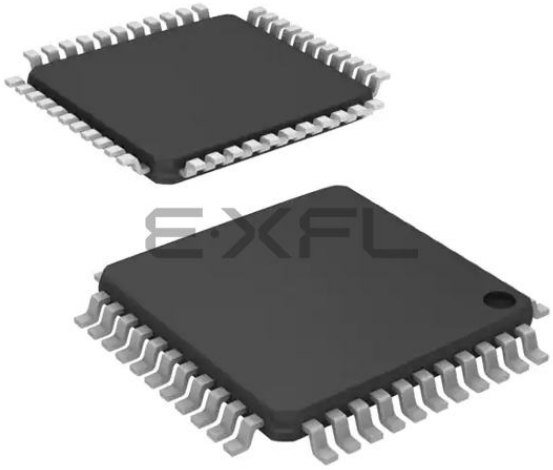

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Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf43k22-i-pt

2.2 Oscillator Control

The OSCCON, OSCCON2 and OSCTUNE registers (Register 2-1 to Register 2-3) control several aspects of the device clock's operation, both in full-power operation and in power-managed modes.

- Main System Clock Selection (SCS)
- Primary Oscillator Circuit Shutdown (PRISD)
- Secondary Oscillator Enable (SOSCGO)
- Primary Clock Frequency 4x multiplier (PLLEN)
- Internal Frequency selection bits (IRCF, INTSRC)
- Clock Status bits (OSTS, HFIOFS, MFIOFS, LFIOFS. SOSCRUN, PLLRDY)
- Power management selection (IDLEN)

2.2.1 MAIN SYSTEM CLOCK SELECTION

The System Clock Select bits, SCS<1:0>, select the main clock source. The available clock sources are

- Primary clock defined by the FOSC<3:0> bits of CONFIG1H. The primary clock can be the primary oscillator, an external clock, or the internal oscillator block.
- Secondary clock (secondary oscillator)
- Internal oscillator block (HFINTOSC, MFINTOSC and LFINTOSC).

The clock source changes immediately after one or more of the bits is written to, following a brief clock transition interval. The SCS bits are cleared to select the primary clock on all forms of Reset.

2.2.2 INTERNAL FREQUENCY SELECTION

The Internal Oscillator Frequency Select bits (IRCF<2:0>) select the frequency output of the internal oscillator block. The choices are the LFINTOSC source (31.25 kHz), the MFINTOSC source (31.25 kHz, 250 kHz or 500 kHz) and the HFINTOSC source (16 MHz) or one of the frequencies derived from the HFINTOSC postscaler (31.25 kHz to 8 MHz). If the internal oscillator block is supplying the main clock, changing the states of these bits will have an immediate change on the internal oscillator's output. On device Resets, the output frequency of the internal oscillator is set to the default frequency of 1 MHz.

2.2.3 LOW FREQUENCY SELECTION

When a nominal output frequency of 31.25 kHz is selected (IRCF<2:0> = 000), users may choose which internal oscillator acts as the source. This is done with the INTSRC bit of the OSCTUNE register and MFIOSEL bit of the OSCCON2 register. See Figure 2-2 and Register 2-1 for specific 31.25 kHz selection. This option allows users to select a 31.25 kHz clock (MFINTOSC or HFINTOSC) that can be tuned using the TUN<5:0> bits in OSCTUNE register, while maintaining power savings with a very low clock speed. LFINTOSC always remains the clock source for features such as the Watchdog Timer and the Fail-Safe Clock Monitor, regardless of the setting of INTSRC and MFIOSEL bits

This option allows users to select the tunable and more precise HFINTOSC as a clock source, while maintaining power savings with a very low clock speed.

2.2.4 POWER MANAGEMENT

The IDLEN bit of the OSCCON register determines whether the device goes into Sleep mode or one of the Idle modes when the SLEEP instruction is executed.

2.11.3 CLOCK SWITCH TIMING

When switching between one oscillator and another, the new oscillator may not be operating which saves power (see Figure 2-9). If this is the case, there is a delay after the SCS<1:0> bits of the OSCCON register are modified before the frequency change takes place. The OSTS and IOFS bits of the OSCCON register will reflect the current active status of the external and HFINTOSC oscillators. The timing of a frequency selection is as follows:

1. SCS<1:0> bits of the OSCCON register are modified.
2. The old clock continues to operate until the new clock is ready.
3. Clock switch circuitry waits for two consecutive rising edges of the old clock after the new clock ready signal goes true.
4. The system clock is held low starting at the next falling edge of the old clock.
5. Clock switch circuitry waits for an additional two rising edges of the new clock.
6. On the next falling edge of the new clock the low hold on the system clock is released and new clock is switched in as the system clock.
7. Clock switch is complete.

See Figure 2-1 for more details.

If the HFINTOSC is the source of both the old and new frequency, there is no start-up delay before the new frequency is active. This is because the old and new frequencies are derived from the HFINTOSC via the postscaler and multiplexer.

Start-up delay specifications are located in **Section 27.0 "Electrical Specifications"**, under AC Specifications (Oscillator Module).

2.12 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device.

This mode allows the application to wake-up from Sleep, perform a few instructions using the HFINTOSC as the clock source and go back to Sleep without waiting for the primary oscillator to become stable.

Note:	Executing a SLEEP instruction will abort the oscillator start-up time and will cause the OSTS bit of the OSCCON register to remain clear.

When the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) is enabled (see **Section 2.5.1 "Oscillator Start-up Timer (OST)"**). The OST will suspend program execution until 1024 oscillations are counted. Two-Speed Start-up mode minimizes the delay in code execution by operating from the internal oscillator as the OST is counting. When the OST count reaches 1024 and the OSTS bit of the OSCCON register is set, program execution switches to the external oscillator.

2.12.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is enabled when all of the following settings are configured as noted:

- Two-Speed Start-up mode is enabled when the IESO of the CONFIG1H Configuration register is set.
- SCS<1:0> (of the OSCCON register) = 00.
- FOSC<2:0> bits of the CONFIG1H Configuration register are configured for LP, XT or HS mode.

Two-Speed Start-up mode becomes active after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

