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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	196
Number of Logic Elements/Cells	784
Total RAM Bits	-
Number of I/O	65
Number of Gates	10000
Voltage - Supply	4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	84-LCC (J-Lead)
Supplier Device Package	84-PLCC (29.31x29.31)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc5206-5pc84c

XC3000 family: XC5200 devices support an additional programming mode: Peripheral Synchronous.

XC3000 family: The XC5200 family does not support Power-down, but offers a Global 3-state input that does not reset any flip-flops.

XC3000 family: The XC5200 family does not provide an on-chip crystal oscillator amplifier, but it does provide an internal oscillator from which a variety of frequencies up to 12 MHz are available.

Architectural Overview

Figure 1 presents a simplified, conceptual overview of the XC5200 architecture. Similar to conventional FPGAs, the XC5200 family consists of programmable IOBs, programmable logic blocks, and programmable interconnect. Unlike other FPGAs, however, the logic and local routing resources of the XC5200 family are combined in flexible VersaBlocks (Figure 2). General-purpose routing connects to the VersaBlock through the General Routing Matrix (GRM).

VersaBlock: Abundant Local Routing Plus Versatile Logic

The basic logic element in each VersaBlock structure is the Logic Cell, shown in Figure 3. Each LC contains a 4-input function generator (F), a storage device (FD), and control logic. There are five independent inputs and three outputs to each LC. The independence of the inputs and outputs allows the software to maximize the resource utilization within each LC. Each Logic Cell also contains a direct feedthrough path that does not sacrifice the use of either the function generator or the register; this feature is a first for FPGAs. The storage device is configurable as either a D flip-flop or a latch. The control logic consists of carry logic for fast implementation of arithmetic functions, which can also be configured as a cascade chain allowing decode of very wide input functions.

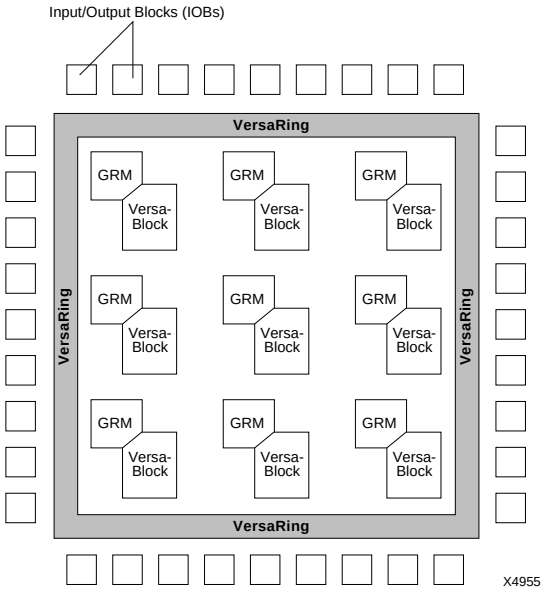

Figure 1: XC5200 Architectural Overview

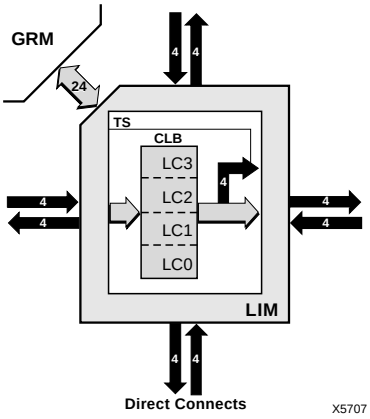

Figure 2: VersaBlock

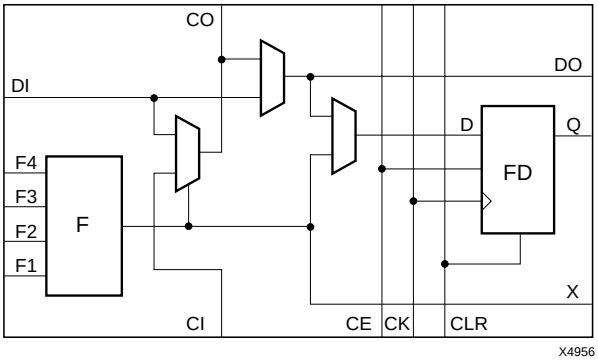

Figure 3: XC5200 Logic Cell (Four LCs per CLB)

The XC5200 CLB consists of four LCs, as shown in Figure 4. Each CLB has 20 independent inputs and 12 independent outputs. The top and bottom pairs of LCs can be configured to implement 5-input functions. The challenge of FPGA implementation software has always been to maximize the usage of logic resources. The XC5200 family addresses this issue by surrounding each CLB with two types of local interconnect — the Local Interconnect Matrix (LIM) and direct connects. These two interconnect resources, combined with the CLB, form the VersaBlock, represented in Figure 2.

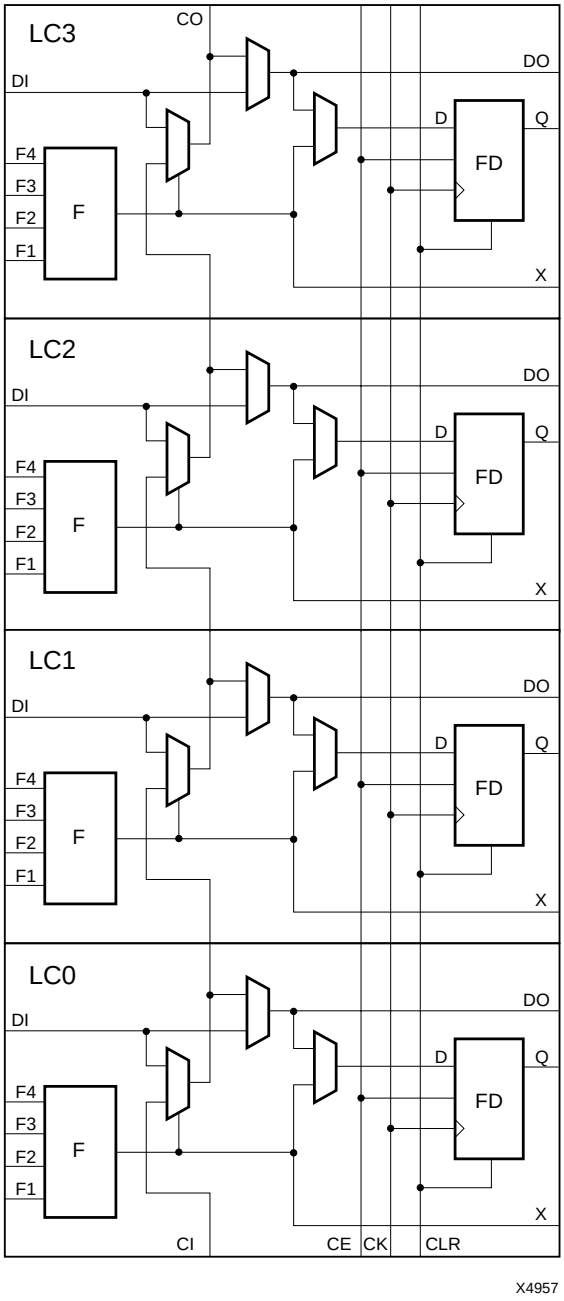

Figure 4: Configurable Logic Block

The LIM provides 100% connectivity of the inputs and outputs of each LC in a given CLB. The benefit of the LIM is that no general routing resources are required to connect feedback paths within a CLB. The LIM connects to the GRM via 24 bidirectional nodes.

The direct connects allow immediate connections to neighboring CLBs, once again without using any of the general interconnect. These two layers of local routing resource improve the granularity of the architecture, effectively making the XC5200 family a “sea of logic cells.” Each Versa-Block has four 3-state buffers that share a common enable line and directly drive horizontal and vertical Longlines, creating robust on-chip bussing capability. The VersaBlock allows fast, local implementation of logic functions, effectively implementing user designs in a hierarchical fashion. These resources also minimize local routing congestion and improve the efficiency of the general interconnect, which is used for connecting larger groups of logic. It is this combination of both fine-grain and coarse-grain architecture attributes that maximize logic utilization in the XC5200 family. This symmetrical structure takes full advantage of the third metal layer, freeing the placement software to pack user logic optimally with minimal routing restrictions.

VersaRing I/O Interface

The interface between the IOBs and core logic has been redesigned in the XC5200 family. The IOBs are completely decoupled from the core logic. The XC5200 IOBs contain dedicated boundary-scan logic for added board-level testability, but do not include input or output registers. This approach allows a maximum number of IOBs to be placed around the device, improving the I/O-to-gate ratio and decreasing the cost per I/O. A “freeway” of interconnect cells surrounding the device forms the VersaRing, which provides connections from the IOBs to the internal logic. These incremental routing resources provide abundant connections from each IOB to the nearest VersaBlock, in addition to Longline connections surrounding the device. The VersaRing eliminates the historic trade-off between high logic utilization and pin placement flexibility. These incremental edge resources give users increased flexibility in preassigning (i.e., locking) I/O pins before completing their logic designs. This ability accelerates time-to-market, since PCBs and other system components can be manufactured concurrent with the logic design.

