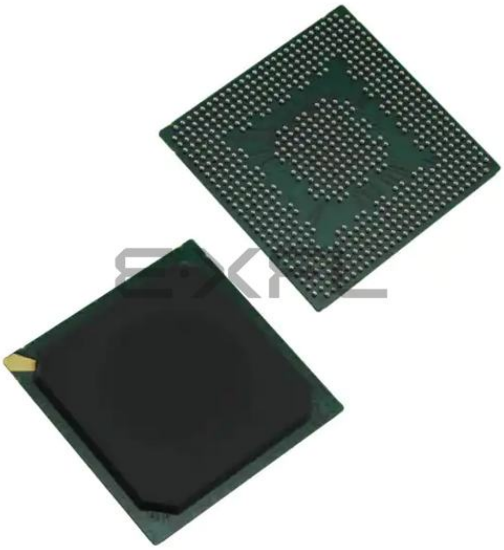

Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	-
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (3)
SATA	-
USB	USB 2.0 + PHY (2)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	620-BBGA Exposed Pad
Supplier Device Package	620-HBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8343zqagdb

NOTE

The information in this document is accurate for revision 3.x silicon and later (in other words, for orderable part numbers ending in A or B). For information on revision 1.1 silicon and earlier versions, see the *MPC8343E PowerQUICC II Pro Integrated Host Processor Hardware Specifications*.

See Section 22.1, “Part Numbers Fully Addressed by This Document,” for silicon revision level determination.

1 Overview

This section provides a high-level overview of the device features. Figure 1 shows the major functional units within the MPC8343EA.

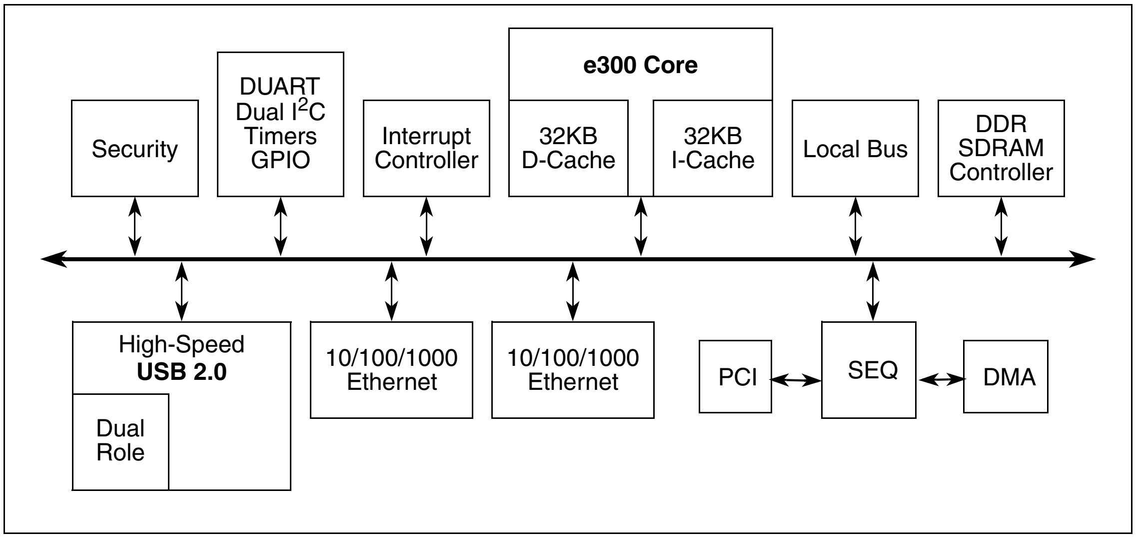

Figure 1. MPC8343EA Block Diagram

Major features of the device are as follows:

- Embedded PowerPC e300 processor core; operates at up to 400 MHz
  - High-performance, superscalar processor core
  - Floating-point, integer, load/store, system register, and branch processing units
  - 32-Kbyte instruction cache, 32-Kbyte data cache
  - Lockable portion of L1 cache
  - Dynamic power management
  - Software-compatible with the other Freescale processor families that implement Power Architecture technology
- Double data rate, DDR1/DDR2 SDRAM memory controller
  - Programmable timing supporting DDR1 and DDR2 SDRAM
  - 32- bit data interface, up to 266 MHz data rate

Figure 3 shows the undershoot and overshoot voltage of the PCI interface of the MPC8343EA for the 3.3-V signals, respectively.

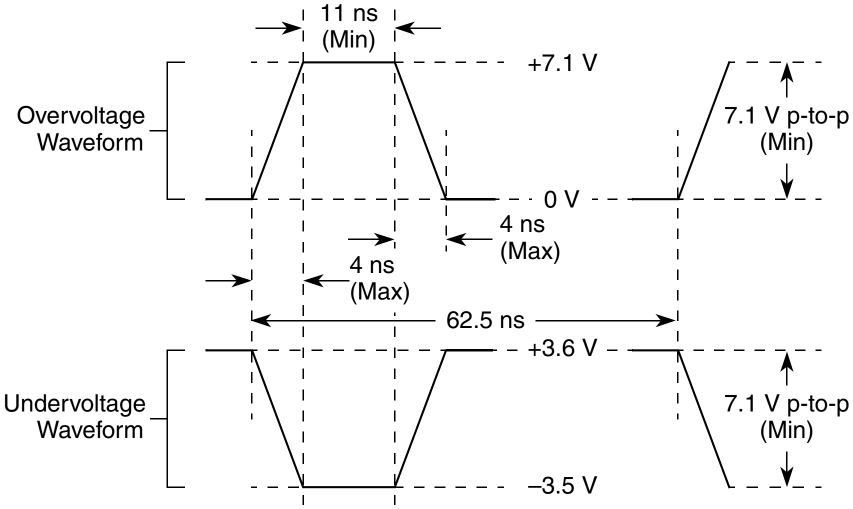

Figure 3. Maximum AC Waveforms on PCI Interface for 3.3-V Signaling

2.1.3 Output Driver Characteristics

Table 3 provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

Table 3. Output Drive Capability

Driver Type	Output Impedance (Ω)	Supply Voltage
Local bus interface utilities signals	40	OV_{DD} = 3.3 V
PCI signals (not including PCI output clocks)	25	
PCI output clocks (including PCI_SYNC_OUT)	40	
DDR signal	18	GV_{DD} = 2.5 V
DDR2 signal	18 36 (half-strength mode)	GV_{DD} = 1.8 V
TSEC/10/100 signals	40	LV_{DD} = 2.5/3.3 V
DUART, system control, I^2C, JTAG, USB	40	OV_{DD} = 3.3 V
GPIO signals	40	OV_{DD} = 3.3 V, LV_{DD} = 2.5/3.3 V

2.2 Power Sequencing

This section details the power sequencing considerations for the MPC8343EA.