TABLE 5-2: REGISTER FILE SUMMARY FOR PIC18(L)F2X/4XK22 DEVICES (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F68h	CCPR2H	Capture/Compare/PWM Register 2, High Byte								xxxx xxxx
F67h	CCPR2L	Capture/Compare/PWM Register 2, Low Byte								xxxx xxxx
F66h	CCP2CON	P2M<1:0>		DC2B<1:0>		CCP2M<3:0>				0000 0000
F65h	PWM2CON	P2RSEN	P2DC<6:0>							0000 0000
F64h	ECCP2AS	CCP2ASE	CCP2AS<2:0>			PSS2AC<1:0>		PSS2BD<1:0>		0000 0000
F63h	PSTR2CON	—	—	—	STR2SYNC	STR2D	STR2C	STR2B	STR2A	---0 0001
F62h	IOCB	IOCB7	IOCB6	IOCB5	IOCB4	—	—	—	—	1111 ----
F61h	WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111
F60h	SLRCON(2)	—	—	—	—	—	SLRC	SLRB	SLRA	---- -111
	SLRCON(1)	—	—	—	SLRE	SLRD	SLRC	SLRB	SLRA	---1 1111
F5Fh	CCPR3H	Capture/Compare/PWM Register 3, High Byte								xxxx xxxx
F5Eh	CCPR3L	Capture/Compare/PWM Register 3, Low Byte								xxxx xxxx
F5Dh	CCP3CON	P3M<1:0>		DC3B<1:0>		CCP3M<3:0>				0000 0000
F5Ch	PWM3CON	P3RSEN	P3DC<6:0>							0000 0000
F5Bh	ECCP3AS	CCP3ASE	CCP3AS<2:0>			PSS3AC<1:0>		PSS3BD<1:0>		0000 0000
F5Ah	PSTR3CON	—	—	—	STR3SYNC	STR3D	STR3C	STR3B	STR3A	---0 0001
F59h	CCPR4H	Capture/Compare/PWM Register 4, High Byte								xxxx xxxx
F58h	CCPR4L	Capture/Compare/PWM Register 4, Low Byte								xxxx xxxx
F57h	CCP4CON	—	—	DC4B<1:0>		CCP4M<3:0>				--00 0000
F56h	CCPR5H	Capture/Compare/PWM Register 5, High Byte								xxxx xxxx
F55h	CCPR5L	Capture/Compare/PWM Register 5, Low Byte								xxxx xxxx
F54h	CCP5CON	—	—	DC5B<1:0>		CCP5M<3:0>				--00 0000
F53h	TMR4	Timer4 Register								0000 0000
F52h	PR4	Timer4 Period Register								1111 1111
F51h	T4CON	—	T4OUTPS<3:0>				TMR4ON	T4CKPS<1:0>		-000 0000
F50h	TMR5H	Holding Register for the Most Significant Byte of the 16-bit TMR5 Register								0000 0000
F4Fh	TMR5L	Least Significant Byte of the 16-bit TMR5 Register								0000 0000
F4Eh	T5CON	TMR5CS<1:0>		T5CKPS<1:0>		T5SOSCEN	$\overline{\text{T5SYNC}}$	T5RD16	TMR5ON	0000 0000
F4Dh	T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/$\overline{\text{DONE}}$	T5GVAL	T5GSS<1:0>		0000 0x00
F4Ch	TMR6	Timer6 Register								0000 0000
F4Bh	PR6	Timer6 Period Register								1111 1111
F4Ah	T6CON	—	T6OUTPS<3:0>				TMR6ON	T6CKPS<1:0>		-000 0000
F49h	CCPTMRS0	C3TSEL<1:0>		—	C2TSEL<1:0>		—	C1TSEL<1:0>		00-0 0-00
F48h	CCPTMRS1	—	—	—	—	C5TSEL<1:0>		C4TSEL<1:0>		---- 0000
F47h	SRCON0	SRLEN	SRCLK<2:0>			SRQEN	SRNQEN	SRPS	SRPR	0000 0000
F46h	SRCON1	SRSPE	SRSCKE	SRSC2E	SRSC1E	SRRPE	SRRCKE	SRRC2E	SRRC1E	0000 0000
F45h	CTMUCONH	CTMUEN	—	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG	0000 0000
F44h	CTMUCONL	EDG2POL	EDG2SEL<1:0>		EDG1POL	EDG1SEL<1:0>		EDG2STAT	EDG1STAT	0000 0000
F43h	CTMUICON	ITRIM<5:0>						IRNG<1:0>		0000 0000
F42h	VREFCON0	FVREN	FVRST	FVRS<1:0>		—	—	—	—	0001 ----
F41h	VREFCON1	DACEN	DACLPS	DACOE	—	DACPSS<1:0>		—	DACNSS	000- 00-0
F40h	VREFCON2	—	—	—	DACR<4:0>					---0 0000
F3Fh	PMD0	UART2MD	UART1MD	TMR6MD	TMR5MD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	0000 0000
F3Eh	PMD1	MSSP2MD	MSSP1MD	—	CCP5MD	CCP4MD	CCP3MD	CCP2MD	CCP1MD	00-0 0000
F3Dh	PMD2	—	—	—	—	CTMUMD	CMP2MD	CMP1MD	ADCMD	---- 0000
F3Ch	ANSELE(1)	—	—	—	—	—	ANSE2	ANSE1	ANSE0	---- -111
F3Bh	ANSELD(1)	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	1111 1111

Legend: x = unknown, u = unchanged, — = unimplemented, q = value depends on condition

Note 1: PIC18(L)F4XK22 devices only.
2: PIC18(L)F2XK22 devices only.
3: PIC18(L)F23/24K22 and PIC18(L)F43/44K22 devices only.
4: PIC18(L)F26K22 and PIC18(L)F46K22 devices only.

REGISTER 12-2: TXGCON: TIMER1/3/5 GATE CONTROL REGISTER

R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W/HC-0/u	R-x/x	R/W-0/u	R/W-0/u
TMRxGE	TxGPOL	TxGTM	TxGSPM	TxGGO/$\overline{\text{DONE}}$	TxGVAL	TxGSS<1:0>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7 **TMRxGE:** Timer1/3/5 Gate Enable bit
If TMRxON = 0:
This bit is ignored
If TMRxON = 1:
1 = Timer1/3/5 counting is controlled by the Timer1/3/5 gate function
0 = Timer1/3/5 counts regardless of Timer1/3/5 gate function

bit 6 **TxGPOL:** Timer1/3/5 Gate Polarity bit
1 = Timer1/3/5 gate is active-high (Timer1/3/5 counts when gate is high)
0 = Timer1/3/5 gate is active-low (Timer1/3/5 counts when gate is low)

bit 5 **TxGTM:** Timer1/3/5 Gate Toggle Mode bit
1 = Timer1/3/5 Gate Toggle mode is enabled
0 = Timer1/3/5 Gate Toggle mode is disabled and toggle flip-flop is cleared
Timer1/3/5 gate flip-flop toggles on every rising edge.

bit 4 **TxGSPM:** Timer1/3/5 Gate Single-Pulse Mode bit
1 = Timer1/3/5 gate Single-Pulse mode is enabled and is controlling Timer1/3/5 gate
0 = Timer1/3/5 gate Single-Pulse mode is disabled

bit 3 **TxGGO/$\overline{\text{DONE}}$:** Timer1/3/5 Gate Single-Pulse Acquisition Status bit
1 = Timer1/3/5 gate single-pulse acquisition is ready, waiting for an edge
0 = Timer1/3/5 gate single-pulse acquisition has completed or has not been started
This bit is automatically cleared when TxGSPM is cleared.

bit 2 **TxGVAL:** Timer1/3/5 Gate Current State bit
Indicates the current state of the Timer1/3/5 gate that could be provided to TMRxH:TMRxL.
Unaffected by Timer1/3/5 Gate Enable (TMRxGE).

bit 1-0 **TxGSS<1:0>:** Timer1/3/5 Gate Source Select bits
00 = Timer1/3/5 Gate pin
01 = Timer2/4/6 Match PR2/4/6 output (See Table 12-5 for proper timer match selection)
10 = Comparator 1 optionally synchronized output (sync_C1OUT)
11 = Comparator 2 optionally synchronized output (sync_C2OUT)

13.1 Timer2/4/6 Operation

The clock input to the Timer2/4/6 module is the system instruction clock (FOSC/4).

TMRx increments from 00h on each clock edge.

A 4-bit counter/prescaler on the clock input allows direct input, divide-by-4 and divide-by-16 prescale options. These options are selected by the prescaler control bits, TxCKPS<1:0> of the TxCON register. The value of TMRx is compared to that of the Period register, PRx, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMRx to 00h on the next cycle and drives the output counter/postscaler (see **Section 13.2 "Timer2/4/6 Interrupt"**).

The TMRx and PRx registers are both directly readable and writable. The TMRx register is cleared on any device Reset, whereas the PRx register initializes to FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMRx register
- a write to the TxCON register
- Power-on Reset (POR)
- Brown-out Reset (BOR)
- $\overline{\text{MCLR}}$ Reset
- Watchdog Timer (WDT) Reset
- Stack Overflow Reset
- Stack Underflow Reset
- RESET Instruction

Note:	TMRx is not cleared when TxCON is written.

13.2 Timer2/4/6 Interrupt

Timer2/4/6 can also generate an optional device interrupt. The Timer2/4/6 output signal (TMRx-to-PRx match) provides the input for the 4-bit counter/postscaler. This counter generates the TMRx match interrupt flag which is latched in TMRxIF of the PIR1/PIR5 registers. The interrupt is enabled by setting the TMRx Match Interrupt Enable bit, TMRxIE of the PIE1/PIE5 registers. Interrupt Priority is selected with the TMRxIP bit in the IPR1/IPR5 registers.

