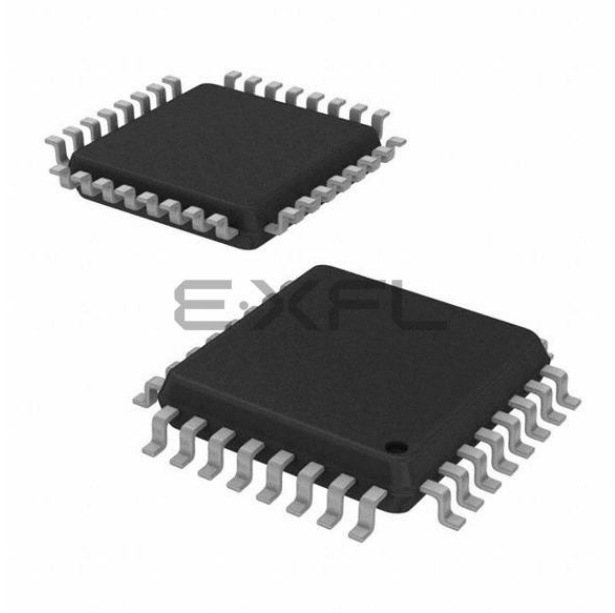

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I²C), SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	29
Program Memory Size	8KB (8K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	1.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 21x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	32-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051t612-gq

1. System Overview

C8051T610/1/2/3/4/5/6/7 devices are fully integrated, mixed-signal, system-on-a-chip MCUs. Highlighted features are listed below. Refer to Table 2.1 for specific product feature selection and part ordering numbers.

- High-speed pipelined 8051-compatible microcontroller core (up to 25 MIPS)
- In-system, full-speed, non-intrusive debug interface (on-chip)
- C8051F310 ISP Flash device is available for quick in-system code development
- 10-bit 500 ksps Single-ended ADC with analog multiplexer and integrated temperature sensor
- Precision calibrated 24.5 MHz internal oscillator
- 16 k or 8 k of on-chip Byte-Programmable EPROM—(512 bytes are reserved on 16k version)
- 1280 bytes of on-chip RAM
- SMBus/I^2C, SPI, and Enhanced UART serial interfaces implemented in hardware
- Four general-purpose 16-bit timers
- Programmable Counter/Timer Array (PCA) with five capture/compare modules and Watchdog Timer function
- On-chip Power-On Reset and V_{DD} Monitor
- On-chip Voltage Comparators (2)
- 29/25/21 Port I/O

With on-chip power-on reset, V_{DD} monitor, watchdog timer, and clock oscillator, the C8051T610/1/2/3/4/5/6/7 devices are truly stand-alone, system-on-a-chip solutions. User software has complete control of all peripherals and may individually shut down any or all peripherals for power savings.

Code written for the C8051T610/1/2/3/4/5/6/7 family of processors will run on the C8051F310 Mixed-Signal ISP Flash microcontroller, providing a quick, cost-effective way to develop code without requiring special emulator circuitry. The C8051T610/1/2/3/4/5/6/7 processors include Silicon Laboratories' 2-Wire C2 Debug and Programming interface, which allows non-intrusive (uses no on-chip resources), full speed, in-circuit debugging using the production MCU installed in the final application. This debug logic supports inspection of memory, viewing and modification of special function registers, setting breakpoints, single stepping, and run and halt commands. All analog and digital peripherals are fully functional while debugging using C2. The two C2 interface pins can be shared with user functions, allowing in-system debugging without occupying package pins.

Each device is specified for 1.8–3.6 V operation over the industrial temperature range (–45 to +85 °C). An internal LDO is used to supply the processor core voltage. The Port I/O and $\overline{RST}$ pins are tolerant of input signals up to 5 V. See Table 2.1 for ordering information. Block diagrams of the devices in the C8051T610/1/2/3/4/5/6/7 family are shown in Figure 1.1, Figure 1.2 and Figure 1.3.

8.1. Output Code Formatting

The ADC measures the input voltage with reference to GND. The registers ADC0H and ADC0L contain the high and low bytes of the output conversion code from the ADC at the completion of each conversion. Data can be right-justified or left-justified, depending on the setting of the AD0LJST bit. Conversion codes are represented as 10-bit unsigned integers. Inputs are measured from 0 to VREF x 1023/1024. Example codes are shown below for both right-justified and left-justified data. Unused bits in the ADC0H and ADC0L registers are set to 0.

Input Voltage	Right-Justified ADC0H:ADC0L (AD0LJST = 0)	Left-Justified ADC0H:ADC0L (AD0LJST = 1)
VREF x 1023/1024	0x03FF	0xFFC0
VREF x 512/1024	0x0200	0x8000
VREF x 256/1024	0x0100	0x4000
0	0x0000	0x0000

8.2. 8-Bit Mode

Setting the ADC08BE bit in register ADC0CF to 1 will put the ADC in 8-bit mode. In 8-bit mode, only the 8 MSBs of data are converted, and the ADC0H register holds the results. The AD0LJST bit is ignored for 8-bit mode. 8-bit conversions take two fewer SAR clock cycles than 10-bit conversions, so the conversion is completed faster, and a 500 ksps sampling rate can be achieved with a slower SAR clock.

8.3. Modes of Operation

ADC0 has a maximum conversion speed of 500 ksps. The ADC0 conversion clock is a divided version of the system clock, determined by the AD0SC bits in the ADC0CF register.

