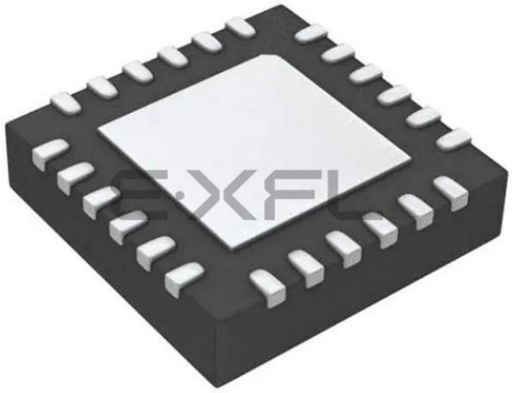

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"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	SMBus (2-Wire/I²C), CANbus, LINbus, SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	18
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.25V
Data Converters	A/D 18x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	24-WFQFN Exposed Pad
Supplier Device Package	24-QFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f554-im

6.3.2. Setting the Gain Value

The three programmable gain registers are accessed indirectly using the ADC0H and ADC0L registers when the GAINEN bit (ADC0CF.0) bit is set. ADC0H acts as the address register, and ADC0L is the data register. The programmable gain registers can only be written to and cannot be read. See Gain Register Definition 6.1, Gain Register Definition 6.2, and Gain Register Definition 6.3 for more information.

The gain is programmed using the following steps:

1. Set the GAINEN bit (ADC0CF.0)
2. Load the ADC0H with the ADC0GNH, ADC0GNL, or ADC0GNA address.
3. Load ADC0L with the desired value for the selected gain register.
4. Reset the GAINEN bit (ADC0CF.0)

Notes:

1. An ADC conversion should not be performed while the GAINEN bit is set.
2. Even with gain enabled, the maximum input voltage must be less than V_{REGIN} and the maximum voltage of the signal after gain must be less than or equal to V_{REF}.

In code, changing the value to 0.44 gain from the previous example looks like:

```
// in 'C':
ADC0CF |= 0x01;          // GAINEN = 1
ADC0H = 0x04;            // Load the ADC0GNH address
ADC0L = 0x6C;            // Load the upper byte of 0x6CA to ADC0GNH
ADC0H = 0x07;            // Load the ADC0GNL address
ADC0L = 0xA0;            // Load the lower nibble of 0x6CA to ADC0GNL
ADC0H = 0x08;            // Load the ADC0GNA address
ADC0L = 0x01;            // Set the GAINADD bit
ADC0CF &= ~0x01;         // GAINEN = 0

; in assembly
ORL ADC0CF,#01H          ; GAINEN = 1
MOV ADC0H,#04H           ; Load the ADC0GNH address
MOV ADC0L,#06CH          ; Load the upper byte of 0x6CA to ADC0GNH
MOV ADC0H,#07H           ; Load the ADC0GNL address
MOV ADC0L,#0A0H          ; Load the lower nibble of 0x6CA to ADC0GNL
MOV ADC0H,#08H           ; Load the ADC0GNA address
MOV ADC0L,#01H           ; Set the GAINADD bit
ANL ADC0CF,#0FEH         ; GAINEN = 0
```

10. CIP-51 Microcontroller

The MCU system controller core is the CIP-51 microcontroller. The CIP-51 is fully compatible with the MCS-51™ instruction set; standard 803x/805x assemblers and compilers can be used to develop software. The MCU family has a superset of all the peripherals included with a standard 8051. The CIP-51 also includes on-chip debug hardware (see description in Section 27), and interfaces directly with the analog and digital subsystems providing a complete data acquisition or control-system solution in a single integrated circuit.

The CIP-51 Microcontroller core implements the standard 8051 organization and peripherals as well as additional custom peripherals and functions to extend its capability (see Figure 10.1 for a block diagram). The CIP-51 includes the following features:

- Fully Compatible with MCS-51 Instruction Set
- 50 MIPS Peak Throughput with 50 MHz Clock
- 0 to 50 MHz Clock Frequency
- Extended Interrupt Handler
- Reset Input
- Power Management Modes
- On-chip Debug Logic
- Program and Data Memory Security

10.1. Performance

The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute, and usually have a maximum system clock of 12 MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with no instructions taking more than eight system clock cycles.

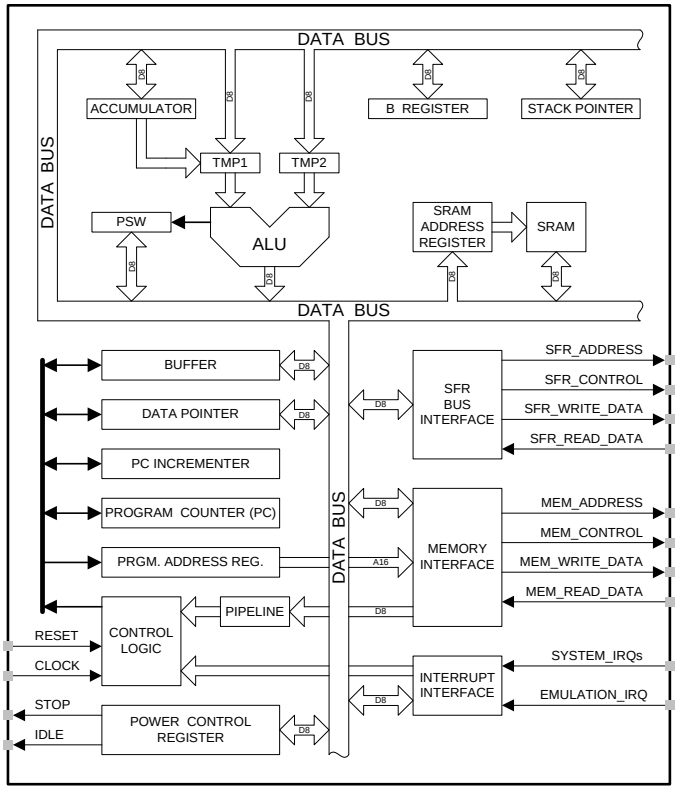

Figure 10.1. CIP-51 Block Diagram

With the CIP-51's maximum system clock at 50 MHz, it has a peak throughput of 50 MIPS. The CIP-51 has a total of 109 instructions. The table below shows the total number of instructions that require each execution time.

Clocks to Execute	1	2	2/3	3	3/4	4	4/5	5	8
Number of Instructions	26	50	5	14	7	3	1	2	1

Programming and Debugging Support

In-system programming of the Flash program memory and communication with on-chip debug support logic is accomplished via the Silicon Labs 2-Wire Development Interface (C2).

