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Details

Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	23
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 5x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf252t-i-sog

TABLE 3-3: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Register	Applicable Devices				Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
IPR2	242	442	252	452	---1 1111	---1 1111	---u uuuu
PIR2	242	442	252	452	---0 0000	---0 0000	---u uuuu ⁽¹⁾
PIE2	242	442	252	452	---0 0000	---0 0000	---u uuuu
IPR1	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
	242	442	252	452	-111 1111	-111 1111	-uuu uuuu
PIR1	242	442	252	452	0000 0000	0000 0000	uuuu uuuu ⁽¹⁾
	242	442	252	452	-000 0000	-000 0000	-uuu uuuu ⁽¹⁾
PIE1	242	442	252	452	0000 0000	0000 0000	uuuu uuuu
	242	442	252	452	-000 0000	-000 0000	-uuu uuuu
TRISE	242	442	252	452	0000 -111	0000 -111	uuuu -uuu
TRISD	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISC	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISB	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISA ^(5,6)	242	442	252	452	-111 1111 ⁽⁵⁾	-111 1111 ⁽⁵⁾	-uuu uuuu ⁽⁵⁾
LATE	242	442	252	452	---- -xxx	---- -uuu	---- -uuu
LATD	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATC	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATB	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATA ^(5,6)	242	442	252	452	-xxx xxxx ⁽⁵⁾	-uuu uuuu ⁽⁵⁾	-uuu uuuu ⁽⁵⁾
PORTE	242	442	252	452	---- -000	---- -000	---- -uuu
PORTD	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTC	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTB	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA ^(5,6)	242	442	252	452	-x0x 0000 ⁽⁵⁾	-u0u 0000 ⁽⁵⁾	-uuu uuuu ⁽⁵⁾

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition.

Shaded cells indicate conditions do not apply for the designated device.

Note 1: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

4: See Table 3-2 for RESET value for specific condition.

5: Bit 6 of PORTA, LATA, and TRISA are enabled in ECIO and RCIO Oscillator modes only. In all other Oscillator modes, they are disabled and read '0'.

6: Bit 6 of PORTA, LATA and TRISA are not available on all devices. When unimplemented, they are read '0'.

4.9 Data Memory Organization

The data memory is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4096 bytes of data memory. Figure 4-6 and Figure 4-7 show the data memory organization for the PIC18FXX2 devices.

The data memory map is divided into as many as 16 banks that contain 256 bytes each. The lower 4 bits of the Bank Select Register (BSR<3:0>) select which bank will be accessed. The upper 4 bits for the BSR are not implemented.

The data memory contains Special Function Registers (SFR) and General Purpose Registers (GPR). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratch pad operations in the user's application. The SFRs start at the last location of Bank 15 (0xFFF) and extend downwards. Any remaining space beyond the SFRs in the Bank may be implemented as GPRs. GPRs start at the first location of Bank 0 and grow upwards. Any read of an unimplemented location will read as '0's.

The entire data memory may be accessed directly or indirectly. Direct addressing may require the use of the BSR register. Indirect addressing requires the use of a File Select Register (FSRn) and a corresponding Indirect File Operand (INDFn). Each FSR holds a 12-bit address value that can be used to access any location in the Data Memory map without banking.

The instruction set and architecture allow operations across all banks. This may be accomplished by indirect addressing or by the use of the `MOVFF` instruction. The `MOVFF` instruction is a two-word/two-cycle instruction that moves a value from one register to another.

To ensure that commonly used registers (SFRs and select GPRs) can be accessed in a single cycle, regardless of the current BSR values, an Access Bank is implemented. A segment of Bank 0 and a segment of Bank 15 comprise the Access RAM. Section 4.10 provides a detailed description of the Access RAM.

4.9.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly or indirectly. Indirect addressing operates using a File Select Register and corresponding Indirect File Operand. The operation of indirect addressing is shown in Section 4.12.

Enhanced MCU devices may have banked memory in the GPR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other RESETS.

Data RAM is available for use as GPR registers by all instructions. The top half of Bank 15 (0xF80 to 0xFFFF) contains SFRs. All other banks of data memory contain GPR registers, starting with Bank 0.

4.9.2 SPECIAL FUNCTION REGISTERS

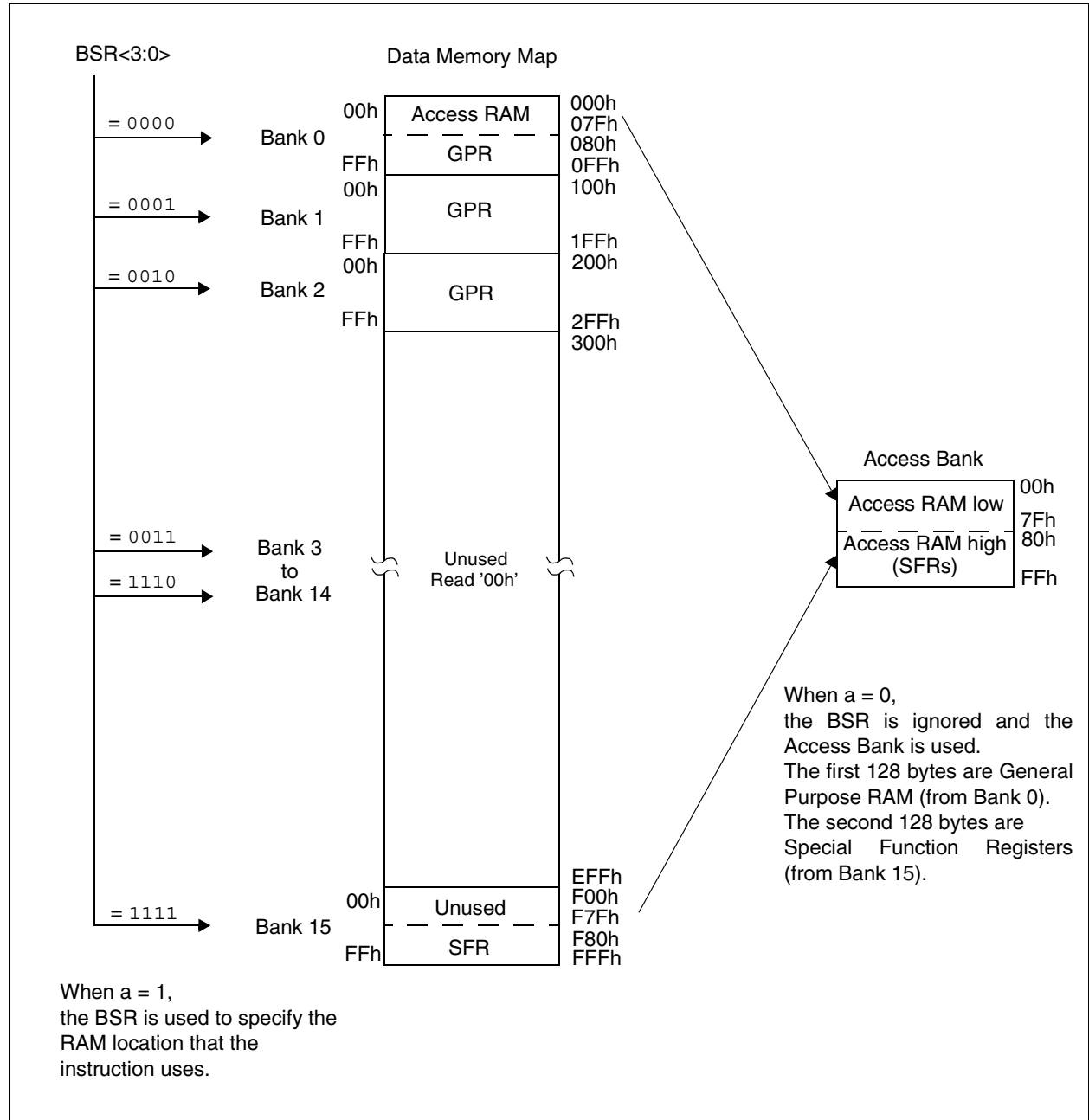
The Special Function Registers (SFRs) are registers used by the CPU and Peripheral Modules for controlling the desired operation of the device. These registers are implemented as static RAM. A list of these registers is given in Table 4-1 and Table 4-2.

The SFRs can be classified into two sets; those associated with the "core" function and those related to the peripheral functions. Those registers related to the "core" are described in this section, while those related to the operation of the peripheral features are described in the section of that peripheral feature.

The SFRs are typically distributed among the peripherals whose functions they control.

The unused SFR locations will be unimplemented and read as '0's. See Table 4-1 for addresses for the SFRs.