8.3.1. Starting a Conversion

A conversion can be initiated in one of six ways, depending on the programmed states of the ADC0 Start of Conversion Mode bits (AD0CM2–0) in register ADC0CN. Conversions may be initiated by one of the following:

1. Writing a 1 to the AD0BUSY bit of register ADC0CN
2. A Timer 0 overflow (i.e., timed continuous conversions)
3. A Timer 2 overflow
4. A Timer 1 overflow
5. A rising edge on the CNVSTR input signal
6. A Timer 3 overflow

Writing a 1 to AD0BUSY provides software control of ADC0 whereby conversions are performed "on-demand". During conversion, the AD0BUSY bit is set to logic 1 and reset to logic 0 when the conversion is complete. The falling edge of AD0BUSY triggers an interrupt (when enabled) and sets the ADC0 interrupt flag (AD0INT). Note: When polling for ADC conversion completions, the ADC0 interrupt flag (AD0INT) should be used. Converted data is available in the ADC0 data registers, ADC0H:ADC0L, when bit AD0INT is logic 1. Note that when Timer 2 or Timer 3 overflows are used as the conversion source, Low Byte overflows are used if Timer 2/3 is in 8-bit mode; High byte overflows are used if Timer 2/3 is in 16-bit mode. See Section “25. Timers” on page 170 for timer configuration.

Important Note About Using CNVSTR: The CNVSTR input pin also functions as a Port I/O pin. When the CNVSTR input is used as the ADC0 conversion source, the associated pin should be skipped by the Digital Crossbar. See Section “21. Port Input/Output” on page 113 for details on Port I/O configuration.

SFR Definition 8.7. ADC0LTH: ADC0 Less-Than Data High Byte

Bit	7	6	5	4	3	2	1	0
Name	ADC0LTH[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC6

Bit	Name	Function
7:0	ADC0LTH[7:0]	**ADC0 Less-Than Data Word High-Order Bits.**

SFR Definition 8.8. ADC0LTL: ADC0 Less-Than Data Low Byte

Bit	7	6	5	4	3	2	1	0
Name	ADC0LTL[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC5

Bit	Name	Function
7:0	ADC0LTL[7:0]	**ADC0 Less-Than Data Word Low-Order Bits.**

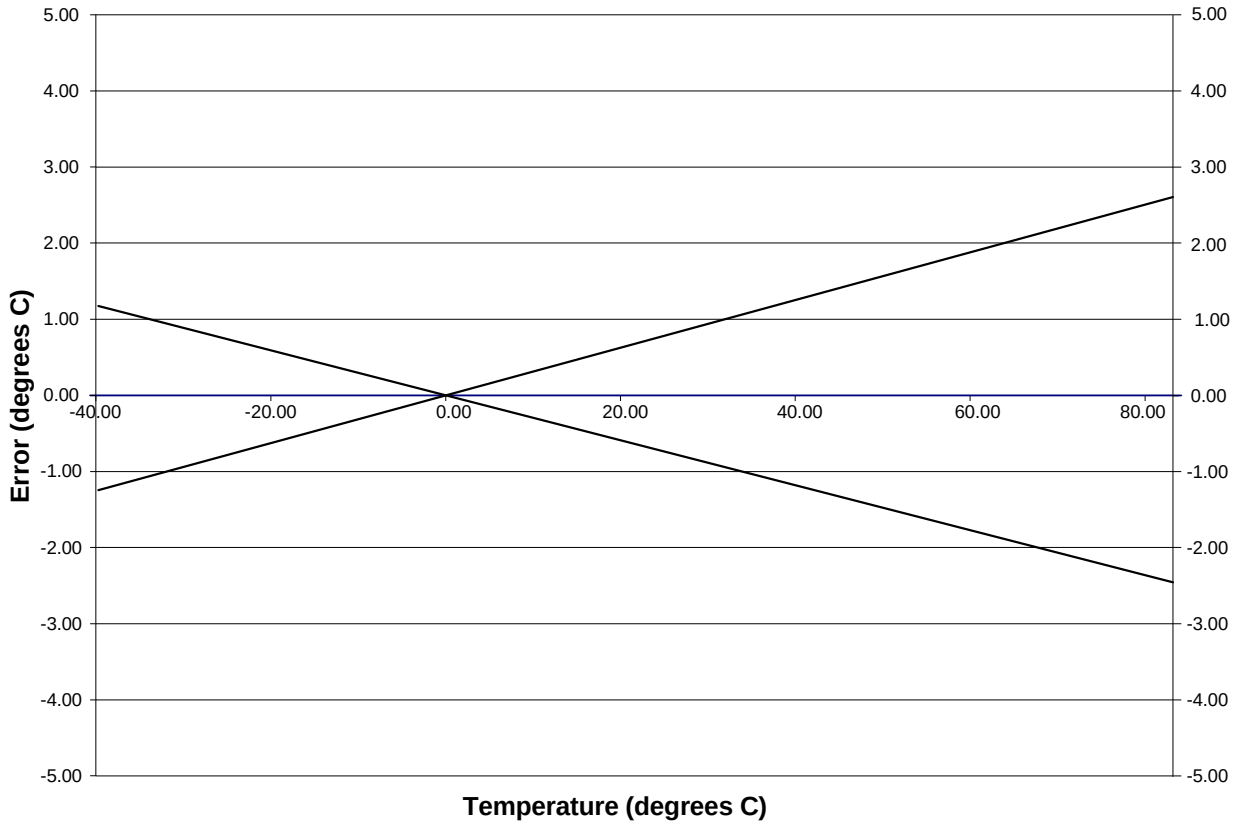

Figure 9.2. Temperature Sensor Error with 1-Point Calibration at 0 Celsius

SFR Definition 12.5. CPT0MX: Comparator0 MUX Selection

Bit	7	6	5	4	3	2	1	0
Name			CMX0N[1:0]				CMX0P[1:0]	
Type	R	R	R/W		R	R	R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x9F

Bit	Name	Function
7:6	Unused	Unused, Read = 00b; Write = Don't Care
5:4	CMX0N[1:0]	**Comparator0 Negative Input MUX Selection.** 00: P1.1 01: P1.5 10: P2.1 11: P2.5
3:2	Unused	Unused, Read = 00b; Write = Don't Care
1:0	CMX0P[1:0]	**Comparator0 Positive Input MUX Selection.** 00: P1.0 01: P1.4 10: P2.0 11: P2.4

16. Interrupts

The C8051T610/1/2/3/4/5/6/7 includes an extended interrupt system supporting a total of 14 interrupt sources with two priority levels. The allocation of interrupt sources between on-chip peripherals and external input pins varies according to the specific version of the device. Each interrupt source has one or more associated interrupt-pending flag(s) located in an SFR. When a peripheral or external source meets a valid interrupt condition, the associated interrupt-pending flag is set to logic 1.

