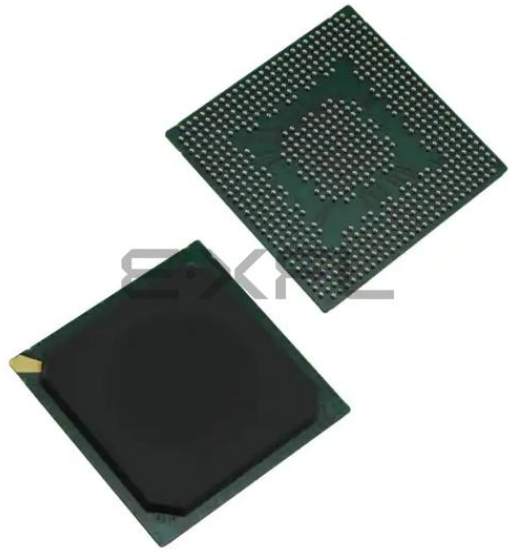

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	266MHz
Co-Processors/DSP	-
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (3)
SATA	-
USB	USB 2.0 + PHY (2)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	620-BBGA Exposed Pad
Supplier Device Package	620-HBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8343vradd

Table 8. EC_GTX_CLK125 AC Timing Specifications

At recommended operating conditions with LV_{DD} = 2.5 ± 0.125 mV/ 3.3 V ± 165 mV (continued)

Parameter	Symbol	Min	Typical	Max	Unit	Notes
EC_GTX_CLK125 duty cycle GMII, TBI 1000Base-T for RGMII, RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	2
EC_GTX_CLK125 jitter	—	—	—	±150	ps	2

Notes:

1. Rise and fall times for EC_GTX_CLK125 are measured from 0.5 and 2.0 V for LV_{DD} = 2.5 V and from 0.6 and 2.7 V for LV_{DD} = 3.3 V.
2. EC_GTX_CLK125 is used to generate the GTX clock for the eTSEC transmitter with 2% degradation. The EC_GTX_CLK125 duty cycle can be loosened from 47%/53% as long as the PHY device can tolerate the duty cycle generated by the eTSEC GTX_CLK. See Section 8.2.2, "RGMII and RTBI AC Timing Specifications for the duty cycle for 10Base-T and 100Base-T reference clock.

5 RESET Initialization

This section describes the DC and AC electrical specifications for the reset initialization timing and electrical requirements of the MPC8343EA.

5.1 RESET DC Electrical Characteristics

Table 9 provides the DC electrical characteristics for the RESET pins of the MPC8343EA.

Table 9. RESET Pins DC Electrical Characteristics[1]

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage[2]	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Notes:

1. This table applies for pins $\overline{\text{PORESET}}$, $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, and $\overline{\text{QUIESCE}}$.
2. $\overline{\text{HRESET}}$ and $\overline{\text{SRESET}}$ are open drain pins, thus V_{OH} is not relevant for those pins.

8.2.1.2 MII Receive AC Timing Specifications

Table 26 provides the MII receive AC timing specifications.

Table 26. MII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RX_CLK clock period 10 Mbps	t_{MRX}	—	400	—	ns
RX_CLK clock period 100 Mbps	t_{MRX}	—	40	—	ns
RX_CLK duty cycle	t_{MRXH}/t_{MRX}	35	—	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise (20%–80%)	t_{MRXR}	1.0	—	4.0	ns
RX_CLK clock fall time (80%–20%)	t_{MRXF}	1.0	—	4.0	ns

Note:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. In general, the clock reference symbol is based on three letters representing the clock of a particular function. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 10 provides the AC test load for TSEC.

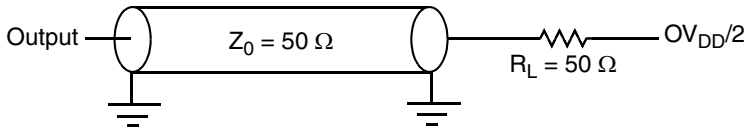

Figure 10. TSEC AC Test Load

Figure 11 shows the MII receive AC timing diagram.

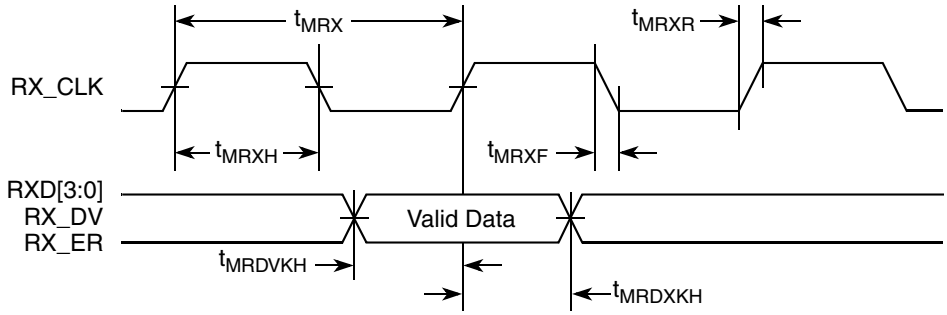

Figure 11. MII Receive AC Timing Diagram

8.2.2 RGMII and RTBI AC Timing Specifications

Table 27 presents the RGMII and RTBI AC timing specifications.

