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Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	24
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	768 x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 19x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SSOP (0.209", 5.30mm Width)
Supplier Device Package	28-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f24k22-e-ss

6.3 Register Definitions: Memory Control

REGISTER 6-1: EECON1: DATA EEPROM CONTROL 1 REGISTER

R/W-x	R/W-x	U-0	R/W-0	R/W-x	R/W-0	R/S-0	R/S-0
EEPGD	CFGS	—	FREE	WRERR	WREN	WR	RD
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit		
S = Bit can be set by software, but not cleared		U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **EEPGD:** Flash Program or Data EEPROM Memory Select bit
- 1 = Access Flash program memory
- 0 = Access data EEPROM memory

bit 6 **CFGS:** Flash Program/Data EEPROM or Configuration Select bit
- 1 = Access Configuration registers
- 0 = Access Flash program or data EEPROM memory

bit 5 **Unimplemented:** Read as '0'

bit 4 **FREE:** Flash Row (Block) Erase Enable bit
- 1 = Erase the program memory block addressed by TBLPTR on the next WR command (cleared by completion of erase operation)
- 0 = Perform write-only

bit 3 **WRERR:** Flash Program/Data EEPROM Error Flag bit(1)
- 1 = A write operation is prematurely terminated (any Reset during self-timed programming in normal operation, or an improper write attempt)
- 0 = The write operation completed

bit 2 **WREN:** Flash Program/Data EEPROM Write Enable bit
- 1 = Allows write cycles to Flash program/data EEPROM
- 0 = Inhibits write cycles to Flash program/data EEPROM

bit 1 **WR:** Write Control bit
- 1 = Initiates a data EEPROM erase/write cycle or a program memory erase cycle or write cycle. (The operation is self-timed and the bit is cleared by hardware once write is complete. The WR bit can only be set (not cleared) by software.)
- 0 = Write cycle to the EEPROM is complete

bit 0 **RD:** Read Control bit
- 1 = Initiates an EEPROM read (Read takes one cycle. RD is cleared by hardware. The RD bit can only be set (not cleared) by software. RD bit cannot be set when EEPGD = 1 or CFGS = 1.)
- 0 = Does not initiate an EEPROM read

Note 1: When a WRERR occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

7.3 Reading the Data EEPROM Memory

To read a data memory location, the user must write the address to the EEADR register, clear the EEPGD control bit of the EECON1 register and then set control bit, RD. The data is available on the very next instruction cycle; therefore, the EEDATA register can be read by the next instruction. EEDATA will hold this value until another read operation, or until it is written to by the user (during a write operation).

The basic process is shown in Example 7-1.

7.4 Writing to the Data EEPROM Memory

To write an EEPROM data location, the address must first be written to the EEADR register and the data written to the EEDATA register. The sequence in Example 7-2 must be followed to initiate the write cycle.

The write will not begin if this sequence is not exactly followed (write 55h to EECON2, write 0AAh to EECON2, then set WR bit) for each byte. It is strongly recommended that interrupts be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable writes. This mechanism prevents accidental writes to data EEPROM due to unexpected code execution (i.e., runaway programs). The WREN bit should be kept clear at all times, except when updating the EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, EECON1, EEADR and EEDATA cannot be modified. The WR bit will be inhibited from being set unless the WREN bit is set. Both WR and WREN cannot be set with the same instruction.

At the completion of the write cycle, the WR bit is cleared by hardware and the EEPROM Interrupt Flag bit, EEIF, is set. The user may either enable this interrupt or poll this bit. EEIF must be cleared by software.

7.5 Write Verify

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

EXAMPLE 7-1: DATA EEPROM READ

```
        MOVLW   DATA_EE_ADDR   ;
        MOVWF   EEADR          ; Data Memory Address to read
        BCF     EECON1, EEPGD  ; Point to DATA memory
        BCF     EECON1, CFGS   ; Access EEPROM
        BSF     EECON1, RD     ; EEPROM Read
        MOVF    EEDATA, W      ; W = EEDATA
```

EXAMPLE 7-2: DATA EEPROM WRITE

```
                MOVLW   DATA_EE_ADDR_LOW   ;
                MOVWF   EEADR              ; Data Memory Address to write
                MOVLW   DATA_EE_ADDR_HI    ;
                MOVWF   EEADRH             ;
                MOVLW   DATA_EE_DATA       ;
                MOVWF   EEDATA             ; Data Memory Value to write
                BCF     EECON1, EEPGD      ; Point to DATA memory
                BCF     EECON1, CFGS       ; Access EEPROM
                BSF     EECON1, WREN       ; Enable writes
                BCF     INTCON, GIE        ; Disable Interrupts
                MOVLW   55h                ;
Required        MOVWF   EECON2             ; Write 55h
Sequence        MOVLW   0AAh               ;
                MOVWF   EECON2             ; Write 0AAh
                BSF     EECON1, WR         ; Set WR bit to begin write
                BSF     INTCON, GIE        ; Enable Interrupts

                                           ; User code execution
                BCF     EECON1, WREN       ; Disable writes on write complete (EEIF set)
```

TABLE 7-1: REGISTERS ASSOCIATED WITH DATA EEPROM MEMORY

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	109
EEADR	EEADR7	EEADR6	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0	—
EEADRH(1)	—	—	—	—	—	—	EEADR9	EEADR8	—
EEDATA	EEPROM Data Register								—
EECON2	EEPROM Control Register 2 (not a physical register)								—
EECON1	EEPGD	CFGS	—	FREE	WRERR	WREN	WR	RD	100
IPR2	OSCFIP	C1IP	C2IP	EEIP	BCL1IP	HLVDIP	TMR3IP	CCP2IP	122
PIR2	OSCFIF	C1IF	C2IF	EEIF	BCL1IF	HLVDIF	TMR3IF	CCP2IF	113
PIE2	OSCFIE	C1IE	C2IE	EEIE	BCL1IE	HLVDIE	TMR3IE	CCP2IE	118

Legend: — = unimplemented, read as '0'. Shaded bits are not used during EEPROM access.

Note 1: PIC18(L)F26K22 and PIC18(L)F46K22 only.

REGISTER 9-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
OSCFIF	C1IF	C2IF	EEIF	BCL1IF	HLVDIF	TMR3IF	CCP2IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **OSCFIF:** Oscillator Fail Interrupt Flag bit
1 = Device oscillator failed, clock input has changed to HFINTOSC (must be cleared by software)
0 = Device clock operating

bit 6 **C1IF:** Comparator C1 Interrupt Flag bit
1 = Comparator C1 output has changed (must be cleared by software)
0 = Comparator C1 output has not changed

bit 5 **C2IF:** Comparator C2 Interrupt Flag bit
1 = Comparator C2 output has changed (must be cleared by software)
0 = Comparator C2 output has not changed

bit 4 **EEIF:** Data EEPROM/Flash Write Operation Interrupt Flag bit
1 = The write operation is complete (must be cleared by software)
0 = The write operation is not complete or has not been started

bit 3 **BCL1IF:** MSSP1 Bus Collision Interrupt Flag bit
1 = A bus collision occurred (must be cleared by software)
0 = No bus collision occurred

bit 2 **HLVDIF:** Low-Voltage Detect Interrupt Flag bit
1 = A low-voltage condition occurred (direction determined by the VDIRMAG bit of the HLVDCON register)
0 = A low-voltage condition has not occurred

bit 1 **TMR3IF:** TMR3 Overflow Interrupt Flag bit
1 = TMR3 register overflowed (must be cleared by software)
0 = TMR3 register did not overflow

bit 0 **CCP2IF:** CCP2 Interrupt Flag bit

Capture mode:
1 = A TMR register capture occurred (must be cleared by software)
0 = No TMR register capture occurred

Compare mode:
1 = A TMR register compare match occurred (must be cleared by software)
0 = No TMR register compare match occurred

PWM mode:
Unused in this mode.