FIGURE 4-6: DATA MEMORY MAP FOR PIC18F242/442



4.12 Indirect Addressing, INDF and FSR Registers

Indirect addressing is a mode of addressing data memory, where the data memory address in the instruction is not fixed. An FSR register is used as a pointer to the data memory location that is to be read or written. Since this pointer is in RAM, the contents can be modified by the program. This can be useful for data tables in the data memory and for software stacks. Figure 4-9 shows the operation of indirect addressing. This shows the moving of the value to the data memory address specified by the value of the FSR register.

Indirect addressing is possible by using one of the INDF registers. Any instruction using the INDF register actually accesses the register pointed to by the File Select Register, FSR. Reading the INDF register itself, indirectly (FSR = 0), will read 00h. Writing to the INDF register indirectly, results in a no operation. The FSR register contains a 12-bit address, which is shown in Figure 4-10.

The INDFn register is not a physical register. Addressing INDFn actually addresses the register whose address is contained in the FSRn register (FSRn is a pointer). This is indirect addressing.

Example 4-4 shows a simple use of indirect addressing to clear the RAM in Bank1 (locations 100h-1FFh) in a minimum number of instructions.

EXAMPLE 4-4: HOW TO CLEAR RAM (BANK1) USING INDIRECT ADDRESSING

```
        LFSR   FSR0, 0x100 ;
NEXT    CLRF   POSTINC0    ; Clear INDF
                                ; register and
                                ; inc pointer
        BTFSS  FSR0H, 1    ; All done with
                                ; Bank1?
        GOTO   NEXT        ; NO, clear next
CONTINUE                                ; YES, continue
```

There are three indirect addressing registers. To address the entire data memory space (4096 bytes), these registers are 12-bit wide. To store the 12-bits of addressing information, two 8-bit registers are required. These indirect addressing registers are:

1. FSR0: composed of FSR0H:FSR0L
2. FSR1: composed of FSR1H:FSR1L
3. FSR2: composed of FSR2H:FSR2L

In addition, there are registers INDF0, INDF1 and INDF2, which are not physically implemented. Reading or writing to these registers activates indirect addressing, with the value in the corresponding FSR register being the address of the data. If an instruction writes a value to INDF0, the value will be written to the address pointed to by FSR0H:FSR0L. A read from INDF1 reads

the data from the address pointed to by FSR1H:FSR1L. INDFn can be used in code anywhere an operand can be used.

If INDF0, INDF1 or INDF2 are read indirectly via an FSR, all '0's are read (zero bit is set). Similarly, if INDF0, INDF1 or INDF2 are written to indirectly, the operation will be equivalent to a NOP instruction and the STATUS bits are not affected.

4.12.1 INDIRECT ADDRESSING OPERATION

Each FSR register has an INDF register associated with it, plus four additional register addresses. Performing an operation on one of these five registers determines how the FSR will be modified during indirect addressing.

When data access is done to one of the five INDFn locations, the address selected will configure the FSRn register to:

- Do nothing to FSRn after an indirect access (no change) - INDFn
- Auto-decrement FSRn after an indirect access (post-decrement) - POSTDECn
- Auto-increment FSRn after an indirect access (post-increment) - POSTINCn
- Auto-increment FSRn before an indirect access (pre-increment) - PREINCn
- Use the value in the WREG register as an offset to FSRn. Do not modify the value of the WREG or the FSRn register after an indirect access (no change) - PLUSWn

When using the auto-increment or auto-decrement features, the effect on the FSR is not reflected in the STATUS register. For example, if the indirect address causes the FSR to equal '0', the Z bit will not be set.

Incrementing or decrementing an FSR affects all 12 bits. That is, when FSRnL overflows from an increment, FSRnH will be incremented automatically.

Adding these features allows the FSRn to be used as a stack pointer, in addition to its uses for table operations in data memory.

Each FSR has an address associated with it that performs an indexed indirect access. When a data access to this INDFn location (PLUSWn) occurs, the FSRn is configured to add the signed value in the WREG register and the value in FSR to form the address before an indirect access. The FSR value is not changed.

If an FSR register contains a value that points to one of the INDFn, an indirect read will read 00h (zero bit is set), while an indirect write will be equivalent to a NOP (STATUS bits are not affected).

If an indirect addressing operation is done where the target address is an FSRnH or FSRnL register, the write operation will dominate over the pre- or post-increment/decrement functions.

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EXAMPLE 5-3: WRITING TO FLASH PROGRAM MEMORY

```
        MOVLW    D'64                ; number of bytes in erase block
        MOVWF    COUNTER
        MOVLW    BUFFER_ADDR_HIGH    ; point to buffer
        MOVWF    FSR0H
        MOVLW    BUFFER_ADDR_LOW
        MOVWF    FSR0L
        MOVLW    CODE_ADDR_UPPER     ; Load TBLPTR with the base
        MOVWF    TBLPTRU              ; address of the memory block
        MOVLW    CODE_ADDR_HIGH
        MOVWF    TBLPTRH
        MOVLW    CODE_ADDR_LOW
        MOVWF    TBLPTRL

READ_BLOCK
        TBLRD*+                ; read into TABLAT, and inc
        MOVF     TABLAT, W        ; get data
        MOVWF    POSTINC0        ; store data
        DECFSZ   COUNTER         ; done?
        BRA      READ_BLOCK      ; repeat

MODIFY_WORD
        MOVLW    DATA_ADDR_HIGH    ; point to buffer
        MOVWF    FSR0H
        MOVLW    DATA_ADDR_LOW
        MOVWF    FSR0L
        MOVLW    NEW_DATA_LOW       ; update buffer word
        MOVWF    POSTINC0
        MOVLW    NEW_DATA_HIGH
        MOVWF    INDF0

ERASE_BLOCK
        MOVLW    CODE_ADDR_UPPER     ; load TBLPTR with the base
        MOVWF    TBLPTRU           ; address of the memory block
        MOVLW    CODE_ADDR_HIGH
        MOVWF    TBLPTRH
        MOVLW    CODE_ADDR_LOW
        MOVWF    TBLPTRL
        BSF      EECON1,EEPGD      ; point to FLASH program memory
        BCF      EECON1,CFGSRW     ; access FLASH program memory
        BSF      EECON1,WREN        ; enable write to memory
        BSF      EECON1,FREE        ; enable Row Erase operation
        BCF      INTCON,GIE        ; disable interrupts
        MOVLW    55h
        MOVWF    EECON2            ; write 55h
        MOVLW    AAh
        MOVWF    EECON2            ; write AAh
        BSF      EECON1,WR          ; start erase (CPU stall)
        BSF      INTCON,GIE        ; re-enable interrupts
        TBLRD*-                ; dummy read decrement

WRITE_BUFFER_BACK
        MOVLW    8                 ; number of write buffer groups of 8 bytes
        MOVWF    COUNTER_HI
        MOVLW    BUFFER_ADDR_HIGH    ; point to buffer
        MOVWF    FSR0H
        MOVLW    BUFFER_ADDR_LOW
        MOVWF    FSR0L

PROGRAM_LOOP
        MOVLW    8                 ; number of bytes in holding register
        MOVWF    COUNTER

WRITE_WORD_TO_HREGS
        MOVF     POSTINC0, W        ; get low byte of buffer data
        MOVWF    TABLAT            ; present data to table latch
        TBLWT+*                    ; write data, perform a short write
                                    ; to internal TBLWT holding register.
        DECFSZ   COUNTER           ; loop until buffers are full
        BRA      WRITE_WORD_TO_HREGS
```

6.3 Reading the Data EEPROM Memory

To read a data memory location, the user must write the address to the EEADR register, clear the EEPGD control bit (EECON1<7>), clear the CFGS control bit

(EECON1<6>), and then set control bit RD (EECON1<0>). The data is available for the very next instruction cycle; therefore, the EEDATA register can be read by the next instruction. EEDATA will hold this value until another read operation, or until it is written to by the user (during a write operation).

EXAMPLE 6-1: DATA EEPROM READ

```
MOVLW DATA_EE_ADDR ;
MOVWF EEADR          ; Data Memory Address to read
BCF    EECON1, EEPGD ; Point to DATA memory
BCF    EECON1, CFGS  ; Access program FLASH or Data EEPROM memory
BSF    EECON1, RD     ; EEPROM Read
MOVF   EEDATA, W      ; W = EEDATA
```

6.4 Writing to the Data EEPROM Memory

To write an EEPROM data location, the address must first be written to the EEADR register and the data written to the EEDATA register. Then the sequence in Example 6-2 must be followed to initiate the write cycle.

