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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	324
Number of Logic Elements/Cells	1296
Total RAM Bits	-
Number of I/O	65
Number of Gates	16000
Voltage - Supply	4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	84-LCC (J-Lead)
Supplier Device Package	84-PLCC (29.31x29.31)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc5210-5pc84c

XC3000 family: XC5200 devices support an additional programming mode: Peripheral Synchronous.

XC3000 family: The XC5200 family does not support Power-down, but offers a Global 3-state input that does not reset any flip-flops.

XC3000 family: The XC5200 family does not provide an on-chip crystal oscillator amplifier, but it does provide an internal oscillator from which a variety of frequencies up to 12 MHz are available.

Architectural Overview

Figure 1 presents a simplified, conceptual overview of the XC5200 architecture. Similar to conventional FPGAs, the XC5200 family consists of programmable IOBs, programmable logic blocks, and programmable interconnect. Unlike other FPGAs, however, the logic and local routing resources of the XC5200 family are combined in flexible VersaBlocks (Figure 2). General-purpose routing connects to the VersaBlock through the General Routing Matrix (GRM).

VersaBlock: Abundant Local Routing Plus Versatile Logic

The basic logic element in each VersaBlock structure is the Logic Cell, shown in Figure 3. Each LC contains a 4-input function generator (F), a storage device (FD), and control logic. There are five independent inputs and three outputs to each LC. The independence of the inputs and outputs allows the software to maximize the resource utilization within each LC. Each Logic Cell also contains a direct feedthrough path that does not sacrifice the use of either the function generator or the register; this feature is a first for FPGAs. The storage device is configurable as either a D flip-flop or a latch. The control logic consists of carry logic for fast implementation of arithmetic functions, which can also be configured as a cascade chain allowing decode of very wide input functions.

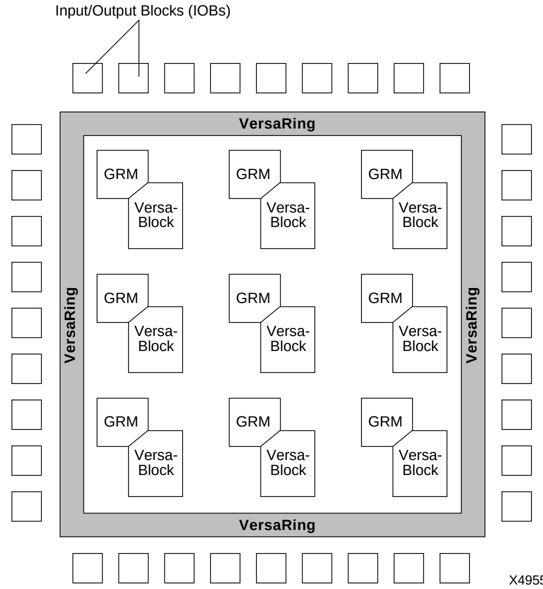

Figure 1: XC5200 Architectural Overview

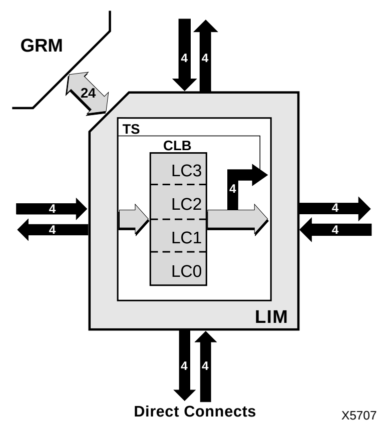

Figure 2: VersaBlock

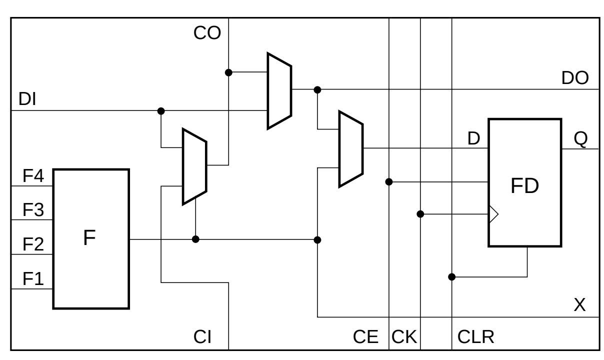

Figure 3: XC5200 Logic Cell (Four LCs per CLB)

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tomized RPMs, freeing the designer from the need to become an expert on architectures.

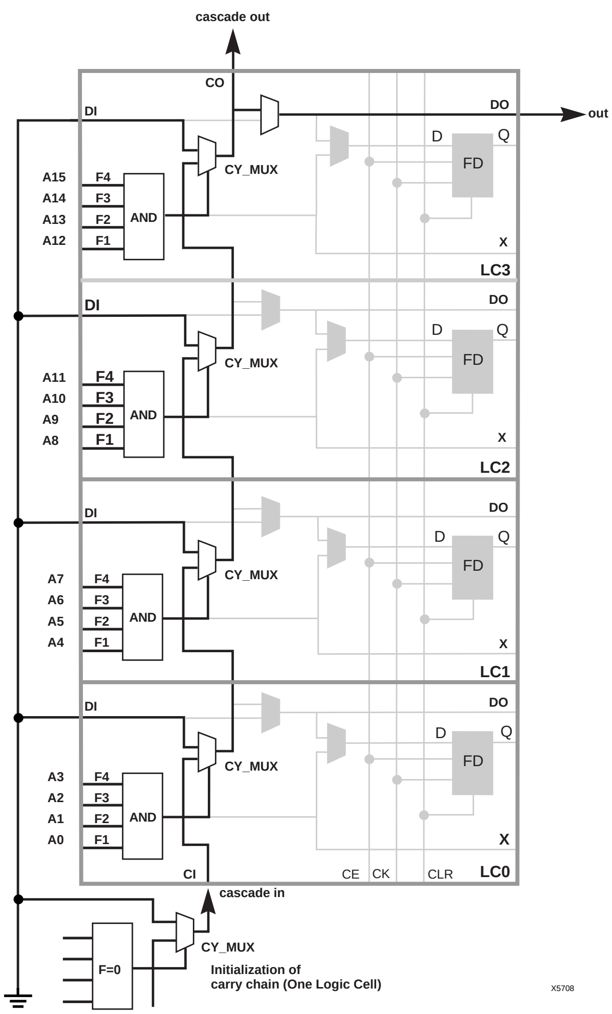


Figure 7: XC5200 CY_MUX Used for Decoder Cascade Logic

Cascade Function

Each CY_MUX can be connected to the CY_MUX in the adjacent LC to provide cascadable decode logic. Figure 7 illustrates how the 4-input function generators can be configured to take advantage of these four cascaded CY_MUXes. Note that AND and OR cascading are specific cases of a general decode. In AND cascading all bits are decoded equal to logic one, while in OR cascading all bits are decoded equal to logic zero. The flexibility of the LUT achieves this result. The XC5200 library contains gate macros designed to take advantage of this function.

CLB Flip-Flops and Latches

The CLB can pass the combinatorial output(s) to the interconnect network, but can also store the combinatorial results or other incoming data in flip-flops, and connect their outputs to the interconnect network as well. The CLB storage elements can also be configured as latches.

Table 3: CLB Storage Element Functionality (active rising edge is shown)

Mode	CK	CE	CLR	D	Q
Power-Up or GR	X	X	X	X	0
Flip-Flop	X	X	1	X	0
	__/	1*	0*	D	D
	0	X	0*	X	Q
Latch	1	1*	0*	X	Q
	0	1*	0*	D	D
Both	X	0	0*	X	Q

Legend:

X	Don't care
__/	Rising edge
0*	Input is Low or unconnected (default value)
1*	Input is High or unconnected (default value)

Data Inputs and Outputs

The source of a storage element data input is programmable. It is driven by the function F, or by the Direct In (DI) block input. The flip-flops or latches drive the Q CLB outputs.

Four fast feed-through paths from DI to DO are available, as shown in Figure 4. This bypass is sometimes used by the automated router to repower internal signals. In addition to the storage element (Q) and direct (DO) outputs, there is a combinatorial output (X) that is always sourced by the Lookup Table.

The four edge-triggered D-type flip-flops or level-sensitive latches have common clock (CK) and clock enable (CE) inputs. Any of the clock inputs can also be permanently enabled. Storage element functionality is described in Table 3.