TABLE 10-11: PORTD I/O SUMMARY

Pin Name	Function	TRIS Setting	ANSEL setting	Pin Type	Buffer Type	Description
RD0/SCK2/SCL2/AN20	RD0	0	0	O	DIG	LATD<0> data output; not affected by analog input.
		1	0	I	ST	PORTD<0> data input; disabled when analog input enabled.
	SCK2	0	0	O	DIG	MSSP2 SPI Clock output.
		1	0	I	ST	MSSP2 SPI Clock input.
	SCL2	0	0	O	DIG	MSSP2 I^2C Clock output.
		1	0	I	I^2C	MSSP2 I^2C Clock input.
	AN20	1	1	I	AN	Analog input 20.
RD1/CCP4/SDI2/SDA2/AN21	RD1	0	0	O	DIG	LATD<1> data output; not affected by analog input.
		1	0	I	ST	PORTD<1> data input; disabled when analog input enabled.
	CCP4	0	0	O	DIG	Compare 4 output/PWM 4 output.
		1	0	I	ST	Capture 4 input.
	SDI2	1	0	I	ST	MSSP2 SPI data input.
	SDA2	0	0	O	DIG	MSSP2 I^2C data output.
		1	0	I	I^2C	MSSP2 I^2C data input.
	AN21	1	1	I	AN	Analog input 21.
RD2/P2B/AN22	RD2	0	0	O	DIG	LATD<2> data output; not affected by analog input.
		1	0	I	ST	PORTD<2> data input; disabled when analog input enabled.
	P2B(1)	0	0	O	DIG	Enhanced CCP2 PWM output 2.
	AN22	1	1	I	AN	Analog input 22.
RD3/P2C/$\overline{SS2}$/AN23	RD3	0	0	O	DIG	LATD<3> data output; not affected by analog input.
		1	0	I	ST	PORTD<3> data input; disabled when analog input enabled.
	P2C	0	0	O	DIG	Enhanced CCP2 PWM output 4.
	$\overline{SS2}$	1	0	I	TTL	MSSP2 SPI slave select input.
	AN23	1	1	I	AN	Analog input 23.
RD4/P2D/SDO2/AN24	RD4	0	0	O	DIG	LATD<4> data output; not affected by analog input.
		1	0	I	ST	PORTD<4> data input; disabled when analog input enabled.
	P2D	0	0	O	DIG	Enhanced CCP2 PWM output 3.
	SDO2	0	0	O	DIG	MSSP2 SPI data output.
	AN24	1	1	I	AN	Analog input 24.
RD5/P1B/AN25	RD5	0	0	O	DIG	LATD<5> data output; not affected by analog input.
		1	0	I	ST	PORTD<5> data input; disabled when analog input enabled.
	P1B	0	0	O	DIG	Enhanced CCP1 PWM output 2.
	AN25	1	1	I	AN	Analog input 25.

Legend: AN = Analog input or output; TTL = TTL compatible input; HV = High Voltage; OD = Open Drain; XTAL = Crystal; CMOS = CMOS compatible input or output; ST = Schmitt Trigger input with CMOS levels; I^2C = Schmitt Trigger input with I^2C.

Note 1: Default pin assignment for P2B, T3CKI, CCP3 and CCP2 when Configuration bits PB2MX, T3CMX, CCP3MX and CCP2MX are set.

15.5.3 SLAVE TRANSMISSION

When the R/$\overline{W}$ bit of the incoming address byte is set and an address match occurs, the R/$\overline{W}$ bit of the SSPxSTAT register is set. The received address is loaded into the SSPxBUF register, and an $\overline{ACK}$ pulse is sent by the slave on the ninth bit.

Following the $\overline{ACK}$, slave hardware clears the CKP bit and the SCLx pin is held low (see **Section 15.5.6 "Clock Stretching"** for more detail). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data.

The transmit data must be loaded into the SSPxBUF register which also loads the SSPxSR register. Then the SCLx pin should be released by setting the CKP bit of the SSPxCON1 register. The eight data bits are shifted out on the falling edge of the SCLx input. This ensures that the SDAx signal is valid during the SCLx high time.

The $\overline{ACK}$ pulse from the master-receiver is latched on the rising edge of the ninth SCLx input pulse. This $\overline{ACK}$ value is copied to the ACKSTAT bit of the SSPxCON2 register. If ACKSTAT is set (not $\overline{ACK}$), then the data transfer is complete. In this case, when the not $\overline{ACK}$ is latched by the slave, the slave goes Idle and waits for another occurrence of the Start bit. If the SDAx line was low ($\overline{ACK}$), the next transmit data must be loaded into the SSPxBUF register. Again, the SCLx pin must be released by setting bit CKP.

An MSSPx interrupt is generated for each data transfer byte. The SSPxIF bit must be cleared by software and the SSPxSTAT register is used to determine the status of the byte. The SSPxIF bit is set on the falling edge of the ninth clock pulse.

15.5.3.1 Slave Mode Bus Collision

A slave receives a Read request and begins shifting data out on the SDAx line. If a bus collision is detected and the SBCDE bit of the SSPxCON3 register is set, the BCLxIF bit of the PIRx register is set. Once a bus collision is detected, the slave goes Idle and waits to be addressed again. User software can use the BCLxIF bit to handle a slave bus collision.

15.5.3.2 7-bit Transmission

A master device can transmit a read request to a slave, and then clock data out of the slave. The list below outlines what software for a slave will need to do to accomplish a standard transmission. Figure 15-18 can be used as a reference to this list.

1. Master sends a Start condition on SDAx and SCLx.
2. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
3. Matching address with R/$\overline{W}$ bit set is received by the slave setting SSPxIF bit.
4. Slave hardware generates an $\overline{ACK}$ and sets SSPxIF.
5. SSPxIF bit is cleared by user.
6. Software reads the received address from SSPxBUF, clearing BF.
7. R/$\overline{W}$ is set so CKP was automatically cleared after the $\overline{ACK}$.
8. The slave software loads the transmit data into SSPxBUF.
9. CKP bit is set releasing SCLx, allowing the master to clock the data out of the slave.
10. SSPxIF is set after the $\overline{ACK}$ response from the master is loaded into the ACKSTAT register.
11. SSPxIF bit is cleared.
12. The slave software checks the ACKSTAT bit to see if the master wants to clock out more data.

Note 1: If the master $\overline{ACK}$s the clock will be stretched.

2: ACKSTAT is the only bit updated on the rising edge of SCLx (9th) rather than the falling.

13. Steps 9-13 are repeated for each transmitted byte.
14. If the master sends a not $\overline{ACK}$; the clock is not held, but SSPxIF is still set.
15. The master sends a Restart condition or a Stop.
16. The slave is no longer addressed.

15.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSPx module configured as an I^2C slave in 10-bit Addressing mode (Figure 15-20) and is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I^2C communication.

1. Bus starts Idle.
2. Master sends Start condition; S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
3. Master sends matching high address with R/$\overline{W}$ bit clear; UA bit of the SSPxSTAT register is set.
4. Slave sends $\overline{ACK}$ and SSPxIF is set.
5. Software clears the SSPxIF bit.
6. Software reads received address from SSPxBUF clearing the BF flag.
7. Slave loads low address into SSPxADD, releasing SCLx.
8. Master sends matching low address byte to the slave; UA bit is set.

Note:	Updates to the SSPxADD register are not allowed until after the $\overline{ACK}$ sequence.

9. Slave sends $\overline{ACK}$ and SSPxIF is set.

Note:	If the low address does not match, SSPxIF and UA are still set so that the slave software can set SSPxADD back to the high address. BF is not set because there is no match. CKP is unaffected.

