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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	EBI/EMI, SMBus (2-Wire/I ² C), CANbus, LINbus, SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	40
Program Memory Size	96KB (96K × 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	8.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.25V
Data Converters	A/D 32x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f584-imr

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



Figure 3.4. QFP-32 Pinout Diagram (Top View)





Figure 4.6. QFN-40 Landing Diagram

Table 4.6. QFN-40 Landing Diagram Dimensions

Dimension	Min	Мах		Dimension	Min	Max
C1	5.80	5.90		X2	4.10	4.20
C2	5.80	5.90		Y1	0.75	0.85
e	0.50	0.50 BSC		Y2	4.10	4.20
X1	0.15	0.25				

Notes:

General

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. Dimension and Tolerancing is per the ANSI Y14.5M-1994 specification.
- 3. This Land Pattern Design is based on the IPC-SM-7351 guidelines.
- 4. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05 mm.

Solder Mask Design

5. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be $60 \ \mu m$ minimum, all the way around the pad.

Stencil Design

- 6. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
- 7. The stencil thickness should be 0.125 mm (5 mils).
- 8. The ratio of stencil aperture to land pad size should be 1:1 for all perimeter pads.
- 9. A 4x4 array of 0.80 mm square openings on a 1.05 mm pitch should be used for the center ground pad.

Card Assembly

- **10.** A No-Clean, Type-3 solder paste is recommended.
- **11.** Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.



been accumulated. Similarly, the Window Comparator will not compare the result to the greater-than and less-than registers until "repeat count" conversions have been accumulated.

Note: When using Burst Mode, care must be taken to issue a convert start signal no faster than once every four SYSCLK periods. This includes external convert start signals.

System Clock																		
Convert Start (AD0BUSY or Timer Overflow)													ſL					
Post-Tracking AD0TM = 01 AD0EN = 0	Powered Down	Powe and	er-Up Idle	Т	С	Т	С	т	С	Т	С	Powered Down	F	owe	ər-L Idle	Jp e	Т	C
Dual-Tracking AD0TM = 11 AD0EN = 0	Powered Down	Powe and	er-Up Track PW/R /	Т	С	т	С	Т	С	т	С	Powered Down	F	owe	er-L Tra	Jp ck	Т	C
Post-Tracking AD0TM = 01 AD0EN = 1	Idle	тС	тс	Т	С	Т	С					Idle	т	С	т	С	т	C
Dual-Tracking AD0TM = 11 AD0EN = 1	Track	тС	ТС	Т	С	Т	С					Track	т	С	т	С	т	C
	T = Tracking C = Converti	ng																
Convert Start (CNVSTR)																		
Post-Tracking AD0TM = 01 AD0EN = 0	Powered Down	Powe and	er-Up Idle	Т	С					F	Pow Do	ered wn	F	owe and	ər-L Idle	Jp e	т	C
Dual-Tracking AD0TM = 11 AD0EN = 0	Powered Down	Powe and	er-Up Track	т	С					F	Pow Do	ered wn	F	owe	er-L Tra	Jp ck	т	C
Post-Tracking AD0TM = 01 AD0EN = 1	Idle	<pre>◆AD0</pre> T <pre>C</pre>	PWR≯	\					Id	le			т	С		Id	le	
Dual-Tracking AD0TM = 11 AD0EN = 1	Track	тС							Tra	ack			Т	С		Tra	ick.	

T = Tracking

C = Converting

Figure 6.4. 12-Bit ADC Burst Mode Example With Repeat Count Set to 4



6.2. Output Code Formatting

The registers ADC0H and ADC0L contain the high and low bytes of the output conversion code. When the repeat count is set to 1, conversion codes are represented in 12-bit unsigned integer format and the output conversion code is updated after each conversion. Inputs are measured from 0 to $V_{REF} \times 4095/4096$. Data can be right-justified or left-justified, depending on the setting of the AD0LJST bit (ADC0CN.2). Unused bits in the ADC0H and ADC0L registers are set to 0. Example codes are shown below for both right-justified and left-justified data.

