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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	533MHz
Co-Processors/DSP	Security; SEC
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 + PHY (2)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	672-LBGA
Supplier Device Package	672-TBGA (35x35)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8349eczuajdb

2.1.1 Absolute Maximum Ratings

Table 1 provides the absolute maximum ratings.

Table 1. Absolute Maximum Ratings[1]

Parameter		Symbol	Max Value	Unit	Notes
Core supply voltage		V_{DD}	–0.3 to 1.32 (1.36 max for 667-MHz core frequency)	V	—
PLL supply voltage		AV_{DD}	–0.3 to 1.32 (1.36 max for 667-MHz core frequency)	V	—
DDR and DDR2 DRAM I/O voltage		GV_{DD}	–0.3 to 2.75 –0.3 to 1.98	V	—
Three-speed Ethernet I/O, MII management voltage		LV_{DD}	–0.3 to 3.63	V	—
PCI, local bus, DUART, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	–0.3 to 3.63	V	—
Input voltage	DDR DRAM signals	MV_{IN}	–0.3 to (GV_{DD} + 0.3)	V	2, 5
	DDR DRAM reference	MV_{REF}	–0.3 to (GV_{DD} + 0.3)	V	2, 5
	Three-speed Ethernet signals	LV_{IN}	–0.3 to (LV_{DD} + 0.3)	V	4, 5
	Local bus, DUART, CLKIN, system control and power management, I^2C, and JTAG signals	OV_{IN}	–0.3 to (OV_{DD} + 0.3)	V	3, 5
	PCI	OV_{IN}	–0.3 to (OV_{DD} + 0.3)	V	6
Storage temperature range		T_{STG}	–55 to 150	°C	—

Notes:

1 Functional and tested operating conditions are given in Table 2. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.

2 **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit can be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

3 **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit can be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

4 **Caution:** LV_{IN} must not exceed LV_{DD} by more than 0.3 V. This limit can be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

5 (M,L,O)V_{IN} and MV_{REF} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.

6 OVIN on the PCI interface can overshoot/undershoot according to the PCI Electrical Specification for 3.3-V operation, as shown in Figure 3.

Table 11 lists the PLL and DLL lock times.

Table 11. PLL and DLL Lock Times

Parameter/Condition	Min	Max	Unit	Notes
PLL lock times	—	100	μs	—
DLL lock times	7680	122,880	csb_clk cycles	1, 2

Notes:
1. DLL lock times are a function of the ratio between the output clock and the coherency system bus clock (csb_clk). A 2:1 ratio results in the minimum and an 8:1 ratio results in the maximum.
2. The csb_clk is determined by the CLKIN and system PLL ratio. See Section 19, "Clocking."

6 DDR and DDR2 SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface of the MPC8349EA. Note that DDR SDRAM is GV_{DD}(typ) = 2.5 V and DDR2 SDRAM is GV_{DD}(typ) = 1.8 V. The AC electrical specifications are the same for DDR and DRR2 SDRAM.

NOTE

The information in this document is accurate for revision 3.0 silicon and later. For information on revision 1.1 silicon and earlier versions see the *MPC8349E PowerQUICC II Pro Integrated Host Processor Hardware Specifications*. See Section 22.1, "Part Numbers Fully Addressed by This Document," for silicon revision level determination.

6.1 DDR and DDR2 SDRAM DC Electrical Characteristics

Table 12 provides the recommended operating conditions for the DDR2 SDRAM component(s) of the MPC8349EA when GV_{DD}(typ) = 1.8 V.

Table 12. DDR2 SDRAM DC Electrical Characteristics for GV_{DD}(typ) = 1.8 V

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O supply voltage	GV_{DD}	1.71	1.89	V	1
I/O reference voltage	MV_{REF}	0.49 × GV_{DD}	0.51 × GV_{DD}	V	2
I/O termination voltage	V_{TT}	MV_{REF} – 0.04	MV_{REF} + 0.04	V	3
Input high voltage	V_{IH}	MV_{REF} + 0.125	GV_{DD} + 0.3	V	—
Input low voltage	V_{IL}	–0.3	MV_{REF} – 0.125	V	—
Output leakage current	I_{OZ}	–9.9	9.9	μA	4
Output high current (V_{OUT} = 1.420 V)	I_{OH}	–13.4	—	mA	—

6.2.2 DDR and DDR2 SDRAM Output AC Timing Specifications

Table 20 shows the DDR and DDR2 output AC timing specifications.

Table 20. DDR and DDR2 SDRAM Output AC Timing Specifications

At recommended operating conditions with GV_{DD} of (1.8 or 2.5 V) ± 5%.

Parameter	Symbol [1]	Min	Max	Unit	Notes
ADDR/CMD/MODT output setup with respect to MCK	t_{DDKHAS}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
ADDR/CMD/MODT output hold with respect to MCK	t_{DDKHAX}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
$\overline{MCS}$(n) output setup with respect to MCK	t_{DDKHCS}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
$\overline{MCS}$(n) output hold with respect to MCK	t_{DDKHCX}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
MCK to MDQS Skew	t_{DDKHMH}	–0.6	0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS	t_{DDKHDS}, t_{DDKLDS}			ps	5
400 MHz		700	—		
333 MHz		775	—		
266 MHz		1100	—		
200 MHz		1200	—		
MDQ/MECC/MDM output hold with respect to MDQS	t_{DDKHDX}, t_{DDKLDX}			ps	5
400 MHz		700	—		
333 MHz		900	—		
266 MHz		1100	—		
200 MHz		1200	—		
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.6$	$-0.5 \times t_{MCK} + 0.6$	ns	6

Table 21. DUART DC Electrical Characteristics (continued)

Parameter	Symbol	Min	Max	Unit
High-level output voltage, I_{OH} = –100 μA	V_{OH}	OV_{DD} – 0.2	—	V
Low-level output voltage, I_{OL} = 100 μA	V_{OL}	—	0.2	V

7.2 DUART AC Electrical Specifications

Table 22 provides the AC timing parameters for the DUART interface of the MPC8349EA.