The write will not initiate if the above sequence is not exactly followed (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. It is strongly recommended that interrupts be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable writes. This mechanism prevents accidental writes to data EEPROM due to unexpected code exe-

cution (i.e., runaway programs). The WREN bit should be kept clear at all times, except when updating the EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, EECON1, EEADR and EEDATA cannot be modified. The WR bit will be inhibited from being set unless the WREN bit is set. The WREN bit must be set on a previous instruction. Both WR and WREN cannot be set with the same instruction.

At the completion of the write cycle, the WR bit is cleared in hardware and the EEPROM Write Complete Interrupt Flag bit (EEIF) is set. The user may either enable this interrupt, or poll this bit. EEIF must be cleared by software.

EXAMPLE 6-2: DATA EEPROM WRITE

	MOVLW DATA_EE_ADDR ;	
	MOVWF EEADR ; Data Memory Address to read	
	MOVLW DATA_EE_DATA ;	
	MOVWF EEDATA ; Data Memory Value to write	
	BCF EECON1, EEPGD ; Point to DATA memory	
	BCF EECON1, CFGS ; Access program FLASH or Data EEPROM memory	
	BSF EECON1, WREN ; Enable writes	
Required Sequence	BCF INTCON, GIE ; Disable interrupts	
	MOVLW 55h ;	
	MOVWF EECON2 ; Write 55h	
	MOVLW AAh ;	
	MOVWF EECON2 ; Write AAh	
	BSF EECON1, WR ; Set WR bit to begin write	
	BSF INTCON, GIE ; Enable interrupts	
	.	; user code execution
	.	
	.	
	BCF EECON1, WREN ; Disable writes on write complete (EEIF set)	

8.6 INT0 Interrupt

External interrupts on the RB0/INT0, RB1/INT1 and RB2/INT2 pins are edge triggered: either rising, if the corresponding INTEDGx bit is set in the INTCON2 register, or falling, if the INTEDGx bit is clear. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit INTxF is set. This interrupt can be disabled by clearing the corresponding enable bit INTxE. Flag bit INTxF must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt. All external interrupts (INT0, INT1 and INT2) can wake-up the processor from SLEEP, if bit INTxE was set prior to going into SLEEP. If the global interrupt enable bit GIE is set, the processor will branch to the interrupt vector following wake-up.

Interrupt priority for INT1 and INT2 is determined by the value contained in the interrupt priority bits, INT1IP (INTCON3<6>) and INT2IP (INTCON3<7>). There is no priority bit associated with INT0. It is always a high priority interrupt source.

8.7 TMR0 Interrupt

In 8-bit mode (which is the default), an overflow (FFh → 00h) in the TMR0 register will set flag bit TMR0IF. In 16-bit mode, an overflow (FFFFh → 0000h) in the TMR0H:TMR0L registers will set flag bit TMR0IF. The interrupt can be enabled/disabled by setting/clearing enable bit T0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit TMR0IP (INTCON2<2>). See Section 10.0 for further details on the Timer0 module.

8.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

8.9 Context Saving During Interrupts

During an interrupt, the return PC value is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the fast return stack. If a fast return from interrupt is not used (See Section 4.3), the user may need to save the WREG, STATUS and BSR registers in software. Depending on the user's application, other registers may also need to be saved. Equation 8-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

EXAMPLE 8-1: SAVING STATUS, WREG AND BSR REGISTERS IN RAM

```
MOVWF  W_TEMP                ; W_TEMP is in virtual bank
MOVFF  STATUS, STATUS_TEMP   ; STATUS_TEMP located anywhere
MOVFF  BSR,    BSR_TEMP      ; BSR located anywhere
;
; USER ISR CODE
;
MOVFF  BSR_TEMP,  BSR        ; Restore BSR
MOVF   W_TEMP,    W          ; Restore WREG
MOVFF  STATUS_TEMP, STATUS    ; Restore STATUS
```


FIGURE 9-5: BLOCK DIAGRAM OF RB2:RB0 PINS

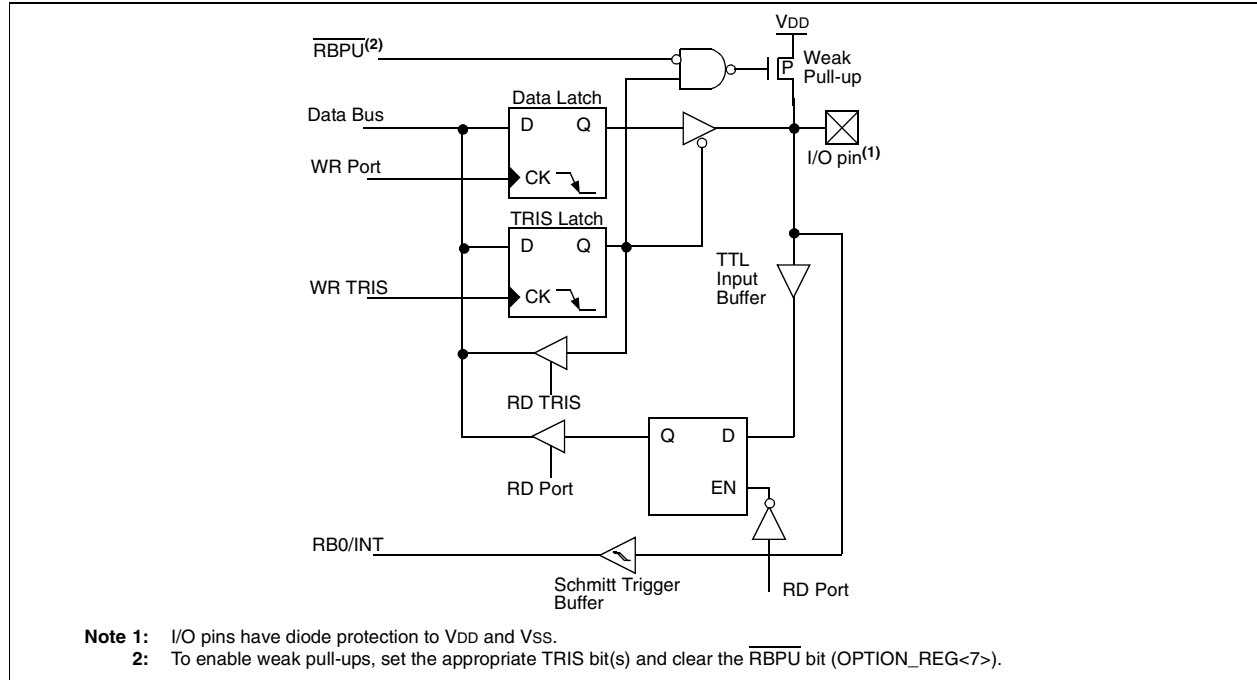
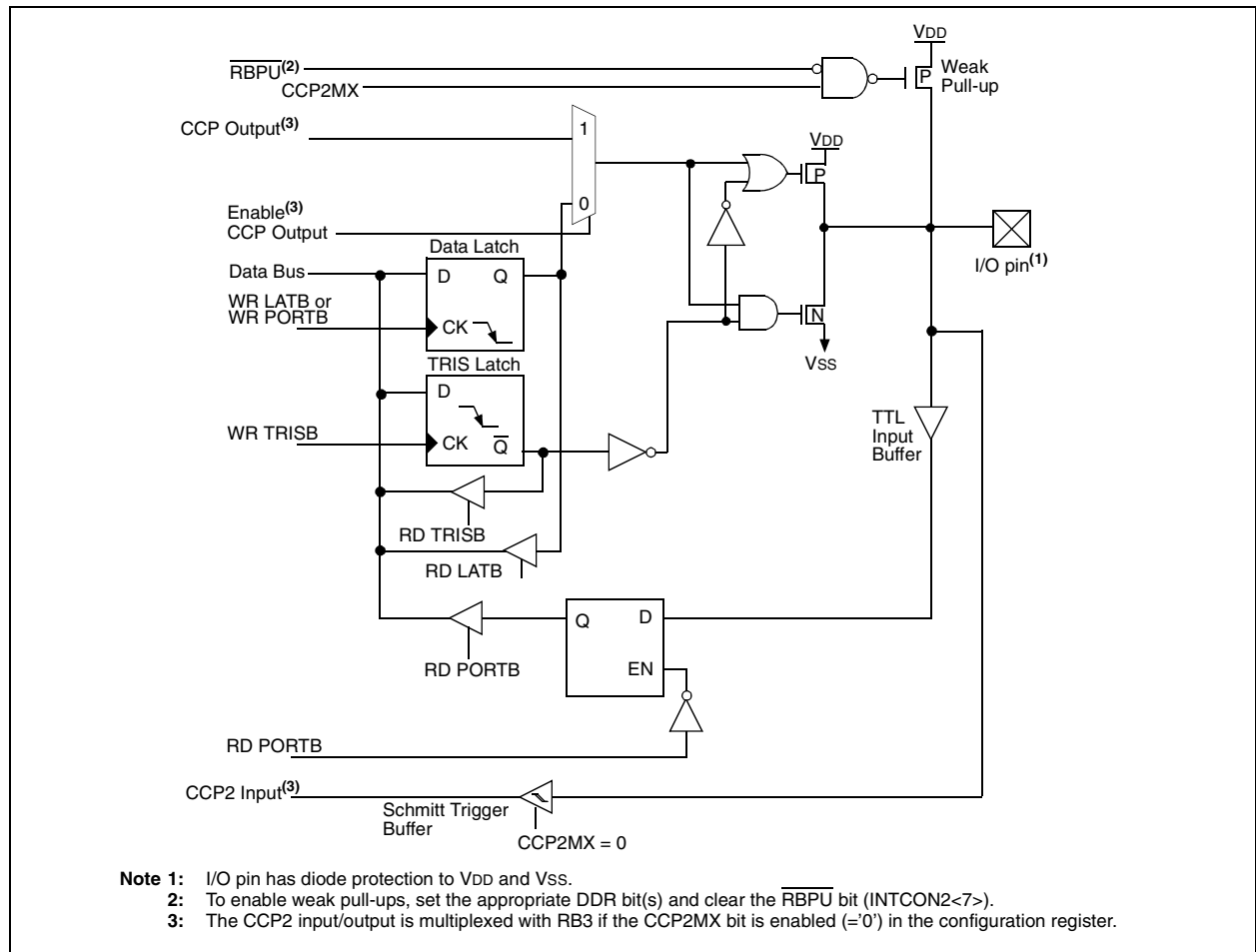


