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Details

Product Status	Not For New Designs
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, I ² C, LINbus, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	79
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	32K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 28x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/spc560c40l3c6e0x

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Figure 1. SPC560B40x/50x and SPC560C40x/50x block diagram

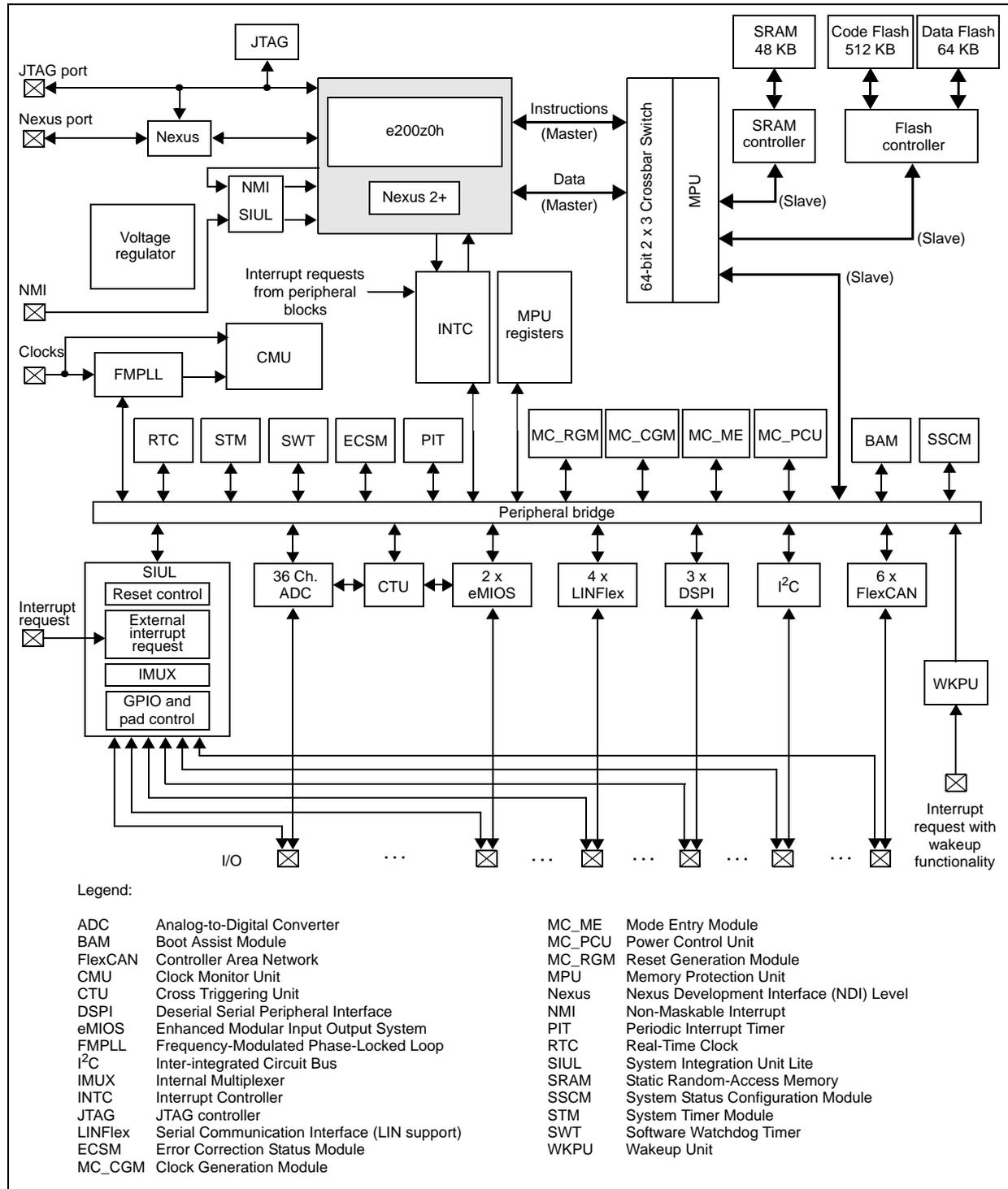


Table 3 summarizes the functions of all blocks present in the SPC560B40x/50x and SPC560C40x/50x series of microcontrollers. Please note that the presence and number of blocks vary by device and package.

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function ⁽¹⁾	Function	Peripheral	I/O direction ⁽²⁾	Pad type	RESET configuration	Pin number			
								LQFP64	LQFP100	LQFP144	LBGA208 ⁽³⁾
PE[11]	PCR[75]	AF0 AF1 AF2 AF3 — —	GPIO[75] — CS4_1 — LIN3RX WKPU[14] ⁽⁴⁾	SIUL — DSPI_1 — LINFlex_3 WKPU	I/O — O — I I	S	Tristate	—	13	17	H2
PE[12]	PCR[76]	AF0 AF1 AF2 AF3 — —	GPIO[76] — E1UC[19] ⁽¹³⁾ — SIN_2 EIRQ[11]	SIUL — eMIOS_1 — DSPI_2 SIUL	I/O — I/O — I I	S	Tristate	—	76	109	C14
PE[13]	PCR[77]	AF0 AF1 AF2 AF3	GPIO[77] SOUT2 E1UC[20] —	SIUL DSPI_2 eMIOS_1 —	I/O O I/O —	S	Tristate	—	—	103	D15
PE[14]	PCR[78]	AF0 AF1 AF2 AF3 —	GPIO[78] SCK_2 E1UC[21] — EIRQ[12]	SIUL DSPI_2 eMIOS_1 — SIUL	I/O I/O I/O — I	S	Tristate	—	—	112	C13
PE[15]	PCR[79]	AF0 AF1 AF2 AF3	GPIO[79] CS0_2 E1UC[22] —	SIUL DSPI_2 eMIOS_1 —	I/O I/O I/O —	M	Tristate	—	—	113	A13
PF[0]	PCR[80]	AF0 AF1 AF2 AF3 —	GPIO[80] E0UC[10] CS3_1 — ANS[8]	SIUL eMIOS_0 DSPI_1 — ADC	I/O I/O O — I	J	Tristate	—	—	55	N10
PF[1]	PCR[81]	AF0 AF1 AF2 AF3 —	GPIO[81] E0UC[11] CS4_1 — ANS[9]	SIUL eMIOS_0 DSPI_1 — I	I/O I/O O — I	J	Tristate	—	—	56	P10

11. Available only on SPC560Cx versions and SPC560B50B2 devices
12. Not available on SPC560B40L3 and SPC560B40L5 devices
13. Not available in 100 LQFP package
14. Available only on SPC560B50B2 devices
15. Not available on SPC560B44L3 devices

3.7 Nexus 2+ pins

In the LBG208 package, eight additional debug pins are available (see [Table 7](#)).

Table 7. Nexus 2+ pin descriptions

Debug pin	Function	I/O direction	Pad type	Function after reset	Pin number		
					LQFP 100	LQFP 144	LBGA 208 ⁽¹⁾
MCKO	Message clock out	O	F	—	—	—	T4
MDO0	Message data out 0	O	M	—	—	—	H15
MDO1	Message data out 1	O	M	—	—	—	H16
MDO2	Message data out 2	O	M	—	—	—	H14
MDO3	Message data out 3	O	M	—	—	—	H13
EVTI	Event in	I	M	Pull-up	—	—	K1
EVTO	Event out	O	M	—	—	—	L4
MSEO	Message start/end out	O	M	—	—	—	G16

1. LBG208 available only as development package for Nexus2+.

3.8 Electrical characteristics

3.9 Introduction

This section contains electrical characteristics of the device as well as temperature and power considerations.