Table 22. DUART AC Timing Specifications

Parameter	Value	Unit	Notes
Minimum baud rate	256	baud	—
Maximum baud rate	> 1,000,000	baud	1
Oversample rate	16	—	2

Notes:
1. Actual attainable baud rate will be limited by the latency of interrupt processing.
2. The middle of a start bit is detected as the 8th sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16th sample.

8 Ethernet: Three-Speed Ethernet, MII Management

This section provides the AC and DC electrical characteristics for three-speeds (10/100/1000 Mbps) and MII management.

8.1 Three-Speed Ethernet Controller (TSEC)—GMII/MII/TBI/RGMII/RTBI Electrical Characteristics

The electrical characteristics specified here apply to gigabit media independent interface (GMII), the media independent interface (MII), ten-bit interface (TBI), reduced gigabit media independent interface (RGMII), and reduced ten-bit interface (RTBI) signals except management data input/output (MDIO) and management data clock (MDC). The MII, GMII, and TBI interfaces are defined for 3.3 V, and the RGMII and RTBI interfaces are defined for 2.5 V. The RGMII and RTBI interfaces follow the Hewlett-Packard *Reduced Pin-Count Interface for Gigabit Ethernet Physical Layer Device Specification*, Version 1.2a (9/22/2000). The electrical characteristics for MDIO and MDC are specified in Section 8.3, "Ethernet Management Interface Electrical Characteristics."

Table 30. TBI Receive AC Timing Specifications (continued)

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RXD[7:0], RX_DV, RX_ER (RCG[9:0]) setup time to rising PMA_RX_CLK	t_{TRDVKH}[2]	2.5	—	—	ns
RXD[7:0], RX_DV, RX_ER (RCG[9:0]) hold time to rising PMA_RX_CLK	t_{TRDXKH}[2]	1.5	—	—	ns
RX_CLK clock rise time (20%–80%)	t_{TRXR}	0.7	—	2.4	ns
RX_CLK clock fall time (80%–20%)	t_{TRXF}	0.7	—	2.4	ns

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{TRDVKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{TRX} clock reference (K) going to the high (H) state or setup time. Also, t_{TRDXKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) went invalid (X) relative to the t_{TRX} clock reference (K) going to the high (H) state. In general, the clock reference symbol is based on three letters representing the clock of a particular function. For example, the subscript of t_{TRX} represents the TBI (T) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall). For symbols representing skews, the subscript SK followed by the clock that is being skewed (TRX).
2. Setup and hold time of even numbered RCG are measured from the riding edge of PMA_RX_CLK1. Setup and hold times of odd-numbered RCG are measured from the riding edge of PMA_RX_CLK0.

Figure 15 shows the TBI receive AC timing diagram.

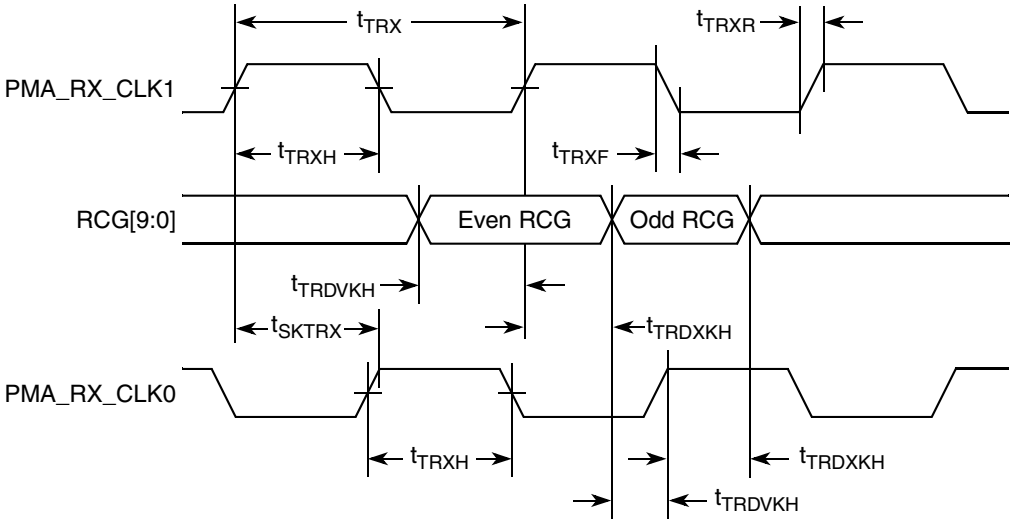


Figure 15. TBI Receive AC Timing Diagram

8.2.4 RGMII and RTBI AC Timing Specifications

Table 31 presents the RGMII and RTBI AC timing specifications.

Table 31. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT}	–0.5	—	0.5	ns
Data to clock input skew (at receiver)[2]	t_{SKRGT}	1.0	—	2.8	ns
Clock cycle duration[3]	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 1000Base-T[4, 5]	t_{RGTH}/t_{RGT}	45	50	55	%
Duty cycle for 10BASE-T and 100BASE-TX[3, 5]	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (80%–20%)	t_{RGTF}	—	—	0.75	ns

Notes:

1. In general, the clock reference symbol for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Also, the notation for rise (R) and fall (F) times follows the clock symbol. For symbols representing skews, the subscript is SK followed by the clock being skewed (RGT).
2. This implies that PC board design requires clocks to be routed so that an additional trace delay of greater than 1.5 ns is added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned.
5. Duty cycle reference is $LV_{DD}/2$.

Figure 21 through Figure 26 show the local bus signals.