General Routing Matrix

The GRM is functionally similar to the switch matrices found in other architectures, but it is novel in its tight coupling to the logic resources contained in the VersaBlocks. Advanced simulation tools were used during the development of the XC5200 architecture to determine the optimal level of routing resources required. The XC5200 family contains six levels of interconnect hierarchy — a series of

single-length lines, double-length lines, and Longlines all routed through the GRM. The direct connects, LIM, and logic-cell feedthrough are contained within each Versa-Block. Throughout the XC5200 interconnect, an efficient multiplexing scheme, in combination with three layer metal (TLM), was used to improve the overall efficiency of silicon usage.

Performance Overview

The XC5200 family has been benchmarked with many designs running synchronous clock rates beyond 66 MHz. The performance of any design depends on the circuit to be implemented, and the delay through the combinatorial and sequential logic elements, plus the delay in the interconnect routing. A rough estimate of timing can be made by assuming 3-6 ns per logic level, which includes direct-connect routing delays, depending on speed grade. More accurate estimations can be made using the information in the Switching Characteristic Guideline section.

Taking Advantage of Reconfiguration

FPGA devices can be reconfigured to change logic function while resident in the system. This capability gives the system designer a new degree of freedom not available with any other type of logic.

Hardware can be changed as easily as software. Design updates or modifications are easy, and can be made to products already in the field. An FPGA can even be reconfigured dynamically to perform different functions at different times.

Reconfigurable logic can be used to implement system self-diagnostics, create systems capable of being reconfigured for different environments or operations, or implement multi-purpose hardware for a given application. As an added benefit, using reconfigurable FPGA devices simplifies hardware design and debugging and shortens product time-to-market.

Detailed Functional Description

Configurable Logic Blocks (CLBs)

Figure 4 shows the logic in the XC5200 CLB, which consists of four Logic Cells (LC[3:0]). Each Logic Cell consists of an independent 4-input Lookup Table (LUT), and a D-Type flip-flop or latch with common clock, clock enable, and clear, but individually selectable clock polarity. Additional logic features provided in the CLB are:

- An independent 5-input LUT by combining two 4-input LUTs.
- High-speed carry propagate logic.
- High-speed pattern decoding.
- High-speed direct connection to flip-flop D-inputs.
- Individual selection of either a transparent, level-sensitive latch or a D flip-flop.
- Four 3-state buffers with a shared Output Enable.

5-Input Functions

Figure 5 illustrates how the outputs from the LUTs from LC0 and LC1 can be combined with a 2:1 multiplexer (F5_MUX) to provide a 5-input function. The outputs from the LUTs of LC2 and LC3 can be similarly combined.

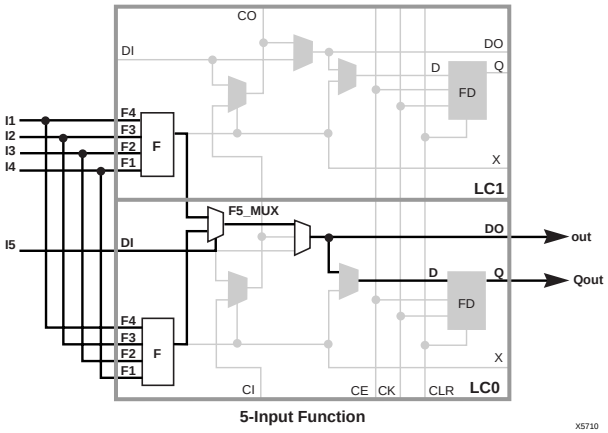

Figure 5: Two LUTs in Parallel Combined to Create a 5-input Function

non-zero hold, attach a NODELAY attribute or property to the flip-flop or input buffer.

IOB Output Signals

Output signals can be optionally inverted within the IOB, and pass directly to the pad. As with the inputs, a CLB flip-flop or latch can be used to store the output signal.

An active-High 3-state signal can be used to place the output buffer in a high-impedance state, implementing 3-state outputs or bidirectional I/O. Under configuration control, the output (OUT) and output 3-state (T) signals can be inverted. The polarity of these signals is independently configured for each IOB.

The XC5200 devices provide a guaranteed output sink current of 8 mA.

Supported destinations for XC5200-Series device outputs are shown in Table 6.(For a detailed discussion of how to interface between 5 V and 3.3 V devices, see the 3V Products section of *The Programmable Logic Data Book*.)

An output can be configured as open-drain (open-collector) by placing an OBUFT symbol in a schematic or HDL code, then tying the 3-state pin (T) to the output signal, and the input pin (I) to Ground. (See Figure 12.)

Table 6: Supported Destinations for XC5200-Series Outputs

Destination	XC5200 Output Mode
	5 V, CMOS
XC5200 device, V_{CC}=3.3 V, CMOS-threshold inputs	√
Any typical device, V_{CC} = 3.3 V, CMOS-threshold inputs	some[1]
Any device, V_{CC} = 5 V, TTL-threshold inputs	√
Any device, V_{CC} = 5 V, CMOS-threshold inputs	√

1. Only if destination device has 5-V tolerant inputs

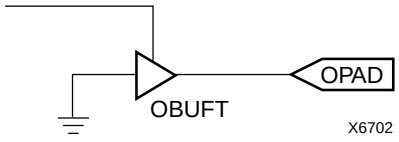

Figure 12: Open-Drain Output

Output Slew Rate

The slew rate of each output buffer is, by default, reduced, to minimize power bus transients when switching non-critical signals. For critical signals, attach a FAST attribute or property to the output buffer or flip-flop.

For XC5200 devices, maximum total capacitive load for simultaneous fast mode switching in the same direction is 200 pF for all package pins between each Power/Ground pin pair. For some XC5200 devices, additional internal Power/Ground pin pairs are connected to special Power and Ground planes within the packages, to reduce ground bounce.

For slew-rate limited outputs this total is two times larger for each device type: 400 pF for XC5200 devices. This maximum capacitive load should not be exceeded, as it can result in ground bounce of greater than 1.5 V amplitude and more than 5 ns duration. This level of ground bounce may cause undesired transient behavior on an output, or in the internal logic. This restriction is common to all high-speed digital ICs, and is not particular to Xilinx or the XC5200 Series.

XC5200-Series devices have a feature called “Soft Start-up,” designed to reduce ground bounce when all outputs are turned on simultaneously at the end of configuration. When the configuration process is finished and the device starts up, the first activation of the outputs is automatically slew-rate limited. Immediately following the initial activation of the I/O, the slew rate of the individual outputs is determined by the individual configuration option for each IOB.

Global Three-State

A separate Global 3-State line (not shown in Figure 11) forces all FPGA outputs to the high-impedance state, unless boundary scan is enabled and is executing an EXTEST instruction. This global net (GTS) does not compete with other routing resources; it uses a dedicated distribution network.

GTS can be driven from any user-programmable pin as a global 3-state input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GTS pin of the STARTUP symbol. A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the Global 3-State signal. Using GTS is similar to Global Reset. See Figure 8 on page 90 for details. Alternatively, GTS can be driven from any internal node.

Other IOB Options

There are a number of other programmable options in the XC5200-Series IOB.

Pull-up and Pull-down Resistors

Programmable IOB pull-up and pull-down resistors are useful for tying unused pins to Vcc or Ground to minimize power consumption and reduce noise sensitivity. The configurable pull-up resistor is a p-channel transistor that pulls

segments span the width and height of the chip, respectively.

Two low-skew horizontal and vertical unidirectional global-line segments span each row and column of the chip, respectively.

Single- and Double-Length Lines

The single- and double-length bidirectional line segments make up the bulk of the routing channels. The double-length lines hop across every other CLB to reduce the propagation delays in speed-critical nets. Regenerating the signal strength is recommended after traversing three or four such segments. Xilinx place-and-route software automatically connects buffers in the path of the signal as necessary. Single- and double-length lines cannot drive onto Longlines and global lines; Longlines and global lines can, however, drive onto single- and double-length lines. As a general rule, Longline and global-line connections to the general routing matrix are unidirectional, with the signal direction from these lines toward the routing matrix.

Longlines

Longlines are used for high-fan-out signals, 3-state busses, low-skew nets, and faraway destinations. Row and column splitter PIPs in the middle of the array effectively double the total number of Longlines by electrically dividing them into two separated half-lines. Longlines are driven by the 3-state buffers in each CLB, and are driven by similar buffers at the periphery of the array from the VersaRing I/O Interface.

Bus-oriented designs are easily implemented by using Longlines in conjunction with the 3-state buffers in the CLB and in the VersaRing. Additionally, weak keeper cells at the periphery retain the last valid logic level on the Longlines when all buffers are in 3-state mode.

Longlines connect to the single-length or double-length lines, or to the logic inside the CLB, through the General Routing Matrix. The only manner in which a Longline can be driven is through the four 3-state buffers; therefore, a Longline-to-Longline or single-line-to-Longline connection through PIPs in the General Routing Matrix is not possible. Again, as a general rule, long- and global-line connections to the General Routing Matrix are unidirectional, with the signal direction from these lines toward the routing matrix.