This product contains devices to protect the inputs against damage due to high static voltages. However, it is advisable to take precautions to avoid applying any voltage higher than the specified maximum rated voltages.

To enhance reliability, unused inputs can be driven to an appropriate logic voltage level (V_{DD} or V_{SS}). This could be done by the internal pull-up and pull-down, which is provided by the product for most general purpose pins.

The parameters listed in the following tables represent the characteristics of the device and its demands on the system.

In the tables where the device logic provides signals with their respective timing characteristics, the symbol “CC” for Controller Characteristics is included in the Symbol column.

Table 13. Recommended operating conditions (3.3 V) (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V _{IN}	SR	Voltage on any GPIO pin with respect to ground (V _{SS})	—	V _{SS} -0.1	—	V
			Relative to V _{DD}	—	V _{DD} +0.1	
I _{INJPAD}	SR	Injected input current on any pin during overload condition	—	-5	5	mA
I _{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	-50	50	
TV _{DD}	SR	V _{DD} slope to ensure correct power up ⁽⁶⁾	—	3.0 ⁽⁷⁾	250 x 10 ³ (0.25 [V/μs])	V/s

- 100 nF capacitance needs to be provided between each V_{DD}/V_{SS} pair
- 330 nF capacitance needs to be provided between each V_{DD_LV}/V_{SS_LV} supply pair.
- 400 nF capacitance needs to be provided between V_{DD_BV} and the nearest V_{SS_LV} (higher value may be needed depending on external regulator characteristics).
- 100 nF capacitance needs to be provided between V_{DD_ADC}/V_{SS_ADC} pair.
- Full electrical specification cannot be guaranteed when voltage drops below 3.0 V. In particular, ADC electrical characteristics and I/Os DC electrical specification may not be guaranteed. When voltage drops below V_{LVDHVL}, device is reset.
- Guaranteed by device validation.
- Minimum value of TV_{DD} must be guaranteed until V_{DD} reaches 2.6 V (maximum value of V_{PORH}).

Table 14. Recommended operating conditions (5.0 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V _{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V _{DD} ⁽¹⁾	SR	Voltage on VDD_HV pins with respect to ground (V _{SS})	—	4.5	5.5	V
			Voltage drop ⁽²⁾	3.0	5.5	
V _{SS_LV} ⁽³⁾	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V _{SS})	—	V _{SS} -0.1	V _{SS} +0.1	V
V _{DD_BV} ⁽⁴⁾	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V _{SS})	—	4.5	5.5	V
			Voltage drop ⁽²⁾	3.0	5.5	
			Relative to V _{DD}	V _{DD} -0.1	V _{DD} +0.1	
V _{SS_ADC}	SR	Voltage on VSS_HV_ADC (ADC reference) pin with respect to ground (V _{SS})	—	V _{SS} -0.1	V _{SS} +0.1	V
V _{DD_ADC} ⁽⁵⁾	SR	Voltage on VDD_HV_ADC pin (ADC reference) with respect to ground (V _{SS})	—	4.5	5.5	V
			Voltage drop ⁽²⁾	3.0	5.5	
			Relative to V _{DD}	V _{DD} -0.1	V _{DD} +0.1	
V _{IN}	SR	Voltage on any GPIO pin with respect to ground (V _{SS})	—	V _{SS} -0.1	—	V
			Relative to V _{DD}	—	V _{DD} +0.1	

Table 21. Output pin transition times (continued)

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit	
				Min	Typ	Max		
t_{tr}	CC	Output transition time output pin ⁽²⁾ MEDIUM configuration	$C_L = 25$ pF	$V_{DD} = 5.0$ V \pm 10%, PAD3V5V = 0 SIUL.PCRx.SRC = 1	—	—	10	ns
			$C_L = 50$ pF		—	—	20	
			$C_L = 100$ pF		—	—	40	
			$C_L = 25$ pF	$V_{DD} = 3.3$ V \pm 10%, PAD3V5V = 1 SIUL.PCRx.SRC = 1	—	—	12	
			$C_L = 50$ pF		—	—	25	
			$C_L = 100$ pF		—	—	40	
t_{tr}	CC	Output transition time output pin ⁽²⁾ FAST configuration	$C_L = 25$ pF	$V_{DD} = 5.0$ V \pm 10%, PAD3V5V = 0	—	—	4	ns
			$C_L = 50$ pF		—	—	6	
			$C_L = 100$ pF		—	—	12	
			$C_L = 25$ pF	$V_{DD} = 3.3$ V \pm 10%, PAD3V5V = 1	—	—	4	
			$C_L = 50$ pF		—	—	7	
			$C_L = 100$ pF		—	—	12	

1. $V_{DD} = 3.3$ V \pm 10% / 5.0 V \pm 10%, $T_A = -40$ to 125 °C, unless otherwise specified

2. C_L includes device and package capacitances ($C_{PKG} < 5$ pF).

3.15.5 I/O pad current specification

The I/O pads are distributed across the I/O supply segment. Each I/O supply segment is associated to a V_{DD}/V_{SS} supply pair as described in [Table 22](#).

Table 22. I/O supply segment

Package	Supply segment					
	1	2	3	4	5	6
LBGA208 ⁽¹⁾	Equivalent to LQFP144 segment pad distribution				MCKO	MDOOn/MSEO
LQFP144	pin20–pin49	pin51–pin99	pin100–pin122	pin 123–pin19	—	—
LQFP100	pin16–pin35	pin37–pin69	pin70–pin83	pin 84–pin15	—	—
LQFP64 ⁽²⁾	pin8–pin26	pin28–pin55	pin56–pin7	—	—	—

