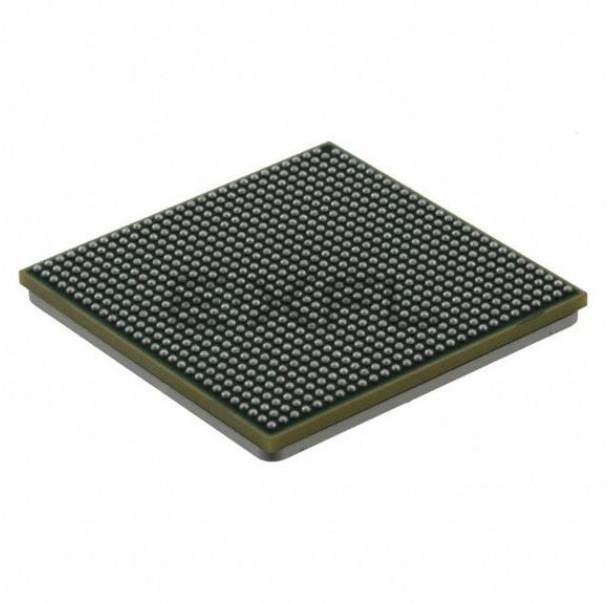

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8544evtaqg

Table 2. Recommended Operating Conditions (continued)

Characteristic		Symbol	Recommended Value	Unit	Notes
Three-speed Ethernet I/O voltage		LV_{DD} (eTSEC1)	3.3 V ± 165 mV 2.5 V ± 125 mV	V	4
		TV_{DD} (eTSEC3)	3.3 V ± 165 mV 2.5 V ± 125 mV		
PCI, DUART, PCI Express, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	3.3 V ± 165 mV	V	3
Local bus I/O voltage		BV_{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV 1.8 V ± 90 mV	V	5
Input voltage	DDR and DDR2 DRAM signals	MV_{IN}	GND to GV_{DD}	V	2
	DDR and DDR2 DRAM reference	MV_{REF}	GND to $GV_{DD}/2$	V	2
	Three-speed Ethernet signals	LV_{IN} TV_{IN}	GND to LV_{DD} GND to TV_{DD}	V	4
	Local bus signals	BV_{IN}	GND to BV_{DD}	V	5
	PCI, Local bus, DUART, SYSCLK, system control and power management, I^2C, and JTAG signals	OV_{IN}	GND to OV_{DD}	V	3
Junction temperature range		T_j	0 to 105	°C	—

Notes:
1. This voltage is the input to the filter discussed in Section 21.2, "PLL Power Supply Filtering," and not necessarily the voltage at the AV_{DD} pin, which may be reduced from V_{DD} by the filter.
2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
3. **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. **Caution:** T/LV_{IN} must not exceed T/ LV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
5. **Caution:** BV_{IN} must not exceed BV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

4.2 Real-Time Clock Timing

The RTC input is sampled by the platform clock (CCB clock). The output of the sampling latch is then used as an input to the counters of the PIC and the TimeBase unit of the e500. There is no jitter specification. The minimum pulse width of the RTC signal should be greater than 2 × the period of the CCB clock. That is, minimum clock high time is $2 \times t_{CCB}$, and minimum clock low time is $2 \times t_{CCB}$. There is no minimum RTC frequency; RTC may be grounded if not needed.

4.3 eTSEC Gigabit Reference Clock Timing

Table 7 provides the eTSEC gigabit reference clocks (EC_GTX_CLK125) AC timing specifications for the MPC8544E.

Table 7. EC_GTX_CLK125 AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Notes
EC_GTX_CLK125 frequency	f_{G125}	—	125	—	MHz	—
EC_GTX_CLK125 cycle time	t_{G125}	—	8	—	ns	—
EC_GTX_CLK rise and fall time LV_{DD}, TV_{DD} = 2.5 V LV_{DD}, TV_{DD} = 3.3 V	t_{G125R}/t_{G125F}	—	—	 0.75 1.0	ns	1
EC_GTX_CLK125 duty cycle GMII, TBI 1000Base-T for RGMII, RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	2

Notes:

1. Rise and fall times for EC_GTX_CLK125 are measured from 0.5 and 2.0 V for L/TV_{DD} = 2.5 V, and from 0.6 and 2.7 V for L/TVDD = 3.3 V.
2. EC_GTX_CLK125 is used to generate the GTX clock for the eTSEC transmitter with 2% degradation. EC_GTX_CLK125 duty cycle can be loosened from 47%/53% as long as the PHY device can tolerate the duty cycle generated by the eTSEC GTX_CLK. See Section 8.7.4, “RGMII and RTBI AC Timing Specifications,” for duty cycle for 10Base-T and 100Base-T reference clock.

4.4 Platform to FIFO Restrictions

Please note the following FIFO maximum speed restrictions based on platform speed.

For FIFO GMII mode:

FIFO TX/RX clock frequency ≤ platform clock frequency ÷ 4.2

For example, if the platform frequency is 533 MHz, the FIFO Tx/Rx clock frequency should be no more than 127 MHz.

For FIFO encoded mode:

FIFO TX/RX clock frequency ≤ platform clock frequency ÷ 3.2

For example, if the platform frequency is 533 MHz, the FIFO Tx/Rx clock frequency should be no more than 167 MHz.