Table 27. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT}	–0.5	—	0.5	ns
Data to clock input skew (at receiver)[2]	t_{SKRGT}	1.0	—	2.8	ns
Clock cycle duration[3]	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 1000Base-T[4, 5]	t_{RGTH}/t_{RGT}	45	50	55	%
Duty cycle for 10BASE-T and 100BASE-TX[3, 5]	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (80%–20%)	t_{RGTF}	—	—	0.75	ns

Notes:

1. In general, the clock reference symbol for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Also, the notation for rise (R) and fall (F) times follows the clock symbol. For symbols representing skews, the subscript is SK followed by the clock being skewed (RGT).
2. This implies that PC board design requires clocks to be routed so that an additional trace delay of greater than 1.5 ns is added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned.
5. Duty cycle reference is $LV_{DD}/2$.

Figure 12 shows the RBMII and RTBI AC timing and multiplexing diagrams.

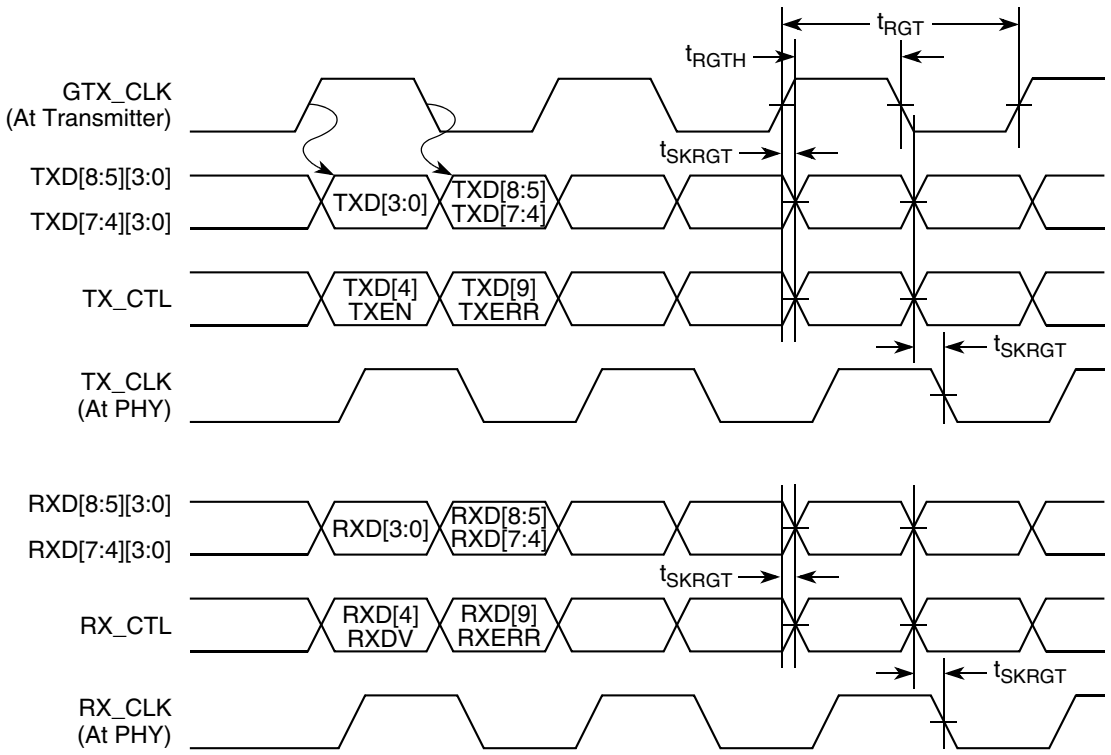

Figure 12. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.3 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to the MII management interface signals management data input/output (MDIO) and management data clock (MDC). The electrical characteristics for GMII, RGMII, TBI and RTBI are specified in Section 8.1, "Three-Speed Ethernet Controller (TSEC)—MII/RGMII/RTBI Electrical Characteristics."

8.3.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 2.5 or 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in Table 28 and Table 29.

Table 28. MII Management DC Electrical Characteristics Powered at 2.5 V

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage (2.5 V)	LV_{DD}	—		2.37	2.63	V
Output high voltage	V_{OH}	I_{OH} = –1.0 mA	LV_{DD} = Min	2.00	LV_{DD} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 1.0 mA	LV_{DD} = Min	GND – 0.3	0.40	V
Input high voltage	V_{IH}	—	LV_{DD} = Min	1.7	—	V
Input low voltage	V_{IL}	—	LV_{DD} = Min	–0.3	0.70	V

Figure 14 and Figure 15 provide the AC test load and signals for the USB, respectively.