1. LBGA208 available only as development package for Nexus2+

2. All LQFP64 information is indicative and must be confirmed during silicon validation.

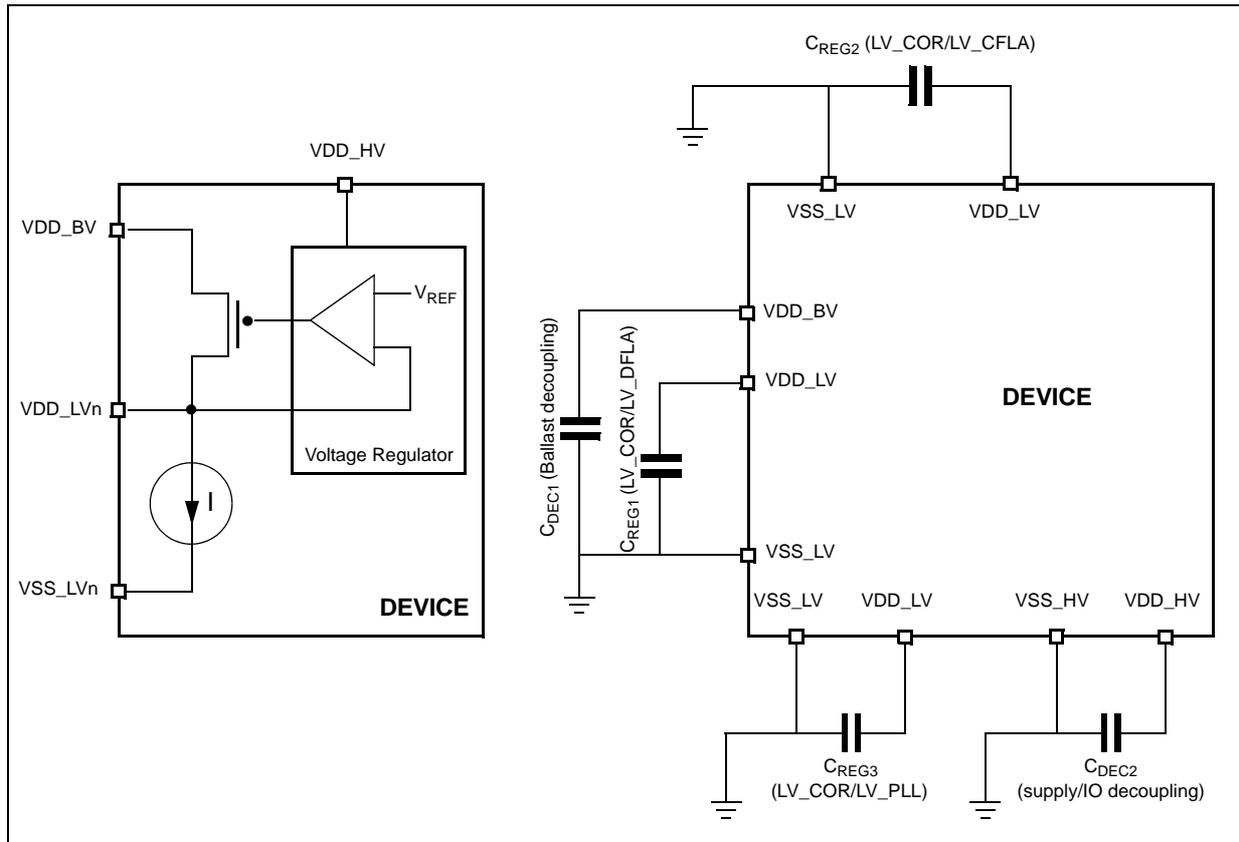
[Table 23](#) provides I/O consumption figures.

In order to ensure device reliability, the average current of the I/O on a single segment should remain below the I_{AVGSEG} maximum value.

Table 24. I/O weight⁽¹⁾ (continued)

Supply segment			Pad	LQFP144/LQFP100				LQFP64 ⁽²⁾			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
LQFP 144	LQFP 100	LQFP 64		SRC ⁽³⁾ = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
2	2	2	PB[13]	10%	—	12%	—	18%	—	21%	—
		—	PD[14]	10%	—	12%	—	—	—	—	—
		2	PB[14]	10%	—	12%	—	18%	—	21%	—
		—	PD[15]	10%	—	11%	—	—	—	—	—
		2	PB[15]	9%	—	11%	—	18%	—	21%	—
			PA[3]	9%	—	11%	—	18%	—	21%	—
	—	—	PG[13]	9%	13%	10%	11%	—	—	—	—
	—	—	PG[12]	9%	12%	10%	11%	—	—	—	—
	—	—	PH[0]	5%	8%	6%	7%	—	—	—	—
	—	—	PH[1]	5%	7%	6%	6%	—	—	—	—
	—	—	PH[2]	5%	6%	5%	6%	—	—	—	—
	—	—	PH[3]	4%	6%	5%	5%	—	—	—	—
	—	—	PG[1]	4%	—	4%	—	—	—	—	—
	—	—	PG[0]	3%	4%	4%	4%	—	—	—	—
3	—	—	PF[15]	3%	—	4%	—	—	—	—	—
	—	—	PF[14]	4%	5%	5%	5%	—	—	—	—
	—	—	PE[13]	4%	—	5%	—	—	—	—	—
	3	2	PA[7]	5%	—	6%	—	16%	—	19%	—
			PA[8]	5%	—	6%	—	16%	—	19%	—
			PA[9]	5%	—	6%	—	15%	—	18%	—
			PA[10]	6%	—	7%	—	15%	—	18%	—
			PA[11]	6%	—	8%	—	14%	—	17%	—
	—	—	PE[12]	7%	—	8%	—	—	—	—	—
	—	—	PG[14]	7%	—	8%	—	—	—	—	—
	—	—	PG[15]	7%	10%	8%	9%	—	—	—	—
	—	—	PE[14]	7%	—	8%	—	—	—	—	—
	—	—	PE[15]	7%	9%	8%	8%	—	—	—	—
	—	—	PG[10]	6%	—	8%	—	—	—	—	—
—	—	PG[11]	6%	9%	7%	8%	—	—	—	—	
3	2	PC[3]	6%	—	7%	—	7%	—	9%	—	
		PC[2]	6%	8%	7%	7%	6%	9%	8%	8%	

Figure 9. Voltage regulator capacitance connection



The internal voltage regulator requires external capacitance (C_{REGn}) to be connected to the device in order to provide a stable low voltage digital supply to the device. Capacitances should be placed on the board as near as possible to the associated pins. Care should also be taken to limit the serial inductance of the board to less than 5 nH.

Each decoupling capacitor must be placed between each of the three V_{DD_LV}/V_{SS_LV} supply pairs to ensure stable voltage (see [Section 3.13: Recommended operating conditions](#)).

The internal voltage regulator requires a controlled slew rate of both V_{DD_HV} and V_{DD_BV} as described in [Figure 10](#).

$$ESR_{STDBY}(MAX) = |\Delta V_{DD}(STDBY)| / (I_{DD_BV} - 200 \text{ mA}) = (30 \text{ mV}) / (100 \text{ mA}) = 0.3 \Omega$$

$$C_{STDBY}(MIN) = (I_{DD_BV} - 200 \text{ mA}) / dV_{DD}(STDBY) / dt = (300 \text{ mA} - 200 \text{ mA}) / (15 \text{ mV} / \mu\text{s}) = 6.7 \mu\text{F}$$

In case optimization is required, $C_{STDBY}(MIN)$ and $ESR_{STDBY}(MAX)$ should be calculated based on the regulator characteristics as well as the board V_{DD} plane characteristics.

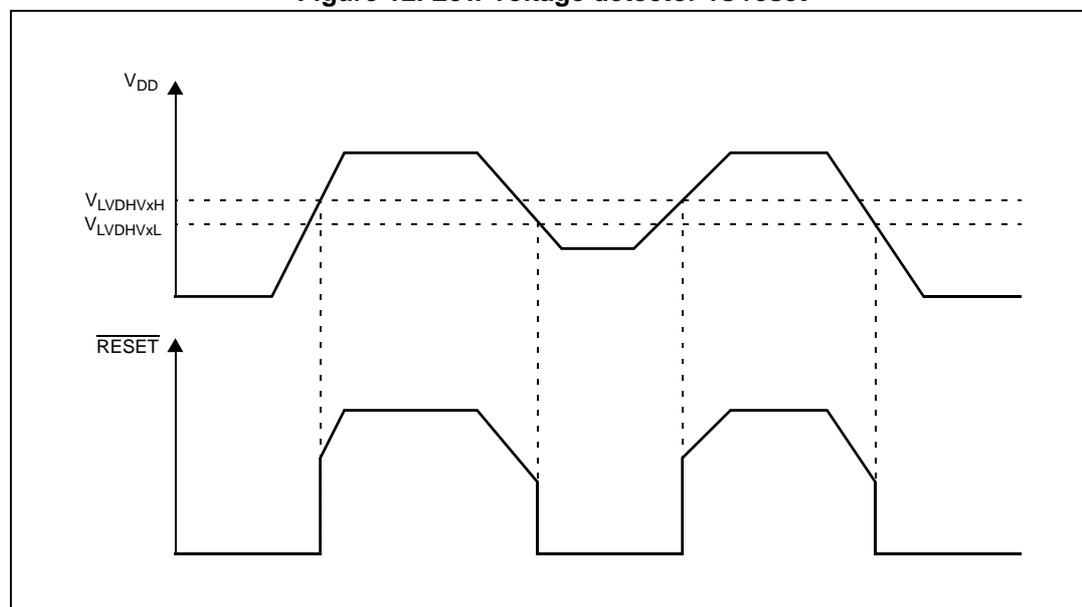
3.17.2 Low voltage detector electrical characteristics