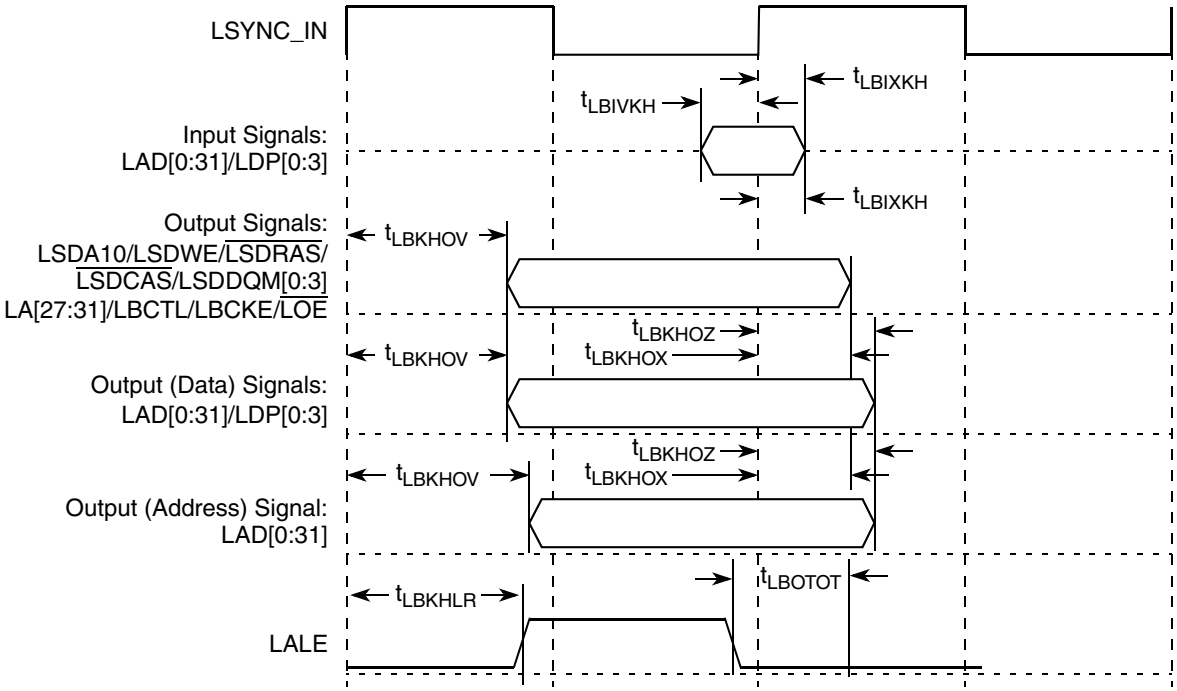


Figure 21. Local Bus Signals, Nonspecial Signals Only (DLL Enabled)

