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### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	54
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	768 x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	64-TQFP (10x10)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/pic18lf6410t-i-pt">https://www.e-xfl.com/product-detail/microchip-technology/pic18lf6410t-i-pt</a>

# PIC18F6310/6410/8310/8410

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**TABLE 11-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	66
LATC	LATC Output Latch Register								66
TRISC	PORTC Data Direction Register								66

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## 17.3.8 SLEEP OPERATION

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode; in the case of the Sleep mode, all clocks are halted.

In most power-managed modes, a clock is provided to the peripherals. That clock should be from the primary clock source, the secondary clock (Timer1 oscillator at 32.768 kHz) or the INTOSC source. See **Section 3.7 “Clock Sources and Oscillator Switching”** for additional information.

In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.

If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.

If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device.

## 17.3.9 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

## 17.3.10 BUS MODE COMPATIBILITY

Table 17-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

**TABLE 17-1: SPI BUS MODES**

Standard SPI Mode Terminology	Control Bits State	
	CKP	CKE
0, 0	0	1
0, 1	0	0
1, 0	1	1
1, 1	1	0

There is also an SMP bit which controls when the data is sampled.

**TABLE 17-2: REGISTERS ASSOCIATED WITH SPI OPERATION**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	63
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSPIF	CCP1IF	TMR2IF	TMR1IF	65
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSPIE	CCP1IE	TMR2IE	TMR1IE	65
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSPIP	CCP1IP	TMR2IP	TMR1IP	65
TRISC	PORTC Data Direction Register								66
TRISF	PORTF Data Direction Register								66
SSPBUF	Master Synchronous Serial Port Receive Buffer/Transmit Register								64
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	64
SSPSTAT	SMP	CKE	D/A	P	S	R/W	UA	BF	64

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the MSSP in SPI mode.



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**TABLE 17-4: REGISTERS ASSOCIATED WITH I<sup>2</sup>C™ OPERATION**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	63
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSPIF	CCP1IF	TMR2IF	TMR1IF	65
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSPIE	CCP1IE	TMR2IE	TMR1IE	65
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSPIP	CCP1IP	TMR2IP	TMR1IP	65
TRISC	PORTC Data Direction Register								66
SSPBUF	Master Synchronous Serial Port Receive Buffer/Transmit Register								64
SSPADD	Master Synchronous Serial Port Receive Buffer/Transmit Register								64
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	64
SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	64
SSPSTAT	SMP	CKE	D/A	P	S	R/W	UA	BF	64

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the MSSP in I<sup>2</sup>C mode.

## 18.2.5 BREAK CHARACTER SEQUENCE

The Enhanced USART module has the capability of sending the special Break character sequences that are required by the LIN/J2602 bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The Frame Break character is sent whenever the SENDB and TXEN bits (TXSTA<3> and TXSTA<5>) are set while the Transmit Shift register is loaded with data. Note that the value of data written to TXREG1 will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN/J2602 specification).

Note that the data value written to the TXREG1 for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 18-10 for the timing of the Break character sequence.

### 18.2.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN/J2602 bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to set up the Break character.

3. Load the TXREG1 with a dummy character to initiate transmission (the value is ignored).
4. Write '55h' to TXREG1 to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREG1 becomes empty, as indicated by the TX1IF bit, the next data byte can be written to TXREG1.