The device implements a Power-on Reset (POR) module to ensure correct power-up initialization, as well as four low voltage detectors (LVDs) to monitor the V_{DD} and the V_{DD_LV} voltage while device is supplied:

- POR monitors V_{DD} during the power-up phase to ensure device is maintained in a safe reset state (refer to RGM Destructive Event Status (RGM_DES) Register flag F_POR in device reference manual)
- LVDHV3 monitors V_{DD} to ensure device reset below minimum functional supply (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD27 in device reference manual)
- LVDHV5 monitors V_{DD} when application uses device in the $5.0 \text{ V} \pm 10\%$ range (refer to RGM Functional Event Status (RGM_FES) Register flag F_LVD45 in device reference manual)
- LVDLVCOR monitors power domain No. 1 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD1 in device reference manual)
- LVDLVBKP monitors power domain No. 0 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD0 in device reference manual)

Note: When enabled, power domain No. 2 is monitored through LVDLVBKP.

Figure 12. Low voltage detector vs reset



Note: [Figure 12: Low voltage detector vs reset](#) does not apply to LVDHV5 low voltage detector because LVDHV5 is automatically disabled during reset and it must be enabled by software again. Once the device is forced to reset by LVDHV5, the LVDHV5 is disabled and reset is

Table 28. Power consumption on VDD_BV and VDD_HV (continued)

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit	
				Min	Typ	Max		
I _{DDSTOP}	CC	STOP mode current ⁽⁷⁾	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	180	700 ⁽⁸⁾	μA
				T _A = 55 °C	—	500	—	
				T _A = 85 °C	—	1	6 ⁽⁸⁾	mA
				T _A = 105 °C	—	2	9 ⁽⁸⁾	
				T _A = 125 °C	—	4.5	12 ⁽⁸⁾	
I _{DDSTDBY2}	CC	STANDBY2 mode current ⁽⁹⁾	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	30	100	μA
				T _A = 55 °C	—	75	—	
				T _A = 85 °C	—	180	700	
				T _A = 105 °C	—	315	1000	
				T _A = 125 °C	—	560	1700	
I _{DDSTDBY1}	CC	STANDBY1 mode current ⁽¹⁰⁾	Slow internal RC oscillator (128 kHz) running	T _A = 25 °C	—	20	60	μA
				T _A = 55 °C	—	45	—	
				T _A = 85 °C	—	100	350	
				T _A = 105 °C	—	165	500	
				T _A = 125 °C	—	280	900	

- V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified
- I_{DDMAX} is drawn only from the V_{DD_BV} pin. Running consumption does not include I/Os toggling which is highly dependent on the application. The given value is thought to be a worst case value with all peripherals running, and code fetched from code flash while modify operation ongoing on data flash. Notice that this value can be significantly reduced by application: switch off not used peripherals (default), reduce peripheral frequency through internal prescaler, fetch from RAM most used functions, use low power mode when possible.
- Higher current may be sinked by device during power-up and standby exit. Please refer to in rush current on [Table 26](#).
- I_{DDRUN} is drawn only from the V_{DD_BV} pin. RUN current measured with typical application with accesses on both flash and RAM.
- Only for the "P" classification: Data and Code Flash in Normal Power. Code fetched from RAM: Serial IPs CAN and LIN in loop back mode, DSPI as Master, PLL as system Clock (4 x Multiplier) peripherals on (eMIOS/CTU/ADC) and running at max frequency, periodic SW/WDG timer reset enabled.
- Data Flash Power Down. Code Flash in Low Power. SIRC (128 kHz) and FIRC (16 MHz) on. 10 MHz XTAL clock. FlexCAN: instances: 0, 1, 2 ON (clocked but not reception or transmission), instances: 4, 5, 6 clock gated. LINFlex: instances: 0, 1, 2 ON (clocked but not reception or transmission), instance: 3 clock gated. eMIOS: instance: 0 ON (16 channels on PA[0]–PA[11] and PC[12]–PC[15]) with PWM 20 kHz, instance: 1 clock gated. DSPI: instance: 0 (clocked but no communication). RTC/API ON. PIT ON. STM ON. ADC ON but not conversion except 2 analog watchdog.
- Only for the "P" classification: No clock, FIRC (16 MHz) off, SIRC (128 kHz) on, PLL off, HPVreg off, ULPVreg/LPVreg on. All possible peripherals off and clock gated. Flash in power down mode.
- When going from RUN to STOP mode and the core consumption is > 6 mA, it is normal operation for the main regulator module to be kept on by the on-chip current monitoring circuit. This is most likely to occur with junction temperatures exceeding 125 °C and under these circumstances, it is possible for the current to initially exceed the maximum STOP specification by up to 2 mA. After entering stop, the application junction temperature will reduce to the ambient level and the main regulator will be automatically switched off when the load current is below 6 mA.
- Only for the "P" classification: ULPreg on, HP/LPVreg off, 32 KB RAM on, device configured for minimum consumption, all possible modules switched off.
- ULPreg on, HP/LPVreg off, 8 KB RAM on, device configured for minimum consumption, all possible modules switched off.