The on-chip debug support logic facilitates full speed in-circuit debugging, allowing the setting of hardware breakpoints, starting, stopping and single stepping through program execution (including interrupt service routines), examination of the program's call stack, and reading/writing the contents of registers and memory. This method of on-chip debugging is completely non-intrusive, requiring no RAM, Stack, timers, or other on-chip resources. C2 details can be found in Section “27. C2 Interface” on page 300.

The CIP-51 is supported by development tools from Silicon Labs and third party vendors. Silicon Labs provides an integrated development environment (IDE) including editor, debugger and programmer. The IDE's debugger and programmer interface to the CIP-51 via the C2 interface to provide fast and efficient in-system device programming and debugging. Third party macro assemblers and C compilers are also available.

Table 12.3. Special Function Registers

SFRs are listed in alphabetical order. All undefined SFR locations are reserved

Register	Address	Description	Page
ACC	0xE0	Accumulator	89
ADC0CF	0xBC	ADC0 Configuration	58
ADC0CN	0xE8	ADC0 Control	60
ADC0GTH	0xC4	ADC0 Greater-Than Compare High	62
ADC0GTL	0xC3	ADC0 Greater-Than Compare Low	62
ADC0H	0xBE	ADC0 High	59
ADC0L	0xBD	ADC0 Low	59
ADC0LTH	0xC6	ADC0 Less-Than Compare Word High	63
ADC0LTL	0xC5	ADC0 Less-Than Compare Word Low	63
ADC0MX	0xBB	ADC0 Mux Configuration	66
ADC0TK	0xBA	ADC0 Tracking Mode Select	61
B	0xF0	B Register	89
CCH0CN	0xE3	Cache Control	134
CKCON	0x8E	Clock Control	260
CLKMUL	0x97	Clock Multiplier	163
CLKSEL	0x8F	Clock Select	158
CPT0CN	0x9A	Comparator0 Control	72
CPT0MD	0x9B	Comparator0 Mode Selection	73
CPT0MX	0x9C	Comparator0 MUX Selection	77
CPT1CN	0x9D	Comparator1 Control	72
CPT1MD	0x9E	Comparator1 Mode Selection	73
CPT1MX	0x9F	Comparator1 MUX Selection	77
DPH	0x83	Data Pointer High	88
DPL	0x82	Data Pointer Low	88
EIE1	0xE6	Extended Interrupt Enable 1	118
EIE2	0xE7	Extended Interrupt Enable 2	118
EIP1	0xF6	Extended Interrupt Priority 1	119
EIP2	0xF7	Extended Interrupt Priority 2	120
EMI0CF	0xB2	External Memory Interface Configuration	148
EMI0CN	0xAA	External Memory Interface Control	147
EMI0TC	0xAA	External Memory Interface Timing Control	152
FLKEY	0xB7	Flash Lock and Key	132
FLSCL	0xB6	Flash Scale	133
IE	0xA8	Interrupt Enable	116
IP	0xB8	Interrupt Priority	117

14.4. Flash Write and Erase Guidelines

Any system which contains routines which write or erase Flash memory from software involves some risk that the write or erase routines will execute unintentionally if the CPU is operating outside its specified operating range of V_{DD}, system clock frequency, or temperature. This accidental execution of Flash modifying code can result in alteration of Flash memory contents causing a system failure that is only recoverable by re-Flashing the code in the device.

The following guidelines are recommended for any system which contains routines which write or erase Flash from code.

14.4.1. V_{DD} Maintenance and the V_{DD} monitor

1. If the system power supply is subject to voltage or current "spikes," add sufficient transient protection devices to the power supply to ensure that the supply voltages listed in the Absolute Maximum Ratings table are not exceeded.
2. Make certain that the minimum VREGIN rise time specification of 1 ms is met. If the system cannot meet this rise time specification, then add an external V_{DD} brownout circuit to the RST pin of the device that holds the device in reset until V_{DD} reaches the minimum threshold and re-asserts RST if V_{DD} drops below the minimum threshold.
3. Enable the on-chip V_{DD} monitor in the high setting and enable the V_{DD} monitor as a reset source as early in code as possible. This should be the first set of instructions executed after the Reset Vector. For C-based systems, this will involve modifying the startup code added by the C compiler. See your compiler documentation for more details. Make certain that there are no delays in software between enabling the V_{DD} monitor in the high setting and enabling the V_{DD} monitor as a reset source. Code examples showing this can be found in “AN201: Writing to Flash from Firmware", available from the Silicon Laboratories web site.
4. As an added precaution, explicitly enable the V_{DD} monitor in the high setting and enable the V_{DD} monitor as a reset source inside the functions that write and erase Flash memory. The V_{DD} monitor enable instructions should be placed just after the instruction to set PSWE to a 1, but before the Flash write or erase operation instruction.

Note: The output of the internal voltage regulator is calibrated by the MCU immediately after any reset event. The output of the un-calibrated internal regulator could be below the high threshold setting of the V_{DD} Monitor. If this is the case *and* the V_{DD} Monitor is set to the high threshold setting *and* if the MCU receives a non-power on reset (POR), the MCU will remain in reset until a POR occurs (i.e., V_{DD} Monitor will keep the device in reset). A POR will force the V_{DD} Monitor to the low threshold setting which is guaranteed to be below the un-calibrated output of the internal regulator. The device will then exit reset and resume normal operation. It is for this reason Silicon Labs strongly recommends that the V_{DD} Monitor is always left in the low threshold setting (i.e. default value upon POR). When programming the Flash in-system, the V_{DD} Monitor must be set to the high threshold setting. For the highest system reliability, the time the V_{DD} Monitor is set to the high threshold setting should be minimized (e.g., setting the V_{DD} Monitor to the high threshold setting just before the Flash write operation and then changing it back to the low threshold setting immediately after the Flash write operation).

5. Make certain that all writes to the RSTSRC (Reset Sources) register use direct assignment operators and explicitly DO NOT use the bit-wise operators (such as AND or OR). For example, "RSTSRC = 0x02" is correct. "RSTSRC |= 0x02" is incorrect.
6. Make certain that all writes to the RSTSRC register explicitly set the PORSF bit to a 1. Areas to check are initialization code which enables other reset sources, such as the Missing Clock Detector or Comparator, for example, and instructions which force a Software Reset. A global search on "RSTSRC" can quickly verify this.