Table 24. DC Transmitter Electrical Characteristics (continued)

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Output differential voltage[2,3,5]	$\lvert V_{OD}\rvert$	323	500	725	mV	Equalization setting: 1.0x
		296	459	665		Equalization setting: 1.09x
		269	417	604		Equalization setting: 1.2x
		243	376	545		Equalization setting: 1.33x
		215	333	483		Equalization setting: 1.5x
		189	292	424		Equalization setting: 1.71x
		162	250	362		Equalization setting: 2.0x
Output offset voltage	V_{OS}	425	500	577.5	mV	1, 4
Output impedance (single ended)	R_O	40	—	60	Ω	—
Mismatch in a pair	ΔR_O	—	—	10	%	—
Change in V_{OD} between 0 and 1	$\Delta\lvert V_{OD}\rvert$	—	—	25	mV	—
Change in V_{OS} between 0 and 1	ΔV_{OS}	—	—	25	mV	—
Output current on short to GND	I_{SA}, I_{SB}	—	—	40	mA	—

Notes:

1. This will not align to DC-coupled SGMII.
2. $\lvert V_{OD}\rvert = \lvert V_{SD2\_TXn} - V_{\overline{SD2\_TXn}}\rvert$. $\lvert V_{OD}\rvert$ is also referred as output differential peak voltage. $V_{TX\text{-}DIFFp\text{-}p} = 2^*\lvert V_{OD}\rvert$.
3. The $\lvert V_{OD}\rvert$ value shown in the table assumes the following transmit equalization setting in the XMITEQ**CD** (for SerDes 2 lane 2 and 3) bit field of MPC8544E SerDes 2 control register 1:
   - The MSbit (bit 0) of the above bit field is set to zero (selecting the full $V_{DD\text{-}DIFF\text{-}p\text{-}p}$ amplitude—power up default);
   - The LSbits (bit [1:3]) of the above bit field is set based on the equalization setting shown in this table.
4. V_{OS} is also referred to as output common mode voltage.
5. The $\lvert V_{OD}\rvert$ value shown in the Typ column is based on the condition of $XV_{DD\_SRDS2\text{-}Typ}$ = 1.0 V, no common mode offset variation (V_{OS} = 500 mV), SerDes2 transmitter is terminated with 100-Ω differential load between SD2_TX[n] and SD2_TX[n].

Table 33. MII Receive AC Timing Specifications (continued)
At recommended operating conditions with L/TVDD of 3.3 V ± 5%.or 2.5 V ± 5%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns	—
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns	—
RX_CLK clock rise (20%–80%)	t_{MRXR}	1.0	—	4.0	ns	—
RX_CLK clock fall time (80%–20%)	t_{MRXF}	1.0	—	4.0	ns	—

Note:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 17 provides the AC test load for eTSEC.

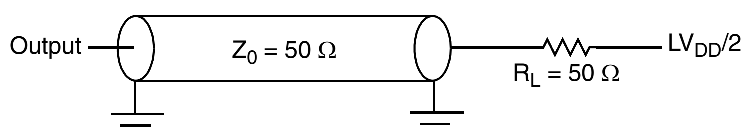


Figure 17. eTSEC AC Test Load

Figure 18 shows the MII receive AC timing diagram.

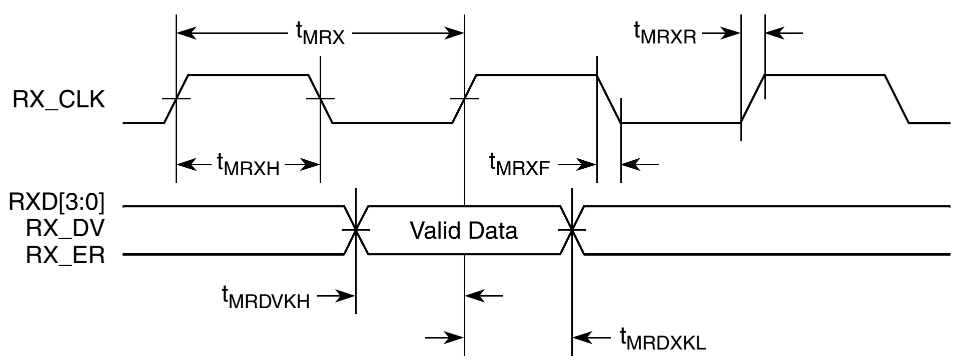


Figure 18. MII Receive AC Timing Diagram

8.7 TBI AC Timing Specifications

This section describes the TBI transmit and receive AC timing specifications.

9 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock). The electrical characteristics for GMII, RGMII, RMII, TBI, and RTBI are specified in "Section 8, "Enhanced Three-Speed Ethernet (eTSEC), MII Management."

9.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in Table 40.

Table 40. MII Management DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit	Notes
Supply voltage (3.3 V)	OV_{DD}	3.135	3.465	V	—
Output high voltage (OV_{DD} = Min, I_{OH} = –1.0 mA)	V_{OH}	2.10	3.60	V	—
Output low voltage (OV_{DD} = Min, I_{OL} = 1.0 mA)	V_{OL}	GND	0.50	V	—
Input high voltage	V_{IH}	1.95	—	V	—
Input low voltage	V_{IL}	—	0.90	V	—
Input high current (OV_{DD} = Max, V_{IN} = 2.1 V)	I_{IH}	—	40	μA	1
Input low current (OV_{DD} = Max, V_{IN} = 0.5 V)	I_{IL}	–600	—	μA	—

Note:
1. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

9.2 MII Management AC Electrical Specifications

Table 41 provides the MII management AC timing specifications.