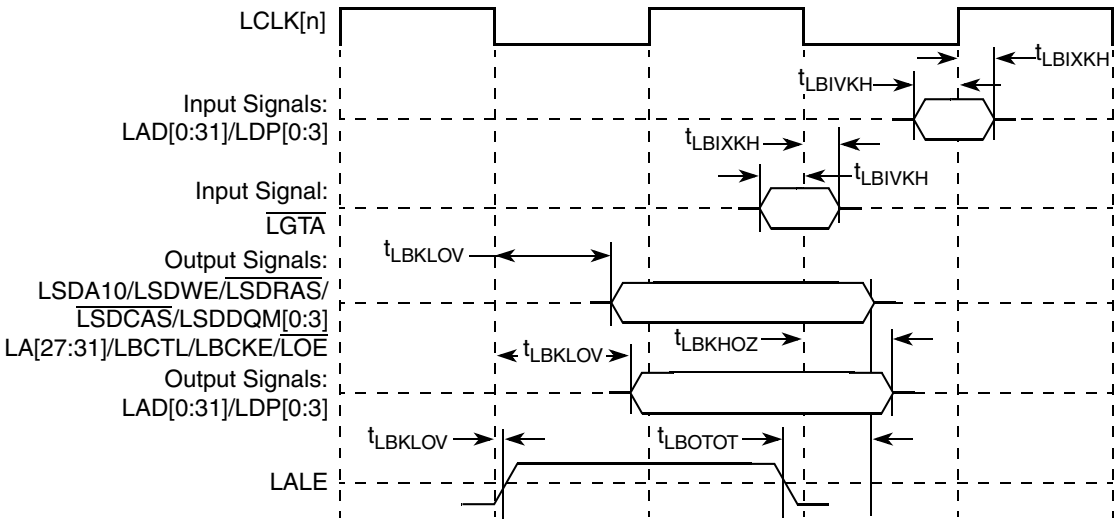


Figure 22. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

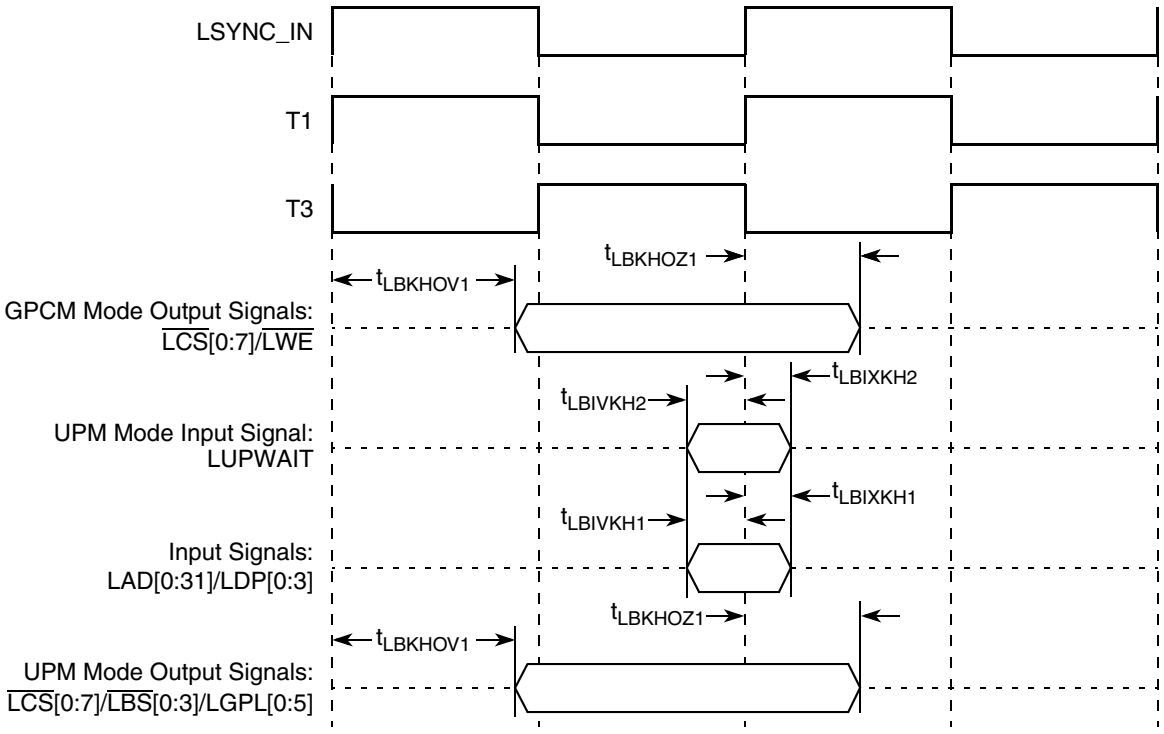


Figure 23. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Enabled)

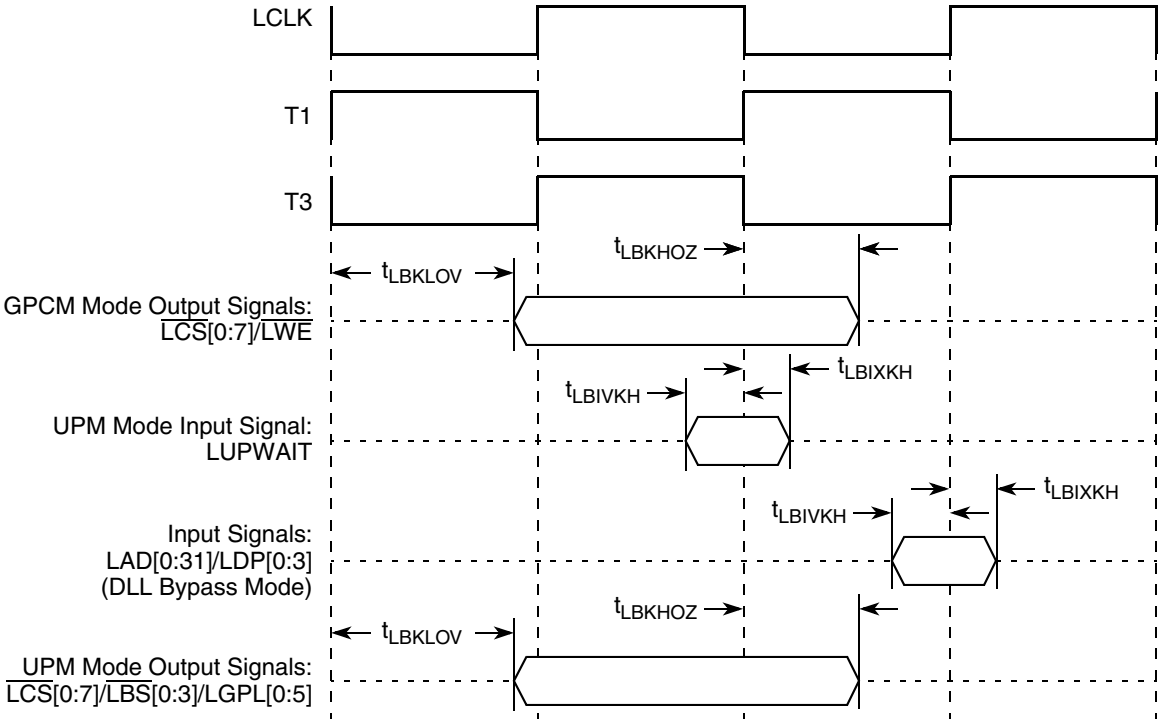


Figure 24. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Bypass Mode)