Figure 13. Crystal oscillator and resonator connection scheme

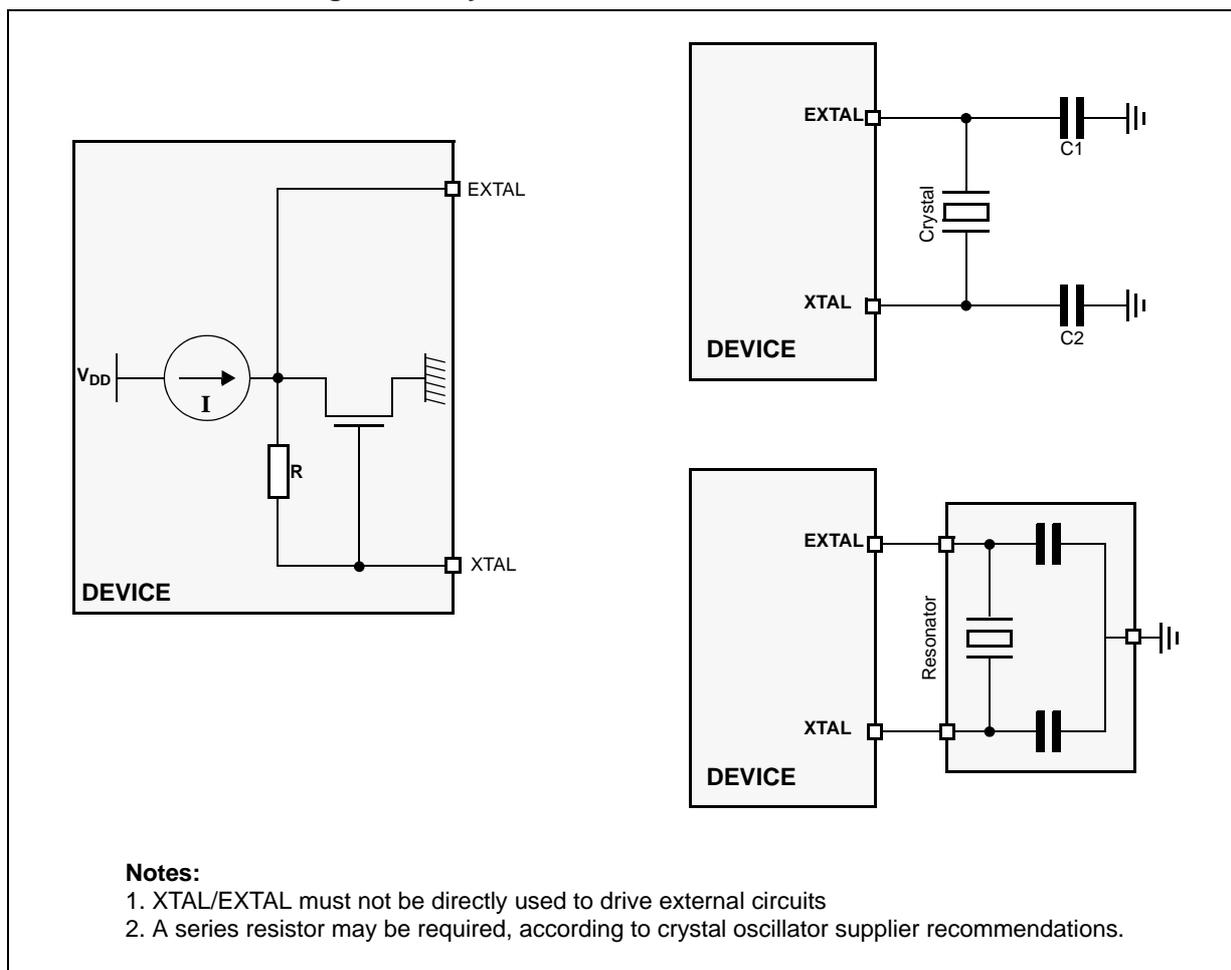


Table 37. Crystal description

Nominal frequency (MHz)	NDK crystal reference	Crystal equivalent series resistance ESR Ω	Crystal motional capacitance (C_m) fF	Crystal motional inductance (L_m) mH	Load on xtalin/xtalout $C1 = C2$ (pF) ⁽¹⁾	Shunt capacitance between xtalout and xtalin $C0$ ⁽²⁾ (pF)
4	NX8045GB	300	2.68	591.0	21	2.93
8	NX5032GA	300	2.46	160.7	17	3.01
10		150	2.93	86.6	15	2.91
12		120	3.11	56.5	15	2.93
16		120	3.90	25.3	10	3.00

1. The values specified for C1 and C2 are the same as used in simulations. It should be ensured that the testing includes all the parasitics (from the board, probe, crystal, etc.) as the AC / transient behavior depends upon them.
2. The value of C0 specified here includes 2 pF additional capacitance for parasitics (to be seen with bond-pads, package, etc.).

Figure 17. Slow external crystal oscillator (32 kHz) timing diagram

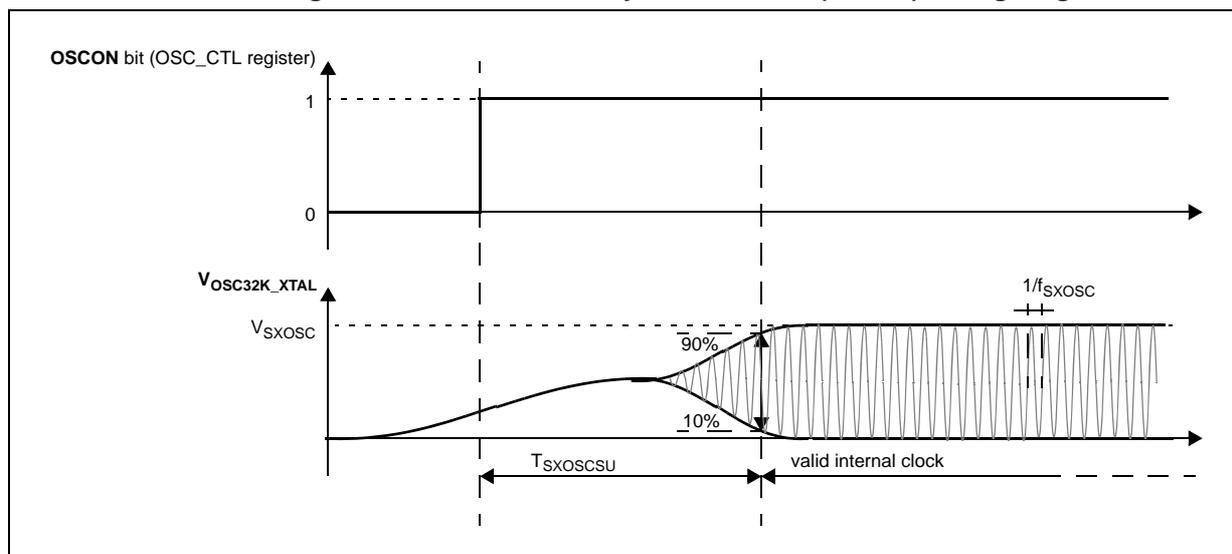


Table 40. Slow external crystal oscillator (32 kHz) electrical characteristics

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit	
				Min	Typ	Max		
f _{SXOSC}	SR	—	Slow external crystal oscillator frequency	—	32	32.768	40	kHz
V _{SXOSC}	CC	T	Oscillation amplitude	—	—	2.1	—	V
I _{SXOSCBIAS}	CC	T	Oscillation bias current	—	—	2.5	—	μA
I _{SXOSC}	CC	T	Slow external crystal oscillator consumption	—	—	—	8	μA
T _{SXOSCSU}	CC	T	Slow external crystal oscillator start-up time	—	—	—	2 ⁽²⁾	s

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified. Values are specified for no neighbor GPIO pin activity. If oscillator is enabled (OSC32K_XTAL and OSC32K_EXTAL pins), neighboring pins should not toggle.
2. Start-up time has been measured with EPSON TOYOCOM MC306 crystal. Variation may be seen with other crystal.

3.23 FMPLL electrical characteristics

The device provides a frequency-modulated phase-locked loop (FMPLL) module to generate a fast system clock from the main oscillator driver.

Table 41. FMPLL electrical characteristics

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit	
				Min	Typ	Max		
f _{PLLIN}	SR	—	FMPLL reference clock ⁽²⁾	—	4	—	64	MHz
Δ _{PLLIN}	SR	—	FMPLL reference clock duty cycle ⁽²⁾	—	40	—	60	%
f _{PLLOUT}	CC	D	FMPLL output clock frequency	—	16	—	64	MHz

Table 41. FMPLL electrical characteristics (continued)

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit			
				Min	Typ	Max				
f _{VCO} ⁽³⁾	CC	P	VCO frequency without frequency modulation	—	256	—	512	MHz		
		C	VCO frequency with frequency modulation	—	245	—	533			
f _{CPU}	SR	—	System clock frequency	—	—	—	64	MHz		
f _{FREE}	CC	P	Free-running frequency	—	20	—	150	MHz		
t _{LOCK}	CC	P	FMPLL lock time	Stable oscillator (f _{PLLIN} = 16 MHz)			—	40	100	μs
Δt _{STJIT}	CC	—	FMPLL short term jitter ⁽⁴⁾	f _{sys} maximum			—4	—	4	%
Δt _{LTJIT}	CC	—	FMPLL long term jitter	f _{PLLIN} = 16 MHz (resonator), f _{PLLCLK} @ 64 MHz, 4000 cycles			—	—	10	ns
I _{PLL}	CC	C	FMPLL consumption	T _A = 25 °C			—	—	4	mA

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = -40 to 125 °C, unless otherwise specified.
2. PLLIN clock retrieved directly from FXOSC clock. Input characteristics are granted when oscillator is used in functional mode. When bypass mode is used, oscillator input clock should verify f_{PLLIN} and Δ_{PLLIN}.
3. Frequency modulation is considered ±4%
4. Short term jitter is measured on the clock rising edge at cycle n and n+4.