Input Voltage	Right-Justified ADC0H:ADC0L (AD0LJST = 0)	Left-Justified ADC0H:ADC0L (AD0LJST = 1)
VREF x 4095/4096	0x0FFF	0xFFF0
VREF x 2048/4096	0x0800	0x8000
VREF x 2047/4096	0x07FF	0x7FF0
0	0x0000	0x0000

When the ADC0 Repeat Count is greater than 1, the output conversion code represents the accumulated result of the conversions performed and is updated after the last conversion in the series is finished. Sets of 4, 8, or 16 consecutive samples can be accumulated and represented in unsigned integer format. The repeat count can be selected using the AD0RPT bits in the ADC0CF register. The value must be right-justified (AD0LJST = 0), and unused bits in the ADC0H and ADC0L registers are set to 0. The following example shows right-justified codes for repeat counts greater than 1. Notice that accumulating 2^n samples is equivalent to left-shifting by *n* bit positions when all samples returned from the ADC have the same value.

Input Voltage	Repeat Count = 4	Repeat Count = 8	Repeat Count = 16
V _{REF} x 4095/4096	0x3FFC	0x7FF8	0xFFF0
V _{REF} x 2048/4096	0x2000	0x4000	0x8000
V _{REF} x 2047/4096	0x1FFC	0x3FF8	0x7FF0
0	0x0000	0x0000	0x0000

6.2.1. Settling Time Requirements

A minimum tracking time is required before an accurate conversion is performed. This tracking time is determined by any series impedance, including the AMUX0 resistance, the ADC0 sampling capacitance, and the accuracy required for the conversion.

Figure 6.5 shows the equivalent ADC0 input circuit. The required ADC0 settling time for a given settling accuracy (SA) may be approximated by Equation 6.1. When measuring the Temperature Sensor output, use the tracking time specified in Table 5.11 on page 52. When measuring V_{DD} with respect to GND, R_{TO-TAL} reduces to R_{MUX} . See Table 5.10 for ADC0 minimum settling time requirements as well as the mux impedance and sampling capacitor values.

$$t = ln\left(\frac{2^{n}}{SA}\right) \times R_{TOTAL}C_{SAMPLE}$$

Equation 6.1. ADC0 Settling Time Requirements

Where:

SA is the settling accuracy, given as a fraction of an LSB (for example, 0.25 to settle within 1/4 LSB) *t* is the required settling time in seconds

 R_{TOTAL} is the sum of the AMUX0 resistance and any external source resistance.

n is the ADC resolution in bits (10).





Figure 6.6. ADC Window Compare Example: Right-Justified Data



Figure 6.7. ADC Window Compare Example: Left-Justified Data



SFR Definition 9.3. CPT1CN: Comparator1 Control

Bit	7	6	5	4	3	2	1	0
Name	CP1EN	CP1OUT	CP1RIF	CP1FIF	CP1HYP[1:0]		CP1H	/N[1:0]
Туре	R/W	R	R/W	R/W	R/W		R/	W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x9D; SFR Page = 0x00

Bit	Name	Function
7	CP1EN	Comparator1 Enable Bit. 0: Comparator1 Disabled. 1: Comparator1 Enabled.
6	CP1OUT	Comparator1 Output State Flag. 0: Voltage on CP1+ < CP1 1: Voltage on CP1+ > CP1
5	CP1RIF	 Comparator1 Rising-Edge Flag. Must be cleared by software. 0: No Comparator1 Rising Edge has occurred since this flag was last cleared. 1: Comparator1 Rising Edge has occurred.
4	CP1FIF	 Comparator1 Falling-Edge Flag. Must be cleared by software. 0: No Comparator1 Falling-Edge has occurred since this flag was last cleared. 1: Comparator1 Falling-Edge has occurred.
3:2	CP1HYP[1:0]	Comparator1 Positive Hysteresis Control Bits. 00: Positive Hysteresis Disabled. 01: Positive Hysteresis = 5 mV. 10: Positive Hysteresis = 10 mV. 11: Positive Hysteresis = 20 mV.
1:0	CP1HYN[1:0]	Comparator1 Negative Hysteresis Control Bits. 00: Negative Hysteresis Disabled. 01: Negative Hysteresis = 5 mV. 10: Negative Hysteresis = 10 mV. 11: Negative Hysteresis = 20 mV.