Table 41. MII Management AC Timing Specifications

At recommended operating conditions with OV_{DD} is 3.3 V ± 5%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
MDC frequency	f_{MDC}	—	2.5	—	MHz	2
MDC period	t_{MDC}	—	400	—	ns	—
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	—
MDC to MDIO delay	t_{MDKHDX}	$(16 \times t_{plb\_clk}) - 3$	—	$(16 \times t_{plb\_clk}) + 3$	ns	3, 4
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	—
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	—
MDC rise time	t_{MDCR}	—	—	10	ns	—

Table 44. Local Bus DC Electrical Characteristics (1.8 V DC) (continued)

Parameter	Symbol	Min	Max	Unit	Notes
High-level output voltage (BV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	1.35	—	V	—
Low-level output voltage (BV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.45	V	—

10.2 Local Bus AC Electrical Specifications

Table 45 describes the general timing parameters of the local bus interface at BV_{DD} = 3.3 V. For information about the frequency range of local bus see Section 19.1, "Clock Ranges."

Table 45. Local Bus General Timing Parameters (BV_{DD} = 3.3 V)—PLL Enabled

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	7.5	12	ns	2
Local bus duty cycle	t_{LBKH}/t_{LBK}	43	57	%	—
LCLK[n] skew to LCLK[m] or LSYNC_OUT	$t_{LBKSKEW}$	—	150	ps	7, 8
Input setup to local bus clock (except LUPWAIT)	$t_{LBIVKH1}$	2.5	—	ns	3, 4
LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	1.85	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)	$t_{LBIXKH1}$	1.0	—	ns	3, 4
LUPWAIT input hold from local bus clock	$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output transition to LAD/LDP output transition (LATCH setup and hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	2.9	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	2.8	ns	—
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	2.7	ns	3
Local bus clock to LALE assertion	$t_{LBKHOV4}$	—	2.7	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	0.7	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	0.7	—	ns	3
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKHOZ1}$	—	2.5	ns	5

13.2 I^2C AC Electrical Specifications

Table 52 provides the AC timing parameters for the I^2C interfaces.

Table 52. I^2C AC Electrical Specifications

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 51).

Parameter	Symbol[1]	Min	Max	Unit	Notes
SCL clock frequency	f_{I2C}	0	400	kHz	—
Low period of the SCL clock	t_{I2CL}	1.3	—	μs	—
High period of the SCL clock	t_{I2CH}	0.6	—	μs	—
Setup time for a repeated START condition	t_{I2SVKH}	0.6	—	μs	—
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL}	0.6	—	μs	—
Data setup time	t_{I2DVKH}	100	—	ns	—
Data hold time: CBUS compatible masters I^2C bus devices	t_{I2DXKL}	 — 0	 — —	μs	2
Data output delay time	t_{I2OVKL}	—	0.9		3
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	μs	—
Rise time of both SDA and SCL signals	t_{I2CR}	20 + 0.1 C_b	300	ns	4
Fall time of both SDA and SCL signals	t_{I2CF}	20 + 0.1 C_b	300	ns	4
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	μs	—
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	0.1 × OV_{DD}	—	V	—
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	0.2 × OV_{DD}	—	V	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I^2C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I^2C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I^2C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The MPC8544E provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IH}min of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DXKL} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.

15.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus. Note that the SYSCLK signal is used as the PCI input clock. Table 56 provides the PCI AC timing specifications at 66 MHz.

Table 56. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol[1]	Min	Max	Unit	Notes
SYSCLK to output valid	t_{PCKHOV}	—	7.4	ns	2, 3
Output hold from SYSCLK	t_{PCKHOX}	2.0	—	ns	2
SYSCLK to output high impedance	t_{PCKHOZ}	—	14	ns	2, 4
Input setup to SYSCLK	t_{PCIVKH}	3.7	—	ns	2, 5
Input hold from SYSCLK	t_{PCIXKH}	0.5	—	ns	2, 5
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$[9] setup time	t_{PCRVRH}	$10 \times t_{SYS}$	—	clocks	6, 7
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	7
$\overline{\text{HRESET}}$ high to first $\overline{\text{FRAME}}$ assertion	t_{PCRHFV}	10	—	clocks	8
Rise time (20%–80%)	t_{PCICLK}	0.6	2.1	ns	—
Fall time (20%–80%)	t_{PCICLK}	0.6	2.1	ns	—

Notes:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the SYSCLK clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.2 Local Bus Specifications*.
3. All PCI signals are measured from $OV_{DD}/2$ of the rising edge of PCI_SYNC_IN to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V PCI signaling levels.
4. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Input timings are measured at the pin.
6. The timing parameter t_{SYS} indicates the minimum and maximum CLK cycle times for the various specified frequencies. The system clock period must be kept within the minimum and maximum defined ranges. For values see Section 19, “Clocking.”
7. The setup and hold time is with respect to the rising edge of $\overline{\text{HRESET}}$.
8. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI 2.2 Local Bus Specifications*.
9. The reset assertion timing requirement for $\overline{\text{HRESET}}$ is 100 μs.

Figure 41 provides the AC test load for PCI.

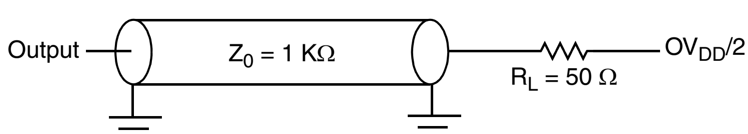


Figure 41. PCI AC Test Load

Figure 42 shows the PCI input AC timing conditions.

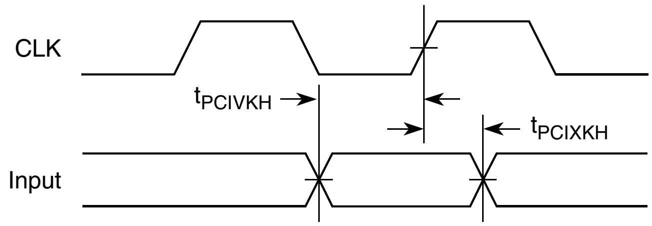


Figure 42. PCI Input AC Timing Measurement Conditions

Figure 43 shows the PCI output AC timing conditions.