SFR Definition 15.1. PCON: Power Control

Bit	7	6	5	4	3	2	1	0
Name	GF[5:0]						STOP	IDLE
Type	R/W						R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x87; SFR Page = All Pages

Bit	Name	Function
7:2	GF[5:0]	**General Purpose Flags 5–0.** These are general purpose flags for use under software control.
1	STOP	**Stop Mode Select.** Setting this bit will place the CIP-51 in Stop mode. This bit will always be read as 0. 1: CPU goes into Stop mode (internal oscillator stopped).
0	IDLE	**IDLE: Idle Mode Select.** Setting this bit will place the CIP-51 in Idle mode. This bit will always be read as 0. 1: CPU goes into Idle mode. (Shuts off clock to CPU, but clock to Timers, Interrupts, Serial Ports, and Analog Peripherals are still active.)

17.5.3. Split Mode with Bank Select

When EMI0CF[3:2] are set to 10, the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. The upper 8-bits of the Address Bus A[15:8] are determined by EMI0CN, and the lower 8-bits of the Address Bus A[7:0] are determined by R0 or R1. All 16-bits of the Address Bus A[15:0] are driven in “Bank Select” mode.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

17.5.4. External Only

When EMI0CF[3:2] are set to 11, all MOVX operations are directed to off-chip space. On-chip XRAM is not visible to the CPU. This mode is useful for accessing off-chip memory located between 0x0000 and the internal XRAM size boundary.

- 8-bit MOVX operations ignore the contents of EMI0CN. The upper Address bits A[15:8] are not driven (identical behavior to an off-chip access in “Split Mode without Bank Select” described above). This allows the user to manipulate the upper address bits at will by setting the Port state directly. The lower 8-bits of the effective address A[7:0] are determined by the contents of R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine the effective address A[15:0]. The full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

17.6. Timing

The timing parameters of the External Memory Interface can be configured to enable connection to devices having different setup and hold time requirements. The Address Setup time, Address Hold time, $\overline{RD}$ and $\overline{WR}$ strobe widths, and in multiplexed mode, the width of the ALE pulse are all programmable in units of SYSCLK periods through EMI0TC, shown in SFR Definition 17.3, and EMI0CF[1:0].

The timing for an off-chip MOVX instruction can be calculated by adding 4 SYSCLK cycles to the timing parameters defined by the EMI0TC register. Assuming non-multiplexed operation, the minimum execution time for an off-chip XRAM operation is 5 SYSCLK cycles (1 SYSCLK for $\overline{RD}$ or $\overline{WR}$ pulse + 4 SYSCLKs). For multiplexed operations, the Address Latch Enable signal will require a minimum of 2 additional SYSCLK cycles. Therefore, the minimum execution time for an off-chip XRAM operation in multiplexed mode is 7 SYSCLK cycles (2 for /ALE + 1 for $\overline{RD}$ or $\overline{WR}$ + 4). The programmable setup and hold times default to the maximum delay settings after a reset. Table 17.2 lists the ac parameters for the External Memory Interface, and Figure 17.3 through Figure 17.5 show the timing diagrams for the different External Memory Interface modes and MOVX operations.