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, TxOUTPS<3:0>, of the TxCON register.

13.3 Timer2/4/6 Output

The unscaled output of TMRx is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode. The timer to be used with a specific CCP module is selected using the CxTSEL<1:0> bits in the CCPTMRS0 and CCPTMRS1 registers.

Timer2 can be optionally used as the shift clock source for the MSSPx modules operating in SPI mode by setting SSPM<3:0> = 0011 in the SSPxCON1 register. Additional information is provided in **Section 15.0 "Master Synchronous Serial Port (MSSP1 and MSSP2) Module"**.

13.4 Timer2/4/6 Operation During Sleep

The Timer2/4/6 timers cannot be operated while the processor is in Sleep mode. The contents of the TMRx and PRx registers will remain unchanged while the processor is in Sleep mode.

13.5 Peripheral Module Disable

When a peripheral module is not used or inactive, the module can be disabled by setting the Module Disable bit in the PMD registers. This will reduce power consumption to an absolute minimum. Setting the PMD bits holds the module in Reset and disconnects the module's clock source. The Module Disable bits for Timer2 (TMR2MD), Timer4 (TMR4MD) and Timer6 (TMR6MD) are in the PMD0 Register. See **Section 3.0 "Power-Managed Modes"** for more information.

REGISTER 14-2: CCPxCON: ENHANCED CCPx CONTROL REGISTER

R/x-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxM<1:0>		DCxB<1:0>		CCPxM<3:0>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Reset
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 **PxM<1:0>**: Enhanced PWM Output Configuration bits

If CCPxM<3:2> = 00, 01, 10: (Capture/Compare modes)

xx = PxA assigned as Capture/Compare input; PxB, PxC, PxD assigned as port pins

Half-Bridge ECCP Modules(1):

If CCPxM<3:2> = 11: (PWM modes)

0x = Single output; PxA modulated; PxB assigned as port pin

1x = Half-Bridge output; PxA, PxB modulated with dead-band control

Full-Bridge ECCP Modules(1):

If CCPxM<3:2> = 11: (PWM modes)

00 = Single output; PxA modulated; PxB, PxC, PxD assigned as port pins

01 = Full-Bridge output forward; PxD modulated; PxA active; PxB, PxC inactive

10 = Half-Bridge output; PxA, PxB modulated with dead-band control; PxC, PxD assigned as port pins

11 = Full-Bridge output reverse; PxB modulated; PxC active; PxA, PxD inactive

bit 5-4 **DCxB<1:0>:** PWM Duty Cycle Least Significant bits

Capture mode:

Unused

Compare mode:

Unused

PWM mode:

These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPRxL.

Note 1: See Table 14-1 to determine full-bridge and half-bridge ECCPs for the device being used.

15.5.2 SLAVE RECEPTION

When the R/$\overline{W}$ bit of a matching received address byte is clear, the R/$\overline{W}$ bit of the SSPxSTAT register is cleared. The received address is loaded into the SSPxBUF register and acknowledged.

When the overflow condition exists for a received address, then not Acknowledge is given. An overflow condition is defined as either bit BF of the SSPxSTAT register is set, or bit SSPxOV of the SSPxCON1 register is set. The BOEN bit of the SSPxCON3 register modifies this operation. For more information see Register 15-5.

An MSSPx interrupt is generated for each transferred data byte. Flag bit, SSPxIF, must be cleared by software.

When the SEN bit of the SSPxCON2 register is set, SCLx will be held low (clock stretch) following each received byte. The clock must be released by setting the CKP bit of the SSPxCON1 register, except sometimes in 10-bit mode. See **Section 15.2.3 "SPI Master Mode"** for more detail.

15.5.2.1 7-bit Addressing Reception

This section describes a standard sequence of events for the MSSPx module configured as an I^2C slave in 7-bit Addressing mode. All decisions made by hardware or software and their effect on reception. Figure 15-14 and Figure 15-5 are used as a visual reference for this description.

This is a step by step process of what typically must be done to accomplish I^2C communication.

1. Start bit detected.
2. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
3. Matching address with R/$\overline{W}$ bit clear is received.
4. The slave pulls SDAx low sending an $\overline{ACK}$ to the master, and sets SSPxIF bit.
5. Software clears the SSPxIF bit.
6. Software reads received address from SSPxBUF clearing the BF flag.
7. If SEN = 1; Slave software sets CKP bit to release the SCLx line.
8. The master clocks out a data byte.
9. Slave drives SDAx low sending an $\overline{ACK}$ to the master, and sets SSPxIF bit.
10. Software clears SSPxIF.
11. Software reads the received byte from SSPxBUF clearing BF.
12. Steps 8-12 are repeated for all received bytes from the master.
13. Master sends Stop condition, setting P bit of SSPxSTAT, and the bus goes Idle.

15.5.2.2 7-bit Reception with AHEN and DHEN

Slave device reception with AHEN and DHEN set operate the same as without these options with extra interrupts and clock stretching added after the 8th falling edge of SCLx. These additional interrupts allow the slave software to decide whether it wants to $\overline{ACK}$ the receive address or data byte, rather than the hardware. This functionality adds support for PMBus™ that was not present on previous versions of this module.

This list describes the steps that need to be taken by slave software to use these options for I^2C communication. Figure 15-16 displays a module using both address and data holding. Figure 15-17 includes the operation with the SEN bit of the SSPxCON2 register set.

1. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
2. Matching address with R/$\overline{W}$ bit clear is clocked in. SSPxIF is set and CKP cleared after the 8th falling edge of SCLx.
3. Slave clears the SSPxIF.
4. Slave can look at the ACKTIM bit of the SSPxCON3 register to determine if the SSPxIF was after or before the $\overline{ACK}$.
5. Slave reads the address value from SSPxBUF, clearing the BF flag.
6. Slave sets $\overline{ACK}$ value clocked out to the master by setting ACKDT.
7. Slave releases the clock by setting CKP.
8. SSPxIF is set after an $\overline{ACK}$, not after a NACK.
9. If SEN = 1 the slave hardware will stretch the clock after the $\overline{ACK}$.
10. Slave clears SSPxIF

.

Note:	SSPxIF is still set after the 9th falling edge of SCLx even if there is no clock stretching and BF has been cleared. Only if NACK is sent to master is SSPxIF not set.

11. SSPxIF set and CKP cleared after 8th falling edge of SCLx for a received data byte.
12. Slave looks at ACKTIM bit of SSPxCON3 to determine the source of the interrupt.
13. Slave reads the received data from SSPxBUF clearing BF.
14. Steps 7-14 are the same for each received data byte.
15. Communication is ended by either the slave sending an $\overline{ACK}$ = 1, or the master sending a Stop condition. If a Stop is sent and Interrupt on Stop detect is disabled, the slave will only know by polling the P bit of the SSTSTAT register.

15.5.3.3 7-bit Transmission with Address Hold Enabled

Setting the AHEN bit of the SSPxCON3 register enables additional clock stretching and interrupt generation after the 8th falling edge of a received matching address. Once a matching address has been clocked in, CKP is cleared and the SSPxIF interrupt is set.

Figure 15-19 displays a standard waveform of a 7-bit Address Slave Transmission with AHEN enabled.

1. Bus starts Idle.
2. Master sends Start condition; the S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
3. Master sends matching address with R/$\overline{W}$ bit set. After the 8th falling edge of the SCLx line the CKP bit is cleared and SSPxIF interrupt is generated.
4. Slave software clears SSPxIF.
5. Slave software reads ACKTIM bit of SSPxCON3 register, and R/$\overline{W}$ and D/$\overline{A}$ of the SSPxSTAT register to determine the source of the interrupt.
6. Slave reads the address value from the SSPxBUF register clearing the BF bit.
7. Slave software decides from this information if it wishes to $\overline{ACK}$ or not $\overline{ACK}$ and sets ACKDT bit of the SSPxCON2 register accordingly.
8. Slave sets the CKP bit releasing SCLx.
9. Master clocks in the $\overline{ACK}$ value from the slave.
10. Slave hardware automatically clears the CKP bit and sets SSPxIF after the $\overline{ACK}$ if the R/$\overline{W}$ bit is set.
11. Slave software clears SSPxIF.
12. Slave loads value to transmit to the master into SSPxBUF setting the BF bit.