The XC5200 family has no pull-ups on the ends of the Longlines sourced by TBUFs, unlike the XC4000 Series. Consequently, wired functions (i.e., WAND and WORAND) and wide multiplexing functions requiring pull-ups for undefined states (i.e., bus applications) must be implemented in a different way. In the case of the wired functions, the same functionality can be achieved by taking advantage of the carry/cascade logic described above, implementing a wide logic function in place of the wired function. In the case of 3-state bus applications, the user must insure that all states of the multiplexing function are defined. This process is as simple as adding an additional TBUF to drive the bus High when the previously undefined states are activated.

Global Lines

Global buffers in Xilinx FPGAs are special buffers that drive a dedicated routing network called Global Lines, as shown in Figure 16. This network is intended for high-fanout clocks or other control signals, to maximize speed and minimize skewing while distributing the signal to many loads.

The XC5200 family has a total of four global buffers (BUFG symbol in the library), each with its own dedicated routing channel. Two are distributed vertically and two horizontally throughout the FPGA.

The global lines provide direct input only to the CLB clock pins. The global lines also connect to the General Routing Matrix to provide access from these lines to the function generators and other control signals.

Four clock input pads at the corners of the chip, as shown in Figure 16, provide a high-speed, low-skew clock network to each of the four global-line buffers. In addition to the dedicated pad, the global lines can be sourced by internal logic. PIPs from several routing channels within the VersaRing can also be configured to drive the global-line buffers.

Details of all the programmable interconnect for a CLB is shown in Figure 17.

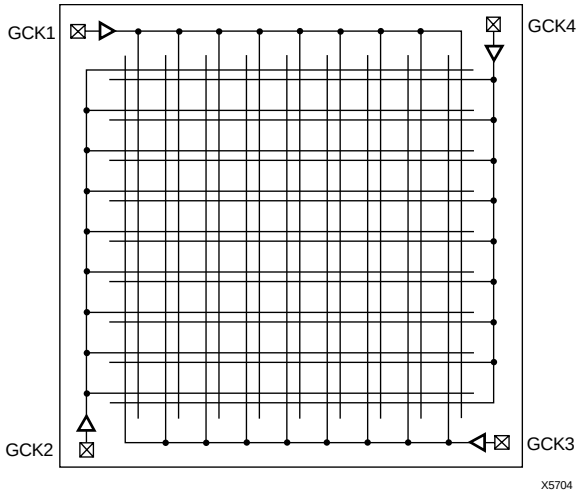

Figure 16: Global Lines

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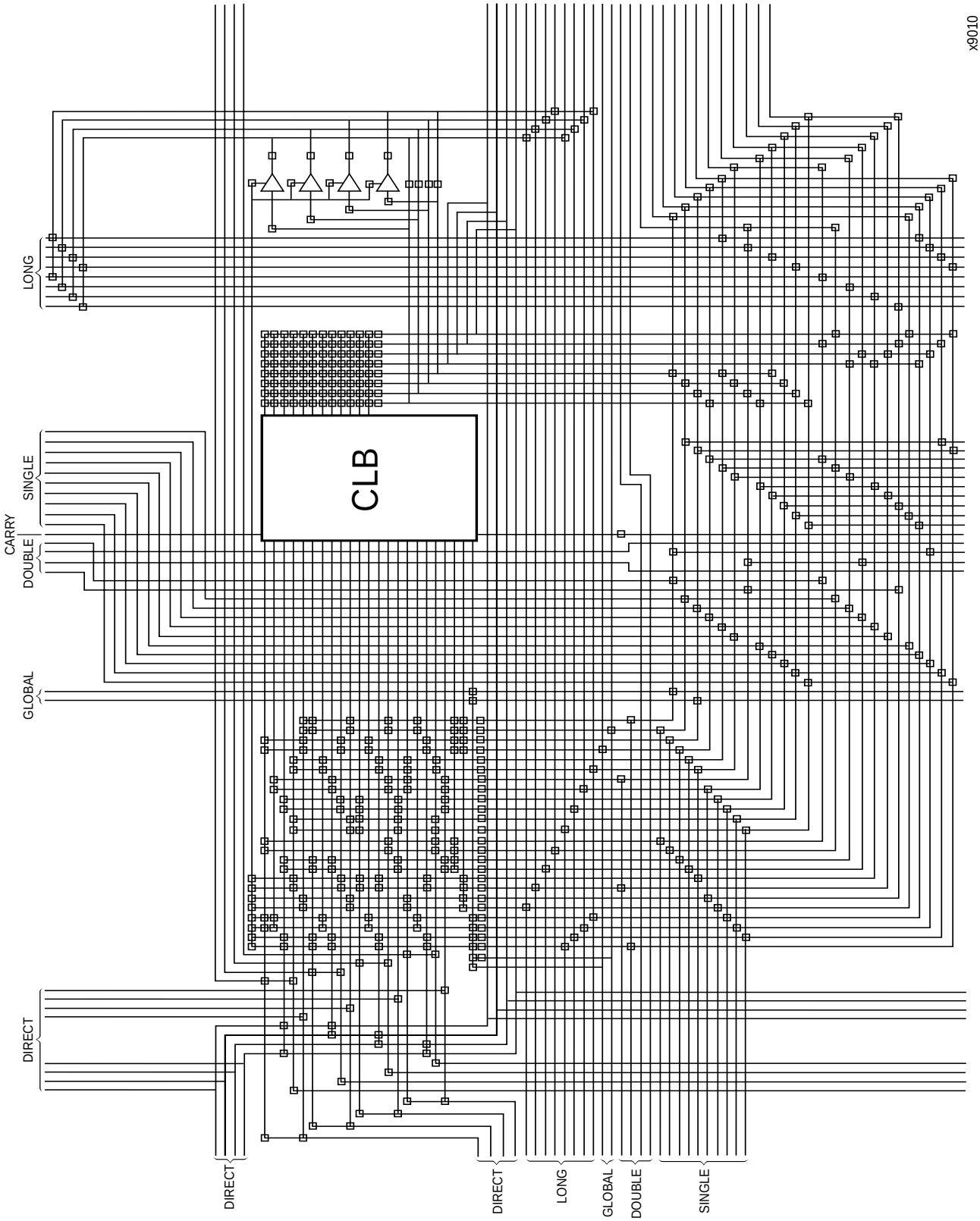


Figure 17: Detail of Programmable Interconnect Associated with XC5200 Series CLB

VersaRing Input/Output Interface

The VersaRing, shown in Figure 18, is positioned between the core logic and the pad ring; it has all the routing resources of a VersaBlock without the CLB logic. The VersaRing decouples the core logic from the I/O pads. Each VersaRing Cell provides up to four pad-cell connections on one side, and connects directly to the CLB ports on the other side.

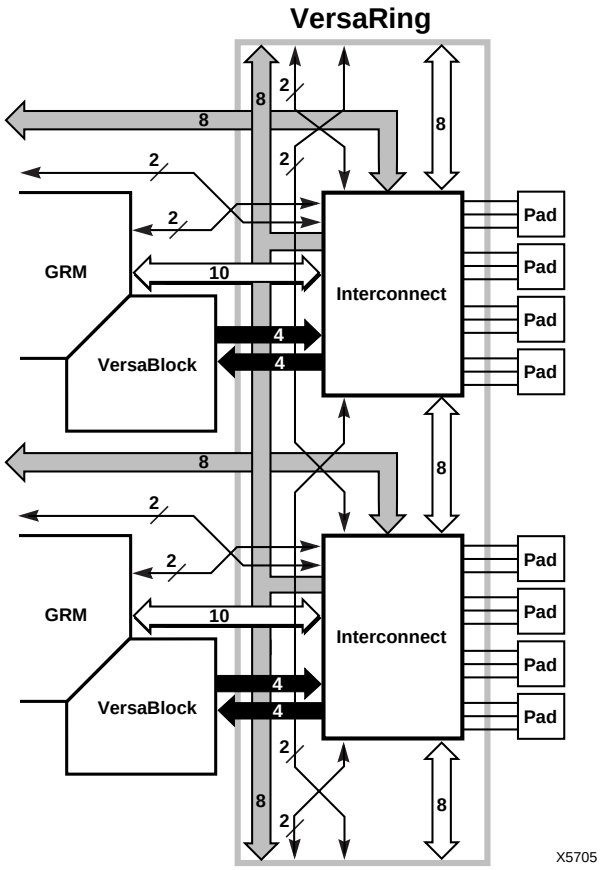


Figure 18: VersaRing I/O Interface

Boundary Scan

The “bed of nails” has been the traditional method of testing electronic assemblies. This approach has become less appropriate, due to closer pin spacing and more sophisticated assembly methods like surface-mount technology and multi-layer boards. The IEEE boundary scan standard 1149.1 was developed to facilitate board-level testing of electronic assemblies. Design and test engineers can imbed a standard test logic structure in their device to achieve high fault coverage for I/O and internal logic. This structure is easily implemented with a four-pin interface on any boundary scan-compatible IC. IEEE 1149.1-compatible devices may be serial daisy-chained together, connected in parallel, or a combination of the two.