17.6.1.2. 8-bit MOVX without Bank Select: EMI0CF[4:2] = 001 or 011

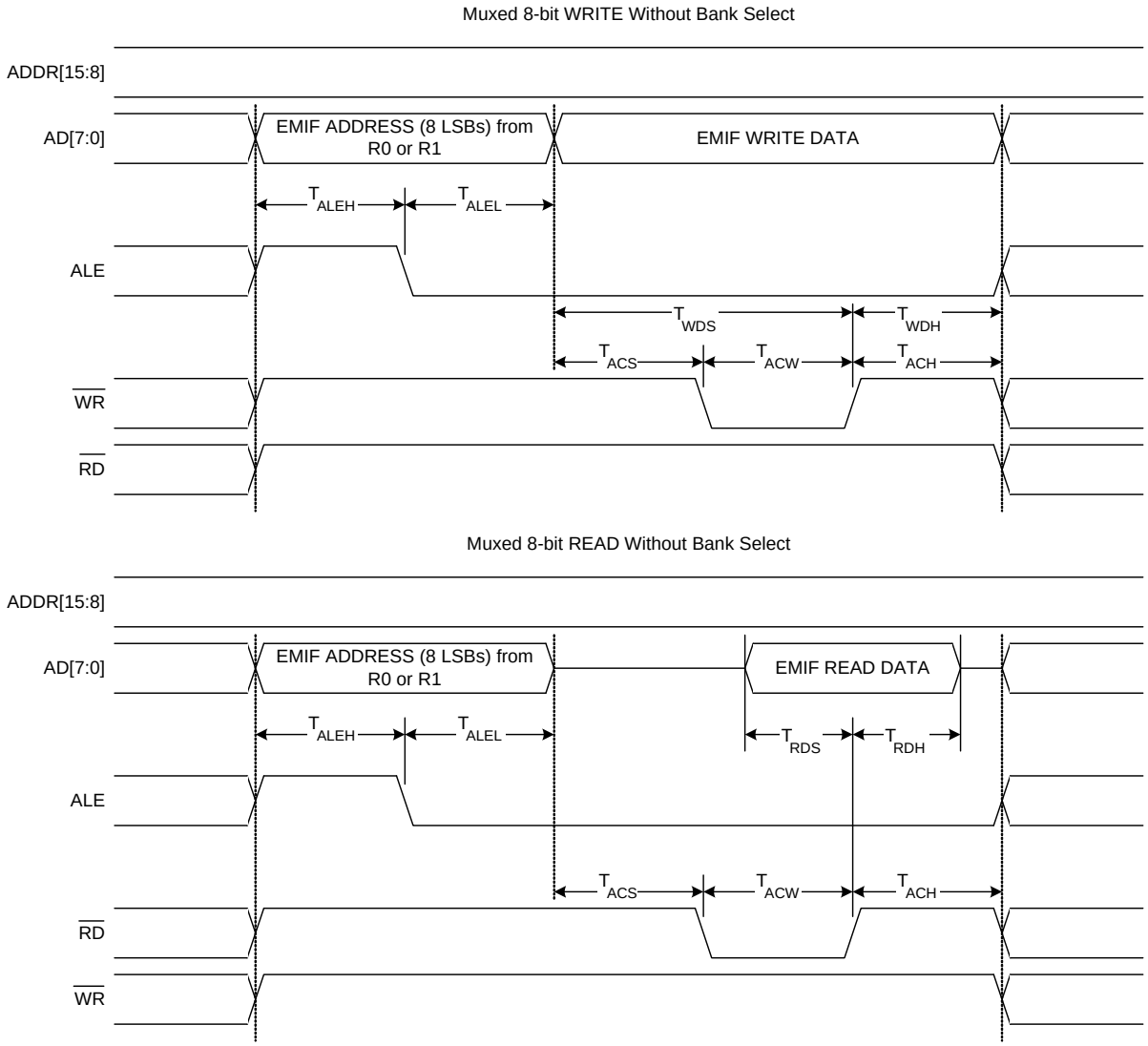

Figure 17.4. Multiplexed 8-bit MOVX without Bank Select Timing

19. Port Input/Output

Digital and analog resources are available through 33 (C8051F568-9 and ‘F570-5), 25 (C8051F550-7) or 18 (C8051F550-7) I/O pins. Port pins P0.0-P4.0 on the C8051F568-9 and ‘F570-5, port pins P0.0-P3.0 on theC8051F560-7, and port pins P0.0-P2.1 on the C8051F550-7 can be defined as general-purpose I/O (GPIO), assigned to one of the internal digital resources, or assigned to an analog function as shown in Figure 19.3. Port pin P4.0 on the C8051F568-9 and ‘F570-5 can be used as GPIO and is shared with the C2 Interface Data signal (C2D). Similarly, port pin P3.0 is shared with C2D on the C8051F560-7 and port pin P2.1 on the C8051F550-7. The designer has complete control over which functions are assigned, limited only by the number of physical I/O pins. This resource assignment flexibility is achieved through the use of a Priority Crossbar Decoder. Note that the state of a Port I/O pin can always be read in the corresponding Port latch, regardless of the Crossbar settings.

The Crossbar assigns the selected internal digital resources to the I/O pins based on the Priority Decoder (Figure 19.3 and Figure 19.4). The registers XBR0, XBR1, XBR2 are defined in SFR Definition 19.1 and SFR Definition 19.2 and are used to select internal digital functions.

The Port I/O cells are configured as either push-pull or open-drain in the Port Output Mode registers (PnMDOUT, where n = 0,1). Complete Electrical Specifications for Port I/O are given in Table 5.3 on page 40.

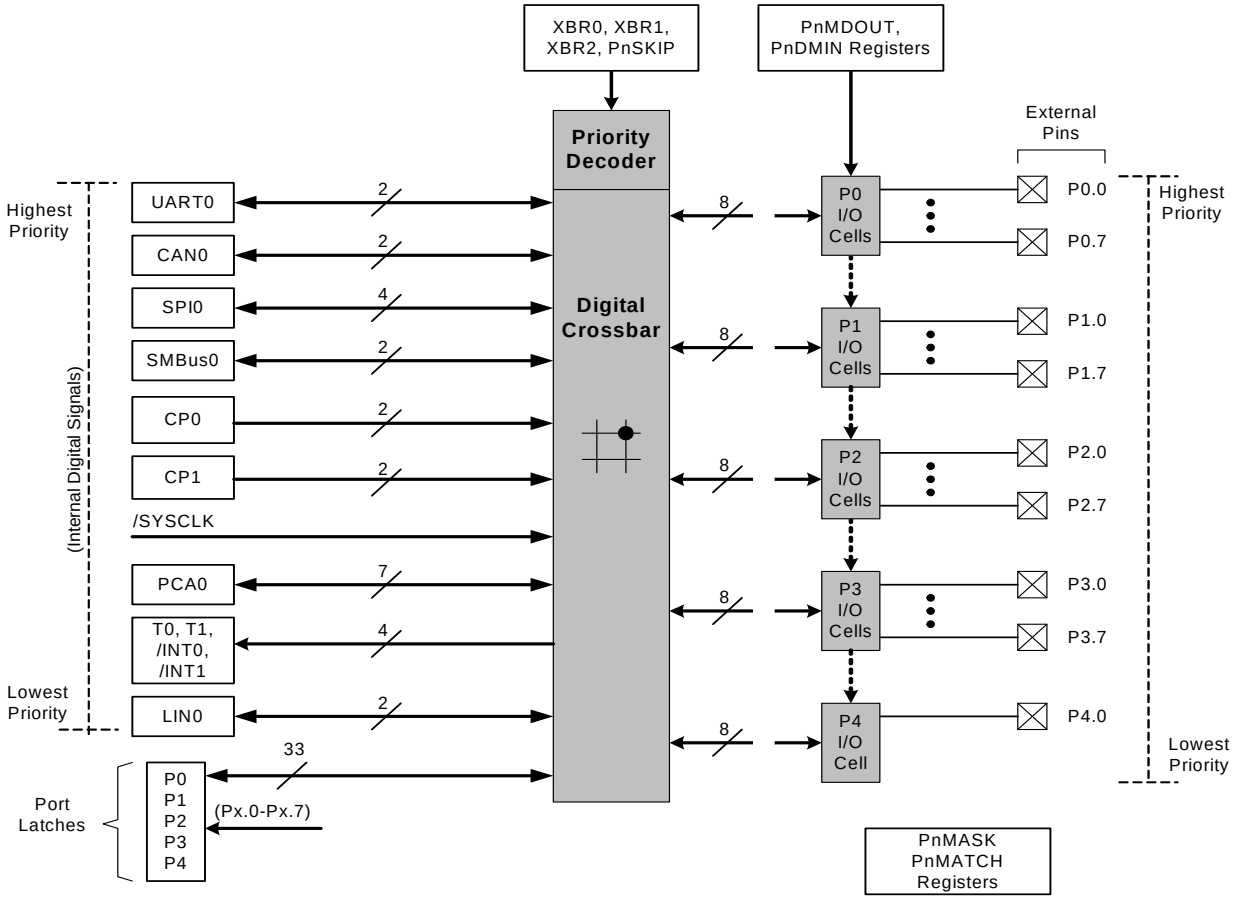

Figure 19.1. Port I/O Functional Block Diagram

SFR Definition 19.1. XBR0: Port I/O Crossbar Register 0

Bit	7	6	5	4	3	2	1	0
Name	CP1AE	CP1E	CP0AE	CP0E	SMB0E	SPI0E	CAN0E	URT0E
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xE1; SFR Page = 0x0F