## 18.2.6 RECEIVING A BREAK CHARACTER

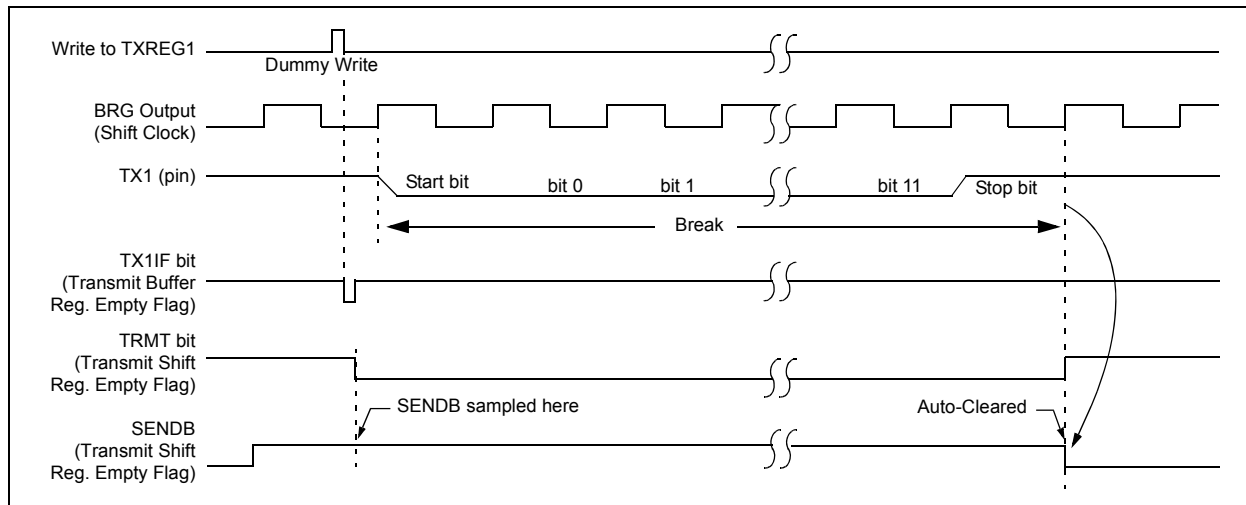
The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in **Section 18.2.4 "Auto-Wake-up on Sync Break Character"**. By enabling this feature, the EUSART will sample the next two transitions on RX1/DT1, cause an RC1IF interrupt and receive the next data byte followed by another interrupt.

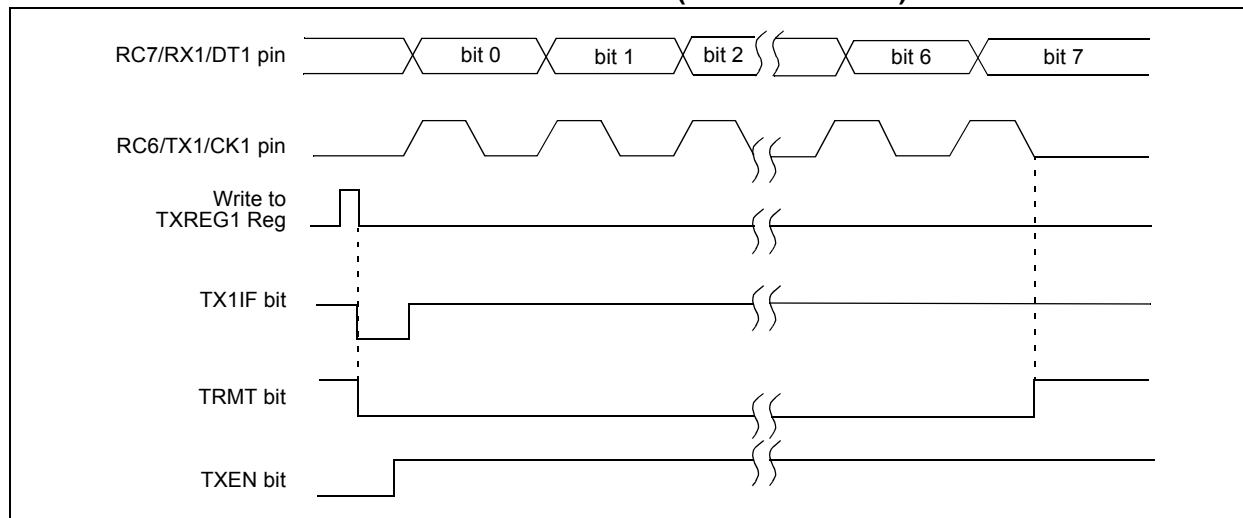
Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABD bit once the TX1IF interrupt is observed.

**FIGURE 18-10: SEND BREAK CHARACTER SEQUENCE**



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**FIGURE 18-12: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)**



**TABLE 18-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	63
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSPIF	CCP1IF	TMR2IF	TMR1IF	65
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSPIE	CCP1IE	TMR2IE	TMR1IE	65
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSPIP	CCP1IP	TMR2IP	TMR1IP	65
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	65
TXREG1	EUSART1 Transmit Register								65
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	65
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	66
SPBRGH1	EUSART1 Baud Rate Generator Register High Byte								66
SPBRG1	EUSART1 Baud Rate Generator Register Low Byte								65

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.