10. Slave clears SSPxIF.
11. Slave reads the received matching address from SSPxBUF clearing BF.
12. Slave loads high address into SSPxADD.
13. Master clocks a data byte to the slave and clocks out the slaves $\overline{ACK}$ on the 9th SCLx pulse; SSPxIF is set.
14. If SEN bit of SSPxCON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSPxIF.
16. Slave reads the received byte from SSPxBUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCLx.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.

15.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSPxADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCLx line is held low are the same. Figure 15-21 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 15-22 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.

TABLE 16-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 0											
	FOSC = 64.000 MHz			FOSC = 18.432 MHz			FOSC = 16.000 MHz			FOSC = 11.0592 MHz		
	Actual Rate	% Error	SPBRGx value (decimal)	Actual Rate	% Error	SPBRGx value (decimal)	Actual Rate	% Error	SPBRGx value (decimal)	Actual Rate	% Error	SPBRGx value (decimal)
300	—	—	—	—	—	—	—	—	—	—	—	—
1200	—	—	—	—	—	—	—	—	—	—	—	—
2400	—	—	—	—	—	—	—	—	—	—	—	—
9600	—	—	—	9600	0.00	119	9615	0.16	103	9600	0.00	71
10417	—	—	—	10378	-0.37	110	10417	0.00	95	10473	0.53	65
19.2k	19.23k	0.16	207	19.20k	0.00	59	19.23k	0.16	51	19.20k	0.00	35
57.6k	57.97k	0.64	68	57.60k	0.00	19	58.82k	2.12	16	57.60k	0.00	11
115.2k	114.29k	-0.79	34	115.2k	0.00	9	111.1k	-3.55	8	115.2k	0.00	5

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 0											
	FOSC = 8.000 MHz			FOSC = 4.000 MHz			FOSC = 3.6864 MHz			FOSC = 1.000 MHz		
	Actual Rate	% Error	SPBRGx value (decimal)	Actual Rate	% Error	SPBRGx value (decimal)	Actual Rate	% Error	SxBRGx value (decimal)	Actual Rate	% Error	SPBRGx value (decimal)
300	—	—	—	—	—	—	—	—	—	300	0.16	207
1200	—	—	—	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	—	—	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19231	0.16	25	19.23k	0.16	12	19.2k	0.00	11	—	—	—
57.6k	55556	-3.55	8	—	—	—	57.60k	0.00	3	—	—	—
115.2k	—	—	—	—	—	—	115.2k	0.00	1	—	—	—

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 1											
	FOSC = 64.000 MHz			FOSC = 18.432 MHz			FOSC = 16.000 MHz			FOSC = 11.0592 MHz		
	Actual Rate	% Error	SPBRGHx: SPBRGx (decimal)	Actual Rate	% Error	SPBRGHx: SPBRGx (decimal)	Actual Rate	% Error	SPBRGHx :SPBRGx (decimal)	Actual Rate	% Error	SPBRGHx: SPBRGx (decimal)
300	300.0	0.00	13332	300.0	0.00	3839	300.03	0.01	3332	300.0	0.00	2303
1200	1200.1	0.01	3332	1200	0.00	959	1200.5	0.04	832	1200	0.00	575
2400	2399	-0.02	1666	2400	0.00	479	2398	-0.08	416	2400	0.00	287
9600	9592	-0.08	416	9600	0.00	119	9615	0.16	103	9600	0.00	71
10417	10417	0.00	383	10378	-0.37	110	10417	0.00	95	10473	0.53	65
19.2k	19.23k	0.16	207	19.20k	0.00	59	19.23k	0.16	51	19.20k	0.00	35
57.6k	57.97k	0.64	68	57.60k	0.00	19	58.82k	2.12	16	57.60k	0.00	11
115.2k	114.29k	-0.79	34	115.2k	0.00	9	111.11k	-3.55	8	115.2k	0.00	5

16.5.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the EUSART for Synchronous slave operation:

- SYNC = 1
- CSRC = 0
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTAx register configures the device for synchronous operation. Clearing the CSRC bit of the TXSTAx register configures the device as a slave. Clearing the SREN and CREN bits of the RCSTAx register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTAx register enables the EUSART. If the RXx/DTx or TXx/CKx pins are shared with an analog peripheral the analog I/O functions must be disabled by clearing the corresponding ANSEL bits.

RXx/DTx and TXx/CKx pin output drivers must be disabled by setting the corresponding TRIS bits.

16.5.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see **Section 16.5.1.3 “Synchronous Master Transmission”)**, except in the case of the Sleep mode.

If two words are written to the TXREGx and then the SLEEP instruction is executed, the following will occur:

1. The first character will immediately transfer to the TSR register and transmit.
2. The second word will remain in TXREGx register.
3. The TXxIF bit will not be set.
4. After the first character has been shifted out of TSR, the TXREGx register will transfer the second character to the TSR and the TXxIF bit will now be set.
5. If the PEIE/GIEL and TXxIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE/GIEH bit is also set, the program will call the Interrupt Service Routine.

16.5.2.2 Synchronous Slave Transmission Setup:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Set the RXx/DTx and TXx/CKx TRIS controls to ‘1’.
3. Clear the CREN and SREN bits.
4. If using interrupts, ensure that the GIE/GIEH and PEIE/GIEL bits of the INTCON register are set and set the TXxIE bit.
5. If 9-bit transmission is desired, set the TX9 bit.
6. Enable transmission by setting the TXEN bit.
7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
8. Start transmission by writing the Least Significant eight bits to the TXREGx register.

17.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/$\overline{\text{DONE}}$ bit
- Set the ADIF flag bit
- Update the ADRESH:ADRESL registers with new conversion result

17.2.3 DISCHARGE

The discharge phase is used to initialize the value of the capacitor array. The array is discharged after every sample. This feature helps to optimize the unity-gain amplifier, as the circuit always needs to charge the capacitor array, rather than charge/discharge based on previous measure values.

17.2.4 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/$\overline{\text{DONE}}$ bit can be cleared by software. The ADRESH:ADRESL registers will not be updated with the partially complete Analog-to-Digital conversion sample. Instead, the ADRESH:ADRESL register pair will retain the value of the previous conversion.

Note:	A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

17.2.5 DELAY BETWEEN CONVERSIONS

After the A/D conversion is completed or aborted, a 2 TAD wait is required before the next acquisition can be started. After this wait, the currently selected channel is reconnected to the charge holding capacitor commencing the next acquisition.

17.2.6 ADC OPERATION IN POWER-MANAGED MODES

The selection of the automatic acquisition time and A/D conversion clock is determined in part by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT<2:0> and ADCS<2:0> bits in ADCON2 should be updated in accordance with the clock source to be used in that mode. After entering the mode, an A/D acquisition or conversion may be started. Once started, the device should continue to be clocked by the same clock source until the conversion has been completed.

If desired, the device may be placed into the corresponding Idle mode during the conversion. If the device clock frequency is less than 1 MHz, the A/D FRC clock source should be selected.

17.2.7 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

17.2.8 SPECIAL EVENT TRIGGER

Two Special Event Triggers are available to start an A/D conversion: CTMU and CCP5. The Special Event Trigger source is selected using the TRIGSEL bit in ADCON1.

When TRIGSEL = 0, the CCP5 module is selected as the Special Event Trigger source. To enable the Special Event Trigger in the CCP module, set CCP5M<3:0> = 1011, in the CCP5CON register.

When TRIGSEL = 1, the CTMU module is selected. The CTMU module requires that the CTTRIG bit in CTMUCONH is set to enable the Special Event Trigger.

In addition to TRIGSEL bit, the following steps are required to start an A/D conversion:

- The A/D module must be enabled (ADON = 1)
- The appropriate analog input channel selected
- The minimum acquisition period set one of these ways:
  - Timing provided by the user
  - Selection made of an appropriate TACQ time

With these conditions met, the trigger sets the GO/$\overline{\text{DONE}}$ bit and the A/D acquisition starts.