Bit	Name	Function
7	CP1AE	**Comparator1 Asynchronous Output Enable.** 0: Asynchronous CP1 unavailable at Port pin. 1: Asynchronous CP1 routed to Port pin.
6	CP1E	**Comparator1 Output Enable.** 0: CP1 unavailable at Port pin. 1: CP1 routed to Port pin.
5	CP0AE	**Comparator0 Asynchronous Output Enable.** 0: Asynchronous CP0 unavailable at Port pin. 1: Asynchronous CP0 routed to Port pin.
4	CP0E	**Comparator0 Output Enable.** 0: CP0 unavailable at Port pin. 1: CP0 routed to Port pin.
3	SMB0E	**SMBus I/O Enable.** 0: SMBus I/O unavailable at Port pins. 1: SMBus I/O routed to Port pins.
2	SPI0E	**SPI I/O Enable.** 0: SPI I/O unavailable at Port pins. 1: SPI I/O routed to Port pins. Note that the SPI can be assigned either 3 or 4 GPIO pins.
1	CAN0E	**CAN I/O Output Enable.** 0: CAN I/O unavailable at Port pins. 1: CAN_TX, CAN_RX routed to Port pins P0.6 and P0.7.
0	URT0E	**UART I/O Output Enable.** 0: UART I/O unavailable at Port pin. 1: UART TX0, RX0 routed to Port pins P0.4 and P0.5.

The application should perform the following steps when an interrupt is requested.

1. Check the DONE bit (LIN0ST.0) and the ERROR bit (LIN0ST.2).
2. If performing a master receive operation and the transfer was successful, read the received data from the data buffer.
3. If the transfer was not successful, check the error register to determine the kind of error. Further error handling has to be done by the application.
4. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

20.4. LIN Slave Mode Operation

When the device is configured for slave mode operation, it must wait for a command from a master node. Access from the firmware to the data buffer and ID registers of the LIN controller is only possible when a data request is pending (DTREQ bit (LIN0ST.4) is 1) and also when the LIN bus is not active (ACTIVE bit (LIN0ST.7) is set to 0).

The LIN controller in slave mode detects the header of the message frame sent by the LIN master. If slave synchronization is enabled (autobaud), the slave synchronizes its internal bit time to the master bit time.

The LIN controller configured for slave mode will generated an interrupt in one of three situations:

1. After the reception of the IDENTIFIER FIELD
2. When an error is detected
3. When the message transfer is completed.

The application should perform the following steps when an interrupt is detected:

1. Check the status of the DTREQ bit (LIN0ST.4). This bit is set when the IDENTIFIER FIELD has been received.
2. If DTREQ (LIN0ST.4) is set, read the identifier from LIN0ID and process it. If DTREQ (LIN0ST.4) is not set, continue to step 7.
3. Set the TXRX bit (LIN0CTRL.5) to 1 if the current frame is a transmit operation for the slave and set to 0 if the current frame is a receive operation for the slave.
4. Load the data length into LIN0SIZE.
5. For a slave transmit operation, load the data to transmit into the data buffer.
6. Set the DTACK bit (LIN0CTRL.4). Continue to step 10.
7. If DTREQ (LIN0ST.4) is not set, check the DONE bit (LIN0ST.0). The transmission was successful if the DONE bit is set.
8. If the transmission was successful and the current frame was a receive operation for the slave, load the received data bytes from the data buffer.
9. If the transmission was not successful, check LIN0ERR to determine the nature of the error. Further error handling has to be done by the application.
10. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

In addition to these steps, the application should be aware of the following:

1. If the current frame is a transmit operation for the slave, steps 1 through 5 must be completed during the IN-FRAME RESPONSE SPACE. If it is not completed in time, a timeout will be detected by the master.
2. If the current frame is a receive operation for the slave, steps 1 through 5 have to be finished until the reception of the first byte after the IDENTIFIER FIELD. Otherwise, the internal receive buffer of the LIN controller will be overwritten and a timeout error will be detected in the LIN controller.

LIN Register Definition 20.4. LIN0DTn: LIN0 Data Byte n

Bit	7	6	5	4	3	2	1	0
Name	DATAn[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

Indirect Address: LIN0DT1 = 0x00, LIN0DT2 = 0x01, LIN0DT3 = 0x02, LIN0DT4 = 0x03, LIN0DT5 = 0x04, LIN0DT6 = 0x05, LIN0DT7 = 0x06, LIN0DT8 = 0x07

Bit	Name	Function
7:0	DATAn[7:0]	**LIN Data Byte n.** Serial Data Byte that is received or transmitted across the LIN interface.

21.2. CAN Registers

CAN registers are classified as follows:

1. **CAN Controller Protocol Registers**: CAN control, interrupt, error control, bus status, test modes.
2. **Message Object Interface Registers**: Used to configure 32 Message Objects, send and receive data to and from Message Objects. The CIP-51 MCU accesses the CAN message RAM via the Message Object Interface Registers. Upon writing a message object number to an IF1 or IF2 Command Request Register, the contents of the associated Interface Registers (IF1 or IF2) will be transferred to or from the message object in CAN RAM.
3. **Message Handler Registers**: These read only registers are used to provide information to the CIP-51 MCU about the message objects (MSGVLD flags, Transmission Request Pending, New Data Flags) and Interrupts Pending (which Message Objects have caused an interrupt or status interrupt condition).

For the registers other than CAN0CFG, refer to the Bosch CAN User's Guide for information on the function and use of the CAN Control Protocol Registers.

21.2.1. CAN Controller Protocol Registers

The CAN Control Protocol Registers are used to configure the CAN controller, process interrupts, monitor bus status, and place the controller in test modes.

The registers are: CAN Control Register (CAN0CN), CAN Clock Configuration (CAN0CFG), CAN Status Register (CAN0STA), CAN Test Register (CAN0TST), Error Counter Register, Bit Timing Register, and the Baud Rate Prescaler (BRP) Extension Register.

21.2.2. Message Object Interface Registers

There are two sets of Message Object Interface Registers used to configure the 32 Message Objects that transmit and receive data to and from the CAN bus. Message objects can be configured for transmit or receive, and are assigned arbitration message identifiers for acceptance filtering by all CAN nodes.

Message Objects are stored in Message RAM, and are accessed and configured using the Message Object Interface Registers.

21.2.3. Message Handler Registers

The Message Handler Registers are read only registers. The message handler registers provide interrupt, error, transmit/receive requests, and new data information.