If interrupts are enabled for the source, an interrupt request is generated when the interrupt-pending flag is set. As soon as execution of the current instruction is complete, the CPU generates an LCALL to a predetermined address to begin execution of an interrupt service routine (ISR). Each ISR must end with an RETI instruction, which returns program execution to the next instruction that would have been executed if the interrupt request had not occurred. If interrupts are not enabled, the interrupt-pending flag is ignored by the hardware and program execution continues as normal. (The interrupt-pending flag is set to logic 1 regardless of the interrupt's enable/disable state.)

Each interrupt source can be individually enabled or disabled through the use of an associated interrupt enable bit in an SFR (IE–EIE1). However, interrupts must first be globally enabled by setting the EA bit (IE.7) to logic 1 before the individual interrupt enables are recognized. Setting the EA bit to logic 0 disables all interrupt sources regardless of the individual interrupt-enable settings.

Note: Any instruction that clears a bit to disable an interrupt should be immediately followed by an instruction that has two or more opcode bytes. Using EA (global interrupt enable) as an example:

```
// in 'C':
EA = 0; // clear EA bit.
EA = 0; // this is a dummy instruction with two-byte opcode.

; in assembly:
CLR EA ; clear EA bit.
CLR EA ; this is a dummy instruction with two-byte opcode.
```

For example, if an interrupt is posted during the execution phase of a "CLR EA" opcode (or any instruction which clears a bit to disable an interrupt source), and the instruction is followed by a single-cycle instruction, the interrupt may be taken. However, a read of the enable bit will return a '0' inside the interrupt service routine. When the bit-clearing opcode is followed by a multi-cycle instruction, the interrupt will not be taken.

Some interrupt-pending flags are automatically cleared by the hardware when the CPU vectors to the ISR. However, most are not cleared by the hardware and must be cleared by software before returning from the ISR. If an interrupt-pending flag remains set after the CPU completes the return-from-interrupt (RETI) instruction, a new interrupt request will be generated immediately and the CPU will re-enter the ISR after the completion of the next instruction.

17. EPROM Memory

Electrically programmable read-only memory (EPROM) is included on-chip for program code storage. The EPROM memory can be programmed via the C2 debug and programming interface when a special programming voltage is applied to the V_{PP} pin. Each location in EPROM memory is programmable only once (i.e., non-erasable). Table 7.6 on page 34 shows the EPROM specifications.

17.1. Programming and Reading the EPROM Memory

Reading and writing the EPROM memory is accomplished through the C2 programming and debug interface. When creating hardware to program the EPROM, it is necessary to follow the programming steps listed below. Refer to the “C2 Interface Specification” available at http://www.silabs.com for details on communicating via the C2 interface. Section “27. C2 Interface” on page 208 has information about C2 register addresses for the C8051T610/1/2/3/4/5/6/7.

17.1.1. EPROM Write Procedure

1. Reset the device using the $\overline{\text{RST}}$ pin.
2. Wait at least 20 µs before sending the first C2 command.
3. Place the device in core reset: Write 0x04 to the DEVCTL register.
4. Set the device to program mode (1st step): Write 0x40 to the EPCTL register.
5. Set the device to program mode (2nd step): Write 0x4A to the EPCTL register.
   Note: Prior to date code 1119, 0x58 should be written to EPCTL.
6. Apply the VPP programming Voltage.
7. Write the first EPROM address for programming to EPADDRH and EPADDRL.
8. Write a data byte to EPDAT. EPADDRH:L will increment by 1 after this write.
9. Use a C2 Address Read command to poll for write completion.
10. (Optional) Check the ERROR bit in register EPSTAT and abort the programming operation if necessary.
11. If programming is not finished, return to Step 8 to write the next address in sequence, or return to Step 7 to program a new address.
12. Remove the VPP programming Voltage.
13. Remove program mode (1st step): Write 0x40 to the EPCTL register.
14. Remove program mode (2nd step): Write 0x00 to the EPCTL register.
15. Reset the device: Write 0x02 and then 0x00 to the DEVCTL register.

Important Note: There is a finite amount of time which V_{PP} can be applied without damaging the device, which is cumulative over the life of the device. Refer to Table 7.1 on page 31 for the V_{PP} timing specification.

18. Power Management Modes

The C8051T610/1/2/3/4/5/6/7 devices have two software programmable power management modes: idle, and stop. Idle mode halts the CPU while leaving the peripherals and clocks active. In stop mode, the CPU is halted, all interrupts and timers (except the missing clock detector) are inactive, and the internal oscillator is stopped (analog peripherals remain in their selected states; the external oscillator is not affected). Since clocks are running in idle mode, power consumption is dependent upon the system clock frequency and the number of peripherals left in active mode before entering Idle. Stop mode consumes the least power because the majority of the device is shut down with no clocks active. SFR Definition 18.1 describes the Power Control Register (PCON) used to control the C8051T610/1/2/3/4/5/6/7's stop and idle power management modes.

