



Welcome to [E-XFL.COM](https://www.e-xfl.com)

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	780
Number of Logic Elements/Cells	15600
Total RAM Bits	419328
Number of I/O	366
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2s15f672c5

One ALM contains two programmable registers. Each register has data, clock, clock enable, synchronous and asynchronous clear, asynchronous load data, and synchronous and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any internal logic can drive the register's clock and clear control signals. Either general-purpose I/O pins or internal logic can drive the clock enable, preset, asynchronous load, and asynchronous load data. The asynchronous load data input comes from the `dataae` or `dataaf` input of the ALM, which are the same inputs that can be used for register packing. For combinational functions, the register is bypassed and the output of the LUT drives directly to the outputs of the ALM.

Each ALM has two sets of outputs that drive the local, row, and column routing resources. The LUT, adder, or register output can drive these output drivers independently (see [Figure 2-6](#)). For each set of output drivers, two ALM outputs can drive column, row, or direct link routing connections, and one of these ALM outputs can also drive local interconnect resources. This allows the LUT or adder to drive one output while the register drives another output. This feature, called register packing, improves device utilization because the device can use the register and the combinational logic for unrelated functions. Another special packing mode allows the register output to feed back into the LUT of the same ALM so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The ALM can also drive out registered and unregistered versions of the LUT or adder output.



See the *Performance & Logic Efficiency Analysis of Stratix II Devices White Paper* for more information on the efficiencies of the Stratix II ALM and comparisons with previous architectures.

ALM Operating Modes

The Stratix II ALM can operate in one of the following modes:

- Normal mode
- Extended LUT mode
- Arithmetic mode
- Shared arithmetic mode

Each mode uses ALM resources differently. In each mode, eleven available inputs to the ALM—the eight data inputs from the LAB local interconnect; `carry-in` from the previous ALM or LAB; the shared arithmetic chain connection from the previous ALM or LAB; and the register chain connection—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous preset/load, synchronous clear,

Clear & Preset Logic Control

LAB-wide signals control the logic for the register's clear and load/preset signals. The ALM directly supports an asynchronous clear and preset function. The register preset is achieved through the asynchronous load of a logic high. The direct asynchronous preset does not require a NOT-gate push-back technique. Stratix II devices support simultaneous asynchronous load/preset, and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one load/preset signal.

In addition to the clear and load/preset ports, Stratix II devices provide a device-wide reset pin (`DEV_CLRn`) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This device-wide reset overrides all other control signals.

MultiTrack Interconnect

In the Stratix II architecture, connections between ALMs, TriMatrix memory, DSP blocks, and device I/O pins are provided by the MultiTrack interconnect structure with DirectDrive™ technology. The MultiTrack interconnect consists of continuous, performance-optimized routing lines of different lengths and speeds used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

DirectDrive technology is a deterministic routing technology that ensures identical routing resource usage for any function regardless of placement in the device. The MultiTrack interconnect and DirectDrive technology simplify the integration stage of block-based designing by eliminating the re-optimization cycles that typically follow design changes and additions.

The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and repeatable performance when migrating through different device densities. Dedicated row interconnects route signals to and from LABs, DSP blocks, and TriMatrix memory in the same row. These row resources include:

- Direct link interconnects between LABs and adjacent blocks
- R4 interconnects traversing four blocks to the right or left
- R24 row interconnects for high-speed access across the length of the device

M512 RAM Block

The M512 RAM block is a simple dual-port memory block and is useful for implementing small FIFO buffers, DSP, and clock domain transfer applications. Each block contains 576 RAM bits (including parity bits). M512 RAM blocks can be configured in the following modes:

- Simple dual-port RAM
- Single-port RAM
- FIFO
- ROM
- Shift register



Violating the setup or hold time on the memory block address registers could corrupt memory contents. This applies to both read and write operations.

When configured as RAM or ROM, you can use an initialization file to pre-load the memory contents.

M512 RAM blocks can have different clocks on its inputs and outputs. The `wren`, `datain`, and write address registers are all clocked together from one of the two clocks feeding the block. The read address, `rden`, and output registers can be clocked by either of the two clocks driving the block. This allows the RAM block to operate in read/write or input/output clock modes. Only the output register can be bypassed. The six `labclk` signals or local interconnect can drive the `inclock`, `outclock`, `wren`, `rden`, and `outclr` signals. Because of the advanced interconnect between the LAB and M512 RAM blocks, ALMs can also control the `wren` and `rden` signals and the RAM clock, clock enable, and asynchronous clear signals. [Figure 2-19](#) shows the M512 RAM block control signal generation logic.

