



Welcome to [E-XFL.COM](https://www.e-xfl.com)

### Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	6627
Number of Logic Elements/Cells	132540
Total RAM Bits	6747840
Number of I/O	1126
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1508-BBGA, FCBGA
Supplier Device Package	1508-FBGA, FC (40x40)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/ep2s130f1508c4">https://www.e-xfl.com/product-detail/intel/ep2s130f1508c4</a>

Copyright © 2011 Altera Corporation. All rights reserved. Altera, The Programmable Solutions Company, the stylized Altera logo, specific device designations, and all other words and logos that are identified as trademarks and/or service marks are, unless noted otherwise, the trademarks and service marks of Altera Corporation in the U.S. and other countries. All other product or service names are the property of their respective holders. Altera products are protected under numerous U.S. and foreign patents and pending applications, maskwork rights, and copyrights. Altera warrants performance of its semiconductor products to current specifications in accordance with Altera's standard warranty, but reserves the right to make changes to any products and services at any time without notice. Altera assumes no responsibility or liability arising out of the application or use of any information, product, or service described herein except as expressly agreed to in writing by Altera Corporation. Altera customers are advised to obtain the latest version of device specifications before relying on any published information and before placing orders for products or services.



R24 row interconnects span 24 LABs and provide the fastest resource for long row connections between LABs, TriMatrix memory, DSP blocks, and Row IOEs. The R24 row interconnects can cross M-RAM blocks. R24 row interconnects drive to other row or column interconnects at every fourth LAB and do not drive directly to LAB local interconnects. R24 row interconnects drive LAB local interconnects via R4 and C4 interconnects. R24 interconnects can drive R24, R4, C16, and C4 interconnects.

The column interconnect operates similarly to the row interconnect and vertically routes signals to and from LABs, TriMatrix memory, DSP blocks, and IOEs. Each column of LABs is served by a dedicated column interconnect. These column resources include:

- Shared arithmetic chain interconnects in an LAB
- Carry chain interconnects in an LAB and from LAB to LAB
- Register chain interconnects in an LAB
- C4 interconnects traversing a distance of four blocks in up and down direction
- C16 column interconnects for high-speed vertical routing through the device

Stratix II devices include an enhanced interconnect structure in LABs for routing shared arithmetic chains and carry chains for efficient arithmetic functions. The register chain connection allows the register output of one ALM to connect directly to the register input of the next ALM in the LAB for fast shift registers. These ALM to ALM connections bypass the local interconnect. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. [Figure 2–17](#) shows the shared arithmetic chain, carry chain and register chain interconnects.

The LAB row source for control signals, data inputs, and outputs is shown in Table 2–7.

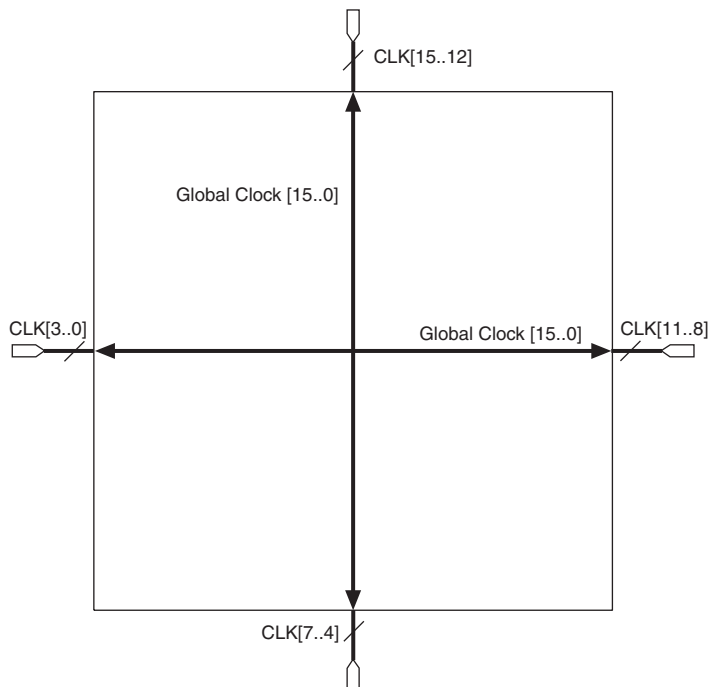
<b>Table 2–7. DSP Block Signal Sources &amp; Destinations</b>			
<b>LAB Row at Interface</b>	<b>Control Signals Generated</b>	<b>Data Inputs</b>	<b>Data Outputs</b>
0	clock0 aclr0 ena0 mult01_saturate addnsub1_round/ accum_round addnsub1 signa sourcea sourceb	A1[17..0] B1[17..0]	OA[17..0] OB[17..0]
1	clock1 aclr1 ena1 accum_saturate mult01_round accum_sload sourcea sourceb mode0	A2[17..0] B2[17..0]	OC[17..0] OD[17..0]
2	clock2 aclr2 ena2 mult23_saturate addnsub3_round/ accum_round addnsub3 sign_b sourcea sourceb	A3[17..0] B3[17..0]	OE[17..0] OF[17..0]
3	clock3 aclr3 ena3 accum_saturate mult23_round accum_sload sourcea sourceb mode1	A4[17..0] B4[17..0]	OG[17..0] OH[17..0]



See the *DSP Blocks in Stratix II & Stratix II GX Devices* chapter in volume 2 of the *Stratix II Device Handbook* or the *Stratix II GX Device Handbook*, for more information on DSP blocks.