XC5200 devices support all the mandatory boundary-scan instructions specified in the IEEE standard 1149.1. A Test Access Port (TAP) and registers are provided that implement the EXTEST, SAMPLE/PRELOAD, and BYPASS instructions. The TAP can also support two USERCODE instructions. When the boundary scan configuration option is selected, three normal user I/O pins become dedicated inputs for these functions. Another user output pin becomes the dedicated boundary scan output.

Boundary-scan operation is independent of individual IOB configuration and package type. All IOBs are treated as independently controlled bidirectional pins, including any unbonded IOBs. Retaining the bidirectional test capability after configuration provides flexibility for interconnect testing.

Also, internal signals can be captured during EXTEST by connecting them to unbonded IOBs, or to the unused outputs in IOBs used as unidirectional input pins. This technique partially compensates for the lack of INTEST support.

The user can serially load commands and data into these devices to control the driving of their outputs and to examine their inputs. This method is an improvement over bed-of-nails testing. It avoids the need to over-drive device outputs, and it reduces the user interface to four pins. An optional fifth pin, a reset for the control logic, is described in the standard but is not implemented in Xilinx devices.

The dedicated on-chip logic implementing the IEEE 1149.1 functions includes a 16-state machine, an instruction register and a number of data registers. The functional details can be found in the IEEE 1149.1 specification and are also discussed in the Xilinx application note XAPP 017: *“Boundary Scan in XC4000 and XC5200 Series devices”*

Figure 19 on page 99 is a diagram of the XC5200-Series boundary scan logic. It includes three bits of Data Register per IOB, the IEEE 1149.1 Test Access Port controller, and the Instruction Register with decodes.

The public boundary-scan instructions are always available prior to configuration. After configuration, the public instructions and any USERCODE instructions are only available if specified in the design. While SAMPLE and BYPASS are available during configuration, it is recommended that boundary-scan operations not be performed during this transitory period.

In addition to the test instructions outlined above, the boundary-scan circuitry can be used to configure the FPGA device, and to read back the configuration data.

All of the XC4000 boundary-scan modes are supported in the XC5200 family. Three additional outputs for the User-Register are provided (Reset, Update, and Shift), repre-

Table 9: Pin Descriptions (Continued)

Pin Name	I/O During Config.	I/O After Config.	Pin Description
Unrestricted User-Programmable I/O Pins			
I/O	Weak Pull-up	I/O	These pins can be configured to be input and/or output after configuration is completed. Before configuration is completed, these pins have an internal high-value pull-up resistor (20 kΩ - 100 kΩ) that defines the logic level as High.

Configuration

Configuration is the process of loading design-specific programming data into one or more FPGAs to define the functional operation of the internal blocks and their interconnections. This is somewhat like loading the command registers of a programmable peripheral chip. XC5200-Series devices use several hundred bits of configuration data per CLB and its associated interconnects. Each configuration bit defines the state of a static memory cell that controls either a function look-up table bit, a multiplexer input, or an interconnect pass transistor. The development system translates the design into a netlist file. It automatically partitions, places and routes the logic and generates the configuration data in PROM format.

Special Purpose Pins

Three configuration mode pins (M2, M1, M0) are sampled prior to configuration to determine the configuration mode. After configuration, these pins can be used as auxiliary I/O connections. The development system does not use these resources unless they are explicitly specified in the design entry. This is done by placing a special pad symbol called MD2, MD1, or MD0 instead of the input or output pad symbol.

In XC5200-Series devices, the mode pins have weak pull-up resistors during configuration. With all three mode pins High, Slave Serial mode is selected, which is the most popular configuration mode. Therefore, for the most common configuration mode, the mode pins can be left unconnected. (Note, however, that the internal pull-up resistor value can be as high as 100 kΩ.) After configuration, these pins can individually have weak pull-up or pull-down resistors, as specified in the design. A pull-down resistor value of 3.3kΩ is recommended.

These pins are located in the lower left chip corner and are near the readback nets. This location allows convenient routing if compatibility with the XC2000 and XC3000 family conventions of M0/RT, M1/RD is desired.

Configuration Modes

XC5200 devices have seven configuration modes. These modes are selected by a 3-bit input code applied to the M2, M1, and M0 inputs. There are three self-loading Master modes, two Peripheral modes, and a Serial Slave mode,

Table 10: Configuration Modes

Mode	M2	M1	M0	CCLK	Data
Master Serial	0	0	0	output	Bit-Serial
Slave Serial	1	1	1	input	Bit-Serial
Master Parallel Up	1	0	0	output	Byte-Wide, increment from 00000
Master Parallel Down	1	1	0	output	Byte-Wide, decrement from 3FFFF
Peripheral Synchronous*	0	1	1	input	Byte-Wide
Peripheral Asynchronous	1	0	1	output	Byte-Wide
Express	0	1	0	input	Byte-Wide
Reserved	0	0	1	—	—

Note :*Peripheral Synchronous can be considered byte-wide Slave Parallel

which is used primarily for daisy-chained devices. The seventh mode, called Express mode, is an additional slave mode that allows high-speed parallel configuration. The coding for mode selection is shown in Table 10.

Note that the smallest package, VQ64, only supports the Master Serial, Slave Serial, and Express modes.A detailed description of each configuration mode, with timing information, is included later in this data sheet. During configuration, some of the I/O pins are used temporarily for the configuration process. All pins used during configuration are shown in Table 13 on page 124.

Master Modes

The three Master modes use an internal oscillator to generate a Configuration Clock (CCLK) for driving potential slave devices. They also generate address and timing for external PROM(s) containing the configuration data.

Master Parallel (Up or Down) modes generate the CCLK signal and PROM addresses and receive byte parallel data. The data is internally serialized into the FPGA data-frame format. The up and down selection generates starting addresses at either zero or 3FFFF, for compatibility with different microprocessor addressing conventions. The

Configuration Timing

The seven configuration modes are discussed in detail in this section. Timing specifications are included.

Slave Serial Mode

In Slave Serial mode, an external signal drives the CCLK input of the FPGA. The serial configuration bitstream must be available at the DIN input of the lead FPGA a short setup time before each rising CCLK edge.

The lead FPGA then presents the preamble data—and all data that overflows the lead device—on its DOUT pin.

There is an internal delay of 0.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge.

Figure 28 shows a full master/slave system. An XC5200-Series device in Slave Serial mode should be connected as shown in the third device from the left.

Slave Serial mode is selected by a <111> on the mode pins (M2, M1, M0). Slave Serial is the default mode if the mode pins are left unconnected, as they have weak pull-up resistors during configuration.

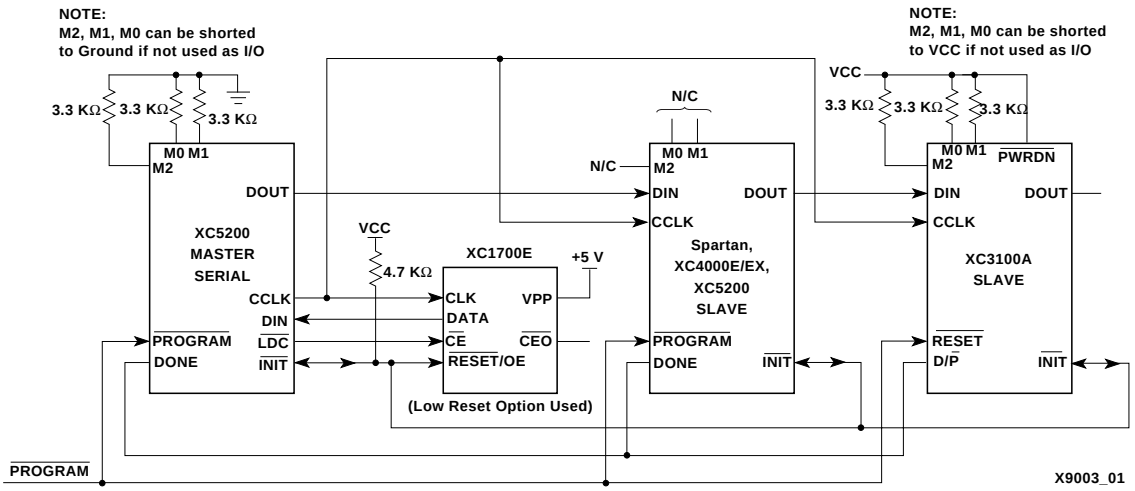
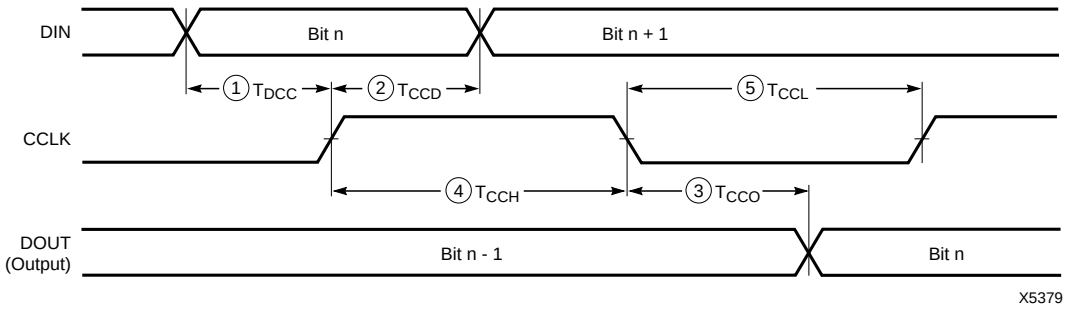


Figure 28: Master/Slave Serial Mode Circuit Diagram



	Description	Symbol		Min	Max	Units
CCLK	DIN setup	1	T_{DCC}	20		ns
	DIN hold	2	T_{CCD}	0		ns
	DIN to DOUT	3	T_{CCO}		30	ns
	High time	4	T_{CCH}	45		ns
	Low time	5	T_{CCL}	45		ns
	Frequency		F_{CC}		10	MHz

Note: Configuration must be delayed until the $\overline{INIT}$ pins of all daisy-chained FPGAs are High.

