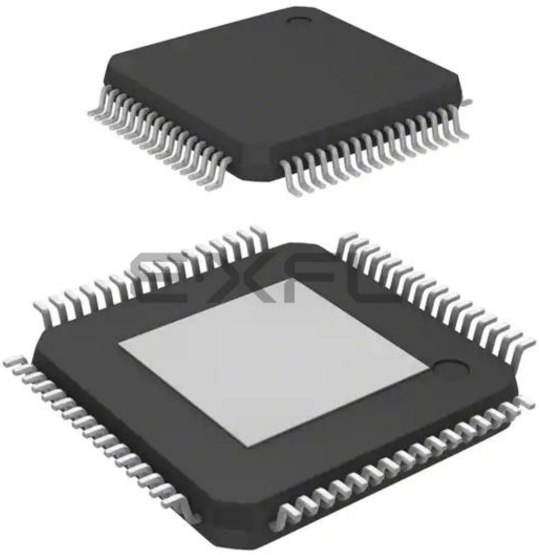

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	40MHz
Connectivity	CANbus, EBI/EMI, I²C, LINbus, SPI, SSC, UART/USART, USI
Peripherals	I²S, POR, PWM, WDT
Number of I/O	38
Program Memory Size	192KB (192K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	26K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 9x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-64-6
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/sak-xc2236n-24f40l

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16/32-Bit

Architecture

XC2232N, XC2234N, XC2236N, XC2238N

16/32-Bit Single-Chip Microcontroller
with 32-Bit Performance
XC2000 Family / Value Line

Data Sheet

V1.5 2013-02

Microcontrollers

- On-Chip Peripheral Modules
  - Two synchronizable A/D Converters with up to 9 channels, 10-bit resolution, conversion time below 1 µs, optional data preprocessing (data reduction, range check), broken wire detection
  - 16-channel general purpose capture/compare unit (CC2)
  - Two capture/compare units for flexible PWM signal generation (CCU6x)
  - Multi-functional general purpose timer unit with 5 timers
  - Up to 6 serial interface channels to be used as UART, LIN, high-speed synchronous channel (SPI/QSPI), IIC bus interface (10-bit addressing, 400 kbit/s), IIS interface
  - On-chip MultiCAN interface (Rev. 2.0B active) with up to 256 message objects (Full CAN/Basic CAN) on 6 CAN node
  - On-chip system timer and on-chip real time clock
- Single power supply from 3.0 V to 5.5 V
- Power reduction and wake-up modes with flexible power management
- Programmable watchdog timer and oscillator watchdog
- Up to 40 general purpose I/O lines
- On-chip bootstrap loaders
- Supported by a full range of development tools including C compilers, macro-assembler packages, emulators, evaluation boards, HLL debuggers, simulators, logic analyzer disassemblers, programming boards
- On-chip debug support via Device Access Port (DAP) or JTAG interface
- 64-pin Green LQFP package, 0.5 mm (19.7 mil) pitch

Ordering Information

The ordering code for an Infineon microcontroller provides an exact reference to a specific product. This ordering code identifies:

- the derivative itself, i.e. its function set, the temperature range, and the supply voltage
- the temperature range:
  - SAF-…: -40°C to 85°C
  - SAH-…: -40°C to 110°C
  - SAK-…: -40°C to 125°C
- the package and the type of delivery.

For ordering codes for the XC223xN please contact your sales representative or local distributor.

This document describes several derivatives of the XC223xN group:

Basic Device Types are readily available and
Special Device Types are only available on request.

As this document refers to all of these derivatives, some descriptions may not apply to a specific product, in particular to the special device types.

For simplicity the term **XC223xN** is used for all derivatives throughout this document.

Table 5 Interface Channel Association (cont'd)

Total Number	Available Channels / Message Objects
3 CAN nodes	CAN0, CAN1, CAN2 64 message objects
6 CAN nodes	CAN0, CAN1, CAN2, CAN3, CAN4, CAN5 256 message objects
2 serial channels	U0C0, U0C1
4 serial channels	U0C0, U0C1, U1C0, U1C1
6 serial channels	U0C0, U0C1, U1C0, U1C1, U2C0, U2C1

The XC223xN types are offered with several SRAM memory sizes. **Figure 1** shows the allocation rules for PSRAM and DSRAM. Note that the rules differ:

- PSRAM allocation starts from the **lower** address
- DSRAM allocation starts from the **higher** address

For example 8 Kbytes of PSRAM will be allocated at E0'0000h-E0'1FFFh and 8 Kbytes of DSRAM will be at 00'C000h-00'DFFFh.