If the A/D module is not enabled (ADON = 0), the module ignores the Special Event Trigger.

17.2.9 PERIPHERAL MODULE DISABLE

When a peripheral module is not used or inactive, the module can be disabled by setting the Module Disable bit in the PMD registers. This will reduce power consumption to an absolute minimum. Setting the PMD bits holds the module in Reset and disconnects the module's clock source. The Module Disable bit for the ADC module is ADCMD in the PMD2 Register. See **Section 3.0 "Power-Managed Modes"** for more information.

19.5 Measuring Time with the CTMU Module

Time can be precisely measured after the ratio (C/I) is measured from the current and capacitance calibration step by following these steps:

1. Initialize the A/D Converter and the CTMU.
2. Set EDG1STAT.
3. Set EDG2STAT.
4. Perform an A/D conversion.
5. Calculate the time between edges as $T = (C/I) * V$, where I is calculated in the current calibration step (**Section 19.3.1 "Current Source Calibration"**), C is calculated in the capacitance calibration step (**Section 19.3.2 "Capacitance Calibration"**) and V is measured by performing the A/D conversion.

It is assumed that the time measured is small enough that the capacitance, C_{OFFSET}, provides a valid voltage to the A/D Converter. For the smallest time measurement, always set the A/D Channel Select register (AD1CHS) to an unused A/D channel; the corresponding pin for which is not connected to any circuit board trace. This minimizes added stray capacitance, keeping the total circuit capacitance close to that of the A/D Converter itself (4-5 pF). To measure longer time intervals, an external capacitor may be connected to an A/D channel and this channel selected when making a time measurement.

FIGURE 19-3: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR TIME MEASUREMENT

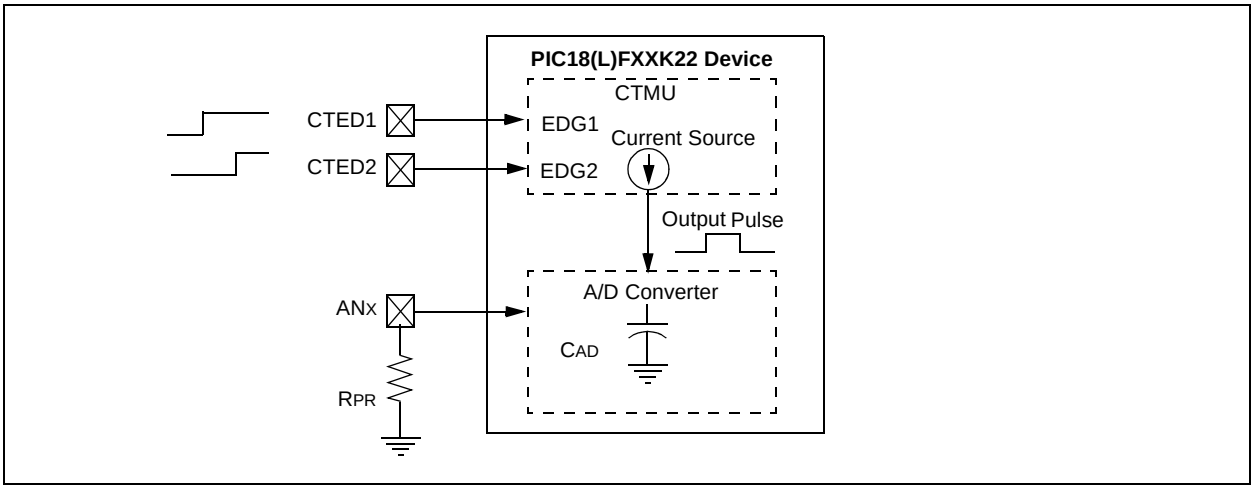


TABLE 24-2: DEVICE ID TABLE FOR THE PIC18(L)F2X/4XK22 FAMILY

DEV<10:3>	DEV<2:0>	Part Number
0101 0100	000	PIC18F46K22
	001	PIC18LF46K22
	010	PIC18F26K22
	011	PIC18LF26K22
0101 0101	000	PIC18F45K22
	001	PIC18LF45K22
	010	PIC18F25K22
	011	PIC18LF25K22
0101 0110	000	PIC18F44K22
	001	PIC18LF44K22
	010	PIC18F24K22
	011	PIC18LF24K22
0101 0111	000	PIC18F43K22
	001	PIC18LF43K22
	010	PIC18F23K22
	011	PIC18LF23K22

BNC **Branch if Not Carry**

Syntax: BNC n

Operands: -128 ≤ n ≤ 127

Operation: if CARRY bit is '0'
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0011	nnnn	nnnn

Description: If the CARRY bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a 2-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: HERE BNC Jump

Before Instruction
PC = address (HERE)
After Instruction
If CARRY = 0;
PC = address (Jump)
If CARRY = 1;
PC = address (HERE + 2)

BNN **Branch if Not Negative**

Syntax: BNN n

Operands: -128 ≤ n ≤ 127

Operation: if NEGATIVE bit is '0'
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0111	nnnn	nnnn

Description: If the NEGATIVE bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a 2-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: HERE BNN Jump

Before Instruction
PC = address (HERE)
After Instruction
If NEGATIVE = 0;
PC = address (Jump)
If NEGATIVE = 1;
PC = address (HERE + 2)

INCFSZ **Increment f, skip if 0**

Syntax: INCFSZ f {,d {,a}}

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) + 1 → dest,
skip if result = 0

Status Affected: None

Encoding:

0011	11da	ffff	ffff

Description: The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).
If the result is '0', the next instruction, which is already fetched, is discarded and a NOP is executed instead, making it a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    INCFSZ  CNT, 1, 0
NZERO   :
ZERO    :
```
Before Instruction
PC = Address (HERE)
After Instruction
CNT = CNT + 1
If CNT = 0;
PC = Address (ZERO)
If CNT ≠ 0;
PC = Address (NZERO)

INFSNZ **Increment f, skip if not 0**

Syntax: INFSNZ f {,d {,a}}

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) + 1 → dest,
skip if result ≠ 0

Status Affected: None

Encoding:

0100	10da	ffff	ffff

Description: The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).
If the result is not '0', the next instruction, which is already fetched, is discarded and a NOP is executed instead, making it a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    INFSNZ  REG, 1, 0
ZERO
NZERO
```
Before Instruction
PC = Address (HERE)
After Instruction
REG = REG + 1
If REG ≠ 0;
PC = Address (NZERO)
If REG = 0;
PC = Address (ZERO)

TBLRD Table Read

Syntax: TBLRD (*; *+; *-; +*)

Operands: None

Operation: if TBLRD *,
(Prog Mem (TBLPTR)) → TABLAT;
TBLPTR – No Change;
if TBLRD *+,
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) + 1 → TBLPTR;
if TBLRD *-,
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) – 1 → TBLPTR;
if TBLRD +*,
(TBLPTR) + 1 → TBLPTR;
(Prog Mem (TBLPTR)) → TABLAT;

Status Affected: None

Encoding:

0000	0000	0000	10nn nn=0 * =1 *+ =2 *- =3 +*

Description: This instruction is used to read the contents of Program Memory (P.M.). To address the program memory, a pointer called Table Pointer (TBLPTR) is used.
The TBLPTR (a 21-bit pointer) points to each byte in the program memory. TBLPTR has a 2-Mbyte address range.