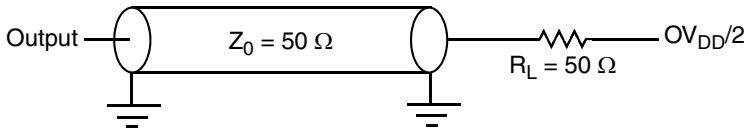

Figure 14. USB AC Test Load

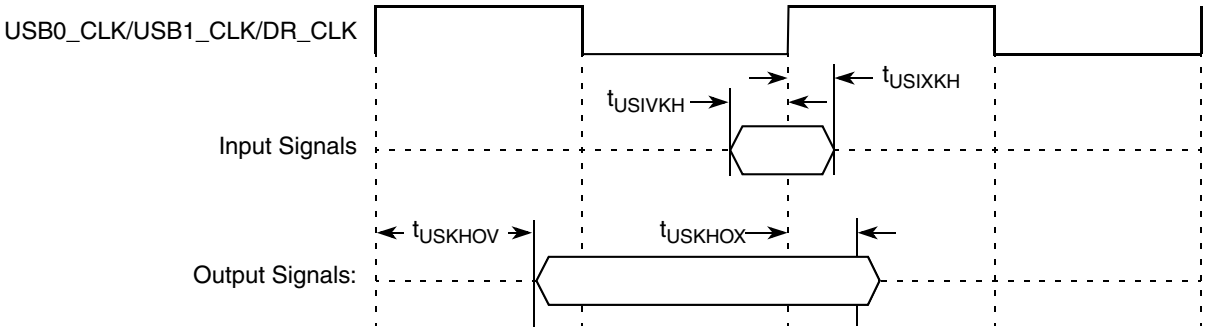

Figure 15. USB Signals

10 Local Bus

This section describes the DC and AC electrical specifications for the local bus interface of the MPC8343EA.

10.1 Local Bus DC Electrical Characteristics

Table 33 provides the DC electrical characteristics for the local bus interface.

Table 33. Local Bus DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current	I_{IN}	—	±5	μA
High-level output voltage, I_{OH} = –100 μA	V_{OH}	OV_{DD} – 0.2	—	V
Low-level output voltage, I_{OL} = 100 μA	V_{OL}	—	0.2	V

10.2 Local Bus AC Electrical Specification

Table 34 and Table 35 describe the general timing parameters of the local bus interface of the MPC8343EA.

Table 34. Local Bus General Timing Parameters—DLL On

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	7.5	—	ns	2
Input setup to local bus clock (except LUPWAIT)	$t_{LBIVKH1}$	1.5	—	ns	3, 4
LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	2.2	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)	$t_{LBIXKH1}$	1.0	—	ns	3, 4
LUPWAIT Input hold from local bus clock	$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7
Local bus clock to LALE rise	t_{LBKHLR}	—	4.5	ns	—
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	4.5	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	4.5	ns	3
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	4.5	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	1	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	1	—	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	3.8	ns	8

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to the rising edge of LSYNC_IN.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN to $0.4 \times OV_{DD}$ of the signal in question for 3.3 V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin equals the load on the LAD output pins.
8. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to that of the leakage current specification.

Table 35. Local Bus General Timing Parameters—DLL Bypass[9]

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	15	—	ns	2
Input setup to local bus clock	t_{LBIVKH}	7	—	ns	3, 4
Input hold from local bus clock	t_{LBIXKH}	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7
Local bus clock to output valid	t_{LBKLOV}	—	3	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	4	ns	8

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to the falling edge of LCLK0 (for all outputs and for $\overline{\text{LGTA}}$ and LUPWAIT inputs) or the rising edge of LCLK0 (for all other inputs).
3. All signals are measured from $OV_{DD}/2$ of the rising/falling edge of LCLK0 to $0.4 \times OV_{DD}$ of the signal in question for 3.3 V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is not set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.the
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is not set and when the load on the LALE output pin equals to the load on the LAD output pins.
8. For purposes of active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
9. DLL bypass mode is not recommended for use at frequencies above 66 MHz.

Figure 16 provides the AC test load for the local bus.

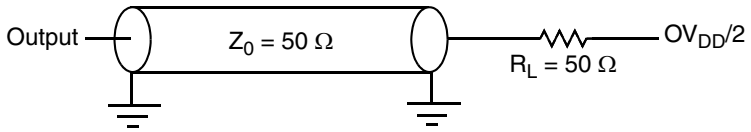

Figure 16. Local Bus C Test Load

Figure 17 through Figure 22 show the local bus signals.