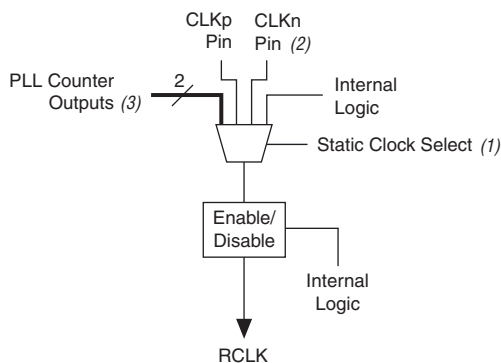
The RAM blocks in Stratix II devices have local interconnects to allow ALMs and interconnects to drive into RAM blocks. The M512 RAM block local interconnect is driven by the R4, C4, and direct link interconnects from adjacent LABs. The M512 RAM blocks can communicate with LABs on either the left or right side through these row interconnects or with LAB columns on the left or right side with the column interconnects. The M512 RAM block has up to 16 direct link input connections from the left adjacent LABs and another 16 from the right adjacent LAB. M512 RAM outputs can also connect to left and right LABs through direct link interconnect. The M512 RAM block has equal opportunity for access and performance to and from LABs on either its left or right side. [Figure 2-20](#) shows the M512 RAM block to logic array interface.

Table 2–4. M-RAM Row Interface Unit Signals

Unit Interface Block	Input Signals	Output Signals
L0	datain_a[14..0] byteena_a[1..0]	dataout_a[11..0]
L1	datain_a[29..15] byteena_a[3..2]	dataout_a[23..12]
L2	datain_a[35..30] addressa[4..0] addr_ena_a clock_a clocken_a renwe_a aclr_a	dataout_a[35..24]
L3	addressa[15..5] datain_a[41..36]	dataout_a[47..36]
L4	datain_a[56..42] byteena_a[5..4]	dataout_a[59..48]
L5	datain_a[71..57] byteena_a[7..6]	dataout_a[71..60]
R0	datain_b[14..0] byteena_b[1..0]	dataout_b[11..0]
R1	datain_b[29..15] byteena_b[3..2]	dataout_b[23..12]
R2	datain_b[35..30] addressb[4..0] addr_ena_b clock_b clocken_b renwe_b aclr_b	dataout_b[35..24]
R3	addressb[15..5] datain_b[41..36]	dataout_b[47..36]
R4	datain_b[56..42] byteena_b[5..4]	dataout_b[59..48]
R5	datain_b[71..57] byteena_b[7..6]	dataout_b[71..60]



See the *TriMatrix Embedded Memory Blocks in Stratix II & Stratix II GX Devices* chapter in volume 2 of the *Stratix II Device Handbook* or the *Stratix II GX Device Handbook* for more information on TriMatrix memory.

Figure 2–38. Regional Clock Control Blocks**Notes to Figure 2–38:**

- (1) These clock select signals can only be set through a configuration file (.sof or .pof) and cannot be dynamically controlled during user mode operation.
- (2) Only the CLKn pins on the top and bottom of the device feed to regional clock select blocks. The clock outputs from corner PLLs cannot be dynamically selected through the global clock control block.
- (3) The clock outputs from corner PLLs cannot be dynamically selected through the global clock control block.

The Quartus II software enables the PLLs and their features without requiring any external devices. Table 2–9 shows the PLLs available for each Stratix II device and their type.

Table 2–9. Stratix II Device PLL Availability												
Device	Fast PLLs								Enhanced PLLs			
	1	2	3	4	7	8	9	10	5	6	11	12
EP2S15	✓	✓	✓	✓					✓	✓		
EP2S30	✓	✓	✓	✓					✓	✓		
EP2S60 (1)	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓
EP2S90 (2)	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓
EP2S130 (3)	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓
EP2S180	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓	✓

Notes to Table 2–9:

- (1) EP2S60 devices in the 1020-pin package contain 12 PLLs. EP2S60 devices in the 484-pin and 672-pin packages contain fast PLLs 1–4 and enhanced PLLs 5 and 6.
- (2) EP2S90 devices in the 1020-pin and 1508-pin packages contain 12 PLLs. EP2S90 devices in the 484-pin and 780-pin packages contain fast PLLs 1–4 and enhanced PLLs 5 and 6.
- (3) EP2S130 devices in the 1020-pin and 1508-pin packages contain 12PLLs. The EP2S130 device in the 780-pin package contains fast PLLs 1–4 and enhanced PLLs 5 and 6.