Note: SSPxBUF cannot be loaded until after the $\overline{ACK}$.

13. Slave sets CKP bit releasing the clock.
14. Master clocks out the data from the slave and sends an $\overline{ACK}$ value on the 9th SCLx pulse.
15. Slave hardware copies the $\overline{ACK}$ value into the ACKSTAT bit of the SSPxCON2 register.
16. Steps 10-15 are repeated for each byte transmitted to the master from the slave.
17. If the master sends a not $\overline{ACK}$ the slave releases the bus allowing the master to send a Stop and end the communication.

Note: Master must send a not $\overline{ACK}$ on the last byte to ensure that the slave releases the SCLx line to receive a Stop.

15.6.5 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition (Figure 15-27) occurs when the RSEN bit of the SSPxCON2 register is programmed high and the master state machine is no longer active. When the RSEN bit is set, the SCLx pin is asserted low. When the SCLx pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDAx pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDAx is sampled high, the SCLx pin will be deasserted (brought high). When SCLx is sampled high, the Baud Rate Generator is reloaded and begins counting. SDAx and SCLx must be sampled high for one TBRG. This action is then followed by assertion of the SDAx pin (SDAx = 0) for one TBRG while SCLx is high. SCLx is asserted low.

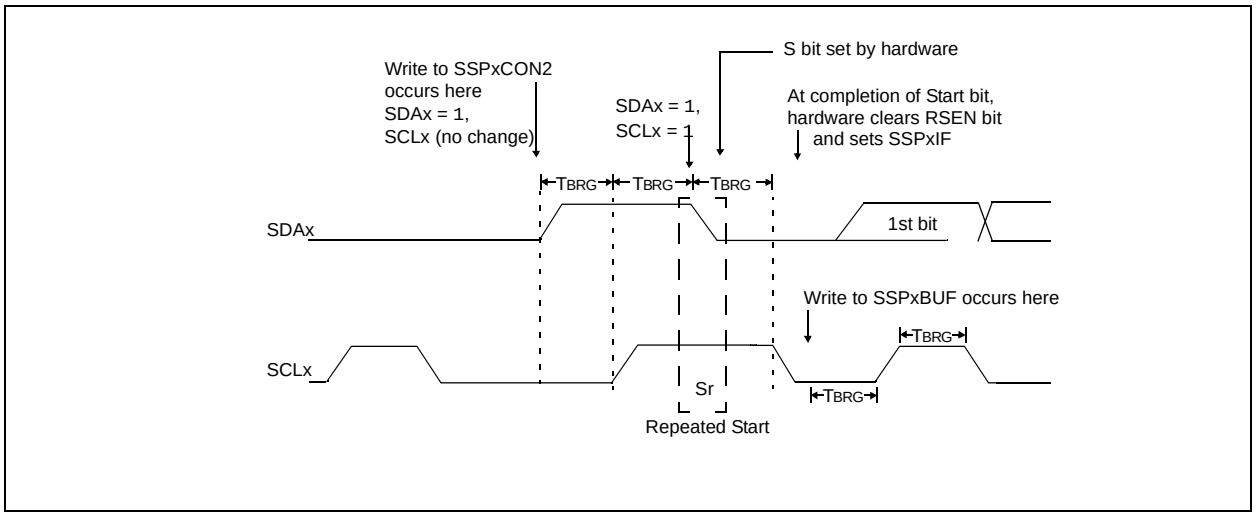
Following this, the RSEN bit of the SSPxCON2 register will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDAx pin held low. As soon as a Start condition is detected on the SDAx and SCLx pins, the S bit of the SSPxSTAT register will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.

2: A bus collision during the Repeated Start condition occurs if:

- SDAx is sampled low when SCLx goes from low-to-high.
- SCLx goes low before SDAx is asserted low. This may indicate that another master is attempting to transmit a data '1'.

FIGURE 15-27: REPEAT START CONDITION WAVEFORM



15.6.13 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDAx pin, arbitration takes place when the master outputs a '1' on SDAx, by letting SDAx float high and another master asserts a '0'. When the SCLx pin floats high, data should be stable. If the expected data on SDAx is a '1' and the data sampled on the SDAx pin is '0', then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLxIF, and reset the I^2C port to its Idle state (Figure 15-32).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDAx and SCLx lines are deasserted and the SSPxBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

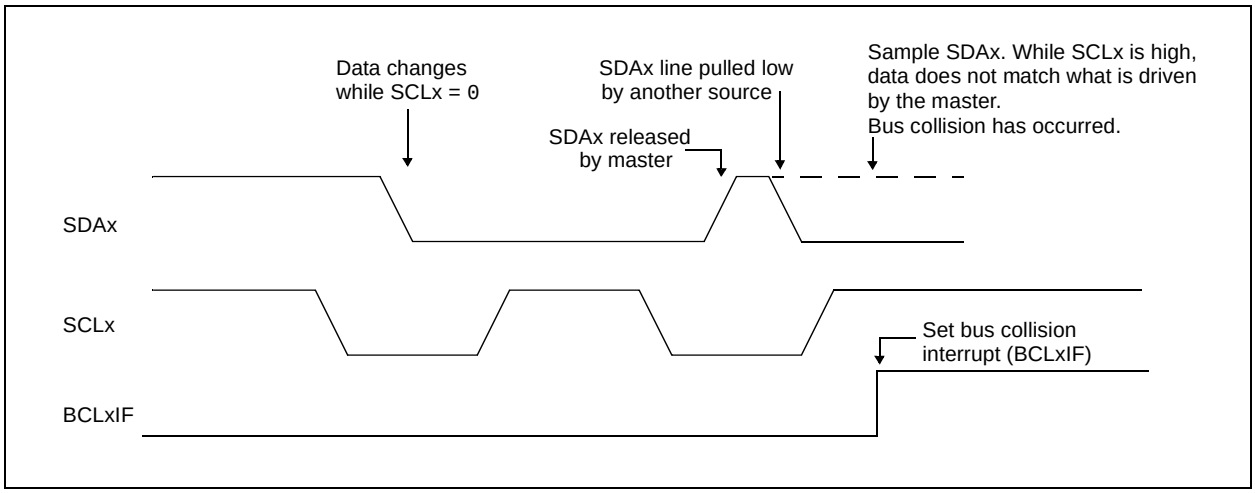
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDAx and SCLx lines are deasserted and the respective control bits in the SSPxCON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDAx and SCLx pins. If a Stop condition occurs, the SSPxIF bit will be set.

A write to the SSPxBUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPxSTAT register, or the bus is Idle and the S and P bits are cleared.

FIGURE 15-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



REGISTER 15-5: SSPxCON3: SSPx CONTROL REGISTER 3

R-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **ACKTIM:** Acknowledge Time Status bit (I^2C mode only)**(3)**

1 = Indicates the I^2C bus is in an Acknowledge sequence, set on 8th falling edge of SCLx clock
0 = Not an Acknowledge sequence, cleared on 9th rising edge of SCLx clock

bit 6 **PCIE**: Stop Condition Interrupt Enable bit (I^2C mode only)

1 = Enable interrupt on detection of Stop condition
0 = Stop detection interrupts are disabled**(2)**

bit 5 **SCIE**: Start Condition Interrupt Enable bit (I^2C mode only)

1 = Enable interrupt on detection of Start or Restart conditions
0 = Start detection interrupts are disabled**(2)**

bit 4 **BOEN:** Buffer Overwrite Enable bit

In SPI Slave mode:**(1)**

1 = SSPxBUF updates every time that a new data byte is shifted in ignoring the BF bit
0 = If new byte is received with BF bit of the SSPxSTAT register already set, SSPxOV bit of the SSPxCON1 register is set, and the buffer is not updated

In I^2C Master mode:

This bit is ignored.