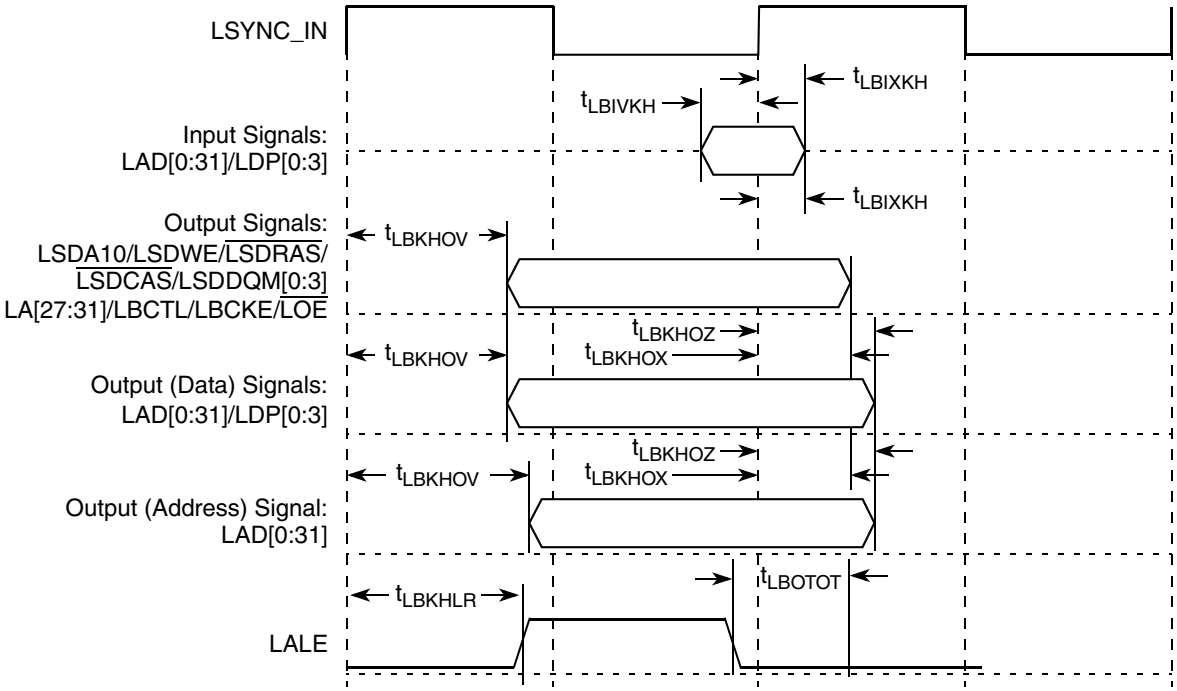

Figure 17. Local Bus Signals, Nonspecial Signals Only (DLL Enabled)

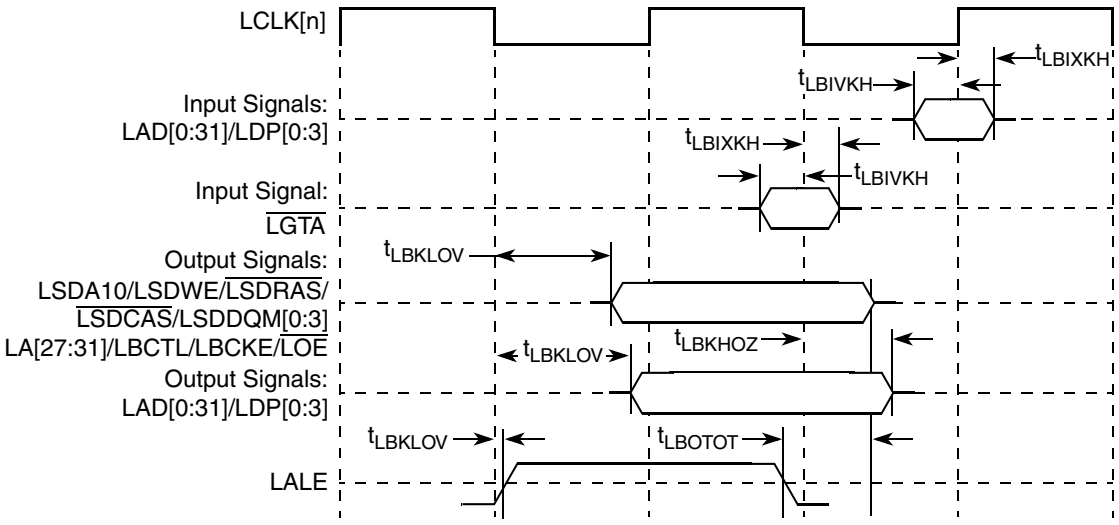

Figure 18. Local Bus Signals, Nonspecial Signals Only (DLL Bypass Mode)

