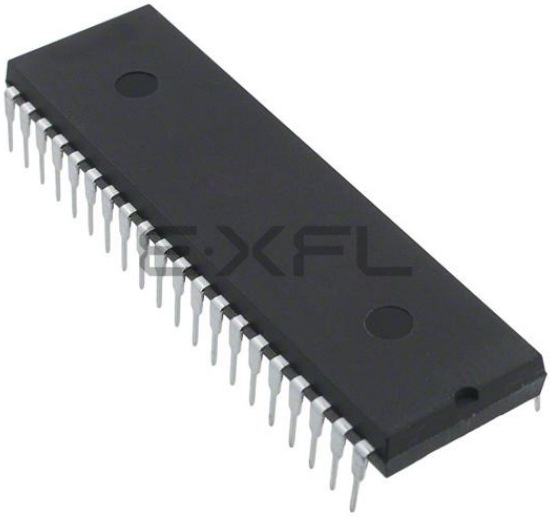

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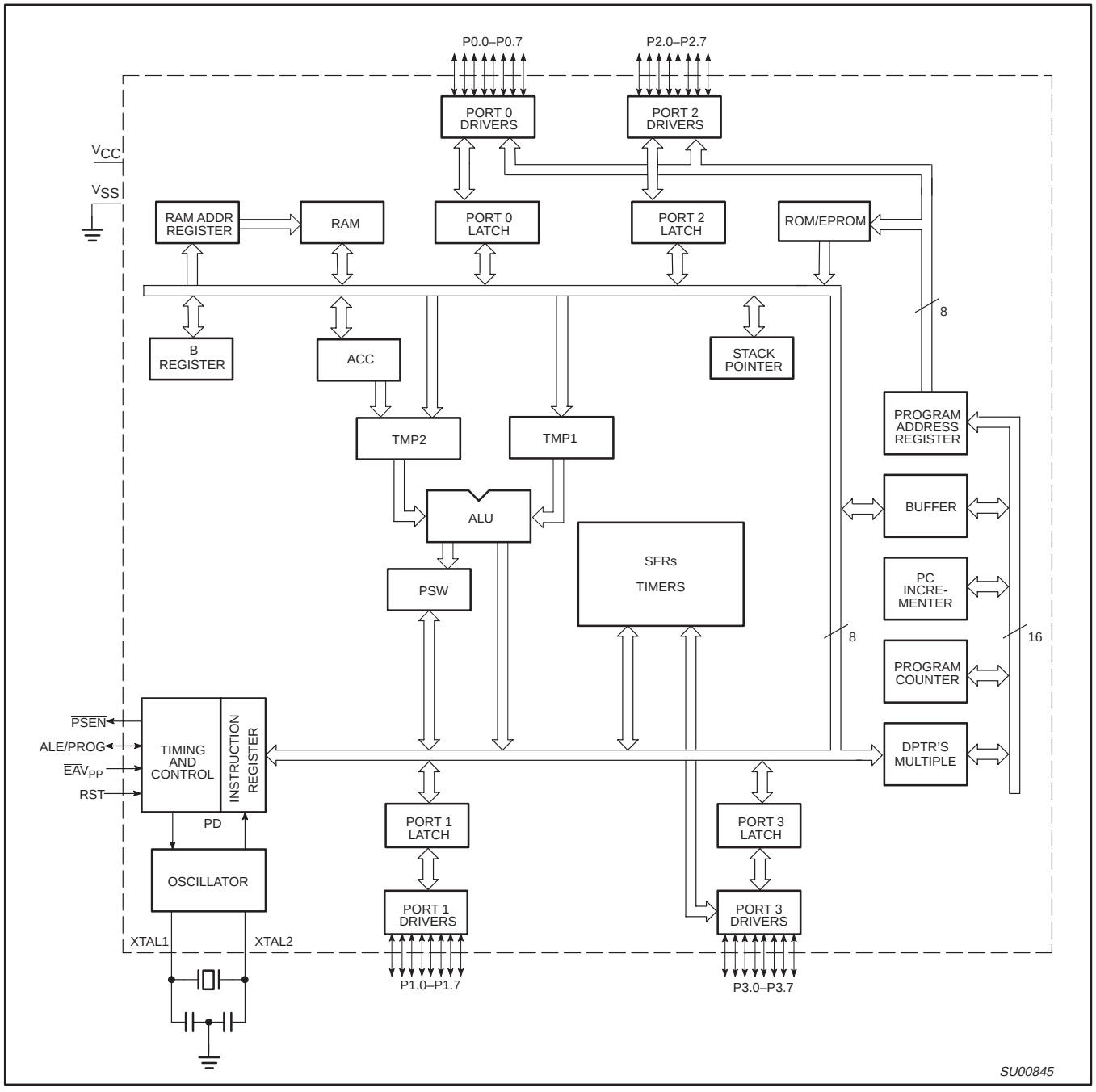
Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	16MHz
Connectivity	EBI/EMI, UART/USART
Peripherals	POR
Number of I/O	32
Program Memory Size	8KB (8K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Through Hole
Package / Case	40-DIP (0.600", 15.24mm)
Supplier Device Package	40-DIP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/p87c52sbpn-112

80C52/87C52 ORDERING INFORMATION

	MEMORY SIZE 8K × 8	TEMPERATURE RANGE °C AND PACKAGE	VOLTAGE RANGE	FREQ. (MHz)	DWG. #
ROM	P80C52SBPN	0 to +70, Plastic Dual In-line Package	2.7 V to 5.5 V	0 to 16	SOT129-1
OTP	P87C52SBPN				
ROM	P80C52SBAA	0 to +70, Plastic Leaded Chip Carrier	2.7 V to 5.5 V	0 to 16	SOT187-2
OTP	P87C52SBAA				
ROM	P80C52SBBB	0 to +70, Plastic Quad Flat Pack	2.7 V to 5.5 V	0 to 16	SOT307-2
OTP	P87C52SBBB				
ROM	P80C52SFPN	–40 to +85, Plastic Dual In-line Package	2.7 V to 5.5 V	0 to 16	SOT129-1
OTP	P87C52SFPN				
ROM	P80C52SFAA	–40 to +85, Plastic Leaded Chip Carrier	2.7 V to 5.5 V	0 to 16	SOT187-2
OTP	P87C52SFAA				
ROM	P80C52SFBB	–40 to +85, Plastic Quad Flat Pack	2.7 V to 5.5 V	0 to 16	SOT307-2
OTP	P87C52SFBB				
ROM	P80C52UBAA	0 to +70, Plastic Leaded Chip Carrier	5 V	0 to 33	SOT187-2
OTP	P87C52UBAA				
ROM	P80C52UBPN	0 to +70, Plastic Dual In-line Package	5 V	0 to 33	SOT129-1
OTP	P87C52UBPN				
ROM	P80C52UFAA	–40 to +85, Plastic Leaded Chip Carrier	5 V	0 to 33	SOT187-2
OTP	P87C52UFAA				

BLOCK DIAGRAM



OSCILLATOR CHARACTERISTICS

XTAL1 and XTAL2 are the input and output, respectively, of an inverting amplifier. The pins can be configured for use as an on-chip oscillator, as shown in the logic symbol.

To drive the device from an external clock source, XTAL1 should be driven while XTAL2 is left unconnected. There are no requirements on the duty cycle of the external clock signal, because the input to the internal clock circuitry is through a divide-by-two flip-flop. However, minimum and maximum high and low times specified in the data sheet must be observed.

Reset

A reset is accomplished by holding the RST pin high for at least two machine cycles (24 oscillator periods), while the oscillator is running. To insure a good power-up reset, the RST pin must be high long enough to allow the oscillator time to start up (normally a few milliseconds) plus two machine cycles.

Stop Clock Mode

The static design enables the clock speed to be reduced down to 0 MHz (stopped). When the oscillator is stopped, the RAM and Special Function Registers retain their values. This mode allows step-by-step utilization and permits reduced system power consumption by lowering the clock frequency down to any value. For lowest power consumption the Power Down mode is suggested.

Idle Mode

In idle mode (see Table 2), the CPU puts itself to sleep while all of the on-chip peripherals stay active. The instruction to invoke the idle mode is the last instruction executed in the normal operating mode before the idle mode is activated. The CPU contents, the on-chip RAM, and all of the special function registers remain intact during this mode. The idle mode can be terminated either by any enabled interrupt (at which time the process is picked up at the interrupt service routine and continued), or by a hardware reset which starts the processor in the same manner as a power-on reset.

Power-Down Mode

To save even more power, a Power Down mode (see Table 2) can be invoked by software. In this mode, the oscillator is stopped and the instruction that invoked Power Down is the last instruction executed. The on-chip RAM and Special Function Registers retain their values down to 2.0 V and care must be taken to return V_{CC} to the minimum specified operating voltages before the Power Down Mode is terminated.

For the 87C51 and 80C51 either a hardware reset or external interrupt can be used to exit from Power Down. Reset redefines all the SFRs but does not change the on-chip RAM. An external interrupt allows both the SFRs and the on-chip RAM to retain their values. WUPD (AUXR1.3–Wakeup from Power Down) enables or disables the wakeup from power down with external interrupt. Where:

WUPD = 0 Disable
WUPD = 1 Enable

To properly terminate Power Down the reset or external interrupt should not be executed before V_{CC} is restored to its normal operating level and must be held active long enough for the oscillator to restart and stabilize (normally less than 10 ms).

With an external interrupt, INT0 or INT1 must be enabled and configured as level-sensitive. Holding the pin low restarts the oscillator but bringing the pin back high completes the exit. Once the interrupt is serviced, the next instruction to be executed after RETI will be the one following the instruction that put the device into Power Down.

LPEP

The eprom array contains some analog circuits that are not required when V_{CC} is less than 4 V, but are required for a V_{CC} greater than 4 V. The LPEP bit (AUXR.4), when set, will powerdown these analog circuits resulting in a reduced supply current. This bit should be set ONLY for applications that operate at a V_{CC} less than 4 V.

Design Consideration

- When the idle mode is terminated by a hardware reset, the device normally resumes program execution, from where it left off, up to two machine cycles before the internal reset algorithm takes control. On-chip hardware inhibits access to internal RAM in this event, but access to the port pins is not inhibited. To eliminate the possibility of an unexpected write when Idle is terminated by reset, the instruction following the one that invokes Idle should not be one that writes to a port pin or to external memory.

ONCE™ Mode

The ONCE (“On-Circuit Emulation”) Mode facilitates testing and debugging of systems without the device having to be removed from the circuit. The ONCE Mode is invoked by:

1. Pull ALE low while the device is in reset and $\overline{\text{PSEN}}$ is high;
2. Hold ALE low as RST is deactivated.