Clock Input

The flip-flops can be triggered on either the rising or falling clock edge. The clock pin is shared by all four storage elements with individual polarity control. Any inverter placed on the clock input is automatically absorbed into the CLB.

Clock Enable

The clock enable signal (CE) is active High. The CE pin is shared by the four storage elements. If left unconnected for any, the clock enable for that storage element defaults to the active state. CE is not invertible within the CLB.

Clear

An asynchronous storage element input (CLR) can be used to reset all four flip-flops or latches in the CLB. This input can also be independently disabled for any flip-flop. CLR is active High. It is not invertible within the CLB.

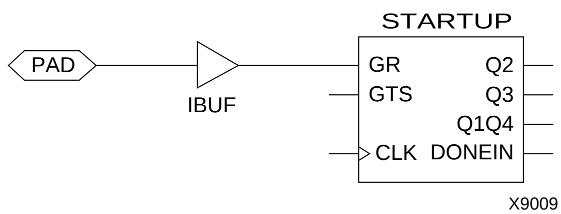


Figure 8: Schematic Symbols for Global Reset

Global Reset

A separate Global Reset line clears each storage element during power-up, reconfiguration, or when a dedicated Reset net is driven active. This global net (GR) does not compete with other routing resources; it uses a dedicated distribution network.

GR can be driven from any user-programmable pin as a global reset input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GR pin of the STARTUP symbol. (See Figure 9.) A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the Global Reset signal. Alternatively, GR can be driven from any internal node.

Using FPGA Flip-Flops and Latches

The abundance of flip-flops in the XC5200 Series invites pipelined designs. This is a powerful way of increasing performance by breaking the function into smaller subfunctions and executing them in parallel, passing on the results through pipeline flip-flops. This method should be seriously considered wherever throughput is more important than latency.

To include a CLB flip-flop, place the appropriate library symbol. For example, FDCE is a D-type flip-flop with clock enable and asynchronous clear. The corresponding latch symbol is called LDCE.

In XC5200-Series devices, the flip-flops can be used as registers or shift registers without blocking the function generators from performing a different, perhaps unrelated task. This ability increases the functional capacity of the devices.

The CLB setup time is specified between the function generator inputs and the clock input CK. Therefore, the specified CLB flip-flop setup time includes the delay through the function generator.

Three-State Buffers

The XC5200 family has four dedicated Three-State Buffers (TBUFs, or BUFTs in the schematic library) per CLB (see Figure 9). The four buffers are individually configurable through four configuration bits to operate as simple non-inverting buffers or in 3-state mode. When in 3-state mode the CLB output enable (TS) control signal drives the enable to all four buffers. Each TBUF can drive up to two horizontal and/or two vertical Longlines. These 3-state buffers can be used to implement multiplexed or bidirectional buses on the horizontal or vertical longlines, saving logic resources.

The 3-state buffer enable is an active-High 3-state (i.e. an active-Low enable), as shown in Table 4.

Table 4: Three-State Buffer Functionality

IN	T	OUT
X	1	Z
IN	0	IN

Another 3-state buffer with similar access is located near each I/O block along the right and left edges of the array.

The longlines driven by the 3-state buffers have a weak keeper at each end. This circuit prevents undefined floating levels. However, it is overridden by any driver. To ensure the longline goes high when no buffers are on, add an additional BUFT to drive the output High during all of the previously undefined states.

Figure 10 shows how to use the 3-state buffers to implement a multiplexer. The selection is accomplished by the buffer 3-state signal.

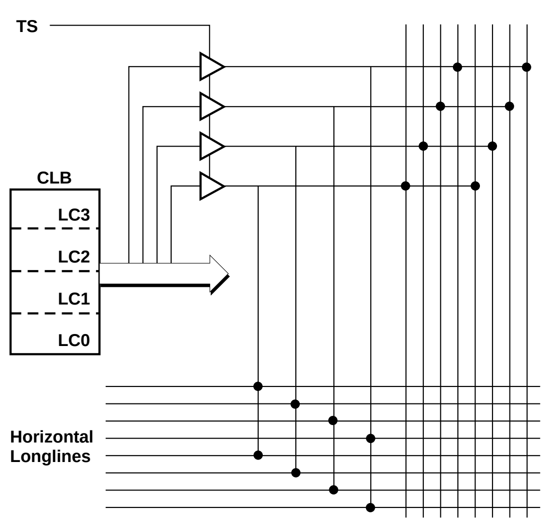


Figure 9: XC5200 3-State Buffers

to Vcc. The configurable pull-down resistor is an n-channel transistor that pulls to Ground.

The value of these resistors is 20 kΩ – 100 kΩ. This high value makes them unsuitable as wired-AND pull-up resistors.

The pull-up resistors for most user-programmable IOBs are active during the configuration process. See Table 13 on page 124 for a list of pins with pull-ups active before and during configuration.

After configuration, voltage levels of unused pads, bonded or unbonded, must be valid logic levels, to reduce noise sensitivity and avoid excess current. Therefore, by default, unused pads are configured with the internal pull-up resistor active. Alternatively, they can be individually configured with the pull-down resistor, or as a driven output, or to be driven by an external source. To activate the internal pull-up, attach the PULLUP library component to the net attached to the pad. To activate the internal pull-down, attach the PULLDOWN library component to the net attached to the pad.

JTAG Support

Embedded logic attached to the IOBs contains test structures compatible with IEEE Standard 1149.1 for boundary scan testing, simplifying board-level testing. More information is provided in "Boundary Scan" on page 98.

Oscillator

XC5200 devices include an internal oscillator. This oscillator is used to clock the power-on time-out, clear configuration memory, and source CCLK in Master configuration modes. The oscillator runs at a nominal 12 MHz frequency that varies with process, Vcc, and temperature. The output CCLK frequency is selectable as 1 MHz (default), 6 MHz, or 12 MHz.

The XC5200 oscillator divides the internal 12-MHz clock or a user clock. The user then has the choice of dividing by 4, 16, 64, or 256 for the "OSC1" output and dividing by 2, 8, 32, 128, 1024, 4096, 16384, or 65536 for the "OSC2" output. The division is specified via a "DIVIDEn_BY=x" attribute on the symbol, where n=1 for OSC1, or n=2 for OSC2. These frequencies can vary by as much as -50% or + 50%.

The OSC5 macro is used where an internal oscillator is required. The CK_DIV macro is applicable when a user clock input is specified (see Figure 13).

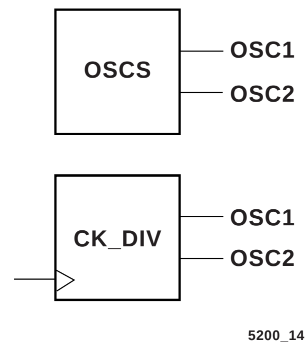


Figure 13: XC5200 Oscillator Macros

VersaBlock Routing

The General Routing Matrix (GRM) connects to the Versa-Block via 24 bidirectional ports (M0-M23). Excluding direct connections, global nets, and 3-statable Longlines, all VersaBlock inputs and outputs connect to the GRM via these 24 ports. Four 3-statable unidirectional signals (TQ0-TQ3) drive out of the VersaBlock directly onto the horizontal and vertical Longlines. Two horizontal global nets and two vertical global nets connect directly to every CLB clock pin; they can connect to other CLB inputs via the GRM. Each CLB also has four unidirectional direct connects to each of its four neighboring CLBs. These direct connects can also feed directly back to the CLB (see Figure 14).

In addition, each CLB has 16 direct inputs, four direct connections from each of the neighboring CLBs. These direct connections provide high-speed local routing that bypasses the GRM.

Local Interconnect Matrix

The Local Interconnect Matrix (LIM) is built from input and output multiplexers. The 13 CLB outputs (12 LC outputs plus a V_{CC}/GND signal) connect to the eight VersaBlock outputs via the output multiplexers, which consist of eight fully populated 13-to-1 multiplexers. Of the eight VersaBlock outputs, four signals drive each neighboring CLB directly, and provide a direct feedback path to the input multiplexers. The four remaining multiplexer outputs can drive the GRM through four TBUFs (TQ0-TQ3). All eight multiplexer outputs can connect to the GRM through the bidirectional M0-M23 signals. All eight signals also connect to the input multiplexers and are potential inputs to that CLB.