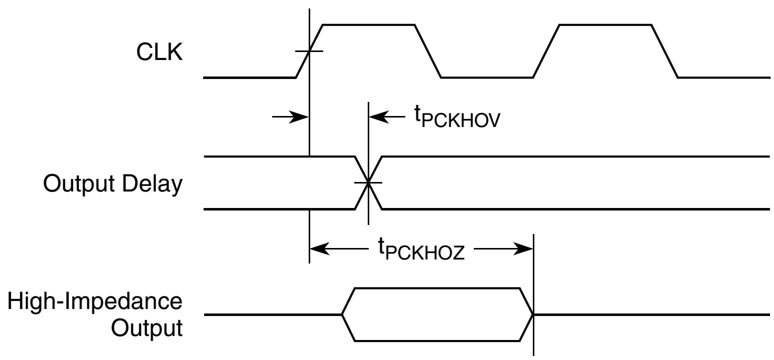


Figure 43. PCI Output AC Timing Measurement Condition

16 High-Speed Serial Interfaces (HSSI)

The MPC8544E features two serializer/deserializer (SerDes) interfaces to be used for high-speed serial interconnect applications.The SerDes1 dedicated for PCI Express data transfers. The SerDes2 can be used for PCI Express and/or SGMII application. This section describes the common portion of SerDes DC electrical specifications, which is the DC requirement for SerDes Reference Clocks. The SerDes data lane's transmitter and receiver reference circuits are also shown.

16.1 Signal Terms Definition

The SerDes utilizes differential signaling to transfer data across the serial link. This section defines terms used in the description and specification of differential signals.

Figure 44 shows how the signals are defined. For illustration purpose, only one SerDes lane is used for description. The figure shows waveform for either a transmitter output (SD*n*_TX and $\overline{\text{SD}n\text{\_TX}}$) or a receiver input (SD*n*_RX and $\overline{\text{SD}n\text{\_RX}}$). Each signal swings between A Volts and B Volts where A > B.

assumes that the LVPECL clock driver's output impedance is 50 Ω. R1 is used to DC-bias the LVPECL outputs prior to AC-coupling. Its value could be ranged from 140 to 240 Ω depending on clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the MPC8544E SerDes reference clock's differential input amplitude requirement (between 200 and 800 mV differential peak). For example, if the LVPECL output's differential peak is 900 mV and the desired SerDes reference clock input amplitude is selected as 600 mV, the attenuation factor is 0.67, which requires R2 = 25 Ω. Please consult clock driver chip manufacturer to verify whether this connection scheme is compatible with a particular clock driver chip.

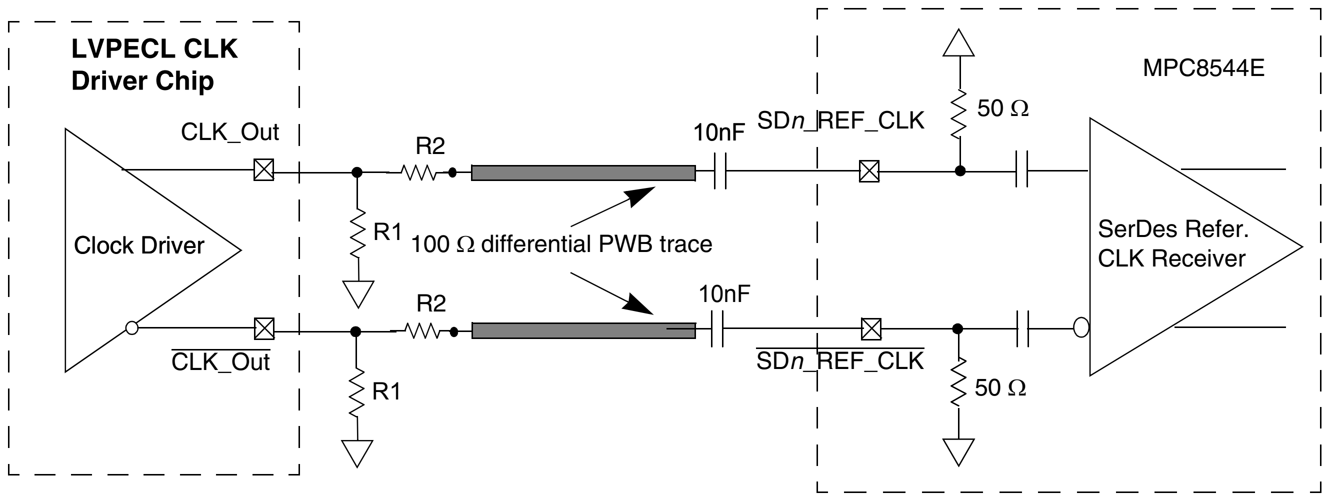


Figure 51. AC-Coupled Differential Connection with LVPECL Clock Driver (Reference Only)

Figure 52 shows the SerDes reference clock connection reference circuits for a single-ended clock driver. It assumes the DC levels of the clock driver are compatible with MPC8544E SerDes reference clock input's DC requirement.