11.2. Instruction Set

The instruction set of the CIP-51 System Controller is fully compatible with the standard MCS-51[™] instruction set. Standard 8051 development tools can be used to develop software for the CIP-51. All CIP-51 instructions are the binary and functional equivalent of their MCS-51[™] counterparts, including opcodes, addressing modes and effect on PSW flags. However, instruction timing is different than that of the standard 8051.

11.2.1. Instruction and CPU Timing

In many 8051 implementations, a distinction is made between machine cycles and clock cycles, with machine cycles varying from 2 to 12 clock cycles in length. However, the CIP-51 implementation is based solely on clock cycle timing. All instruction timings are specified in terms of clock cycles.

Due to the pipelined architecture of the CIP-51, most instructions execute in the same number of clock cycles as there are program bytes in the instruction. Conditional branch instructions take one less clock cycle to complete when the branch is not taken as opposed to when the branch is taken. Table 11.1 is the CIP-51 Instruction Set Summary, which includes the mnemonic, number of bytes, and number of clock cycles for each instruction.



Mnemonic	Description	Bytes	Clock Cycles			
XRL A, #data	Exclusive-OR immediate to A	2	2			
XRL direct, A	Exclusive-OR A to direct byte	2	2			
XRL direct, #data	Exclusive-OR immediate to direct byte	3	3			
CLR A	Clear A	1	1			
CPL A	Complement A	1	2			
RL A	Rotate A left	1	1			
RLC A	Rotate A left through Carry	1	1			
RR A	Rotate A right	1	1			
RRC A	Rotate A right through Carry	1	1			
SWAP A	Swap nibbles of A	1	1			
Data Transfer						
MOV A, Rn	Move Register to A	1	1			
MOV A, direct	Move direct byte to A	2	2			
MOV A, @Ri	Move indirect RAM to A	1	2			
MOV A, #data	Move immediate to A	2	2			
MOV Rn, A	Move A to Register	1	1			
MOV Rn, direct	Move direct byte to Register	2	2			
MOV Rn, #data	Move immediate to Register	2	2			
MOV direct, A	Move A to direct byte	2	2			
MOV direct, Rn	Move Register to direct byte	2	2			
MOV direct, direct	Move direct byte to direct byte	3	3			
MOV direct, @Ri	Move indirect RAM to direct byte	2	2			
MOV direct, #data	Move immediate to direct byte	3	3			
MOV @Ri, A	Move A to indirect RAM	1	2			
MOV @Ri, direct	Move direct byte to indirect RAM	2	2			
MOV @Ri, #data	Move immediate to indirect RAM	2	2			
MOV DPTR, #data16	Load DPTR with 16-bit constant	3	3			
MOVC A, @A+DPTR	Move code byte relative DPTR to A	1	4-7*			
MOVC A, @A+PC	Move code byte relative PC to A	1	3			
MOVX A, @Ri	Move external data (8-bit address) to A	1	3			
MOVX @Ri, A	Move A to external data (8-bit address)	1	3			
MOVX A, @DPTR	Move external data (16-bit address) to A	1	3			
MOVX @DPTR, A	Move A to external data (16-bit address)	1	3			
PUSH direct	Push direct byte onto stack	2	2			
POP direct	Pop direct byte from stack	2	2			
XCH A, Rn	Exchange Register with A	1	1			
XCH A, direct	Exchange direct byte with A	2	2			
XCH A, @Ri	Exchange indirect RAM with A	1	2			
XCHD A, @Ri	Exchange low nibble of indirect RAM with A	1	2			
Boolean Manipulation						
CLR C	Clear Carry	1	1			
CLR bit	Clear direct bit	2	2			
Note: Certain instructions take a variable number of clock cycles to execute depending on instruction alignment and the FLRT setting (SFR Definition 15.3).						

Table 11.1. CIP-51 Instruction Set Summary (Prefetch-Enabled) (Continued)





Figure 13.4. SFR Page Stack Upon PCA Interrupt Occurring During a CAN0 ISR

On exit from the PCA interrupt service routine, the CIP-51 will return to the CAN0 ISR. On execution of the RETI instruction, SFR Page 0x00 used to access the PCA registers will be automatically popped off of the SFR Page Stack, and the contents of the SFRNEXT register will be moved to the SFRPAGE register. Software in the CAN0 ISR can continue to access SFRs as it did prior to the PCA interrupt. Likewise, the contents of SFRLAST are moved to the SFRNEXT register. Recall this was the SFR Page value 0x00 being used to access SPI0DAT before the CAN0 interrupt occurred. See Figure 13.5.