TBLPTR[0] = 0: Least Significant Byte of Program Memory Word
TBLPTR[0] = 1: Most Significant Byte of Program Memory Word

The TBLRD instruction can modify the value of TBLPTR as follows:

- no change
- post-increment
- post-decrement
- pre-increment

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	No operation	No operation
No operation	No operation (Read Program Memory)	No operation	No operation (Write TABLAT)

TBLRD Table Read (Continued)

Example1:
```
TBLRD   *+ ;
```
Before Instruction
TABLAT = 55h
TBLPTR = 00A356h
MEMORY (00A356h) = 34h
After Instruction
TABLAT = 34h
TBLPTR = 00A357h

Example2:
```
TBLRD   +* ;
```
Before Instruction
TABLAT = AAh
TBLPTR = 01A357h
MEMORY (01A357h) = 12h
MEMORY (01A358h) = 34h
After Instruction
TABLAT = 34h
TBLPTR = 01A358h

FIGURE 28-17: PIC18LF2X/4XK22 DELTA IPD FVR

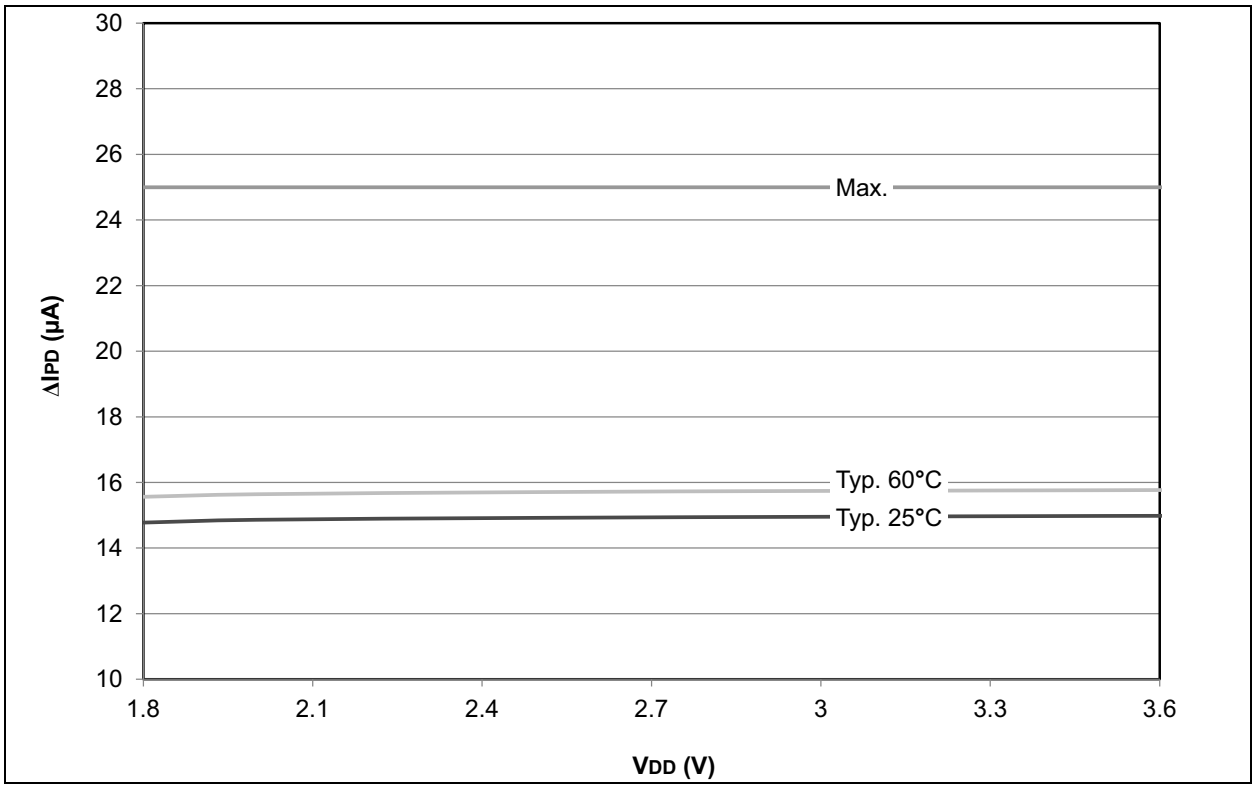
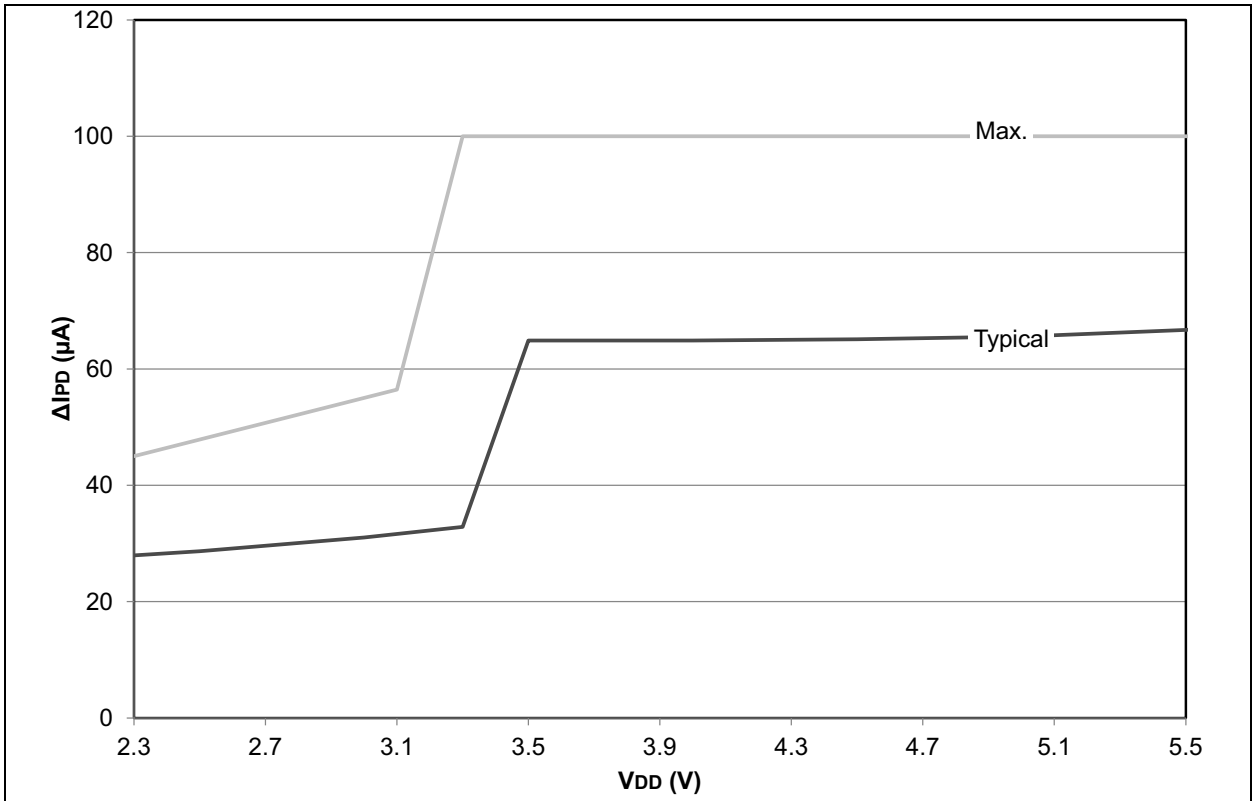


FIGURE 28-18: PIC18F2X/4XK22 DELTA IPD FVR



Note 1: On the PIC18F2X/4XK22, enabling the FVR results in significantly more Sleep current when the part enters Voltage Regulation mode at VDD ~ 3.2V.