Table 2–17 shows the Stratix II on-chip termination support per I/O bank.

Table 2–17. On-Chip Termination Support by I/O Banks (Part 1 of 2)			
On-Chip Termination Support	I/O Standard Support	Top & Bottom Banks	Left & Right Banks
Series termination without calibration	3.3-V LVTTTL	✓	✓
	3.3-V LVCMOS	✓	✓
	2.5-V LVTTTL	✓	✓
	2.5-V LVCMOS	✓	✓
	1.8-V LVTTTL	✓	✓
	1.8-V LVCMOS	✓	✓
	1.5-V LVTTTL	✓	✓
	1.5-V LVCMOS	✓	✓
	SSTL-2 Class I and II	✓	✓
	SSTL-18 Class I	✓	✓
	SSTL-18 Class II	✓	
	1.8-V HSTL Class I	✓	✓
	1.8-V HSTL Class II	✓	
	1.5-V HSTL Class I	✓	✓
	1.2-V HSTL	✓	

The PLL_ENA pin and the configuration input pins (Table 3–4) have a dual buffer design: a 3.3-V/2.5-V input buffer and a 1.8-V/1.5-V input buffer. The VCCSEL input pin selects which input buffer is used. The 3.3-V/2.5-V input buffer is powered by V_{CCPD}, while the 1.8-V/1.5-V input buffer is powered by V_{CCIO}. Table 3–4 shows the pins affected by VCCSEL.

Table 3–4. Pins Affected by the Voltage Level at VCCSEL

Pin	VCCSEL = LOW (connected to GND)	VCCSEL = HIGH (connected to V _{CCPD})
nSTATUS (when used as an input)	3.3/2.5-V input buffer is selected. Input buffer is powered by V _{CCPD} .	1.8/1.5-V input buffer is selected. Input buffer is powered by V _{CCIO} of the I/O bank.
nCONFIG		
CONF_DONE (when used as an input)		
DATA[7..0]		
nCE		
DCLK (when used as an input)		
CS		
nWS		
nRS		
nCS		
CLKUSR		
DEV_OE		
DEV_CLRn		
RUnLU		
PLL_ENA		

VCCSEL is sampled during power-up. Therefore, the VCCSEL setting cannot change on the fly or during a reconfiguration. The VCCSEL input buffer is powered by V_{CCINT} and must be hardwired to V_{CCPD} or ground. A logic high VCCSEL connection selects the 1.8-V/1.5-V input buffer, and a logic low selects the 3.3-V/2.5-V input buffer. VCCSEL should be set to comply with the logic levels driven out of the configuration device or MAX[®] II/microprocessor.

If you need to support configuration input voltages of 3.3 V/2.5 V, you should set the VCCSEL to a logic low; you can set the V_{CCIO} of the I/O bank that contains the configuration inputs to any supported voltage. If

you need to support configuration input voltages of 1.8 V/1.5 V, you should set the VCCSEL to a logic high and the V_{CCIO} of the bank that contains the configuration inputs to 1.8 V/1.5 V.



For more information on multi-volt support, including information on using TDO and nCEO in multi-volt systems, refer to the *Stratix II Architecture* chapter in volume 1 of the *Stratix II Device Handbook*.

Configuration Schemes

You can load the configuration data for a Stratix II device with one of five configuration schemes (see Table 3–5), chosen on the basis of the target application. You can use a configuration device, intelligent controller, or the JTAG port to configure a Stratix II device. A configuration device can automatically configure a Stratix II device at system power-up.

You can configure multiple Stratix II devices in any of the five configuration schemes by connecting the configuration enable (nCE) and configuration enable output (nCEO) pins on each device.

Stratix II FPGAs offer the following:

- Configuration data decompression to reduce configuration file storage
- Design security using configuration data encryption to protect your designs
- Remote system upgrades for remotely updating your Stratix II designs

Table 3–5 summarizes which configuration features can be used in each configuration scheme.