While the device is in ONCE Mode, the Port 0 pins go into a float state, and the other port pins and ALE and $\overline{\text{PSEN}}$ are weakly pulled high. The oscillator circuit remains active. While the device is in this mode, an emulator or test CPU can be used to drive the circuit. Normal operation is restored when a normal reset is applied.

Table 2. External Pin Status During Idle and Power-Down Modes

MODE	PROGRAM MEMORY	ALE	$\overline{\text{PSEN}}$	PORT 0	PORT 1	PORT 2	PORT 3
Idle	Internal	1	1	Data	Data	Data	Data
Idle	External	1	1	Float	Data	Address	Data
Power-down	Internal	0	0	Data	Data	Data	Data
Power-down	External	0	0	Float	Data	Data	Data

Programmable Clock-Out

A 50% duty cycle clock can be programmed to come out on P1.0. This pin, besides being a regular I/O pin, has two alternate functions. It can be programmed:

1. to input the external clock for Timer/Counter 2, or
2. to output a 50% duty cycle clock ranging from 61 Hz to 4 MHz at a 16 MHz operating frequency.

To configure the Timer/Counter 2 as a clock generator, bit C/T2 (in T2CON) must be cleared and bit T20E in T2MOD must be set. Bit TR2 (T2CON.2) also must be set to start the timer.

The Clock-Out frequency depends on the oscillator frequency and the reload value of Timer 2 capture registers (RCAP2H, RCAP2L) as shown in this equation:

$$\frac{\text{Oscillator Frequency}}{4 \times (65536 - \text{RCAP2H, RCAP2L})}$$

Where:

(RCAP2H,RCAP2L) = the content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

In the Clock-Out mode Timer 2 roll-overs will not generate an interrupt. This is similar to when it is used as a baud-rate generator. It is possible to use Timer 2 as a baud-rate generator and a clock generator simultaneously. Note, however, that the baud-rate and the Clock-Out frequency will be the same.

TIMER 2 OPERATION

Timer 2

Timer 2 is a 16-bit Timer/Counter which can operate as either an event timer or an event counter, as selected by C/$\overline{T}$2* in the special function register T2CON (see Figure 1). Timer 2 has three operating modes:Capture, Auto-reload (up or down counting) ,and Baud Rate Generator, which are selected by bits in the T2CON as shown in Table 3.

Capture Mode

In the capture mode there are two options which are selected by bit EXEN2 in T2CON. If EXEN2=0, then timer 2 is a 16-bit timer or counter (as selected by C/T2* in T2CON) which, upon overflowing sets bit TF2, the timer 2 overflow bit. This bit can be used to generate an interrupt (by enabling the Timer 2 interrupt bit in the IE register). If EXEN2= 1, Timer 2 operates as described above, but with the added feature that a 1- to -0 transition at external input T2EX causes the current value in the Timer 2 registers, TL2 and TH2, to be captured into registers RCAP2L and RCAP2H, respectively. In addition, the transition at T2EX causes bit EXF2 in T2CON to be set, and EXF2 like TF2 can generate an interrupt (which vectors to the same location as Timer 2 overflow interrupt. The Timer 2 interrupt service routine can interrogate TF2 and EXF2 to determine which event caused the interrupt). The capture mode is illustrated in Figure 2 (There is no reload value for TL2 and TH2 in this mode. Even when a capture event occurs from T2EX, the counter keeps on counting T2EX pin transitions or osc/12 pulses.).

Auto-Reload Mode (Up or Down Counter)

In the 16-bit auto-reload mode, Timer 2 can be configured (as either a timer or counter (C/T2* in T2CON)) then programmed to count up or down. The counting direction is determined by bit DCEN (Down Counter Enable) which is located in the T2MOD register (see Figure 3). When reset is applied the DCEN=0 which means Timer 2 will default to counting up. If DCEN bit is set, Timer 2 can count up or down depending on the value of the T2EX pin.

Figure 4 shows Timer 2 which will count up automatically since DCEN=0. In this mode there are two options selected by bit EXEN2 in T2CON register. If EXEN2=0, then Timer 2 counts up to 0FFFFH and sets the TF2 (Overflow Flag) bit upon overflow. This causes the Timer 2 registers to be reloaded with the 16-bit value in RCAP2L and RCAP2H. The values in RCAP2L and RCAP2H are preset by software means.

If EXEN2=1, then a 16-bit reload can be triggered either by an overflow or by a 1-to-0 transition at input T2EX. This transition also sets the EXF2 bit. The Timer 2 interrupt, if enabled, can be generated when either TF2 or EXF2 are 1.

In Figure 5 DCEN=1 which enables Timer 2 to count up or down. This mode allows pin T2EX to control the direction of count. When a logic 1 is applied at pin T2EX Timer 2 will count up. Timer 2 will overflow at 0FFFFH and set the TF2 flag, which can then generate an interrupt, if the interrupt is enabled. This timer overflow also causes the 16–bit value in RCAP2L and RCAP2H to be reloaded into the timer registers TL2 and TH2.

When a logic 0 is applied at pin T2EX this causes Timer 2 to count down. The timer will underflow when TL2 and TH2 become equal to the value stored in RCAP2L and RCAP2H. Timer 2 underflow sets the TF2 flag and causes 0FFFFH to be reloaded into the timer registers TL2 and TH2.

The external flag EXF2 toggles when Timer 2 underflows or overflows. This EXF2 bit can be used as a 17th bit of resolution if needed. The EXF2 flag does not generate an interrupt in this mode of operation.

Table 3. Timer 2 Operating Modes

RCLK + TCLK	CP/$\overline{RL2}$	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	X	1	Baud rate generator
X	X	0	(off)

T2MOD Address = 0C9H Reset Value = XXXX XX00B

Not Bit Addressable

—	—	—	—	—	—	T2OE	DCEN
Bit 7	6	5	4	3	2	1	0

Symbol	Function
—	Not implemented, reserved for future use.*
T2OE	Timer 2 Output Enable bit.
DCEN	Down Count Enable bit. When set, this allows Timer 2 to be configured as an up/down counter.

* User software should not write 1s to reserved bits. These bits may be used in future 8051 family products to invoke new features. In that case, the reset or inactive value of the new bit will be 0, and its active value will be 1. The value read from a reserved bit is indeterminate.

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Figure 3. Timer 2 Mode (T2MOD) Control Register

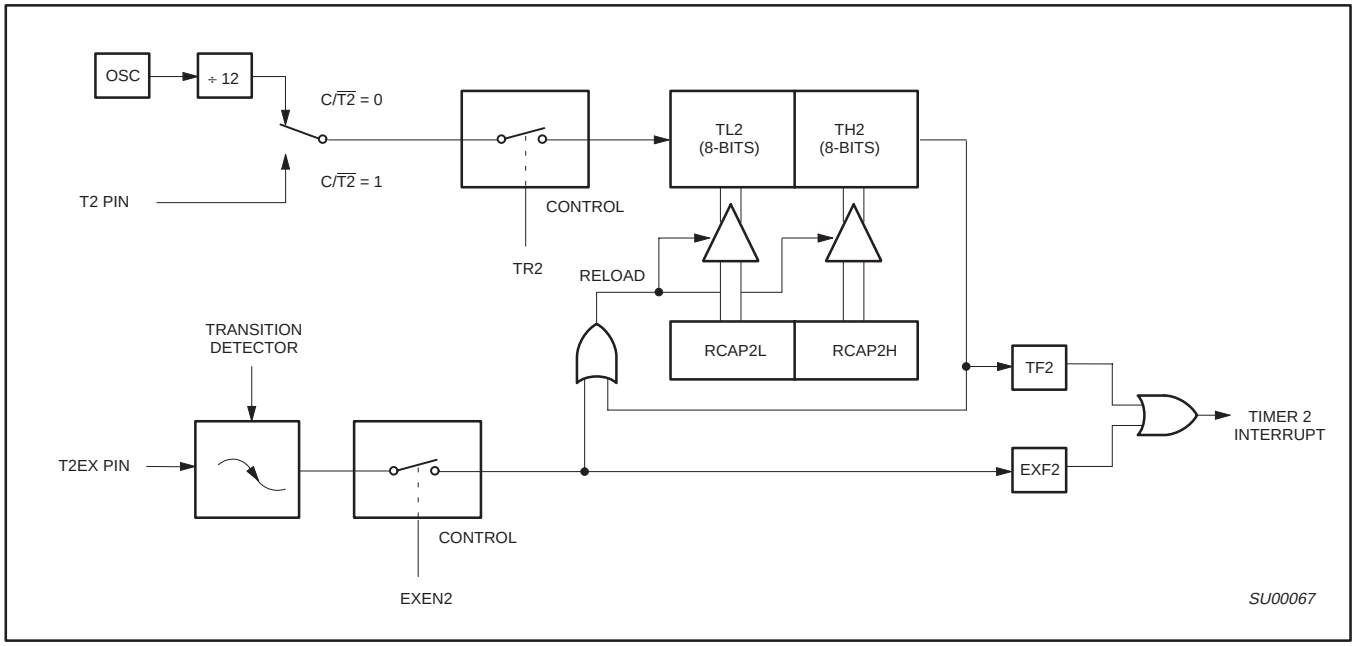

Figure 4. Timer 2 in Auto-Reload Mode (DCEN = 0)

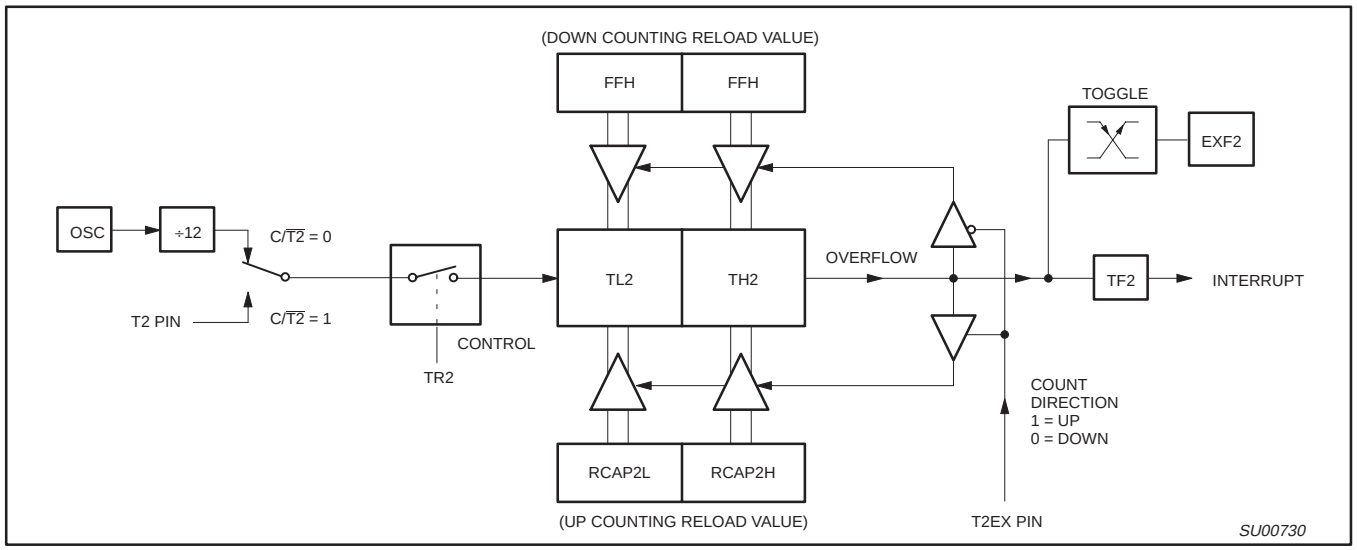

Figure 5. Timer 2 Auto Reload Mode (DCEN = 1)