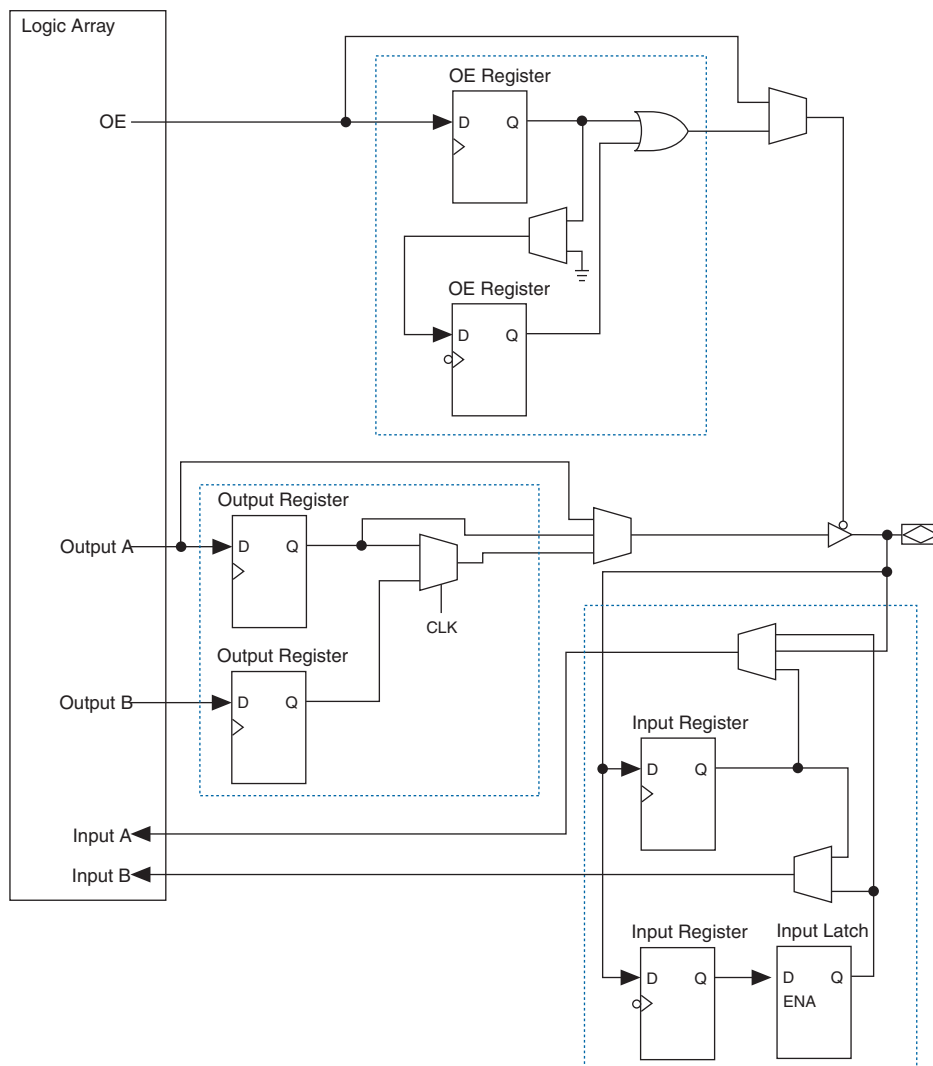
global clock networks can also be driven by internal logic for internally generated global clocks and asynchronous clears, clock enables, or other control signals with large fanout. Figure 2–31 shows the 16 dedicated CLK pins driving global clock networks.

**Figure 2–31. Global Clocking**



### *Regional Clock Network*

There are eight regional clock networks  $RCLK[7..0]$  in each quadrant of the Stratix II device that are driven by the dedicated  $CLK[15..0]$  input pins, by PLL outputs, or by internal logic. The regional clock networks provide the lowest clock delay and skew for logic contained in a single quadrant. The CLK clock pins symmetrically drive the RCLK networks in a particular quadrant, as shown in Figure 2–32.

**Figure 2–46. Stratix II IOE Structure**

The IOEs are located in I/O blocks around the periphery of the Stratix II device. There are up to four IOEs per row I/O block and four IOEs per column I/O block. The row I/O blocks drive row, column, or direct link interconnects. The column I/O blocks drive column interconnects.

Figure 2–47 shows how a row I/O block connects to the logic array.