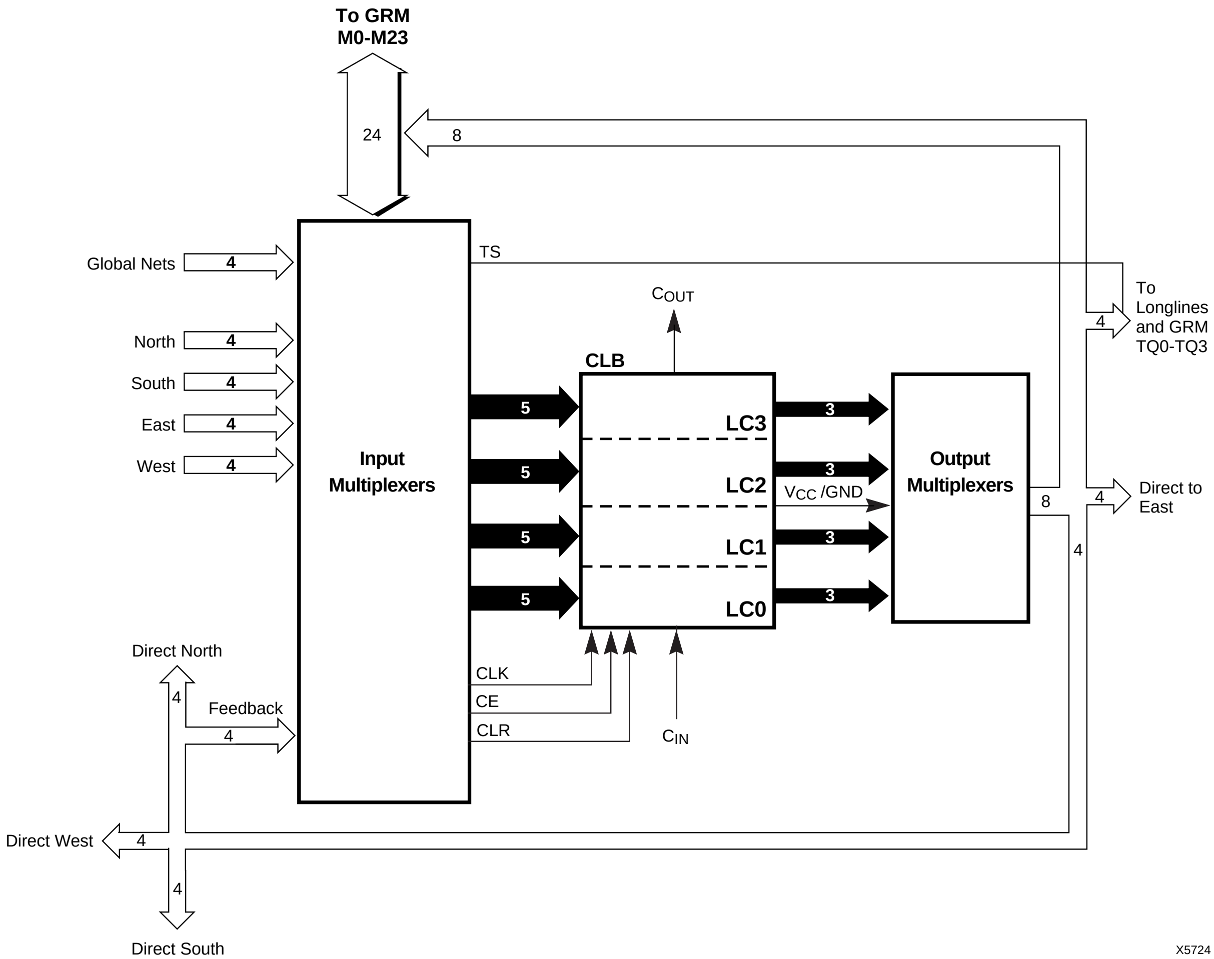


Figure 14: VersaBlock Details

CLB inputs have several possible sources: the 24 signals from the GRM, 16 direct connections from neighboring VersaBlocks, four signals from global, low-skew buffers, and the four signals from the CLB output multiplexers. Unlike the output multiplexers, the input multiplexers are not fully populated; i.e., only a subset of the available signals can be connected to a given CLB input. The flexibility of LUT input swapping and LUT mapping compensates for this limitation. For example, if a 2-input NAND gate is required, it can be mapped into any of the four LUTs, and use any two of the four inputs to the LUT.

Direct Connects

The unidirectional direct-connect segments are connected to the logic input/output pins through the CLB input and output multiplexer arrays, and thus bypass the general routing matrix altogether. These lines increase the routing channel utilization, while simultaneously reducing the delay incurred in speed-critical connections.

The direct connects also provide a high-speed path from the edge CLBs to the VersaRing input/output buffers, and thus reduce pin-to-pin set-up time, clock-to-out, and combinational propagation delay. Direct connects from the input buffers to the CLB DI pin (direct flip-flop input) are only available on the left and right edges of the device. CLB look-up table inputs and combinatorial/registered outputs have direct connects to input/output buffers on all four sides.

The direct connects are ideal for developing customized RPM cells. Using direct connects improves the macro performance, and leaves the other routing channels intact for improved routing. Direct connects can also route through a CLB using one of the four cell-feedthrough paths.

General Routing Matrix

The General Routing Matrix, shown in Figure 15, provides flexible bidirectional connections to the Local Interconnect

Table 9: Pin Descriptions (Continued)

Pin Name	I/O During Config.	I/O After Config.	Pin Description
Unrestricted User-Programmable I/O Pins			
I/O	Weak Pull-up	I/O	These pins can be configured to be input and/or output after configuration is completed. Before configuration is completed, these pins have an internal high-value pull-up resistor (20 kΩ - 100 kΩ) that defines the logic level as High.

Configuration

Configuration is the process of loading design-specific programming data into one or more FPGAs to define the functional operation of the internal blocks and their interconnections. This is somewhat like loading the command registers of a programmable peripheral chip. XC5200-Series devices use several hundred bits of configuration data per CLB and its associated interconnects. Each configuration bit defines the state of a static memory cell that controls either a function look-up table bit, a multiplexer input, or an interconnect pass transistor. The development system translates the design into a netlist file. It automatically partitions, places and routes the logic and generates the configuration data in PROM format.

Special Purpose Pins

Three configuration mode pins (M2, M1, M0) are sampled prior to configuration to determine the configuration mode. After configuration, these pins can be used as auxiliary I/O connections. The development system does not use these resources unless they are explicitly specified in the design entry. This is done by placing a special pad symbol called MD2, MD1, or MD0 instead of the input or output pad symbol.

In XC5200-Series devices, the mode pins have weak pull-up resistors during configuration. With all three mode pins High, Slave Serial mode is selected, which is the most popular configuration mode. Therefore, for the most common configuration mode, the mode pins can be left unconnected. (Note, however, that the internal pull-up resistor value can be as high as 100 kΩ.) After configuration, these pins can individually have weak pull-up or pull-down resistors, as specified in the design. A pull-down resistor value of 3.3kΩ is recommended.

These pins are located in the lower left chip corner and are near the readback nets. This location allows convenient routing if compatibility with the XC2000 and XC3000 family conventions of M0/RT, M1/RD is desired.

Configuration Modes

XC5200 devices have seven configuration modes. These modes are selected by a 3-bit input code applied to the M2, M1, and M0 inputs. There are three self-loading Master modes, two Peripheral modes, and a Serial Slave mode,

Table 10: Configuration Modes

Mode	M2	M1	M0	CCLK	Data
Master Serial	0	0	0	output	Bit-Serial
Slave Serial	1	1	1	input	Bit-Serial
Master Parallel Up	1	0	0	output	Byte-Wide, increment from 00000
Master Parallel Down	1	1	0	output	Byte-Wide, decrement from 3FFFF
Peripheral Synchronous*	0	1	1	input	Byte-Wide
Peripheral Asynchronous	1	0	1	output	Byte-Wide
Express	0	1	0	input	Byte-Wide
Reserved	0	0	1	—	—

Note :*Peripheral Synchronous can be considered byte-wide Slave Parallel

which is used primarily for daisy-chained devices. The seventh mode, called Express mode, is an additional slave mode that allows high-speed parallel configuration. The coding for mode selection is shown in Table 10.