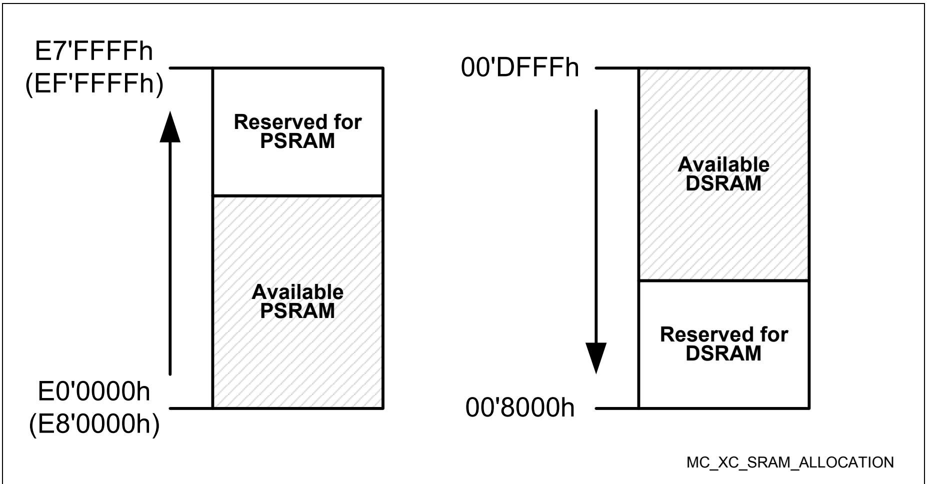

Figure 1 SRAM Allocation

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
31	P2.6	O0 / I	St/B	**Bit 6 of Port 2, General Purpose Input/Output**
	U0C0_SELO 0	O1	St/B	**USIC0 Channel 0 Select/Control 0 Output**
	U0C1_SELO 1	O2	St/B	**USIC0 Channel 1 Select/Control 1 Output**
	CC2_CC19	O3 / I	St/B	**CAPCOM2 CC19IO Capture Inp./ Compare Out.**
	U0C0_DX2D	I	St/B	**USIC0 Channel 0 Shift Control Input**
	RxDC0D	I	St/B	**CAN Node 0 Receive Data Input**
	ESR2_6	I	St/B	**ESR2 Trigger Input 6**
35	P2.7	O0 / I	St/B	**Bit 7 of Port 2, General Purpose Input/Output**
	U0C1_SELO 0	O1	St/B	**USIC0 Channel 1 Select/Control 0 Output**
	U0C0_SELO 1	O2	St/B	**USIC0 Channel 0 Select/Control 1 Output**
	CC2_CC20	O3 / I	St/B	**CAPCOM2 CC20IO Capture Inp./ Compare Out.**
	U0C1_DX2C	I	St/B	**USIC0 Channel 1 Shift Control Input**
	RxDC1C	I	St/B	**CAN Node 1 Receive Data Input**
	ESR2_7	I	St/B	**ESR2 Trigger Input 7**
36	P2.8	O0 / I	DP/B	**Bit 8 of Port 2, General Purpose Input/Output**
	U0C1_SCLK OUT	O1	DP/B	**USIC0 Channel 1 Shift Clock Output**
	EXTCLK	O2	DP/B	**Programmable Clock Signal Output**
	CC2_CC21	O3 / I	DP/B	**CAPCOM2 CC21IO Capture Inp./ Compare Out.**
	U0C1_DX1D	I	DP/B	**USIC0 Channel 1 Shift Clock Input**