Although the C8051T610/1/2/3/4/5/6/7 has idle and stop modes available, more control over the device power can be achieved by enabling/disabling individual peripherals as needed. Each analog peripheral can be disabled when not in use and placed in low power mode. Digital peripherals, such as timers or serial buses, draw little power when they are not in use.

18.1. Idle Mode

Setting the Idle Mode Select bit (PCON.0) causes the hardware to halt the CPU and enter idle mode as soon as the instruction that sets the bit completes execution. All internal registers and memory maintain their original data. All analog and digital peripherals can remain active during idle mode.

Idle mode is terminated when an enabled interrupt is asserted or a reset occurs. The assertion of an enabled interrupt will cause the Idle Mode Selection bit (PCON.0) to be cleared and the CPU to resume operation. The pending interrupt will be serviced and the next instruction to be executed after the return from interrupt (RETI) will be the instruction immediately following the one that set the Idle Mode Select bit. If Idle mode is terminated by an internal or external reset, the CIP-51 performs a normal reset sequence and begins program execution at address 0x0000.

If the instruction following the write of the IDLE bit is a single-byte instruction and an interrupt occurs during the execution phase of the instruction that sets the IDLE bit, the CPU may not wake from idle mode when a future interrupt occurs. Therefore, instructions that set the IDLE bit should be followed by an instruction that has two or more opcode bytes, for example:

```
// in 'C':
PCON |= 0x01;                  // set IDLE bit
PCON = PCON;                   // ... followed by a 3-cycle dummy instruction

; in assembly:
ORL PCON, #01h                 ; set IDLE bit
MOV PCON, PCON                 ; ... followed by a 3-cycle dummy instruction
```

If enabled, the watchdog timer (WDT) will eventually cause an internal watchdog reset and thereby terminate the idle mode. This feature protects the system from an unintended permanent shutdown in the event of an inadvertent write to the PCON register. If this behavior is not desired, the WDT may be disabled by software prior to entering the idle mode if the WDT was initially configured to allow this operation. This provides the opportunity for additional power savings, allowing the system to remain in the idle mode indefinitely, waiting for an external stimulus to wake up the system. Refer to Section "19.6. PCA Watchdog Timer Reset" on page 104 for more information on the use and configuration of the WDT.

19. Reset Sources

Reset circuitry allows the controller to be easily placed in a predefined default condition. On entry to this reset state, the following occur:

- CIP-51 halts program execution
- Special Function Registers (SFRs) are initialized to their defined reset values
- External Port pins are forced to a known state
- Interrupts and timers are disabled

All SFRs are reset to the predefined values noted in the SFR detailed descriptions. The contents of internal data memory are unaffected during a reset; any previously stored data is preserved. However, since the stack pointer SFR is reset, the stack is effectively lost, even though the data on the stack is not altered.

The Port I/O latches are reset to 0xFF (all logic ones) in open-drain mode. Weak pullups are enabled during and after the reset. For V_{DD} Monitor and power-on resets, the $\overline{RST}$ pin is driven low until the device exits the reset state.

On exit from the reset state, the program counter (PC) is reset, and the system clock defaults to the internal oscillator. The Watchdog Timer is enabled with the system clock divided by 12 as its clock source. Program execution begins at location 0x0000.

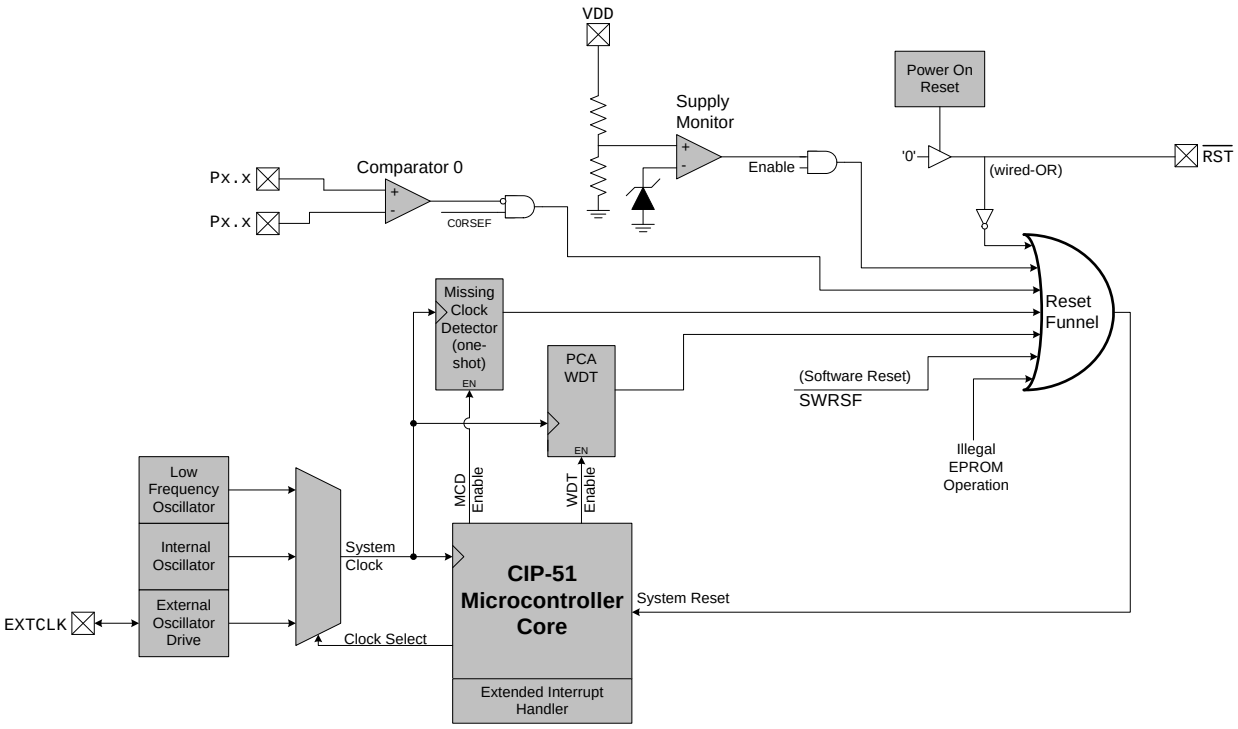

Figure 19.1. Reset Sources

SFR Definition 20.1. CLKSEL: Clock Select

Bit	7	6	5	4	3	2	1	0
Name								CLKSL0
Type	R	R	R	R	R	R	R	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xA9