Note that the smallest package, VQ64, only supports the Master Serial, Slave Serial, and Express modes.A detailed description of each configuration mode, with timing information, is included later in this data sheet. During configuration, some of the I/O pins are used temporarily for the configuration process. All pins used during configuration are shown in Table 13 on page 124.

Master Modes

The three Master modes use an internal oscillator to generate a Configuration Clock (CCLK) for driving potential slave devices. They also generate address and timing for external PROM(s) containing the configuration data.

Master Parallel (Up or Down) modes generate the CCLK signal and PROM addresses and receive byte parallel data. The data is internally serialized into the FPGA data-frame format. The up and down selection generates starting addresses at either zero or 3FFFF, for compatibility with different microprocessor addressing conventions. The

Master Serial mode generates CCLK and receives the configuration data in serial form from a Xilinx serial-configuration PROM.

CCLK speed is selectable as 1 MHz (default), 6 MHz, or 12 MHz. Configuration always starts at the default slow frequency, then can switch to the higher frequency during the first frame. Frequency tolerance is -50% to +50%.

Peripheral Modes

The two Peripheral modes accept byte-wide data from a bus. A RDY/$\overline{\text{BUSY}}$ status is available as a handshake signal. In Asynchronous Peripheral mode, the internal oscillator generates a CCLK burst signal that serializes the byte-wide data. CCLK can also drive slave devices. In the synchronous mode, an externally supplied clock input to CCLK serializes the data.

Slave Serial Mode

In Slave Serial mode, the FPGA receives serial configuration data on the rising edge of CCLK and, after loading its configuration, passes additional data out, resynchronized on the next falling edge of CCLK.

Multiple slave devices with identical configurations can be wired with parallel DIN inputs. In this way, multiple devices can be configured simultaneously.

Serial Daisy Chain

Multiple devices with different configurations can be connected together in a “daisy chain,” and a single combined bitstream used to configure the chain of slave devices.

To configure a daisy chain of devices, wire the CCLK pins of all devices in parallel, as shown in Figure 28 on page 114. Connect the DOUT of each device to the DIN of the next. The lead or master FPGA and following slaves each passes resynchronized configuration data coming from a single source. The header data, including the length count, is passed through and is captured by each FPGA when it recognizes the 0010 preamble. Following the length-count data, each FPGA outputs a High on DOUT until it has received its required number of data frames.

After an FPGA has received its configuration data, it passes on any additional frame start bits and configuration data on DOUT. When the total number of configuration clocks applied after memory initialization equals the value of the 24-bit length count, the FPGAs begin the start-up sequence and become operational together. FPGA I/O are normally released two CCLK cycles after the last configuration bit is received. Figure 25 on page 109 shows the start-up timing for an XC5200-Series device.

The daisy-chained bitstream is not simply a concatenation of the individual bitstreams. The PROM file formatter must be used to combine the bitstreams for a daisy-chained configuration.

Multi-Family Daisy Chain

All Xilinx FPGAs of the XC2000, XC3000, XC4000, and XC5200 Series use a compatible bitstream format and can, therefore, be connected in a daisy chain in an arbitrary sequence. There is, however, one limitation. If the chain contains XC5200-Series devices, the master normally cannot be an XC2000 or XC3000 device.

The reason for this rule is shown in Figure 25 on page 109. Since all devices in the chain store the same length count value and generate or receive one common sequence of CCLK pulses, they all recognize length-count match on the same CCLK edge, as indicated on the left edge of Figure 25. The master device then generates additional CCLK pulses until it reaches its finish point F. The different families generate or require different numbers of additional CCLK pulses until they reach F. Not reaching F means that the device does not really finish its configuration, although DONE may have gone High, the outputs became active, and the internal reset was released. For the XC5200-Series device, not reaching F means that readback cannot be initiated and most boundary scan instructions cannot be used.

The user has some control over the relative timing of these events and can, therefore, make sure that they occur at the proper time and the finish point F is reached. Timing is controlled using options in the bitstream generation software.

XC5200 devices always have the same number of CCLKs in the power up delay, independent of the configuration mode, unlike the XC3000/XC4000 Series devices. To guarantee all devices in a daisy chain have finished the power-up delay, tie the INIT pins together, as shown in Figure 27.

XC3000 Master with an XC5200-Series Slave

Some designers want to use an XC3000 lead device in peripheral mode and have the I/O pins of the XC5200-Series devices all available for user I/O. Figure 22 provides a solution for that case.

This solution requires one CLB, one IOB and pin, and an internal oscillator with a frequency of up to 5 MHz as a clock source. The XC3000 master device must be configured with late Internal Reset, which is the default option.

One CLB and one IOB in the lead XC3000-family device are used to generate the additional CCLK pulse required by the XC5200-Series devices. When the lead device removes the internal RESET signal, the 2-bit shift register responds to its clock input and generates an active Low output signal for the duration of the subsequent clock period. An external connection between this output and CCLK thus creates the extra CCLK pulse.

Configuration Timing

The seven configuration modes are discussed in detail in this section. Timing specifications are included.

Slave Serial Mode

In Slave Serial mode, an external signal drives the CCLK input of the FPGA. The serial configuration bitstream must be available at the DIN input of the lead FPGA a short setup time before each rising CCLK edge.

The lead FPGA then presents the preamble data—and all data that overflows the lead device—on its DOUT pin.

There is an internal delay of 0.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge.

Figure 28 shows a full master/slave system. An XC5200-Series device in Slave Serial mode should be connected as shown in the third device from the left.

Slave Serial mode is selected by a <111> on the mode pins (M2, M1, M0). Slave Serial is the default mode if the mode pins are left unconnected, as they have weak pull-up resistors during configuration.

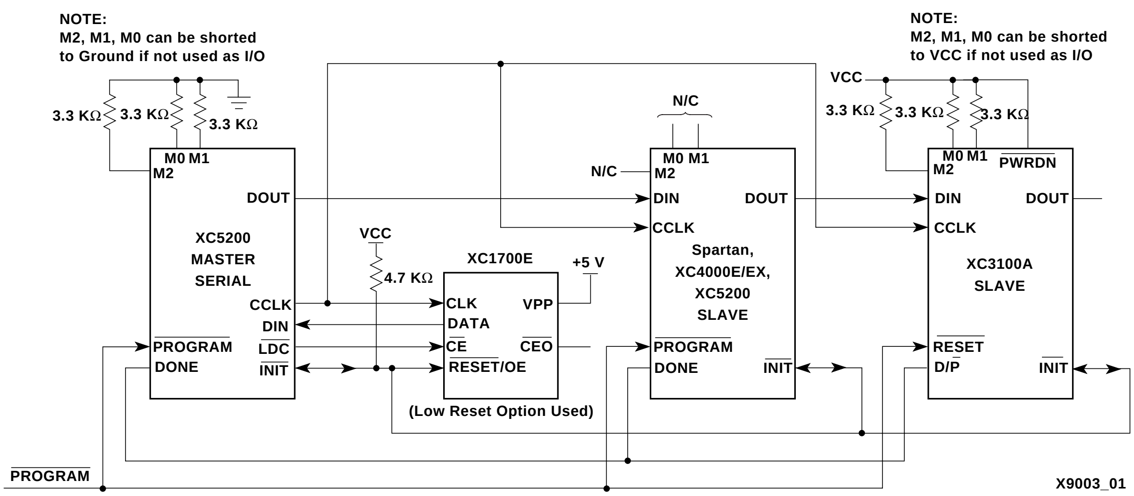
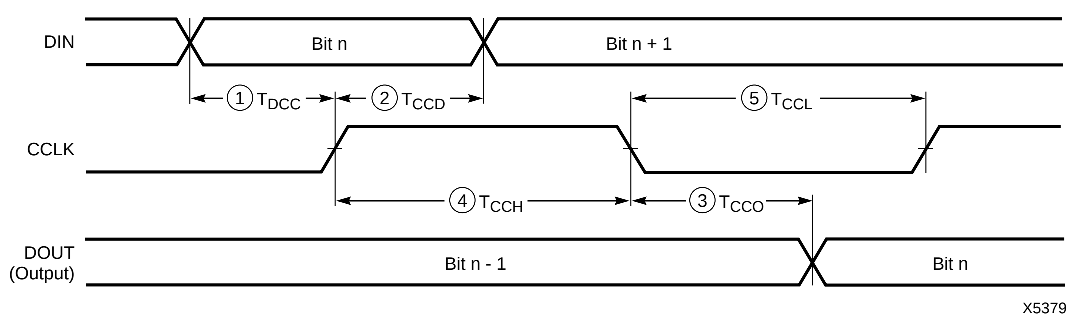