## 19.0 ADDRESSABLE UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (AUSART)

The Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART) module is very similar in function to the Enhanced USART module, discussed in the previous chapter. It is provided as an additional channel for serial communication with external devices, for those situations that do not require Auto-Baud Detection (ABD) or LIN/J2602 bus support.

The AUSART can be configured in the following modes:

- Asynchronous (full-duplex)
- Synchronous – Master (half-duplex)
- Synchronous – Slave (half-duplex)

The pins of the AUSART module are multiplexed with the functions of PORTG (RG1/TX2/CK2 and RG2/RX2/DT2, respectively). In order to configure these pins as an AUSART:

- SPEN bit (RCSTA2<7>) must be set (= 1)
- TRISG<2> bit must be set (= 1)
- TRISG<1> bit must be cleared (= 0) for Asynchronous and Synchronous Master modes
- TRISG<1> bit must be set (= 1) for Synchronous Slave mode

<b>Note:</b> The USART control will automatically reconfigure the pin from input to output as needed.
---

The operation of the Addressable USART module is controlled through two registers: TXSTA2 and RXSTA2. These are detailed in Register 19-1 and Register 19-2 respectively.

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## REGISTER 19-2: RCSTA2: AUSART2 RECEIVE STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
bit 7							bit 0

### Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

- bit 7      **SPEN:** Serial Port Enable bit  
1 = Serial port is enabled (configures RXx/DTx and TXx/CKx pins as serial port pins)  
0 = Serial port is disabled (held in Reset)
- bit 6      **RX9:** 9-Bit Receive Enable bit  
1 = Selects 9-bit reception  
0 = Selects 8-bit reception
- bit 5      **SREN:** Single Receive Enable bit  
Asynchronous mode:  
Don't care.  
Synchronous mode – Master:  
1 = Enables single receive  
0 = Disables single receive  
This bit is cleared after reception is complete.  
Synchronous mode – Slave:  
Don't care.
- bit 4      **CREN:** Continuous Receive Enable bit  
Asynchronous mode:  
1 = Enables receiver  
0 = Disables receiver  
Synchronous mode:  
1 = Enables continuous receive until enable bit, CREN, is cleared (CREN overrides SREN)  
0 = Disables continuous receive
- bit 3      **ADDEN:** Address Detect Enable bit  
Asynchronous mode 9-Bit (RX9 = 1):  
1 = Enables address detection, enables interrupt and loads the receive buffer when RSR<8> are set  
0 = Disables address detection, all bytes are received and ninth bit can be used as a parity bit  
Asynchronous mode 9-Bit (RX9 = 0):  
Don't care.
- bit 2      **FERR:** Framing Error bit  
1 = Framing error (can be updated by reading RCREG1 register and receiving next valid byte)  
0 = No framing error
- bit 1      **OERR:** Overrun Error bit  
1 = Overrun error (can be cleared by clearing bit, CREN)  
0 = No overrun error
- bit 0      **RX9D:** 9th bit of Received Data bit  
This can be address/data bit or a parity bit and must be calculated by user firmware.

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## REGISTER 20-3: ADCON2: A/D CONTROL REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0

bit 7

bit 0

### Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

bit 7 **ADFM:** A/D Result Format Select bit

1 = Right justified

0 = Left justified

bit 6 **Unimplemented:** Read as '0'

bit 5-3 **ACQT<2:0>:** A/D Acquisition Time Select bits

111 = 20 TAD

110 = 16 TAD

101 = 12 TAD

100 = 8 TAD

011 = 6 TAD

010 = 4 TAD

001 = 2 TAD

000 = 0 TAD<sup>(1)</sup>

bit 2-0 **ADCS<2:0>:** A/D Conversion Clock Select bits

111 = FRC (clock derived from A/D RC oscillator)<sup>(1)</sup>

110 = FOSC/64

101 = FOSC/16

100 = FOSC/4

011 = FRC (clock derived from A/D RC oscillator)<sup>(1)</sup>

010 = FOSC/32

001 = FOSC/8

000 = FOSC/2

**Note 1:** If the A/D FRC clock source is selected, a delay of one T<sub>CY</sub> (instruction cycle) is added before the A/D clock starts. This allows the **SLEEP** instruction to be executed before starting a conversion.