FIGURE 9-6: BLOCK DIAGRAM OF RB3 PIN



13.0 TIMER3 MODULE

The Timer3 module timer/counter has the following features:

- 16-bit timer/counter
(two 8-bit registers; TMR3H and TMR3L)
- Readable and writable (both registers)
- Internal or external clock select
- Interrupt-on-overflow from FFFFh to 0000h
- RESET from CCP module trigger

Figure 13-1 is a simplified block diagram of the Timer3 module.

Register 13-1 shows the Timer3 control register. This register controls the Operating mode of the Timer3 module and sets the CCP clock source.

Register 11-1 shows the Timer1 control register. This register controls the Operating mode of the Timer1 module, as well as contains the Timer1 oscillator enable bit (T1OSCEN), which can be a clock source for Timer3.

REGISTER 13-1: T3CON: TIMER3 CONTROL REGISTER

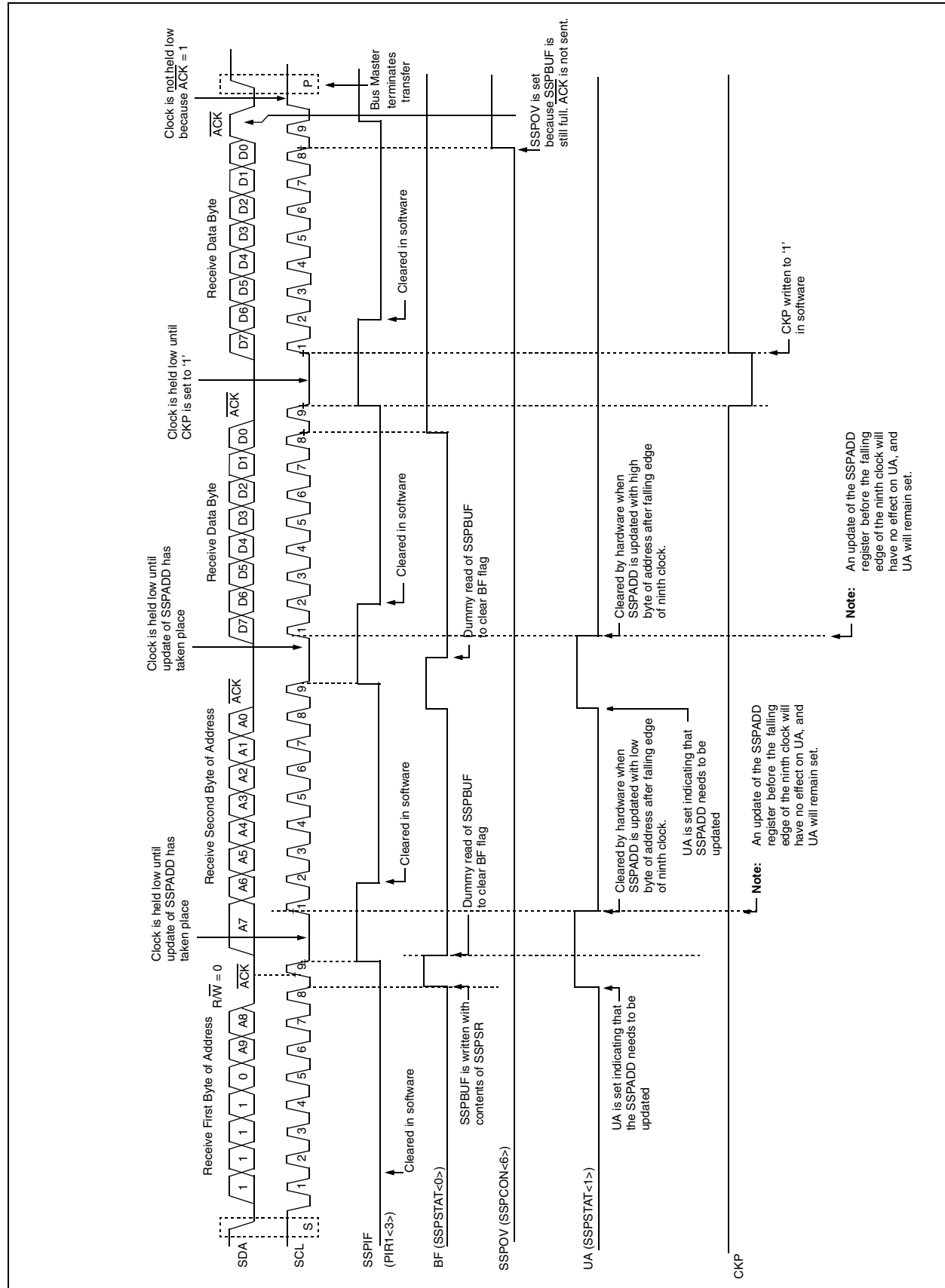
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYN \bar{C}	TMR3CS	TMR3ON
bit 7							bit 0

- bit 7 **RD16:** 16-bit Read/Write Mode Enable bit
 1 = Enables register Read/Write of Timer3 in one 16-bit operation
 0 = Enables register Read/Write of Timer3 in two 8-bit operations
- bit 6-3 **T3CCP2:T3CCP1:** Timer3 and Timer1 to CCPx Enable bits
 1x = Timer3 is the clock source for compare/capture CCP modules
 01 = Timer3 is the clock source for compare/capture of CCP2,
 Timer1 is the clock source for compare/capture of CCP1
 00 = Timer1 is the clock source for compare/capture CCP modules
- bit 5-4 **T3CKPS1:T3CKPS0:** Timer3 Input Clock Prescale Select bits
 11 = 1:8 Prescale value
 10 = 1:4 Prescale value
 01 = 1:2 Prescale value
 00 = 1:1 Prescale value
- bit 2 **T3SYN \bar{C} :** Timer3 External Clock Input Synchronization Control bit
 (Not usable if the system clock comes from Timer1/Timer3)
When TMR3CS = 1:
 1 = Do not synchronize external clock input
 0 = Synchronize external clock input
When TMR3CS = 0:
 This bit is ignored. Timer3 uses the internal clock when TMR3CS = 0.
- bit 1 **TMR3CS:** Timer3 Clock Source Select bit
 1 = External clock input from Timer1 oscillator or T1CKI
 (on the rising edge after the first falling edge)
 0 = Internal clock (FOSC/4)
- bit 0 **TMR3ON:** Timer3 On bit
 1 = Enables Timer3
 0 = Stops Timer3

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

FIGURE 15-14: I²C SLAVE MODE TIMING SEN = 1 (RECEPTION, 10-BIT ADDRESS)



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15.4.17.1 Bus Collision During a START Condition

During a START condition, a bus collision occurs if:

- SDA or SCL are sampled low at the beginning of the START condition (Figure 15-26).
- SCL is sampled low before SDA is asserted low (Figure 15-27).