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
56	P10.13	O0 / I	St/B	**Bit 13 of Port 10, General Purpose Input/Output**
	U1C0_DOUT	O1	St/B	**USIC1 Channel 0 Shift Data Output**
	TxDC3	O2	St/B	**CAN Node 3 Transmit Data Output**
	U1C0_SELO 3	O3	St/B	**USIC1 Channel 0 Select/Control 3 Output**
	U1C0_DX0D	I	St/B	**USIC1 Channel 0 Shift Data Input**
58	P10.14	O0 / I	St/B	**Bit 14 of Port 10, General Purpose Input/Output**
	U1C0_SELO 1	O1	St/B	**USIC1 Channel 0 Select/Control 1 Output**
	U0C1_DOUT	O2	St/B	**USIC0 Channel 1 Shift Data Output**
	ESR2_2	I	St/B	**ESR2 Trigger Input 2**
	U0C1_DX0C	I	St/B	**USIC0 Channel 1 Shift Data Input**
	RxDC3C	I	St/B	**CAN Node 3 Receive Data Input**
59	P10.15	O0 / I	St/B	**Bit 15 of Port 10, General Purpose Input/Output**
	U1C0_SELO 2	O1	St/B	**USIC1 Channel 0 Select/Control 2 Output**
	U0C1_DOUT	O2	St/B	**USIC0 Channel 1 Shift Data Output**
	U1C0_DOUT	O3	St/B	**USIC1 Channel 0 Shift Data Output**
	U0C1_DX1C	I	St/B	**USIC0 Channel 1 Shift Clock Input**
60	XTAL2	O	Sp/M	**Crystal Oscillator Amplifier Output**
61	XTAL1	I	Sp/M	**Crystal Oscillator Amplifier Input** To clock the device from an external source, drive XTAL1, while leaving XTAL2 unconnected. Voltages on XTAL1 must comply to the core supply voltage V_{DDIM}.
	ESR2_9	I	St/B	**ESR2 Trigger Input 9**
62	$\overline{\text{PORST}}$	I	In/B	**Power On Reset Input** A low level at this pin resets the XC223xN completely. A spike filter suppresses input pulses <10 ns. Input pulses >100 ns safely pass the filter. The minimum duration for a safe recognition should be 120 ns. An internal pull-up device will hold this pin high when nothing is driving it.

3 Functional Description

The architecture of the XC223xN combines advantages of RISC, CISC, and DSP processors with an advanced peripheral subsystem in a well-balanced design. On-chip memory blocks allow the design of compact systems-on-silicon with maximum performance suited for computing, control, and communication.

The on-chip memory blocks (program code memory and SRAM, dual-port RAM, data SRAM) and the generic peripherals are connected to the CPU by separate high-speed buses. Another bus, the LXBus, connects additional on-chip resources and external resources. This bus structure enhances overall system performance by enabling the concurrent operation of several subsystems of the XC223xN.

The block diagram gives an overview of the on-chip components and the advanced internal bus structure of the XC223xN.