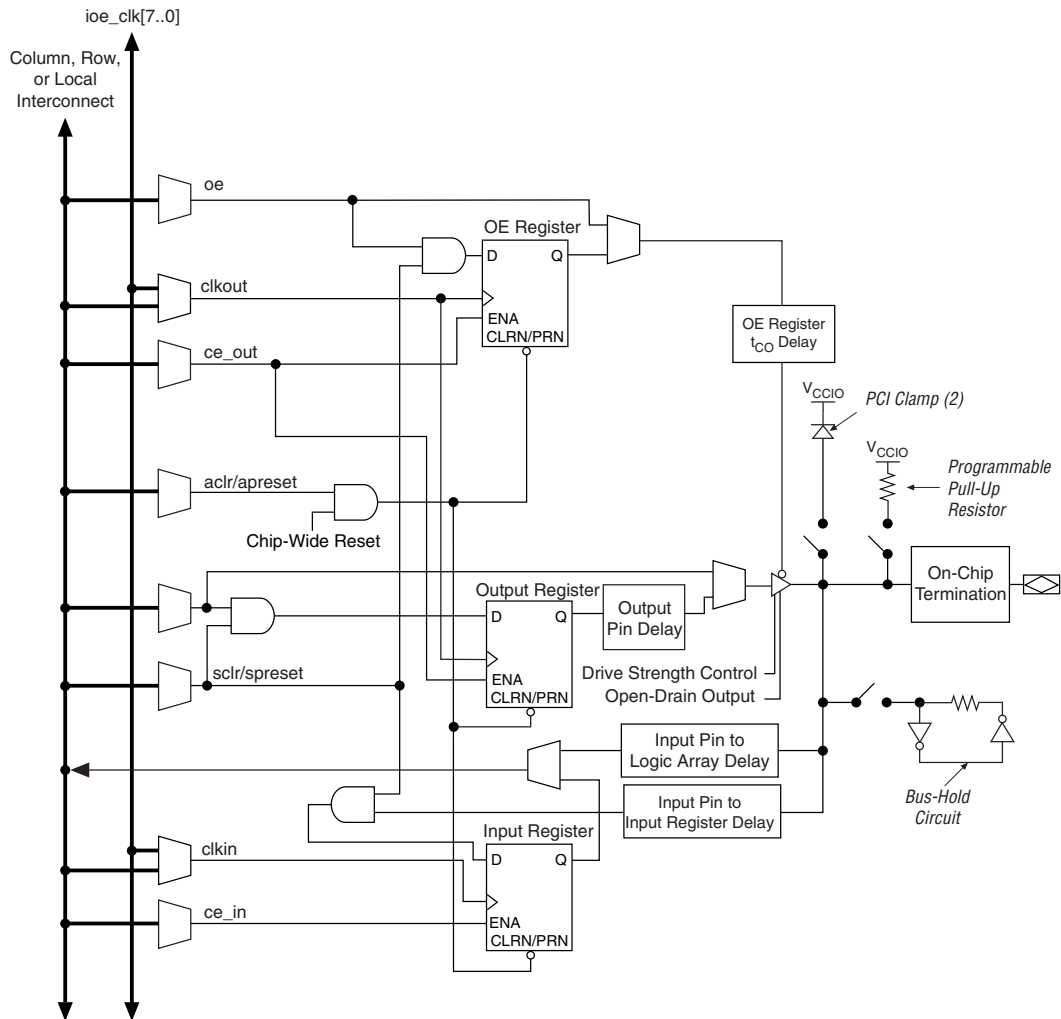
Figure 2–48 shows how a column I/O block connects to the logic array.



- 2-73  
Stratix II Device Handbook, Volume 1

Figure 2–51 shows the IOE in bidirectional configuration.

**Figure 2–51. Stratix II IOE in Bidirectional I/O Configuration** *Note (1)*



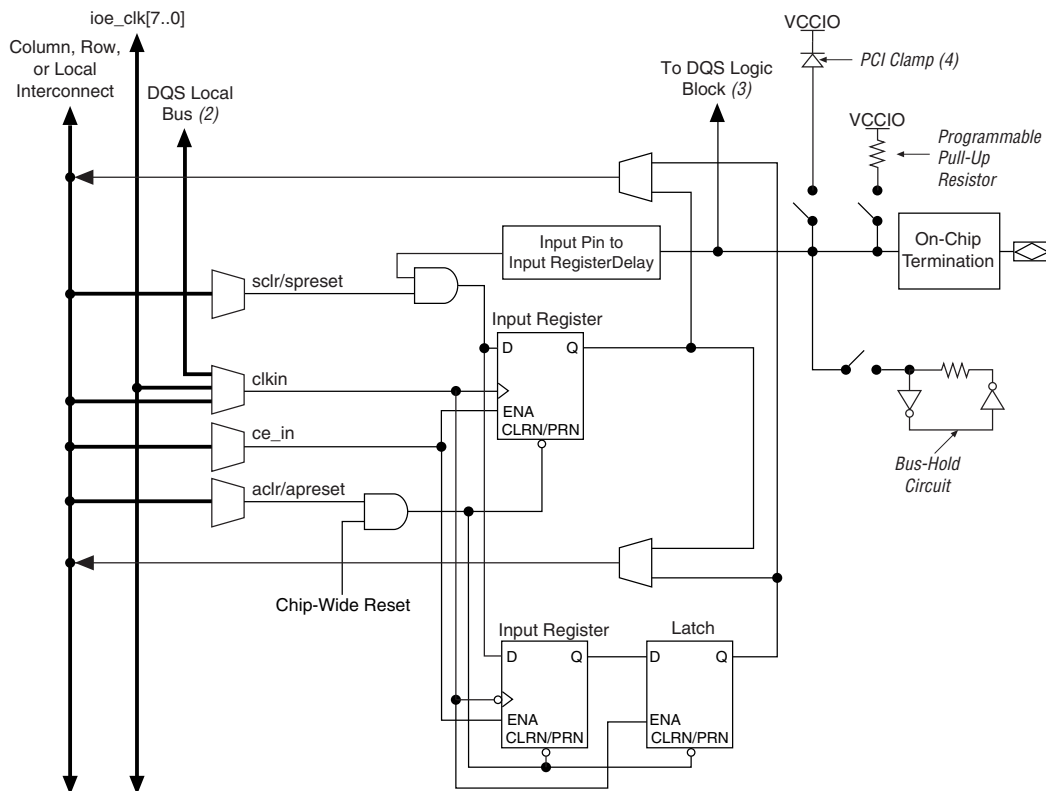
**Notes to Figure 2–51:**

- (1) All input signals to the IOE can be inverted at the IOE.
- (2) The optional PCI clamp is only available on column I/O pins.