3.24 Fast internal RC oscillator (16 MHz) electrical characteristics

The device provides a 16 MHz fast internal RC oscillator. This is used as the default clock at the power-up of the device.

Table 42. Fast internal RC oscillator (16 MHz) electrical characteristics

Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit		
				Min	Typ	Max			
f _{FIRC}	CC	P	Fast internal RC oscillator high frequency	T _A = 25 °C, trimmed			MHz		
	SR			—	16	—			
I _{FIRC} _{RUN} ⁽²⁾	CC	T	Fast internal RC oscillator high frequency current in running mode	T _A = 25 °C, trimmed			μA		
I _{FIRC} _{PWD}	CC	D	Fast internal RC oscillator high frequency current in power down mode	T _A = 125 °C			μA		
I _{FIRC} _{STOP}	CC	T	Fast internal RC oscillator high frequency and system clock current in stop mode	T _A = 25 °C	sysclk = off	—	500	—	μA
					sysclk = 2 MHz	—	600	—	
					sysclk = 4 MHz	—	700	—	
					sysclk = 8 MHz	—	900	—	
					sysclk = 16 MHz	—	1250	—	

To preserve the accuracy of the A/D converter, it is necessary that analog input pins have low AC impedance. Placing a capacitor with good high frequency characteristics at the input pin of the device can be effective: the capacitor should be as large as possible, ideally infinite. This capacitor contributes to attenuating the noise present on the input pin; furthermore, it sources charge during the sampling phase, when the analog signal source is a high-impedance source.

A real filter can typically be obtained by using a series resistance with a capacitor on the input pin (simple RC filter). The RC filtering may be limited according to the value of source impedance of the transducer or circuit supplying the analog signal to be measured. The filter at the input pins must be designed taking into account the dynamic characteristics of the input signal (bandwidth) and the equivalent input impedance of the ADC itself.

In fact a current sink contributor is represented by the charge sharing effects with the sampling capacitance: being C_S and C_{p2} substantially two switched capacitances, with a frequency equal to the conversion rate of the ADC, it can be seen as a resistive path to ground. For instance, assuming a conversion rate of 1 MHz, with C_S+C_{p2} equal to 3 pF, a resistance of 330 k Ω is obtained ($R_{EQ} = 1 / (f_c \times (C_S+C_{p2}))$), where f_c represents the conversion rate at the considered channel). To minimize the error induced by the voltage partitioning between this resistance (sampled voltage on C_S+C_{p2}) and the sum of $R_S + R_F$, the external circuit must be designed to respect the [Equation 4](#):

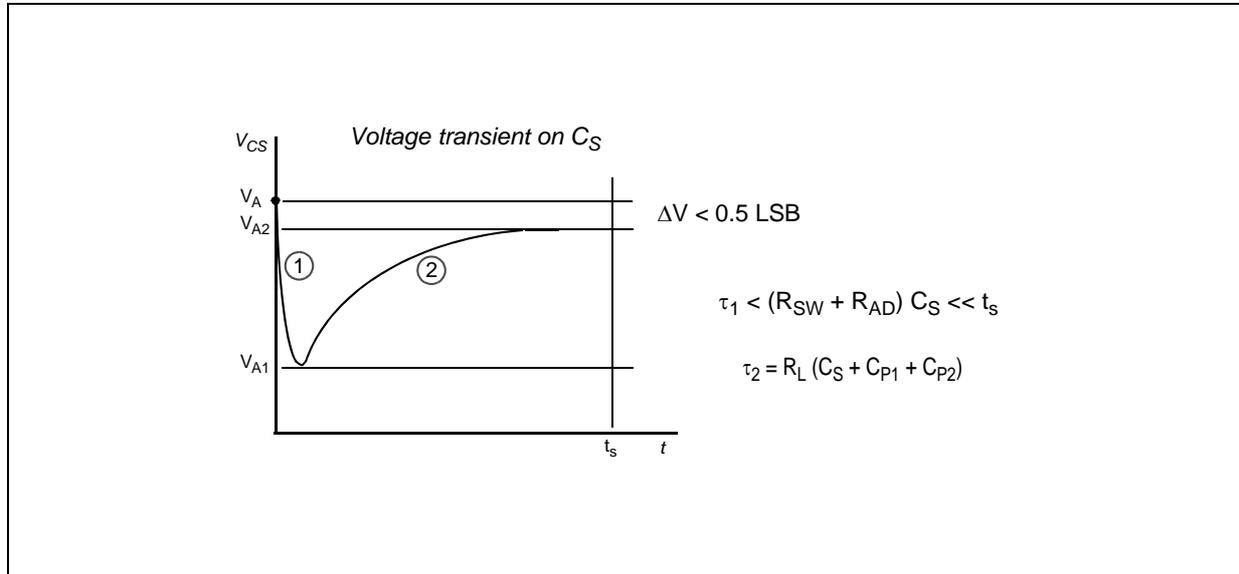
Equation 4

$$V_A \cdot \frac{R_S + R_F}{R_{EQ}} < \frac{1}{2} \text{LSB}$$

[Equation 4](#) generates a constraint for external network design, in particular on a resistive path.

A second aspect involving the capacitance network shall be considered. Assuming the three capacitances C_F , C_{P1} and C_{P2} are initially charged at the source voltage V_A (refer to the equivalent circuit in [Figure 19](#)): A charge sharing phenomenon is installed when the sampling phase is started (A/D switch close).

Figure 21. Transient behavior during sampling phase



In particular two different transient periods can be distinguished:

1. A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Equation 5

$$\tau_1 = (R_{SW} + R_{AD}) \cdot \frac{C_P \cdot C_S}{C_P + C_S}$$

[Equation 5](#) can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time t_s is always much longer than the internal time constant:

Equation 6

$$\tau_1 < (R_{SW} + R_{AD}) \cdot C_S \ll t_s$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S , determining a new value of the voltage V_{A1} on the capacitance according to [Equation 7](#):

Table 47. DSPI characteristics⁽¹⁾

No.	Symbol	C	Parameter	DSPI0/DSPI1			DSPI2			Unit		
				Min	Typ	Max	Min	Typ	Max			
1	t_{SCK}	SR	SCK cycle time	D	Master mode (MTFE = 0)	125	—	—	333	—	—	ns
				D	Slave mode (MTFE = 0)	125	—	—	333	—	—	
				D	Master mode (MTFE = 1)	83	—	—	125	—	—	
				D	Slave mode (MTFE = 1)	83	—	—	125	—	—	
—	f_{DSPI}	SR	D	DSPI digital controller frequency		—	—	f_{CPU}	—	—	f_{CPU}	MHz
—	Δt_{CSC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1→0	Master mode	—	—	130 ⁽²⁾	—	—	15 ⁽³⁾	ns
—	Δt_{ASC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1→1	Master mode	—	—	130 ⁽³⁾	—	—	130 ⁽³⁾	ns
2	$t_{CSCext}^{(4)}$	SR	D	CS to SCK delay	Slave mode	32	—	—	32	—	—	ns
3	$t_{ASCext}^{(5)}$	SR	D	After SCK delay	Slave mode	$1/f_{DSPI} + 5$	—	—	$1/f_{DSPI} + 5$	—	—	ns
4	t_{SDC}	CC	D	SCK duty cycle	Master mode	—	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	ns
					Slave mode	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	—	
5	t_A	SR	D	Slave access time	Slave mode	—	—	$1/f_{DSPI} + 70$	—	—	$1/f_{DSPI} + 130$	ns
6	t_{DI}	SR	D	Slave SOUT disable time	Slave mode	7	—	—	7	—	—	ns
7	t_{PCSC}	SR	D	PCSx to PCSS time		0	—	—	0	—	—	ns
8	t_{PASC}	SR	D	PCSS to PCSx time		0	—	—	0	—	—	ns