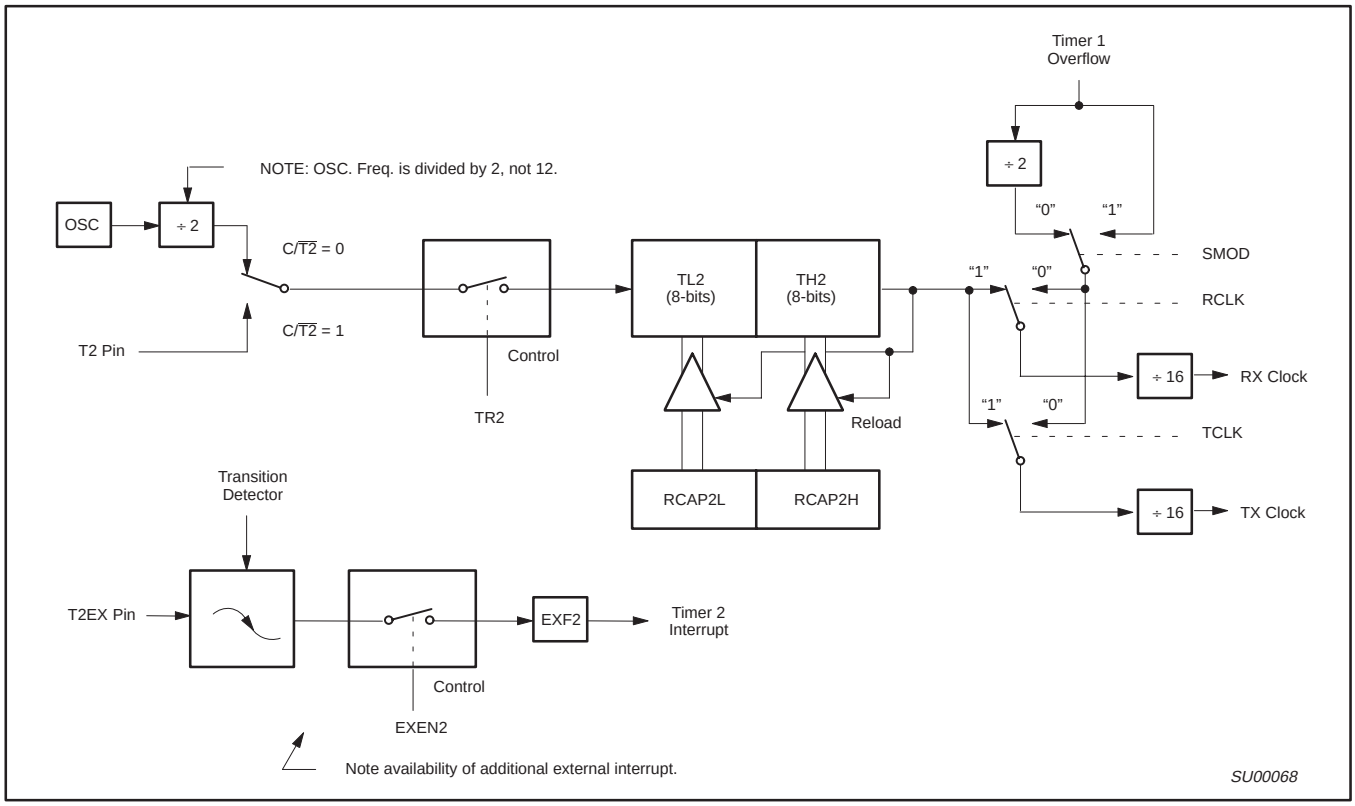

Figure 6. Timer 2 in Baud Rate Generator Mode

Baud Rate Generator Mode

Bits TCLK and/or RCLK in T2CON (Table 3) allow the serial port transmit and receive baud rates to be derived from either Timer 1 or Timer 2. When TCLK= 0, Timer 1 is used as the serial port transmit baud rate generator. When TCLK= 1, Timer 2 is used as the serial port transmit baud rate generator. RCLK has the same effect for the serial port receive baud rate. With these two bits, the serial port can have different receive and transmit baud rates – one generated by Timer 1, the other by Timer 2.

Figure 6 shows the Timer 2 in baud rate generation mode. The baud rate generation mode is like the auto-reload mode, in that a rollover in TH2 causes the Timer 2 registers to be reloaded with the 16-bit value in registers RCAP2H and RCAP2L, which are preset by software.

The baud rates in modes 1 and 3 are determined by Timer 2's overflow rate given below:

$$\text{Modes 1 and 3 Baud Rates} = \frac{\text{Timer 2 Overflow Rate}}{16}$$

The timer can be configured for either "timer" or "counter" operation. In many applications, it is configured for "timer" operation (C/$\overline{T}$2*=0). Timer operation is different for Timer 2 when it is being used as a baud rate generator.

Usually, as a timer it would increment every machine cycle (i.e., 1/12 the oscillator frequency). As a baud rate generator, it increments every state time (i.e., 1/2 the oscillator frequency). Thus the baud rate formula is as follows:

$$\text{Modes 1 and 3 Baud Rates} = \frac{\text{Oscillator Frequency}}{[32 \times [65536 - (\text{RCAP2H}, \text{RCAP2L})]]}$$

Where: (RCAP2H, RCAP2L)= The content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

The Timer 2 as a baud rate generator mode shown in Figure 6, is valid only if RCLK and/or TCLK = 1 in T2CON register. Note that a rollover in TH2 does not set TF2, and will not generate an interrupt. Thus, the Timer 2 interrupt does not have to be disabled when Timer 2 is in the baud rate generator mode. Also if the EXEN2 (T2 external enable flag) is set, a 1-to-0 transition in T2EX (Timer/counter 2 trigger input) will set EXF2 (T2 external flag) but will not cause a reload from (RCAP2H, RCAP2L) to (TH2,TL2). Therefore when Timer 2 is in use as a baud rate generator, T2EX can be used as an additional external interrupt, if needed.

When Timer 2 is in the baud rate generator mode, one should not try to read or write TH2 and TL2. As a baud rate generator, Timer 2 is incremented every state time (osc/2) or asynchronously from pin T2; under these conditions, a read or write of TH2 or TL2 may not be accurate. The RCAP2 registers may be read, but should not be written to, because a write might overlap a reload and cause write and/or reload errors. The timer should be turned off (clear TR2) before accessing the Timer 2 or RCAP2 registers.

Table 4 shows commonly used baud rates and how they can be obtained from Timer 2.

Table 4. Timer 2 Generated Commonly Used Baud Rates

Baud Rate	Osc Freq	Timer 2	
		RCAP2H	RCAP2L
375 K	12 MHz	FF	FF
9.6 K	12 MHz	FF	D9
2.8 K	12 MHz	FF	B2
2.4 K	12 MHz	FF	64
1.2 K	12 MHz	FE	C8
300	12 MHz	FB	1E
110	12 MHz	F2	AF
300	6 MHz	FD	8F
110	6 MHz	F9	57

Summary Of Baud Rate Equations

Timer 2 is in baud rate generating mode. If Timer 2 is being clocked through pin T2(P1.0) the baud rate is:

$$\text{Baud Rate} = \frac{\text{Timer 2 Overflow Rate}}{16}$$

If Timer 2 is being clocked internally, the baud rate is:

$$\text{Baud Rate} = \frac{f_{OSC}}{[32 \times [65536 - (\text{RCAP2H}, \text{RCAP2L})]]}$$

Where f_{OSC}= Oscillator Frequency

To obtain the reload value for RCAP2H and RCAP2L, the above equation can be rewritten as:

$$\text{RCAP2H}, \text{RCAP2L} = 65536 - \left(\frac{f_{OSC}}{32 \times \text{Baud Rate}}\right)$$

Timer/Counter 2 Set-up

Except for the baud rate generator mode, the values given for T2CON do not include the setting of the TR2 bit. Therefore, bit TR2 must be set, separately, to turn the timer on. See Table 5 for set-up of Timer 2 as a timer. Also see Table 6 for set-up of Timer 2 as a counter.

Table 5. Timer 2 as a Timer

MODE	T2CON	
	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)
16-bit Auto-Reload	00H	08H
16-bit Capture	01H	09H
Baud rate generator receive and transmit same baud rate	34H	36H
Receive only	24H	26H
Transmit only	14H	16H

Table 6. Timer 2 as a Counter

MODE	TMOD	
	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)
16-bit	02H	0AH
Auto-Reload	03H	0BH

NOTES:
1. Capture/reload occurs only on timer/counter overflow.
2. Capture/reload occurs on timer/counter overflow and a 1-to-0 transition on T2EX (P1.1) pin except when Timer 2 is used in the baud rate generator mode.

Enhanced UART

The UART operates in all of the usual modes that are described in the first section of *Data Handbook IC20, 80C51-Based 8-Bit Microcontrollers*. In addition the UART can perform framing error detect by looking for missing stop bits, and automatic address recognition. The UART also fully supports multiprocessor communication.

When used for framing error detect the UART looks for missing stop bits in the communication. A missing bit will set the FE bit in the SCON register. The FE bit shares the SCON.7 bit with SM0 and the function of SCON.7 is determined by PCON.6 (SMOD0) (see Figure 7). If SMOD0 is set then SCON.7 functions as FE. SCON.7 functions as SM0 when SMOD0 is cleared. When used as FE SCON.7 can only be cleared by software. Refer to Figure 8.

Automatic Address Recognition

Automatic Address Recognition is a feature which allows the UART to recognize certain addresses in the serial bit stream by using hardware to make the comparisons. This feature saves a great deal of software overhead by eliminating the need for the software to examine every serial address which passes by the serial port. This feature is enabled by setting the SM2 bit in SCON. In the 9 bit UART modes, mode 2 and mode 3, the Receive Interrupt flag (RI) will be automatically set when the received byte contains either the “Given” address or the “Broadcast” address. The 9 bit mode requires that the 9th information bit is a 1 to indicate that the received information is an address and not data. Automatic address recognition is shown in Figure 9.