During a START condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

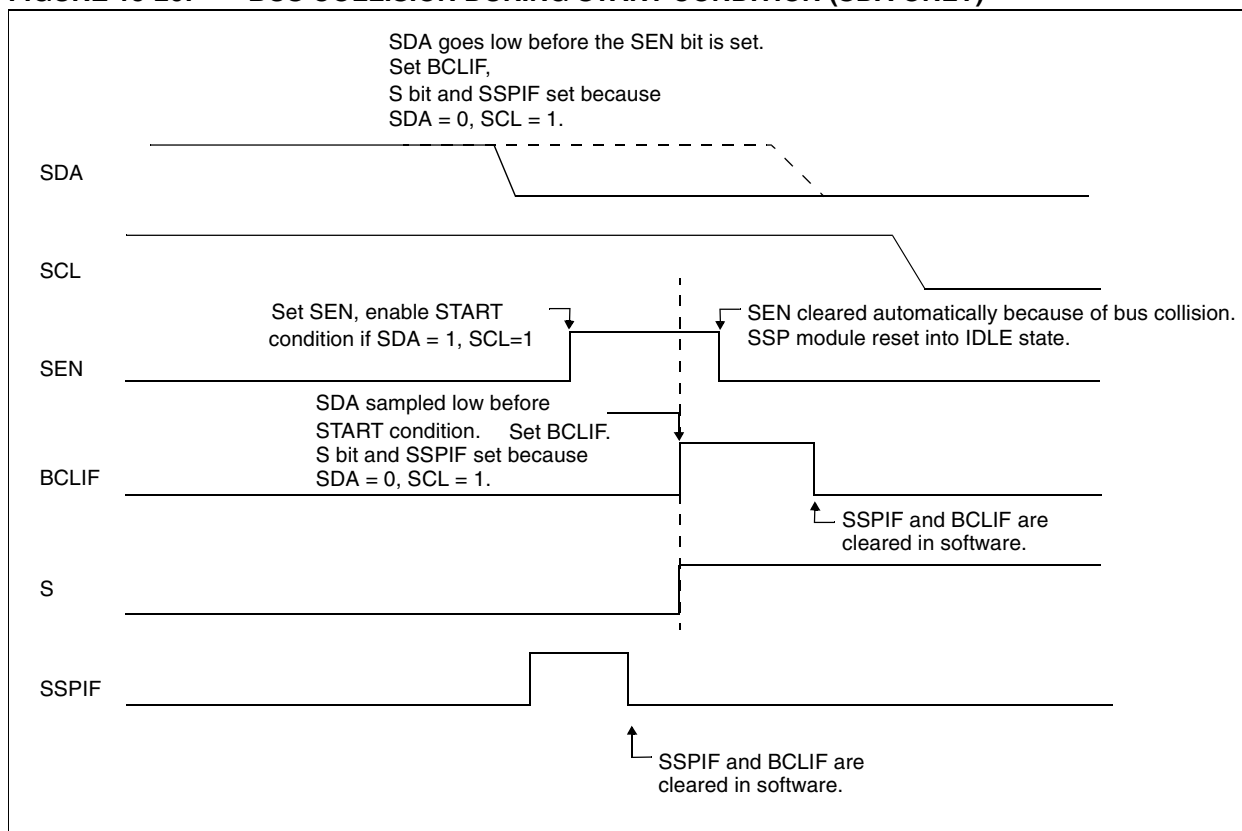
- the START condition is aborted,
- the BCLIF flag is set, and
- the MSSP module is reset to its IDLE state (Figure 15-26).

The START condition begins with the SDA and SCL pins de-asserted. When the SDA pin is sampled high, the baud rate generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the START condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 15-28). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The baud rate generator is then reloaded and counts down to 0, and during this time, if the SCL pins are sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a START condition is that no two bus masters can assert a START condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision, because the two masters must be allowed to arbitrate the first address following the START condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated START or STOP conditions.

FIGURE 15-26: BUS COLLISION DURING START CONDITION (SDA ONLY)



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NOTES:

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16.1 USART Baud Rate Generator (BRG)

The BRG supports both the Asynchronous and Synchronous modes of the USART. It is a dedicated 8-bit baud rate generator. The SPBRG register controls the period of a free running 8-bit timer. In Asynchronous mode, bit BRGH (TXSTA<2>) also controls the baud rate. In Synchronous mode, bit BRGH is ignored. Table 16-1 shows the formula for computation of the baud rate for different USART modes, which only apply in Master mode (internal clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRG register can be calculated using the formula in Table 16-1. From this, the error in baud rate can be determined.

Example 16-1 shows the calculation of the baud rate error for the following conditions:

- FOSC = 16 MHz
- Desired Baud Rate = 9600
- BRGH = 0
- SYNC = 0

It may be advantageous to use the high baud rate (BRGH = 1) even for slower baud clocks. This is because the $F_{OSC}/(16(X + 1))$ equation can reduce the baud rate error in some cases.

Writing a new value to the SPBRG register causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

16.1.1 SAMPLING

The data on the RC7/RX/DT pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX pin.

EXAMPLE 16-1: CALCULATING BAUD RATE ERROR

Desired Baud Rate	=	$F_{OSC} / (64 (X + 1))$
Solving for X:		
X	=	$((F_{OSC} / \text{Desired Baud Rate}) / 64) - 1$
X	=	$((16000000 / 9600) / 64) - 1$
X	=	$[25.042] = 25$
Calculated Baud Rate	=	$16000000 / (64 (25 + 1))$
	=	9615
Error	=	$\frac{(\text{Calculated Baud Rate} - \text{Desired Baud Rate})}{\text{Desired Baud Rate}}$
	=	$(9615 - 9600) / 9600$
	=	0.16%

TABLE 16-1: BAUD RATE FORMULA

SYNC	BRGH = 0 (Low Speed)	BRGH = 1 (High Speed)
0	(Asynchronous) Baud Rate = $F_{OSC}/(64(X+1))$	Baud Rate = $F_{OSC}/(16(X+1))$
1	(Synchronous) Baud Rate = $F_{OSC}/(4(X+1))$	N/A

Legend: X = value in SPBRG (0 to 255)

TABLE 16-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on All Other RESETS
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used by the BRG.

20.1 Instruction Set

ADDLW ADD literal to W

Syntax:	[<i>label</i>] ADDLW k				
Operands:	$0 \leq k \leq 255$				
Operation:	$(W) + k \rightarrow W$				
Status Affected:	N, OV, C, DC, Z				
Encoding:	<table><tr><td>0000</td><td>1111</td><td>kkkk</td><td>kkkk</td></tr></table>	0000	1111	kkkk	kkkk
0000	1111	kkkk	kkkk		
Description:	The contents of W are added to the 8-bit literal 'k' and the result is placed in W.				
Words:	1				
Cycles:	1				

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to W

Example: ADDLW 0x15

Before Instruction

W = 0x10

After Instruction

W = 0x25

ADDWF ADD W to f

Syntax:	[<i>label</i>] ADDWF f [,d [,a]]				
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$ $a \in [0,1]$				
Operation:	$(W) + (f) \rightarrow \text{dest}$				
Status Affected:	N, OV, C, DC, Z				
Encoding:	<table><tr><td>0010</td><td>01da</td><td>ffff</td><td>ffff</td></tr></table>	0010	01da	ffff	ffff
0010	01da	ffff	ffff		
Description:	Add W to register 'f'. If 'd' is 0, the result is stored in W. If 'd' is 1, the result is stored back in register 'f' (default). If 'a' is 0, the Access Bank will be selected. If 'a' is 1, the BSR is used.				

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: ADDWF REG, 0, 0

Before Instruction

W = 0x17

REG = 0xC2

After Instruction

W = 0xD9

REG = 0xC2

NEGf		Negate f					
Syntax:	[<i>label</i>] NEGf f [,a]						
Operands:	$0 \leq f \leq 255$ $a \in [0,1]$						
Operation:	$(\bar{f}) + 1 \rightarrow f$						
Status Affected:	N, OV, C, DC, Z						
Encoding:	0110		110a		ffff		ffff
Description:	Location 'f' is negated using two's complement. The result is placed in the data memory location 'f'. If 'a' is 0, the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value.						
Words:	1						
Cycles:	1						
Q Cycle Activity:							
	Q1		Q2		Q3		Q4
	Decode		Read register 'f'		Process Data		Write register 'f'

Example: NEGF REG, 1

Before Instruction

REG = 0011 1010 [0x3A]

After Instruction

REG = 1100 0110 [0xC6]

NOP		No Operation										
Syntax:	[<i>label</i>] NOP											
Operands:	None											
Operation:	No operation											
Status Affected:	None											
Encoding:	<table><tr><td>0000</td><td>0000</td><td>0000</td><td>0000</td></tr><tr><td>1111</td><td>xxxx</td><td>xxxx</td><td>xxxx</td></tr></table>				0000	0000	0000	0000	1111	xxxx	xxxx	xxxx
0000	0000	0000	0000									
1111	xxxx	xxxx	xxxx									
Description:	No operation.											
Words:	1											
Cycles:	1											
Q Cycle Activity:												
	Q1	Q2	Q3	Q4								
	Decode	No operation	No operation	No operation								

Example:

None.