Bit	Name	Function
7:1	Unused	Unused. Read = 0000000b; Write = Don't Care
0	CLKSL0	**System Clock Source Select Bit.** 0: SYSCLK derived from the Internal High-Frequency Oscillator and scaled per the IFCN bits in register OSCICN. 1: SYSCLK derived from the External Oscillator circuit.

20.3.1. External RC Example

If an RC network is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 20.1, “RC Mode”. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, first select the RC network value to produce the desired frequency of oscillation, according to Equation 20.1, where f = the frequency of oscillation in MHz, C = the capacitor value in pF, and R = the pull-up resistor value in kΩ.

Equation 20.1. RC Mode Oscillator Frequency

$$f = 1.23 \times 10^3 / (R \times C)$$

For example: If the frequency desired is 100 kHz, let R = 246 kΩ and C = 50 pF:

f = 1.23(10^3) / RC = 1.23 (10^3) / [246 x 50] = 0.1 MHz = 100 kHz

Referring to the table in SFR Definition 20.4, the required XFCN setting is 010b.

20.3.2. External Capacitor Example

If a capacitor is used as an external oscillator for the MCU, the circuit should be configured as shown in Figure 20.1, “C Mode”. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, select the capacitor to be used and find the frequency of oscillation according to Equation 20.2, where f = the frequency of oscillation in MHz, C = the capacitor value in pF, and V_{DD} = the MCU power supply in Volts.

Equation 20.2. C Mode Oscillator Frequency

$$f = (KF) / (C \times V_{DD})$$

For example: Assume V_{DD} = 3.0 V and f = 150 kHz:

f = KF / (C x VDD)
0.150 MHz = KF / (C x 3.0)

Since the frequency of roughly 150 kHz is desired, select the K Factor from the table in SFR Definition 20.4 (OSCXCN) as KF = 22:

0.150 MHz = 22 / (C x 3.0)
C x 3.0 = 22 / 0.150 MHz
C = 146.6 / 3.0 pF = 48.8 pF

Therefore, the XFCN value to use in this example is 011b and C = 50 pF.

21.2. Assigning Port I/O Pins to Analog and Digital Functions

Port I/O pins can be assigned to various analog, digital, and external interrupt functions. The Port pins assigned to analog functions should be configured for analog I/O, and Port pins assigned to digital or external interrupt functions should be configured for digital I/O.

21.2.1. Assigning Port I/O Pins to Analog Functions

Table 21.1 shows all available analog functions that require Port I/O assignments. **Port pins selected for these analog functions should have their corresponding bit in PnSKIP set to 1.** This reserves the pin for use by the analog function and does not allow it to be claimed by the Crossbar. Table 21.1 shows the potential mapping of Port I/O to each analog function.

Table 21.1. Port I/O Assignment for Analog Functions

Analog Function	Potentially Assignable Port Pins	SFR(s) used for Assignment
ADC Input	P1.0–P3.4	AMX0P, PnSKIP
Comparator Inputs	P1.0–P2.7	CPT0MX, CPT1MX, PnSKIP
Voltage Reference (VREF0)	P0.0	REF0CN, PnSKIP
External Oscillator in RC or C Mode (EXTCLK)	P0.3	OSCXCN, PnSKIP

21.2.2. Assigning Port I/O Pins to Digital Functions

Any Port pins not assigned to analog functions may be assigned to digital functions or used as GPIO. Most digital functions rely on the Crossbar for pin assignment; however, some digital functions bypass the Crossbar in a manner similar to the analog functions listed above. **Port pins used by these digital functions and any Port pins selected for use as GPIO should have their corresponding bit in PnSKIP set to 1.** Table 21.2 shows all available digital functions and the potential mapping of Port I/O to each digital function.

Table 21.2. Port I/O Assignment for Digital Functions

Digital Function	Potentially Assignable Port Pins	SFR(s) used for Assignment
UART0, SPI0, SMBus, CP0, CP0A, CP1, CP1A, SYSCLK, PCA0 (CEX0-4 and ECI), T0 or T1.	Any Port pin available for assignment by the Crossbar. This includes P0.0 - P2.3 pins which have their PnSKIP bit set to 0. **Note:** The Crossbar will always assign UART0 pins to P0.4 and P0.5.	XBR0, XBR1
Any pin used for GPIO	P0.0–P3.4	PnSKIP

SFR Definition 21.10. P1SKIP: Port 1 Skip

Bit	7	6	5	4	3	2	1	0
Name	P1SKIP[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xD5

Bit	Name	Function
7:0	P1SKIP[7:0]	**Port 1 Crossbar Skip Enable Bits.** These bits select Port 1 pins to be skipped by the Crossbar Decoder. Port pins used for analog, special functions or GPIO should be skipped by the Crossbar. 0: Corresponding P1.n pin is not skipped by the Crossbar. 1: Corresponding P1.n pin is skipped by the Crossbar.
Note: P1.6 and P1.7 are not connected to external pins on the C8051T616/7 devices. When writing code for the C8051T616/7, P1SKIP[6:7] should be set to 11b to skip these two pins on the crossbar.		