The 8 bit mode is called Mode 1. In this mode the RI flag will be set if SM2 is enabled and the information received has a valid stop bit following the 8 address bits and the information is either a Given or Broadcast address.

Mode 0 is the Shift Register mode and SM2 is ignored.

Using the Automatic Address Recognition feature allows a master to selectively communicate with one or more slaves by invoking the Given slave address or addresses. All of the slaves may be contacted by using the Broadcast address. Two special Function Registers are used to define the slave’s address, SADDR, and the address mask, SADEN. SADEN is used to define which bits in the SADDR are to b used and which bits are “don’t care”. The SADEN mask can be logically ANDed with the SADDR to create the “Given” address which the master will use for addressing each of the slaves. Use of the Given address allows multiple slaves to be recognized while excluding others. The following examples will help to show the versatility of this scheme:

Slave 0	SADDR	=	1100 0000
	SADEN	=	1111 1101
	Given	=	1100 00X0
Slave 1	SADDR	=	1100 0000
	SADEN	=	1111 1110
	Given	=	1100 000X

In the above example SADDR is the same and the SADEN data is used to differentiate between the two slaves. Slave 0 requires a 0 in bit 0 and it ignores bit 1. Slave 1 requires a 0 in bit 1 and bit 0 is ignored. A unique address for Slave 0 would be 1100 0010 since slave 1 requires a 0 in bit 1. A unique address for slave 1 would be 1100 0001 since a 1 in bit 0 will exclude slave 0. Both slaves can be selected at the same time by an address which has bit 0 = 0 (for slave 0) and bit 1 = 0 (for slave 1). Thus, both could be addressed with 1100 0000.

In a more complex system the following could be used to select slaves 1 and 2 while excluding slave 0:

Slave 0	SADDR	=	1100 0000
	SADEN	=	1111 1001
	Given	=	1100 0XX0
Slave 1	SADDR	=	1110 0000
	SADEN	=	1111 1010
	Given	=	1110 0X0X
Slave 2	SADDR	=	1110 0000
	SADEN	=	1111 1100
	Given	=	1110 00XX

In the above example the differentiation among the 3 slaves is in the lower 3 address bits. Slave 0 requires that bit 0 = 0 and it can be uniquely addressed by 1110 0110. Slave 1 requires that bit 1 = 0 and it can be uniquely addressed by 1110 and 0101. Slave 2 requires that bit 2 = 0 and its unique address is 1110 0011. To select Slaves 0

and 1 and exclude Slave 2 use address 1110 0100, since it is necessary to make bit 2 = 1 to exclude slave 2.

The Broadcast Address for each slave is created by taking the logical OR of SADDR and SADEN. Zeros in this result are trended as don't-cares. In most cases, interpreting the don't-cares as ones, the broadcast address will be FF hexadecimal.

Upon reset SADDR (SFR address 0A9H) and SADEN (SFR address 0B9H) are leaded with 0s. This produces a given address of all "don't cares" as well as a Broadcast address of all "don't cares". This effectively disables the Automatic Addressing mode and allows the microcontroller to use standard 80C51 type UART drivers which do not make use of this feature.

SCON Address = 98H — Reset Value = 0000 0000B

Bit Addressable

	SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI
Bit:	7	6	5	4	3	2	1	0
	(SMOD0 = 0/1)*							

Symbol	Function
FE	Framing Error bit. This bit is set by the receiver when an invalid stop bit is detected. The FE bit is not cleared by valid frames but should be cleared by software. The SMOD0 bit must be set to enable access to the FE bit.
SM0	Serial Port Mode Bit 0, (SMOD0 must = 0 to access bit SM0)
SM1	Serial Port Mode Bit 1

SM0	SM1	Mode	Description	Baud Rate**
0	0	0	shift register	$f_{OSC}/12$
0	1	1	8-bit UART	variable
1	0	2	9-bit UART	$f_{OSC}/64$ or $f_{OSC}/32$
1	1	3	9-bit UART	variable

SM2	Enables the Automatic Address Recognition feature in Modes 2 or 3. If SM2 = 1 then RI will not be set unless the received 9th data bit (RB8) is 1, indicating an address, and the received byte is a Given or Broadcast Address. In Mode 1, if SM2 = 1 then RI will not be activated unless a valid stop bit was received, and the received byte is a Given or Broadcast Address. In Mode 0, SM2 should be 0.
REN	Enables serial reception. Set by software to enable reception. Clear by software to disable reception.
TB8	The 9th data bit that will be transmitted in Modes 2 and 3. Set or clear by software as desired.
RB8	In modes 2 and 3, the 9th data bit that was received. In Mode 1, if SM2 = 0, RB8 is the stop bit that was received. In Mode 0, RB8 is not used.
TI	Transmit interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or at the beginning of the stop bit in the other modes, in any serial transmission. Must be cleared by software.
RI	Receive interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or halfway through the stop bit time in the other modes, in any serial reception (except see SM2). Must be cleared by software.

NOTE:
*SMOD0 is located at PCON6.
**f_{OSC} = oscillator frequency

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Figure 7. SCON: Serial Port Control Register

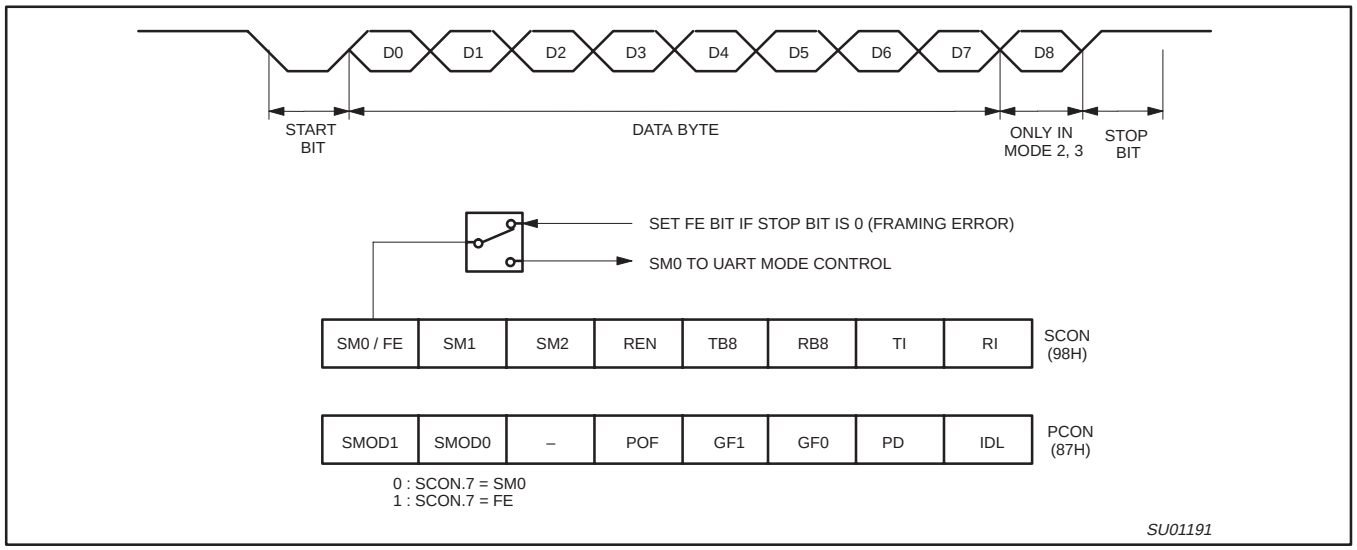


Figure 8. UART Framing Error Detection

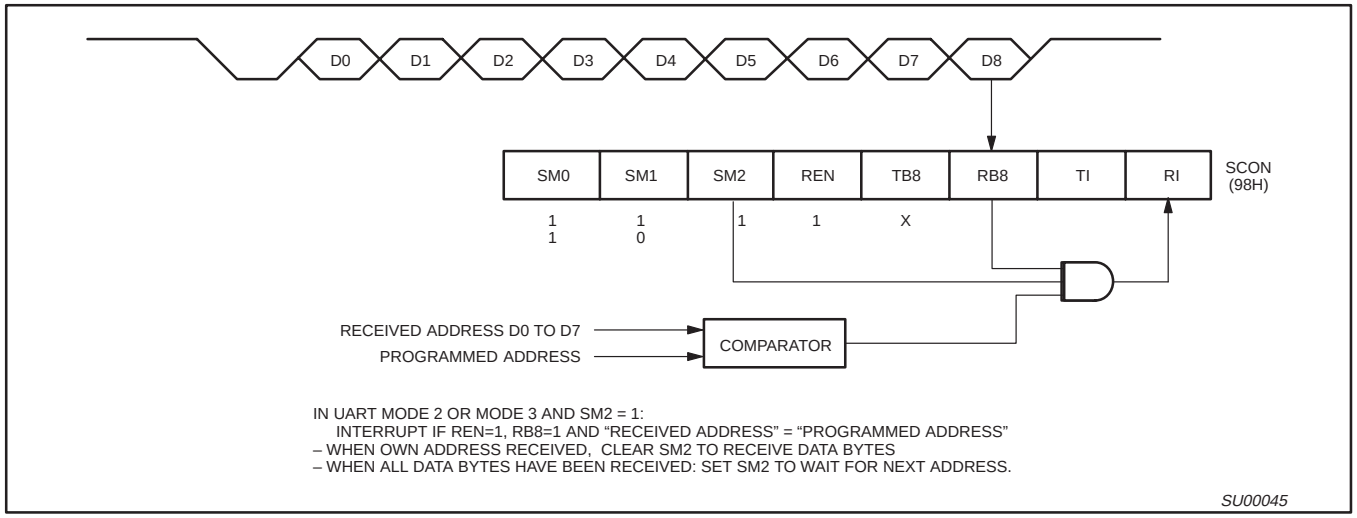


Figure 9. UART Multiprocessor Communication, Automatic Address Recognition

	7	6	5	4	3	2	1	0
IP (0B8H)	—	—	PT2	PS	PT1	PX1	PT0	PX0

Priority Bit = 1 assigns higher priority
Priority Bit = 0 assigns lower priority

BIT	SYMBOL	FUNCTION
IP.7	—	Not implemented, reserved for future use.
IP.6	—	Not implemented, reserved for future use.
IP.5	PT2	Timer 2 interrupt priority bit.
IP.4	PS	Serial Port interrupt priority bit.
IP.3	PT1	Timer 1 interrupt priority bit.
IP.2	PX1	External interrupt 1 priority bit.
IP.1	PT0	Timer 0 interrupt priority bit.
IP.0	PX0	External interrupt 0 priority bit.