FIGURE 28-20: PIC18LF2X/4XK22 TYPICAL IDD: RC_RUN LF-INTOSC 31 kHz

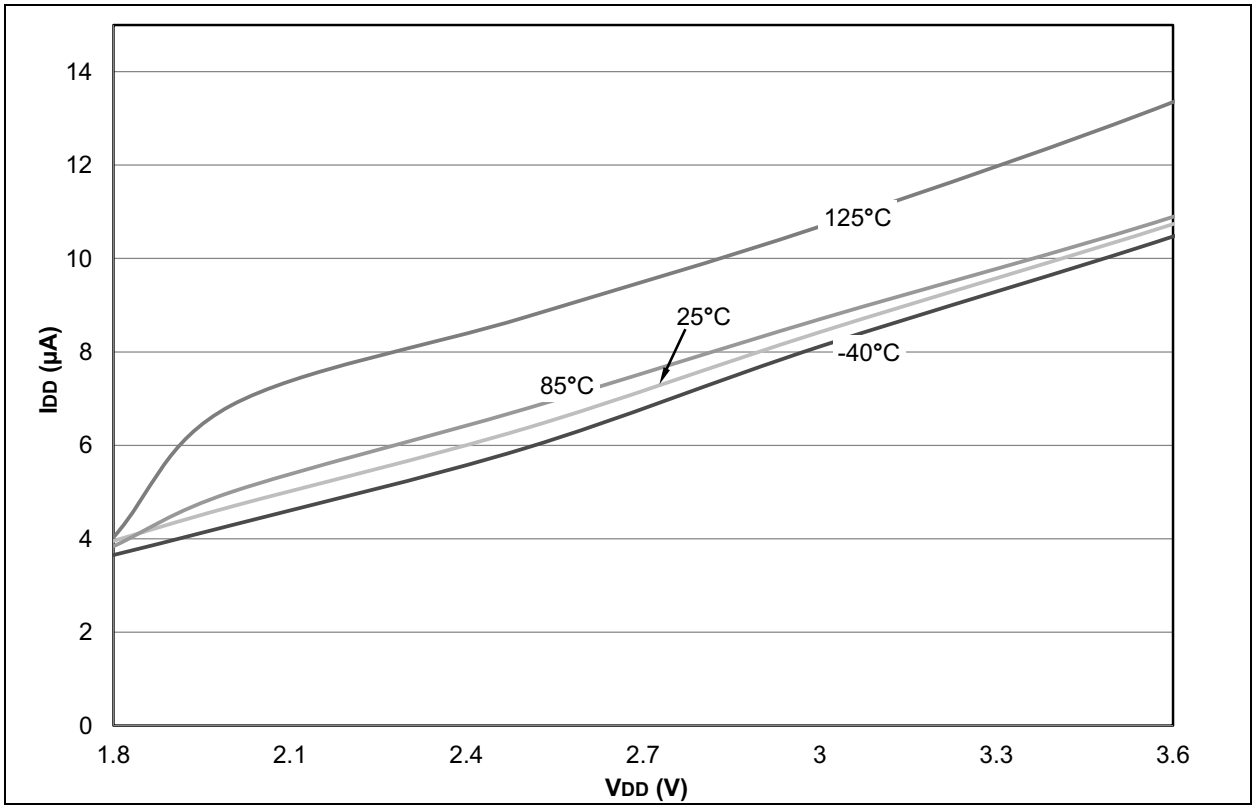


FIGURE 28-21: PIC18LF2X/4XK22 MAXIMUM IDD: RC_RUN LF-INTOSC 31 kHz

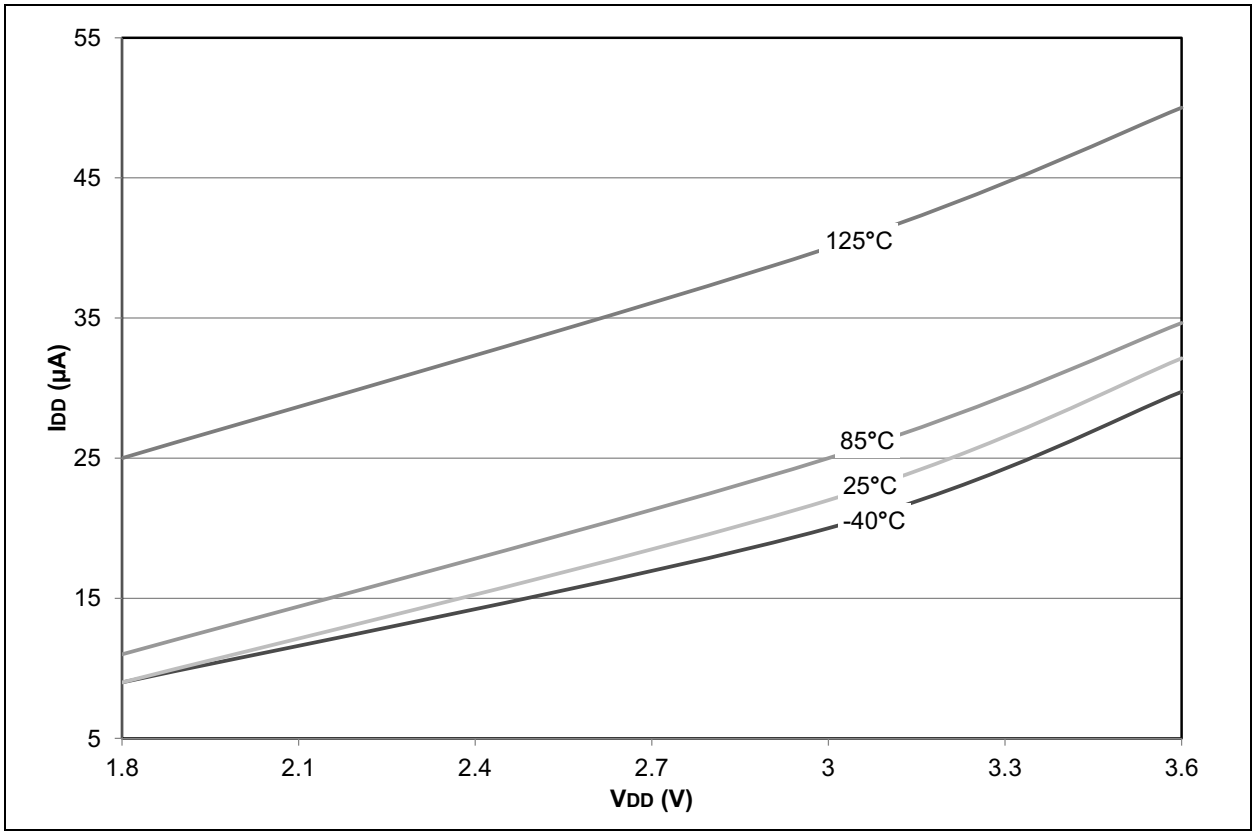


FIGURE 28-24: PIC18LF2X/4XK22 IDD: RC_RUN MF-INTOSC 500 kHz