Figure 24. DSPI classic SPI timing – master, CPHA = 1

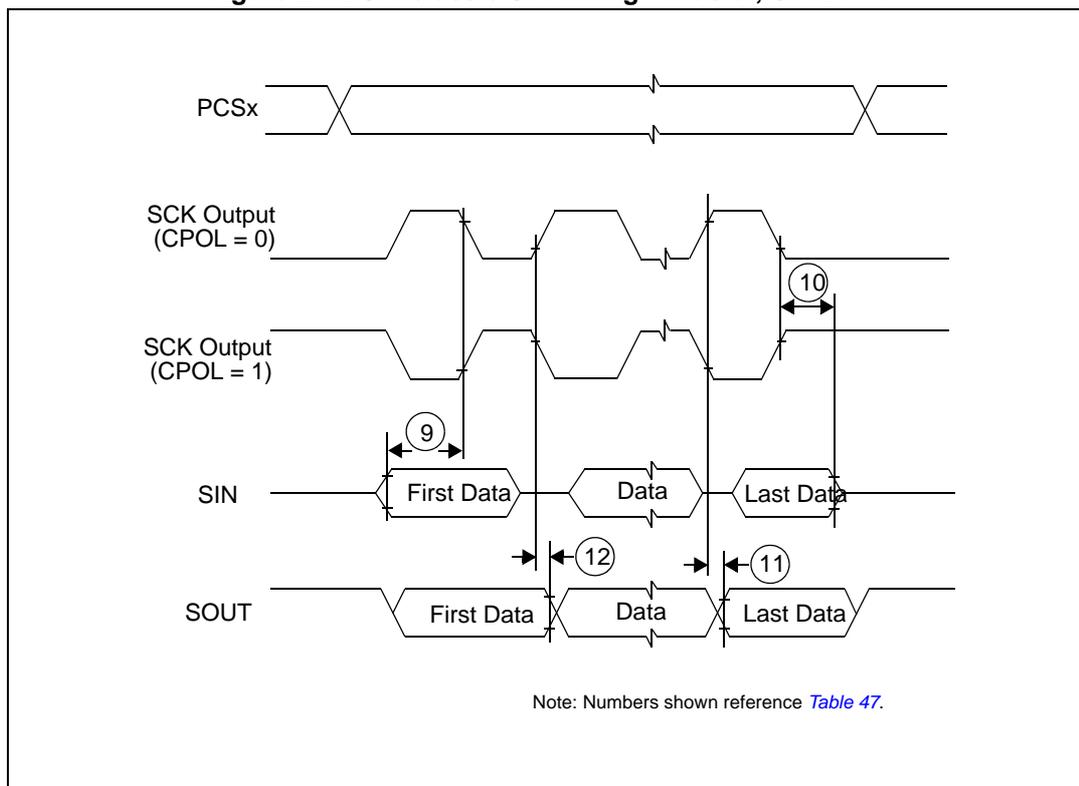


Figure 25. DSPI classic SPI timing – slave, CPHA = 0

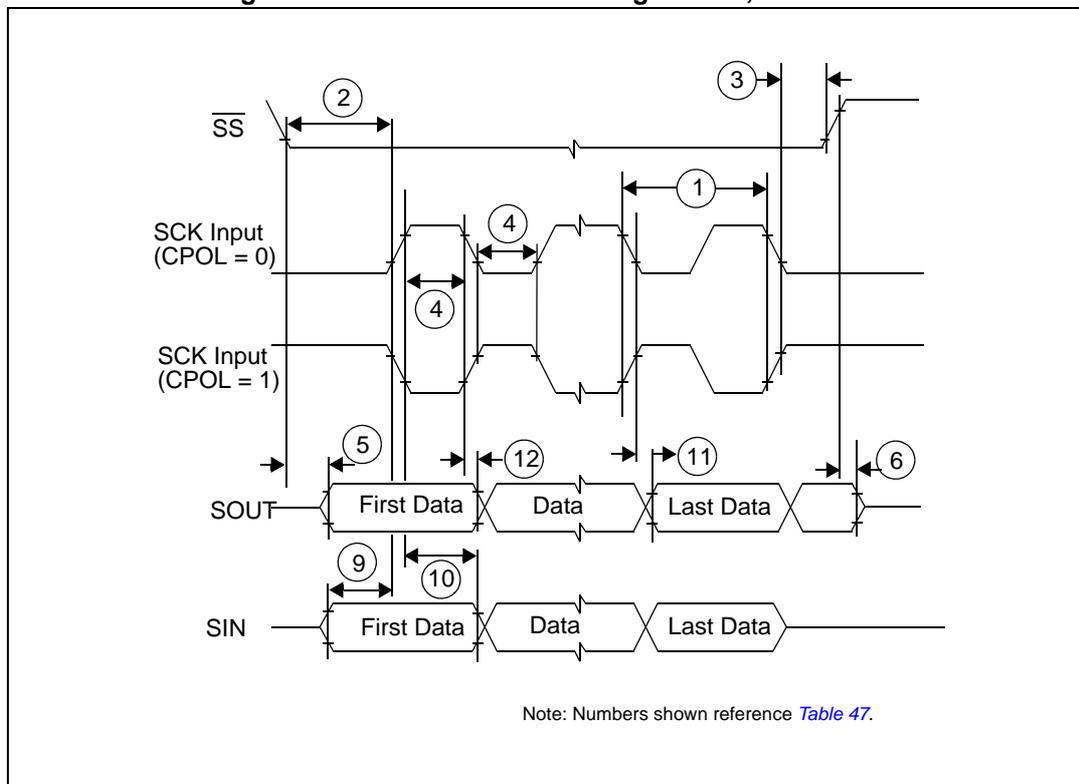


Figure 30. DSPI modified transfer format timing – slave, CPHA = 1

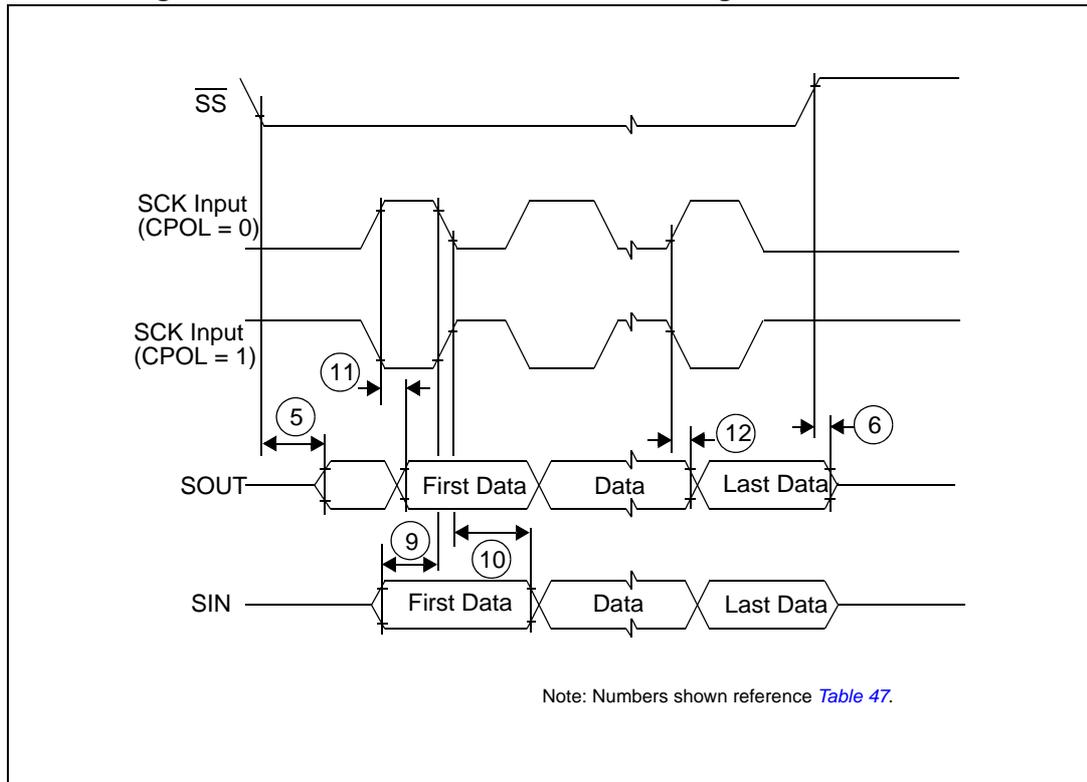
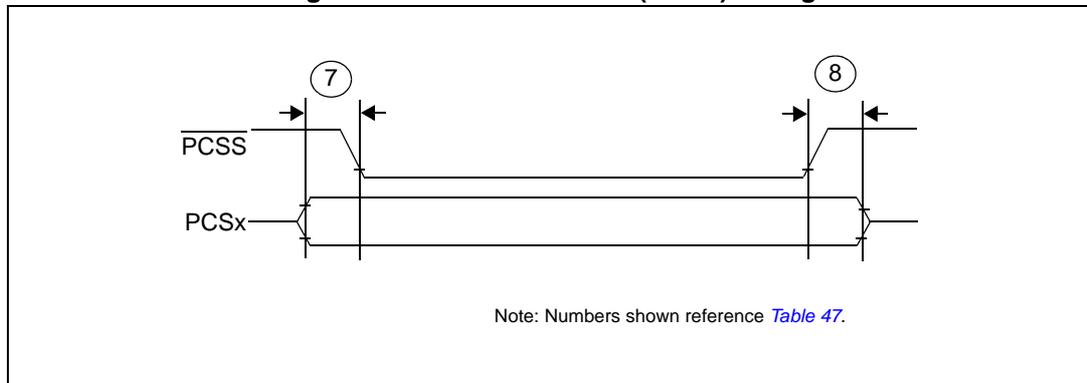


Figure 31. DSPI PCS strobe (PCSS) timing



3.27.3 Nexus characteristics

Table 48. Nexus characteristics

No.	Symbol	C	Parameter	Value			Unit
				Min	Typ	Max	
1	t_{TCYC}	CC	D TCK cycle time	64	—	—	ns