SFR Definition 14.4. EIP1: Extended Interrupt Priority 1

Bit	7	6	5	4	3	2	1	0
Name	PLIN0	PT3	PCP1	PCP0	PPCA0	PADC0	PWADC0	PSMB0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xF6; SFR Page = All Pages

Bit	Name	Function
7	PLIN0	LINO Interrupt Priority Control. This bit sets the priority of the LINO interrupt. 0: LINO interrupts set to low priority level. 1: LINO interrupts set to high priority level.
6	PT3	Timer 3 Interrupt Priority Control.This bit sets the priority of the Timer 3 interrupt.0: Timer 3 interrupts set to low priority level.1: Timer 3 interrupts set to high priority level.
5	PCP1	Comparator0 (CP1) Interrupt Priority Control. This bit sets the priority of the CP1 interrupt. 0: CP1 interrupt set to low priority level. 1: CP1 interrupt set to high priority level.
4	PCP0	Comparator0 (CP0) Interrupt Priority Control. This bit sets the priority of the CP0 interrupt. 0: CP0 interrupt set to low priority level. 1: CP0 interrupt set to high priority level.
3	PPCA0	 Programmable Counter Array (PCA0) Interrupt Priority Control. This bit sets the priority of the PCA0 interrupt. 0: PCA0 interrupt set to low priority level. 1: PCA0 interrupt set to high priority level.
2	PADC0	 ADC0 Conversion Complete Interrupt Priority Control. This bit sets the priority of the ADC0 Conversion Complete interrupt. 0: ADC0 Conversion Complete interrupt set to low priority level. 1: ADC0 Conversion Complete interrupt set to high priority level.
1	PWADC0	 ADC0 Window Comparator Interrupt Priority Control. This bit sets the priority of the ADC0 Window interrupt. 0: ADC0 Window interrupt set to low priority level. 1: ADC0 Window interrupt set to high priority level.
0	PSMB0	 SMBus (SMB0) Interrupt Priority Control. This bit sets the priority of the SMB0 interrupt. 0: SMB0 interrupt set to low priority level. 1: SMB0 interrupt set to high priority level.



SFR Definition 15.1. PSCTL: Program Store R/W Control

Bit	7	6	5	4	3	2	1	0
Name							PSEE	PSWE
Туре	R	R	R	R	R	R	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x8F; SFR Page = 0x00

Bit	Name	Function
7:2	Unused	Read = 000000b, Write = don't care.
1	PSEE	Program Store Erase Enable.
		 Setting this bit (in combination with PSWE) allows an entire page of Flash program memory to be erased. If this bit is logic 1 and Flash writes are enabled (PSWE is logic 1), a write to Flash memory using the MOVX instruction will erase the entire page that contains the location addressed by the MOVX instruction. The value of the data byte written does not matter. 0: Flash program memory erasure disabled. 1: Flash program memory erasure enabled.
0	PSWE	Program Store Write Enable.
		 Setting this bit allows writing a byte of data to the Flash program memory using the MOVX write instruction. The Flash location should be erased before writing data. 0: Writes to Flash program memory disabled. 1: Writes to Flash program memory enabled; the MOVX write instruction targets Flash memory.



Rev. 1.3

18.4.2. Non-multiplexed Configuration

In Non-multiplexed mode, the Data Bus and the Address Bus pins are not shared. An example of a Nonmultiplexed Configuration is shown in Figure 18.2. See Section "18.6.1. Non-Multiplexed Mode" on page 169 for more information about Non-multiplexed operation.