# PIC18F6310/6410/8310/8410

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NOTES:

## 23.6 Operation During Sleep

When enabled, the HLVD circuitry continues to operate during Sleep. If the device voltage crosses the trip point, the HLVDIF bit will be set and the device will wake-up from Sleep. Device execution will continue from the interrupt vector address if interrupts have been globally enabled.

## 23.7 Effects of a Reset

A device Reset forces all registers to their Reset state. This forces the HLVD module to be turned off.

**TABLE 23-1: REGISTERS ASSOCIATED WITH HIGH/LOW-VOLTAGE DETECT MODULE**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
HLVDCON	VDIRMAG	—	IRVST	HLVDEN	HLVDL3	HLVDL2	HLVDL1	HLVDL0	64
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	63
PIR2	OSCFIF	CMIF	—	—	BCLIF	HLVDIF	TMR3IF	CCP2IF	65
PIE2	OCSFIE	CMIE	—	—	BCLIE	HLVDIE	TMR3IE	CCP2IE	65
IPR2	OSCFIP	CMIP	—	—	BCLIP	HLVDIP	TMR3IP	CCP2IP	65

**Legend:** — = unimplemented, read as '0'. Shaded cells are unused by the HLVD module.

## 24.3 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period from oscillator start-up to code execution by allowing the microcontroller to use the INTRC oscillator as a clock source until the primary clock source is available. It is enabled by setting the IESO Configuration bit.

Two-Speed Start-up should be enabled only if the primary oscillator mode is LP, XT, HS or HSPLL (Crystal-Based modes). Other sources do not require a OST start-up delay; for these, Two-Speed Start-up should be disabled.

When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a Power-on Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI\_RUN mode.

To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits, IRCF<2:0>, immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting the IRCF<2:0> bits prior to entering Sleep mode.

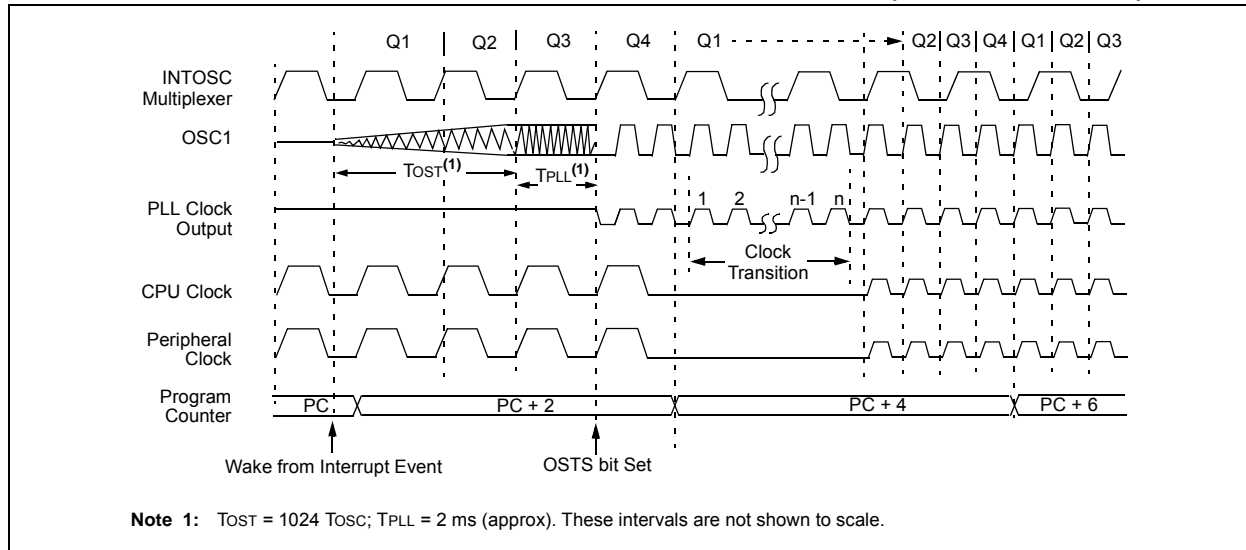
In all other power-managed modes, Two-Speed Start-up is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

### 24.3.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTRC oscillator in Two-Speed Start-up, the device still obeys the normal command sequences for entering power-managed modes, including serial `SLEEP` instructions (refer to **Section 4.1.2 “Entering Power-Managed Modes”**). In practice, this means that user code can change the SCS<1:0> bits setting or issue `SLEEP` instructions before the OST times out. This would allow an application to briefly wake-up, perform routine “housekeeping” tasks and return to Sleep before the device starts to operate from the primary oscillator.