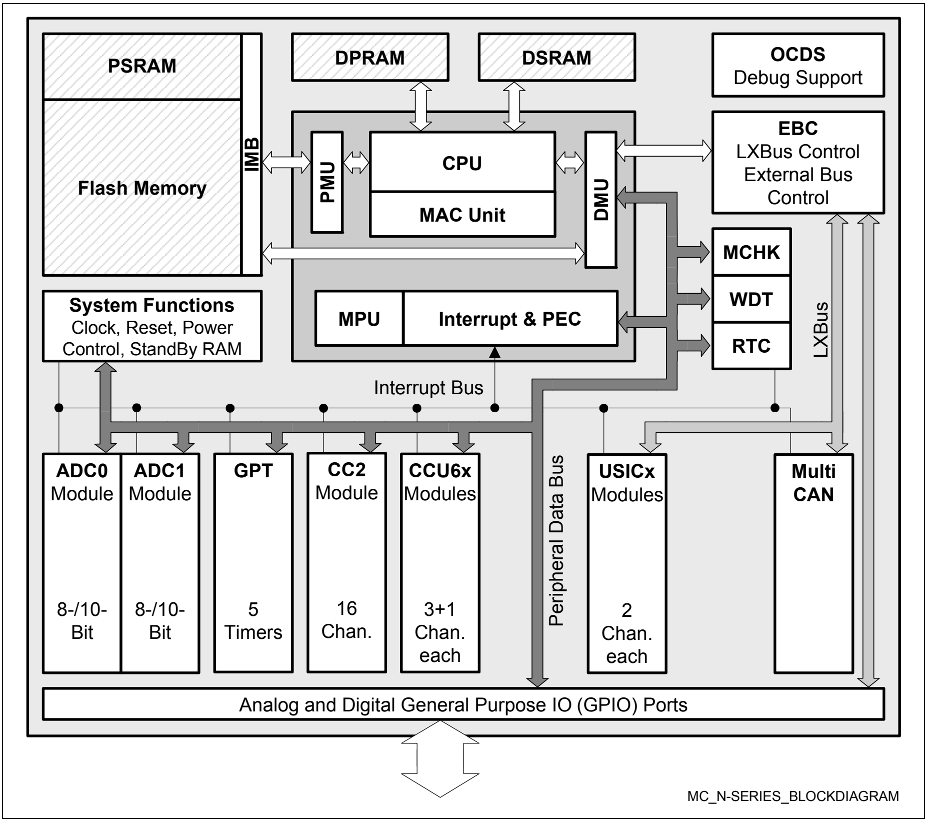

Figure 4 Block Diagram

Table 8 XC223xN Memory Map (cont’d)[1]

Address Area	Start Loc.	End Loc.	Area Size[2]	Notes
Reserved for DSRAM	00’8000$_H$	00’9FFF$_H$	8 Kbytes	
External memory area	00’0000$_H$	00’7FFF$_H$	32 Kbytes	

1) Accesses to the shaded areas are reserved. In devices with external bus interface these accesses generate external bus accesses.
2) The areas marked with “<” are slightly smaller than indicated, see column “Notes”.
3) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use (C0’F000$_H$ to C0’FFFF$_H$).
4) Several pipeline optimizations are not active within the external IO area. This is necessary to control external peripherals properly.

This common memory space consists of 16 Mbytes organized as 256 segments of 64 Kbytes; each segment contains four data pages of 16 Kbytes. The entire memory space can be accessed bytewise or wordwise. Portions of the on-chip DPRAM and the register spaces (ESFR/SFR) additionally are directly bit addressable.

The internal data memory areas and the Special Function Register areas (SFR and ESFR) are mapped into segment 0, the system segment.

The Program Management Unit (PMU) handles all code fetches and, therefore, controls access to the program memories such as Flash memory and PSRAM.

The Data Management Unit (DMU) handles all data transfers and, therefore, controls access to the DSRAM and the on-chip peripherals.

Both units (PMU and DMU) are connected to the high-speed system bus so that they can exchange data. This is required if operands are read from program memory, code or data is written to the PSRAM, code is fetched from external memory, or data is read from or written to external resources. These include peripherals on the LXBus such as USIC or MultiCAN. The system bus allows concurrent two-way communication for maximum transfer performance.

Up to 16 Kbytes of on-chip Program SRAM (PSRAM) are provided to store user code or data. The PSRAM is accessed via the PMU and is optimized for code fetches. A section of the PSRAM with programmable size can be write-protected.

Note: The actual size of the PSRAM depends on the quoted device type.

3.7 Capture/Compare Unit (CC2)

The CAPCOM unit supports generation and control of timing sequences on up to 16 channels with a maximum resolution of one system clock cycle (eight cycles in staggered mode). The CAPCOM unit is typically used to handle high-speed I/O tasks such as pulse and waveform generation, pulse width modulation (PWM), digital to analog (D/A) conversion, software timing, or time recording with respect to external events.

Two 16-bit timers with reload registers provide two independent time bases for the capture/compare register array.

The input clock for the timers is programmable to several prescaled values of the internal system clock, or may be derived from an overflow/underflow of timer T6 in module GPT2. This provides a wide range of variation for the timer period and resolution and allows precise adjustments to the application specific requirements. In addition, external count inputs allow event scheduling for the capture/compare registers relative to external events.

The capture/compare register array contains 16 dual purpose capture/compare registers, each of which may be individually allocated to either CAPCOM timer and programmed for capture or compare function.
All registers have each one port pin associated with it which serves as an input pin for triggering the capture function, or as an output pin to indicate the occurrence of a compare event.

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched (‘captured’) into the capture/compare register in response to an external event at the port pin which is associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers which have been selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the selected compare mode.

Table 9 Compare Modes

Compare Modes	Function
Mode 0	Interrupt-only compare mode; Several compare interrupts per timer period are possible
Mode 1	Pin toggles on each compare match; Several compare events per timer period are possible

With its maximum resolution of 2 system clock cycles, the **GPT2 module** provides precise event control and time measurement. It includes two timers (T5, T6) and a capture/reload register (CAPREL). Both timers can be clocked with an input clock which is derived from the CPU clock via a programmable prescaler or with external signals. The counting direction (up/down) for each timer can be programmed by software or altered dynamically with an external signal on a port pin (TxEUD). Concatenation of the timers is supported with the output toggle latch (T6OTL) of timer T6, which changes its state on each timer overflow/underflow.