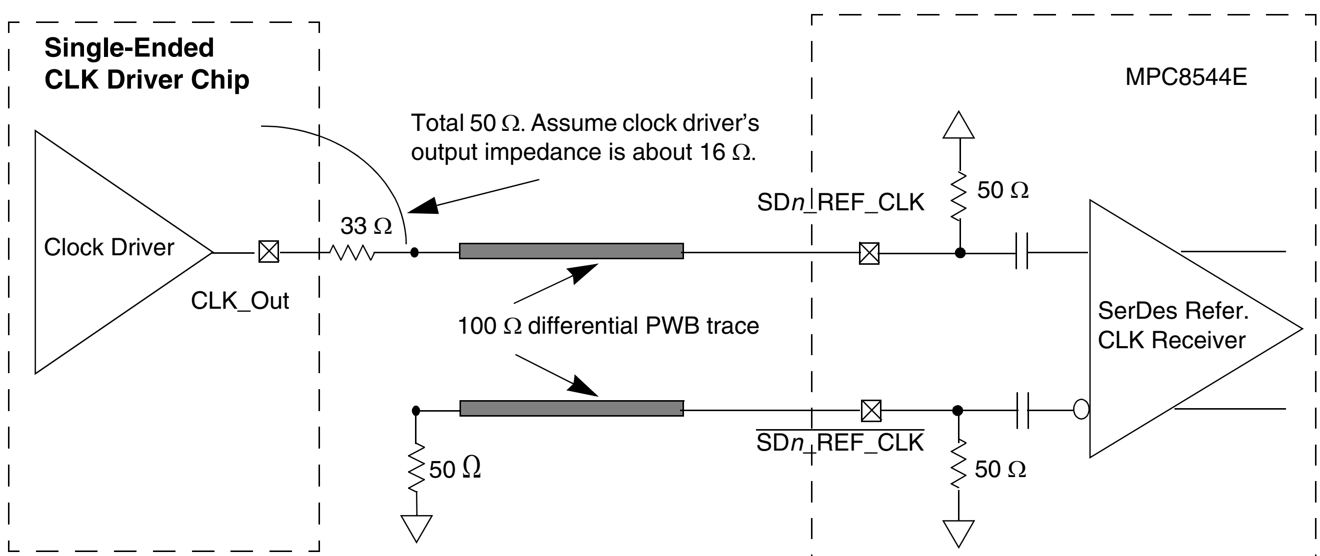


Figure 52. Single-Ended Connection (Reference Only)

16.2.4 AC Requirements for SerDes Reference Clocks

The clock driver selected should provide a high quality reference clock with low phase noise and cycle-to-cycle jitter. Phase noise less than 100 kHz can be tracked by the PLL and data recovery loops and is less of a problem. Phase noise above 15 MHz is filtered by the PLL. The most problematic phase noise occurs in the 1–15 MHz range. The source impedance of the clock driver should be 50 Ω to match the transmission line and reduce reflections which are a source of noise to the system.

Table 57 describes some AC parameters common to SGMII, and PCI Express protocols.

Table 57. SerDes Reference Clock Common AC Parameters

Parameter	Symbol	Min	Max	Unit	Notes
Rising Edge Rate	Rise Edge Rate	1.0	4.0	V/ns	2, 3
Falling Edge Rate	Fall Edge Rate	1.0	4.0	V/ns	2, 3
Differential Input High Voltage	V_{IH}	+200	—	mV	2
Differential Input Low Voltage	V_{IL}	—	–200	mV	2
Rising edge rate (SD*n*_REF_CLK) to falling edge rate ($\overline{\text{SD}n\text{\_REF\_CLK}}$) matching	Rise-Fall Matching	—	20	%	1, 4

Notes:

1. Measurement taken from single ended waveform.
2. Measurement taken from differential waveform.
3. Measured from –200 mV to +200 mV on the differential waveform (derived from SD*n*_REF_CLK minus $\overline{\text{SD}n\text{\_REF\_CLK}}$). The signal must be monotonic through the measurement region for rise and fall time. The 400 mV measurement window is centered on the differential zero crossing. See Figure 53.
4. Matching applies to rising edge rate for SD*n*_REF_CLK and falling edge rate for $\overline{\text{SD}n\text{\_REF\_CLK}}$. It is measured using a 200 mV window centered on the median cross point where SD*n*_REF_CLK rising meets $\overline{\text{SD}n\text{\_REF\_CLK}}$ falling. The median cross point is used to calculate the voltage thresholds the oscilloscope is to use for the edge rate calculations. The rise edge rate of SD*n*_REF_CLK should be compared to the fall edge rate of $\overline{\text{SD}n\text{\_REF\_CLK}}$, the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 54.

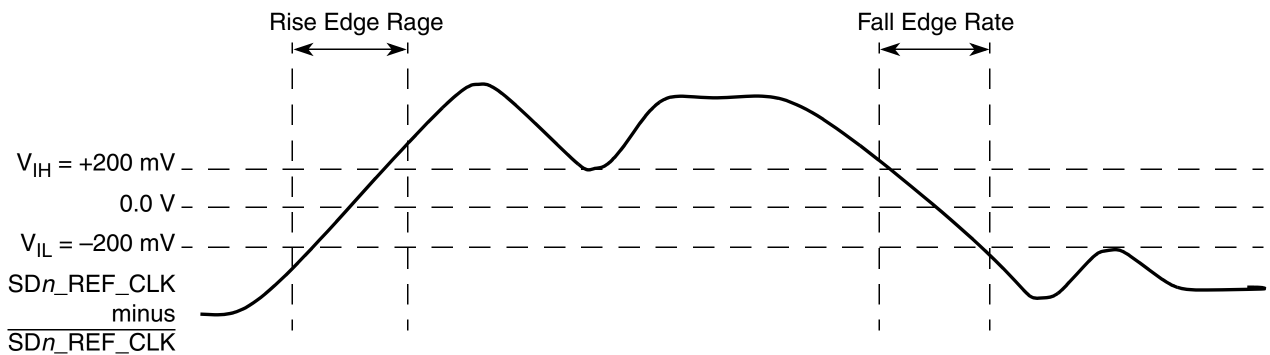


Figure 53. Differential Measurement Points for Rise and Fall Time

17 PCI Express

This section describes the DC and AC electrical specifications for the PCI Express bus of the MPC8544.