When using the IOE for DDR inputs, the two input registers clock double rate input data on alternating edges. An input latch is also used in the IOE for DDR input acquisition. The latch holds the data that is present during the clock high times. This allows both bits of data to be synchronous with the same clock edge (either rising or falling). Figure 2–52 shows an IOE configured for DDR input. Figure 2–53 shows the DDR input timing diagram.

**Figure 2–52. Stratix II IOE in DDR Input I/O Configuration** Notes (1), (2), (3)



Notes to Figure 2–52:

- (1) All input signals to the IOE can be inverted at the IOE.
- (2) This signal connection is only allowed on dedicated DQ function pins.
- (3) This signal is for dedicated DQS function pins only.
- (4) The optional PCI clamp is only available on column I/O pins.

**Table 2–14. DQS & DQ Bus Mode Support (Part 2 of 2)** *Note (1)*

Device	Package	Number of ×4 Groups	Number of ×8/×9 Groups	Number of ×16/×18 Groups	Number of ×32/×36 Groups
EP2S90	484-pin Hybrid FineLine BGA	8	4	0	0
	780-pin FineLine BGA	18	8	4	0
	1,020-pin FineLine BGA	36	18	8	4
	1,508-pin FineLine BGA	36	18	8	4
EP2S130	780-pin FineLine BGA	18	8	4	0
	1,020-pin FineLine BGA	36	18	8	4
	1,508-pin FineLine BGA	36	18	8	4
EP2S180	1,020-pin FineLine BGA	36	18	8	4
	1,508-pin FineLine BGA	36	18	8	4

**Notes to Table 2–14:**

- (1) Check the pin table for each DQS/DQ group in the different modes.

A compensated delay element on each DQS pin automatically aligns input DQS synchronization signals with the data window of their corresponding DQ data signals. The DQS signals drive a local DQS bus in the top and bottom I/O banks. This DQS bus is an additional resource to the I/O clocks and is used to clock DQ input registers with the DQS signal.

The Stratix II device has two phase-shifting reference circuits, one on the top and one on the bottom of the device. The circuit on the top controls the compensated delay elements for all DQS pins on the top. The circuit on the bottom controls the compensated delay elements for all DQS pins on the bottom.

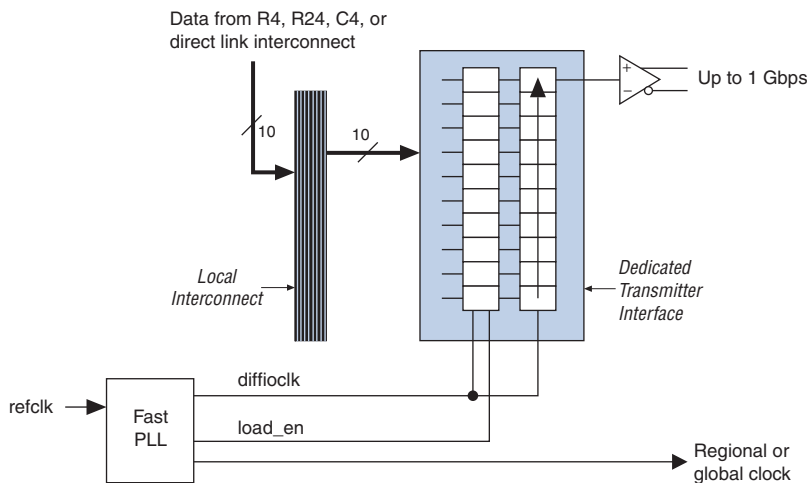
Each phase-shifting reference circuit is driven by a system reference clock, which must have the same frequency as the DQS signal. Clock pins CLK[15 . . 12] p feed the phase circuitry on the top of the device and clock pins CLK[7 . . 4] p feed the phase circuitry on the bottom of the device. In addition, PLL clock outputs can also feed the phase-shifting reference circuits.

Figure 2–56 illustrates the phase-shift reference circuit control of each DQS delay shift on the top of the device. This same circuit is duplicated on the bottom of the device.