SFR Definition 21.11. P2: Port 2

Bit	7	6	5	4	3	2	1	0
Name	P2[7:0]							
Type	R/W							
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xA0; Bit-Addressable

Bit	Name	Description	Write	Read
7:0	P2[7:0]	**Port 2 Data.** Sets the Port latch logic value or reads the Port pin logic state in Port cells configured for digital I/O.	0: Set output latch to logic LOW. 1: Set output latch to logic HIGH.	0: P2.n Port pin is logic LOW. 1: P2.n Port pin is logic HIGH.
Note: P2.6 and P2.7 are not connected to external pins on the C8051T616/7 devices.				

22.5.4. Read Sequence (Slave)

During a read sequence, an SMBus master reads data from a slave device. The slave in this transfer will be a receiver during the address byte, and a transmitter during all data bytes. When slave events are enabled (INH = 0), the interface enters Slave Receiver Mode (to receive the slave address) when a START followed by a slave address and direction bit (READ in this case) is received. Upon entering Slave Receiver Mode, an interrupt is generated and the ACKRQ bit is set. The software must respond to the received slave address with an ACK, or ignore the received slave address with a NACK.

If the received slave address is ignored by software (by NACKing the address), slave interrupts will be inhibited until the next START is detected. If the received slave address is acknowledged, zero or more data bytes are transmitted. If the received slave address is acknowledged, data should be written to SMB0DAT to be transmitted. The interface enters slave transmitter mode, and transmits one or more bytes of data. After each byte is transmitted, the master sends an acknowledge bit; if the acknowledge bit is an ACK, SMB0DAT should be written with the next data byte. If the acknowledge bit is a NACK, SMB0DAT should not be written to before SI is cleared (an error condition may be generated if SMB0DAT is written following a received NACK while in slave transmitter mode). The interface exits slave transmitter mode after receiving a STOP. Note that the interface will switch to slave receiver mode if SMB0DAT is not written following a Slave Transmitter interrupt. Figure 22.8 shows a typical slave read sequence. Two transmitted data bytes are shown, though any number of bytes may be transmitted. Notice that all of the “data byte transferred” interrupts occur **after** the ACK cycle in this mode.

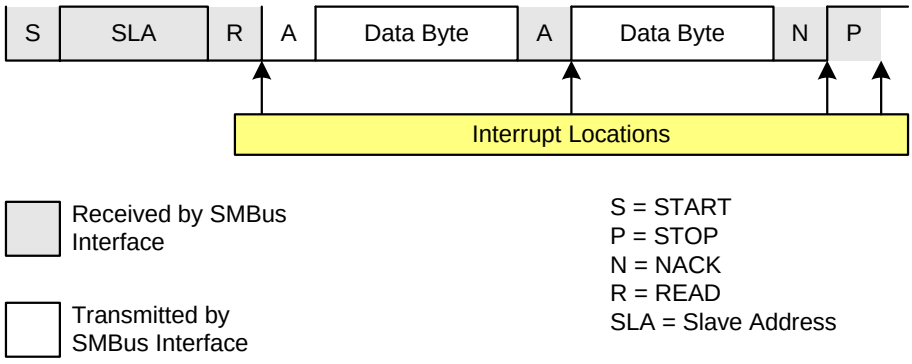

Figure 22.8. Typical Slave Read Sequence

22.6. SMBus Status Decoding

The current SMBus status can be easily decoded using the SMB0CN register. Table 22.4 describes the typical actions taken by firmware on each condition. In the table, STATUS VECTOR refers to the four upper bits of SMB0CN: MASTER, TXMODE, STA, and STO. The shown response options are only the typical responses; application-specific procedures are allowed as long as they conform to the SMBus specification. Highlighted responses are allowed by hardware but do not conform to the SMBus specification.

23.2.2. 9-Bit UART

9-bit UART mode uses a total of eleven bits per data byte: a start bit, 8 data bits (LSB first), a programmable ninth data bit, and a stop bit. The state of the ninth transmit data bit is determined by the value in TB80 (SCON0.3), which is assigned by user software. It can be assigned the value of the parity flag (bit P in register PSW) for error detection, or used in multiprocessor communications. On receive, the ninth data bit goes into RB80 (SCON0.2) and the stop bit is ignored.

Data transmission begins when an instruction writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to 1. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: (1) RI0 must be logic 0, and (2) if MCE0 is logic 1, the 9th bit must be logic 1 (when MCE0 is logic 0, the state of the ninth data bit is unimportant). If these conditions are met, the eight bits of data are stored in SBUF0, the ninth bit is stored in RB80, and the RI0 flag is set to 1. If the above conditions are not met, SBUF0 and RB80 will not be loaded and the RI0 flag will not be set to 1. A UART0 interrupt will occur if enabled when either TI0 or RI0 is set to 1.

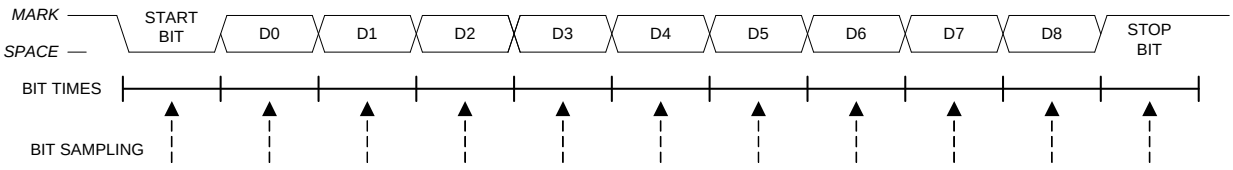

Figure 23.5. 9-Bit UART Timing Diagram

24.2. SPI0 Master Mode Operation

A SPI master device initiates all data transfers on a SPI bus. SPI0 is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CN.6). Writing a byte of data to the SPI0 data register (SPI0DAT) when in master mode writes to the transmit buffer. If the SPI shift register is empty, the byte in the transmit buffer is moved to the shift register, and a data transfer begins. The SPI0 master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic 1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. While the SPI0 master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers the contents of its shift register to the SPI master on the MISO line in a full-duplex operation. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data byte received from the slave is transferred MSB-first into the master's shift register. When a byte is fully shifted into the register, it is moved to the receive buffer where it can be read by the processor by reading SPI0DAT.