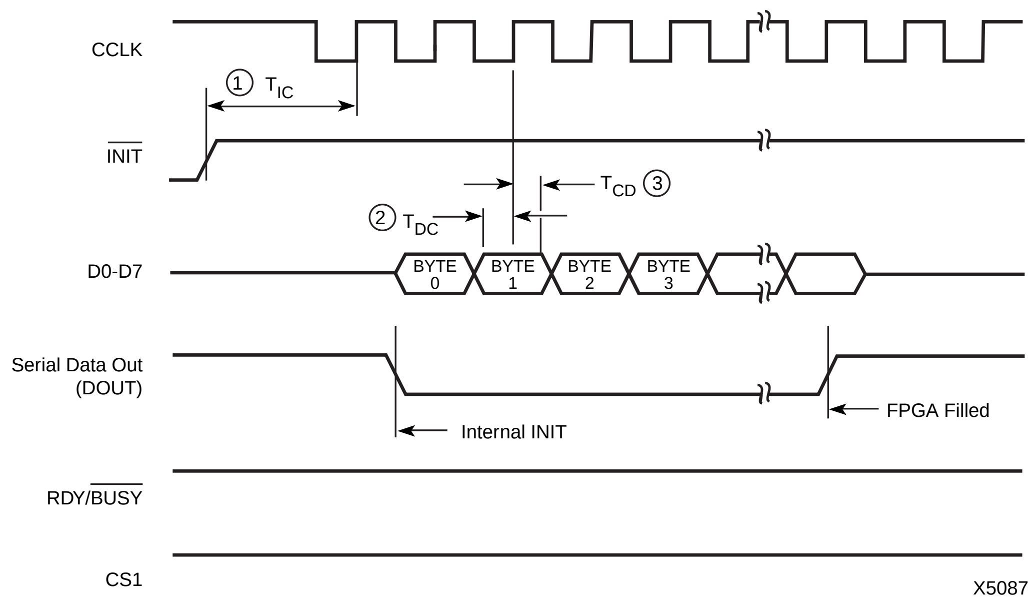
Figure 28: Master/Slave Serial Mode Circuit Diagram



	Description	Symbol		Min	Max	Units
CCLK	DIN setup	1	T_{DCC}	20		ns
	DIN hold	2	T_{CCD}	0		ns
	DIN to DOUT	3	T_{CCO}		30	ns
	High time	4	T_{CCH}	45		ns
	Low time	5	T_{CCL}	45		ns
	Frequency		F_{CC}		10	MHz

Note: Configuration must be delayed until the $\overline{\text{INIT}}$ pins of all daisy-chained FPGAs are High.

Figure 29: Slave Serial Mode Programming Switching Characteristics



	Description	Symbol		Min	Max	Units
CCLK	INIT (High) Setup time required	1	T_{IC}	5		µs
	DIN Setup time required	2	T_{DC}	30		ns
	DIN hold time required	3	T_{CD}	0		ns
	CCLK High time		T_{CCH}	30		ns
	CCLK Low time		T_{CCL}	30		ns
	CCLK frequency		F_{CC}		10	MHz

Note: If not driven by the preceding DOUT, CS1 must remain high until the device is fully configured.

Figure 38: Express Mode Programming Switching Characteristics

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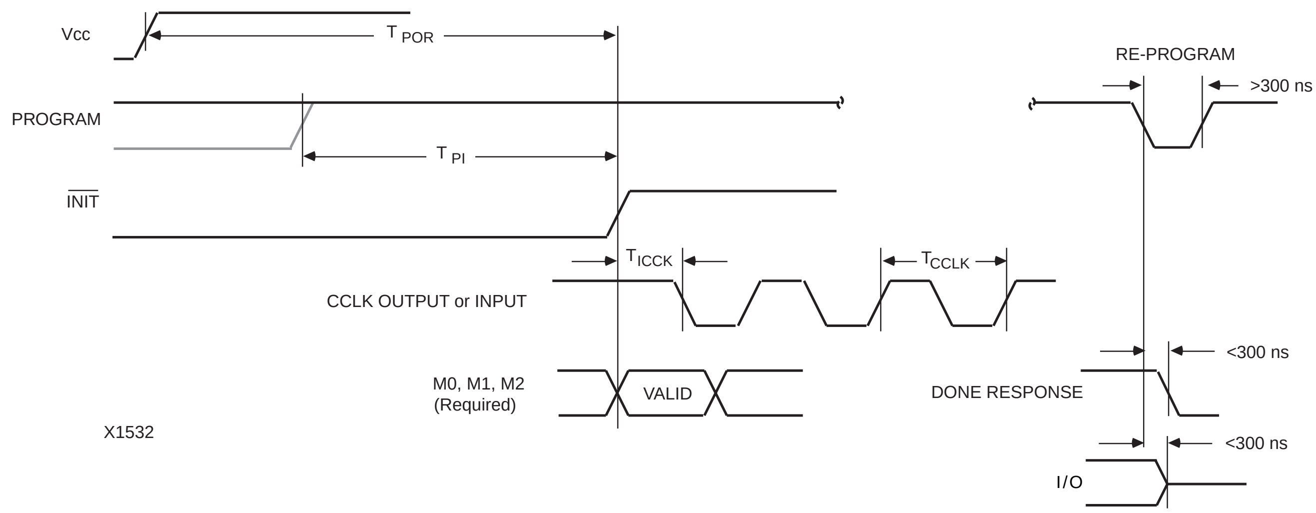
Configuration Switching Characteristics



Master Modes

Description	Symbol	Min	Max	Units
Power-On-Reset	T_{POR}	2	15	ms
Program Latency	T_{PI}	6	70	µs per CLB column
CCLK (output) Delay	T_{ICCK}	40	375	µs
period (slow)	T_{CCLK}	640	3000	ns
period (fast)	T_{CCLK}	100	375	ns

Slave and Peripheral Modes

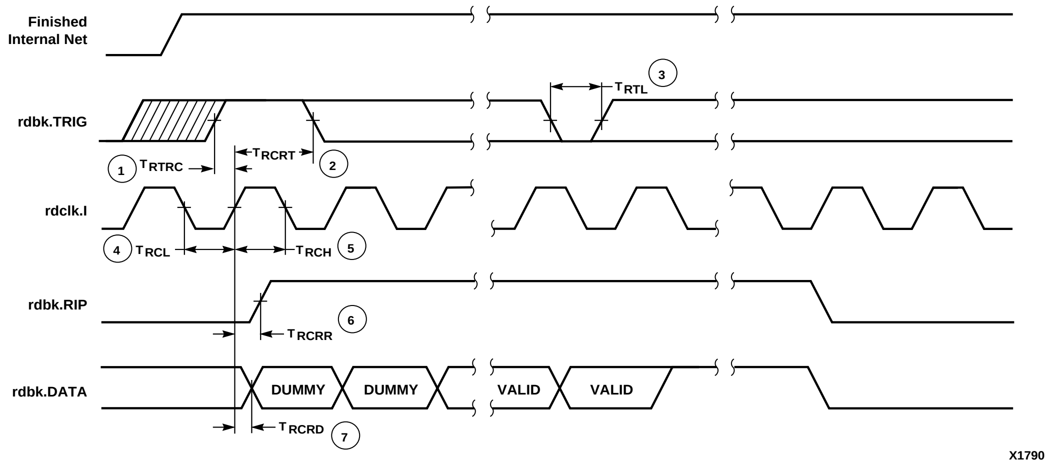
Description	Symbol	Min	Max	Units
Power-On-Reset	T_{POR}	2	15	ms
Program Latency	T_{PI}	6	70	µs per CLB column
CCLK (input) Delay (required)	T_{ICCK}	5		µs
period (required)	T_{CCLK}	100		ns

Note: At power-up, V_{CC} must rise from 2.0 to V_{CC} min in less than 15 ms, otherwise delay configuration using $\overline{\text{PROGRAM}}$ until V_{CC} is valid.