2.2.1 Power-Up Sequencing

MPC8343EAdoes not require the core supply voltage (V_{DD} and AV_{DD}) and I/O supply voltages (GV_{DD}, LV_{DD}, and OV_{DD}) to be applied in any particular order. During the power ramp up, before the power

Table 13 provides the DDR2 capacitance when GV_{DD}(typ) = 1.8 V.

Table 13. DDR2 SDRAM Capacitance for GV_{DD}(typ) = 1.8 V

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{DIO}	—	0.5	pF	1

Note:

1. This parameter is sampled. GV_{DD} = 1.8 V ± 0.090 V, f = 1 MHz, T_A = 25°C, V_{OUT} = GV_{DD}/2, V_{OUT} (peak-to-peak) = 0.2 V.

Table 14 provides the recommended operating conditions for the DDR SDRAM component(s) when GV_{DD}(typ) = 2.5 V.

Table 14. DDR SDRAM DC Electrical Characteristics for GV_{DD}(typ) = 2.5 V

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O supply voltage	GV_{DD}	2.375	2.625	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} - 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.18$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	–0.3	$MV_{REF} - 0.18$	V	—
Output leakage current	I_{OZ}	–9.9	–9.9	μA	4
Output high current (V_{OUT} = 1.95 V)	I_{OH}	–15.2	—	mA	—
Output low current (V_{OUT} = 0.35 V)	I_{OL}	15.2	—	mA	—

Notes:

1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. MV_{REF} is expected to be equal to $0.5 \times GV_{DD}$, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF}. This rail should track variations in the DC level of MV_{REF}.
4. Output leakage is measured with all outputs disabled, $0 \text{ V} \leq V_{OUT} \leq GV_{DD}$.

Table 15 provides the DDR capacitance when GV_{DD}(typ) = 2.5 V.

Table 15. DDR SDRAM Capacitance for GV_{DD}(typ) = 2.5 V

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Input/output capacitance: DQ, DQS	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS	C_{DIO}	—	0.5	pF	1

Note:

1. This parameter is sampled. GV_{DD} = 2.5 V ± 0.125 V, f = 1 MHz, T_A = 25°C, V_{OUT} = GV_{DD}/2, V_{OUT} (peak-to-peak) = 0.2 V.

Table 20. DDR and DDR2 SDRAM Output AC Timing Specifications (continued)

At recommended operating conditions with GV_{DD} of (1.8 or 2.5 V) ± 5%.

Parameter	Symbol [1]	Min	Max	Unit	Notes
ADDR/CMD/MODT output hold with respect to MCK	t_{DDKHAX}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
$\overline{MCS}$(n) output setup with respect to MCK	t_{DDKHCS}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
$\overline{MCS}$(n) output hold with respect to MCK	t_{DDKHCX}			ns	3
400 MHz		1.95	—		
333 MHz		2.40	—		
266 MHz		3.15	—		
200 MHz		4.20	—		
MCK to MDQS Skew	t_{DDKHMH}	–0.6	0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS	t_{DDKHDS}, t_{DDKLDS}			ps	5
400 MHz		700	—		
333 MHz		775	—		
266 MHz		1100	—		
200 MHz		1200	—		
MDQ/MECC/MDM output hold with respect to MDQS	t_{DDKHDX}, t_{DDKLDX}			ps	5
400 MHz		700	—		
333 MHz		900	—		
266 MHz		1100	—		
200 MHz		1200	—		
MDQS preamble start	t_{DDKHMP}	$-0.5 \times t_{MCK} - 0.6$	$-0.5 \times t_{MCK} + 0.6$	ns	6

Table 21. DUART DC Electrical Characteristics (continued)

Parameter	Symbol	Min	Max	Unit
High-level output voltage, I_{OH} = –100 μA	V_{OH}	OV_{DD} – 0.2	—	V
Low-level output voltage, I_{OL} = 100 μA	V_{OL}	—	0.2	V

7.2 DUART AC Electrical Specifications

Table 22 provides the AC timing parameters for the DUART interface of the MPC8343EA.

Table 22. DUART AC Timing Specifications

Parameter	Value	Unit	Notes
Minimum baud rate	256	baud	—
Maximum baud rate	> 1,000,000	baud	1
Oversample rate	16	—	2

Notes:

1. Actual attainable baud rate will be limited by the latency of interrupt processing.
2. The middle of a start bit is detected as the 8th sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16th sample.

8 Ethernet: Three-Speed Ethernet, MII Management

This section provides the AC and DC electrical characteristics for three-speeds (10/100/1000 Mbps) and MII management.

8.1 Three-Speed Ethernet Controller (TSEC)—MII/RGMII/RTBI Electrical Characteristics

The electrical characteristics specified here apply to media independent interface (MII), reduced gigabit media independent interface (RGMII), and reduced ten-bit interface (RTBI) signals except management data input/output (MDIO) and management data clock (MDC). The MII interface is defined for 3.3 V, and the RGMII and RTBI interfaces are defined for 2.5 V. The RGMII and RTBI interfaces follow the Hewlett-Packard *Reduced Pin-Count Interface for Gigabit Ethernet Physical Layer Device Specification*, Version 1.2a (9/22/2000). The electrical characteristics for MDIO and MDC are specified in Section 8.3, "Ethernet Management Interface Electrical Characteristics."

10.2 Local Bus AC Electrical Specification

Table 34 and Table 35 describe the general timing parameters of the local bus interface of the MPC8343EA.

Table 34. Local Bus General Timing Parameters—DLL On

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	7.5	—	ns	2
Input setup to local bus clock (except LUPWAIT)	$t_{LBIVKH1}$	1.5	—	ns	3, 4
LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	2.2	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)	$t_{LBIXKH1}$	1.0	—	ns	3, 4
LUPWAIT Input hold from local bus clock	$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7
Local bus clock to LALE rise	t_{LBKHLR}	—	4.5	ns	—
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	4.5	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	4.5	ns	3
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	4.5	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	1	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	1	—	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	3.8	ns	8

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to the rising edge of LSYNC_IN.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN to $0.4 \times OV_{DD}$ of the signal in question for 3.3 V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin equals the load on the LAD output pins.
8. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to that of the leakage current specification.