17.1 DC Requirements for PCI Express SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$

For more information, see Section 16.2, “SerDes Reference Clocks.”

17.2 AC Requirements for PCI Express SerDes Clocks

Table 58 provides the AC requirements for the PCI Express SerDes clocks.

Table 58. SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ AC Requirements

Symbol[2]	Parameter Description	Min	Typ	Max	Units	Notes
t_{REF}	REFCLK cycle time	—	10	—	ns	1
t_{REFCJ}	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles	—	—	100	ps	—
t_{REFPJ}	Phase jitter. Deviation in edge location with respect to mean edge location	–50	—	50	ps	—

Notes:
1. Typical based on *PCI Express Specification 2.0*.
2. Guaranteed by characterization.

17.3 Clocking Dependencies

The ports on the two ends of a link must transmit data at a rate that is within 600 parts per million (ppm) of each other at all times. This is specified to allow bit rate clock sources with a ±300 ppm tolerance.

17.4 Physical Layer Specifications

The following is a summary of the specifications for the physical layer of PCI Express on this device. For further details as well as the specifications of the transport and data link layer please refer to the *PCI Express Base Specification. Rev. 1.0a*.

Table 59. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$V_{TX\text{-}RCV\text{-}DETECT}$	Amount of voltage change allowed during receiver detection	—	—	600	mV	The total amount of voltage change that a transmitter can apply to sense whether a low impedance receiver is present. See Note 6.
$V_{TX\text{-}DC\text{-}CM}$	TX DC common mode voltage	0	—	3.6	V	The allowed DC common mode voltage under any conditions. See Note 6.
$I_{TX\text{-}SHORT}$	TX short circuit current limit	—	—	90	mA	The total current the transmitter can provide when shorted to its ground.
$T_{TX\text{-}IDLE\text{-}MIN}$	Minimum time spent in electrical idle	50	—	—	UI	Minimum time a transmitter must be in electrical idle utilized by the receiver to start looking for an electrical idle exit after successfully receiving an electrical idle ordered set.
$T_{TX\text{-}IDLE\text{-}SET\text{-}TO\text{-}IDLE}$	Maximum time to transition to a valid electrical idle after sending an electrical Idle ordered set	—	—	20	UI	After sending an electrical idle ordered set, the transmitter must meet all electrical idle specifications within this time. This is considered a debounce time for the transmitter to meet electrical idle after transitioning from LO.
$T_{TX\text{-}IDLE\text{-}TO\text{-}DIFF\text{-}DATA}$	Maximum time to transition to valid TX specifications after leaving an electrical idle condition	—	—	20	UI	Maximum time to meet all TX specifications when transitioning from electrical idle to sending differential data. This is considered a debounce time for the TX to meet all TX specifications after leaving electrical idle.
$RL_{TX\text{-}DIFF}$	Differential return loss	12	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4.
$RL_{TX\text{-}CM}$	Common mode return loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4.
$Z_{TX\text{-}DIFF\text{-}DC}$	DC differential TX impedance	80	100	120	Ω	TX DC differential mode low impedance.
$Z_{TX\text{-}DC}$	Transmitter DC impedance	40	—	—	Ω	Required TX D+ as well as D– DC Impedance during all states.
$L_{TX\text{-}SKEW}$	Lane-to-lane output skew	—	—	500 + 2 UI	ps	Static skew between any two transmitter lanes within a single link.
C_{TX}	AC coupling capacitor	75	—	200	nF	All transmitters shall be AC coupled. The AC coupling is required either within the media or within the transmitting component itself.

Table 59. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$T_{crosslink}$	Crosslink random timeout	0	—	1	ms	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one downstream and one upstream port. See Note 7.

Notes:
1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 58 and measured over any 250 consecutive TX UIs. (Also refer to the transmitter compliance eye diagram shown in Figure 56.)
3. A $T_{TX\text{-}EYE}$ = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX\text{-}JITTER\text{-}MAX}$ = 0.30 UI for the transmitter collected over any 250 consecutive TX UIs. The $T_{TX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$ median is less than half of the total TX jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The transmitter input impedance shall result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50-Ω probes—see Figure 58.) Note that the series capacitors C_{TX} is optional for the return loss measurement.
5. Measured between 20%–80% at transmitter package pins into a test load as shown in Figure 58 for both $V_{TX\text{-}D+}$ and $V_{TX\text{-}D-}$.
6. See Section 4.3.1.8 of the *PCI Express Base Specifications, Rev 1.0a.*
7. See Section 4.2.6.3 of the *PCI Express Base Specifications, Rev 1.0a.*

17.4.2 Transmitter Compliance Eye Diagrams

The TX eye diagram in Figure 56 is specified using the passive compliance/test measurement load (see Figure 58) in place of any real PCI Express interconnect +RX component.

There are two eye diagrams that must be met for the transmitter. Both eye diagrams must be aligned in time using the jitter median to locate the center of the eye diagram. The different eye diagrams will differ in voltage depending whether it is a transition bit or a de-emphasized bit. The exact reduced voltage level of the de-emphasized bit will always be relative to the transition bit.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.

NOTE

It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function (that is, least squares and median deviation fits).