2	t_{MCYC}	CC	D MCKO cycle time	32	—	—	ns
3	t_{MDOV}	CC	D MCKO low to MDO data valid	—	—	8	ns

4.2.2 LQFP100

Figure 35. LQFP100 package mechanical drawing

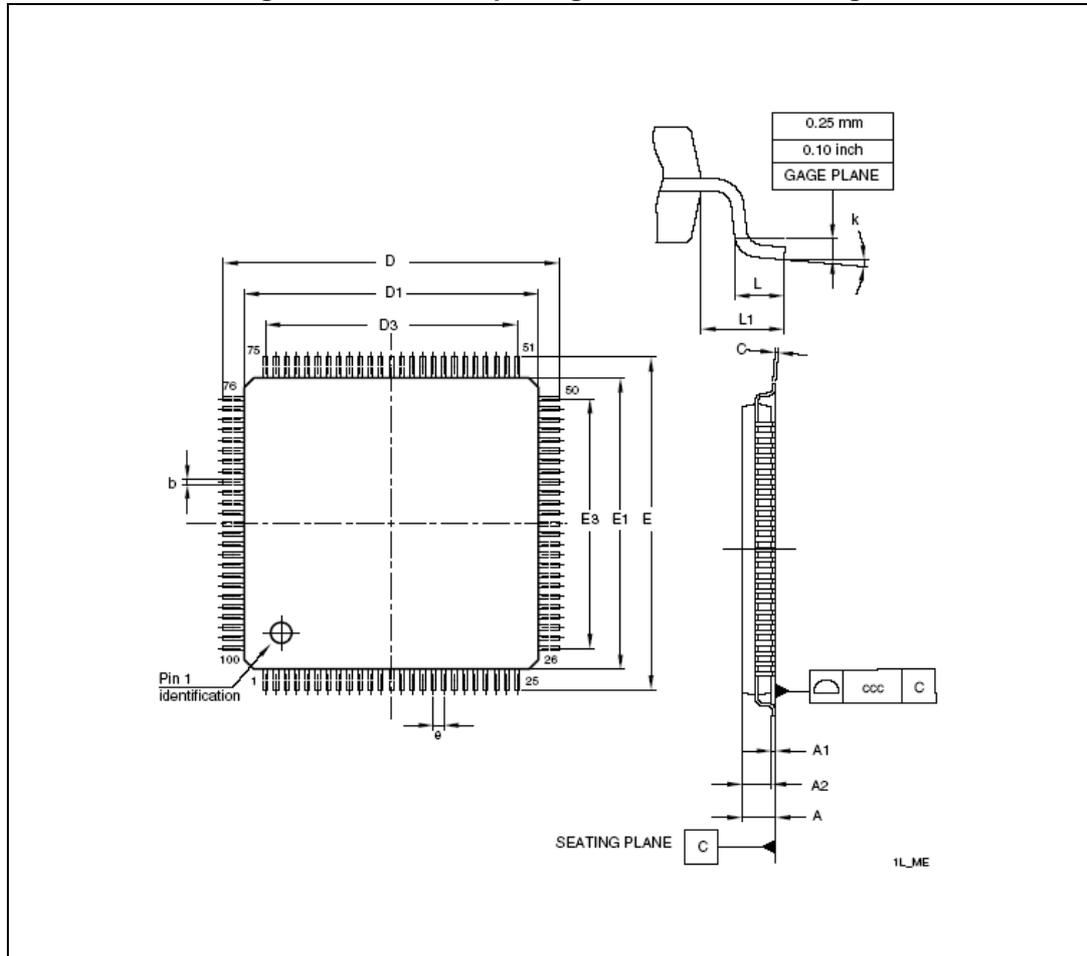


Table 51. LQFP100 mechanical data

Symbol	mm			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	—	—	1.600	—	—	0.0630
A1	0.050	—	0.150	0.0020	—	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	—	0.200	0.0035	—	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	—	12.000	—	—	0.4724	—
E	15.800	16.000	16.200	0.6220	0.6299	0.6378