The state of this latch may be used to clock timer T5, and/or it may be output on pin T6OUT. The overflows/underflows of timer T6 can also be used to clock the CAPCOM2 timers and to initiate a reload from the CAPREL register.

The CAPREL register can capture the contents of timer T5 based on an external signal transition on the corresponding port pin (CAPIN); timer T5 may optionally be cleared after the capture procedure. This allows the XC223xN to measure absolute time differences or to perform pulse multiplication without software overhead.

The capture trigger (timer T5 to CAPREL) can also be generated upon transitions of GPT1 timer T3 inputs T3IN and/or T3EUD. This is especially advantageous when T3 operates in Incremental Interface Mode.

3.10 Real Time Clock

The Real Time Clock (RTC) module of the XC223xN can be clocked with a clock signal selected from internal sources or external sources (pins).

The RTC basically consists of a chain of divider blocks:

- Selectable 32:1 and 8:1 dividers (on - off)
- The reloadable 16-bit timer T14
- The 32-bit RTC timer block (accessible via registers RTCH and RTCL) consisting of:
  - a reloadable 10-bit timer
  - a reloadable 6-bit timer
  - a reloadable 6-bit timer
  - a reloadable 10-bit timer

All timers count up. Each timer can generate an interrupt request. All requests are combined to a common node request.

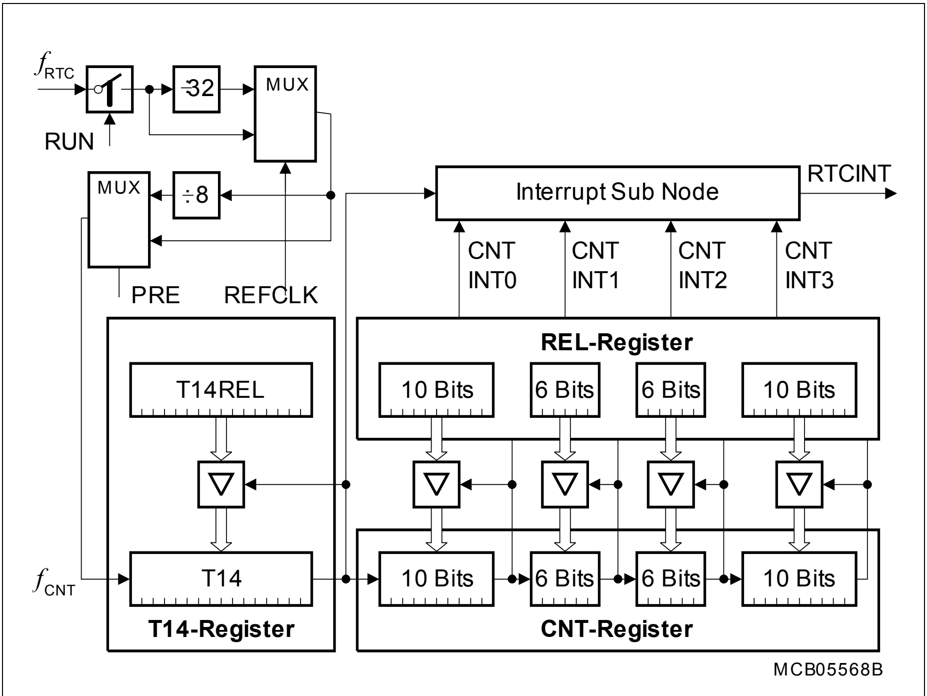

Figure 10 RTC Block Diagram

Note: The registers associated with the RTC are only affected by a power reset.

MultiCAN Features

- CAN functionality conforming to CAN specification V2.0 B active for each CAN node (compliant to ISO 11898)
- Independent CAN nodes
- Set of independent message objects (shared by the CAN nodes)
- Dedicated control registers for each CAN node
- Data transfer rate up to 1 Mbit/s, individually programmable for each node
- Flexible and powerful message transfer control and error handling capabilities
- Full-CAN functionality for message objects:
  - Can be assigned to one of the CAN nodes
  - Configurable as transmit or receive objects, or as message buffer FIFO
  - Handle 11-bit or 29-bit identifiers with programmable acceptance mask for filtering
  - Remote Monitoring Mode, and frame counter for monitoring
- Automatic Gateway Mode support
- 16 individually programmable interrupt nodes
- Analyzer mode for CAN bus monitoring

3.14 System Timer

The System Timer consists of a programmable prescaler and two concatenated timers (10 bits and 6 bits). Both timers can generate interrupt requests. The clock source can be selected and the timers can also run during power reduction modes.