XC5200 Program Readback Switching Characteristic Guidelines

Testing of the switching parameters is modeled after testing methods specified by MIL-M-38510/605. All devices are 100% functionally tested. Internal timing parameters are not measured directly. They are derived from benchmark timing patterns that are taken at device introduction, prior to any process improvements.

The following guidelines reflect worst-case values over the recommended operating conditions.



	Description	Symbol		Min	Max	Units
rdbk.TRIG	rdbk.TRIG setup to initiate and abort Readback	1	T_{RTRC}	200	-	ns
	rdbk.TRIG hold to initiate and abort Readback	2	T_{RCRT}	50	-	ns
rdclk.1	rdbk.DATA delay	7	T_{RCRD}	-	250	ns
	rdbk.RIP delay	6	T_{RCRR}	-	250	ns
	High time	5	T_{RCH}	250	500	ns
	Low time	4	T_{RCL}	250	500	ns

Note 1: Timing parameters apply to all speed grades.

Note 2: rdbk.TRIG is High prior to Finished, Finished will trigger the first Readback

XC5200 CLB Switching Characteristic Guidelines

Testing of the switching parameters is modeled after testing methods specified by MIL-M-38510/605. All devices are 100% functionally tested. Since many internal timing parameters cannot be measured directly, they are derived from benchmark timing patterns. The following guidelines reflect worst-case values over the recommended operating conditions. For more detailed, more precise, and more up-to-date timing information, use the values provided by the timing calculator and used in the simulator.

Speed Grade		-6		-5		-4		-3	
Description	**Symbol**	**Min (ns)**	**Max (ns)**	**Min (ns)**	**Max (ns)**	**Min (ns)**	**Max (ns)**	**Min (ns)**	**Max (ns)**
Combinatorial Delays									
F inputs to X output	T_{ILO}		5.6		4.6		3.8		3.0
F inputs via transparent latch to Q	T_{ITO}		8.0		6.6		5.4		4.3
DI inputs to DO output (Logic-Cell Feedthrough)	T_{IDO}		4.3		3.5		2.8		2.4
F inputs via F5_MUX to DO output	T_{IMO}		7.2		5.8		5.0		4.3
Carry Delays									
Incremental delay per bit	T_{CY}		0.7		0.6		0.5		0.5
Carry-in overhead from DI	T_{CYDI}		1.8		1.6		1.5		1.4
Carry-in overhead from F	T_{CYL}		3.7		3.2		2.9		2.4
Carry-out overhead to DO	T_{CYO}		4.0		3.2		2.5		2.1
Sequential Delays									
Clock (CK) to out (Q) (Flip-Flop)	T_{CKO}		5.8		4.9		4.0		4.0
Gate (Latch enable) going active to out (Q)	T_{GO}		9.2		7.4		5.9		5.5
Set-up Time Before Clock (CK)									
F inputs	T_{ICK}	2.3		1.8		1.4		1.3	
F inputs via F5_MUX	T_{MICK}	3.8		3.0		2.5		2.4	
DI input	T_{DICK}	0.8		0.5		0.4		0.4	
CE input	T_{EICK}	1.6		1.2		0.9		0.9	
Hold Times After Clock (CK)									
F inputs	T_{CKI}	0		0		0		0	
F inputs via F5_MUX	T_{CKMI}	0		0		0		0	
DI input	T_{CKDI}	0		0		0		0	
CE input	T_{CKEI}	0		0		0		0	
Clock Widths									
Clock High Time	T_{CH}	6.0		6.0		6.0		6.0	
Clock Low Time	T_{CL}	6.0		6.0		6.0		6.0	
Toggle Frequency (MHz) (Note 3)	F_{TOG}		83		83		83		83
Reset Delays									
Width (High)	T_{CLRW}	6.0		6.0		6.0		6.0	
Delay from CLR to Q (Flip-Flop)	T_{CLR}		7.7		6.3		5.1		4.0
Delay from CLR to Q (Latch)	T_{CLRL}		6.5		5.2		4.2		3.0
Global Reset Delays									
Width (High)	T_{GCLRW}	6.0		6.0		6.0		6.0	
Delay from internal GR to Q	T_{GCLR}		14.7		12.1		9.1		8.0

Note:
1. The CLB K to Q output delay (T_{CKO}) of any CLB, plus the shortest possible interconnect delay, is always longer than the Data In hold-time requirement (T_{CKDI}) of any CLB on the same die.
2. Timing is based upon the XC5215 device. For other devices, see Timing Calculator.
3. Maximum flip-flop toggle rate for export control purposes.

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Product Obsolete or Under Obsolescence

XC5200 Boundary Scan (JTAG) Switching Characteristic Guidelines

The following guidelines reflect worst-case values over the recommended operating conditions. They are expressed in units of nanoseconds and apply to all XC5200 devices unless otherwise noted.

Speed Grade		-6		-5		-4		-3	
Description	**Symbol**	**Min**	**Max**	**Min**	**Max**	**Min**	**Max**	**Min**	**Max**
Setup and Hold									
Input (TDI) to clock (TCK) setup time	T_{TDITCK}	30.0		30.0		30.0		30.0	
Input (TDI) to clock (TCK) hold time	T_{TCKTDI}	0		0		0		0	
Input (TMS) to clock (TCK) setup time	T_{TMSTCK}	15.0		15.0		15.0		15.0	
Input (TMS) to clock (TCK) hold time	T_{TCKTMS}	0		0		0		0	
Propagation Delay									
Clock (TCK) to Pad (TDO)	T_{TCKPO}		30.0		30.0		30.0		30.0
Clock									
Clock (TCK) High	T_{TCKH}	30.0		30.0		30.0		30.0	
Clock (TCK) Low	T_{TCKL}	30.0		30.0		30.0		30.0	
F_{MAX} (MHz)	F_{MAX}		10.0		10.0		10.0		10.0

Note 1: Input pad setup and hold times are specified with respect to the internal clock.

Device-Specific Pinout Tables

Device-specific tables include all packages for each XC5200-Series device. They follow the pad locations around the die, and include boundary scan register locations.

Pin Locations for XC5202 Devices

The following table may contain pinout information for unsupported device/package combinations. Please see the availability charts elsewhere in the XC5200 Series data sheet for availability information.

Pin	Description	VQ64*	PC84	PQ100	VQ100	TQ144	PG156	Boundary Scan Order
	VCC	-	2	92	89	128	H3	-
1.	I/O (A8)	57	3	93	90	129	H1	51
2.	I/O (A9)	58	4	94	91	130	G1	54
3.	I/O	-	-	95	92	131	G2	57
4.	I/O	-	-	96	93	132	G3	63
5.	I/O (A10)	-	5	97	94	133	F1	66
6.	I/O (A11)	59	6	98	95	134	F2	69
	GND	-	-	-	-	137	F3	-
7.	I/O (A12)	60	7	99	96	138	E3	78
8.	I/O (A13)	61	8	100	97	139	C1	81
9.	I/O (A14)	62	9	1	98	142	B1	90
10.	I/O (A15)	63	10	2	99	143	B2	93
	VCC	64	11	3	100	144	C3	-
	GND	-	12	4	1	1	C4	-
11.	GCK1 (A16, I/O)	1	13	5	2	2	B3	102
12.	I/O (A17)	2	14	6	3	3	A1	105
13.	I/O (TDI)	3	15	7	4	6	B4	111
14.	I/O (TCK)	4	16	8	5	7	A3	114
	GND	-	-	-	-	8	C6	-
15.	I/O (TMS)	5	17	9	6	11	A5	117
16.	I/O	6	18	10	7	12	C7	123
17.	I/O	-	-	-	-	13	B7	126
18.	I/O	-	-	11	8	14	A6	129
19.	I/O	-	19	12	9	15	A7	135
20.	I/O	7	20	13	10	16	A8	138
	GND	8	21	14	11	17	C8	-
	VCC	9	22	15	12	18	B8	-
21.	I/O	-	23	16	13	19	C9	141
22.	I/O	10	24	17	14	20	B9	147
23.	I/O		-	18	15	21	A9	150
24.	I/O		-	-	-	22	B10	153
25.	I/O	-	25	19	16	23	C10	159
26.	I/O	11	26	20	17	24	A10	162
	GND		-	-	-	27	C11	-
27.	I/O	12	27	21	18	28	B12	165
28.	I/O		-	22	19	29	A13	171
29.	I/O	13	28	23	20	32	B13	174
30.	I/O	14	29	24	21	33	B14	177
31.	M1 (I/O)	15	30	25	22	34	A15	186
	GND	-	31	26	23	35	C13	-
32.	M0 (I/O)	16	32	27	24	36	A16	189
	VCC	-	33	28	25	37	C14	-
33.	M2 (I/O)	17	34	29	26	38	B15	192
34.	GCK2 (I/O)	18	35	30	27	39	B16	195