All transactions are initiated by a master, with one or more addressed slave devices as the target. The master generates the START condition and then transmits the slave address and direction bit. If the transaction is a WRITE operation from the master to the slave, the master transmits the data a byte at a time waiting for an ACK from the slave at the end of each byte. For READ operations, the slave transmits the data waiting for an ACK from the master at the end of each byte. At the end of the data transfer, the master generates a STOP condition to terminate the transaction and free the bus. Figure 22.3 illustrates a typical SMBus transaction.

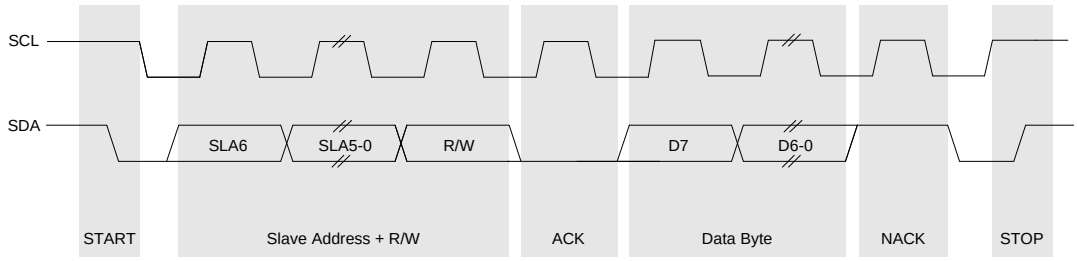

Figure 22.3. SMBus Transaction

22.3.1. Transmitter Vs. Receiver

On the SMBus communications interface, a device is the “transmitter” when it is sending an address or data byte to another device on the bus. A device is a “receiver” when an address or data byte is being sent to it from another device on the bus. The transmitter controls the SDA line during the address or data byte. After each byte of address or data information is sent by the transmitter, the receiver sends an ACK or NACK bit during the ACK phase of the transfer, during which time the receiver controls the SDA line.

22.3.2. Arbitration

A master may start a transfer only if the bus is free. The bus is free after a STOP condition or after the SCL and SDA lines remain high for a specified time (see Section “22.3.5. SCL High (SMBus Free) Timeout” on page 221). In the event that two or more devices attempt to begin a transfer at the same time, an arbitration scheme is employed to force one master to give up the bus. The master devices continue transmitting until one attempts a HIGH while the other transmits a LOW. Since the bus is open-drain, the bus will be pulled LOW. The master attempting the HIGH will detect a LOW SDA and lose the arbitration. The winning master continues its transmission without interruption; the losing master becomes a slave and receives the rest of the transfer if addressed. This arbitration scheme is non-destructive: one device always wins, and no data is lost.

22.3.3. Clock Low Extension

SMBus provides a clock synchronization mechanism, similar to I^2C, which allows devices with different speed capabilities to coexist on the bus. A clock-low extension is used during a transfer in order to allow slower slave devices to communicate with faster masters. The slave may temporarily hold the SCL line LOW to extend the clock low period, effectively decreasing the serial clock frequency.

22.3.4. SCL Low Timeout

If the SCL line is held low by a slave device on the bus, no further communication is possible. Furthermore, the master cannot force the SCL line high to correct the error condition. To solve this problem, the SMBus protocol specifies that devices participating in a transfer must detect any clock cycle held low longer than 25 ms as a “timeout” condition. Devices that have detected the timeout condition must reset the communication no later than 10 ms after detecting the timeout condition.

When the SMBTOE bit in SMB0CF is set, Timer 3 is used to detect SCL low timeouts. Timer 3 is forced to reload when SCL is high, and allowed to count when SCL is low. With Timer 3 enabled and configured to

SFR Definition 25.8. TMR2CN: Timer 2 Control

Bit	7	6	5	4	3	2	1	0
Name	TF2H	TF2L	TF2LEN	TF2CEN	T2SPLIT	TR2		T2XCLK
Type	R/W	R/W	R/W	R/W	R/W	R/W	R	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC8; Bit-Addressable; SFR Page = 0x00

Bit	Name	Function
7	TF2H	**Timer 2 High Byte Overflow Flag.** Set by hardware when the Timer 2 high byte overflows from 0xFF to 0x00. In 16 bit mode, this will occur when Timer 2 overflows from 0xFFFF to 0x0000. When the Timer 2 interrupt is enabled, setting this bit causes the CPU to vector to the Timer 2 interrupt service routine. This bit is not automatically cleared by hardware.
6	TF2L	**Timer 2 Low Byte Overflow Flag.** Set by hardware when the Timer 2 low byte overflows from 0xFF to 0x00. TF2L will be set when the low byte overflows regardless of the Timer 2 mode. This bit is not automatically cleared by hardware.
5	TF2LEN	**Timer 2 Low Byte Interrupt Enable.** When set to 1, this bit enables Timer 2 Low Byte interrupts. If Timer 2 interrupts are also enabled, an interrupt will be generated when the low byte of Timer 2 overflows.
4	TF2CEN	**Timer 2 Capture Mode Enable.** 0: Timer 2 Capture Mode is disabled. 1: Timer 2 Capture Mode is enabled.
3	T2SPLIT	**Timer 2 Split Mode Enable.** When this bit is set, Timer 2 operates as two 8-bit timers with auto-reload. 0: Timer 2 operates in 16-bit auto-reload mode. 1: Timer 2 operates as two 8-bit auto-reload timers.
2	TR2	**Timer 2 Run Control.** Timer 2 is enabled by setting this bit to 1. In 8-bit mode, this bit enables/disables TMR2H only; TMR2L is always enabled in split mode.
1	Unused	Read = 0b; Write = Don’t Care
0	T2XCLK	**Timer 2 External Clock Select.** This bit selects the external clock source for Timer 2. If Timer 2 is in 8-bit mode, this bit selects the external oscillator clock source for both timer bytes. However, the Timer 2 Clock Select bits (T2MH and T2ML in register CKCON) may still be used to select between the external clock and the system clock for either timer. 0: Timer 2 clock is the system clock divided by 12. 1: Timer 2 clock is the external clock divided by 8 (synchronized with SYSCLK).

SFR Definition 25.16. TMR3L: Timer 3 Low Byte

Bit	7	6	5	4	3	2	1	0
Name	TMR3L[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x94; SFR Page = 0x00

Bit	Name	Function
7:0	TMR3L[7:0]	**Timer 3 Low Byte.** In 16-bit mode, the TMR3L register contains the low byte of the 16-bit Timer 3. In 8-bit mode, TMR3L contains the 8-bit low byte timer value.