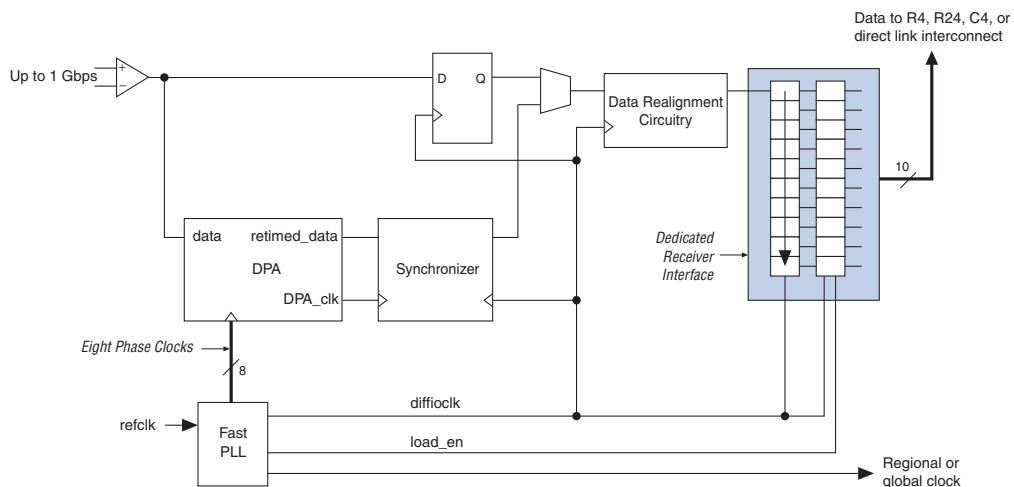
## Dedicated Circuitry with DPA Support

Stratix II devices support source-synchronous interfacing with LVDS or HyperTransport signaling at up to 1 Gbps. Stratix II devices can transmit or receive serial channels along with a low-speed or high-speed clock. The receiving device PLL multiplies the clock by an integer factor  $W = 1$  through 32. For example, a HyperTransport technology application where the data rate is 1,000 Mbps and the clock rate is 500 MHz would require that  $W$  be set to 2. The SERDES factor  $J$  determines the parallel data width to deserialize from receivers or to serialize for transmitters. The SERDES factor  $J$  can be set to 4, 5, 6, 7, 8, 9, or 10 and does not have to equal the PLL clock-multiplication  $W$  value. A design using the dynamic phase aligner also supports all of these  $J$  factor values. For a  $J$  factor of 1, the Stratix II device bypasses the SERDES block. For a  $J$  factor of 2, the Stratix II device bypasses the SERDES block, and the DDR input and output registers are used in the IOE. Figure 2-58 shows the block diagram of the Stratix II transmitter channel.

**Figure 2-58. Stratix II Transmitter Channel**



Each Stratix II receiver channel features a DPA block for phase detection and selection, a SERDES, a synchronizer, and a data realigner circuit. You can bypass the dynamic phase aligner without affecting the basic source-synchronous operation of the channel. In addition, you can dynamically switch between using the DPA block or bypassing the block via a control signal from the logic array. Figure 2-59 shows the block diagram of the Stratix II receiver channel.

**Figure 2–59. Stratix II Receiver Channel**

An external pin or global or regional clock can drive the fast PLLs, which can output up to three clocks: two multiplied high-speed clocks to drive the SERDES block and/or external pin, and a low-speed clock to drive the logic array. In addition, eight phase-shifted clocks from the VCO can feed to the DPA circuitry.



For more information on the fast PLL, see the *PLLs in Stratix II & Stratix II GX Devices* chapter in volume 2 of the *Stratix II Device Handbook* or the *Stratix II GX Device Handbook*.

The eight phase-shifted clocks from the fast PLL feed to the DPA block. The DPA block selects the closest phase to the center of the serial data eye to sample the incoming data. This allows the source-synchronous circuitry to capture incoming data correctly regardless of the channel-to-channel or clock-to-channel skew. The DPA block locks to a phase closest to the serial data phase. The phase-aligned DPA clock is used to write the data into the synchronizer.

The synchronizer sits between the DPA block and the data realignment and SERDES circuitry. Since every channel utilizing the DPA block can have a different phase selected to sample the data, the synchronizer is needed to synchronize the data to the high-speed clock domain of the data realignment and the SERDES circuitry.

## Operating Modes

The Stratix II architecture uses SRAM configuration elements that require configuration data to be loaded each time the circuit powers up. The process of physically loading the SRAM data into the device is called configuration. During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called command mode. Normal device operation is called user mode.

SRAM configuration elements allow Stratix II devices to be reconfigured in-circuit by loading new configuration data into the device. With real-time reconfiguration, the device is forced into command mode with a device pin. The configuration process loads different configuration data, reinitializes the device, and resumes user-mode operation. You can perform in-field upgrades by distributing new configuration files either within the system or remotely.

PORSEL is a dedicated input pin used to select POR delay times of 12 ms or 100 ms during power-up. When the PORSEL pin is connected to ground, the POR time is 100 ms; when the PORSEL pin is connected to  $V_{CC}$ , the POR time is 12 ms.

The nIO PULLUP pin is a dedicated input that chooses whether the internal pull-ups on the user I/O pins and dual-purpose configuration I/O pins (nCS0, ASDO, DATA [7 . . 0], nWS, nRS, RDYnBSY, nCS, CS, RUnLU, PGM [2 . . 0], CLKUSR, INIT\_DONE, DEV\_OE, DEV\_CLR) are on or off before and during configuration. A logic high (1.5, 1.8, 2.5, 3.3 V) turns off the weak internal pull-ups, while a logic low turns them on.

Stratix II devices also offer a new power supply,  $V_{CCPD}$ , which must be connected to 3.3 V in order to power the 3.3-V/2.5-V buffer available on the configuration input pins and JTAG pins.  $V_{CCPD}$  applies to all the JTAG input pins (TCK, TMS, TDI, and TRST) and the configuration input pins when VCCSEL is connected to ground. See [Table 3–4](#) for more information on the pins affected by VCCSEL.

The VCCSEL pin allows the  $V_{CCIO}$  setting (of the banks where the configuration inputs reside) to be independent of the voltage required by the configuration inputs. Therefore, when selecting the  $V_{CCIO}$ , the  $V_{IL}$  and  $V_{IH}$  levels driven to the configuration inputs do not have to be a concern.

the Device & Pin Options dialog box in the Quartus II software uses a 32-bit CRC circuit to ensure data reliability and is one of the best options for mitigating SEU.

You can implement the error detection CRC feature with existing circuitry in Stratix II devices, eliminating the need for external logic. For Stratix II devices, CRC is computed by the device during configuration and checked against an automatically computed CRC during normal operation. The CRC\_ERROR pin reports a soft error when configuration SRAM data is corrupted, triggering device reconfiguration.

Custom-Built Circuitry

Dedicated circuitry is built in the Stratix II devices to perform error detection automatically. This error detection circuitry in Stratix II devices constantly checks for errors in the configuration SRAM cells while the device is in user mode. You can monitor one external pin for the error and use it to trigger a re-configuration cycle. You can select the desired time between checks by adjusting a built-in clock divider.