Figure 17 through Figure 22 show the local bus signals.

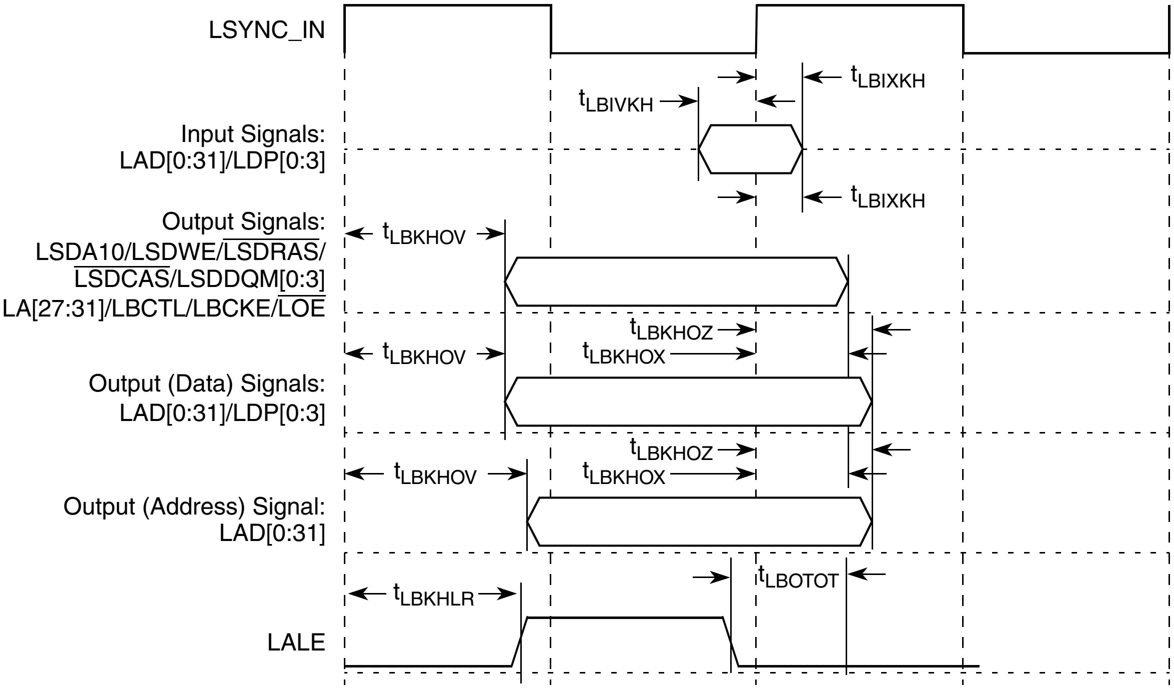

Figure 17. Local Bus Signals, Nonspecial Signals Only (DLL Enabled)