SU00572

Figure 11. IP Registers

	7	6	5	4	3	2	1	0
IPH (B7H)	—	—	PT2H	PSH	PT1H	PX1H	PT0H	PX0H

Priority Bit = 1 assigns higher priority
Priority Bit = 0 assigns lower priority

BIT	SYMBOL	FUNCTION
IPH.7	—	Not implemented, reserved for future use.
IPH.6	—	Not implemented, reserved for future use.
IPH.5	PT2H	Timer 2 interrupt priority bit high.
IPH.4	PSH	Serial Port interrupt priority bit high.
IPH.3	PT1H	Timer 1 interrupt priority bit high.
IPH.2	PX1H	External interrupt 1 priority bit high.
IPH.1	PT0H	Timer 0 interrupt priority bit high.
IPH.0	PX0H	External interrupt 0 priority bit high.

SU01058

Figure 12. IPH Registers

Reduced EMI

All port pins of the 8xC51 and 8xC52 have slew rate controlled outputs. This is to limit noise generated by quickly switching output signals. The slew rate is factory set to approximately 10 ns rise and fall times.

Reduced EMI Mode

The AO bit (AUXR.0) in the AUXR register when set disables the ALE output.

AUXR (8EH)

7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	AO

AUXR.0 AO Turns off ALE output.

Dual DPTR

The dual DPTR structure (see Figure 13) enables a way to specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1/bit0 that allows the program code to switch between them.

- New Register Name: AUXR1#
- SFR Address: A2H
- Reset Value: xxx000x0B

AUXR1 (A2H)

7	6	5	4	3	2	1	0
–	–	–	LPEP	WUPD	0	–	DPS

Where:
DPS = AUXR1/bit0 = Switches between DPTR0 and DPTR1.

Select Reg	DPS
DPTR0	0
DPTR1	1

The DPS bit status should be saved by software when switching between DPTR0 and DPTR1.

Note that bit 2 is not writable and is always read as a zero. This allows the DPS bit to be quickly toggled simply by executing an INC DPTR instruction without affecting the WOPD or LPEP bits.

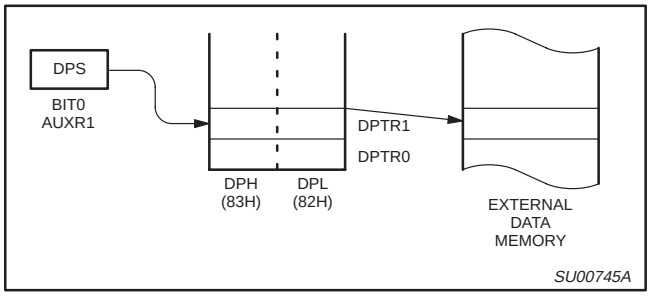


Figure 13.

DPTR Instructions

The instructions that refer to DPTR refer to the data pointer that is currently selected using the AUXR1/bit 0 register. The six instructions that use the DPTR are as follows:

INC DPTR	Increments the data pointer by 1
MOV DPTR, #data16	Loads the DPTR with a 16-bit constant
MOV A, @ A+DPTR	Move code byte relative to DPTR to ACC
MOVX A, @ DPTR	Move external RAM (16-bit address) to ACC
MOVX @ DPTR , A	Move ACC to external RAM (16-bit address)
JMP @ A + DPTR	Jump indirect relative to DPTR

The data pointer can be accessed on a byte-by-byte basis by specifying the low or high byte in an instruction which accesses the SFRs. See application note AN458 for more details.

DC ELECTRICAL CHARACTERISTICS

T_{amb} = 0°C to +70°C or –40°C to +85°C, V_{CC} = 2.7 V to 5.5 V, V_{SS} = 0 V (16 MHz devices)

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP[1]	MAX	
V_{IL}	Input low voltage[11]	4.0 V < V_{CC} < 5.5 V	–0.5		0.2 V_{CC}–0.1	V
		2.7 V<V_{CC}< 4.0 V	–0.5		0.7	V
V_{IH}	Input high voltage (ports 0, 1, 2, 3, $\overline{EA}$)		0.2 V_{CC}+0.9		V_{CC}+0.5	V
V_{IH1}	Input high voltage, XTAL1, RST[11]		0.7 V_{CC}		V_{CC}+0.5	V
V_{OL}	Output low voltage, ports 1, 2, [8]	V_{CC} = 2.7 V I_{OL} = 1.6 mA[2]			0.4	V
V_{OL1}	Output low voltage, port 0, ALE, $\overline{PSEN}$[8, 7]	V_{CC} = 2.7 V I_{OL} = 3.2 mA[2]			0.4	V
V_{OH}	Output high voltage, ports 1, 2, 3 [3]	V_{CC} = 2.7 V I_{OH} = –20 µA	V_{CC} – 0.7			V
		V_{CC} = 4.5 V I_{OH} = –30 µA	V_{CC} – 0.7			V
V_{OH1}	Output high voltage (port 0 in external bus mode), ALE[9], $\overline{PSEN}$[3]	V_{CC} = 2.7 V I_{OH} = –3.2 mA	V_{CC} – 0.7			V
I_{IL}	Logical 0 input current, ports 1, 2, 3	V_{IN} = 0.4 V	–1		–50	µA
I_{TL}	Logical 1-to-0 transition current, ports 1, 2, 3[6]	V_{IN} = 2.0 V See note 4			–650	µA
I_{LI}	Input leakage current, port 0	0.45 < V_{IN} < V_{CC} – 0.3			±10	µA
I_{CC}	Power supply current (see Figure 21): Active mode @ 16 MHz Idle mode @ 16 MHz Power-down mode or clock stopped (see Figure 25 for conditions)	See note 5 T_{amb} = 0°C to 70°C T_{amb} = –40°C to +85°C		 3	 50 75	 µA µA µA µA
R_{RST}	Internal reset pull-down resistor		40		225	kΩ
C_{IO}	Pin capacitance[10] (except $\overline{EA}$)				15	pF

NOTES:

1. Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
2. Capacitive loading on ports 0 and 2 may cause spurious noise to be superimposed on the V_{OL}s of ALE and ports 1 and 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operations. In the worst cases (capacitive loading > 100 pF), the noise pulse on the ALE pin may exceed 0.8 V. In such cases, it may be desirable to qualify ALE with a Schmitt Trigger, or use an address latch with a Schmitt Trigger STROBE input. I_{OL} can exceed these conditions provided that no single output sinks more than 5 mA and no more than two outputs exceed the test conditions.
3. Capacitive loading on ports 0 and 2 may cause the V_{OH} on ALE and $\overline{PSEN}$ to momentarily fall below the V_{CC}–0.7 specification when the address bits are stabilizing.
4. Pins of ports 1, 2 and 3 source a transition current when they are being externally driven from 1 to 0. The transition current reaches its maximum value when V_{IN} is approximately 2 V.
5. See Figures 22 through 25 for I_{CC} test conditions.
   Active mode: I_{CC} = 0.9 × FREQ. + 1.1 mA
   Idle mode: I_{CC} = 0.18 × FREQ. +1.01 mA; See Figure 21.
6. This value applies to T_{amb} = 0°C to +70°C. For T_{amb} = –40°C to +85°C, I_{TL} = –750 µA.
7. Load capacitance for port 0, ALE, and $\overline{PSEN}$ = 100 pF, load capacitance for all other outputs = 80 pF.
8. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 15 mA (*NOTE: This is 85°C specification.)
   Maximum I_{OL} per 8-bit port: 26 mA
   Maximum total I_{OL} for all outputs: 71 mA
   If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
9. ALE is tested to V_{OH1}, except when ALE is off then V_{OH} is the voltage specification.
10. Pin capacitance is characterized but not tested. Pin capacitance is less than 25 pF. Pin capacitance of ceramic package is less than 15 pF (except $\overline{EA}$ is 25 pF).
11. To improve noise rejection a nominal 100 ns glitch rejection circuitry has been added to the RST pin, and a nominal 15 ns glitch rejection circuitry has been added to the $\overline{INT0}$ and $\overline{INT1}$ pins. Previous devices provided only an inherent 5 ns of glitch rejection.

AC ELECTRICAL CHARACTERISTICS

T_{amb} = 0°C to +70°C or –40°C to +85°C, V_{CC} = +2.7 V to +5.5 V, V_{SS} = 0 V[1, 2, 3]

SYMBOL	FIGURE	PARAMETER	16 MHz CLOCK		VARIABLE CLOCK		UNIT
			MIN	MAX	MIN	MAX	
$1/t_{CLCL}$	14	Oscillator frequency[5] Speed versions :S			3.5	16	MHz
t_{LHLL}	14	ALE pulse width	85		$2t_{CLCL}$–40		ns
t_{AVLL}	14	Address valid to ALE low	22		t_{CLCL}–40		ns
t_{LLAX}	14	Address hold after ALE low	32		t_{CLCL}–30		ns
t_{LLIV}	14	ALE low to valid instruction in		150		$4t_{CLCL}$–100	ns
t_{LLPL}	14	ALE low to $\overline{PSEN}$ low	32		t_{CLCL}–30		ns
t_{PLPH}	14	$\overline{PSEN}$ pulse width	142		$3t_{CLCL}$–45		ns
t_{PLIV}	14	$\overline{PSEN}$ low to valid instruction in		82		$3t_{CLCL}$–105	ns
t_{PXIX}	14	Input instruction hold after $\overline{PSEN}$	0		0		ns
t_{PXIZ}	14	Input instruction float after $\overline{PSEN}$		37		t_{CLCL}–25	ns
t_{AVIV} [4]	14	Address to valid instruction in		207		$5t_{CLCL}$–105	ns
t_{PLAZ}	14	$\overline{PSEN}$ low to address float		10		10	ns
Data Memory							
t_{RLRH}	15, 16	$\overline{RD}$ pulse width	275		$6t_{CLCL}$–100		ns
t_{WLWH}	15, 16	$\overline{WR}$ pulse width	275		$6t_{CLCL}$–100		ns
t_{RLDV}	15, 16	$\overline{RD}$ low to valid data in		147		$5t_{CLCL}$–165	ns
t_{RHDX}	15, 16	Data hold after $\overline{RD}$	0		0		ns
t_{RHDZ}	15, 16	Data float after $\overline{RD}$		65		$2t_{CLCL}$–60	ns
t_{LLDV}	15, 16	ALE low to valid data in		350		$8t_{CLCL}$–150	ns
t_{AVDV}	15, 16	Address to valid data in		397		$9t_{CLCL}$–165	ns
t_{LLWL}	15, 16	ALE low to $\overline{RD}$ or $\overline{WR}$ low	137	239	$3t_{CLCL}$–50	$3t_{CLCL}$+50	ns
t_{AVWL}	15, 16	Address valid to $\overline{WR}$ low or $\overline{RD}$ low	122		$4t_{CLCL}$–130		ns
t_{QVWX}	15, 16	Data valid to $\overline{WR}$ transition	13		t_{CLCL}–50		ns
t_{WHQX}	15, 16	Data hold after $\overline{WR}$	13		t_{CLCL}–50		ns
t_{QVWH}	16	Data valid to $\overline{WR}$ high	287		$7t_{CLCL}$–150		ns
t_{RLAZ}	15, 16	$\overline{RD}$ low to address float		0		0	ns
t_{WHLH}	15, 16	$\overline{RD}$ or $\overline{WR}$ high to ALE high	23	103	t_{CLCL}–40	t_{CLCL}+40	ns
External Clock							
t_{CHCX}	18	High time	20		20	t_{CLCL}–t_{CLCX}	ns
t_{CLCX}	18	Low time	20		20	t_{CLCL}–t_{CHCX}	ns
t_{CLCH}	18	Rise time		20		20	ns
t_{CHCL}	18	Fall time		20		20	ns
Shift Register							
t_{XLXL}	17	Serial port clock cycle time	750		$12t_{CLCL}$		ns
t_{QVXH}	17	Output data setup to clock rising edge	492		$10t_{CLCL}$–133		ns
t_{XHQX}	17	Output data hold after clock rising edge	8		$2t_{CLCL}$–117		ns
t_{XHDX}	17	Input data hold after clock rising edge	0		0		ns
t_{XHDV}	17	Clock rising edge to input data valid		492		$10t_{CLCL}$–133	ns