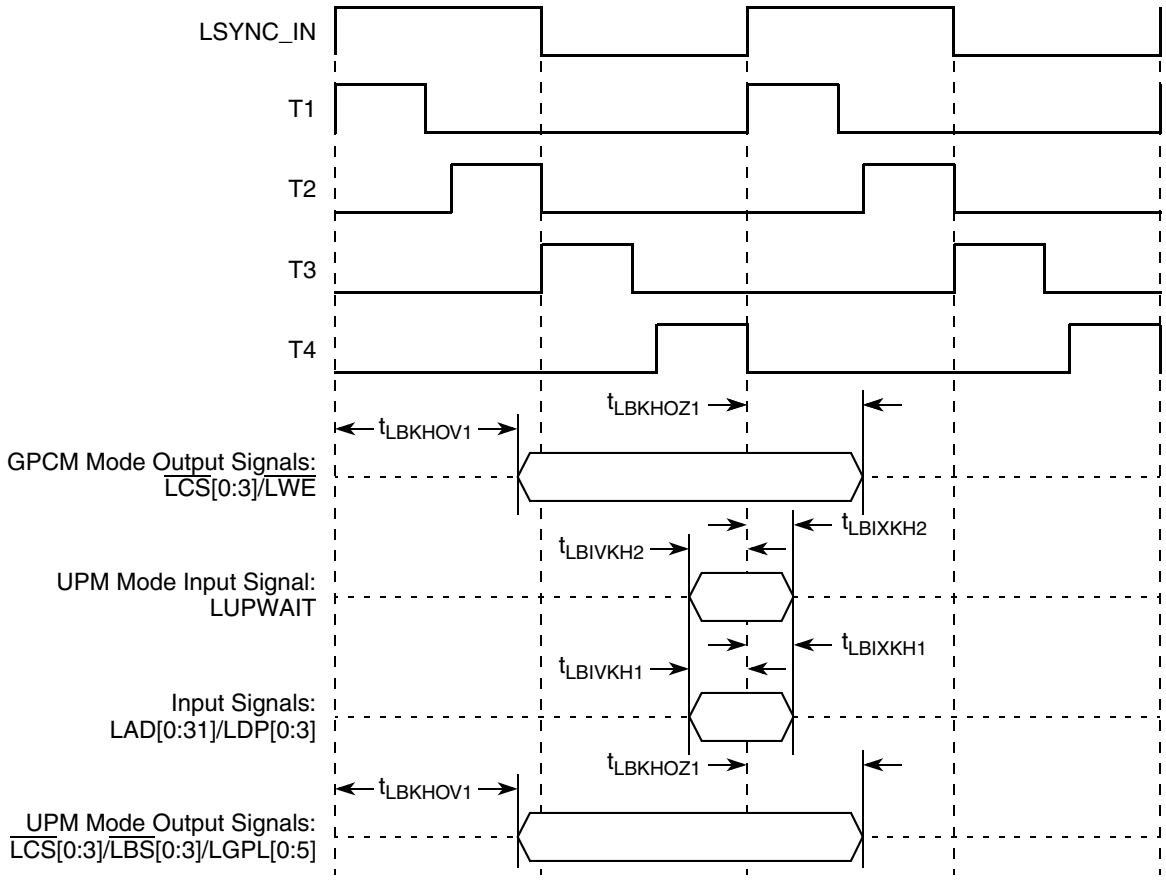


Figure 26. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (DLL Enabled)

11 JTAG

This section describes the DC and AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8349EA.

11.1 JTAG DC Electrical Characteristics

Table 40 provides the DC electrical characteristics for the IEEE Std. 1149.1 (JTAG) interface of the MPC8349EA.

Table 40. JTAG Interface DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	OV_{DD} – 0.3	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V

13 PCI

This section describes the DC and AC electrical specifications for the PCI bus of the MPC8349EA.

13.1 PCI DC Electrical Characteristics

Table 44 provides the DC electrical characteristics for the PCI interface of the MPC8349EA.

Table 44. PCI DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \geq V_{OH}$ (min) or	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	$V_{OUT} \leq V_{OL}$ (max)	–0.3	0.8	V
Input current	I_{IN}	V_{IN}[1]= 0 V or $V_{IN} = OV_{DD}$	—	±5	μA
High-level output voltage	V_{OH}	OV_{DD} = min, I_{OH} = –100 μA	OV_{DD} – 0.2	—	V
Low-level output voltage	V_{OL}	OV_{DD} = min, I_{OL} = 100 μA	—	0.2	V

Note:
1. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1.

13.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus of the MPC8349EA. Note that the PCI_CLK or PCI_SYNC_IN signal is used as the PCI input clock depending on whether the device is configured as a host or agent device. Table 45 provides the PCI AC timing specifications at 66 MHz.

Table 45. PCI AC Timing Specifications at 66 MHz[1]

Parameter	Symbol[2]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	6.0	ns	3
Output hold from clock	t_{PCKHOX}	1	—	ns	3
Clock to output high impedance	t_{PCKHOZ}	—	14	ns	3, 4
Input setup to clock	t_{PCIVKH}	3.0	—	ns	3, 5
Input hold from clock	t_{PCIXKH}	0	—	ns	3, 5
$\overline{\text{REQ64}}$ to $\overline{\text{PORESET}}$ setup time	t_{PCRVRH}	5	—	clocks	6

17 SPI

This section describes the SPI DC and AC electrical specifications.

17.1 SPI DC Electrical Characteristics

Table 53 provides the SPI DC electrical characteristics.

Table 53. SPI DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

17.2 SPI AC Timing Specifications

Table 54 provides the SPI input and output AC timing specifications.

Table 54. SPI AC Timing Specifications[1]

Parameter	Symbol[2]	Min	Max	Unit
SPI outputs valid—Master mode (internal clock) delay	t_{NIKHOV}	—	6	ns
SPI outputs hold—Master mode (internal clock) delay	t_{NIKHOX}	0.5	—	ns
SPI outputs valid—Slave mode (external clock) delay	t_{NEKHOV}	—	8	ns
SPI outputs hold—Slave mode (external clock) delay	t_{NEKHOX}	2	—	ns
SPI inputs—Master mode (internal clock input setup time	t_{NIIVKH}	4	—	ns
SPI inputs—Master mode (internal clock input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns

Notes:

1. Output specifications are measured from the 50 percent level of the rising edge of CLKIN to the 50 percent level of the signal. Timings are measured at the pin.
2. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOX} symbolizes the internal timing (NI) for the time SPICLK clock reference (K) goes to the high state (H) until outputs (O) are invalid (X).

Figure 37 provides the AC test load for the SPI.