FIGURE 22-9: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

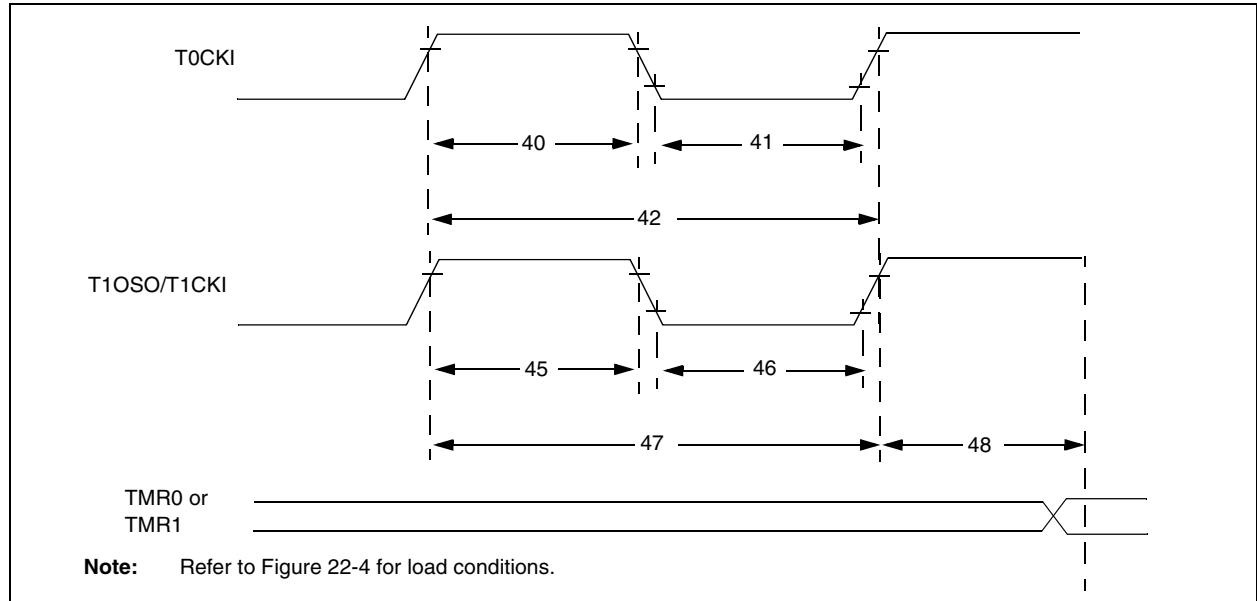


TABLE 22-8: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
40	Tt0H	T0CKI High Pulse Width	No Prescaler	$0.5T_{CY} + 20$	—	ns	
			With Prescaler	10	—	ns	
41	Tt0L	T0CKI Low Pulse Width	No Prescaler	$0.5T_{CY} + 20$	—	ns	
			With Prescaler	10	—	ns	
42	Tt0P	T0CKI Period	No Prescaler	$T_{CY} + 10$	—	ns	N = prescale value (1, 2, 4, ..., 256)
			With Prescaler	Greater of: $20 \text{ ns or } \frac{T_{CY} + 40}{N}$	—	ns	
45	Tt1H	T1CKI High Time	Synchronous, no prescaler	$0.5T_{CY} + 20$	—	ns	
			Synchronous, with prescaler	PIC18FXXX	10	—	ns
				PIC18LFXXX	25	—	ns
			Asynchronous	PIC18FXXX	30	—	ns
				PIC18LFXXX	50	—	ns
46	Tt1L	T1CKI Low Time	Synchronous, no prescaler	$0.5T_{CY} + 5$	—	ns	
			Synchronous, with prescaler	PIC18FXXX	10	—	ns
				PIC18LFXXX	25	—	ns
			Asynchronous	PIC18FXXX	30	—	ns
				PIC18LFXXX	50	—	ns
47	Tt1P	T1CKI input period	Synchronous	Greater of: $20 \text{ ns or } \frac{T_{CY} + 40}{N}$	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous	60	—	ns	
	Ft1	T1CKI oscillator input frequency range		DC	50	kHz	
48	Tcke2tmr1	Delay from external T1CKI clock edge to timer increment		$2 T_{OSC}$	$7 T_{OSC}$	—	

FIGURE 23-9: TYPICAL I_{DD} vs. F_{osc} OVER V_{DD} (EC MODE)

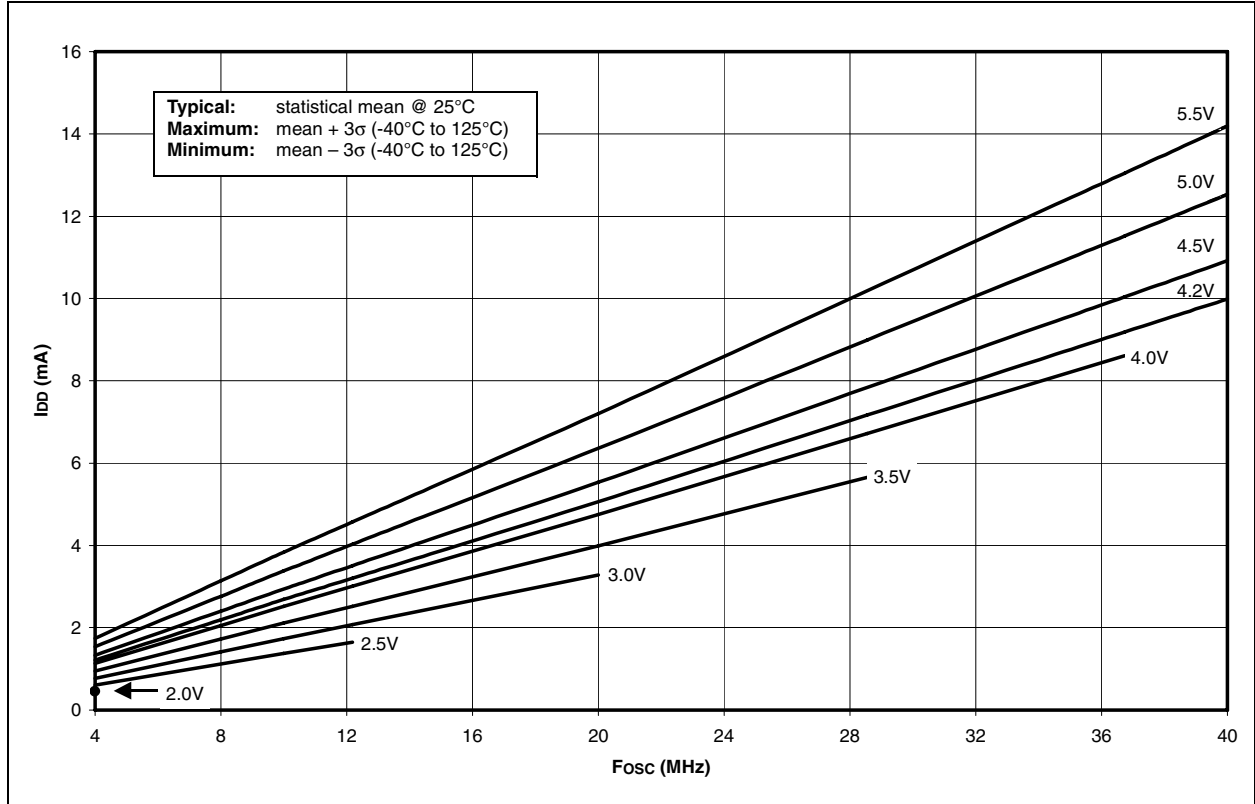
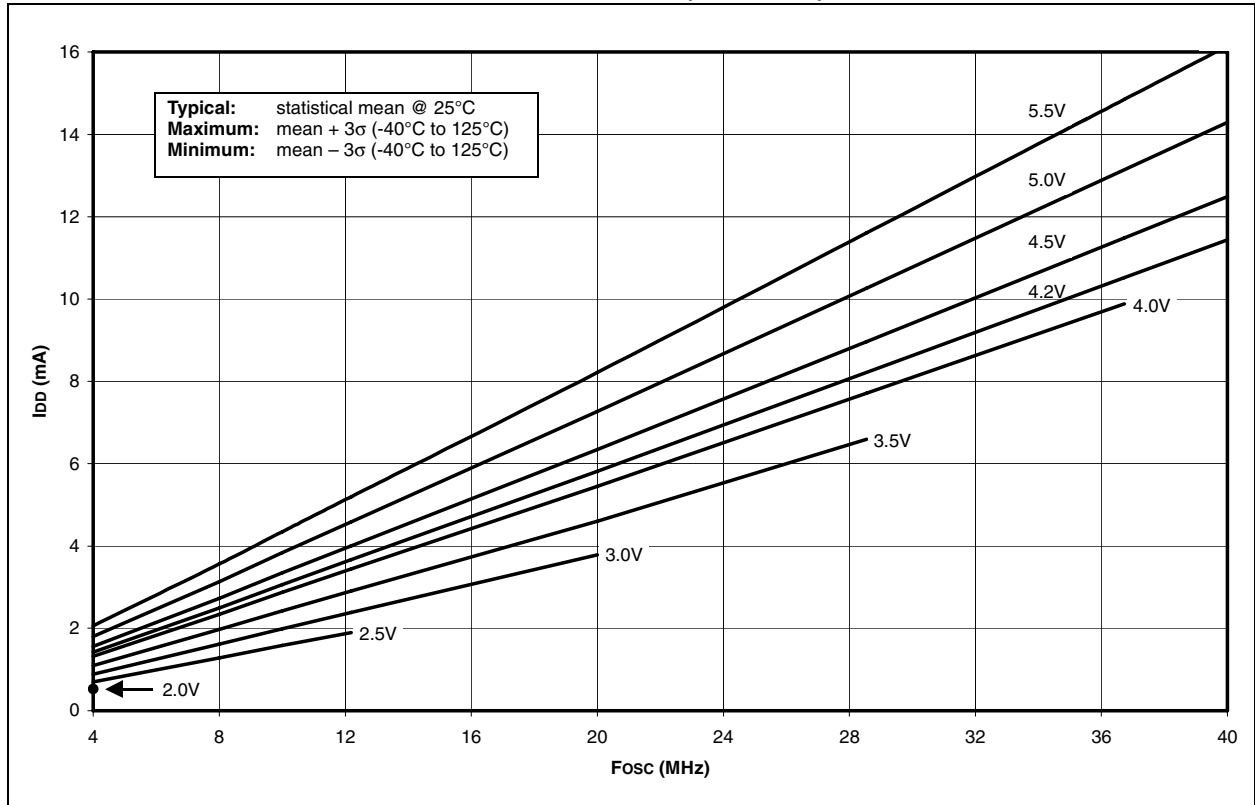