18 Package Description

This section details package parameters, pin assignments, and dimensions.

18.1 Package Parameters for the MPC8544E FC-PBGA

The package parameters for flip chip plastic ball grid array (FC-PBGA) are provided in Table 61.

Table 61. Package Parameters

Parameter	PBGA[1]
Package outline	29 mm × 29 mm
Interconnects	783
Ball pitch	1 mm
Ball diameter (typical)	0.6 mm
Solder ball (Pb-free)	96.5% Sn 3.5% Ag

Note:
1. (FC-PBGA) without a lid.

Table 62. MPC8544E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
DFT				
L1_TSTCLK	AC20	I	OV_{DD}	18
L2_TSTCLK	AE17	I	OV_{DD}	18
$\overline{\text{LSSD\_MODE}}$	AH19	I	OV_{DD}	18
$\overline{\text{TEST\_SEL}}$	AH13	I	OV_{DD}	3
Thermal Management				
TEMP_ANODE	Y3	—	—	13
TEMP_CATHODE	AA3	—	—	13
Power Management				
ASLEEP	AH17	O	OV_{DD}	8, 15, 21
Power and Ground Signals				
GND	D5, M10, F4, D26, D23, C12, C15, E20, D8, B10, E3, J14, K21, F8, A3, F16, E12, E15, D17, L1, F21, H1, G13, G15, G18, C6, A14, A7, G25, H4, C20, J12, J15, J17, F27, M5, J27, K11, L26, K7, K8, L12, L15, M14, M16, M18, N13, N15, N17, N2, P5, P14, P16, P18, R13, R15, R17, T14, T16, T18, U13, U15, U17, AA8, U6, Y10, AC21, AA17, AC16, V4, AD7, AD18, AE23, AF11, AF14, AG23, AH9, A27, B28, C27	—	—	—
OV_{DD}[1:17]	Y16, AB7, AB10, AB13, AC6, AC18, AD9, AD11, AE13, AD15, AD20, AE5, AE22, AF10, AF20, AF24, AF27	Power for PCI and other standards (3.3 V)	OV_{DD}	—
LV_{DD}[1:2]	R4, U3	Power for TSEC1 interfaces (2.5 V, 3.3 V)	LV_{DD}	—
TV_{DD}[1:2]	N8, R10	Power for TSEC3 interfaces (2.5 V, 3.3 V)	TV_{DD}	—
GV_{DD}	B1, B11, C7, C9, C14, C17, D4, D6, R3, D15, E2, E8,C24, E18, F5, E14, C21, G3, G7, G9, G11, H5, H12, E22, F15, J10, K3, K12, K14, H14, D20, E11, M1, N5	Power for DDR1 and DDR2 DRAM I/O voltage (1.8 V, 2.5 V)	GV_{DD}	—
BV_{DD}	L23, J18, J19, F20, F23, H26, J21, J23	Power for local bus (1.8 V, 2.5 V, 3.3 V)	BV_{DD}	—

Table 71 provides the thermal resistance with heat sink in open flow.

Table 71. Thermal Resistance with Heat Sink in Open Flow

Heat Sink with Thermal Grease	Air Flow	Thermal Resistance (°C/W)
Wakefield 53 × 53 × 25 mm pin fin	Natural convection	6.1
Wakefield 53 × 53 × 25 mm pin fin	1 m/s	3.0
Aavid 35 × 31 × 23 mm pin fin	Natural convection	8.1
Aavid 35 × 31 × 23 mm pin fin	1 m/s	4.3
Aavid 30 × 30 × 9.4 mm pin fin	Natural convection	11.6
Aavid 30 × 30 × 9.4 mm pin fin	1 m/s	6.7
Aavid 43 × 41 × 16.5 mm pin fin	Natural convection	8.3
Aavid 43 × 41 × 16.5 mm pin fin	1 m/s	4.3

Simulations with heat sinks were done with the package mounted on the 2s2p thermal test board. The thermal interface material was a typical thermal grease such as Dow Corning 340 or Wakefield 120 grease. For system thermal modeling, the MPC8544E thermal model without a lid is shown in Figure 60. The substrate is modeled as a block 29 × 29 × 1.18 mm with an in-plane conductivity of 18.0 W/m•K and a through-plane conductivity of 1.0 W/m•K. The solder balls and air are modeled as a single block 29 × 29 × 0.58 mm with an in-plane conductivity of 0.034 W/m•K and a through plane conductivity of 12.1 W/m•K. The die is modeled as 7.6 × 8.4 mm with a thickness of 0.75 mm. The bump/underfill layer is modeled as a collapsed thermal resistance between the die and substrate assuming a conductivity of 6.5 W/m•K in the thickness dimension of 0.07 mm. The die is centered on the substrate. The thermal model uses approximate dimensions to reduce grid. Please refer to Figure 59 for actual dimensions.

20.2 Recommended Thermal Model

Table 72 shows the MPC8544E thermal model.

Table 72. MPC8544EThermal Model

Conductivity	Value	Units
Die (7.6 × 8.4 × 0.75mm)		
Silicon	Temperature dependent	—
Bump/Underfill (7.6 × 8.4 × 0.070 mm) Collapsed Thermal Resistance		
Kz	6.5	W/m•K
Substrate (29 × 29 × 1.18 mm)		
Kx	18	W/m•K
Ky	18	
Kz	1.0	

where:

I_{fw} = Forward current

I_s = Saturation current

V_d = Voltage at diode

V_f = Voltage forward biased

V_H = Diode voltage while I_H is flowing

V_L = Diode voltage while I_L is flowing

I_H = Larger diode bias current

I_L = Smaller diode bias current

q = Charge of electron (1.6×10^{-19} C)

n = Ideality factor (normally 1.0)

K = Boltzman's constant (1.38×10^{-23} Joules/K)

T = Temperature (Kelvins)

The ratio of I_H to I_L is usually selected to be 10:1. The above simplifies to the following:

$$\mathbf{V_H - V_L = 1.986 \times 10^{-4} \times nT}$$

Solving for T, the equation becomes:

$$\mathbf{nT} = \frac{V_H - V_L}{1.986 \times 10^{-4}}$$

21 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8544E.