Figure 29: Slave Serial Mode Programming Switching Characteristics

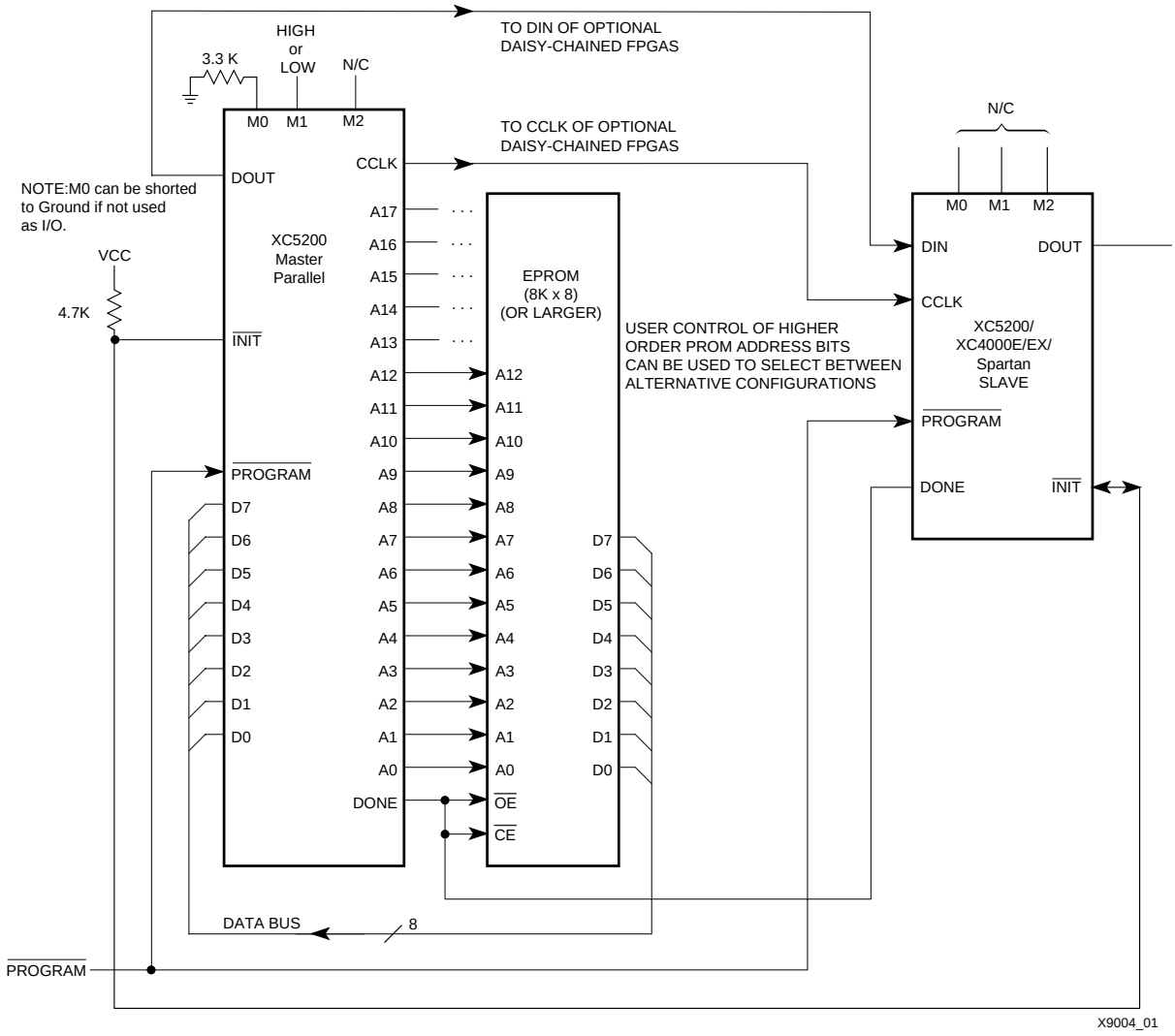
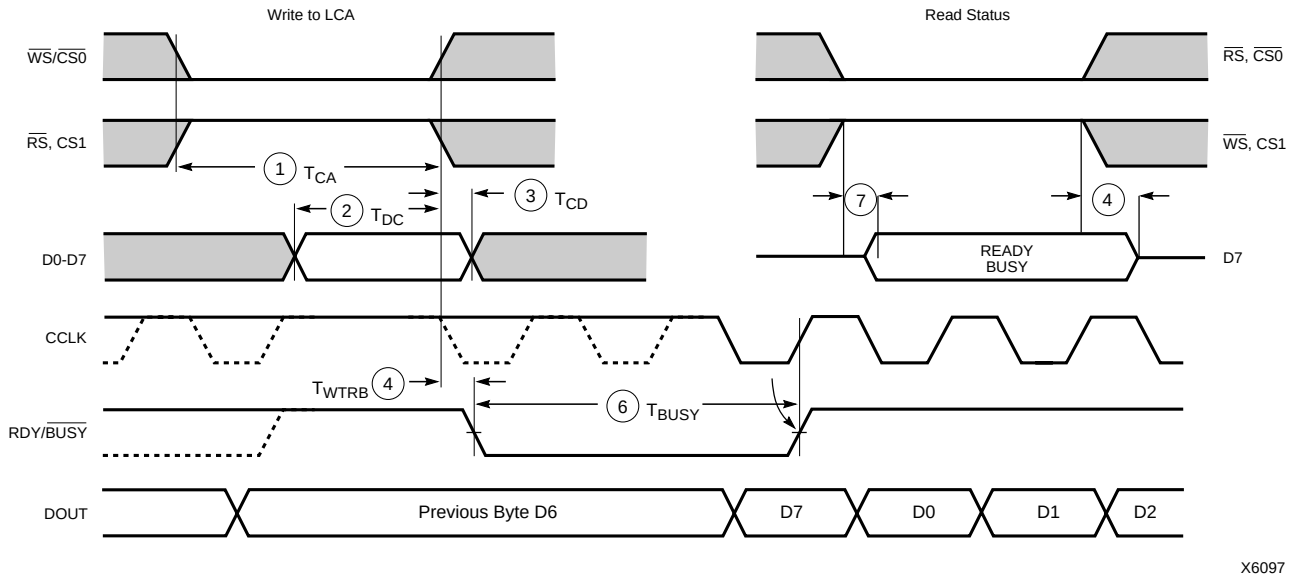


Figure 31: Master Parallel Mode Circuit Diagram

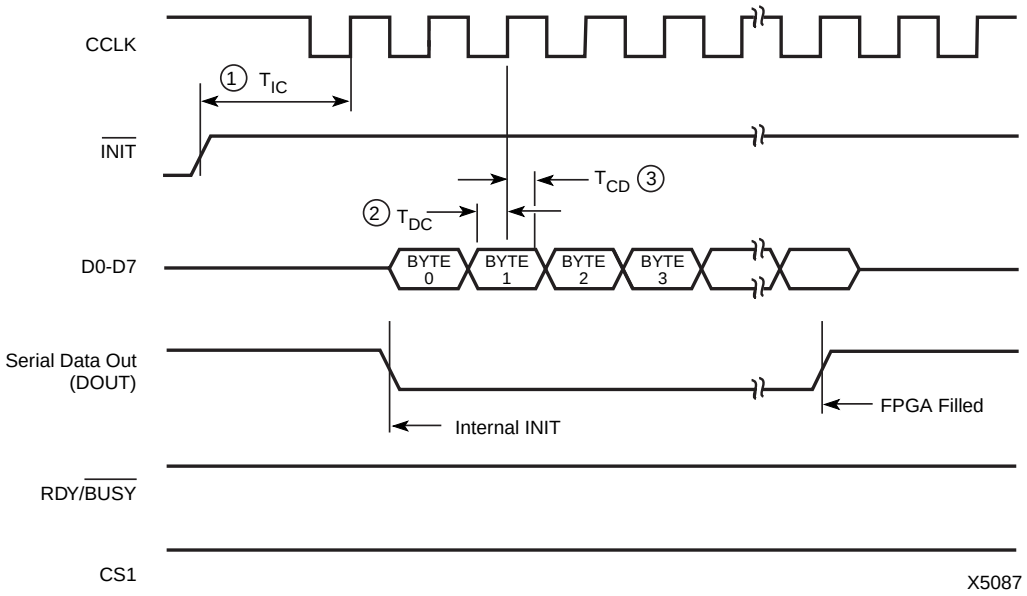


	Description	Symbol		Min	Max	Units
Write	Effective Write time ($\overline{CSO}$, $\overline{WS}$=Low; $\overline{RS}$, CS1=High	1	T_{CA}	100		ns
	DIN setup time	2	T_{DC}	60		ns
	DIN hold time	3	T_{CD}	0		ns
RDY	RDY/$\overline{BUSY}$ delay after end of Write or Read	4	T_{WTRB}		60	ns
	RDY/$\overline{BUSY}$ active after beginning of Read	7			60	ns
	RDY/$\overline{BUSY}$ Low output (Note 4)	6	T_{BUSY}	2	9	CCLK periods