Therefore, the System Timer enables the software to maintain the current time for scheduling functions or for the implementation of a clock.

3.15 Watchdog Timer

The Watchdog Timer is one of the fail-safe mechanisms which have been implemented to prevent the controller from malfunctioning for longer periods of time.

The Watchdog Timer is always enabled after an application reset of the chip. It can be disabled and enabled at any time by executing the instructions DISWDT and ENWDT respectively. The software has to service the Watchdog Timer before it overflows. If this is not the case because of a hardware or software failure, the Watchdog Timer overflows, generating a prewarning interrupt and then a reset request.

The Watchdog Timer is a 16-bit timer clocked with the system clock divided by 16,384 or 256. The Watchdog Timer register is set to a prespecified reload value (stored in WDTREL) in order to allow further variation of the monitored time interval. Each time it is serviced by the application software, the Watchdog Timer is reloaded and the prescaler is cleared.

Time intervals between 3.2 μs and 13.4 s can be monitored (@ 80 MHz).

The default Watchdog Timer interval after power-up is 6.5 ms (@ 10 MHz).

7) An overload current (I_{OV}) through a pin injects a certain error current (I_{INJ}) into the adjacent pins. This error current adds to the respective pins leakage current (I_{OZ}). The amount of error current depends on the overload current and is defined by the overload coupling factor K_{OV}. The polarity of the injected error current is inverse compared to the polarity of the overload current that produces it.The total current through a pin is $|I_{TOT}| = |I_{OZ}| + (|I_{OV}|\ \ K_{OV})$. The additional error current may distort the input voltage on analog inputs.

8) Value is controlled by on-chip regulator

4.2 Voltage Range definitions

The XC223xN timing depends on the supply voltage. If such a dependency exists the timing values are given for 2 voltage areas commonly used. The voltage areas are defined in the following tables.

Table 14 Upper Voltage Range Definition

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	4.5	5	5.5	V	

Table 15 Lower Voltage Range Definition

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	3.0	3.3	4.5	V	

4.2.1 Parameter Interpretation

The parameters listed in the following include both the characteristics of the XC223xN and its demands on the system. To aid in correctly interpreting the parameters when evaluating them for a design, they are marked accordingly in the column “Symbol”:

CC (**C**ontroller **C**haracteristics):
The logic of the XC223xN provides signals with the specified characteristics.

SR (**S**ystem **R**equirement):
The external system must provide signals with the specified characteristics to the XC223xN.

4.3 DC Parameters

These parameters are static or average values that may be exceeded during switching transitions (e.g. output current).

The XC223xN can operate within a wide supply voltage range from 3.0 V to 5.5 V. However, during operation this supply voltage must remain within 10 percent of the selected nominal supply voltage. It cannot vary across the full operating voltage range.

Because of the supply voltage restriction and because electrical behavior depends on the supply voltage, the parameters are specified separately for the upper and the lower voltage range.

During operation, the supply voltages may only change with a maximum speed of dV/dt < 1 V/ms.

Leakage current is strongly dependent on the operating temperature and the voltage level at the respective pin. The maximum values in the following tables apply under worst case conditions, i.e. maximum temperature and an input level equal to the supply voltage.
The value for the leakage current in an application can be determined by using the respective leakage derating formula (see tables) with values from that application.

The pads of the XC223xN are designed to operate in various driver modes. The DC parameter specifications refer to the pad current limits specified in **Section 4.7.4**.

During operation domain A draws a maximum current of 1.5 mA for each active A/D converter module from V_{DDPA}.

In Fast Startup Mode (with the Flash modules deactivated), the typical current is reduced to 3 + 0.6×f_{SYS} mA.

