9	Power for system PLL (1.2 V)	$AV_{DD}6$	—
GND	C16, D11, D21, E24, F7, J10, J12, J15, J16, J17, J28, K11, K13, K14, K17, K18, L4, L9, L11, L12, L13, L14, L15, L16, L17, L18, L19, L24, M10, M11, M14, M15, M18, M19, N11, N18, N25, P9, P11, P18, P19, R9, R11, R14, R15, R18, R19, R26, T10, T11, T14, T15, T18, T25, U10, U11, U18, V9, V11, V14, V15, V18, V24, V27, W18, W19, Y11, Y14, Y18, Y19, Y25, Y27, AB4, AB27, AC27, AE20, AE24, AF5, AF15, AG10	—	—	—

Note that *lb_clk* is not the external local bus or DDRC2 frequency; *lb_clk* passes through the a LB clock divider to create the external local bus clock outputs (LSYNC_OUT and LCLK[0:2]). The LB clock divider ratio is controlled by LCRR[CLKDIV].

In addition, some of the internal units may be required to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory mapped register after the device comes out of reset. Table 66 specifies which units have a configurable clock frequency.

Table 66. Configurable Clock Units

Unit	Default Frequency	Options
Security core	*csb_clk*/3	Off, *csb_clk*[1], *csb_clk*/2, *csb_clk*/3
PCI and DMA complex	csb_clk	Off, *csb_clk*

1 With limitation, only for slow csb_clk rates, up to 166 MHz.

Table 67 provides the operating frequencies for the PBGA package under recommended operating conditions (see Table 2). All frequency combinations shown in the table below may not be available. Maximum operating frequencies depend on the part ordered, see Section 25.1, "Part Numbers Fully Addressed by this Document," for part ordering details and contact your Freescale sales representative or authorized distributor for more information.

Table 67. Operating Frequencies for the PBGA Package

Characteristic[1]	400 MHz	Unit
e300 core frequency (*core_clk*)	266–400	MHz
Coherent system bus frequency (*csb_clk*)	133–266	MHz
QUICC Engine frequency (*ce_clk*)	266–400	MHz
DDR and DDR2 memory bus frequency (MCLK)[2]	100–133	MHz
Local bus frequency (LCLK*n*)[3]	16.67–133	MHz
PCI input frequency (CLKIN or PCI_CLK)	25–66.67	MHz
Security core maximum internal operating frequency	133	MHz

1 The CLKIN frequency, RCWL[SPMF], and RCWL[COREPLL] settings must be chosen such that the resulting *csb_clk*, MCLK, LCLK[0:2], and *core_clk* frequencies do not exceed their respective maximum or minimum operating frequencies.

2 The DDR data rate is 2x the DDR memory bus frequency.

3 The local bus frequency is 1/2, 1/4, or 1/8 of the *lb_clk* frequency (depending on LCRR[CLKDIV]) which is in turn 1x or 2x the *csb_clk* frequency (depending on RCWL[LBCM]).

Table 72. QUICC Engine Block PLL Multiplication Factors (continued)

RCWL[CEPMF]	RCWL[CEPDF]	QUICC Engine PLL Multiplication Factor = RCWL[CEPMF]/ (1 + RCWL[CEPDF])
00101	0	× 5
00110	0	× 6
00111	0	× 7
01000	0	× 8
01001	0	× 9
01010	0	× 10
01011	0	× 11
01100	0	× 12
01101	0	× 13
01110	0	× 14
01111	0	× 15
10000	0	× 16
10001	0	× 17
10010	0	× 18
10011	0	× 19
10100	0	× 20
10101	0	× 21
10110	0	× 22
10111	0	× 23
11000	0	× 24
11001	0	× 25
11010	0	× 26
11011	0	× 27
11100	0	× 28
11101	0	× 29
11110	0	× 30
11111	0	× 31
00011	1	× 1.5
00101	1	× 2.5
00111	1	× 3.5
01001	1	× 4.5

23.2.1 Estimation of Junction Temperature with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)

T_A = ambient temperature for the package (°C)

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. As a general statement, the value obtained on a single-layer board is appropriate for a tightly packed printed-circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

23.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

The thermal performance of a device cannot be adequately predicted from the junction-to-ambient thermal resistance. The thermal performance of any component is strongly dependent on the power dissipation of surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package will be approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device. At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature (°C)

T_B = board temperature at the package perimeter (°C)

$R_{\theta JA}$ = junction to board thermal resistance (°C/W) per JESD51-8

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

23.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (°C)

T_C = case temperature of the package (°C)

$R_{\theta JC}$ = junction to case thermal resistance (°C/W)

P_D = power dissipation (W)

24 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8358E. Additional information can be found in *MPC8360E/MPC8358E PowerQUICC Design Checklist* (AN3097).

24.1 System Clocking

The device includes two PLLs, as follows.

- The platform PLL ($AV_{DD}1$) generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 22.1, “System PLL Configuration.”
- The e300 core PLL ($AV_{DD}2$) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 22.2, “Core PLL Configuration.”

24.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages will be derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide five independent filter circuits as illustrated in Figure 54, one to each of the five AV_{DD} pins. By providing independent filters to each PLL, the opportunity to cause noise injection from one PLL to the other is reduced.

24.5 Output Buffer DC Impedance

The device drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 55). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

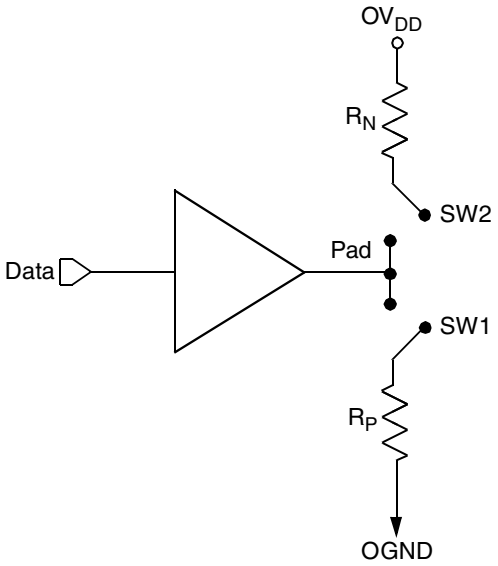

Figure 55. Driver Impedance Measurement

The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = 1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

Table 77 summarizes the signal impedance targets. The driver impedance are targeted at minimum V_{DD}, nominal OV_{DD}, 105°C.

Table 77. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R_N	42 Target	25 Target	20 Target	Z_0	W
R_P	42 Target	25 Target	20 Target	Z_0	W