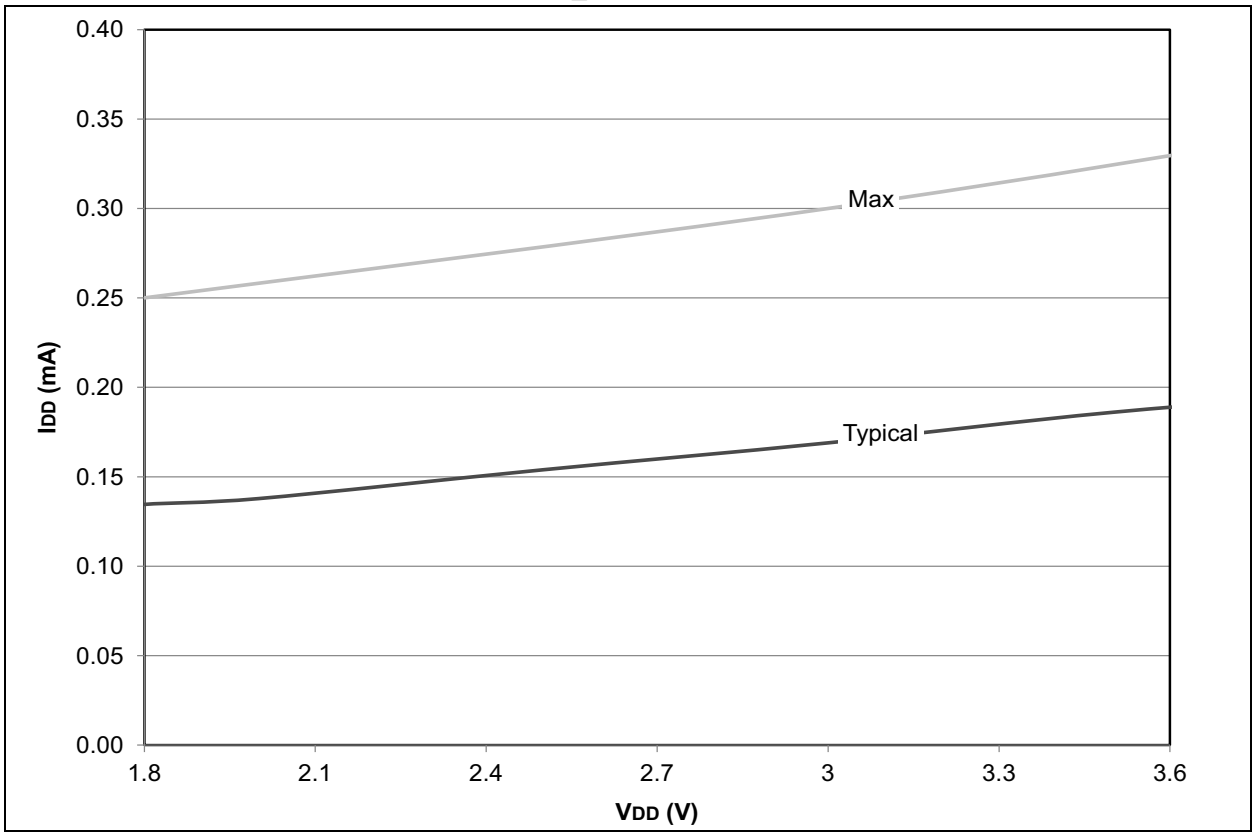


FIGURE 28-25: PIC18F2X/4XK22 IDD: RC_RUN MF-INTOSC 500 kHz

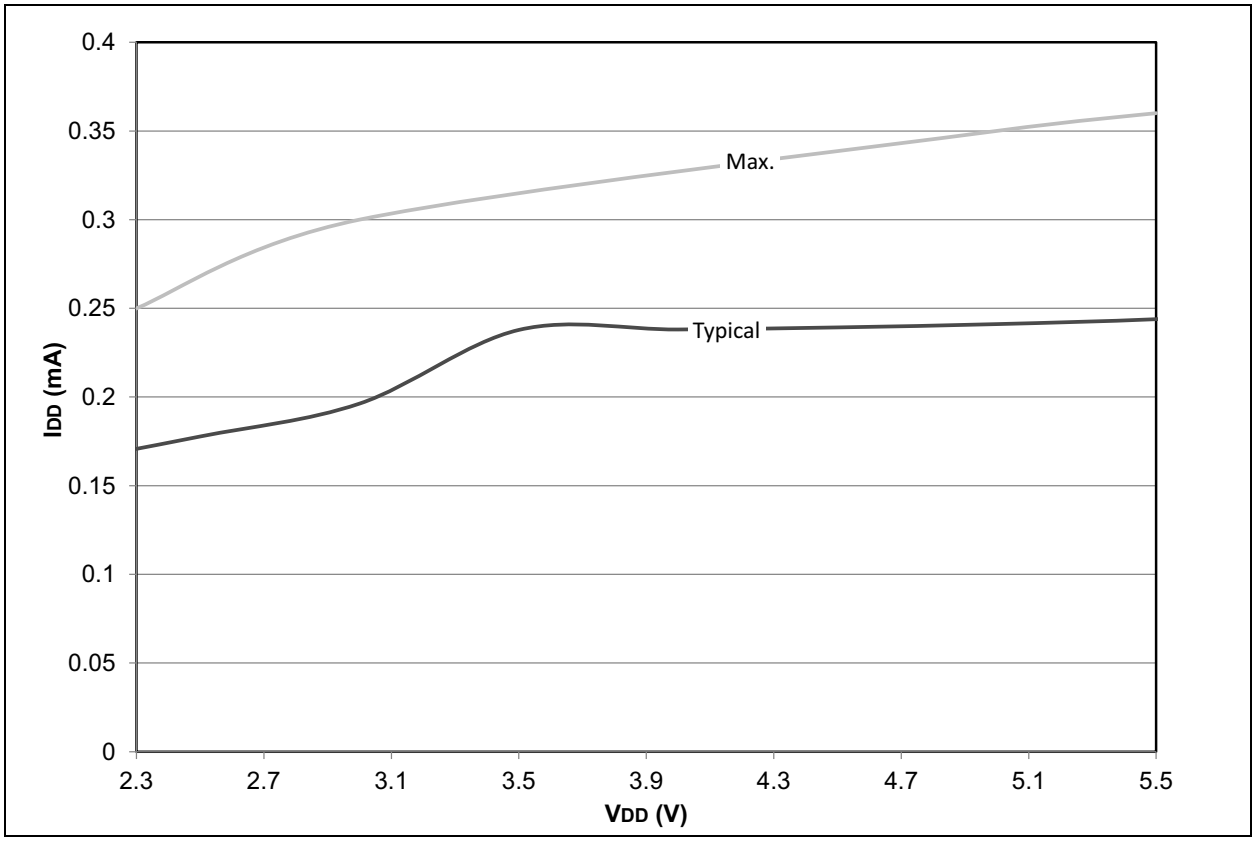


FIGURE 28-36: PIC18F2X/4XK22 TYPICAL IDD: RC_IDLE LF-INTOSC 31 kHz

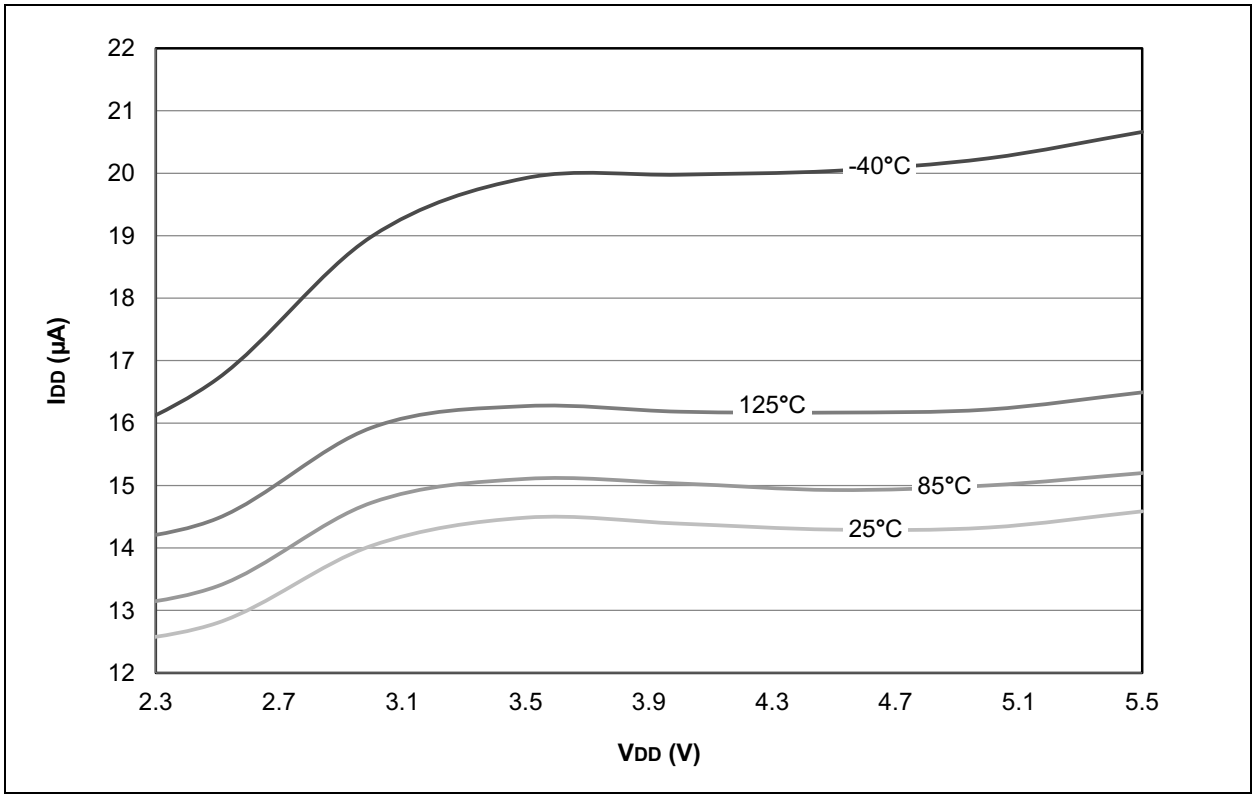