21.1 System Clocking

This device includes six PLLs:

- The platform PLL generates the platform clock from the externally supplied SYSCLK input. The frequency ratio between the platform and SYSCLK is selected using the platform PLL ratio configuration bits as described in Section 19.2, "CCB/SYSCLK PLL Ratio."
- The e500 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e500 core clock and the platform clock is selected using the e500 PLL ratio configuration bits as described in Section 19.3, "e500 Core PLL Ratio."
- The PCI PLL generates the clocking for the PCI bus.
- The local bus PLL generates the clock for the local bus.
- There are two PLLs for the SerDes block.

resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N) \div 2$.

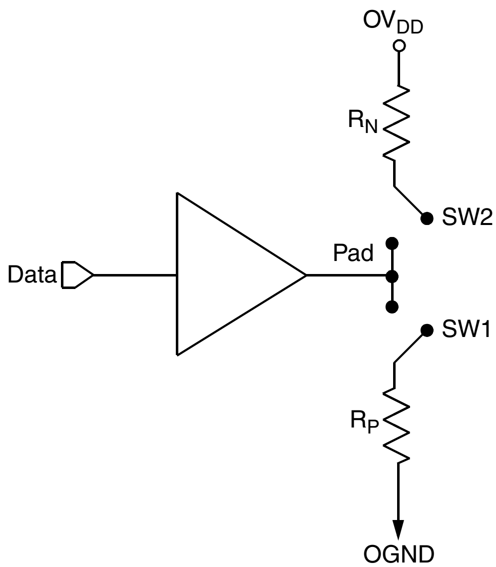


Figure 67. Driver Impedance Measurement

Table 73 summarizes the signal impedance targets. The driver impedances are targeted at minimum V_{DD}, nominal OV_{DD}, 90°C.

Table 73. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R_N	43 Target	25 Target	20 Target	Z_0	W
R_P	43 Target	25 Target	20 Target	Z_0	W

Note: Nominal supply voltages. See Table 1.

21.8 Configuration Pin Muxing

The MPC8544E provides the user with power-on configuration options which can be set through the use of external pull-up or pull-down resistors of 4.7 kΩ on certain output pins (see customer visible configuration pins). These pins are generally used as output only pins in normal operation.

While $\overline{\text{HRESET}}$ is asserted however, these pins are treated as inputs. The value presented on these pins while $\overline{\text{HRESET}}$ is asserted, is latched when $\overline{\text{HRESET}}$ deasserts, at which time the input receiver is disabled and the I/O circuit takes on its normal function. Most of these sampled configuration pins are equipped with an on-chip gated resistor of approximately 20 kΩ. This value should permit the 4.7-kΩ resistor to pull the configuration pin to a valid logic low level. The pull-up resistor is enabled only during $\overline{\text{HRESET}}$ (and for platform /system clocks after $\overline{\text{HRESET}}$ deassertion to ensure capture of the reset value). When the input receiver is disabled the pull-up is also, thus allowing functional operation of the pin as an output with minimal signal quality or delay disruption. The default value for all configuration bits treated this way has

22.2 Nomenclature of Parts Fully Addressed by this Document

Table 75 provides the Freescale part numbering nomenclature for the MPC8544E.

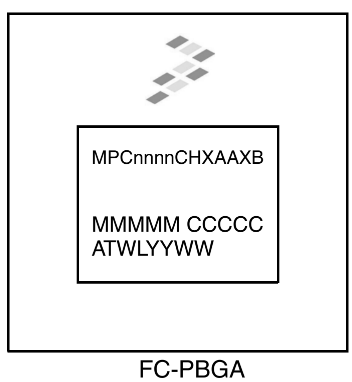
Table 75. Device Nomenclature

MPC	*nnnn*	*E*	*C*	*HX*	*AA*	*X*	*B*
Product Code	**Part Identifier**	**Encryption Acceleration**	**Temperature Range**	**Package[1]**	**Processor Frequency[2]**	**Platform Frequency**	**Revision Level**
MPC	8544	Blank = not included E = included	B or Blank = Industrial Tier standard temp range(0° to 105°C) C = Industrial Tier Extended temp range(–40° to 105°C)	VT = FC-PBGA (lead-free) VJ = lead-free FC-PBGA	AL = 667 MHz AN = 800 MHz AQ = 1000 MHz AR = 1067 MHz	F = 333 MHz G = 400 MHz J = 533 MHz	Blank = Rev. 1.1 1.1.1 A = Rev. 2.1

Notes:
1. See Section 18, "Package Description," for more information on available package types.
2. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.
3. The VT part number is ROHS-compliant, with the permitted exception of the C4 die bumps.
4. The VJ part number is entirely lead-free. This includes the C4 die bumps.

22.3 Part Marking

Parts are marked as in the example shown in Figure 70.



Notes:
MMMMM is the 5-digit mask number.
ATWLYYWW is the traceability code.
CCCCC is the country of assembly. This space is left blank if parts are assembled in the United States.

Figure 70. Part Marking for FC-PBGA Device