SFR Definition 25.17. TMR3H Timer 3 High Byte

Bit	7	6	5	4	3	2	1	0
Name	TMR3H[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x95; SFR Page = 0x00

Bit	Name	Function
7:0	TMR3H[7:0]	**Timer 3 High Byte.** In 16-bit mode, the TMR3H register contains the high byte of the 16-bit Timer 3. In 8-bit mode, TMR3H contains the 8-bit high byte timer value.

26.1. PCA Counter/Timer

The 16-bit PCA counter/timer consists of two 8-bit SFRs: PCA0L and PCA0H. PCA0H is the high byte (MSB) of the 16-bit counter/timer and PCA0L is the low byte (LSB). Reading PCA0L automatically latches the value of PCA0H into a “snapshot” register; the following PCA0H read accesses this “snapshot” register. **Reading the PCA0L Register first guarantees an accurate reading of the entire 16-bit PCA0 counter.** Reading PCA0H or PCA0L does not disturb the counter operation. The CPS[2:0] bits in the PCA0MD register select the timebase for the counter/timer as shown in Table 26.1.

When the counter/timer overflows from 0xFFFF to 0x0000, the Counter Overflow Flag (CF) in PCA0MD is set to logic 1 and an interrupt request is generated if CF interrupts are enabled. Setting the ECF bit in PCA0MD to logic 1 enables the CF flag to generate an interrupt request. The CF bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Clearing the CIDL bit in the PCA0MD register allows the PCA to continue normal operation while the CPU is in Idle mode.

Table 26.1. PCA Timebase Input Options

CPS2	CPS1	CPS0	Timebase
0	0	0	System clock divided by 12.
0	0	1	System clock divided by 4.
0	1	0	Timer 0 overflow.
0	1	1	High-to-low transitions on ECI (max rate = system clock divided by 4).
1	0	0	System clock.
1	0	1	External oscillator source divided by 8.*
1	1	x	Reserved.
***Note:** External oscillator source divided by 8 is synchronized with the system clock.			

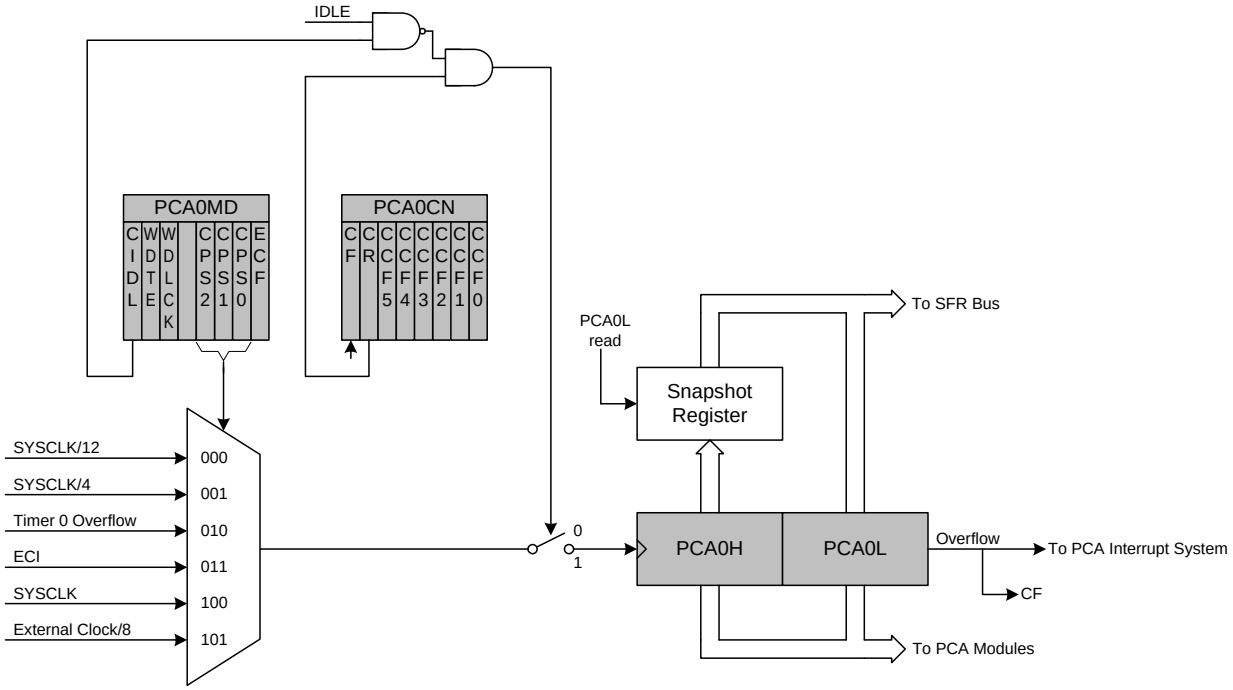

Figure 26.2. PCA Counter/Timer Block Diagram

26.2. PCA0 Interrupt Sources

Figure 26.3 shows a diagram of the PCA interrupt tree. There are five independent event flags that can be used to generate a PCA0 interrupt. They are as follows: the main PCA counter overflow flag (CF), which is set upon a 16-bit overflow of the PCA0 counter, an intermediate overflow flag (COVF), which can be set on an overflow from the 8th, 9th, 10th, or 11th bit of the PCA0 counter, and the individual flags for each PCA channel (CCF0, CCF1, CCF2, CCF3, CCF4, and CCF5), which are set according to the operation mode of that module. These event flags are always set when the trigger condition occurs. Each of these flags can be individually selected to generate a PCA0 interrupt, using the corresponding interrupt enable flag (ECF for CF, ECOV for COVF, and ECCFn for each CCFn). PCA0 interrupts must be globally enabled before any individual interrupt sources are recognized by the processor. PCA0 interrupts are globally enabled by setting the EA bit and the EPCA0 bit to logic 1.