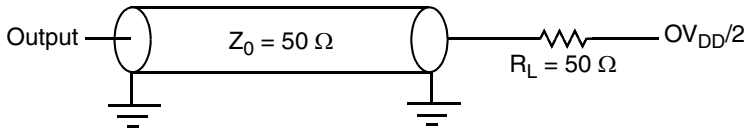
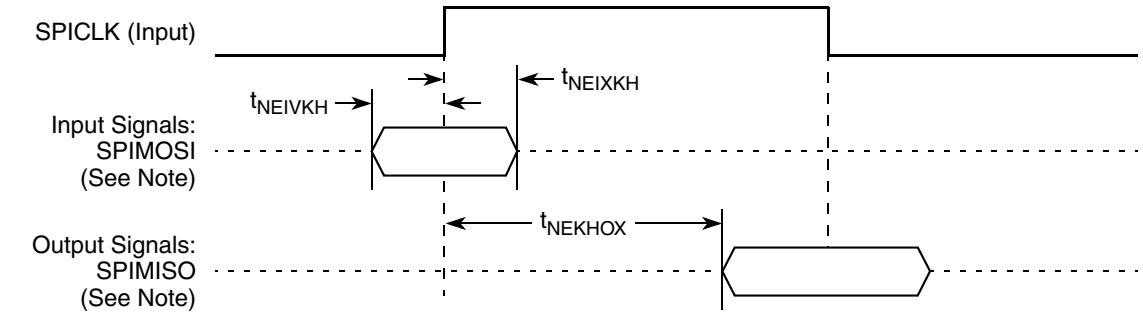


Figure 37. SPI AC Test Load

Figure 38 and Figure 39 represent the AC timings from Table 54. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

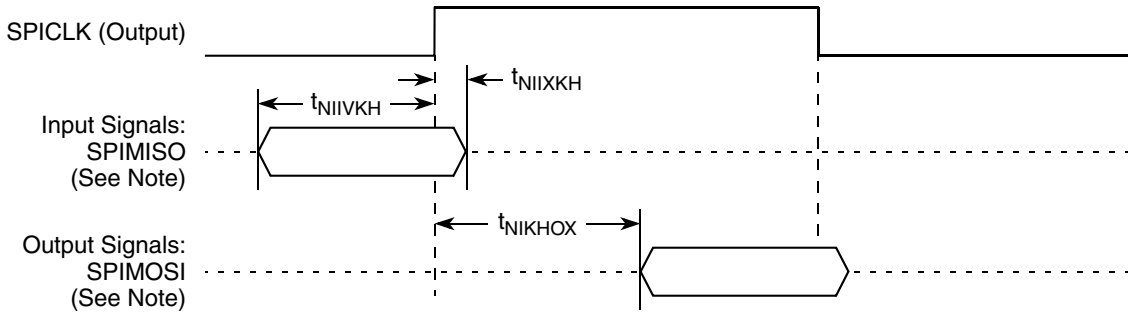
Figure 38 shows the SPI timings in slave mode (external clock).



Note: The clock edge is selectable on SPI.

Figure 38. SPI AC Timing in Slave Mode (External Clock) Diagram

Figure 39 shows the SPI timings in master mode (internal clock).



Note: The clock edge is selectable on SPI.

Figure 39. SPI AC Timing in Master Mode (Internal Clock) Diagram

18 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions. The MPC8349EA is available in a tape ball grid array (TBGA). See Section 18.1, “Package Parameters for the MPC8349EA TBGA” and Section 18.2, “Mechanical Dimensions for the MPC8349EA TBGA.

18.1 Package Parameters for the MPC8349EA TBGA

The package parameters are provided in the following list. The package type is 35 mm × 35 mm, 672 tape ball grid array (TBGA).

Package outline	35 mm × 35 mm
Interconnects	672
Pitch	1.00 mm
Module height (typical)	1.46 mm
Solder balls	62 Sn/36 Pb/2 Ag (ZU package) 96.5 Sn/3.5Ag (VV package)
Ball diameter (typical)	0.64 mm

Table 55. MPC8349EA (TBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
GPIO1[2]/DMA_DDONE0/ GTM1_TOUT1	B25	I/O	OV_{DD}	—
GPIO1[3]/DMA_DREQ1/GTM1_TIN2/ GTM2_TIN1	D24	**I/O**	OV_{DD}	—
GPIO1[4]/DMA_DACK1/ GTM1_TGATE2/GTM2_TGATE1	A25	**I/O**	OV_{DD}	—
GPIO1[5]/DMA_DDONE1/ GTM1_TOUT2/GTM2_TOUT1	B24	**I/O**	OV_{DD}	—
GPIO1[6]/DMA_DREQ2/GTM1_TIN3/ GTM2_TIN4	A24	**I/O**	OV_{DD}	—
GPIO1[7]/DMA_DACK2/ GTM1_TGATE3/GTM2_TGATE4	D23	**I/O**	OV_{DD}	—
GPIO1[8]/DMA_DDONE2/ GTM1_TOUT3	B23	**I/O**	OV_{DD}	—
GPIO1[9]/DMA_DREQ3/GTM1_TIN4/ GTM2_TIN3	A23	**I/O**	OV_{DD}	—
GPIO1[10]/DMA_DACK3/ GTM1_TGATE4/GTM2_TGATE3	F22	**I/O**	OV_{DD}	—
GPIO1[11]/DMA_DDONE3/ GTM1_TOUT4/GTM2_TOUT3	E22	**I/O**	OV_{DD}	—
USB Port 1				
MPH1_D0_ENABLEN/ DR_D0_ENABLEN	A26	I/O	OV_{DD}	—
MPH1_D1_SER_TXD/ DR_D1_SER_TXD	B26	I/O	OV_{DD}	—
MPH1_D2_VMO_SE0/ DR_D2_VMO_SE0	D25	I/O	OV_{DD}	—
MPH1_D3_SPEED/DR_D3_SPEED	A27	I/O	OV_{DD}	—
MPH1_D4_DP/DR_D4_DP	B27	I/O	OV_{DD}	—
MPH1_D5_DM/DR_D5_DM	C27	I/O	OV_{DD}	—
MPH1_D6_SER_RCV/ DR_D6_SER_RCV	D26	I/O	OV_{DD}	—
MPH1_D7_DRVVBUS/ DR_D7_DRVVBUS	E26	I/O	OV_{DD}	—
MPH1_NXT/DR_SESS_VLD_NXT	D27	I	OV_{DD}	—
MPH1_DIR_DPPULLUP/ DR_XCVR_SEL_DPPULLUP	A28	I/O	OV_{DD}	—
MPH1_STP_SUSPEND/ DR_STP_SUSPEND	F26	O	OV_{DD}	—