Figure 18.2. Non-multiplexed Configuration Example



SFR Definition 20.20. P1SKIP: Port 1 Skip

Bit	7	6	5	4	3	2	1	0
Name		P1SKIP[7:0]						
Туре		R/W						
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xD5; SFR Page = 0x0F

Bit	Name	Function
7:0	P1SKIP[7:0]	Port 1 Crossbar Skip Enable Bits.
		 These bits select Port 1 pins to be skipped by the Crossbar Decoder. Port pins used for analog, special functions or GPIO should be skipped by the Crossbar. 0: Corresponding P1.n pin is not skipped by the Crossbar. 1: Corresponding P1.n pin is skipped by the Crossbar.

SFR Definition 20.21. P2: Port 2

Bit	7	6	5	4	3	2	1	0
Name		P2[7:0]						
Туре				R/	W			
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xA0; SFR Page = All Pages; Bit-Addressable

Bit	Name	Description	Write	Read
7:0	P2[7:0]	Port 2Data. Sets the Port latch logic value or reads the Port pin logic state in Port cells con- figured for digital I/O.	0: Set output latch to logic LOW. 1: Set output latch to logic HIGH.	0: P2.n Port pin is logic LOW. 1: P2.n Port pin is logic HIGH.



SFR Definition 20.26. P3MDIN: Port 3 Input Mode

Bit	7	6	5	4	3	2	1	0
Name	P3MDIN[7:0]							
Туре				R/	W			
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xF4; SFR Page = 0x0F

Bit	Name	Function
7:0	P3MDIN[7:0]	Analog Configuration Bits for P3.7–P3.0 (respectively).
		Port pins configured for analog mode have their weak pull-up and digital receiver disabled. For analog mode, the pin also needs to be configured for open-drain mode in the P3MDOUT register. 0: Corresponding P3.n pin is configured for analog mode. 1: Corresponding P3.n pin is not configured for analog mode.
Note:	Port P3.1–P3.7 a	re only available on the 48-pin and 40-pin packages.

SFR Definition 20.27. P3MDOUT: Port 3 Output Mode

Bit	7	6	5	4	3	2	1	0
Name		P3MDOUT[7:0]						
Туре				R/	W			
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xAE; SFR Page = 0x0F

Bit	Name	Function
7:0	P3MDOUT[7:0]	Output Configuration Bits for P3.7–P3.0 (respectively).
		These bits are ignored if the corresponding bit in register P3MDIN is logic 0. 0: Corresponding P3.n Output is open-drain. 1: Corresponding P3.n Output is push-pull.
Note:	Port P3.1-P3.7 a	re only available on the 48-pin and 40-pin packages.



LIN Register Definition 21.9. LIN0DIV: LIN0 Divider Register

Bit	7	6	5	4	3	2	1	0
Name				DIVLS	B[3:0]			
Туре				R/	W			
Reset								
Indirect Address = 0x0C								

Bit	Name	Function
7:0	DIVLSB	LIN Baud Rate Divider Least Significant Bits. The 8 least significant bits for the baud rate divider. The 9th and most significant bit is the DIV9 bit (LIN0MUL.0). The valid range for the divider is 200 to 511.

LIN Register Definition 21.10. LIN0MUL: LIN0 Multiplier Register

Bit	7	6	5	4	3	2	1	0
Name	PRESCL[1:0]			DIV9				
Туре	R/W				R/W			R/W
Reset	1	1	1	1	1	1	1	1

Indirect Address = 0x0D

Bit	Name	Function
7:6	PRESCL[1:0]	LIN Baud Rate Prescaler Bits.
		These bits are the baud rate prescaler bits.
5:1	LINMUL[4:0]	LIN Baud Rate Multiplier Bits.
		These bits are the baud rate multiplier bits. These bits are not used in slave mode.
0	DIV9	LIN Baud Rate Divider Most Significant Bit.
		The most significant bit of the baud rate divider. The 8 least significant bits are in LIN0DIV. The valid range for the divider is 200 to 511.



Data transmission begins when an instruction writes a data byte to the SBUF1 register. The TI1 Transmit Interrupt Flag (SCON1.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN1 Receive Enable bit (SCON1.4) is set to 1. After the stop bit is received, the data byte will be loaded into the SBUF1 receive register if the following conditions are met: (1) RI1 must be logic 0, and (2) if MCE1 is logic 1, the 9th bit must be logic 1 (when MCE1 is logic 0, the state of the ninth data bit is unimportant). If these conditions are met, the eight bits of data are stored in SBUF1, the ninth bit is stored in RB81, and the RI1 flag is set to 1. A UART1 interrupt will occur if enabled when either TI1 or RI1 is set to '1'.