NOTES:
1. Parameters are valid over operating temperature range unless otherwise specified.
2. Load capacitance for port 0, ALE, and $\overline{PSEN}$ = 100 pF, load capacitance for all other outputs = 80 pF.
3. Interfacing the 87C51, 80C51, 87C52, or 80C52 to devices with float times up to 45 ns is permitted. This limited bus contention will not cause damage to Port 0 drivers.
4. See application note AN457 for external memory interface.
5. Parts are guaranteed to operate down to 0 Hz. When an external clock source is used, the RST pin should be held high for a minimum of 20 μs for power-on or wakeup from power down.

EXPLANATION OF THE AC SYMBOLS

Each timing symbol has five characters. The first character is always 't' (= time). The other characters, depending on their positions, indicate the name of a signal or the logical status of that signal. The designations are:

A – Address
C – Clock
D – Input data
H – Logic level high
I – Instruction (program memory contents)
L – Logic level low, or ALE
P – $\overline{\text{PSEN}}$
Q – Output data
R – $\overline{\text{RD}}$ signal
t – Time
V – Valid
W – $\overline{\text{WR}}$ signal
X – No longer a valid logic level
Z – Float

Examples: t_{AVLL} = Time for address valid to ALE low.
t_{LLPL} =Time for ALE low to $\overline{\text{PSEN}}$ low.