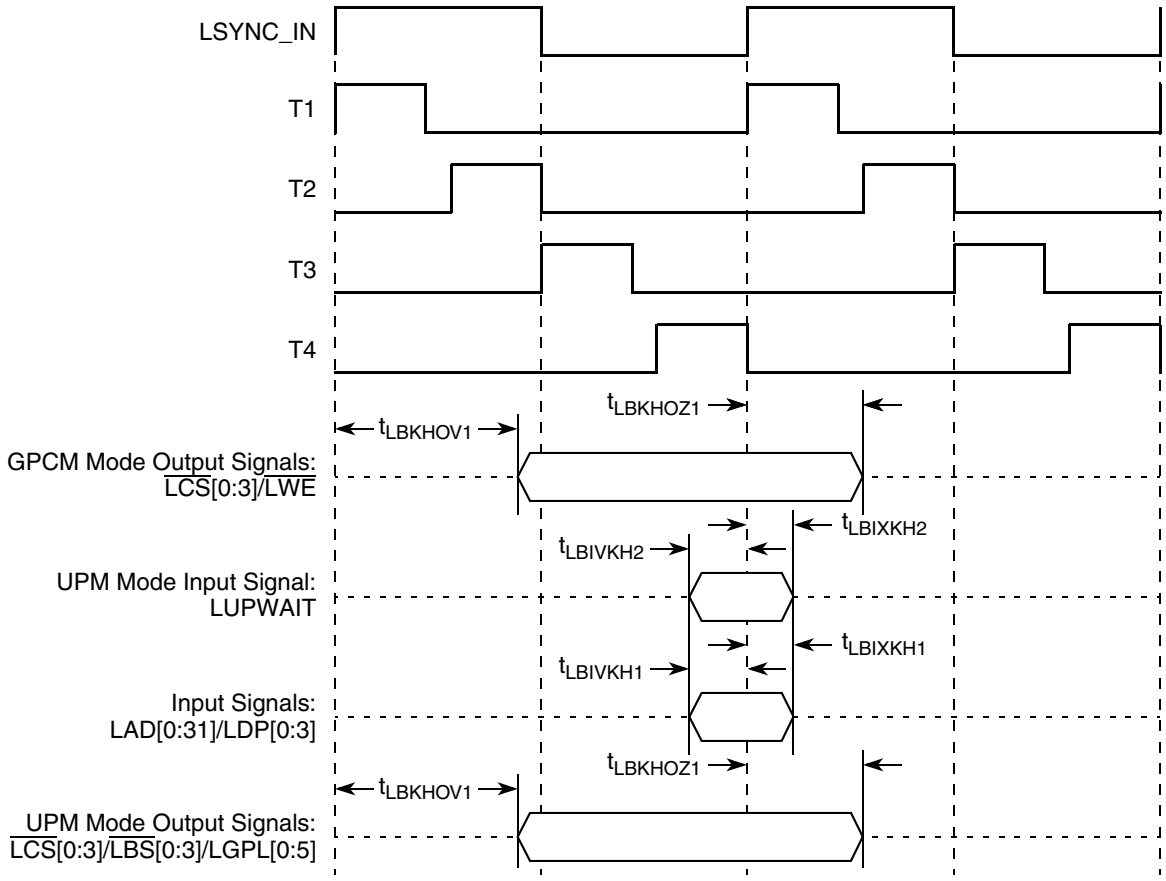

Figure 22. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (DLL Enabled)

11 JTAG

This section describes the DC and AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8343EA.

11.1 JTAG DC Electrical Characteristics

Table 36 provides the DC electrical characteristics for the IEEE Std. 1149.1 (JTAG) interface of the MPC8343EA.

Table 36. JTAG Interface DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	OV_{DD} – 0.3	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V

Table 36. JTAG Interface DC Electrical Characteristics (continued)

Parameter	Symbol	Condition	Min	Max	Unit
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

11.2 JTAG AC Timing Specifications

This section describes the AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8343EA. Table 37 provides the JTAG AC timing specifications as defined in Figure 24 through Figure 27.

Table 37. JTAG AC Timing Specifications (Independent of CLKIN)[1]

At recommended operating conditions (see Table 2).

Parameter	Symbol[2]	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	15	—	ns	—
JTAG external clock rise and fall times	t_{JTGR}, t_{JTGF}	0	2	ns	—
$\overline{\text{TRST}}$ assert time	t_{TRST}	25	—	ns	3
Input setup times: Boundary-scan data TMS, TDI	 t_{JTDVKH} t_{JTIVKH}	 4 4	 — —	ns	4
Input hold times: Boundary-scan data TMS, TDI	 t_{JTDXKH} t_{JTIXKH}	 10 10	 — —	ns	4
Valid times: Boundary-scan data TDO	 t_{JTKLDV} t_{JTKLOV}	 2 2	 11 11	ns	5
Output hold times: Boundary-scan data TDO	 t_{JTKLDX} t_{JTKLOX}	 2 2	 — —	ns	5

Table 39. I²C AC Electrical Specifications (continued)

Parameter	Symbol[1]	Min	Max	Unit
Fall time of both SDA and SCL signals[5]	t_{I2CF}	—	300	ns
Setup time for STOP condition	t_{I2PVKH}	0.6	—	μs
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	μs
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	$0.1 \times OV_{DD}$	—	V
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	$0.2 \times OV_{DD}$	—	V

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I²C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I²C timing (I2) for the time that the data with respect to the start condition (S) goes invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I²C timing (I2) for the time that the data with respect to the stop condition (P) reaches the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The device provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IH}(min) of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DVKH} must be met only if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.
5.)The device does not follow the “I2C-BUS Specifications” version 2.1 regarding the t_{I2CF} AC parameter.

Figure 28 provides the AC test load for the I²C.

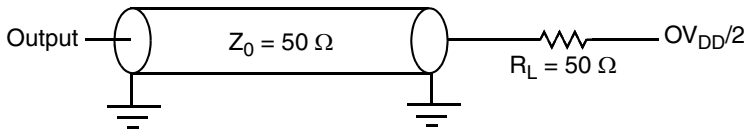

Figure 28. I²C AC Test Load

Figure 29 shows the AC timing diagram for the I²C bus.

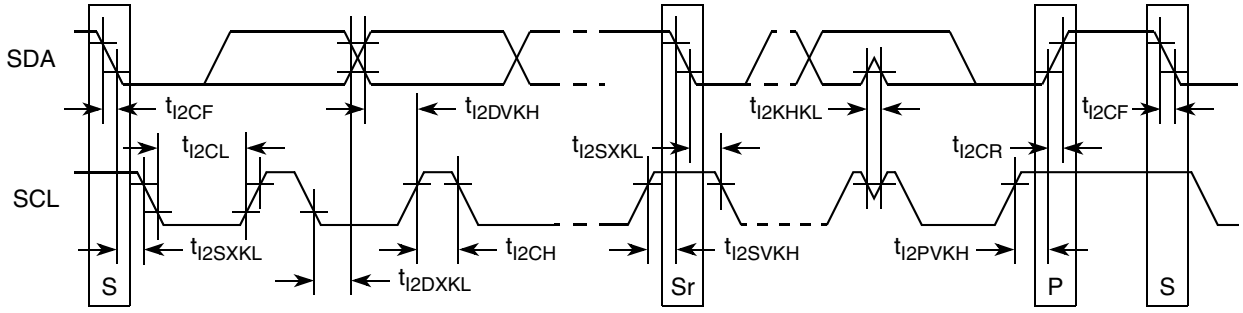

Figure 29. I²C Bus AC Timing Diagram

Table 49. SPI DC Electrical Characteristics (continued)

Parameter	Symbol	Condition	Min	Max	Unit
Input current	I_{IN}	—	—	±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

17.2 SPI AC Timing Specifications

Table 50 provides the SPI input and output AC timing specifications.