Table 3–5. Stratix II Configuration Features (Part 1 of 2)

Configuration Scheme	Configuration Method	Design Security	Decompression	Remote System Upgrade
FPP	MAX II device or microprocessor and flash device	✓ (1)	✓ (1)	✓
	Enhanced configuration device		✓ (2)	✓
AS	Serial configuration device	✓	✓	✓ (3)
PS	MAX II device or microprocessor and flash device	✓	✓	✓
	Enhanced configuration device	✓	✓	✓
	Download cable (4)	✓	✓	

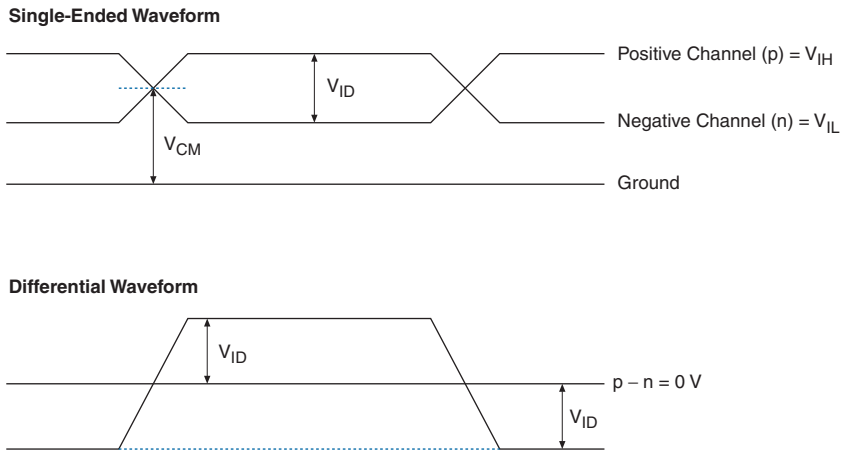
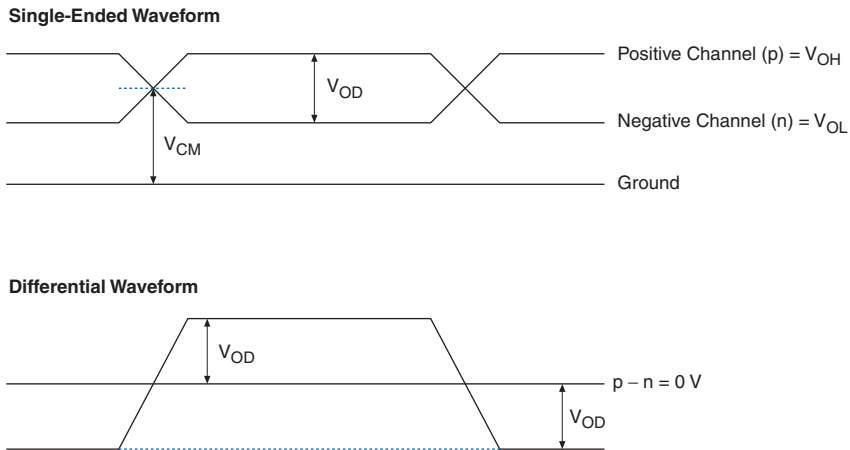
Figure 5–1. Receiver Input Waveforms for Differential I/O Standards**Figure 5–2. Transmitter Output Waveforms for Differential I/O Standards**

Table 5–25. 1.5-V HSTL Class I & II Differential Specifications

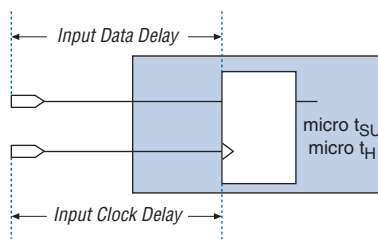
Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	I/O supply voltage		1.425	1.500	1.575	V
V_{DIF} (DC)	DC input differential voltage		0.2			V
V_{CM} (DC)	DC common mode input voltage		0.68		0.90	V
V_{DIF} (AC)	AC differential input voltage		0.4			V
V_{OX} (AC)	AC differential cross point voltage		0.68		0.90	V