User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit (OSCCON<3>). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

**FIGURE 24-2: TIMING TRANSITION FOR TWO-SPEED START-UP (INTOSC TO HSPLL)**





# PIC18F6310/6410/8310/8410

## BCF Bit Clear f

**Syntax:** BCF f, b {,a}

**Operands:**  $0 \leq f \leq 255$   
 $0 \leq b \leq 7$   
 $a \in [0,1]$

**Operation:**  $0 \rightarrow f \leftarrow b$

**Status Affected:** None

**Encoding:**

1001	bbba	ffff	ffff
------	------	------	------

**Description:** Bit 'b' in register 'f' is cleared.  
 If 'a' is '0', the Access Bank is selected.  
 If 'a' is '1', the BSR is used to select the GPR bank.  
 If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever  $f \leq 95$  (5Fh). See **Section 25.2.3** for details.

**Words:** 1

**Cycles:** 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

**Example:** BCF FLAG\_REG, 7, 0

Before Instruction  
 FLAG\_REG = C7h  
 After Instruction  
 FLAG\_REG = 47h

## BN Branch if Negative

**Syntax:** BN n

**Operands:**  $-128 \leq n \leq 127$

**Operation:** if Negative bit is '1',  
 $(PC) + 2 + 2n \rightarrow PC$

**Status Affected:** None

**Encoding:**

1110	0110	nnnn	nnnn
------	------	------	------

**Description:** If the Negative bit is '1', then the program will branch.  
 The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be  $PC + 2 + 2n$ . This instruction is then a two-cycle instruction.

**Words:** 1

**Cycles:** 1(2)

Q Cycle Activity:

If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

**Example:** HERE BN Jump

Before Instruction  
 PC = address (HERE)  
 After Instruction  
 If Negative = 1;  
 PC = address (Jump)  
 If Negative = 0;  
 PC = address (HERE + 2)



# PIC18F6310/6410/8310/8410

## LFSR Load FSR

Syntax: LFSR f, k

Operands:  $0 \leq f \leq 2$   
 $0 \leq k \leq 4095$

Operation:  $k \rightarrow \text{FSRf}$

Status Affected: None

Encoding:

1110	1110	00ff	$k_{11}kkk$
1111	0000	$k_7kkk$	$kkkk$

Description: The 12-bit literal 'k' is loaded into the file select register pointed to by 'f'.

Words: 2

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k' MSB	Process Data	Write literal 'k' MSB to FSRfH
Decode	Read literal 'k' LSB	Process Data	Write literal 'k' to FSRfL

**Example:** LFSR 2, 3ABh

After Instruction

FSR2H = 03h  
 FSR2L = ABh

## MOVF Move f

Syntax: MOVF f {,d {,a}}

Operands:  $0 \leq f \leq 255$   
 $d \in [0,1]$   
 $a \in [0,1]$

Operation:  $f \rightarrow \text{dest}$

Status Affected: N, Z

Encoding:

0101	00da	ffff	ffff
------	------	------	------

Description: The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'. Location 'f' can be anywhere in the 256-byte bank. If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank. If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever  $f \leq 95$  (5Fh). See **Section 25.2.3** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write W

**Example:** MOVF REG, 0, 0

Before Instruction

REG = 22h  
 W = FFh

After Instruction

REG = 22h  
 W = 22h

# PIC18F6310/6410/8310/8410

## CALLW Subroutine Call Using WREG

Syntax:	CALLW				
Operands:	None				
Operation:	(PC + 2) → TOS, (W) → PCL, (PCLATH) → PCH, (PCLATU) → PCU				
Status Affected:	None				
Encoding:	<table border="1"><tr><td>0000</td><td>0000</td><td>0001</td><td>0100</td></tr></table>	0000	0000	0001	0100
0000	0000	0001	0100		
Description:	<p>First, the return address (PC + 2) is pushed onto the return stack. Next, the contents of W are written to PCL; the existing value is discarded. Then, the contents of PCLATH and PCLATU are latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while the new next instruction is fetched. Unlike CALL, there is no option to update W, STATUS or BSR.</p>				
Words:	1				
Cycles:	2				
Q Cycle Activity:					