Notes:
1. Configuration must be delayed until $\overline{INIT}$ pins of all daisy-chained FPGAs are high.
2. The time from the end of $\overline{WS}$ to CCLK cycle for the new byte of data depends on the completion of previous byte processing and the phase of internal timing generator for CCLK.
3. CCLK and DOUT timing is tested in slave mode.
4. T_{BUSY} indicates that the double-buffered parallel-to-serial converter is not yet ready to receive new data. The shortest T_{BUSY} occurs when a byte is loaded into an empty parallel-to-serial converter. The longest T_{BUSY} occurs when a new word is loaded into the input register before the second-level buffer has started shifting out data.

This timing diagram shows very relaxed requirements. Data need not be held beyond the rising edge of $\overline{WS}$. RDY/$\overline{BUSY}$ will go active within 60 ns after the end of $\overline{WS}$. A new write may be asserted immediately after RDY/$\overline{BUSY}$ goes Low, but write may not be terminated until RDY/$\overline{BUSY}$ has been High for one CCLK period.

Figure 36: Asynchronous Peripheral Mode Programming Switching Characteristics



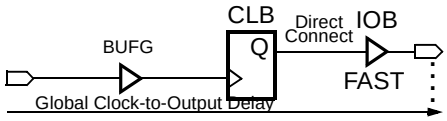
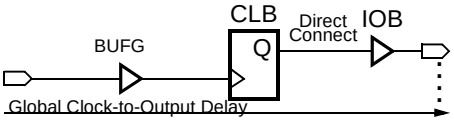
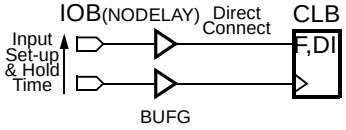
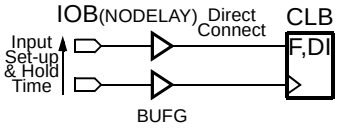
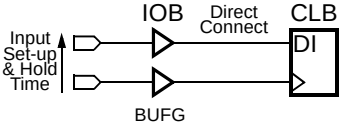
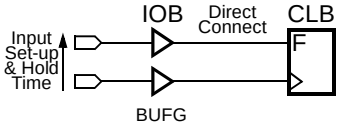
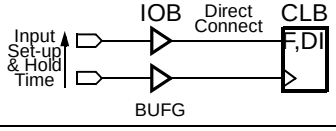
	Description	Symbol		Min	Max	Units
CCLK	INIT (High) Setup time required	1	T_{IC}	5		µs
	DIN Setup time required	2	T_{DC}	30		ns
	DIN hold time required	3	T_{CD}	0		ns
	CCLK High time		T_{CCH}	30		ns
	CCLK Low time		T_{CCL}	30		ns
	CCLK frequency		F_{CC}		10	MHz

Note: If not driven by the preceding DOUT, CS1 must remain high until the device is fully configured.

Figure 38: Express Mode Programming Switching Characteristics

XC5200 Guaranteed Input and Output Parameters (Pin-to-Pin)

All values listed below are tested directly, and guaranteed over the operating conditions. The same parameters can also be derived indirectly from the Global Buffer specifications. The delay calculator uses this indirect method, and may overestimate because of worst-case assumptions. When there is a discrepancy between these two methods, the values listed below should be used, and the derived values should be considered conservative overestimates.

		Speed Grade	-6	-5	-4	-3
Description	**Symbol**	**Device**	**Max (ns)**	**Max (ns)**	**Max (ns)**	**Max (ns)**
Global Clock to Output Pad (fast) 	T_{ICKOF} (Max)	XC5202	16.9	15.1	10.9	9.8
		XC5204	17.1	15.3	11.3	9.9
		XC5206	17.2	15.4	11.9	10.8
		XC5210	17.2	15.4	12.8	11.2
		XC5215	19.0	17.0	12.8	11.7
Global Clock to Output Pad (slew-limited) 	T_{ICKO} (Max)	XC5202	21.4	18.7	12.6	11.5
		XC5204	21.6	18.9	13.3	11.9
		XC5206	21.7	19.0	13.6	12.5
		XC5210	21.7	19.0	15.0	12.9
		XC5215	24.3	21.2	15.0	13.1
Input Set-up Time (no delay) to CLB Flip-Flop 	T_{PSUF} (Min)	XC5202	2.5	2.0	1.9	1.9
		XC5204	2.3	1.9	1.9	1.9
		XC5206	2.2	1.9	1.9	1.9
		XC5210	2.2	1.9	1.9	1.8
		XC5215	2.0	1.8	1.7	1.7
Input Hold Time (no delay) to CLB Flip-Flop 	T_{PHF} (Min)	XC5202	3.8	3.8	3.5	3.5
		XC5204	3.9	3.9	3.8	3.6
		XC5206	4.4	4.4	4.4	4.3
		XC5210	5.1	5.1	4.9	4.8
		XC5215	5.8	5.8	5.7	5.6
Input Set-up Time (with delay) to CLB Flip-Flop DI Input 	T_{PSU}	XC5202	7.3	6.6	6.6	6.6
		XC5204	7.3	6.6	6.6	6.6
		XC5206	7.2	6.5	6.4	6.3
		XC5210	7.2	6.5	6.0	6.0
		XC5215	6.8	5.7	5.7	5.7
Input Set-up Time (with delay) to CLB Flip-Flop F Input 	T_{PSUL} (Min)	XC5202	8.8	7.7	7.5	7.5
		XC5204	8.6	7.5	7.5	7.5
		XC5206	8.5	7.4	7.4	7.4
		XC5210	8.5	7.4	7.4	7.3
		XC5215	8.5	7.4	7.4	7.2
Input Hold Time (with delay) to CLB Flip-Flop 	T_{PH} (Min)	XC52xx	0	0	0	0

Note: 1. These measurements assume that the CLB flip-flop uses a direct interconnect to or from the IOB. The INREG/ OUTREG properties, or XACT-Performance, can be used to assure that direct connects are used. t_{PSU} applies only to the CLB input DI that bypasses the look-up table, which only offers direct connects to IOBs on the left and right edges of the die. t_{PSUL} applies to the CLB inputs F that feed the look-up table, which offers direct connect to IOBs on all four edges, as do the CLB Q outputs.

2. When testing outputs (fast or slew-limited), half of the outputs on one side of the device are switching.

Pin	Description	VQ64*	PC84	PQ100	VQ100	TQ144	PG156	Boundary Scan Order
	CCLK	48	73	77	74	107	R2	-
	VCC	-	74	78	75	108	P3	-
74.	I/O (TDO)	49	75	79	76	109	T1	0
	GND	-	76	80	77	110	N3	-
75.	I/O (A0, $\overline{WS}$)	50	77	81	78	111	R1	9
76.	GCK4 (A1, I/O)	51	78	82	79	112	P2	15
77.	I/O (A2, CS1)	52	79	83	80	115	P1	18
78.	I/O (A3)	-	80	84	81	116	N1	21
	GND	-	-	-	-	118	L3	-
79.	I/O (A4)	-	81	85	82	121	K3	27
80.	I/O (A5)	53	82	86	83	122	K2	30
81.	I/O	-	-	87	84	123	K1	33
82.	I/O	-	-	88	85	124	J1	39
83.	I/O (A6)	54	83	89	86	125	J2	42
84.	I/O (A7)	55	84	90	87	126	J3	45
	GND	56	1	91	88	127	H2	-

* VQ64 package supports Master Serial, Slave Serial, and Express configuration modes only.

Additional No Connect (N.C.) Connections on TQ144 Package

TQ144					
135	9	41	67	98	117
136	10	42	68	99	119
140	25	46	77	103	120
141	26	47	78	104	
4	30	62	82	113	
5	31	63	83	114	

Notes: Boundary Scan Bit 0 = TDO.T
Boundary Scan Bit 1 = TDO.O
Boundary Scan Bit 1056 = BSCAN.UPD

Pin Locations for XC5204 Devices

The following table may contain pinout information for unsupported device/package combinations. Please see the availability charts elsewhere in the XC5200 Series data sheet for availability information.