Software Interface

In the Quartus II software version 4.1 and later, you can turn on the automated error detection CRC feature in the Device & Pin Options dialog box. This dialog box allows you to enable the feature and set the internal frequency of the CRC between 400 kHz to 50 MHz. This controls the rate that the CRC circuitry verifies the internal configuration SRAM bits in the FPGA device.

For more information on CRC, refer to *AN 357: Error Detection Using CRC in Altera FPGA Devices*.

Document  
Revision History

Table 3–7 shows the revision history for this chapter.

Table 3–7. Document Revision History (Part 1 of 2)		
Date and Document Version	Changes Made	Summary of Changes
May 2007, v4.2	Moved Document Revision History section to the end of the chapter.	—
	Updated the “Temperature Sensing Diode (TSD)” section.	—

**Table 3–7. Document Revision History (Part 2 of 2)**

<b>Date and Document Version</b>	<b>Changes Made</b>	<b>Summary of Changes</b>
April 2006, v4.1	Updated “Device Security Using Configuration Bitstream Encryption” section.	—
December 2005, v4.0	Updated “Software Interface” section.	—
May 2005, v3.0	<ul style="list-style-type: none"> <li>• Updated “IEEE Std. 1149.1 JTAG Boundary-Scan Support” section.</li> <li>• Updated “Operating Modes” section.</li> </ul>	—
January 2005, v2.1	Updated JTAG chain device limits.	—
January 2005, v2.0	Updated Table 3–3.	—
July 2004, v1.1	<ul style="list-style-type: none"> <li>• Added “Automated Single Event Upset (SEU) Detection” section.</li> <li>• Updated “Device Security Using Configuration Bitstream Encryption” section.</li> <li>• Updated Figure 3–2.</li> </ul>	—
February 2004, v1.0	Added document to the Stratix II Device Handbook.	—

$I_{IOPIN}$  is the current at any user I/O pin on the device. This specification takes into account the pin capacitance, but not board trace and external loading capacitance. Additional capacitance for trace, connector, and loading needs must be considered separately. For the AC specification, the peak current duration is 10 ns or less because of power-up transients. For more information, refer to the *Hot-Socketing & Power-Sequencing Feature & Testing for Altera Devices* white paper.

A possible concern regarding hot-socketing is the potential for latch-up. Latch-up can occur when electrical subsystems are hot-socketed into an active system. During hot-socketing, the signal pins may be connected and driven by the active system before the power supply can provide current to the device's  $V_{CC}$  and ground planes. This condition can lead to latch-up and cause a low-impedance path from  $V_{CC}$  to ground within the device. As a result, the device extends a large amount of current, possibly causing electrical damage. Nevertheless, Stratix II devices are immune to latch-up when hot-socketing.

## Hot Socketing Feature Implementation in Stratix II Devices

The hot socketing feature turns off the output buffer during the power-up event (either  $V_{CCINT}$ ,  $V_{CCIO}$ , or  $V_{CCPD}$  supplies) or power down. The hot-socket circuit will generate an internal `HOTSCKT` signal when either  $V_{CCINT}$ ,  $V_{CCIO}$ , or  $V_{CCPD}$  is below threshold voltage. The `HOTSCKT` signal will cut off the output buffer to make sure that no DC current (except for weak pull up leaking) leaks through the pin. When  $V_{CC}$  ramps up very slowly,  $V_{CC}$  is still relatively low even after the POR signal is released and the configuration is finished. The `CONF_DONE`, `nCEO`, and `nSTATUS` pins fail to respond, as the output buffer can not flip from the state set by the hot socketing circuit at this low  $V_{CC}$  voltage. Therefore, the hot socketing circuit has been removed on these configuration pins to make sure that they are able to operate during configuration. It is expected behavior for these pins to drive out during power-up and power-down sequences.

Each I/O pin has the following circuitry shown in [Figure 4-1](#).



### Operating Conditions

Stratix® II devices are offered in both commercial and industrial grades. Industrial devices are offered in -4 speed grades and commercial devices are offered in -3 (fastest), -4, -5 speed grades.

Tables 5–1 through 5–32 provide information about absolute maximum ratings, recommended operating conditions, DC electrical characteristics, and other specifications for Stratix II devices.

### Absolute Maximum Ratings

Table 5–1 contains the absolute maximum ratings for the Stratix II device family.

<b>Table 5–1. Stratix II Device Absolute Maximum Ratings</b> <i>Notes (1), (2), (3)</i>					
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>	<b>Minimum</b>	<b>Maximum</b>	<b>Unit</b>
$V_{CCINT}$	Supply voltage	With respect to ground	–0.5	1.8	V
$V_{CCIO}$	Supply voltage	With respect to ground	–0.5	4.6	V
$V_{CCPD}$	Supply voltage	With respect to ground	–0.5	4.6	V
$V_{CCA}$	Analog power supply for PLLs	With respect to ground	–0.5	1.8	V
$V_{CCD}$	Digital power supply for PLLs	With respect to ground	–0.5	1.8	V
$V_I$	DC input voltage (4)		–0.5	4.6	V
$I_{OUT}$	DC output current, per pin		–25	40	mA
$T_{STG}$	Storage temperature	No bias	–65	150	°C
$T_J$	Junction temperature	BGA packages under bias	–55	125	°C

#### Notes to Tables 5–1

- (1) See the *Operating Requirements for Altera Devices Data Sheet*.
- (2) Conditions beyond those listed in Table 5–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse effects on the device.
- (3) Supply voltage specifications apply to voltage readings taken at the device pins, not at the power supply.
- (4) During transitions, the inputs may overshoot to the voltage shown in Table 5–2 based upon the input duty cycle. The DC case is equivalent to 100% duty cycle. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns.