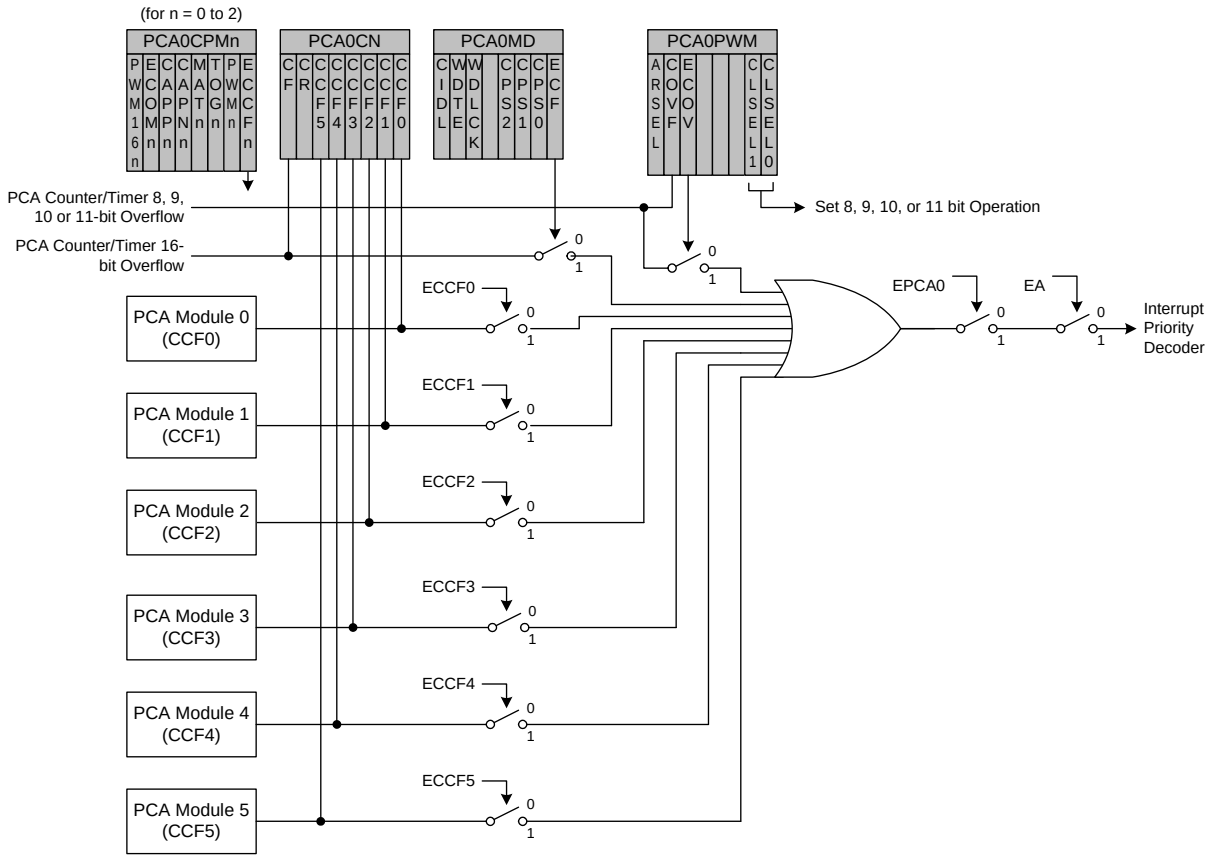

Figure 26.3. PCA Interrupt Block Diagram

26.3. Capture/Compare Modules

Each module can be configured to operate independently in one of six operation modes: Edge-triggered Capture, Software Timer, High Speed Output, Frequency Output, 8 to 11-Bit Pulse Width Modulator, or 16-Bit Pulse Width Modulator. Each module has Special Function Registers (SFRs) associated with it in the CIP-51 system controller. These registers are used to exchange data with a module and configure the module's mode of operation. Table 26.2 summarizes the bit settings in the PCA0CPMn and PCA0PWM registers used to select the PCA capture/compare module's operating mode. All modules set to use 8, 9, 10, or 11-bit PWM mode must use the same cycle length (8-11 bits). Setting the ECCFn bit in a PCA0CPMn register enables the module's CCFn interrupt.

Table 26.2. PCA0CPM and PCA0PWM Bit Settings for PCA Capture/Compare Modules

Operational Mode	PCA0CPMn								PCA0PWM				
Bit Number	7	6	5	4	3	2	1	0	7	6	5	4–2	1–0
Capture triggered by positive edge on CEXn	X	X	1	0	0	0	0	A	0	X	B	XXX	XX
Capture triggered by negative edge on CEXn	X	X	0	1	0	0	0	A	0	X	B	XXX	XX
Capture triggered by any transition on CEXn	X	X	1	1	0	0	0	A	0	X	B	XXX	XX
Software Timer	X	C	0	0	1	0	0	A	0	X	B	XXX	XX
High Speed Output	X	C	0	0	1	1	0	A	0	X	B	XXX	XX
Frequency Output	X	C	0	0	0	1	1	A	0	X	B	XXX	XX
8-Bit Pulse Width Modulator (7)	0	C	0	0	E	0	1	A	0	X	B	XXX	00
9-Bit Pulse Width Modulator (7)	0	C	0	0	E	0	1	A	D	X	B	XXX	01
10-Bit Pulse Width Modulator (7)	0	C	0	0	E	0	1	A	D	X	B	XXX	10
11-Bit Pulse Width Modulator (7)	0	C	0	0	E	0	1	A	D	X	B	XXX	11
16-Bit Pulse Width Modulator	1	C	0	0	E	0	1	A	0	X	B	XXX	XX

Notes:
1. X = Don't Care (no functional difference for individual module if 1 or 0).
2. A = Enable interrupts for this module (PCA interrupt triggered on CCFn set to 1).
3. B = Enable 8th, 9th, 10th or 11th bit overflow interrupt (Depends on setting of CLSEL[1:0]).
4. C = When set to 0, the digital comparator is off. For high speed and frequency output modes, the associated pin will not toggle. In any of the PWM modes, this generates a 0% duty cycle (output = 0).
5. D = Selects whether the Capture/Compare register (0) or the Auto-Reload register (1) for the associated channel is accessed via addresses PCA0CPHn and PCA0CPLn.
6. E = When set, a match event will cause the CCFn flag for the associated channel to be set.
7. All modules set to 8, 9, 10 or 11-bit PWM mode use the same cycle length setting.

26.3.1. Edge-triggered Capture Mode

In this mode, a valid transition on the CEXn pin causes the PCA to capture the value of the PCA counter/timer and load it into the corresponding module's 16-bit capture/compare register (PCA0CPLn and PCA0CPHn). The CAPPn and CAPNn bits in the PCA0CPMn register are used to select the type of transition that triggers the capture: low-to-high transition (positive edge), high-to-low transition (negative edge), or either transition (positive or negative edge). When a capture occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1. An interrupt request is generated if the CCFn interrupt for that module is enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. If both CAPPn and CAPNn bits are set to logic 1, then the state of the Port pin associated with CEXn can be read directly to determine whether a rising-edge or falling-edge caused the capture.