25.3. Multiprocessor Communications

9-Bit UART mode supports multiprocessor communication between a master processor and one or more slave processors by special use of the ninth data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its ninth bit is logic 1; in a data byte, the ninth bit is always set to logic 0.

Setting the MCE1 bit (SCON1.5) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the ninth bit is logic 1 (RB81 = 1) signifying an address byte has been received. In the UART interrupt handler, software will compare the received address with the slave's own assigned 8-bit address. If the addresses match, the slave will clear its MCE1 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their MCE1 bits set and do not generate interrupts on the reception of the following data byte(s) addressed slave resets its MCE1 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).



Figure 25.6. UART Multi-Processor Mode Interconnect Diagram



26.3. SPI0 Slave Mode Operation

When SPI0 is enabled and not configured as a master, it will operate as a SPI slave. As a slave, bytes are shifted in through the MOSI pin and out through the MISO pin by a master device controlling the SCK signal. A bit counter in the SPI0 logic counts SCK edges. When 8 bits have been shifted through the shift register, the SPIF flag is set to logic 1, and the byte is copied into the receive buffer. Data is read from the receive buffer by reading SPI0DAT. A slave device cannot initiate transfers. Data to be transferred to the master device is pre-loaded into the shift register by writing to SPI0DAT. Writes to SPI0DAT are double-buffered, and are placed in the transmit buffer first. If the shift register is empty, the contents of the transmit buffer will immediately be transferred into the shift register. When the shift register already contains data, the SPI will load the shift register with the transmit buffer's contents after the last SCK edge of the next (or current) SPI transfer.

When configured as a slave, SPI0 can be configured for 4-wire or 3-wire operation. The default, 4-wire slave mode, is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In 4-wire mode, the NSS signal is routed to a port pin and configured as a digital input. SPI0 is enabled when NSS is logic 0, and disabled when NSS is logic 1. The bit counter is reset on a falling edge of NSS. Note that the NSS signal must be driven low at least 2 system clocks before the first active edge of SCK for each byte transfer. Figure 26.4 shows a connection diagram between two slave devices in 4-wire slave mode and a master device.

3-wire slave mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. NSS is not used in this mode, and is not mapped to an external port pin through the crossbar. Since there is no way of uniquely addressing the device in 3-wire slave mode, SPI0 must be the only slave device present on the bus. It is important to note that in 3-wire slave mode there is no external means of resetting the bit counter that determines when a full byte has been received. The bit counter can only be reset by disabling and re-enabling SPI0 with the SPIEN bit. Figure 26.3 shows a connection diagram between a slave device in 3-wire slave mode and a master device.

26.4. SPI0 Interrupt Sources

When SPI0 interrupts are enabled, the following four flags will generate an interrupt when they are set to logic 1:

All of the following bits must be cleared by software.

- 1. The SPI Interrupt Flag, SPIF (SPI0CN.7) is set to logic 1 at the end of each byte transfer. This flag can occur in all SPI0 modes.
- 2. The Write Collision Flag, WCOL (SPI0CN.6) is set to logic 1 if a write to SPI0DAT is attempted when the transmit buffer has not been emptied to the SPI shift register. When this occurs, the write to SPI0DAT will be ignored, and the transmit buffer will not be written. This flag can occur in all SPI0 modes.
- 3. The Mode Fault Flag MODF (SPI0CN.5) is set to logic 1 when SPI0 is configured as a master, and for multi-master mode and the NSS pin is pulled low. When a Mode Fault occurs, the MSTEN and SPIEN bits in SPI0CN are set to logic 0 to disable SPI0 and allow another master device to access the bus.
- 4. The Receive Overrun Flag RXOVRN (SPI0CN.4) is set to logic 1 when configured as a slave, and a transfer is completed and the receive buffer still holds an unread byte from a previous transfer. The new byte is not transferred to the receive buffer, allowing the previously received data byte to be read. The data byte which caused the overrun is lost.