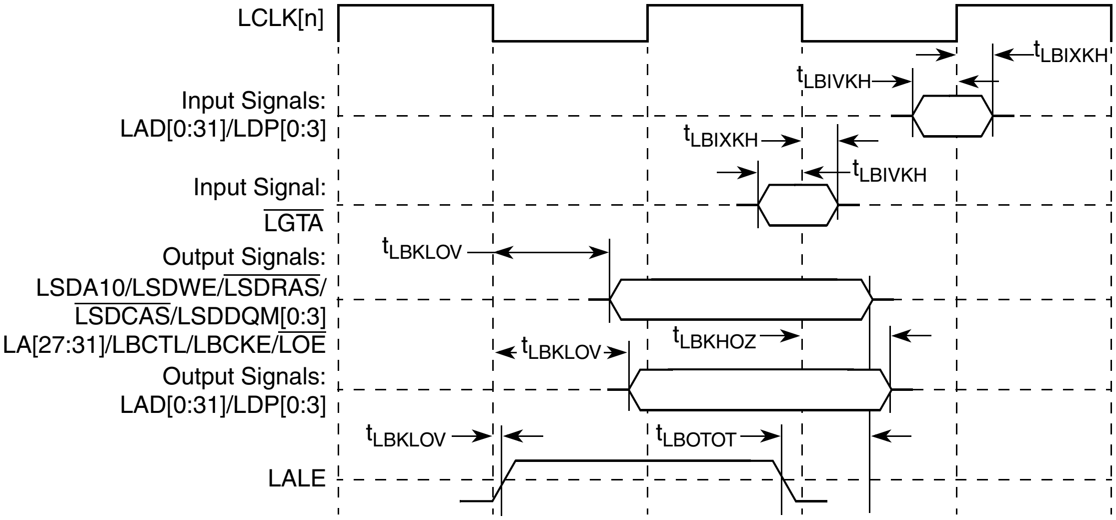

Figure 18. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

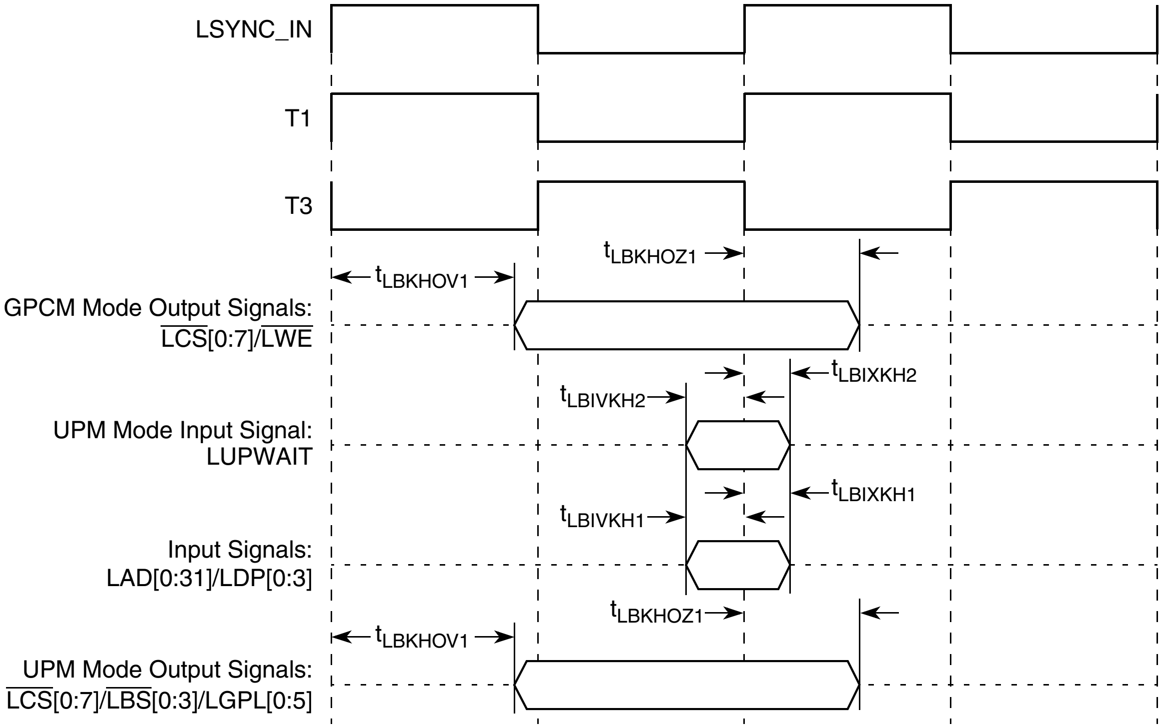

Figure 19. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Enabled)

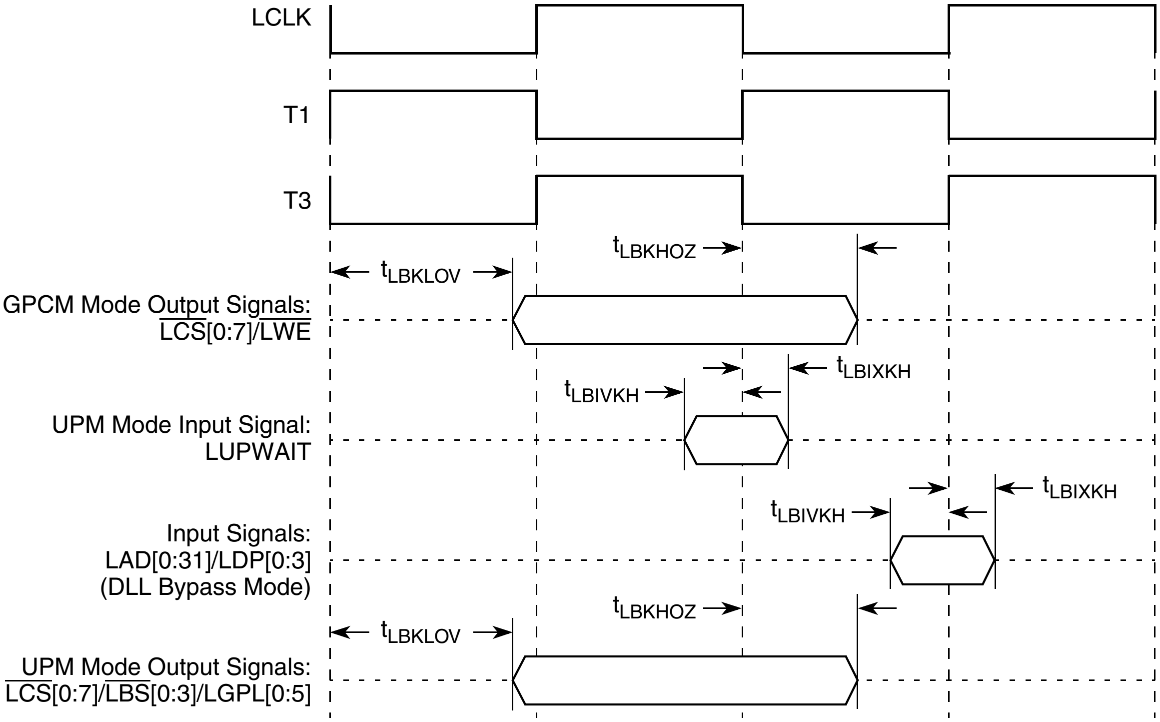

Figure 20. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 2 (DLL Bypass Mode)

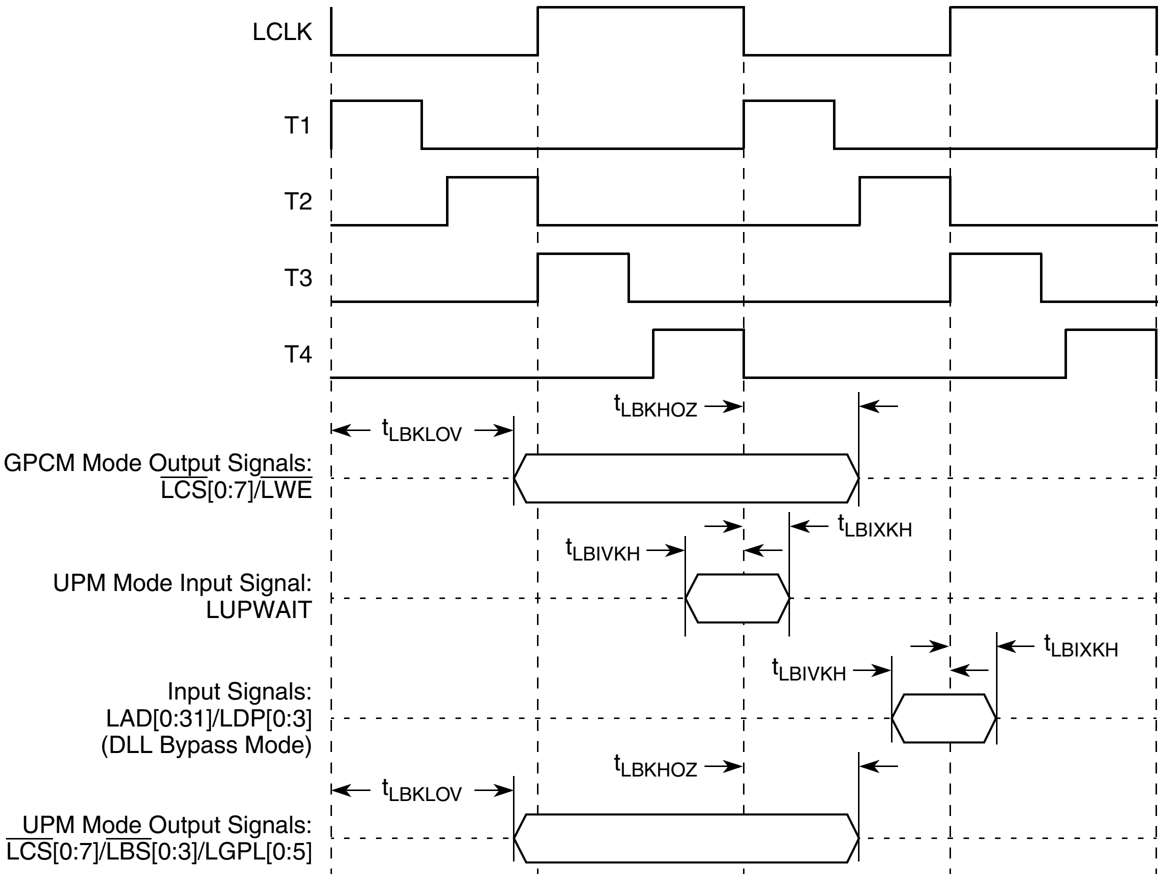

Figure 21. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (DLL Bypass Mode)

Figure 26 provides the boundary-scan timing diagram.