FIGURE 23-10: MAXIMUM I_{DD} vs. F_{osc} OVER V_{DD} (EC MODE)



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FIGURE 23-15: I_{PD} vs. V_{DD}, -40°C TO +125°C (SLEEP MODE, ALL PERIPHERALS DISABLED)

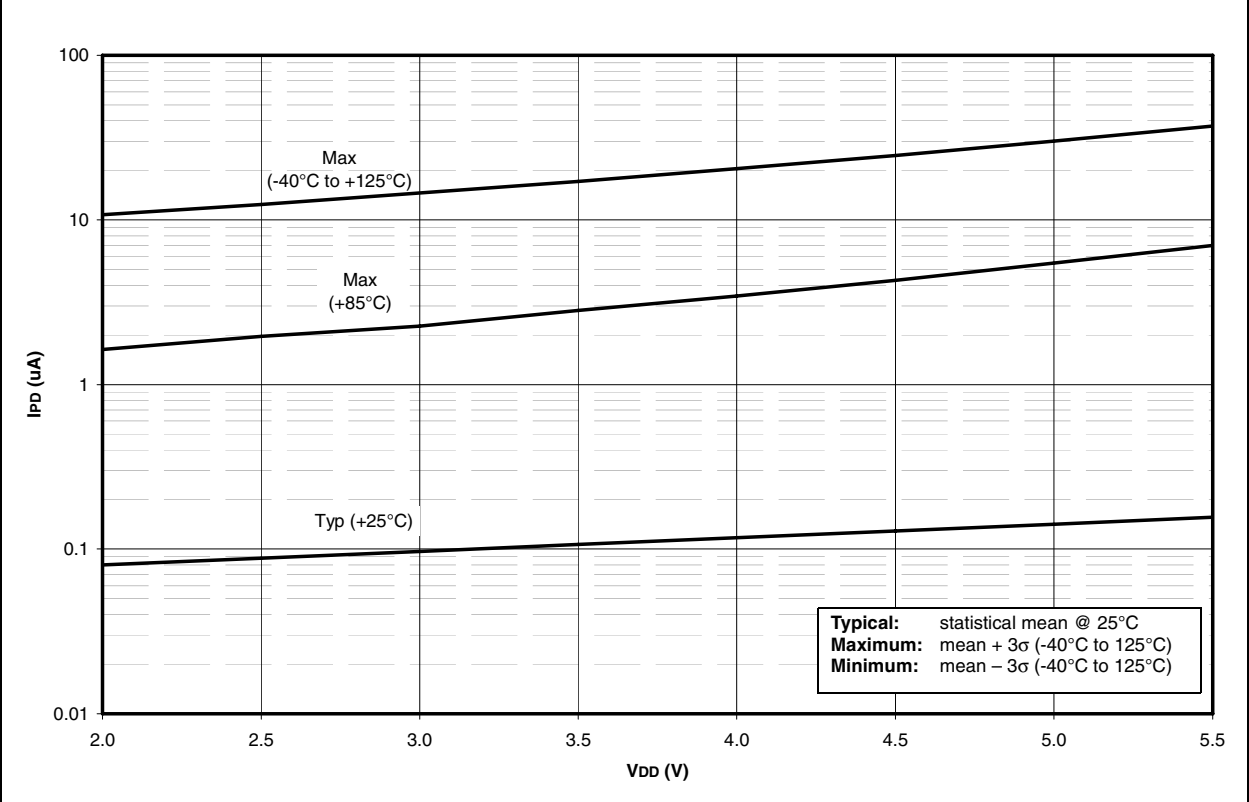
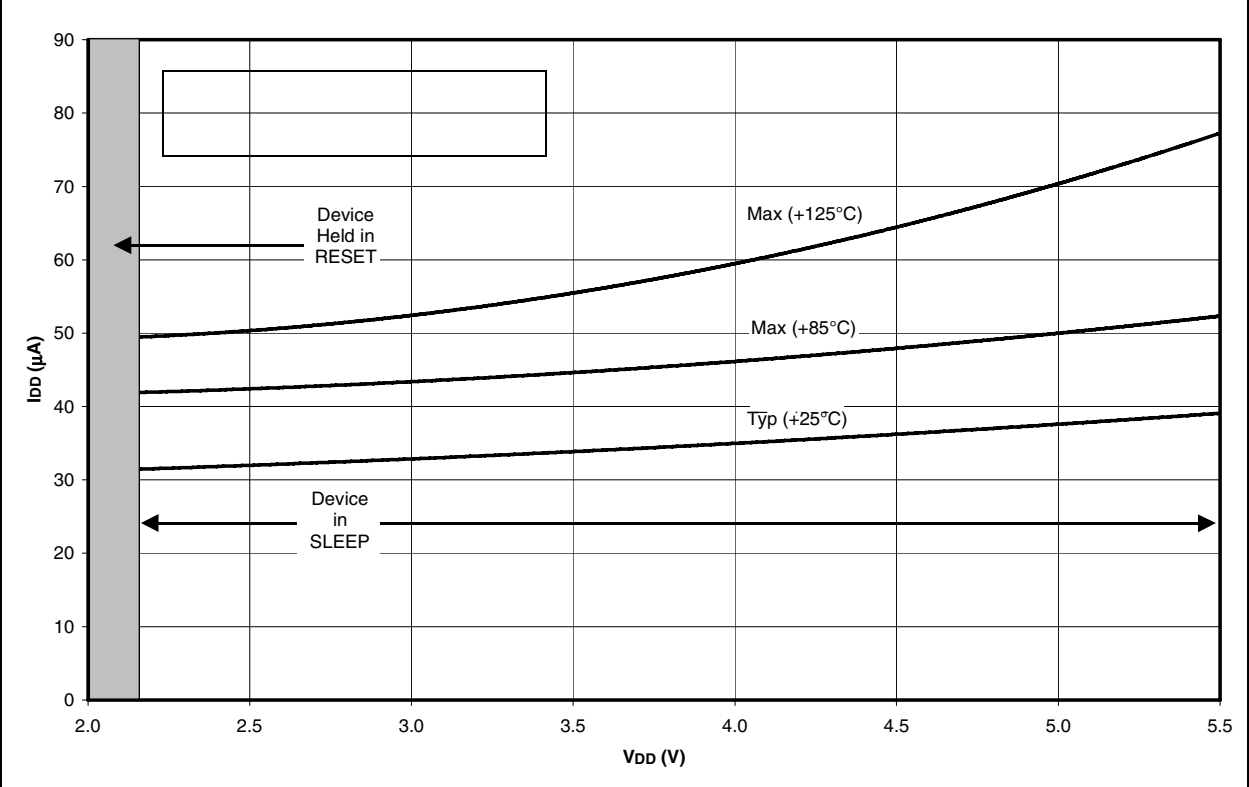