Table 55. MPC8349EA (TBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
TSEC2_TX_ER/GPIO1[24]	F14	I/O	OV_{DD}	—
TSEC2_TX_EN/GPIO1[12]	C5	I/O	LV_{DD2}	—
TSEC2_TX_CLK/GPIO1[30]	E14	I/O	OV_{DD}	—
DUART				
UART_SOUT[1:2]/MSRCID[0:1]/ LSRCID[0:1]	AK27, AN29	O	OV_{DD}	—
UART_SIN[1:2]/MSRCID[2:3]/ LSRCID[2:3]	AL28, AM29	I/O	OV_{DD}	—
$\overline{\text{UART\_CTS}}$[1]/MSRCID4/LSRCID4	AP30	I/O	OV_{DD}	—
$\overline{\text{UART\_CTS}}$[2]/MDVAL/ LDVAL	AN30	I/O	OV_{DD}	—
$\overline{\text{UART\_RTS}}$[1:2]	AP31, AM30	O	OV_{DD}	—
I^2C interface				
IIC1_SDA	AK29	I/O	OV_{DD}	2
IIC1_SCL	AP32	I/O	OV_{DD}	2
IIC2_SDA	AN31	I/O	OV_{DD}	2
IIC2_SCL	AM31	I/O	OV_{DD}	2
SPI				
SPIMOSI/$\overline{\text{LCS}}$[6]	AN32	I/O	OV_{DD}	—
SPIMISO/$\overline{\text{LCS}}$[7]	AP33	I/O	OV_{DD}	—
SPICLK	AK30	I/O	OV_{DD}	—
SPISEL	AL31	I	OV_{DD}	—
Clocks				
PCI_CLK_OUT[0:2]	AN9, AP9, AM10,	O	OV_{DD}	—
PCI_CLK_OUT[3]/$\overline{\text{LCS}}$[6]	AN10	O	OV_{DD}	—
PCI_CLK_OUT[4]/$\overline{\text{LCS}}$[7]	AJ11	O	OV_{DD}	—
PCI_CLK_OUT[5:7]	AP10, AL11, AM11	O	OV_{DD}	—
PCI_SYNC_IN/PCI_CLOCK	AK12	I	OV_{DD}	—
PCI_SYNC_OUT	AP11	O	OV_{DD}	3
RTC/PIT_CLOCK	AM32	I	OV_{DD}	—
CLKIN	AM9	I	OV_{DD}	—
JTAG				
TCK	E20	I	OV_{DD}	—
TDI	F20	I	OV_{DD}	4

19 Clocking

Figure 41 shows the internal distribution of the clocks.