Table 50. SPI AC Timing Specifications[1]

Parameter	Symbol[2]	Min	Max	Unit
SPI outputs valid—Master mode (internal clock) delay	t_{NIKHOV}	—	6	ns
SPI outputs hold—Master mode (internal clock) delay	t_{NIKHOX}	0.5	—	ns
SPI outputs valid—Slave mode (external clock) delay	t_{NEKHOV}	—	8	ns
SPI outputs hold—Slave mode (external clock) delay	t_{NEKHOX}	2	—	ns
SPI inputs—Master mode (internal clock input setup time	t_{NIIVKH}	4	—	ns
SPI inputs—Master mode (internal clock input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns

Notes:
1. Output specifications are measured from the 50 percent level of the rising edge of CLKIN to the 50 percent level of the signal. Timings are measured at the pin.
2. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOX} symbolizes the internal timing (NI) for the time SPICLK clock reference (K) goes to the high state (H) until outputs (O) are invalid (X).

Figure 33 provides the AC test load for the SPI.

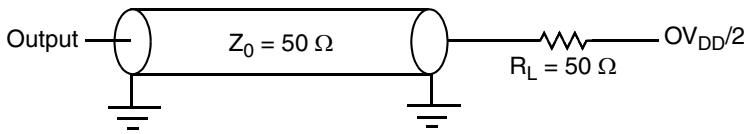

Figure 33. SPI AC Test Load

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
MECC[0:4]/MSRCID[0:4]	AG13, AE14, AH12, AH10, AE15	I/O	GV_{DD}	—
MECC[5]/MDVAL	AH14	I/O	GV_{DD}	—
MECC[6:7]	AE13, AH11	I/O	GV_{DD}	—
MDM[0:3]	AG28, AG24, AF20, AG17	O	GV_{DD}	—
MDM[8]	AG12	O	GV_{DD}	—
MDQS[0:3]	AE27, AE26, AE20, AH18	I/O	GV_{DD}	—
MDQS[8]	AH13	I/O	GV_{DD}	—
MBA[0:1]	AF10, AF11	O	GV_{DD}	—
MA[0:14]	AF13, AF15, AG16, AD16, AF17, AH20, AH19, AH21, AD18, AG21, AD13, AF21, AF22, AE1, AA5	O	GV_{DD}	—
$\overline{\text{MWE}}$	AD10	O	GV_{DD}	—
$\overline{\text{MRAS}}$	AF7	O	GV_{DD}	—
$\overline{\text{MCAS}}$	AG6	O	GV_{DD}	—
$\overline{\text{MCS}}$[0:3]	AE7, AH7, AH4, AF2	O	GV_{DD}	—
MCKE[0:1]	AG23, AH23	O	GV_{DD}	3
MCK[0:3]	AH15, AE24, AE2, AF14	O	GV_{DD}	—
$\overline{\text{MCK}}$[0:3]	AG15, AD23, AE3, AG14	O	GV_{DD}	—
MODT[0:3]	AG5, AD4, AH6, AF4	O	GV_{DD}	—
MBA[2]	AD22	O	GV_{DD}	—
MDIC0	AG11	I/O	—	9
MDIC1	AF12	I/O	—	9
Local Bus Controller Interface				
LAD[0:31]	T4, T5, T1, R2, R3, T2, R1, R4, P1, P2, P3, P4, N1, N4, N2, N3, M1, M2, M3, N5, M4, L1, L2, L3, K1, M5, K2, K3, J1, J2, L5, J3	I/O	OV_{DD}	—
LDP[0]/$\overline{\text{CKSTOP\_OUT}}$	H1	I/O	OV_{DD}	—
LDP[1]/$\overline{\text{CKSTOP\_IN}}$	K5	I/O	OV_{DD}	—
LDP[2]/$\overline{\text{LCS}}$[4]	H2	I/O	OV_{DD}	—
LDP[3]/$\overline{\text{LCS}}$[5]	G1	I/O	OV_{DD}	—
LA[27:31]	J4, H3, G2, F1, G3	O	OV_{DD}	—
$\overline{\text{LCS}}$[0:3]	J5, H4, F2, E1	O	OV_{DD}	—
$\overline{\text{LWE}}$[0:3]/LSDDQM[0:3]/$\overline{\text{LBS}}$[0:3]	F3, G4, D1, E2	O	OV_{DD}	—