SFR Definition 27.11. TMR2L: Timer 2 Low Byte

Bit	7	6	5	4	3	2	1	0
Name	TMR2L[7:0]							
Туре	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xCC; SFR Page = 0x00

Bit	Name	Function
7:0	TMR2L[7:0]	Timer 2 Low Byte.
		In 16-bit mode, the TMR2L register contains the low byte of the 16-bit Timer 2. In 8- bit mode, TMR2L contains the 8-bit low byte timer value.

SFR Definition 27.12. TMR2H Timer 2 High Byte

Bit	7	6	5	4	3	2	1	0
Name				TMR2	H[7:0]			
Туре	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xCD; SFR Page = 0x00

Bit	Name	Function
7:0	TMR2H[7:0]	Timer 2 High Byte.
		In 16-bit mode, the TMR2H register contains the high byte of the 16-bit Timer 2. In 8- bit mode, TMR2H contains the 8-bit high byte timer value.





Figure 28.7. PCA0 Frequency Output Mode

28.3.5. 8-bit, 9-bit, 10-bit and 11-bit Pulse Width Modulator Modes

Each module can be used independently to generate a pulse width modulated (PWM) output on its associated CEXn pin. The frequency of the output is dependent on the timebase for the PCA0 counter/timer, and the setting of the PWM cycle length (8, 9, 10 or 11-bits). For backwards-compatibility with the 8-bit PWM mode available on other devices, the 8-bit PWM mode operates slightly different than 9, 10 and 11-bit PWM modes. It is important to note that all channels configured for 8/9/10/11-bit PWM mode will use the same cycle length. It is not possible to configure one channel for 8-bit PWM mode and another for 11bit mode (for example). However, other PCA0 channels can be configured to Pin Capture, High-Speed Output, Software Timer, Frequency Output, or 16-bit PWM mode independently.

28.3.5.1. 8-bit Pulse Width Modulator Mode

The duty cycle of the PWM output signal in 8-bit PWM mode is varied using the module's PCA0CPLn capture/compare register. When the value in the low byte of the PCA0 counter/timer (PCA0L) is equal to the value in PCA0CPLn, the output on the CEXn pin will be set. When the count value in PCA0L overflows, the CEXn output will be reset (see Figure 28.8). Also, when the counter/timer low byte (PCA0L) overflows from 0xFF to 0x00, PCA0CPLn is reloaded automatically with the value stored in the module's capture/compare high byte (PCA0CPHn) without software intervention. Setting the ECOMn and PWMn bits in the PCA0CPMn register, and setting the CLSEL bits in register PCA0PWM to 00b enables 8-Bit Pulse Width Modulator mode. If the MATn bit is set to 1, the CCFn flag for the module will be set each time an 8-bit comparator match (rising edge) occurs. The COVF flag in PCA0PWM can be used to detect the overflow (falling edge), which will occur every 256 PCA0 clock cycles. The duty cycle for 8-Bit PWM Mode is given in Equation 28.2.

Important Note About Capture/Compare Registers: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to 0; writing to PCA0CPHn sets ECOMn to 1.

Duty Cycle = $\frac{(256 - PCA0CPHn)}{256}$

200

Equation 28.2. 8-Bit PWM Duty Cycle

Using Equation 28.2, the largest duty cycle is 100% (PCA0CPHn = 0), and the smallest duty cycle is 0.39% (PCA0CPHn = 0xFF). A 0% duty cycle may be generated by clearing the ECOMn bit to 0.



29.3.2. Software Timer (Compare) Mode

In Software Timer mode, the PCA1 counter/timer value is compared to the module's 16-bit capture/compare register (PCA1CPHn and PCA1CPLn). When a match occurs, the Capture/Compare Flag (CCFn) in PCA1CN is set to logic 1. An interrupt request is generated if the CCFn interrupt for that module is enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Setting the ECOM1n and MAT1n bits in the PCA1CPMn register enables Software Timer mode.

Important Note About Capture/Compare Registers: When writing a 16-bit value to the PCA1 Capture/Compare registers, the low byte should always be written first. Writing to PCA1CPLn clears the ECOM1n bit to 0; writing to PCA1CPHn sets ECOM1n to 1.



Figure 29.5. PCA1 Software Timer Mode Diagram