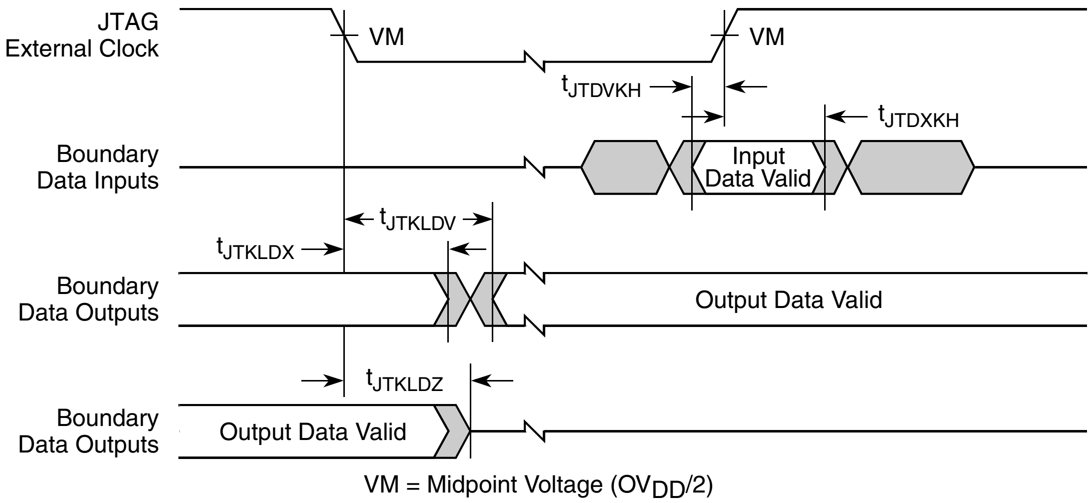

Figure 26. Boundary-Scan Timing Diagram

Figure 27 provides the test access port timing diagram.

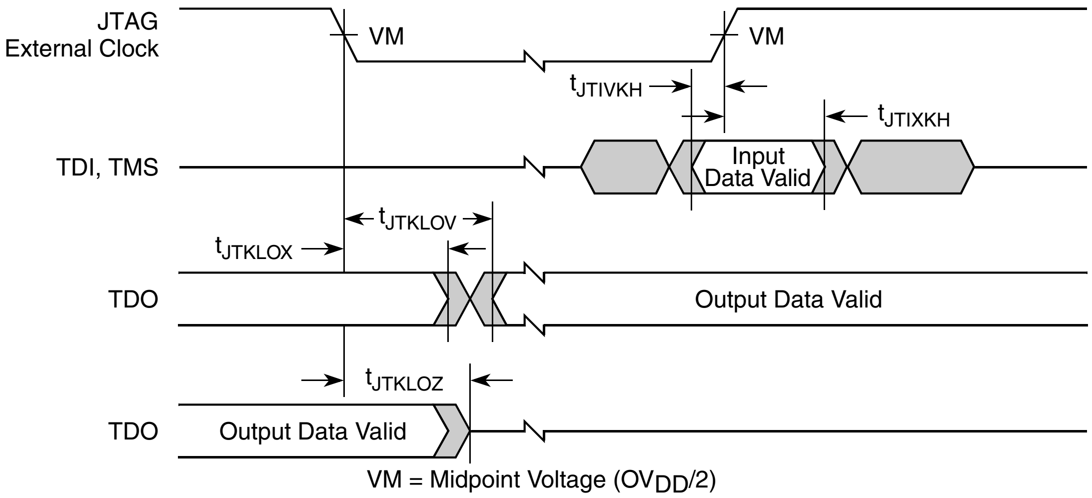

Figure 27. Test Access Port Timing Diagram

Table 41. PCI AC Timing Specifications at 66 MHz[1] (continued)

Parameter	Symbol[2]	Min	Max	Unit	Notes
Input hold from clock	t_{PCIXKH}	0	—	ns	3, 5

Notes:

1. PCI timing depends on M66EN and the ratio between PCI1/PCI2. Refer to the PCI chapter of the reference manual for a description of M66EN.
2. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
3. See the timing measurement conditions in the *PCI 2.3 Local Bus Specifications*.
4. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Input timings are measured at the pin.

Table 42 provides the PCI AC timing specifications at 33 MHz.

Table 42. PCI AC Timing Specifications at 33 MHz

Parameter	Symbol[1]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	11	ns	2
Output hold from clock	t_{PCKHOX}	2	—	ns	2
Clock to output high impedance	t_{PCKHOZ}	—	14	ns	2, 3
Input setup to clock	t_{PCIVKH}	3.0	—	ns	2, 4
Input hold from clock	t_{PCIXKH}	0	—	ns	2, 4

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.3 Local Bus Specifications*.
3. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
4. Input timings are measured at the pin.

Figure 30 provides the AC test load for PCI.

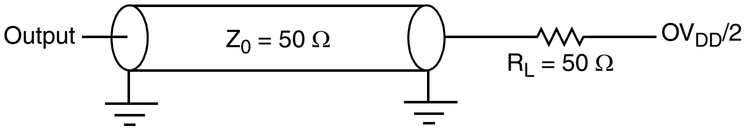

Figure 30. PCI AC Test Load

Figure 31 shows the PCI input AC timing diagram.