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
TSEC1_RX_DV	U24	I	LV_{DD1}	—
TSEC1_RX_ER/GPIO2[26]	L28	I/O	OV_{DD}	—
TSEC1_RXD[3:0]	W26, W24, Y28, Y27	I	LV_{DD1}	—
TSEC1_TX_CLK	N25	I	OV_{DD}	—
TSEC1_TXD[3:0]	V28, V27, V26, W28	O	LV_{DD1}	10
TSEC1_TX_EN	W27	O	LV_{DD1}	—
TSEC1_TX_ER/GPIO2[31]	N24	I/O	OV_{DD}	—
Three-Speed Ethernet Controller (Gigabit Ethernet 2)				
TSEC2_COL/GPIO1[21]	P28	I/O	OV_{DD}	—
TSEC2_CRS/GPIO1[22]	AC28	I/O	LV_{DD2}	—
TSEC2_GTX_CLK	AC27	O	LV_{DD2}	—
TSEC2_RX_CLK	AB25	I	LV_{DD2}	—
TSEC2_RX_DV/GPIO1[23]	AC26	I/O	LV_{DD2}	—
TSEC2_RXD[3:0]/GPIO1[13:16]	AA25, AA26, AA27, AA28	I/O	LV_{DD2}	—
TSEC2_RX_ER/GPIO1[25]	R25	I/O	OV_{DD}	—
TSEC2_TXD[3:0]/GPIO1[17:20]	AB26, AB27, AA24, AB28	I/O	LV_{DD2}	—
TSEC2_TX_ER/GPIO1[24]	R27	I/O	OV_{DD}	—
TSEC2_TX_EN/GPIO1[12]	AD28	I/O	LV_{DD2}	3
TSEC2_TX_CLK/GPIO1[30]	R26	I/O	OV_{DD}	—
DUART				
UART_SOUT[1:2]/MSRCID[0:1]/ LSRCID[0:1]	B4, A4	O	OV_{DD}	—
UART_SIN[1:2]/MSRCID[2:3]/ LSRCID[2:3]	D5, C5	I/O	OV_{DD}	—
$\overline{\text{UART\_CTS}}$[1]/MSRCID4/LSRCID4	B5	I/O	OV_{DD}	—
$\overline{\text{UART\_CTS}}$[2]/MDVAL/LDVAL	A5	I/O	OV_{DD}	—
$\overline{\text{UART\_RTS}}$[1:2]	D6, C6	O	OV_{DD}	—
I^2C interface				
IIC1_SDA	E5	I/O	OV_{DD}	2
IIC1_SCL	A6	I/O	OV_{DD}	2
IIC2_SDA	B6	I/O	OV_{DD}	2
IIC2_SCL	E7	I/O	OV_{DD}	2

Table 51. MPC8343EA (PBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
OV_{DD}	B27, D3, D11, D19, E15, E23, F5, F8, F11, F14, F17, F20, G24, H23, H24, J6, J14, J17, J18, K4, L9, L20, L23, L25, M6, M9, M20, P5, P20, P23, R6, R9, R24, U23, V4, V6	PCI, 10/100 Ethernet, and other standard (3.3 V)	OV_{DD}	—
MVREF1	AF19	I	DDR reference voltage	—
MVREF2	AE10	I	DDR reference voltage	—
No Connection				
NC	A22, A23, A24, B22, B23, B24, C21, C22, C23, C24, D21, D22, D23, D24, E21, M27, M28, N26, N27, N28, P25, P26, P27, R28, T24, T25, T26, T27, T28, U27, U28, Y3, Y4, Y5, AA1, AA2, AA3, AA4, AB1, AB2, AB3, AB4, AC1, AC2, AC3, AC4, AD1, AD2, AD3, AD5, AD7, AD11, AD12, AE4, AE6, AE8, AE9, AE23, AF1, AF5, AF6, AF8, AF24, AG1, AG3, AG4, AG7, AG8, AG9, AG10, AH2, AH3, AH5, AH8, AH9, V5, V2, V1	—	—	—

Notes:

1. This pin is an open-drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to OV_{DD}.
2. This pin is an open-drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to OV_{DD}.
3. During reset, this output is actively driven rather than three-stated.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull-up if the chip is in PCI host mode. Follow the PCI specifications.
6. This pin must be always be tied to GND.
7. This pin must always be pulled up to OV_{DD}.
8. Thermal sensitive resistor.
9. It is recommended that MDIC0 be tied to GND using an 18.2 Ω resistor and MDIC1 be tied to DDR power using an 18.2 Ω resistor.
10. TSEC1_TXD[3] is required an external pull-up resistor. For proper functionality of the device, this pin must be pulled up or actively driven high during a hard reset. No external pull-down resistors are allowed to be attached to this net.
11. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to LV_{DD1}.
12. For systems that boot from local bus (GPCM)-controlled NOR flash, a pull up on LGPL4 is required.

As shown in Figure 37, the primary clock input (frequency) is multiplied up by the system phase-locked loop (PLL) and the clock unit to create the coherent system bus clock (*csb_clk*), the internal clock for the DDR controller (*ddr_clk*), and the internal clock for the local bus interface unit (*lbiu_clk*).

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$csb\_clk = \{PCI\_SYNC\_IN \times (1 + CFG\_CLKIN\_DIV)\} \times SPMF$$

In PCI host mode, PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the CLKIN frequency.

The *csb_clk* serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies the *csb_clk* frequency to create the internal clock for the e300 core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL), which is loaded at power-on reset or by one of the hard-coded reset options. See the chapter on reset, clocking, and initialization in the *MPC8349EA Reference Manual* for more information on the clock subsystem.

The internal *ddr_clk* frequency is determined by the following equation:

$$ddr\_clk = csb\_clk \times (1 + RCWL[DDRCM])$$

ddr_clk is not the external memory bus frequency; *ddr_clk* passes through the DDR clock divider (÷2) to create the differential DDR memory bus clock outputs (MCK and $\overline{\text{MCK}}$). However, the data rate is the same frequency as *ddr_clk*.

The internal *lbiu_clk* frequency is determined by the following equation:

$$lbiu\_clk = csb\_clk \times (1 + RCWL[LBIUCM])$$

lbiu_clk is not the external local bus frequency; *lbiu_clk* passes through the LBIU clock divider to create the external local bus clock outputs (LSYNC_OUT and LCLK[0:2]). The LBIU clock divider ratio is controlled by LCCR[CLKDIV].

In addition, some of the internal units may have to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory-mapped register after the device exits reset. Table 52 specifies which units have a configurable clock frequency.