Table 5–75. Stratix II I/O Output Delay for Column Pins (Part 3 of 8)

I/O Standard	Drive Strength	Parameter	Minimum Timing		-3 Speed Grade (3)	-3 Speed Grade (4)	-4 Speed Grade	-5 Speed Grade	Unit
			Industrial	Commercial					
1.8 V	2 mA	t <sub>OP</sub>	1042	1093	2904	3048	3338	3472	ps
		t <sub>DIP</sub>	1062	1115	2970	3118	3414	3562	ps
	4 mA	t <sub>OP</sub>	1047	1098	2248	2359	2584	2698	ps
		t <sub>DIP</sub>	1067	1120	2314	2429	2660	2788	ps
	6 mA	t <sub>OP</sub>	974	1022	2024	2124	2326	2434	ps
		t <sub>DIP</sub>	994	1044	2090	2194	2402	2524	ps
	8 mA	t <sub>OP</sub>	976	1024	1947	2043	2238	2343	ps
		t <sub>DIP</sub>	996	1046	2013	2113	2314	2433	ps
	10 mA	t <sub>OP</sub>	933	978	1882	1975	2163	2266	ps
		t <sub>DIP</sub>	953	1000	1948	2045	2239	2356	ps
	12 mA (1)	t <sub>OP</sub>	934	979	1833	1923	2107	2209	ps
		t <sub>DIP</sub>	954	1001	1899	1993	2183	2299	ps
1.5 V	2 mA	t <sub>OP</sub>	1023	1073	2505	2629	2879	3002	ps
		t <sub>DIP</sub>	1043	1095	2571	2699	2955	3092	ps
	4 mA	t <sub>OP</sub>	963	1009	2023	2123	2325	2433	ps
		t <sub>DIP</sub>	983	1031	2089	2193	2401	2523	ps
	6 mA	t <sub>OP</sub>	966	1012	1923	2018	2210	2315	ps
		t <sub>DIP</sub>	986	1034	1989	2088	2286	2405	ps
	8 mA (1)	t <sub>OP</sub>	926	971	1878	1970	2158	2262	ps
		t <sub>DIP</sub>	946	993	1944	2040	2234	2352	ps
SSTL-2 Class I	8 mA	t <sub>OP</sub>	913	957	1715	1799	1971	2041	ps
		t <sub>DIP</sub>	933	979	1781	1869	2047	2131	ps
	12 mA (1)	t <sub>OP</sub>	896	940	1672	1754	1921	1991	ps
		t <sub>DIP</sub>	916	962	1738	1824	1997	2081	ps
SSTL-2 Class II	16 mA	t <sub>OP</sub>	876	918	1609	1688	1849	1918	ps
		t <sub>DIP</sub>	896	940	1675	1758	1925	2008	ps
	20 mA	t <sub>OP</sub>	877	919	1598	1676	1836	1905	ps
		t <sub>DIP</sub>	897	941	1664	1746	1912	1995	ps
	24 mA (1)	t <sub>OP</sub>	872	915	1596	1674	1834	1903	ps
		t <sub>DIP</sub>	892	937	1662	1744	1910	1993	ps

**Table 5–78. Maximum Output Toggle Rate on Stratix II Devices (Part 1 of 5)** *Note (1)*

I/O Standard	Drive Strength	Column I/O Pins (MHz)			Row I/O Pins (MHz)			Clock Outputs (MHz)		
		-3	-4	-5	-3	-4	-5	-3	-4	-5
3.3-V LVTTTL	4 mA	270	225	210	270	225	210	270	225	210
	8 mA	435	355	325	435	355	325	435	355	325
	12 mA	580	475	420	580	475	420	580	475	420
	16 mA	720	594	520	-	-	-	720	594	520
	20 mA	875	700	610	-	-	-	875	700	610
	24 mA	1,030	794	670	-	-	-	1,030	794	670
3.3-V LVCMOS	4 mA	290	250	230	290	250	230	290	250	230
	8 mA	565	480	440	565	480	440	565	480	440
	12 mA	790	710	670	-	-	-	790	710	670
	16 mA	1,020	925	875	-	-	-	1,020	925	875
	20 mA	1,066	985	935	-	-	-	1,066	985	935
	24 mA	1,100	1,040	1,000	-	-	-	1,100	1,040	1,000
2.5-V LVTTTL/LVCMOS	4 mA	230	194	180	230	194	180	230	194	180
	8 mA	430	380	380	430	380	380	430	380	380
	12 mA	630	575	550	630	575	550	630	575	550
	16 mA	930	845	820	-	-	-	930	845	820
1.8-V LVTTTL/LVCMOS	2 mA	120	109	104	120	109	104	120	109	104
	4 mA	285	250	230	285	250	230	285	250	230
	6 mA	450	390	360	450	390	360	450	390	360
	8 mA	660	570	520	660	570	520	660	570	520
	10 mA	905	805	755	-	-	-	905	805	755
	12 mA	1,131	1,040	990	-	-	-	1,131	1,040	990
1.5-V LVTTTL/LVCMOS	2 mA	244	200	180	244	200	180	244	200	180
	4 mA	470	370	325	470	370	325	470	370	325
	6 mA	550	430	375	-	-	-	550	430	375
	8 mA	625	495	420	-	-	-	625	495	420
SSTL-2 Class I	8 mA	400	300	300	-	-	-	400	300	300
	12 mA	400	400	350	400	350	350	400	400	350
SSTL-2 Class II	16 mA	350	350	300	350	350	300	350	350	300
	20 mA	400	350	350	-	-	-	400	350	350
	24 mA	400	400	350	-	-	-	400	400	350