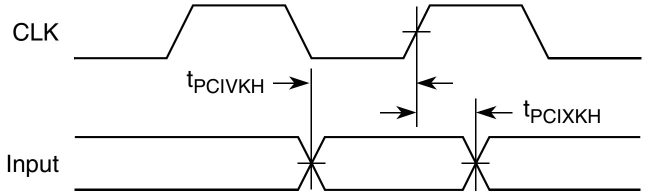

Figure 31. PCI Input AC Timing Diagram

Figure 32 shows the PCI output AC timing diagram.

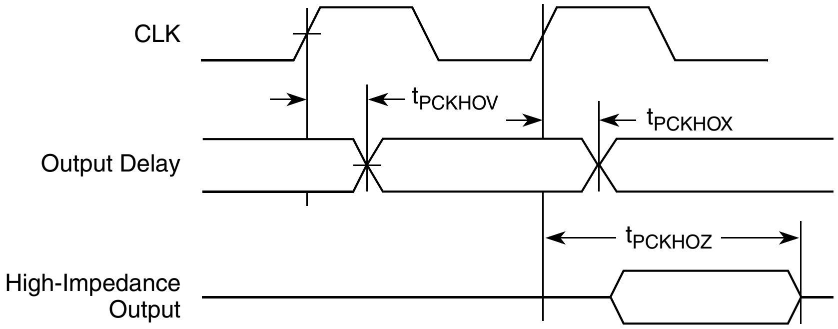

Figure 32. PCI Output AC Timing Diagram

14 Timers

This section describes the DC and AC electrical specifications for the timers.

14.1 Timer DC Electrical Characteristics

Table 43 provides the DC electrical characteristics for the MPC8343EA timer pins, including TIN, $\overline{\text{TOUT}}$, $\overline{\text{TGATE}}$, and RTC_CLK.

Table 43. Timer DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
MECC[0:4]/MSRCID[0:4]	AG13, AE14, AH12, AH10, AE15	I/O	GV_{DD}	—
MECC[5]/MDVAL	AH14	I/O	GV_{DD}	—
MECC[6:7]	AE13, AH11	I/O	GV_{DD}	—
MDM[0:3]	AG28, AG24, AF20, AG17	O	GV_{DD}	—
MDM[8]	AG12	O	GV_{DD}	—
MDQS[0:3]	AE27, AE26, AE20, AH18	I/O	GV_{DD}	—
MDQS[8]	AH13	I/O	GV_{DD}	—
MBA[0:1]	AF10, AF11	O	GV_{DD}	—
MA[0:14]	AF13, AF15, AG16, AD16, AF17, AH20, AH19, AH21, AD18, AG21, AD13, AF21, AF22, AE1, AA5	O	GV_{DD}	—
$\overline{\text{MWE}}$	AD10	O	GV_{DD}	—
$\overline{\text{MRAS}}$	AF7	O	GV_{DD}	—
$\overline{\text{MCAS}}$	AG6	O	GV_{DD}	—
$\overline{\text{MCS}}$[0:3]	AE7, AH7, AH4, AF2	O	GV_{DD}	—
MCKE[0:1]	AG23, AH23	O	GV_{DD}	3
MCK[0:3]	AH15, AE24, AE2, AF14	O	GV_{DD}	—
$\overline{\text{MCK}}$[0:3]	AG15, AD23, AE3, AG14	O	GV_{DD}	—
MODT[0:3]	AG5, AD4, AH6, AF4	O	GV_{DD}	—
MBA[2]	AD22	O	GV_{DD}	—
MDIC0	AG11	I/O	—	9
MDIC1	AF12	I/O	—	9
Local Bus Controller Interface				
LAD[0:31]	T4, T5, T1, R2, R3, T2, R1, R4, P1, P2, P3, P4, N1, N4, N2, N3, M1, M2, M3, N5, M4, L1, L2, L3, K1, M5, K2, K3, J1, J2, L5, J3	I/O	OV_{DD}	—
LDP[0]/$\overline{\text{CKSTOP\_OUT}}$	H1	I/O	OV_{DD}	—
LDP[1]/$\overline{\text{CKSTOP\_IN}}$	K5	I/O	OV_{DD}	—
LDP[2]/$\overline{\text{LCS}}$[4]	H2	I/O	OV_{DD}	—
LDP[3]/$\overline{\text{LCS}}$[5]	G1	I/O	OV_{DD}	—
LA[27:31]	J4, H3, G2, F1, G3	O	OV_{DD}	—
$\overline{\text{LCS}}$[0:3]	J5, H4, F2, E1	O	OV_{DD}	—
$\overline{\text{LWE}}$[0:3]/LSDDQM[0:3]/$\overline{\text{LBS}}$[0:3]	F3, G4, D1, E2	O	OV_{DD}	—

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
USB				
DR_D0_ENABLEN	C28	I/O	OV_{DD}	—
DR_D1_SER_TXD	F25	I/O	OV_{DD}	—
DR_D2_VMO_SE0	B28	I/O	OV_{DD}	—
DR_D3_SPEED	C27	I/O	OV_{DD}	—
DR_D4_DP	D26	I/O	OV_{DD}	—
DR_D5_DM	E25	I/O	OV_{DD}	—
DR_D6_SER_RCV	C26	I/O	OV_{DD}	—
DR_D7_DRVVBUS	D25	I/O	OV_{DD}	—
DR_SESS_VLD_NXT	B26	I	OV_{DD}	—
DR_XCVR_SEL_DPPULLUP	E24	I/O	OV_{DD}	—
DR_STP_SUSPEND	A27	O	OV_{DD}	—
DR_RX_ERROR_PWRFAULT	C25	I	OV_{DD}	—
DR_TX_VALID_PCTL0	A26	O	OV_{DD}	—
DR_TX_VALIDH_PCTL1	B25	O	OV_{DD}	—
DR_CLK	A25	I	OV_{DD}	—
Programmable Interrupt Controller				
$\overline{\text{MCP\_OUT}}$	E8	O	OV_{DD}	2
$\overline{\text{IRQ}}$0/$\overline{\text{MCP_IN}}$/GPIO2[12]	J28	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[1:5]/GPIO2[13:17]	K25, J25, H26, L24, G27	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[6]/GPIO2[18]/CKSTOP_OUT	G28	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[7]/GPIO2[19]/CKSTOP_IN	J26	I/O	OV_{DD}	—
Ethernet Management Interface				
EC_MDC	Y24	O	LV_{DD1}	—
EC_MDIO	Y25	I/O	LV_{DD1}	11
Gigabit Reference Clock				
EC_GTX_CLK125	Y26	I	LV_{DD1}	—
Three-Speed Ethernet Controller (Gigabit Ethernet 1)				
TSEC1_COL/GPIO2[20]	M26	I/O	OV_{DD}	—
TSEC1_CRS/GPIO2[21]	U25	I/O	LV_{DD1}	—
TSEC1_GTX_CLK	V24	O	LV_{DD1}	3
TSEC1_RX_CLK	U26	I	LV_{DD1}	—