When configured as a master, SPI0 can operate in one of three different modes: multi-master mode, 3-wire single-master mode, and 4-wire single-master mode. The default, multi-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In this mode, NSS is an input to the device, and is used to disable the master SPI0 when another master is accessing the bus. When NSS is pulled low in this mode, MSTEN (SPI0CN.6) and SPIEN (SPI0CN.0) are set to 0 to disable the SPI master device, and a Mode Fault is generated (MODF, SPI0CN.5 = 1). Mode Fault will generate an interrupt if enabled. SPI0 must be manually re-enabled in software under these circumstances. In multi-master systems, devices will typically default to being slave devices while they are not acting as the system master device. In multi-master mode, slave devices can be addressed individually (if needed) using general-purpose I/O pins. Figure 24.2 shows a connection diagram between two master devices in multiple-master mode.

3-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. In this mode, NSS is not used, and is not mapped to an external port pin through the crossbar. Any slave devices that must be addressed in this mode should be selected using general-purpose I/O pins. Figure 24.3 shows a connection diagram between a master device in 3-wire master mode and a slave device.

4-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 1. In this mode, NSS is configured as an output pin, and can be used as a slave-select signal for a single SPI device. In this mode, the output value of NSS is controlled (in software) with the bit NSSMD0 (SPI0CN.2). Additional slave devices can be addressed using general-purpose I/O pins. Figure 24.4 shows a connection diagram for a master device in 4-wire master mode and two slave devices.

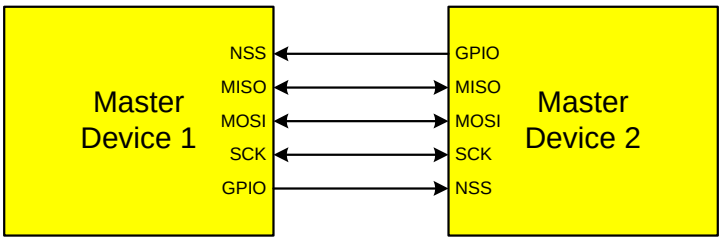

Figure 24.2. Multiple-Master Mode Connection Diagram

25.2.2. 8-bit Timers with Auto-Reload

When T2SPLIT is set, Timer 2 operates as two 8-bit timers (TMR2H and TMR2L). Both 8-bit timers operate in auto-reload mode as shown in Figure 25.5. TMR2RLL holds the reload value for TMR2L; TMR2RLH holds the reload value for TMR2H. The TR2 bit in TMR2CN handles the run control for TMR2H. TMR2L is always running when configured for 8-bit Mode.

Each 8-bit timer may be configured to use SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. The Timer 2 Clock Select bits (T2MH and T2ML in CKCON) select either SYSCLK or the clock defined by the Timer 2 External Clock Select bit (T2XCLK in TMR2CN), as follows:

T2MH	T2XCLK	TMR2H Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	X	SYSCLK

T2ML	T2XCLK	TMR2L Clock Source
0	0	SYSCLK / 12
0	1	External Clock / 8
1	X	SYSCLK

The TF2H bit is set when TMR2H overflows from 0xFF to 0x00; the TF2L bit is set when TMR2L overflows from 0xFF to 0x00. When Timer 2 interrupts are enabled, an interrupt is generated each time TMR2H overflows. If Timer 2 interrupts are enabled and TF2LEN (TMR2CN.5) is set, an interrupt is generated each time either TMR2L or TMR2H overflows. When TF2LEN is enabled, software must check the TF2H and TF2L flags to determine the source of the Timer 2 interrupt. The TF2H and TF2L interrupt flags are not cleared by hardware and must be manually cleared by software.