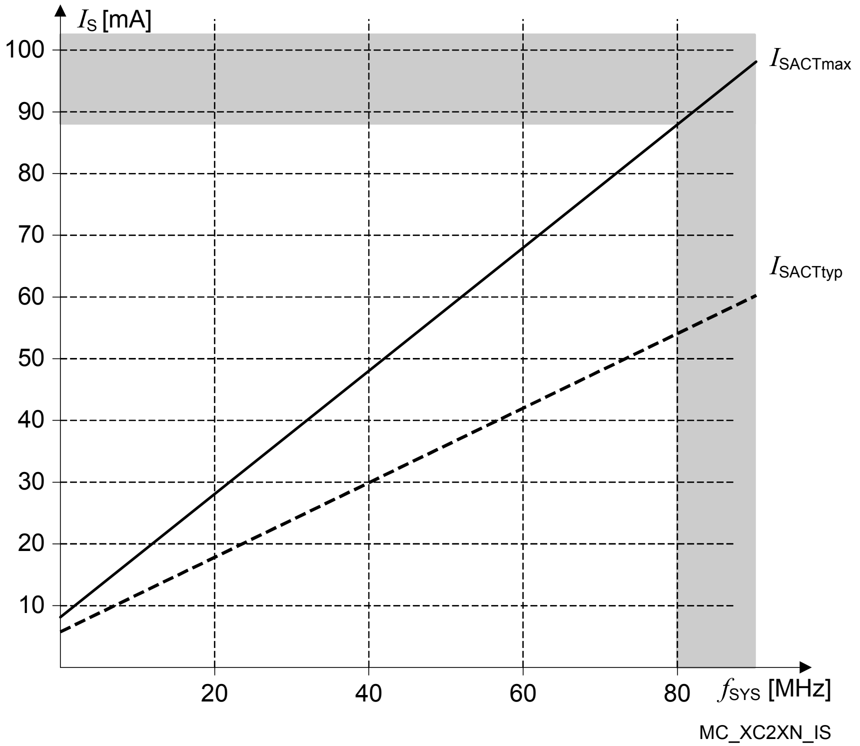

Figure 14 Supply Current in Active Mode as a Function of Frequency

Note: Operating Conditions apply.

The timing in the AC Characteristics refers to TCSs. Timing must be calculated using the minimum TCS possible under the given circumstances.
The actual minimum value for TCS depends on the jitter of the PLL. Because the PLL is constantly adjusting its output frequency to correspond to the input frequency (from crystal or oscillator), the accumulated jitter is limited. This means that the relative deviation for periods of more than one TCS is lower than for a single TCS (see formulas and **Figure 20**).

This is especially important for bus cycles using waitstates and for the operation of timers, serial interfaces, etc. For all slower operations and longer periods (e.g. pulse train generation or measurement, lower baudrates, etc.) the deviation caused by the PLL jitter is negligible.

The value of the accumulated PLL jitter depends on the number of consecutive VCO output cycles within the respective timeframe. The VCO output clock is divided by the output prescaler K2 to generate the system clock signal f_{SYS}. The number of VCO cycles is K2 $\times$ **T**, where **T** is the number of consecutive f_{SYS} cycles (TCS).

The maximum accumulated jitter (long-term jitter) D_{Tmax} is defined by:

$\mathbf{D_{Tmax}}$ **[ns] =** $\pm(220 / (K2 \times f_{SYS}) + 4.3)$

This maximum value is applicable, if either the number of clock cycles T > (f_{SYS} / 1.2) or the prescaler value K2 > 17.

In all other cases for a timeframe of **T** $\times$ TCS the accumulated jitter D_T is determined by:

$\mathbf{D_T}$ **[ns] =** $D_{Tmax} \times [(1 - 0.058 \times K2) \times (T - 1) / (0.83 \times f_{SYS} - 1) + 0.058 \times K2]$

f_{SYS} in [MHz] in all formulas.

Example, for a period of 3 TCSs @ 33 MHz and K2 = 4:

$D_{max} = \pm(220 / (4 \times 33) + 4.3) = 5.97$ ns (Not applicable directly in this case!)

$D_3 = 5.97 \times [(1 - 0.058 \times 4) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 4]$
$= 5.97 \times [0.768 \times 2 / 26.39 + 0.232]$
$= 1.7$ ns

Example, for a period of 3 TCSs @ 33 MHz and K2 = 2:

$D_{max} = \pm(220 / (2 \times 33) + 4.3) = 7.63$ ns (Not applicable directly in this case!)