In I^2C Slave mode:

1 = SSPxBUF is updated and $\overline{ACK}$ is generated for a received address/data byte, ignoring the state of the SSPxOV bit only if the BF bit = 0.
0 = SSPxBUF is only updated when SSPxOV is clear

bit 3 **SDAHT:** SDAx Hold Time Selection bit (I^2C mode only)

1 = Minimum of 300 ns hold time on SDAx after the falling edge of SCLx
0 = Minimum of 100 ns hold time on SDAx after the falling edge of SCLx

bit 2 **SBCDE:** Slave Mode Bus Collision Detect Enable bit (I^2C Slave mode only)

If on the rising edge of SCLx, SDAx is sampled low when the module is outputting a high state, the BCLxIF bit of the PIR2 register is set, and bus goes idle

1 = Enable slave bus collision interrupts
0 = Slave bus collision interrupts are disabled

bit 1 **AHEN:** Address Hold Enable bit (I^2C Slave mode only)

1 = Following the 8th falling edge of SCLx for a matching received address byte; CKP bit of the SSPxCON1 register will be cleared and the SCLx will be held low.
0 = Address holding is disabled

Note 1: For daisy-chained SPI operation; allows the user to ignore all but the last received byte. SSPxOV is still set when a new byte is received and BF = 1, but hardware continues to write the most recent byte to SSPxBUF.

2: This bit has no effect in Slave modes for which Start and Stop condition detection is explicitly listed as enabled.

3: The ACKTIM Status bit is active only when the AHEN bit or DHEN bit is set.

REGISTER 15-5: SSPxCON3: SSPx CONTROL REGISTER 3 (CONTINUED)

bit 0 **DHEN:** Data Hold Enable bit (I^2C Slave mode only)
1 = Following the 8th falling edge of SCLx for a received data byte; slave hardware clears the CKP bit of the SSPxCON1 register and SCLx is held low.
0 = Data holding is disabled

Note 1: For daisy-chained SPI operation; allows the user to ignore all but the last received byte. SSPxOV is still set when a new byte is received and BF = 1, but hardware continues to write the most recent byte to SSPxBUF.
2: This bit has no effect in Slave modes for which Start and Stop condition detection is explicitly listed as enabled.
3: The ACKTIM Status bit is active only when the AHEN bit or DHEN bit is set.

REGISTER 15-6: SSPxMSK: SSPx MASK REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-1 **MSK<7:1>:** Mask bits
1 = The received address bit n is compared to SSPxADD<n> to detect I^2C address match
0 = The received address bit n is not used to detect I^2C address match

bit 0 **MSK<0>:** Mask bit for I^2C Slave mode, 10-bit Address
I^2C Slave mode, 10-bit address (SSPxM<3:0> = 0111 or 1111):
1 = The received address bit 0 is compared to SSPxADD<0> to detect I^2C address match
0 = The received address bit 0 is not used to detect I^2C address match
I^2C Slave mode, 7-bit address, the bit is ignored

16.4.2 AUTO-BAUD OVERFLOW

During the course of automatic baud detection, the ABDOVF bit of the BAUDCONx register will be set if the baud rate counter overflows before the fifth rising edge is detected on the RX pin. The ABDOVF bit indicates that the counter has exceeded the maximum count that can fit in the 16 bits of the SPBRGHx:SPBRGx register pair. After the ABDOVF has been set, the counter continues to count until the fifth rising edge is detected on the RXx/DTx pin. Upon detecting the fifth RXx/DTx edge, the hardware will set the RCxIF interrupt flag and clear the ABDEN bit of the BAUDCONx register. The RCxIF flag can be subsequently cleared by reading the RCREGx. The ABDOVF flag can be cleared by software directly.

To terminate the auto-baud process before the RCxIF flag is set, clear the ABDEN bit then clear the ABDOVF bit. The ABDOVF bit will remain set if the ABDEN bit is not cleared first.

16.4.3 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RXx/DTx line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUDCONx register. Once set, the normal receive sequence on RXx/DTx is disabled, and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RXx/DTx line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCxIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 16-7), and asynchronously if the device is in Sleep mode (Figure 16-8). The interrupt condition is cleared by reading the RCREGx register.

The WUE bit is automatically cleared by the low-to-high transition on the RXx line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

16.4.3.1 Special Considerations

Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all ‘0’s. This must be 10 or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

Oscillator Start-up Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

WUE Bit

The wake-up event causes a receive interrupt by setting the RCxIF bit. The WUE bit is cleared by hardware by a rising edge on RXx/DTx. The interrupt condition is then cleared by software by reading the RCREGx register and discarding its contents.

To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

CPFSGT **Compare f with W, skip if f > W**

Syntax: CPFSGT f {,a}

Operands: $0 \le f \le 255$
$a \in [0,1]$

Operation: (f) – (W),
skip if (f) > (W)
(unsigned comparison)

Status Affected: None

Encoding:

0110	010a	ffff	ffff

Description: Compares the contents of data memory location 'f' to the contents of the W by performing an unsigned subtraction.
If the contents of 'f' are greater than the contents of WREG, then the fetched instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	No operation

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE      CPFSGT REG, 0
NGREATER  :
GREATER   :
```

Before Instruction
PC = Address (HERE)
W = ?
After Instruction
If REG > W;
PC = Address (GREATER)
If REG ≤ W;
PC = Address (NGREATER)

CPFSLT **Compare f with W, skip if f < W**

Syntax: CPFSLT f {,a}

Operands: $0 \le f \le 255$
$a \in [0,1]$

Operation: (f) – (W),
skip if (f) < (W)
(unsigned comparison)

Status Affected: None

Encoding:

0110	000a	ffff	ffff

Description: Compares the contents of data memory location 'f' to the contents of W by performing an unsigned subtraction.
If the contents of 'f' are less than the contents of W, then the fetched instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	No operation

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    CPFSLT REG, 1
NLESS   :
LESS    :
```

Before Instruction
PC = Address (HERE)
W = ?
After Instruction
If REG < W;
PC = Address (LESS)
If REG ≥ W;
PC = Address (NLESS)

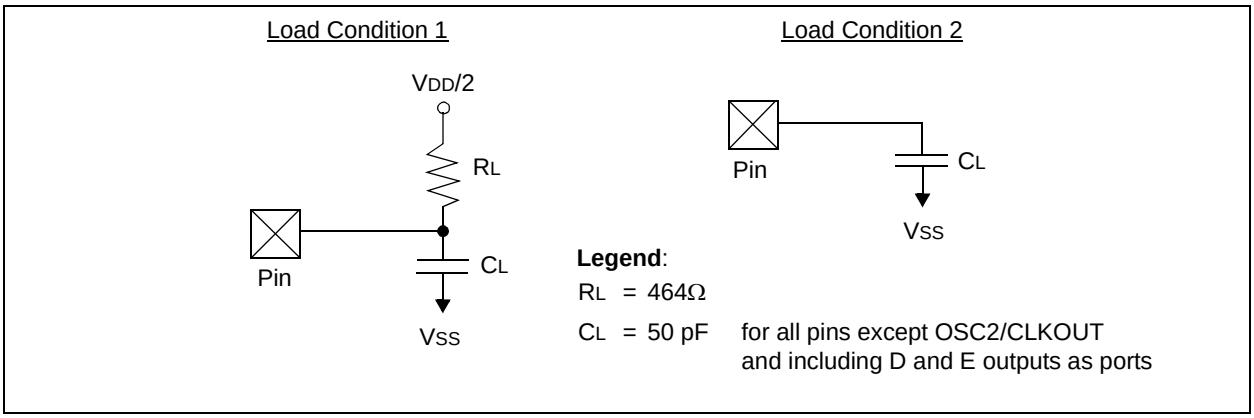
27.11.2 TIMING CONDITIONS

The temperature and voltages specified in Table 27-6 apply to all timing specifications unless otherwise noted. Figure 27-6 specifies the load conditions for the timing specifications.