FIGURE 28-37: PIC18F2X/4XK22 MAXIMUM IDD: RC_IDLE LF-INTOSC 31 kHz

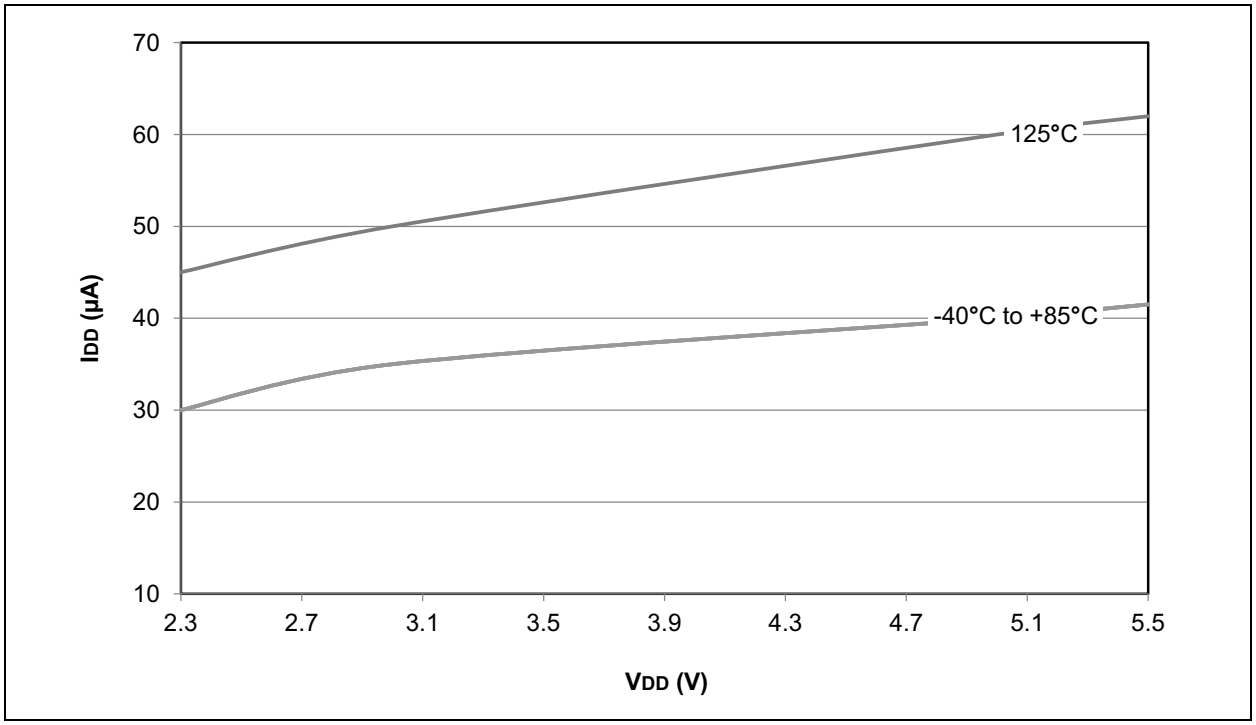


FIGURE 28-50: PIC18F2X/4XK22 TYPICAL IDD: PRI_RUN EC MEDIUM POWER

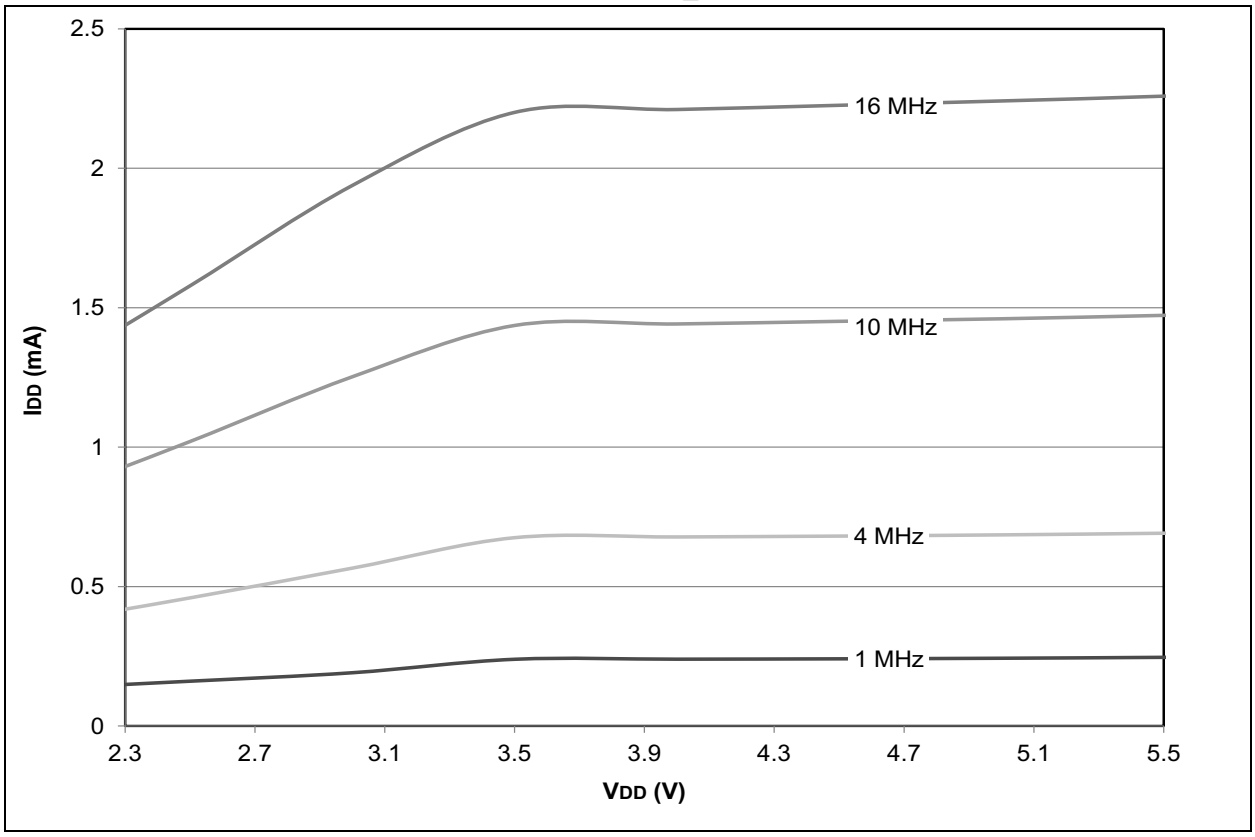


FIGURE 28-51: PIC18F2X/4XK22 MAXIMUM IDD: PRI_RUN EC MEDIUM POWER