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Power and Ground Signals				
$AV_{DD}1$	C15	Power for e300 PLL (1.2 V)	$AV_{DD}1$	—
$AV_{DD}2$	U1	Power for system PLL (1.2 V)	$AV_{DD}2$	—
$AV_{DD}3$	AF9	Power for DDR DLL (1.2 V)	—	—
$AV_{DD}4$	U2	Power for LBIU DLL (1.2 V)	$AV_{DD}4$	—
GND	A2, B1, B2, D10, D18, E6, E14, E22, F9, F12, F15, F18, F21, F24, G5, H6, J23, L4, L6, L12, L13, L14, L15, L16, L17, M11, M12, M13, M14, M15, M16, M17, M18, M23, N11, N12, N13, N14, N15, N16, N17, N18, P6, P11, P12, P13, P14, P15, P16, P17, P18, P24, R5, R23, R11, R12, R13, R14, R15, R16, R17, R18, T11, T12, T13, T14, T15, T16, T17, T18, U6, U11, U12, U13, U14, U15, U16, U17, U18, V12, V13, V14, V15, V16, V17, V23, V25, W4, Y6, AA23, AB24, AC5, AC8, AC11, AC14, AC17, AC20, AD9, AD15, AD21, AE12, AE18, AF3, AF26	—	—	—
GV_{DD}	U9, V9, W10, W19, Y11, Y12, Y14, Y15, Y17, Y18, AA6, AB5, AC9, AC12, AC15, AC18, AC21, AC24, AD6, AD8, AD14, AD20, AE5, AE11, AE17, AG2, AG27	Power for DDR DRAM I/O voltage (2.5 V)	GV_{DD}	—
LV_{DD1}	U20, W25	Power for three speed Ethernet #1 and for Ethernet management interface I/O (2.5V, 3.3V)	LV_{DD1}	—
LV_{DD2}	V20, Y23	Power for three speed Ethernet #2 I/O (2.5 V, 3.3 V)	LV_{DD2}	—
V_{DD}	J11, J12, J15, K10, K11, K12, K13, K14, K15, K16, K17, K18, K19, L10, L11, L18, L19, M10, M19, N10, N19, P9, P10, P19, R10, R19, R20, T10, T19, U10, U19, V10, V11, V18, V19, W11, W12, W13, W14, W15, W16, W17, W18	Power for core (1.2 V)	V_{DD}	—

19 Clocking

Figure 37 shows the internal distribution of the clocks.

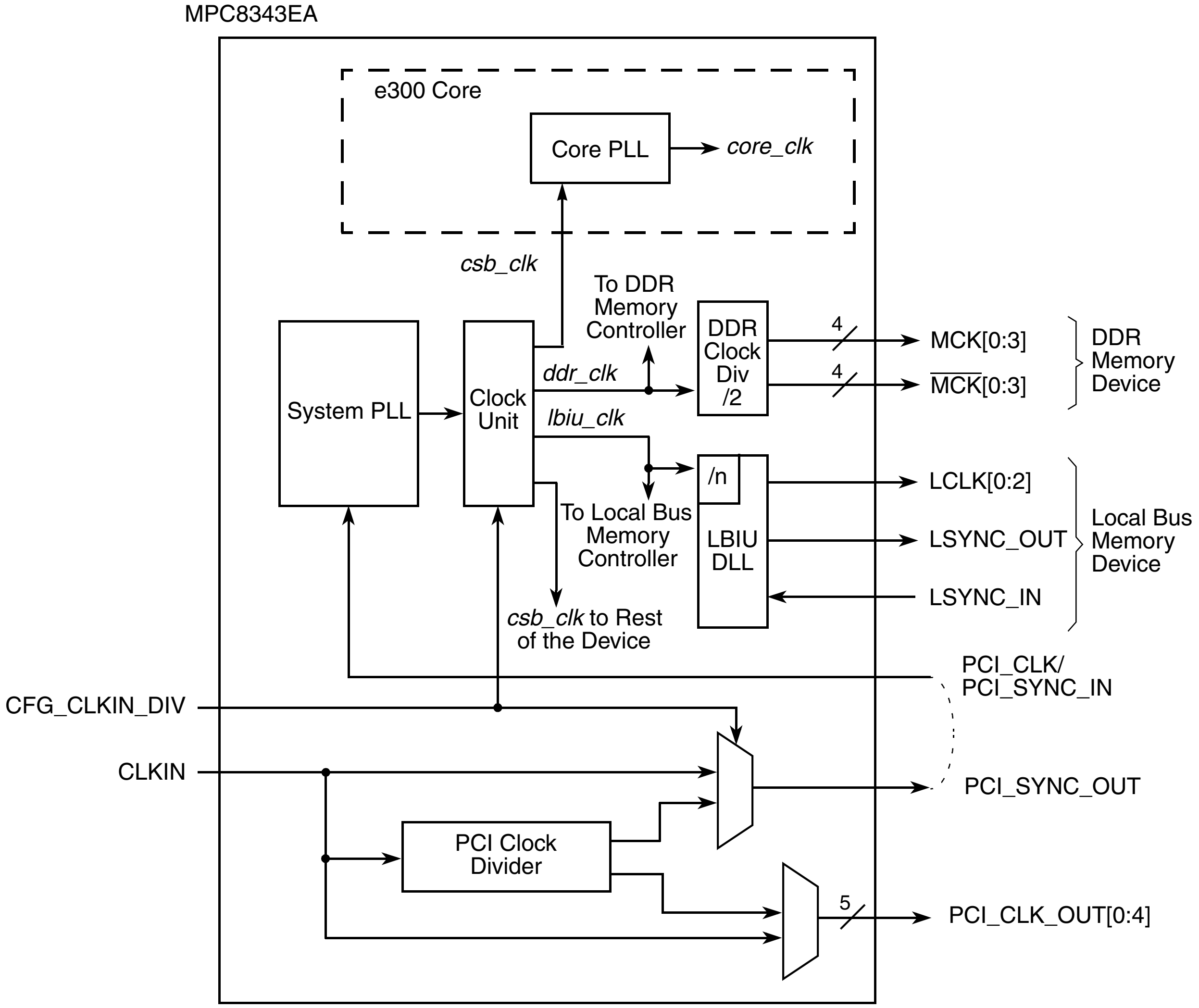

Figure 37. MPC8343EA Clock Subsystem

The primary clock source can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. When the MPC8343EA is configured as a PCI host device, CLKIN is its primary input clock. CLKIN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCICD*n*] parameters select whether CLKIN or CLKIN/2 is driven out on the PCI_CLK_OUT*n* signals.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the MPC8343EA to function. When the device is configured as a PCI agent device, PCI_CLK is the primary input clock and the CLKIN signal should be tied to GND.