TABLE 27-6: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

AC CHARACTERISTICS	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +125°C Operating voltage VDD range as described in **Section 27.1 "DC Characteristics: Supply Voltage, PIC18(L)F2X/4XK22"** and **Section 27.9 "Memory Programming Requirements"**.

FIGURE 27-6: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS



27.11.3 TIMING DIAGRAMS AND SPECIFICATIONS

FIGURE 27-7: EXTERNAL CLOCK TIMING (ALL MODES EXCEPT PLL)

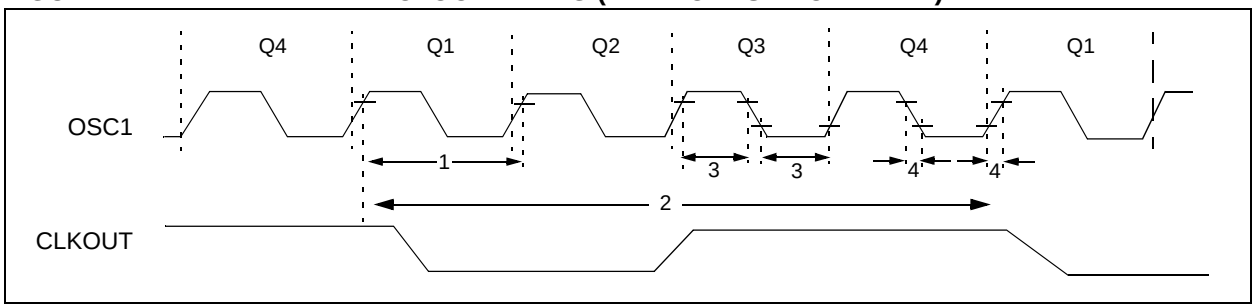


TABLE 27-7: EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
1A	FOSC	External CLKIN Frequency(1)	DC	0.5	MHz	EC, ECIO Oscillator mode (low power)
			DC	16	MHz	EC, ECIO Oscillator mode (medium power)
			DC	64	MHz	EC, ECIO Oscillator mode (high power)
		Oscillator Frequency(1)	DC	4	MHz	RC Oscillator mode
			5	200	kHz	LP Oscillator mode
			0.1	4	MHz	XT Oscillator mode
			4	4	MHz	HS Oscillator mode, VDD < 2.7V
			4	16	MHz	HS Oscillator mode, VDD ≥ 2.7V, Medium-Power mode (HSMP)
			4	20	MHz	HS Oscillator mode, VDD ≥ 2.7V, High-Power mode (HSHP)
1	TOSC	External CLKIN Period(1)	2.0	—	µs	EC, ECIO Oscillator mode (low power)
			62.5	—	ns	EC, ECIO Oscillator mode (medium power)
			15.6	—	ns	EC, ECIO Oscillator mode (high power)
		Oscillator Period(1)	250	—	ns	RC Oscillator mode
			5	200	µs	LP Oscillator mode
			0.25	10	µs	XT Oscillator mode
			250	250	ns	HS Oscillator mode, VDD < 2.7V
			62.5	250	ns	HS Oscillator mode, VDD ≥ 2.7V, Medium-Power mode (HSMP)
			50	250	ns	HS Oscillator mode, VDD ≥ 2.7V, High-Power mode (HSHP)
2	TCY	Instruction Cycle Time(1)	62.5	—	ns	TCY = 4/FOSC
3	TOSL, TOSH	External Clock in (OSC1) High or Low Time	2.5	—	µs	LP Oscillator mode
			30	—	ns	XT Oscillator mode
			10	—	ns	HS Oscillator mode
4	TOSR, TOSF	External Clock in (OSC1) Rise or Fall Time	—	50	ns	LP Oscillator mode
			—	20	ns	XT Oscillator mode
			—	7.5	ns	HS Oscillator mode

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min.” values with an external clock applied to the OSC1/CLKIN pin. When an external clock input is used, the “max.” cycle time limit is “DC” (no clock) for all devices.