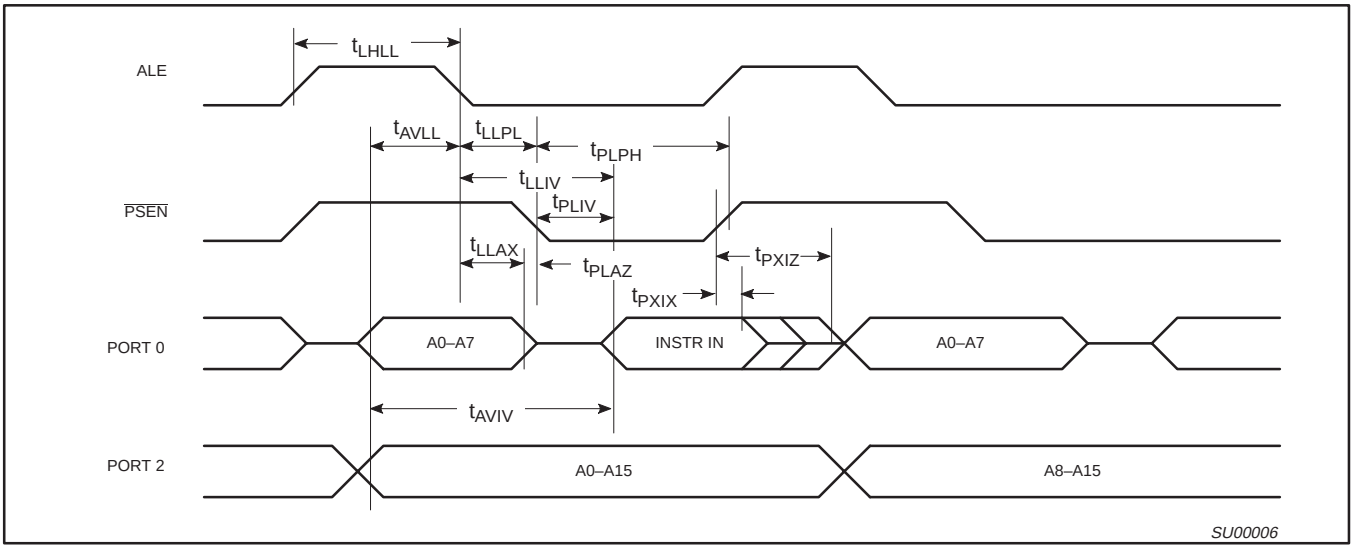


Figure 14. External Program Memory Read Cycle

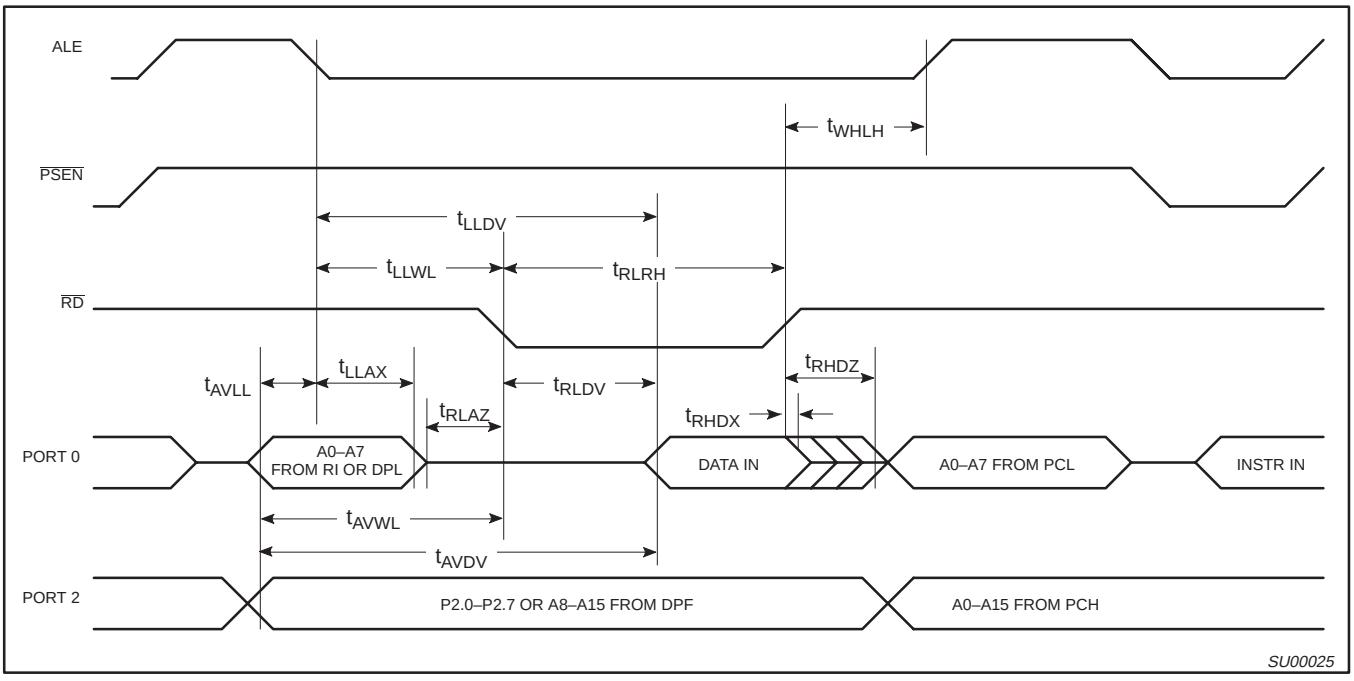


Figure 15. External Data Memory Read Cycle

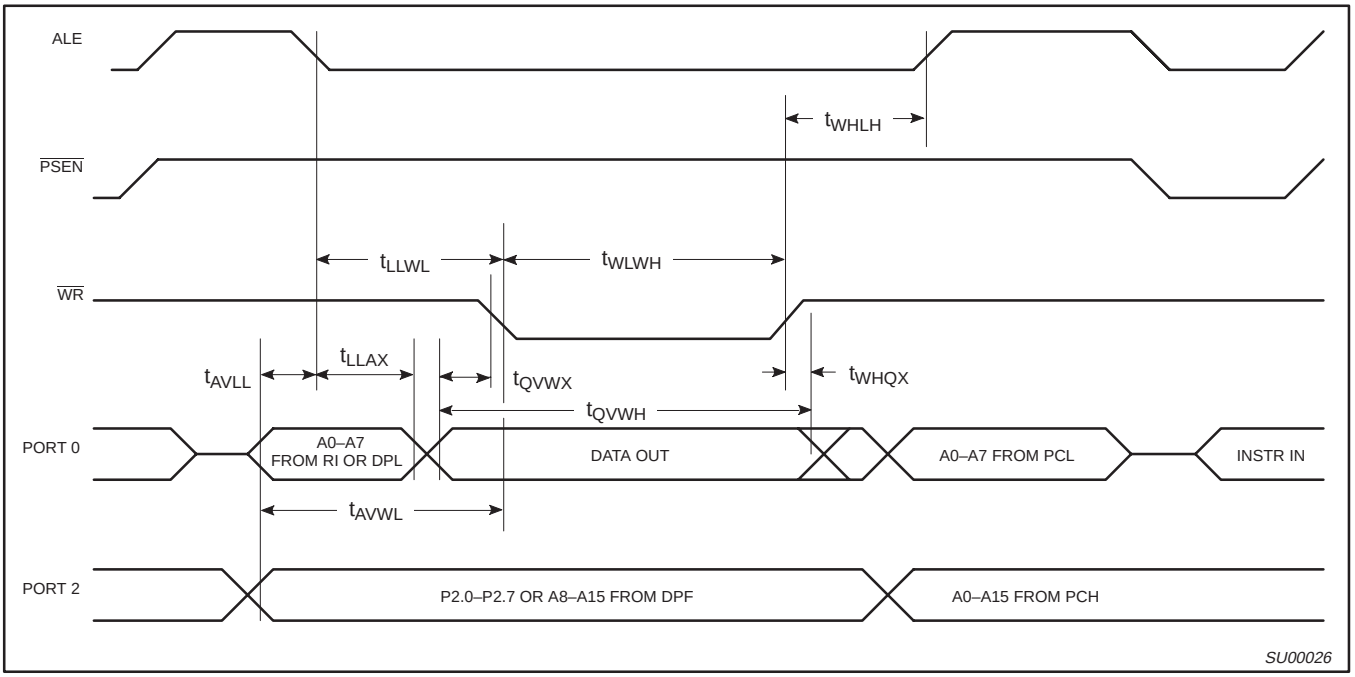


Figure 16. External Data Memory Write Cycle

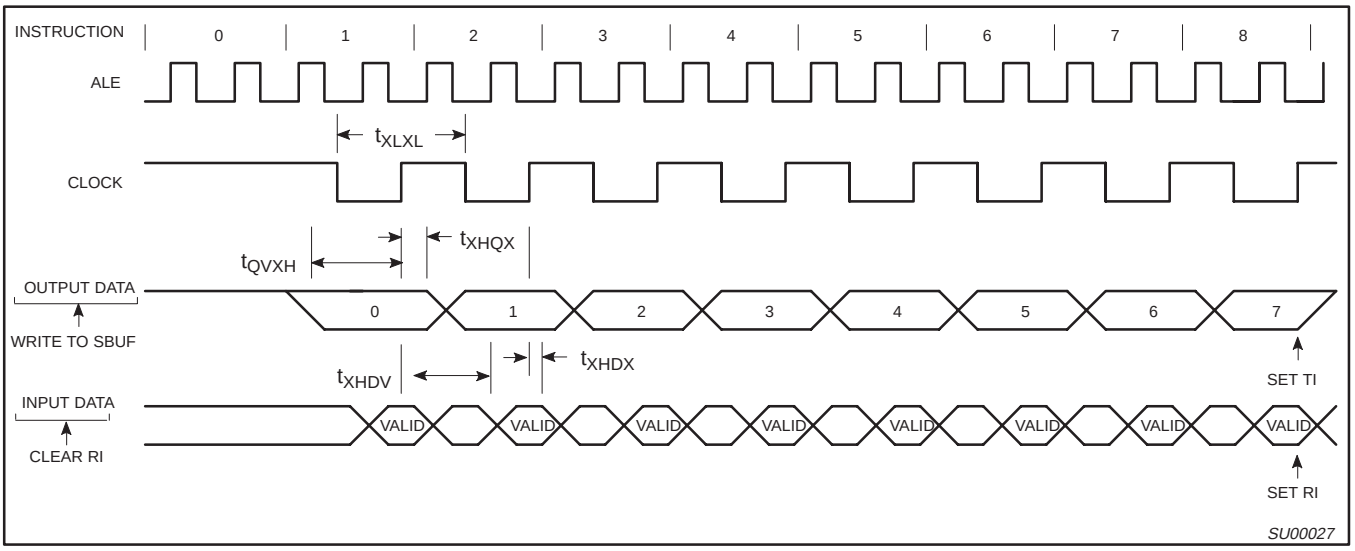


Figure 17. Shift Register Mode Timing

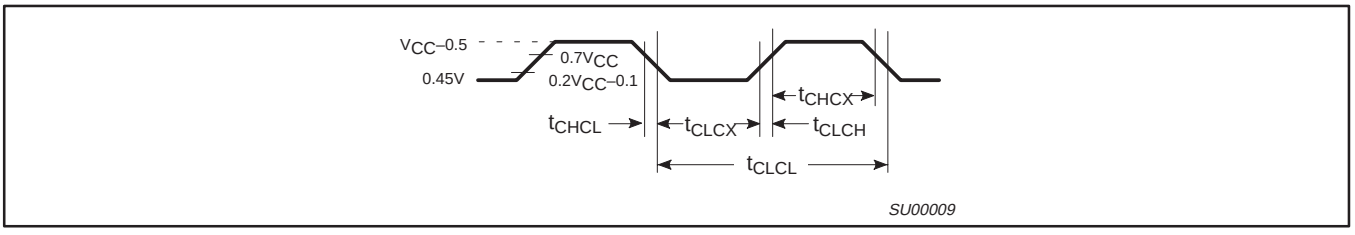


Figure 18. External Clock Drive

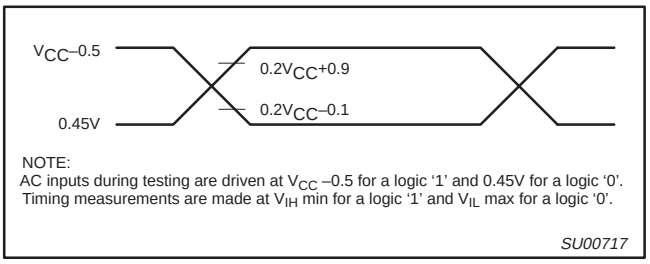


Figure 19. AC Testing Input/Output

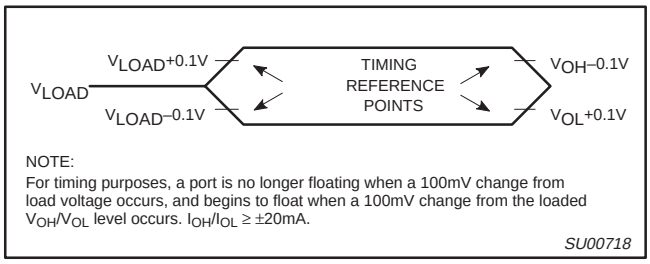


Figure 20. Float Waveform

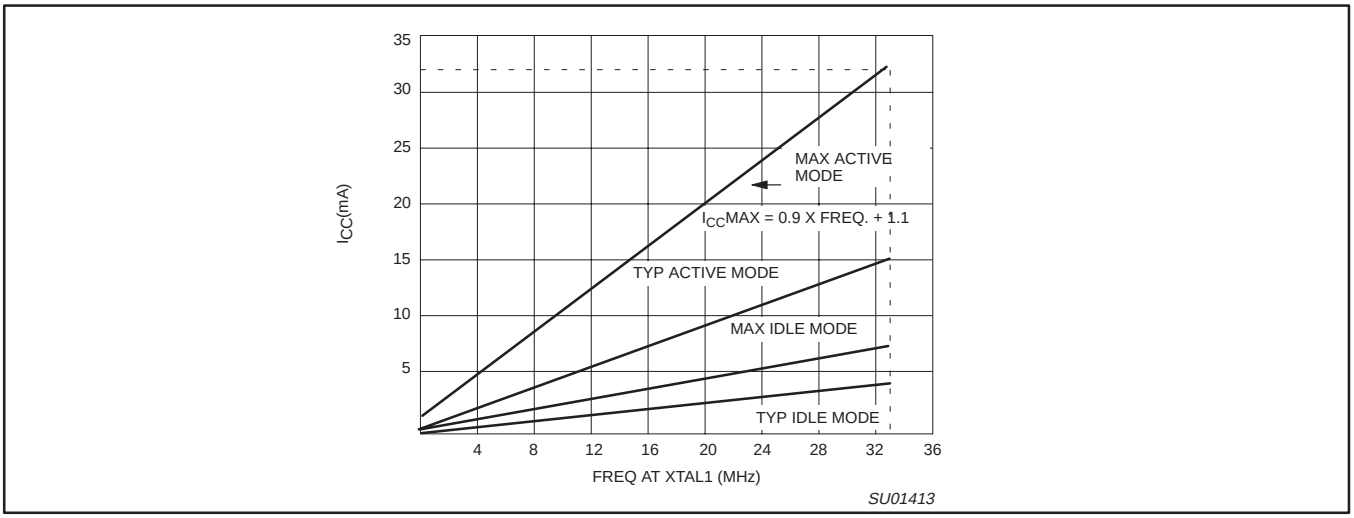


Figure 21. I_{CC} vs. FREQ
Valid only within frequency specifications of the device under test

EPROM CHARACTERISTICS

These devices can be programmed by using a modified Improved Quick-Pulse Programming™ algorithm. It differs from older methods in the value used for V_{PP} (programming supply voltage) and in the width and number of the ALE/$\overline{PROG}$ pulses.

The family contains two signature bytes that can be read and used by an EPROM programming system to identify the device. The signature bytes identify the device as being manufactured by Philips.

Table 8 shows the logic levels for reading the signature byte, and for programming the program memory, the encryption table, and the security bits. The circuit configuration and waveforms for quick-pulse programming are shown in Figures 26 and 27. Figure 28 shows the circuit configuration for normal program memory verification.

Quick-Pulse Programming

The setup for microcontroller quick-pulse programming is shown in Figure 26. Note that the device is running with a 4 to 6 MHz oscillator. The reason the oscillator needs to be running is that the device is executing internal address and program data transfers.

The address of the EPROM location to be programmed is applied to ports 1 and 2, as shown in Figure 26. The code byte to be programmed into that location is applied to port 0. RST, $\overline{PSEN}$ and pins of ports 2 and 3 specified in Table 8 are held at the 'Program Code Data' levels indicated in Table 8. The ALE/$\overline{PROG}$ is pulsed low 5 times as shown in Figure 27.

To program the encryption table, repeat the 5 pulse programming sequence for addresses 0 through 1FH, using the 'Pgm Encryption Table' levels. Do not forget that after the encryption table is programmed, verification cycles will produce only encrypted data.

To program the security bits, repeat the 5 pulse programming sequence using the 'Pgm Security Bit' levels. After one security bit is programmed, further programming of the code memory and encryption table is disabled. However, the other security bits can still be programmed.

Note that the $\overline{EA}$/V_{PP} pin must not be allowed to go above the maximum specified V_{PP} level for any amount of time. Even a narrow glitch above that voltage can cause permanent damage to the device. The V_{PP} source should be well regulated and free of glitches and overshoot.

Program Verification

If security bits 2 and 3 have not been programmed, the on-chip program memory can be read out for program verification. The address of the program memory locations to be read is applied to ports 1 and 2 as shown in Figure 28. The other pins are held at the 'Verify Code Data' levels indicated in Table 8. The contents of the address location will be emitted on port 0. External pull-ups are required on port 0 for this operation.

If the 64 byte encryption table has been programmed, the data presented at port 0 will be the exclusive NOR of the program byte with one of the encryption bytes. The user will have to know the encryption table contents in order to correctly decode the verification data. The encryption table itself cannot be read out.

Reading the Signature Bytes

The signature bytes are read by the same procedure as a normal verification of locations 030H and 031H, except that P3.6 and P3.7 need to be pulled to a logic low. The values are:
(030H) = 15H indicates manufactured by Philips
(031H) = 92H indicates 87C51

Program/Verify Algorithms

Any algorithm in agreement with the conditions listed in Table 8, and which satisfies the timing specifications, is suitable.