Table 53 provides the operating frequencies for the MPC8343EA PBGA under recommended operating conditions.

Table 53. Operating Frequencies for PBGA

Parameter[1]	266 MHz	333 MHz	400 MHz	Unit
e300 core frequency (*core_clk*)	200–266	200–333	200–400	MHz
Coherent system bus frequency (*csb_clk*)	100–266			MHz
DDR1 memory bus frequency (MCK)[2]	100–133			MHz
DDR2 memory bus frequency (MCK)[3]	100–133			MHz
Local bus frequency (LCLK*n*)[4]	16.67–133			MHz
PCI input frequency (CLKIN or PCI_CLK)	25–66			MHz
Security core maximum internal operating frequency	133			MHz
USB_DR, USB_MPH maximum internal operating frequency	133			MHz

[1] The CLKIN frequency, RCWL[SPMF], and RCWL[COREPLL] settings must be chosen so that the resulting *csb_clk*, MCLK, LCLK[0:2], and *core_clk* frequencies do not exceed their respective maximum or minimum operating frequencies. The value of SCCR[ENCCM], SCCR[USBDRCM], and SCCR[USBMPHCM] must be programmed so that the maximum internal operating frequency of the Security core and USB modules does not exceed the respective values listed in this table.

[2] The DDR data rate is 2× the DDR memory bus frequency.

[3] The DDR data rate is 2× the DDR memory bus frequency.

[4] The local bus frequency is ½, ¼, or 1/8 of the *lbiu_clk* frequency (depending on LCCR[CLKDIV]) which is in turn 1× or 2× the *csb_clk* frequency (depending on RCWL[LBIUCM]).

19.1 System PLL Configuration

The system PLL is controlled by the RCWL[SPMF] parameter. Table 54 shows the multiplication factor encodings for the system PLL.

Table 54. System PLL Multiplication Factors

RCWL[SPMF]	System PLL Multiplication Factor
0000	× 16
0001	Reserved
0010	× 2
0011	× 3
0100	× 4
0101	× 5
0110	× 6
0111	× 7
1000	× 8
1001	× 9
1010	× 10

The junction-to-ambient thermal resistance is an industry-standard value that provides a quick and easy estimation of thermal performance. Generally, the value obtained on a single-layer board is appropriate for a tightly packed printed-circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

20.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

The thermal performance of a device cannot be adequately predicted from the junction-to-ambient thermal resistance. The thermal performance of any component is strongly dependent on the power dissipation of surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)

T_A = ambient temperature for the package (°C)

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

20.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature and a measure of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)

T_T = thermocouple temperature on top of package (°C)

Ψ_{JT} = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per the JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

20.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

Some application environments require a heat sink to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)
$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)
$R_{\theta CA}$ = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ is device-related and cannot be influenced by the user. The user controls the thermal environment to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the size of the heat sink, the air flow around the device, the interface material, the mounting arrangement on printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device.

The thermal performance of devices with heat sinks has been simulated with a few commercially available heat sinks. The heat sink choice is determined by the application environment (temperature, air flow, adjacent component power dissipation) and the physical space available. Because there is not a standard application environment, a standard heat sink is not required.

Table 60 shows heat sink thermal resistance for PBGA of the MPC8343EA.

Table 60. Heat Sink and Thermal Resistance of MPC8343EA (PBGA)

Heat Sink Assuming Thermal Grease	Air Flow	29 × 29 mm PBGA
		Thermal Resistance
AAVID 30 × 30 × 9.4 mm pin fin	Natural convection	13.5
AAVID 30 × 30 × 9.4 mm pin fin	1 m/s	9.6
AAVID 30 × 30 × 9.4 mm pin fin	2 m/s	8.8
AAVID 31 × 35 × 23 mm pin fin	Natural convection	11.3
AAVID 31 × 35 × 23 mm pin fin	1 m/s	8.1
AAVID 31 × 35 × 23 mm pin fin	2 m/s	7.5
Wakefield, 53 × 53 × 25 mm pin fin	Natural convection	9.1
Wakefield, 53 × 53 × 25 mm pin fin	1 m/s	7.1
Wakefield, 53 × 53 × 25 mm pin fin	2 m/s	6.5
MEI, 75 × 85 × 12 no adjacent board, extrusion	Natural convection	10.1

$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)

P_D = power dissipation (W)

21 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8343EA.

21.1 System Clocking

The MPC8343EA includes two PLLs:

1. The platform PLL generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 19.1, “System PLL Configuration.”
2. The e300 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 19.2, “Core PLL Configuration.”

21.2 PLL Power Supply Filtering

Each PLL gets power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always equal to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme.

There are a number of ways to provide power reliably to the PLLs, but the recommended solution is to provide four independent filter circuits as illustrated in Figure 38, one to each of the four AV_{DD} pins. Independent filters to each PLL reduce the opportunity to cause noise injection from one PLL to the other.

The circuit filters noise in the PLL resonant frequency range from 500 kHz to 10 MHz. It should be built with surface mount capacitors with minimum effective series inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

To minimize noise coupled from nearby circuits, each circuit should be placed as closely as possible to the specific AV_{DD} pin being supplied. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

Figure 38 shows the PLL power supply filter circuit.

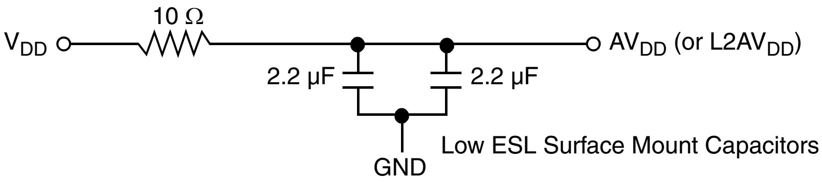

Figure 38. PLL Power Supply Filter Circuit