FIGURE 28-1: PIC18LF2X/4XK22 BASE IPD

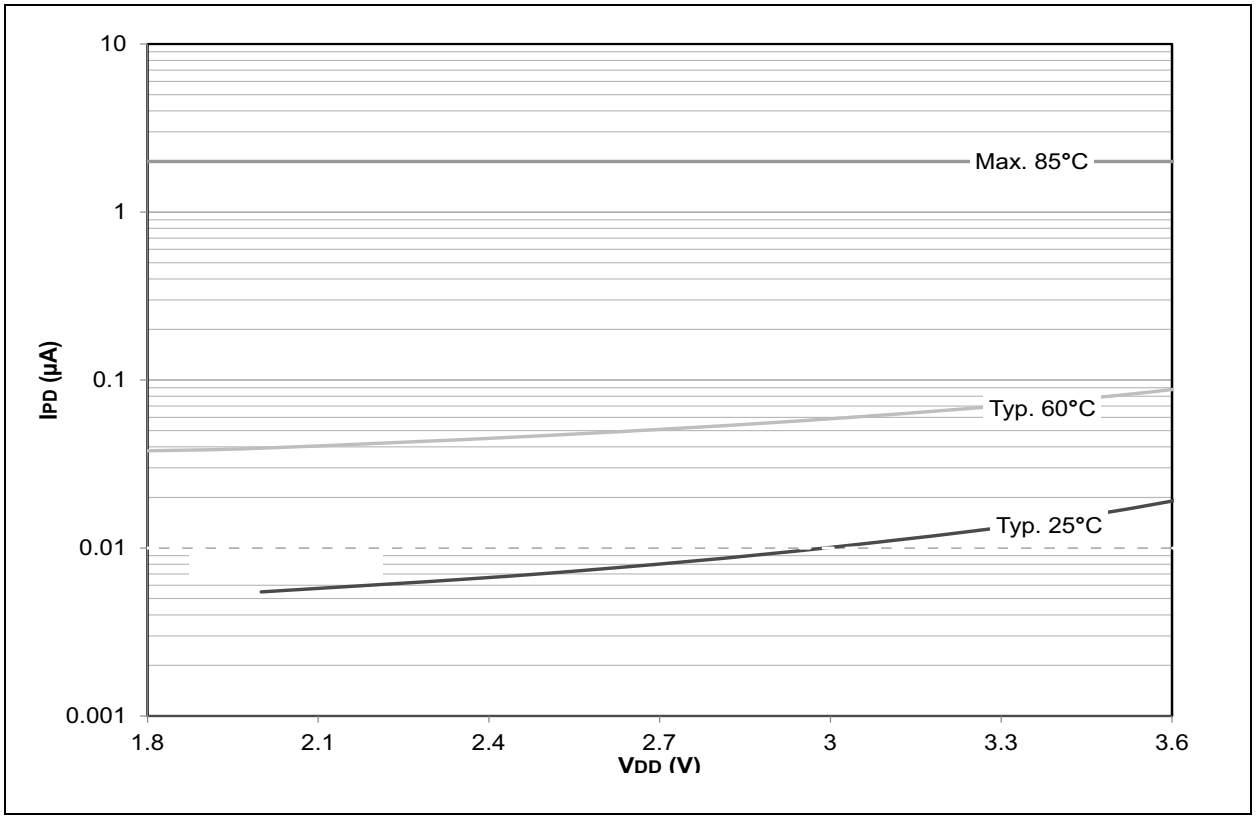


FIGURE 28-2: PIC18F2X/4XK22 BASE IPD

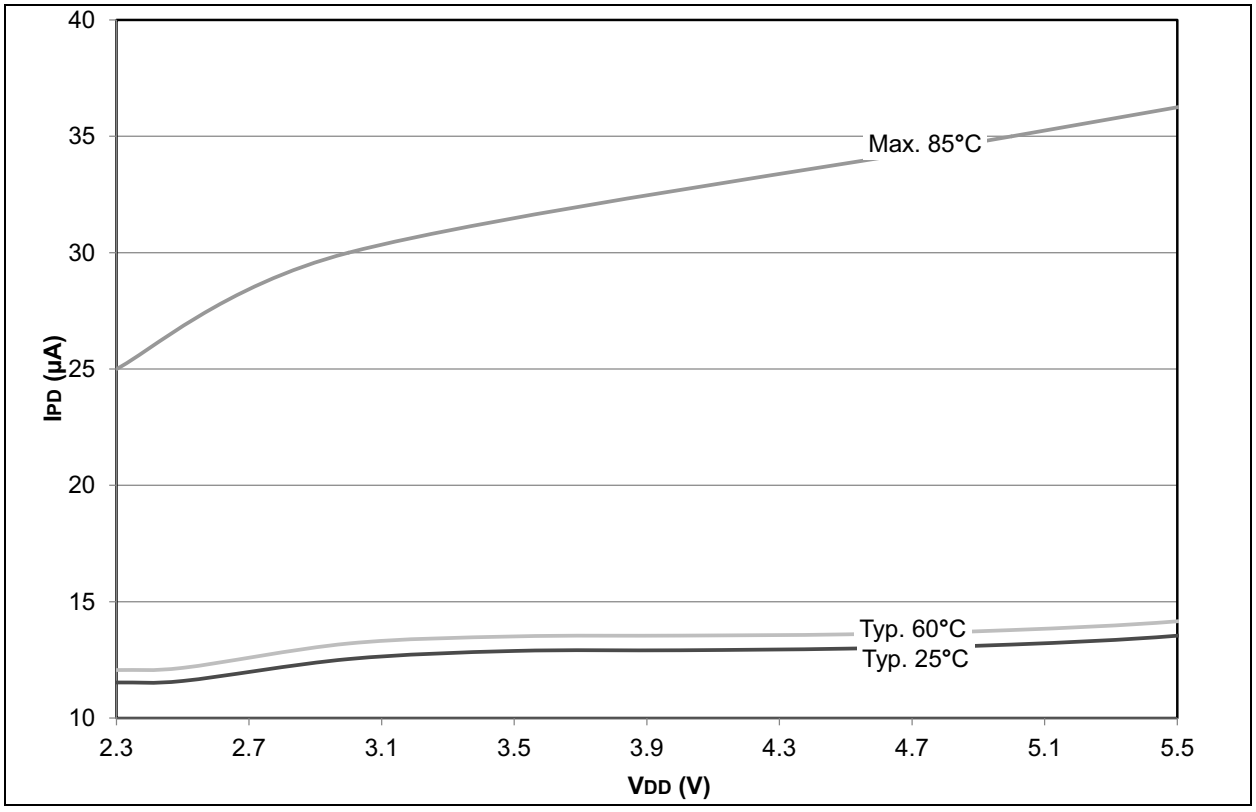


FIGURE 28-17: PIC18LF2X/4XK22 DELTA IPD FVR

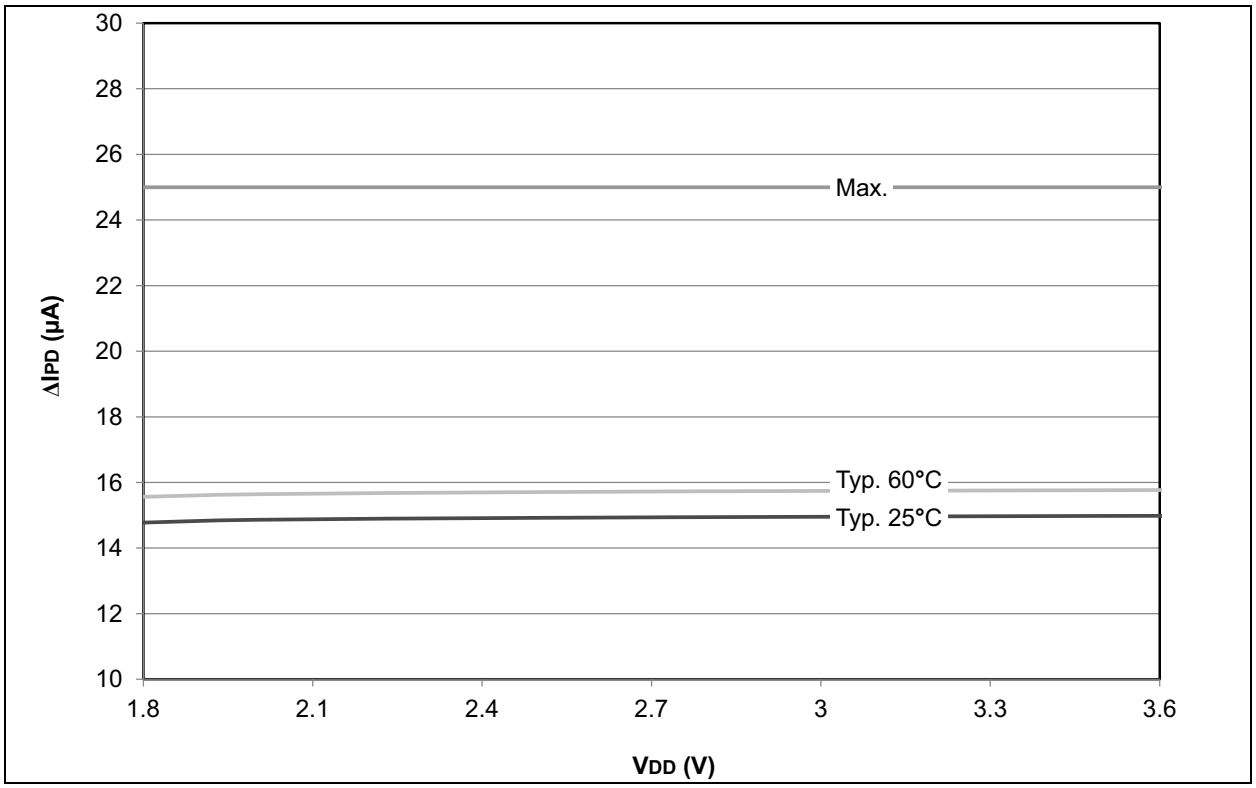
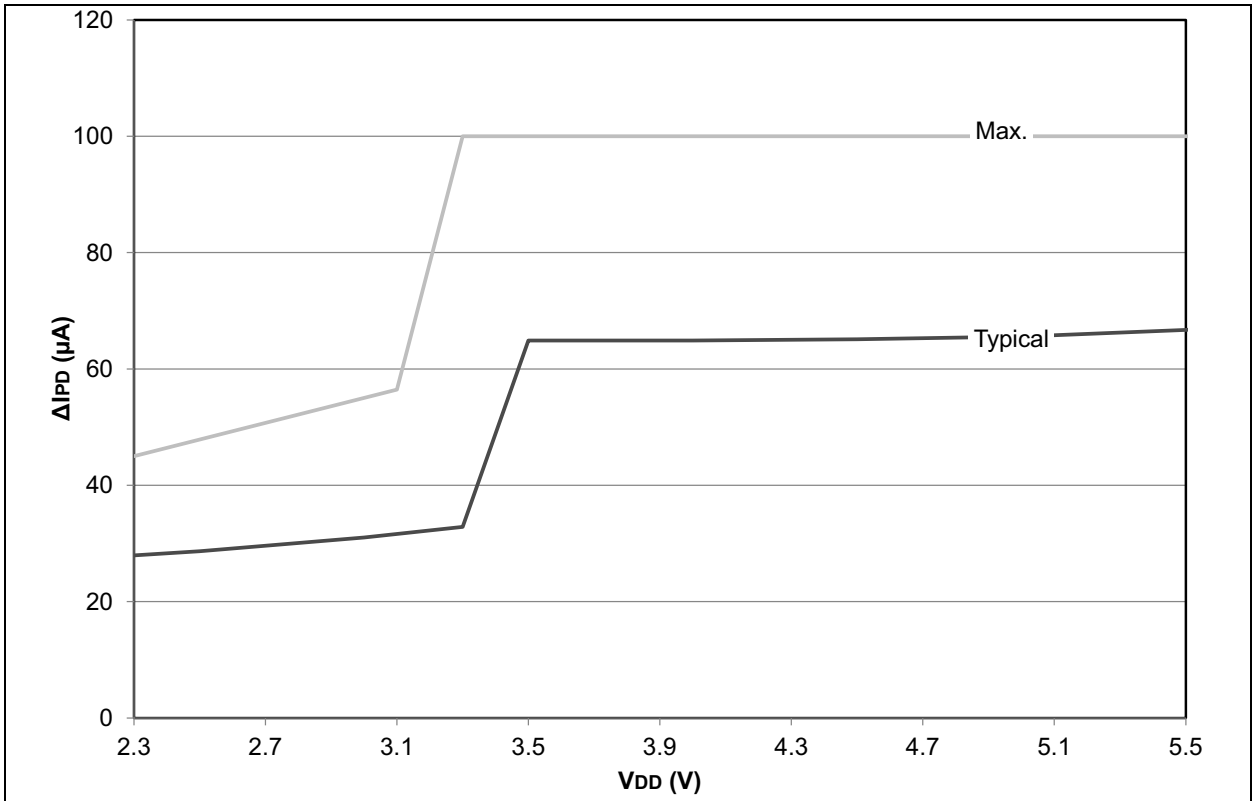