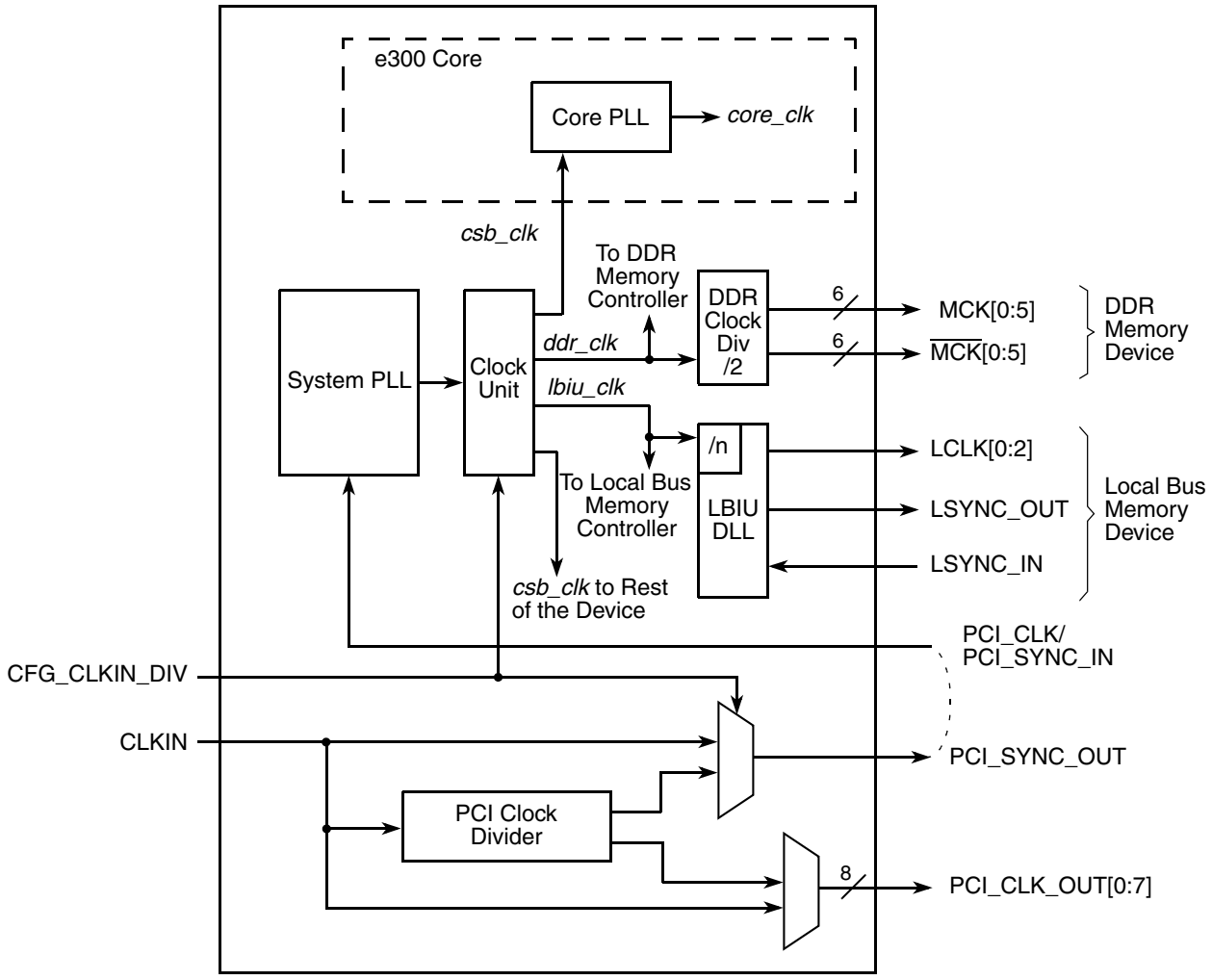


Figure 41. MPC8349EA Clock Subsystem

The primary clock source can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. When the MPC8349EA is configured as a PCI host device, CLKIN is its primary input clock. CLKIN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCICD*n*] parameters select whether CLKIN or CLKIN/2 is driven out on the PCI_CLK_OUT*n* signals.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the MPC8349EA to function. When the device is configured as a PCI agent device, PCI_CLK is the primary input clock and the CLKIN signal should be tied to GND.

As shown in Figure 41, the primary clock input (frequency) is multiplied up by the system phase-locked loop (PLL) and the clock unit to create the coherent system bus clock (*csb_clk*), the internal clock for the DDR controller (*ddr_clk*), and the internal clock for the local bus interface unit (*lbiu_clk*).

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$csb\_clk = \{PCI\_SYNC\_IN \times (1 + CFG\_CLKIN\_DIV)\} \times SPMF$$

In PCI host mode, PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the CLKIN frequency.

The *csb_clk* serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies the *csb_clk* frequency to create the internal clock for the e300 core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL), which is loaded at power-on reset or by one of the hard-coded reset options. See the chapter on reset, clocking, and initialization in the *MPC8349EA Reference Manual* for more information on the clock subsystem.

The internal *ddr_clk* frequency is determined by the following equation:

$$ddr\_clk = csb\_clk \times (1 + RCWL[DDRCM])$$

ddr_clk is not the external memory bus frequency; *ddr_clk* passes through the DDR clock divider (÷2) to create the differential DDR memory bus clock outputs (MCK and $\overline{\text{MCK}}$). However, the data rate is the same frequency as *ddr_clk*.

The internal *lbiu_clk* frequency is determined by the following equation:

$$lbiu\_clk = csb\_clk \times (1 + RCWL[LBIUCM])$$

lbiu_clk is not the external local bus frequency; *lbiu_clk* passes through the LBIU clock divider to create the external local bus clock outputs (LSYNC_OUT and LCLK[0:2]). The LBIU clock divider ratio is controlled by LCCR[CLKDIV].

In addition, some of the internal units may have to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory-mapped register after the device exits reset. Table 56 specifies which units have a configurable clock frequency.

Table 56. Configurable Clock Units

Unit	Default Frequency	Options
TSEC1	*csb_clk/3*	Off, *csb_clk, csb_clk/2, csb_clk/3*
TSEC2, I^2C1	*csb_clk/3*	Off, *csb_clk, csb_clk/2, csb_clk/3*
Security core	*csb_clk/3*	Off, *csb_clk,* csb_clk/2, *csb_clk/3*
USB DR, USB MPH	*csb_clk/3*	Off, csb_clk, csb_clk/2, *csb_clk/3*
PCI1, PCI2 and DMA complex	*csb_clk*	Off, *csb_clk*

Table 59. CSB Frequency Options for Host Mode (continued)

CFG_CLKIN_DIV at Reset[1]	SPMF	*csb_clk* : Input Clock Ratio[2]	Input Clock Frequency (MHz)[2]			
			16.67	25	33.33	66.67
			csb_clk Frequency (MHz)			
Low	0110	6 : 1	100	150	200	
Low	0111	7 : 1	116	175	233	
Low	1000	8 : 1	133	200	266	
Low	1001	9 : 1	150	225	300	
Low	1010	10 : 1	166	250	333	
Low	1011	11 : 1	183	275		
Low	1100	12 : 1	200	300		
Low	1101	13 : 1	216	325		
Low	1110	14 : 1	233			
Low	1111	15 : 1	250			
Low	0000	16 : 1	266			
High	0010	2 : 1				133
High	0011	3 : 1			100	200
High	0100	4 : 1			133	266
High	0101	5 : 1			166	333
High	0110	6 : 1			200	
High	0111	7 : 1			233	
High	1000	8 : 1				

1 CFG_CLKIN_DIV selects the ratio between CLKIN and PCI_SYNC_OUT.
2 CLKIN is the input clock in host mode; PCI_CLK is the input clock in agent mode.

Table 60. CSB Frequency Options for Agent Mode

CFG_CLKIN_DIV at Reset[1]	SPMF	*csb_clk* : Input Clock Ratio[2]	Input Clock Frequency (MHz)[2]			
			16.67	25	33.33	66.67
			csb_clk Frequency (MHz)			
Low	0010	2 : 1				133
Low	0011	3 : 1			100	200
Low	0100	4 : 1		100	133	266
Low	0101	5 : 1		125	166	333

(edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)

T_A = ambient temperature for the package (°C)

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

20.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature and a measure of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)

T_T = thermocouple temperature on top of package (°C)

Ψ_{JT} = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per the JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

20.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

Some application environments require a heat sink to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

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20.3 Heat Sink Attachment

When heat sinks are attached, an interface material is required, preferably thermal grease and a spring clip. The spring clip should connect to the printed-circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces that can lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. The recommended maximum force on the top of the package is 10 lb force (4.5 kg force). Any adhesive attachment should attach to painted or plastic surfaces, and its performance should be verified under the application requirements.

20.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When a heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimize the size of the clearance to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (°C)
T_C = case temperature of the package (°C)
$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)
P_D = power dissipation (W)

21 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8349EA.

21.1 System Clocking

The MPC8349EA includes two PLLs:

1. The platform PLL generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 19.1, “System PLL Configuration.”