Pin	Description		PC84	PQ100	VQ100	TQ144	PG156	PQ160	Boundary Scan Order
	VCC		2	92	89	128	H3	142	-
1.	I/O (A8)		3	93	90	129	H1	143	78
2.	I/O (A9)		4	94	91	130	G1	144	81
3.	I/O		-	95	92	131	G2	145	87
4.	I/O		-	96	93	132	G3	146	90
5.	I/O (A10)		5	97	94	133	F1	147	93
6.	I/O (A11)		6	98	95	134	F2	148	99
7.	I/O		-	-	-	135	E1	149	102
8.	I/O		-	-	-	136	E2	150	105
	GND		-	-	-	137	F3	151	-
9.	I/O		-	-	-	-	D1	152	111
10.	I/O		-	-	-	-	D2	153	114
11.	I/O (A12)		7	99	96	138	E3	154	117
12.	I/O (A13)		8	100	97	139	C1	155	123
13.	I/O		-	-	-	140	C2	156	126

Pin Locations for XC5206 Devices

The following table may contain pinout information for unsupported device/package combinations. Please see the availability charts elsewhere in the XC5200 Series data sheet for availability information.

Pin	Description	PC84	PQ100	VQ100	TQ144	PQ160	TQ176	PG191	PQ208	Boundary Scan Order
	VCC	2	92	89	128	142	155	J4	183	-
1.	I/O (A8)	3	93	90	129	143	156	J3	184	87
2.	I/O (A9)	4	94	91	130	144	157	J2	185	90
3.	I/O	-	95	92	131	145	158	J1	186	93
4.	I/O	-	96	93	132	146	159	H1	187	99
5.	I/O	-	-	-	-	-	160	H2	188	102
6.	I/O	-	-	-	-	-	161	H3	189	105
7.	I/O (A10)	5	97	94	133	147	162	G1	190	111
8.	I/O (A11)	6	98	95	134	148	163	G2	191	114
9.	I/O	-	-	-	135	149	164	F1	192	117
10.	I/O	-	-	-	136	150	165	E1	193	123
	GND	-	-	-	137	151	166	G3	194	-
11.	I/O	-	-	-	-	152	168	C1	197	126
12.	I/O	-	-	-	-	153	169	E2	198	129
13.	I/O (A12)	7	99	96	138	154	170	F3	199	138
14.	I/O (A13)	8	100	97	139	155	171	D2	200	141
15.	I/O	-	-	-	140	156	172	B1	201	150
16.	I/O	-	-	-	141	157	173	E3	202	153
17.	I/O (A14)	9	1	98	142	158	174	C2	203	162
18.	I/O (A15)	10	2	99	143	159	175	B2	204	165
	VCC	11	3	100	144	160	176	D3	205	-
	GND	12	4	1	1	1	1	D4	2	-
19.	GCK1 (A16, I/O)	13	5	2	2	2	2	C3	4	174
20.	I/O (A17)	14	6	3	3	3	3	C4	5	177
21.	I/O	-	-	-	4	4	4	B3	6	183
22.	I/O	-	-	-	5	5	5	C5	7	186
23.	I/O (TDI)	15	7	4	6	6	6	A2	8	189
24.	I/O (TCK)	16	8	5	7	7	7	B4	9	195
25.	I/O	-	-	-	-	8	8	C6	10	198
26.	I/O	-	-	-	-	9	9	A3	11	201
	GND	-	-	-	8	10	10	C7	14	-
27.	I/O	-	-	-	9	11	11	A4	15	207
28.	I/O	-	-	-	10	12	12	A5	16	210
29.	I/O (TMS)	17	9	6	11	13	13	B7	17	213
30.	I/O	18	10	7	12	14	14	A6	18	219
31.	I/O	-	-	-	-	-	15	C8	19	222
32.	I/O	-	-	-	-	-	16	A7	20	225
33.	I/O	-	-	-	13	15	17	B8	21	234
34.	I/O	-	11	8	14	16	18	A8	22	237
35.	I/O	19	12	9	15	17	19	B9	23	246
36.	I/O	20	13	10	16	18	20	C9	24	249
	GND	21	14	11	17	19	21	D9	25	-
	VCC	22	15	12	18	20	22	D10	26	-
37.	I/O	23	16	13	19	21	23	C10	27	255
38.	I/O	24	17	14	20	22	24	B10	28	258
39.	I/O	-	18	15	21	23	25	A9	29	261
40.	I/O	-	-	-	22	24	26	A10	30	267
41.	I/O	-	-	-	-	-	27	A11	31	270

Pin	Description	PC84	TQ144	PQ160	TQ176	PQ208	PG223	BG225	PQ240	Boundary Scan Order
95.	I/O	-	-	-	-	-	H15	P7	85	519
96.	I/O	-	50	56	62	74	H18	R7	86	522
97.	I/O	-	51	57	63	75	J18	L7	87	528
98.	I/O	40	52	58	64	76	J17	N8	88	531
99.	I/O ($\overline{\text{ERR}}$, $\overline{\text{INIT}}$)	41	53	59	65	77	J16	P8	89	534
	VCC	42	54	60	66	78	J15	VCC*	90	-
	GND	43	55	61	67	79	K15	GND*	91	-
100.	I/O	44	56	62	68	80	K16	L8	92	540
101.	I/O	45	57	63	69	81	K17	P9	93	543
102.	I/O	-	58	64	70	82	K18	R9	94	546
103.	I/O	-	59	65	71	83	L18	N9	95	552
104.	I/O	-	-	-	72	84	L17	M9	96	555
105.	I/O	-	-	-	73	85	L16	L9	97	558
106.	I/O	-	-	-	-	-	L15	R10	99	564
107.	I/O	-	-	-	-	-	M15	P10	100	567
	VCC	-	-	-	-	-	-	VCC*	101	-
108.	I/O	46	60	66	74	86	M18	N10	102	570
109.	I/O	47	61	67	75	87	M17	K9	103	576
110.	I/O	-	62	68	76	88	N18	R11	104	579
111.	I/O	-	63	69	77	89	P18	P11	105	588
	GND	-	64	70	78	90	M16	GND*	106	-
112.	I/O	-	-	-	-	-	N15	M10	107	591
113.	I/O	-	-	-	-	-	P15	N11	108	600
114.	I/O	-	-	-	-	91	N17	R12	109	603
115.	I/O	-	-	-	-	92	R18	L10	110	606
116.	I/O	-	-	71	79	93	T18	P12	111	612
117.	I/O	-	-	72	80	94	P17	M11	112	615
118.	I/O	48	65	73	81	95	N16	R13	113	618
119.	I/O	49	66	74	82	96	T17	N12	114	624
120.	I/O	-	67	75	83	97	R17	P13	115	627
121.	I/O	-	68	76	84	98	P16	K10	116	630
122.	I/O	50	69	77	85	99	U18	R14	117	636
123.	I/O	51	70	78	86	100	T16	N13	118	639
	GND	52	71	79	87	101	R16	GND*	119	-
	DONE	53	72	80	88	103	U17	P14	120	-
	VCC	54	73	81	89	106	R15	VCC*	121	-
	PROG	55	74	82	90	108	V18	M12	122	-
124.	I/O (D7)	56	75	83	91	109	T15	P15	123	648
125.	GCK3 (I/O)	57	76	84	92	110	U16	N14	124	651
126.	I/O	-	77	85	93	111	T14	L11	125	660
127.	I/O	-	78	86	94	112	U15	M13	126	663
128.	I/O	-	-	-	-	-	R14	N15	127	666
129.	I/O	-	-	-	-	-	R13	M14	128	672
130.	I/O (D6)	58	79	87	95	113	V17	J10	129	675
131.	I/O	-	80	88	96	114	V16	L12	130	678
132.	I/O	-	-	89	97	115	T13	M15	131	684
133.	I/O	-	-	90	98	116	U14	L13	132	687
134.	I/O	-	-	-	-	117	V15	L14	133	690
135.	I/O	-	-	-	-	118	V14	K11	134	696
	GND	-	81	91	99	119	T12	GND*	135	-
136.	I/O	-	-	-	-	-	R12	L15	136	699