Table 52. Configurable Clock Units

Unit	Default Frequency	Options
TSEC1	*csb_clk/3*	Off, *csb_clk, csb_clk/2, csb_clk/3*
TSEC2, I^2C1	*csb_clk/3*	Off, *csb_clk, csb_clk/2, csb_clk/3*
Security core	*csb_clk/3*	Off, *csb_clk,* csb_clk/2, *csb_clk/3*
USB DR, USB MPH	*csb_clk/3*	Off, csb_clk, csb_clk/2, *csb_clk/3*
PCI and DMA complex	*csb_clk*	Off, *csb_clk*

All frequency combinations shown in the table below may not be available. Maximum operating frequencies depend on the part ordered, see Section 22.1, “Part Numbers Fully Addressed by This Document,” for part ordering details and contact your Freescale Sales Representative or authorized distributor for more information.

The junction-to-ambient thermal resistance is an industry-standard value that provides a quick and easy estimation of thermal performance. Generally, the value obtained on a single-layer board is appropriate for a tightly packed printed-circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

20.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

The thermal performance of a device cannot be adequately predicted from the junction-to-ambient thermal resistance. The thermal performance of any component is strongly dependent on the power dissipation of surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)

T_A = ambient temperature for the package (°C)

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

20.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature and a measure of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)

T_T = thermocouple temperature on top of package (°C)

Ψ_{JT} = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per the JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so

that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

20.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

Some application environments require a heat sink to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)
$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)
$R_{\theta CA}$ = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ is device-related and cannot be influenced by the user. The user controls the thermal environment to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the size of the heat sink, the air flow around the device, the interface material, the mounting arrangement on printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device.

The thermal performance of devices with heat sinks has been simulated with a few commercially available heat sinks. The heat sink choice is determined by the application environment (temperature, air flow, adjacent component power dissipation) and the physical space available. Because there is not a standard application environment, a standard heat sink is not required.

Table 60 shows heat sink thermal resistance for PBGA of the MPC8343EA.

Table 60. Heat Sink and Thermal Resistance of MPC8343EA (PBGA)

Heat Sink Assuming Thermal Grease	Air Flow	29 × 29 mm PBGA
		Thermal Resistance
AAVID 30 × 30 × 9.4 mm pin fin	Natural convection	13.5
AAVID 30 × 30 × 9.4 mm pin fin	1 m/s	9.6
AAVID 30 × 30 × 9.4 mm pin fin	2 m/s	8.8
AAVID 31 × 35 × 23 mm pin fin	Natural convection	11.3
AAVID 31 × 35 × 23 mm pin fin	1 m/s	8.1
AAVID 31 × 35 × 23 mm pin fin	2 m/s	7.5
Wakefield, 53 × 53 × 25 mm pin fin	Natural convection	9.1
Wakefield, 53 × 53 × 25 mm pin fin	1 m/s	7.1
Wakefield, 53 × 53 × 25 mm pin fin	2 m/s	6.5
MEI, 75 × 85 × 12 no adjacent board, extrusion	Natural convection	10.1

Interface material vendors include the following:

Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01801 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Dow-Corning Electronic Materials P.O. Box 994 Midland, MI 48686-0997 Internet: www.dowcorning.com	800-248-2481
Shin-Etsu MicroSi, Inc. 10028 S. 51st St. Phoenix, AZ 85044 Internet: www.microsi.com	888-642-7674
The Bergquist Company 18930 West 78th St. Chanhassen, MN 55317 Internet: www.bergquistcompany.com	800-347-4572

20.3 Heat Sink Attachment

When heat sinks are attached, an interface material is required, preferably thermal grease and a spring clip. The spring clip should connect to the printed-circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces that can lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. The recommended maximum force on the top of the package is 10 lb force (4.5 kg force). Any adhesive attachment should attach to painted or plastic surfaces, and its performance should be verified under the application requirements.

20.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When a heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimize the size of the clearance to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (°C)

T_C = case temperature of the package (°C)

$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)
P_D = power dissipation (W)

21 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8343EA.

21.1 System Clocking

The MPC8343EA includes two PLLs:

1. The platform PLL generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 19.1, “System PLL Configuration.”
2. The e300 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 19.2, “Core PLL Configuration.”

21.2 PLL Power Supply Filtering

Each PLL gets power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always equal to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme.

There are a number of ways to provide power reliably to the PLLs, but the recommended solution is to provide four independent filter circuits as illustrated in Figure 38, one to each of the four AV_{DD} pins. Independent filters to each PLL reduce the opportunity to cause noise injection from one PLL to the other.

The circuit filters noise in the PLL resonant frequency range from 500 kHz to 10 MHz. It should be built with surface mount capacitors with minimum effective series inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

To minimize noise coupled from nearby circuits, each circuit should be placed as closely as possible to the specific AV_{DD} pin being supplied. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

Figure 38 shows the PLL power supply filter circuit.

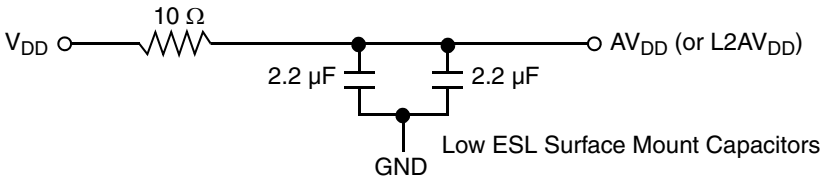

Figure 38. PLL Power Supply Filter Circuit