$D_3 = 7.63 \times [(1 - 0.058 \times 2) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 2]$
$= 7.63 \times [0.884 \times 2 / 26.39 + 0.116]$
$= 1.4$ ns

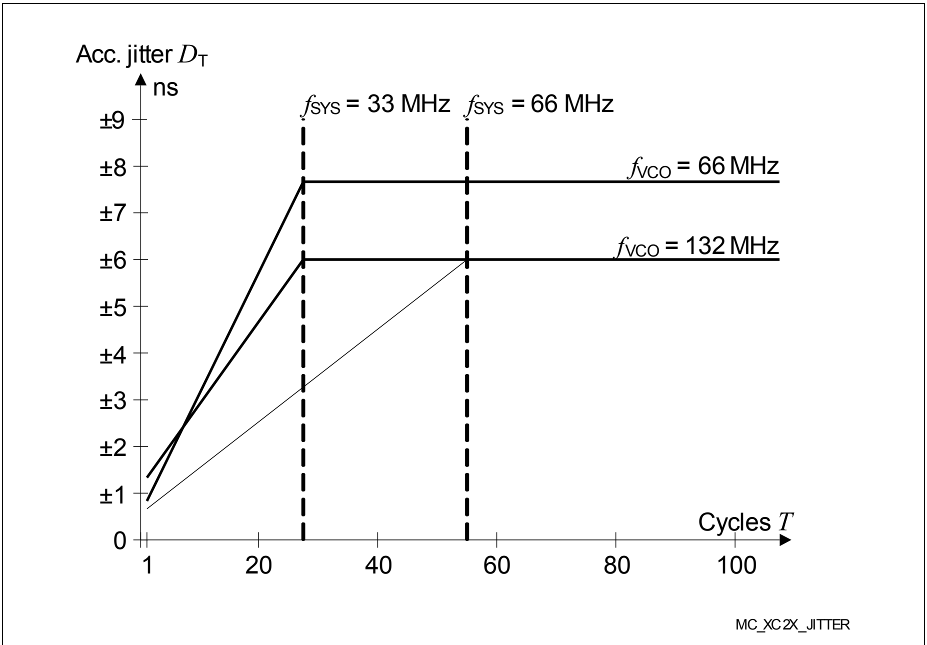

Figure 20 Approximated Accumulated PLL Jitter

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed C_L *= 20 pF.*
The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} *pin 64 and* V_{SS} *pin 1) is limited to a peak-to-peak voltage of* V_{PP} *= 50 mV. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.*

PLL frequency band selection

Different frequency bands can be selected for the VCO so that the operation of the PLL can be adjusted to a wide range of input and output frequencies:

1) The amplitude voltage V_{AX1} refers to the offset voltage V_{OFF}. This offset voltage must be stable during the operation and the resulting voltage peaks must remain within the limits defined by V_{IX1}.
2) Overload conditions must not occur on pin XTAL1.

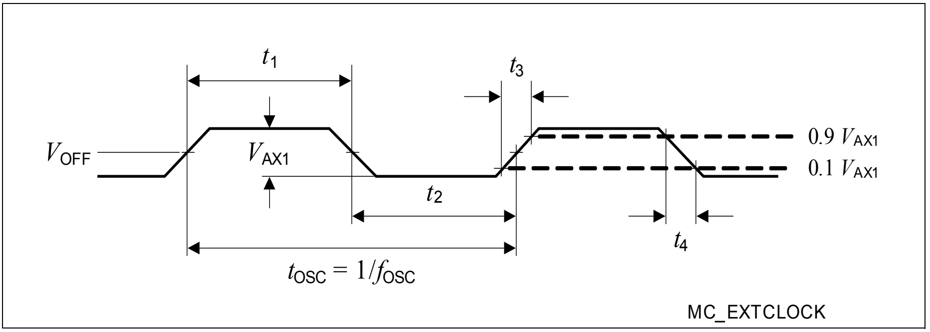

Figure 21 External Clock Drive XTAL1

Note: For crystal or ceramic resonator operation, it is strongly recommended to measure the oscillation allowance (negative resistance) in the final target system (layout) to determine the optimum parameters for oscillator operation.
The manufacturers of crystals and ceramic resonators offer an oscillator evaluation service. This evaluation checks the crystal/resonator specification limits to ensure a reliable oscillator operation.