FIGURE 23-16: ΔI_{BOR} vs. V_{DD} OVER TEMPERATURE (BOR ENABLED, V_{BOR} = 2.00 - 2.16V)



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FIGURE 23-27: MINIMUM AND MAXIMUM V_{IN} vs. V_{DD} (I^2C INPUT, $-40^{\circ}C$ TO $+125^{\circ}C$)

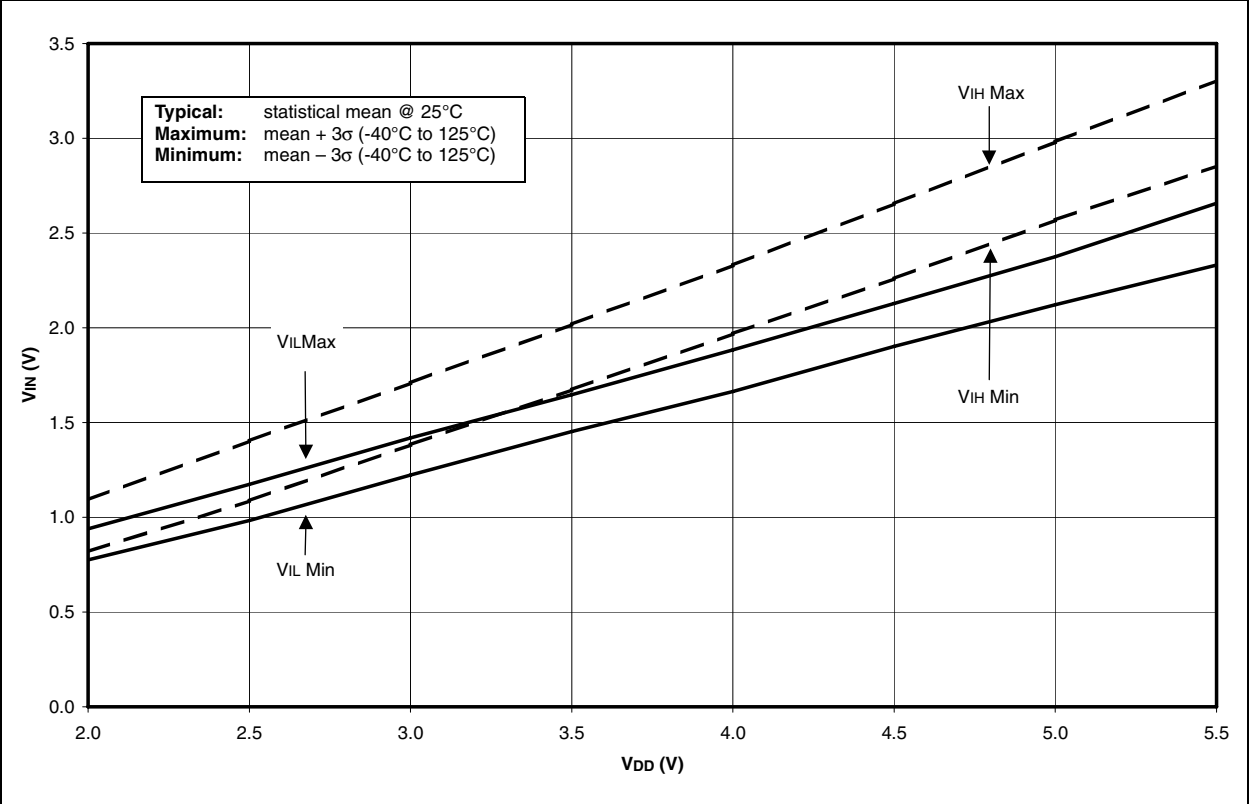
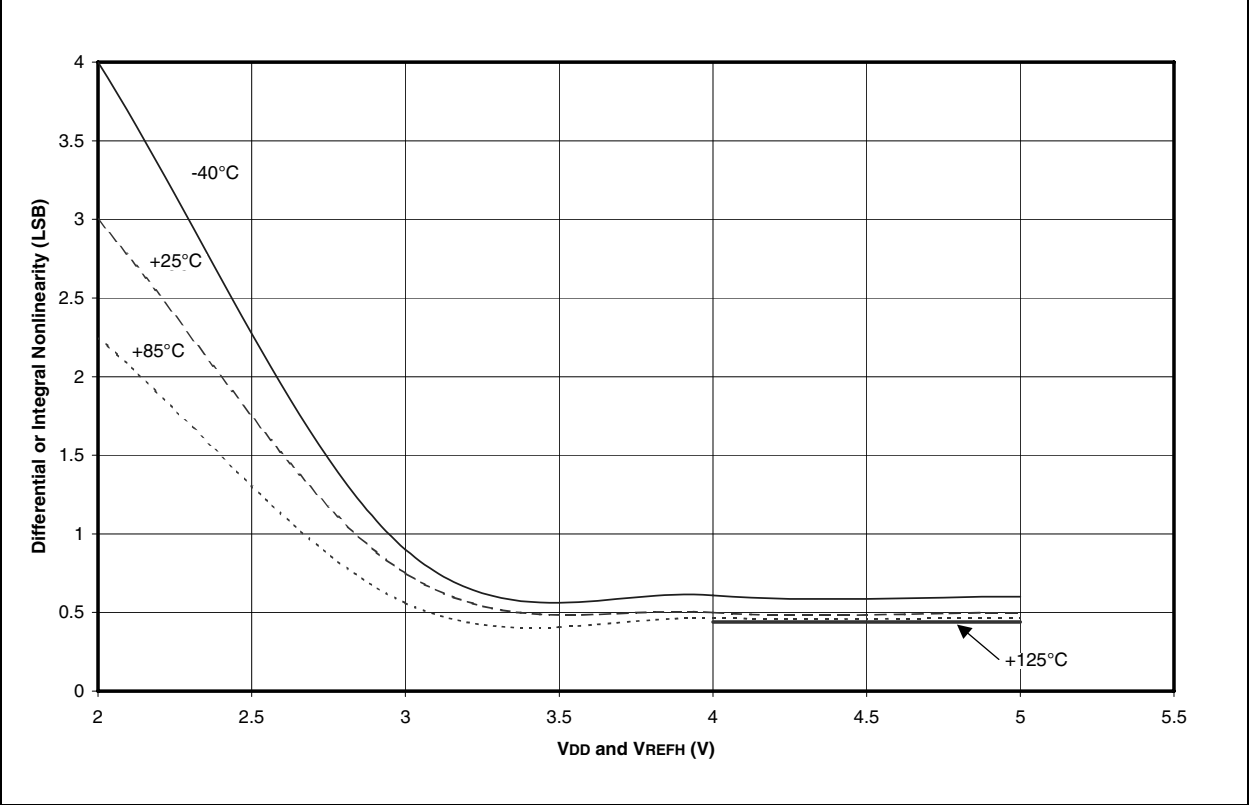


FIGURE 23-28: A/D NON-LINEARITY vs. V_{REFH} ($V_{DD} = V_{REFH}$, $-40^{\circ}C$ TO $+125^{\circ}C$)



PIC18FXX2 PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<u>PART NO.</u>	—	<u>X</u>	<u>/XX</u>	<u>XXX</u>
Device		Temperature Range	Package	Pattern
Device	PIC18FXX2 ⁽¹⁾ , PIC18FXX2T ⁽²⁾ ; VDD range 4.2V to 5.5V PIC18LFXX2 ⁽¹⁾ , PIC18LFXX2T ⁽²⁾ ; VDD range 2.5V to 5.5V			
Temperature Range	I	= -40°C to +85°C (Industrial)		
	E	= -40°C to +125°C (Extended)		
Package	PT	= TQFP (Thin Quad Flatpack)		
	SO	= SOIC		
	SP	= Skinny Plastic DIP		
	P	= PDIP		
	L	= PLCC		
Pattern	QTP, SQTP, Code or Special Requirements (blank otherwise)			

Examples:

- a) PIC18LF452 - I/P 301 = Industrial temp., PDIP package, Extended VDD limits, QTP pattern #301.
- b) PIC18LF242 - I/SO = Industrial temp., SOIC package, Extended VDD limits.
- c) PIC18F442 - E/P = Extended temp., PDIP package, normal VDD limits.

Note 1: F = Standard Voltage range
 LF = Wide Voltage Range

2: T = in tape and reel - SOIC, PLCC, and TQFP packages only.