Q1	Q2	Q3	Q4
Decode	Read WREG	Push PC to stack	No operation
No operation	No operation	No operation	No operation

**Example:**                    HERE        CALLW

Before Instruction

PC        =    address (HERE)  
PCLATH =    10h  
PCLATU =    00h  
W        =    06h

After Instruction

PC        =    001006h  
TOS       =    address (HERE + 2)  
PCLATH =    10h  
PCLATU =    00h  
W        =    06h

## MOVSF Move Indexed to f

Syntax:	MOVSF [z <sub>s</sub> ], f <sub>d</sub>			
Operands:	0 ≤ z <sub>s</sub> ≤ 127 0 ≤ f <sub>d</sub> ≤ 4095			
Operation:	((FSR2) + z <sub>s</sub> ) → f <sub>d</sub>			
Status Affected:	None			
Encoding:				
1st word (source)	1110	1011	0zzz	zzzz <sub>s</sub>
2nd word (destin.)	1111	ffff	ffff	ffff <sub>d</sub>
Description:	<p>The contents of the source register are moved to destination register 'f<sub>d</sub>'. The actual address of the source register is determined by adding the 7-bit literal offset 'z<sub>s</sub>' in the first word to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f<sub>d</sub>' in the second word. Both addresses can be anywhere in the 4096-byte data space (000h to FFFh).</p> <p>The <b>MOVSF</b> instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.</p> <p>If the resultant source address points to an indirect addressing register, the value returned will be 00h.</p>			

Q1	Q2	Q3	Q4
Decode	Determine source addr	Determine source addr	Read source reg
Decode	No operation No dummy read	No operation	Write register 'f' (dest)

**Example:**                    MOVSF    [05h], REG2

Before Instruction

FSR2        =    80h  
Contents of 85h =    33h  
REG2        =    11h

After Instruction

FSR2        =    80h  
Contents of 85h =    33h  
REG2        =    33h

## 25.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

**Note:** Enabling the PIC18 instruction set extension may cause legacy applications to behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset addressing (**Section 6.5.1 “Indexed Addressing with Literal Offset”**). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank ( $a = 0$ ) or in a GPR bank designated by the BSR ( $a = 1$ ). When the extended instruction set is enabled and  $a = 0$ , however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see **Section 25.2.3.1 “Extended Instruction Syntax with Standard PIC18 Commands”**).

Although the Indexed Literal Offset mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

### 25.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument ‘f’ in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value ‘k’. As already noted, this occurs only when f is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets (“[ ]”). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM Assembler.

If the index argument is properly bracketed for Indexed Literal Offset addressing, the Access RAM argument is never specified; it will automatically be assumed to be ‘0’. This is in contrast to standard operation (extended instruction set disabled), when ‘a’ is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM assembler.

The destination argument ‘d’ functions as before.

In the latest versions of the MPASM assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option `/Y`, or the PE directive in the source listing.

## 25.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F6310/6410/8310/8410, it is very important to consider the type of code. A large, re-entrant application that is written in C and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

## 26.0 DEVELOPMENT SUPPORT

The PIC® microcontrollers and dsPIC® digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
  - MPLAB® IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB C Compiler for Various Device Families
  - HI-TECH C for Various Device Families
  - MPASM™ Assembler
  - MPLINK™ Object Linker/  
MPLIB™ Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
  - MPLAB ICD 3
  - PICKit™ 3 Debug Express
- Device Programmers
  - PICKit™ 2 Programmer
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

## 26.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows® operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - In-Circuit Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
  - Source files (C or assembly)
  - Mixed C and assembly
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

## 26.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC® DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

## 26.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC® Flash MCUs and dsPIC® Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

## 26.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC® Flash microcontrollers and dsPIC® DSCs with the powerful, yet easy-to-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

## 26.10 PICkit 3 In-Circuit Debugger/Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC® and dsPIC® Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming™.

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.