Pin	Description	PC84	TQ144	PQ160	TQ176	PQ208	PG223	BG225	PQ240	Boundary Scan Order
137.	I/O	-	-	-	-	-	R11	K12	137	708
138.	I/O	-	82	92	100	120	U13	K13	138	711
139.	I/O	-	83	93	101	121	V13	K14	139	714
	VCC	-	-	-	-	-	-	VCC*	140	-
140.	I/O (D5)	59	84	94	102	122	U12	K15	141	720
141.	I/O ($\overline{CS0}$)	60	85	95	103	123	V12	J12	142	723
142.	I/O	-	-	-	104	124	T11	J13	144	726
143.	I/O	-	-	-	105	125	U11	J14	145	732
144.	I/O	-	86	96	106	126	V11	J15	146	735
145.	I/O	-	87	97	107	127	V10	J11	147	738
146.	I/O (D4)	61	88	98	108	128	U10	H13	148	744
147.	I/O	62	89	99	109	129	T10	H14	149	747
	VCC	63	90	100	110	130	R10	VCC*	150	-
	GND	64	91	101	111	131	R9	GND*	151	-
148.	I/O (D3)	65	92	102	112	132	T9	H12	152	756
149.	I/O ($\overline{RS}$)	66	93	103	113	133	U9	H11	153	759
150.	I/O	-	94	104	114	134	V9	G14	154	768
151.	I/O	-	95	105	115	135	V8	G15	155	771
152.	I/O	-	-	-	116	136	U8	G13	156	780
153.	I/O	-	-	-	117	137	T8	G12	157	783
154.	I/O (D2)	67	96	106	118	138	V7	G11	159	786
155.	I/O	68	97	107	119	139	U7	F15	160	792
	VCC	-	-	-	-	-	-	VCC*	161	-
156.	I/O	-	98	108	120	140	V6	F14	162	795
157.	I/O	-	99	109	121	141	U6	F13	163	798
158.	I/O	-	-	-	-	-	R8	G10	164	804
159.	I/O	-	-	-	-	-	R7	E15	165	807
	GND	-	100	110	122	142	T7	GND*	166	-
160.	I/O	-	-	-	-	-	R6	E14	167	810
161.	I/O	-	-	-	-	-	R5	F12	168	816
162.	I/O	-	-	-	-	143	V5	E13	169	819
163.	I/O	-	-	-	-	144	V4	D15	170	822
164.	I/O	-	-	111	123	145	U5	F11	171	828
165.	I/O	-	-	112	124	146	T6	D14	172	831
166.	I/O (D1)	69	101	113	125	147	V3	E12	173	834
167.	I/O ($\overline{RCLK}$-$\overline{BUSY}$/RDY)	70	102	114	126	148	V2	C15	174	840
168.	I/O	-	103	115	127	149	U4	D13	175	843
169.	I/O	-	104	116	128	150	T5	C14	176	846
170.	I/O (D0, DIN)	71	105	117	129	151	U3	F10	177	855
171.	I/O (DOUT)	72	106	118	130	152	T4	B15	178	858
	CCLK	73	107	119	131	153	V1	C13	179	-
	VCC	74	108	120	132	154	R4	VCC*	180	-
172.	I/O (TDO)	75	109	121	133	159	U2	A15	181	-
	GND	76	110	122	134	160	R3	GND*	182	-
173.	I/O (A0, $\overline{WS}$)	77	111	123	135	161	T3	A14	183	9
174.	GCK4 (A1, I/O)	78	112	124	136	162	U1	B13	184	15
175.	I/O	-	113	125	137	163	P3	E11	185	18
176.	I/O	-	114	126	138	164	R2	C12	186	21
177.	I/O (CS1, A2)	79	115	127	139	165	T2	A13	187	27
178.	I/O (A3)	80	116	128	140	166	N3	B12	188	30
179.	I/O	-	-	-	-	-	P4	F9	189	33

Pin	Description	PC84	TQ144	PQ160	TQ176	PQ208	PG223	BG225	PQ240	Boundary Scan Order
180.	I/O	-	-	-	-	-	N4	D11	190	39
181.	I/O	-	117	129	141	167	P2	A12	191	42
182.	I/O	-	-	130	142	168	T1	C11	192	45
183.	I/O	-	-	-	-	169	R1	B11	193	51
184.	I/O	-	-	-	-	170	N2	E10	194	54
	-	-	-	-	-	-	-	GND*		-
	GND	-	118	131	143	171	M3	-	196	-
185.	I/O	-	119	132	144	172	P1	A11	197	57
186.	I/O	-	120	133	145	173	N1	D10	198	66
187.	I/O	-	-	-	-	-	M4	C10	199	69
188.	I/O	-	-	-	-	-	L4	B10	200	75
	VCC	-	-	-	-	-	-	VCC*	201	-
189.	I/O (A4)	81	121	134	146	174	M2	A10	202	78
190.	I/O (A5)	82	122	135	147	175	M1	D9	203	81
191.	I/O	-	-	-	148	176	L3	C9	205	87
192.	I/O	-	-	136	149	177	L2	B9	206	90
193.	I/O	-	123	137	150	178	L1	A9	207	93
194.	I/O	-	124	138	151	179	K1	E9	208	99
195.	I/O (A6)	83	125	139	152	180	K2	C8	209	102
196.	I/O (A7)	84	126	140	153	181	K3	B8	210	105
	GND	1	127	141	154	182	K4	GND*	211	-

Additional No Connect (N.C.) Connections for PQ208 and PQ240 Packages

PQ208					PQ240		
1	53	105	157	208	22	143	219
3	54	107	158		37	158	
51	102	155	206		83	195	
52	104	156	207		98	204	

Notes: * Pins labeled VCC* are internally bonded to a VCC plane within the BG225 package. The external pins are: B2, D8, H15, R8, B14, R1, H1, and R15.
Pins labeled GND* are internally bonded to a ground plane within the BG225 package. The external pins are: A1, D12, G7, G9, H6, H8, H10, J8, K8, A8, F8, G8, H2, H7, H9, J7, J9, M8.
Boundary Scan Bit 0 = TDO.T
Boundary Scan Bit 1 = TDO.O
Boundary Scan Bit 1056 = BSCAN.UPD

Pin Locations for XC5215 Devices

The following table may contain pinout information for unsupported device/package combinations. Please see the availability charts elsewhere in the XC5200 Series data sheet for availability information.

Pin	Description	PQ160	HQ208	HQ240	PG299	BG225	BG352	Boundary Scan Order
	VCC	142	183	212	K1	VCC*	VCC*	-
1.	I/O (A8)	143	184	213	K2	E8	D14	138
2.	I/O (A9)	144	185	214	K3	B7	C14	141
3.	I/O	145	186	215	K5	A7	A15	147
4.	I/O	146	187	216	K4	C7	B15	150
5.	I/O	-	188	217	J1	D7	C15	153
6.	I/O	-	189	218	J2	E7	D15	159
7.	I/O (A10)	147	190	220	H1	A6	A16	162

Product Availability

	PINS	64	84	100	100	144	156	160	176	191	208	208	223	225	240	240	299	352
	TYPE	Plast. VQFP	Plast. PLCC	Plast. PQFP	Plast. VQFP	Plast. TQFP	Ceram. PGA	Plast. PQFP	Plast. TQFP	Ceram. PGA	High-Perf. QFP	Plast. PQFP	Ceram. PGA	Plast. BGA	High-Perf. QFP	Plast. PQFP	Ceram. PGA	Plast. BGA
	CODE	VQ64*	PC84	PQ100	VQ100	TQ144	PG156	PQ160	TQ176	PG191	HQ208	PQ208	PG223	BG225	HQ240	PQ240	PG299	BG352
XC5202	-6	CI	CI	CI	CI	CI	CI											
	-5	CI	CI	CI	CI	CI	CI											
	-4	C	C	C	C	C	C											
	-3	C	C	C	C	C	C											
XC5204	-6		CI	CI	CI	CI	CI	CI										
	-5		CI	CI	CI	CI	CI	CI										
	-4		C	C	C	C	C	C										
	-3		C	C	C	C	C	C										
XC5206	-6		CI	CI	CI	CI		CI	CI	CI		CI						
	-5		CI	CI	CI	CI		CI	CI	CI		CI						
	-4		C	C	C	C		C	C	C		C						
	-3		C	C	C	C		C	C	C		C						
XC5210	-6		CI			CI		CI	CI			CI	CI	CI		CI		
	-5		CI			CI		CI	CI			CI	CI	CI		CI		
	-4		C			C		C	C			C	C	C		C		
	-3		C			C		C	C			C	C	C		C		
XC5215	-6							CI			CI			CI	CI		CI	CI
	-5							C			C			C	C		C	C
	-4							C			C			C	C		C	C
	-3							C			C			C	C		C	C

7/8/98

C = Commercial T_J = 0° to +85°C

I= Industrial T_J = -40°C to +100°C

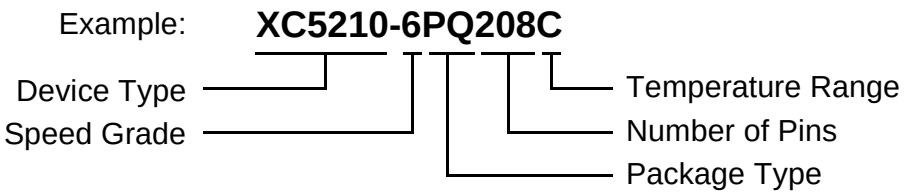
* VQ64 package supports Master Serial, Slave Serial, and Express configuration modes only.

User I/O Per Package

Device	Max I/O	Package Type																
		VQ64	PC84	PQ100	VQ100	TQ144	PG156	PQ160	TQ176	PG191	HQ208	PQ208	PG223	BG225	HQ240	PQ240	PG299	BG352
XC5202	84	52	65	81	81	84	84											
XC5204	124		65	81	81	117	124	124										
XC5206	148		65	81	81	117		133	148	148		148						
XC5210	196		65			117		133	149			164	196	196		196		
XC5215	244							133			164			196	197		244	244

7/8/98

Ordering Information



Revisions

Version	Description
12/97	Rev 5.0 added -3, -4 specification
7/98	Rev 5.1 added Spartan family to comparison, removed HQ304
11/98	Rev 5.2 All specifications made final.