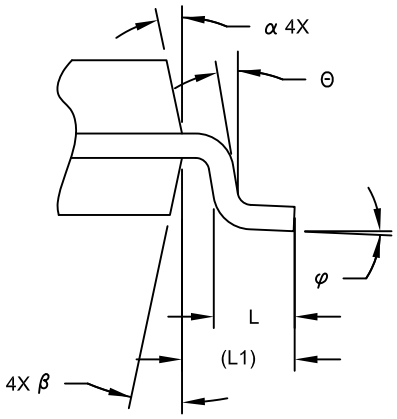
FIGURE 28-18: PIC18F2X/4XK22 DELTA IPD FVR



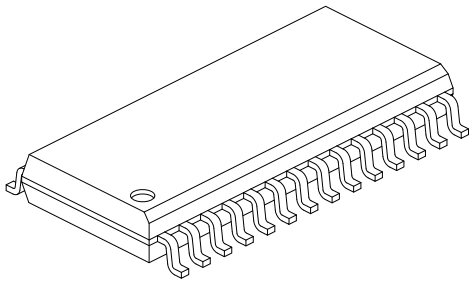
Note 1: On the PIC18F2X/4XK22, enabling the FVR results in significantly more Sleep current when the part enters Voltage Regulation mode at VDD ~ 3.2V.

28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



VIEW C



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		28	
Pitch	e		1.27 BSC	
Overall Height	A	-	-	2.65
Molded Package Thickness	A2	2.05	-	-
Standoff §	A1	0.10	-	0.30
Overall Width	E		10.30 BSC	
Molded Package Width	E1		7.50 BSC	
Overall Length	D		17.90 BSC	
Chamfer (Optional)	h	0.25	-	0.75
Foot Length	L	0.40	-	1.27
Footprint	L1		1.40 REF	
Lead Angle	Θ	0°	-	-
Foot Angle	φ	0°	-	8°
Lead Thickness	c	0.18	-	0.33
Lead Width	b	0.31	-	0.51
Mold Draft Angle Top	α	5°	-	15°
Mold Draft Angle Bottom	β	5°	-	15°

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic
3. Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M

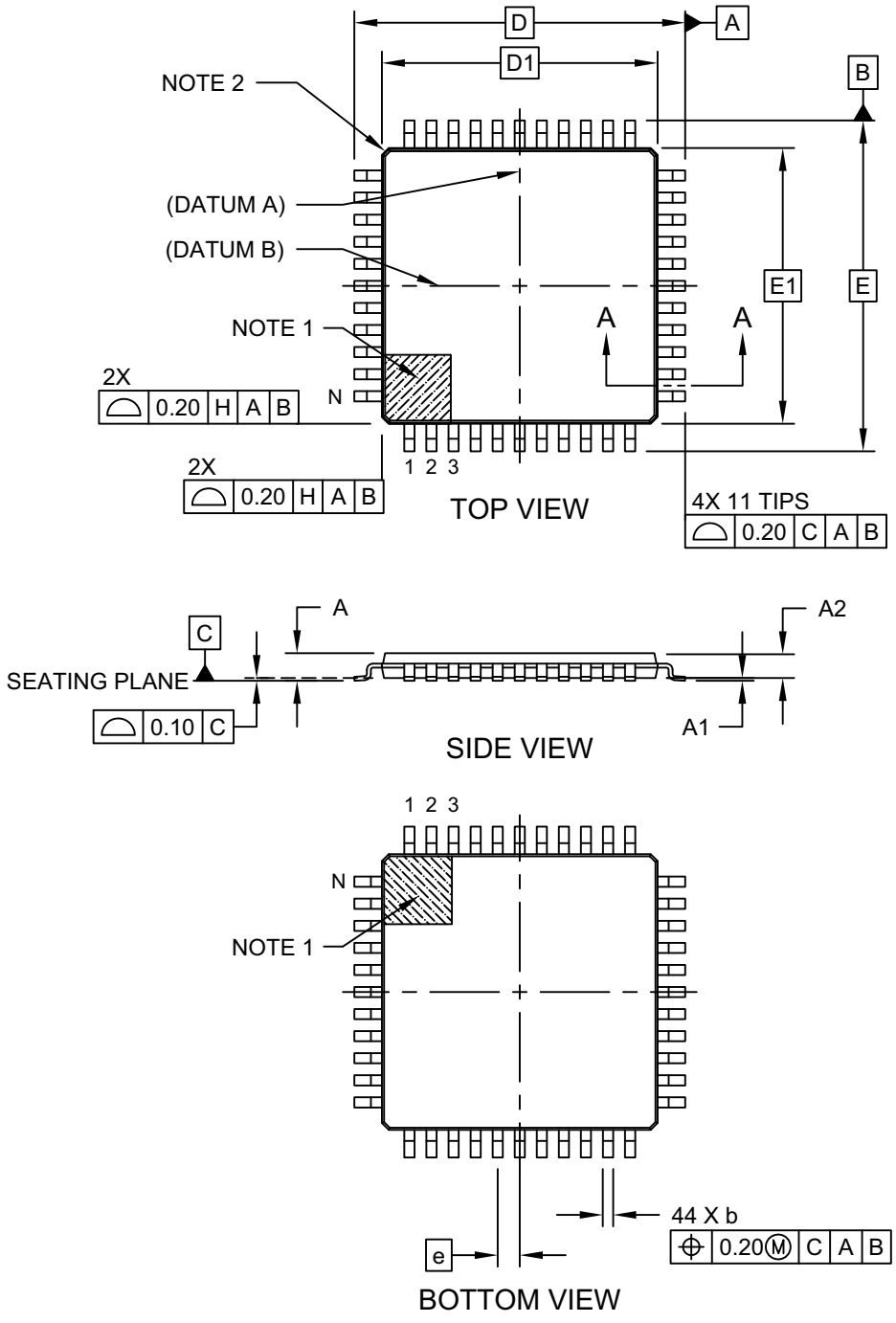
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

   REF: Reference Dimension, usually without tolerance, for information purposes only.
5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing C04-052C Sheet 2 of 2

44-Lead Plastic Thin Quad Flatpack (PT) - 10x10x1.0 mm Body [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-076C Sheet 1 of 2