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Pin	Description		PC84	PQ100	VQ100	TQ144	PG156	PQ160	Boundary Scan Order
57.	I/O		-	-	-	47	E16	53	306
58.	I/O		38	34	31	48	F16	54	312
59.	I/O		39	35	32	49	G14	55	315
60.	I/O		-	36	33	50	G15	56	318
61.	I/O		-	37	34	51	G16	57	324
62.	I/O		40	38	35	52	H16	58	327
63.	I/O ($\overline{\text{ERR}}$, $\overline{\text{INIT}}$)		41	39	36	53	H15	59	330
	VCC		42	40	37	54	H14	60	-
	GND		43	41	38	55	J14	61	-
64.	I/O		44	42	39	56	J15	62	336
65.	I/O		45	43	40	57	J16	63	339
66.	I/O		-	44	41	58	K16	64	348
67.	I/O		-	45	42	59	K15	65	351
68.	I/O		46	46	43	60	K14	66	354
69.	I/O		47	47	44	61	L16	67	360
70.	I/O		-	-	-	62	M16	68	363
71.	I/O		-	-	-	63	L15	69	366
	GND		-	-	-	64	L14	70	-
72.	I/O		-	-	-	-	N16	71	372
73.	I/O		-	-	-	-	M15	72	375
74.	I/O		48	48	45	65	P16	73	378
75.	I/O		49	49	46	66	M14	74	384
76.	I/O		-	-	-	67	N15	75	387
77.	I/O		-	-	-	68	P15	76	390
78.	I/O		50	50	47	69	N14	77	396
79.	I/O		51	51	48	70	R16	78	399
	GND		52	52	49	71	P14	79	-
	DONE		53	53	50	72	R15	80	-
	VCC		54	54	51	73	P13	81	-
	PROG		55	55	52	74	R14	82	-
80.	I/O (D7)		56	56	53	75	T16	83	408
81.	GCK3 (I/O)		57	57	54	76	T15	84	411
82.	I/O		-	-	-	77	R13	85	420
83.	I/O		-	-	-	78	P12	86	423
84.	I/O (D6)		58	58	55	79	T14	87	426
85.	I/O		-	59	56	80	T13	88	432
	GND		-	-	-	81	P11	91	-
86.	I/O		-	-	-	82	R11	92	435
87.	I/O		-	-	-	83	T11	93	438
88.	I/O (D5)		59	60	57	84	T10	94	444
89.	I/O ($\overline{\text{CS0}}$)		60	61	58	85	P10	95	447
90.	I/O		-	62	59	86	R10	96	450
91.	I/O		-	63	60	87	T9	97	456
92.	I/O (D4)		61	64	61	88	R9	98	459
93.	I/O		62	65	62	89	P9	99	462
	VCC		63	66	63	90	R8	100	-
	GND		64	67	64	91	P8	101	-
94.	I/O (D3)		65	68	65	92	T8	102	468
95.	I/O ($\overline{\text{RS}}$)		66	69	66	93	T7	103	471
96.	I/O		-	70	67	94	T6	104	474
97.	I/O		-	-	-	95	R7	105	480
98.	I/O (D2)		67	71	68	96	P7	106	483

Pin	Description		PC84	PQ100	VQ100	TQ144	PG156	PQ160	Boundary Scan Order
99.	I/O		68	72	69	97	T5	107	486
100.	I/O		-	-	-	98	R6	108	492
101.	I/O		-	-	-	99	T4	109	495
	GND		-	-	-	100	P6	110	-
102.	I/O (D1)		69	73	70	101	T3	113	498
103.	I/O (RCLK-BUSY/RDY)		70	74	71	102	P5	114	504
104.	I/O		-	-	-	103	R4	115	507
105.	I/O		-	-	-	104	R3	116	510
106.	I/O (D0, DIN)		71	75	72	105	P4	117	516
107.	I/O (DOUT)		72	76	73	106	T2	118	519
	CCLK		73	77	74	107	R2	119	-
	VCC		74	78	75	108	P3	120	-
108.	I/O (TDO)		75	79	76	109	T1	121	0
	GND		76	80	77	110	N3	122	-
109.	I/O (A0, $\overline{WS}$)		77	81	78	111	R1	123	9
110.	GCK4 (A1, I/O)		78	82	79	112	P2	124	15
111.	I/O		-	-	-	113	N2	125	18
112.	I/O		-	-	-	114	M3	126	21
113.	I/O (A2, CS1)		79	83	80	115	P1	127	27
114.	I/O (A3)		80	84	81	116	N1	128	30
115.	I/O		-	-	-	117	M2	129	33
116.	I/O		-	-	-	-	M1	130	39
	GND		-	-	-	118	L3	131	-
117.	I/O		-	-	-	119	L2	132	42
118.	I/O		-	-	-	120	L1	133	45
119.	I/O (A4)		81	85	82	121	K3	134	51
120.	I/O (A5)		82	86	83	122	K2	135	54
121.	I/O		-	87	84	123	K1	137	57
122.	I/O		-	88	85	124	J1	138	63
123.	I/O (A6)		83	89	86	125	J2	139	66
124.	I/O (A7)		84	90	87	126	J3	140	69
	GND		1	91	88	127	H2	141	-

Additional No Connect (N.C.) Connections for PQ160 Package

PQ160				
8	30	89	111	136
9	31	90	112	

Notes: Boundary Scan Bit 0 = TDO.T
Boundary Scan Bit 1 = TDO.O
Boundary Scan Bit 1056 = BSCAN.UPD

Pin	Description	PC84	PQ100	VQ100	TQ144	PQ160	TQ176	PG191	PQ208	Boundary Scan Order
87.	I/O	-	-	-	-	72	80	P17	94	468
88.	I/O	48	48	45	65	73	81	N16	95	471
89.	I/O	49	49	46	66	74	82	T17	96	480
90.	I/O	-	-	-	67	75	83	R17	97	483
91.	I/O	-	-	-	68	76	84	P16	98	486
92.	I/O	50	50	47	69	77	85	U18	99	492
93.	I/O	51	51	48	70	78	86	T16	100	495
	GND	52	52	49	71	79	87	R16	101	-
	DONE	53	53	50	72	80	88	U17	103	-
	VCC	54	54	51	73	81	89	R15	106	-
	PROG	55	55	52	74	82	90	V18	108	-
94.	I/O (D7)	56	56	53	75	83	91	T15	109	504
95.	GCK3 (I/O)	57	57	54	76	84	92	U16	110	507
96.	I/O	-	-	-	77	85	93	T14	111	516
97.	I/O	-	-	-	78	86	94	U15	112	519
98.	I/O (D6)	58	58	55	79	87	95	V17	113	522
99.	I/O	-	59	56	80	88	96	V16	114	528
100.	I/O	-	-	-	-	89	97	T13	115	531
101.	I/O	-	-	-	-	90	98	U14	116	534
	GND	-	-	-	81	91	99	T12	119	-
102.	I/O	-	-	-	82	92	100	U13	120	540
103.	I/O	-	-	-	83	93	101	V13	121	543
104.	I/O (D5)	59	60	57	84	94	102	U12	122	552
105.	I/O ($\overline{\text{CS0}}$)	60	61	58	85	95	103	V12	123	555
106.	I/O	-	-	-	-	-	104	T11	124	558
107.	I/O	-	-	-	-	-	105	U11	125	564
108.	I/O	-	62	59	86	96	106	V11	126	567
109.	I/O	-	63	60	87	97	107	V10	127	570
110.	I/O (D4)	61	64	61	88	98	108	U10	128	576
111.	I/O	62	65	62	89	99	109	T10	129	579
	VCC	63	66	63	90	100	110	R10	130	-
	GND	64	67	64	91	101	111	R9	131	-
112.	I/O (D3)	65	68	65	92	102	112	T9	132	588
113.	I/O ($\overline{\text{RS}}$)	66	69	66	93	103	113	U9	133	591
114.	I/O	-	70	67	94	104	114	V9	134	600
115.	I/O	-	-	-	95	105	115	V8	135	603
116.	I/O	-	-	-	-	-	116	U8	136	612
117.	I/O	-	-	-	-	-	117	T8	137	615
118.	I/O (D2)	67	71	68	96	106	118	V7	138	618
119.	I/O	68	72	69	97	107	119	U7	139	624
120.	I/O	-	-	-	98	108	120	V6	140	627
121.	I/O	-	-	-	99	109	121	U6	141	630
	GND	-	-	-	100	110	122	T7	142	-
122.	I/O	-	-	-	-	111	123	U5	145	636
123.	I/O	-	-	-	-	112	124	T6	146	639
124.	I/O (D1)	69	73	70	101	113	125	V3	147	642
125.	I/O (RCLK-BUSY/RDY)	70	74	71	102	114	126	V2	148	648
126.	I/O	-	-	-	103	115	127	U4	149	651
127.	I/O	-	-	-	104	116	128	T5	150	654
128.	I/O (D0, DIN)	71	75	72	105	117	129	U3	151	660
129.	I/O (DOUT)	72	76	73	106	118	130	T4	152	663