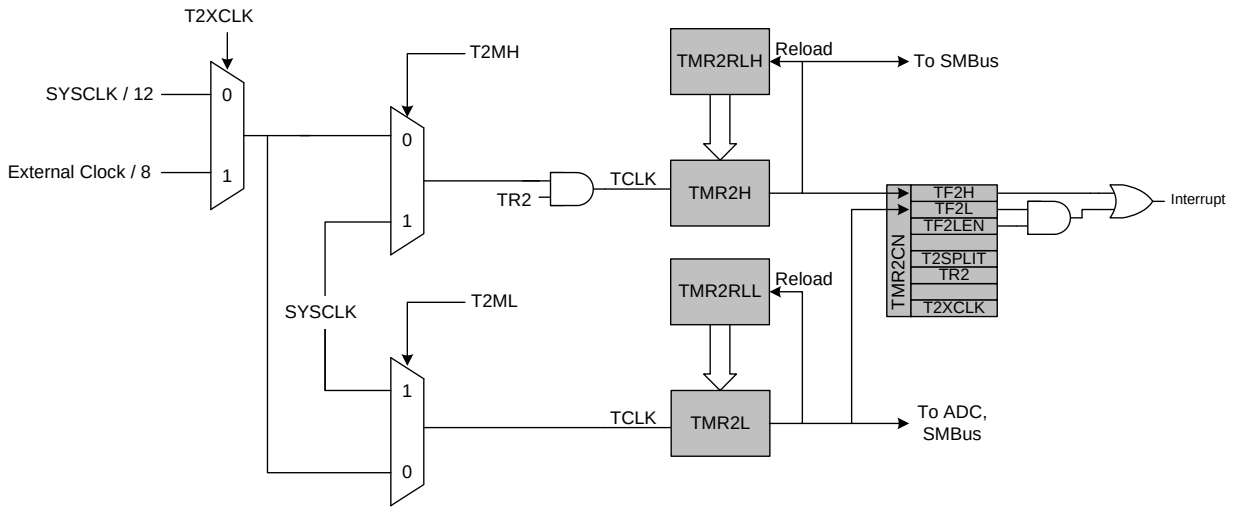

Figure 25.5. Timer 2 8-Bit Mode Block Diagram

26.3.3. High-Speed Output Mode

In High-Speed Output mode, a module's associated CEXn pin is toggled each time a match occurs between the PCA Counter and the module's 16-bit capture/compare register (PCA0CPHn and PCA0CPLn). When a match occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1. An interrupt request is generated if the CCFn interrupt for that module is enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Setting the TOGn, MATn, and ECOMn bits in the PCA0CPMn register enables the High-Speed Output mode. If ECOMn is cleared, the associated pin will retain its state, and not toggle on the next match event.

Important Note About Capture/Compare Registers: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to 0; writing to PCA0CPHn sets ECOMn to 1.

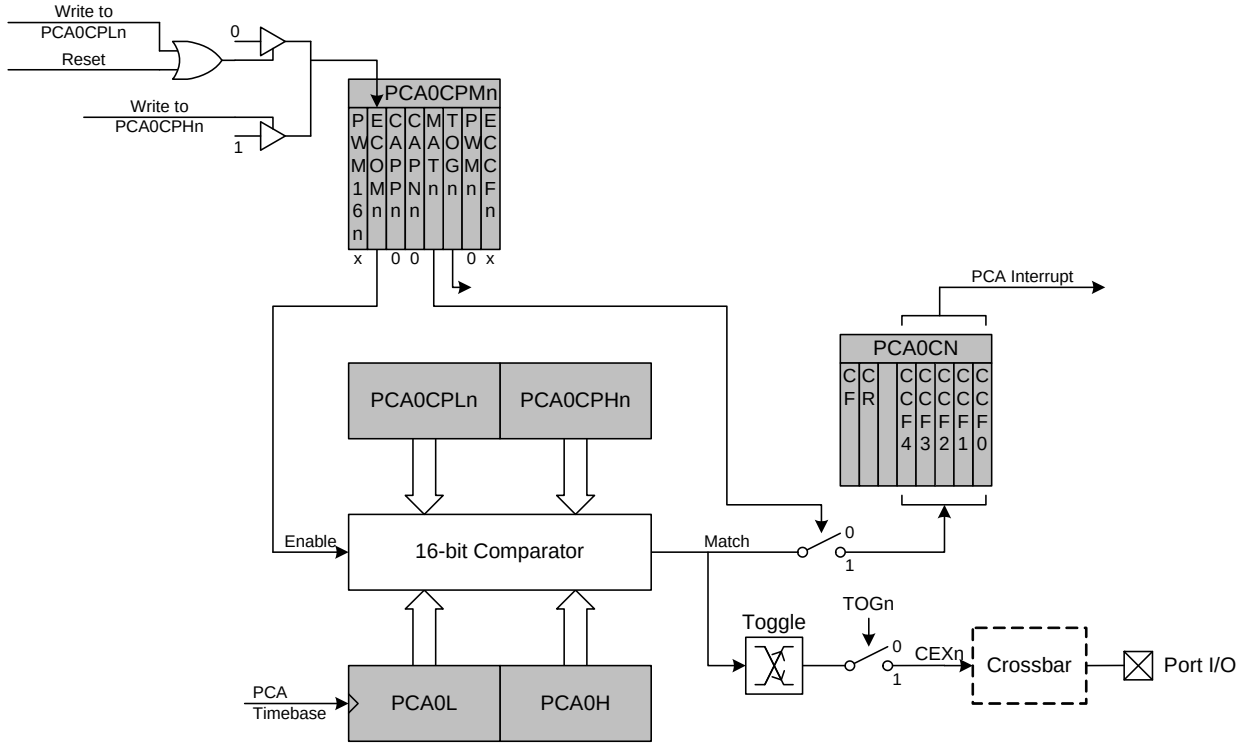

Figure 26.6. PCA High-Speed Output Mode Diagram

SFR Definition 26.4. PCA0L: PCA Counter/Timer Low Byte

Bit	7	6	5	4	3	2	1	0
Name	PCA0[7:0]							
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xF9

Bit	Name	Function
7:0	PCA0[7:0]	**PCA Counter/Timer Low Byte.** The PCA0L register holds the low byte (LSB) of the 16-bit PCA Counter/Timer.

Note: When the WDTE bit is set to 1, the PCA0L register cannot be modified by software. To change the contents of the PCA0L register, the Watchdog Timer must first be disabled.

SFR Definition 26.5. PCA0H: PCA Counter/Timer High Byte

Bit	7	6	5	4	3	2	1	0
Name	PCA0[15:8]							
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xFA

Bit	Name	Function
7:0	PCA0[15:8]	**PCA Counter/Timer High Byte.** The PCA0H register holds the high byte (MSB) of the 16-bit PCA Counter/Timer. Reads of this register will read the contents of a “snapshot” register, whose contents are updated only when the contents of PCA0L are read (see Section 26.1).

Note: When the WDTE bit is set to 1, the PCA0H register cannot be modified by software. To change the contents of the PCA0H register, the Watchdog Timer must first be disabled.