Security Bits

With none of the security bits programmed the code in the program memory can be verified. If the encryption table is programmed, the code will be encrypted when verified. When only security bit 1 (see Table 9) is programmed, MOVC instructions executed from external program memory are disabled from fetching code bytes from the internal memory, EA is latched on Reset and all further programming of the EPROM is disabled. When security bits 1 and 2 are programmed, in addition to the above, verify mode is disabled. When all three security bits are programmed, all of the conditions above apply and all external program memory execution is disabled.

Encryption Array

64 bytes of encryption array are initially unprogrammed (all 1s).

Table 8. EPROM Programming Modes

MODE	RST	$\overline{PSEN}$	ALE/$\overline{PROG}$	$\overline{EA}$/V_{PP}	P2.7	P2.6	P3.7	P3.6
Read signature	1	0	1	1	0	0	0	0
Program code data	1	0	0*	V_{PP}	1	0	1	1
Verify code data	1	0	1	1	0	0	1	1
Pgm encryption table	1	0	0*	V_{PP}	1	0	1	0
Pgm security bit 1	1	0	0*	V_{PP}	1	1	1	1
Pgm security bit 2	1	0	0*	V_{PP}	1	1	0	0
Pgm security bit 3	1	0	0*	V_{PP}	0	1	0	1

NOTES:
1. '0' = Valid low for that pin, '1' = valid high for that pin.
2. V_{PP} = 12.75 V ±0.25 V.
3. V_{CC} = 5 V±10% during programming and verification.

* ALE/$\overline{PROG}$ receives 5 programming pulses for code data (also for user array; 5 pulses for encryption or security bits) while V_{PP} is held at 12.75 V. Each programming pulse is low for 100 µs (±10 µs) and high for a minimum of 10 µs.

™Trademark phrase of Intel Corporation.

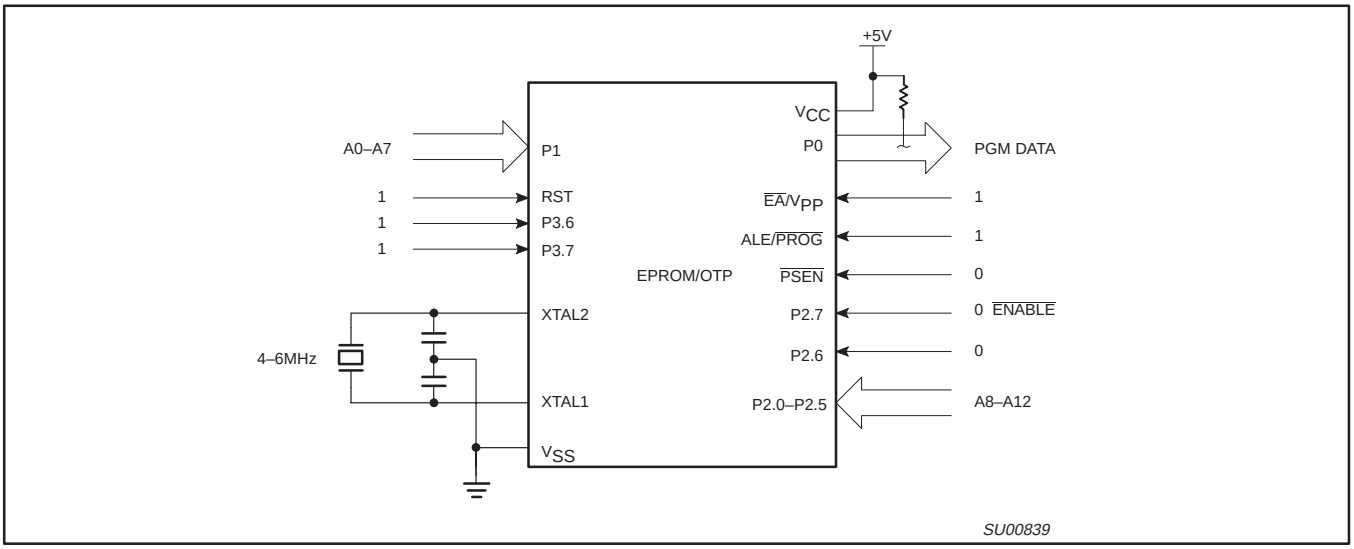


Figure 28. Program Verification

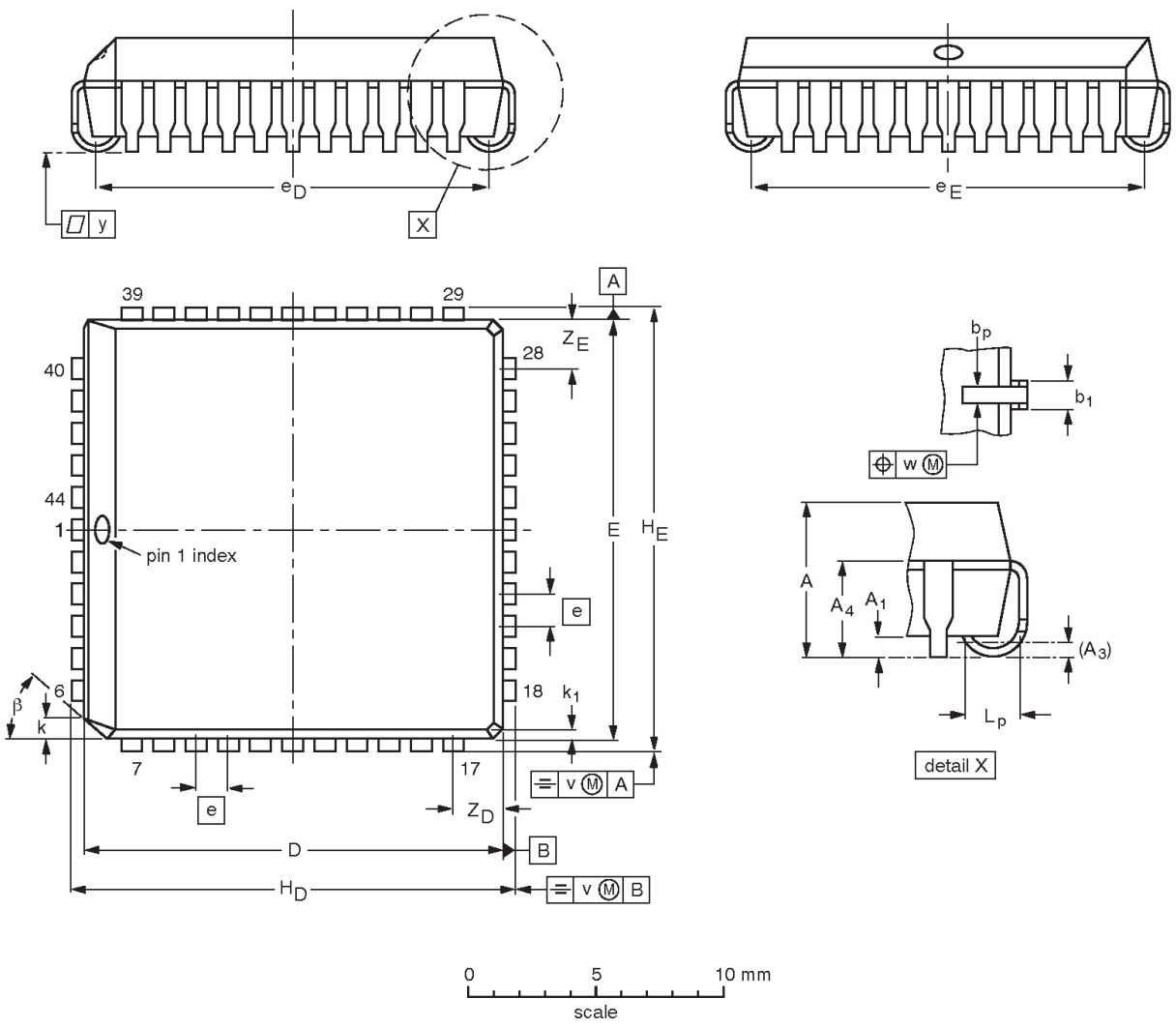
EPROM PROGRAMMING AND VERIFICATION CHARACTERISTICS

T_{amb} = 21°C to +27°C, V_{CC} = 5 V±10%, V_{SS} = 0 V (See Figure 29)

SYMBOL	PARAMETER	MIN	MAX	UNIT
V_{PP}	Programming supply voltage	12.5	13.0	V
I_{PP}	Programming supply current		50 [1]	mA
$1/t_{CLCL}$	Oscillator frequency	4	6	MHz
t_{AVGL}	Address setup to $\overline{PROG}$ low	$48t_{CLCL}$		
t_{GHAX}	Address hold after $\overline{PROG}$	$48t_{CLCL}$		
t_{DVGL}	Data setup to $\overline{PROG}$ low	$48t_{CLCL}$		
t_{GHDX}	Data hold after $\overline{PROG}$	$48t_{CLCL}$		
t_{EHSH}	P2.7 ($\overline{ENABLE}$) high to V_{PP}	$48t_{CLCL}$		
t_{SHGL}	V_{PP} setup to $\overline{PROG}$ low	10		µs
t_{GHSL}	V_{PP} hold after $\overline{PROG}$	10		µs
t_{GLGH}	$\overline{PROG}$ width	90	110	µs
t_{AVQV}	Address to data valid		$48t_{CLCL}$	
t_{ELQZ}	$\overline{ENABLE}$ low to data valid		$48t_{CLCL}$	
t_{EHQZ}	Data float after $\overline{ENABLE}$	0	$48t_{CLCL}$	
t_{GHGL}	$\overline{PROG}$ high to $\overline{PROG}$ low	10		µs

NOTE:
1. Not tested.

PLCC44: plastic leaded chip carrier; 44 leads SOT187-2



DIMENSIONS (millimetre dimensions are derived from the original inch dimensions)

UNIT	A	A_1 min.	A_3	A_4 max.	b_p	b_1	$D^{(1)}$	$E^{(1)}$	e	e_D	e_E	H_D	H_E	k	k_1 max.	L_p	v	w	y	$Z_D^{(1)}$ max.	$Z_E^{(1)}$ max.	β
mm	4.57 4.19	0.51	0.25	3.05	0.53 0.33	0.81 0.66	16.66 16.51	16.66 16.51	1.27	16.00 14.99	16.00 14.99	17.65 17.40	17.65 17.40	1.22 1.07	0.51	1.44 1.02	0.18	0.18	0.10	2.16	2.16	45^o
inches	0.180 0.165	0.020	0.01	0.12	0.021 0.013	0.032 0.026	0.656 0.650	0.656 0.650	0.05	0.630 0.590	0.630 0.590	0.695 0.685	0.695 0.685	0.048 0.042	0.020	0.057 0.040	0.007	0.007	0.004	0.085	0.085	

Note

1. Plastic or metal protrusions of 0.01 inches maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT187-2	112E10	MO-047				~~97-12-16~~ 99-12-27