Pin	Description	PC84	PQ100	VQ100	TQ144	PQ160	TQ176	PG191	PQ208	Boundary Scan Order
	CCLK	73	77	74	107	119	131	V1	153	-
	VCC	74	78	75	108	120	132	R4	154	-
130.	I/O (TDO)	75	79	76	109	121	133	U2	159	-
	GND	76	80	77	110	122	134	R3	160	-
131.	I/O (A0, $\overline{WS}$)	77	81	78	111	123	135	T3	161	9
132.	GCK4 (A1, I/O)	78	82	79	112	124	136	U1	162	15
133.	I/O	-	-	-	113	125	137	P3	163	18
134.	I/O	-	-	-	114	126	138	R2	164	21
135.	I/O (A2, CS1)	79	83	80	115	127	139	T2	165	27
136.	I/O (A3)	80	84	81	116	128	140	N3	166	30
137.	I/O	-	-	-	117	129	141	P2	167	33
138.	I/O	-	-	-	-	130	142	T1	168	42
	GND	-	-	-	118	131	143	M3	171	-
139.	I/O	-	-	-	119	132	144	P1	172	45
140.	I/O	-	-	-	120	133	145	N1	173	51
141.	I/O (A4)	81	85	82	121	134	146	M2	174	54
142.	I/O (A5)	82	86	83	122	135	147	M1	175	57
143.	I/O	-	-	-	-	-	148	L3	176	63
144.	I/O	-	-	-	-	136	149	L2	177	66
145.	I/O	-	87	84	123	137	150	L1	178	69
146.	I/O	-	88	85	124	138	151	K1	179	75
147.	I/O (A6)	83	89	86	125	139	152	K2	180	78
148.	I/O (A7)	84	90	87	126	140	153	K3	181	81
	GND	1	91	88	127	141	154	K4	182	-

Additional No Connect (N.C.) Connections for PQ208 and TQ176 Packages

PQ208							TQ176
195	1	39	65	104	143	158	167
196	3	51	66	105	144	169	
206	12	52	91	107	155	170	
207	13	53	92	117	156		
208	38	54	102	118	157		

Notes: Boundary Scan Bit 0 = TDO.T
Boundary Scan Bit 1 = TDO.O
Boundary Scan Bit 1056 = BSCAN.UPD

Pin Locations for XC5210 Devices

The following table may contain pinout information for unsupported device/package combinations. Please see the availability charts elsewhere in the XC5200 Series data sheet for availability information.

Pin	Description	PC84	TQ144	PQ160	TQ176	PQ208	PG223	BG225	PQ240	Boundary Scan Order
	VCC	2	128	142	155	183	J4	VCC*	212	-
1.	I/O (A8)	3	129	143	156	184	J3	E8	213	111
2.	I/O (A9)	4	130	144	157	185	J2	B7	214	114
3.	I/O	-	131	145	158	186	J1	A7	215	117
4.	I/O	-	132	146	159	187	H1	C7	216	123
5.	I/O	-	-	-	160	188	H2	D7	217	126
6.	I/O	-	-	-	161	189	H3	E7	218	129

Pin	Description	PQ160	HQ208	HQ240	PG299	BG225	BG352	Boundary Scan Order
146.	I/O	-	-	-	R17	-	AD6	750
147.	I/O	-	-	-	T18	-	AC7	756
148.	I/O	73	95	113	U19	R13	AF4	759
149.	I/O	74	96	114	V19	N12	AF3	768
150.	I/O	75	97	115	R16	P13	AD5	771
151.	I/O	76	98	116	T17	K10	AE3	774
152.	I/O	77	99	117	U18	R14	AD4	780
153.	I/O	78	100	118	X20	N13	AC5	783
	GND	79	101	119	W20	GND*	GND*	-
	DONE	80	103	120	V18	P14	AD3	-
	VCC	81	106	121	X19	VCC*	VCC*	-
	PROG	82	108	122	U17	M12	AC4	-
154.	I/O (D7)	83	109	123	W19	P15	AD2	792
155.	GCK3 (I/O)	84	110	124	W18	N14	AC3	795
156.	I/O	85	111	125	T15	L11	AB4	804
157.	I/O	86	112	126	U16	M13	AD1	807
158.	I/O	-	-	127	V17	N15	AA4	810
159.	I/O	-	-	128	X18	M14	AA3	816
160.	I/O	-	-	-	U15	-	AB2	819
161.	I/O	-	-	-	T14	-	AC1	828
162.	I/O (D6)	87	113	129	W17	J10	Y3	831
163.	I/O	88	114	130	V16	L12	AA2	834
164.	I/O	89	115	131	X17	M15	AA1	840
165.	I/O	90	116	132	U14	L13	W4	843
166.	I/O	-	117	133	V15	L14	W3	846
167.	I/O	-	118	134	T13	K11	Y2	852
168.	I/O	-	-	-	W16	-	Y1	855
169.	I/O	-	-	-	W15	-	V4	858
	GND	91	119	135	X16	GND*	GND*	-
170.	I/O	-	-	136	U13	L15	V3	864
171.	I/O	-	-	137	V14	K12	W2	867
172.	I/O	92	120	138	W14	K13	U4	870
173.	I/O	93	121	139	V13	K14	U3	876
	VCC	-	-	140	X15	VCC*	VCC*	-
174.	I/O (D5)	94	122	141	T12	K15	V2	879
175.	I/O ($\overline{CS0}$)	95	123	142	X14	J12	V1	882
176.	I/O	-	-	-	X13	-	T1	888
177.	I/O	-	-	-	V12	-	R4	891
178.	I/O	-	124	144	W12	J13	R3	894
179.	I/O	-	125	145	T11	J14	R2	900
180.	I/O	96	126	146	X12	J15	R1	903
181.	I/O	97	127	147	U11	J11	P3	906
182.	I/O (D4)	98	128	148	V11	H13	P2	912
183.	I/O	99	129	149	W11	H14	P1	915
	VCC	100	130	150	X10	VCC*	VCC*	-
	GND	101	131	151	X11	GND*	GND*	-
184.	I/O (D3)	102	132	152	W10	H12	N2	924
185.	I/O ($\overline{RS}$)	103	133	153	V10	H11	N4	927
186.	I/O	104	134	154	T10	G14	N3	936
187.	I/O	105	135	155	U10	G15	M1	939
188.	I/O	-	136	156	X9	G13	M2	942
189.	I/O	-	137	157	W9	G12	M3	948

Revisions

Version	Description
12/97	Rev 5.0 added -3, -4 specification
7/98	Rev 5.1 added Spartan family to comparison, removed HQ304
11/98	Rev 5.2 All specifications made final.