**Table 5–78. Maximum Output Toggle Rate on Stratix II Devices (Part 3 of 5)** *Note (1)*

I/O Standard	Drive Strength	Column I/O Pins (MHz)			Row I/O Pins (MHz)			Clock Outputs (MHz)		
		-3	-4	-5	-3	-4	-5	-3	-4	-5
Differential SSTL-18 Class I (3)	4 mA	200	150	150	200	150	150	200	150	150
	6 mA	350	250	200	350	250	200	350	250	200
	8 mA	450	300	300	450	300	300	450	300	300
	10 mA	500	400	400	500	400	400	500	400	400
	12 mA	700	550	400	350	350	297	650	550	400
Differential SSTL-18 Class II (3)	8 mA	200	200	150	-	-	-	200	200	150
	16 mA	400	350	350	-	-	-	400	350	350
	18 mA	450	400	400	-	-	-	450	400	400
	20 mA	550	500	450	-	-	-	550	500	450
1.8-V Differential HSTL Class I (3)	4 mA	300	300	300	-	-	-	300	300	300
	6 mA	500	450	450	-	-	-	500	450	450
	8 mA	650	600	600	-	-	-	650	600	600
	10 mA	700	650	600	-	-	-	700	650	600
	12 mA	700	700	650	-	-	-	700	700	650
1.8-V Differential HSTL Class II (3)	16 mA	500	500	450	-	-	-	500	500	450
	18 mA	550	500	500	-	-	-	550	500	500
	20 mA	650	550	550	-	-	-	550	550	550
1.5-V Differential HSTL Class I (3)	4 mA	350	300	300	-	-	-	350	300	300
	6 mA	500	500	450	-	-	-	500	500	450
	8 mA	700	650	600	-	-	-	700	650	600
	10 mA	700	700	650	-	-	-	700	700	650
	12 mA	700	700	700	-	-	-	700	700	700
1.5-V Differential HSTL Class II (3)	16 mA	600	600	550	-	-	-	600	600	550
	18 mA	650	600	600	-	-	-	650	600	600
	20 mA	700	650	600	-	-	-	700	650	600
3.3-V PCI		1,000	790	670	-	-	-	1,000	790	670
3.3-V PCI-X		1,000	790	670	-	-	-	1,000	790	670
LVDS (6)		-	-	-	500	500	500	450	400	300
HyperTransport technology (4), (6)					500	500	500	-	-	-
LVPECL (5)		-	-	-	-	-	-	450	400	300
3.3-V LVTTTL	OCT 50 $\Omega$	400	400	350	400	400	350	400	400	350
2.5-V LVTTTL	OCT 50 $\Omega$	350	350	300	350	350	300	350	350	300

## High-Speed I/O Specifications

Table 5–88 provides high-speed timing specifications definitions.

<b>Table 5–88. High-Speed Timing Specifications &amp; Definitions</b>	
<b>High-Speed Timing Specifications</b>	<b>Definitions</b>
$t_C$	High-speed receiver/transmitter input and output clock period.
$f_{HSCLK}$	High-speed receiver/transmitter input and output clock frequency.
J	Deserialization factor (width of parallel data bus).
W	PLL multiplication factor.
$t_{RISE}$	Low-to-high transmission time.
$t_{FALL}$	High-to-low transmission time.
Timing unit interval (TUI)	The timing budget allowed for skew, propagation delays, and data sampling window. (TUI = $1/(\text{Receiver Input Clock Frequency} \times \text{Multiplication Factor}) = t_C/w$ ).
$f_{HSDR}$	Maximum/minimum LVDS data transfer rate ( $f_{HSDR} = 1/\text{TUI}$ ), non-DPA.
$f_{HSDRDPA}$	Maximum/minimum LVDS data transfer rate ( $f_{HSDRDPA} = 1/\text{TUI}$ ), DPA.
Channel-to-channel skew (TCCS)	The timing difference between the fastest and slowest output edges, including $t_{CO}$ variation and clock skew. The clock is included in the TCCS measurement.
Sampling window (SW)	The period of time during which the data must be valid in order to capture it correctly. The setup and hold times determine the ideal strobe position within the sampling window.
Input jitter	Peak-to-peak input jitter on high-speed PLLs.
Output jitter	Peak-to-peak output jitter on high-speed PLLs.
$t_{DUTY}$	Duty cycle on high-speed transmitter output clock.
$t_{LOCK}$	Lock time for high-speed transmitter and receiver PLLs.

Table 5–89 shows the high-speed I/O timing specifications for -3 speed grade Stratix II devices.

<b>Table 5–89. High-Speed I/O Specifications for -3 Speed Grade (Part 1 of 2)</b> <i>Notes (1), (2)</i>					
Symbol	Conditions	-3 Speed Grade			Unit
		Min	Typ	Max	
$f_{HSCLK}$ (clock frequency) $f_{HSCLK} = f_{HSDR} / W$	W = 2 to 32 (LVDS, HyperTransport technology) (3)	16		520	MHz
	W = 1 (SERDES bypass, LVDS only)	16		500	MHz
	W = 1 (SERDES used, LVDS only)	150		717	MHz