Table 5–26. 1.8-V HSTL Class I Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.71	1.80	1.89	V
V_{REF}	Input reference voltage		0.85	0.90	0.95	V
V_{TT}	Termination voltage		0.85	0.90	0.95	V
V_{IH} (DC)	DC high-level input voltage		$V_{REF} + 0.1$			V
V_{IL} (DC)	DC low-level input voltage		–0.3		$V_{REF} - 0.1$	V
V_{IH} (AC)	AC high-level input voltage		$V_{REF} + 0.2$			V
V_{IL} (AC)	AC low-level input voltage				$V_{REF} - 0.2$	V
V_{OH}	High-level output voltage	$I_{OH} = 8 \text{ mA}$ (1)	$V_{CCIO} - 0.4$			V
V_{OL}	Low-level output voltage	$I_{OH} = -8 \text{ mA}$ (1)			0.4	V

Note to Table 5–26:

- (1) This specification is supported across all the programmable drive settings available for this I/O standard as shown in the *Stratix II Architecture* chapter in volume 1 of the *Stratix II Device Handbook*.

Figure 5–3. Input Register Setup & Hold Timing Diagram

For output timing, different I/O standards require different baseline loading techniques for reporting timing delays. Altera characterizes timing delays with the required termination for each I/O standard and with 0 pF (except for PCI and PCI-X which use 10 pF) loading and the timing is specified up to the output pin of the FPGA device. The Quartus II software calculates the I/O timing for each I/O standard with a default baseline loading as specified by the I/O standards.

The following measurements are made during device characterization. Altera measures clock-to-output delays (t_{CO}) at worst-case process, minimum voltage, and maximum temperature (PVT) for default loading conditions shown in Table 5–34. Use the following equations to calculate clock pin to output pin timing for Stratix II devices.

t_{CO} from clock pin to I/O pin = delay from clock pad to I/O output register + IOE output register clock-to-output delay + delay from output register to output pin + I/O output delay

t_{xz}/t_{zx} from clock pin to I/O pin = delay from clock pad to I/O output register + IOE output register clock-to-output delay + delay from output register to output pin + I/O output delay + output enable pin delay

Simulation using IBIS models is required to determine the delays on the PCB traces in addition to the output pin delay timing reported by the Quartus II software and the timing model in the device handbook.

1. Simulate the output driver of choice into the generalized test setup, using values from Table 5–34.
2. Record the time to V_{MEAS} .
3. Simulate the output driver of choice into the actual PCB trace and load, using the appropriate IBIS model or capacitance value to represent the load.

Table 5–39. DSP Block Internal Timing Microparameters (Part 1 of 2)

Symbol	Parameter	-3 Speed Grade (1)		-3 Speed Grade (2)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (3)	Max	Min (3)	Max	Min (4)	Max	Min (3)	Max	
t_{SU}	Input, pipeline, and output register setup time before clock	50		52		57 57		67		ps
t_H	Input, pipeline, and output register hold time after clock	180		189		206 206		241		ps
t_{CO}	Input, pipeline, and output register clock-to-output delay	0	0	0	0	0 0	0	0	0	ps
$t_{INREG2PIPE9}$	Input register to DSP block pipeline register in 9×9 -bit mode	1,312	2,030	1,312	2,030	1,250 1,312	2,334	1,312	2,720	ps
$t_{INREG2PIPE18}$	Input register to DSP block pipeline register in 18×18 -bit mode	1,302	2,010	1,302	2,110	1,240 1,302	2,311	1,302	2,693	ps
$t_{INREG2PIPE36}$	Input register to DSP block pipeline register in 36×36 -bit mode	1,302	2,010	1,302	2,110	1,240 1,302	2,311	1,302	2,693	ps
$t_{PIPE2OUTREG2ADD}$	DSP block pipeline register to output register delay in two-multipliers adder mode	924	1,450	924	1,522	880 924	1,667	924	1,943	ps
$t_{PIPE2OUTREG4ADD}$	DSP block pipeline register to output register delay in four-multipliers adder mode	1,134	1,850	1,134	1,942	1,080 1,134	2,127	1,134	2,479	ps
t_{PD9}	Combinational input to output delay for 9×9	2,100	2,880	2,100	3,024	2,000 2,100	3,312	2,100	3,859	ps
t_{PD18}	Combinational input to output delay for 18×18	2,110	2,990	2,110	3,139	2,010 2,110	3,438	2,110	4,006	ps
t_{PD36}	Combinational input to output delay for 36×36	2,939	4,450	2,939	4,672	2,800 2,939	5,117	2,939	5,962	ps
t_{CLR}	Minimum clear pulse width	2,212		2,322		2,543 2,543		2,964		ps

Table 5–39. DSP Block Internal Timing Microparameters (Part 2 of 2)

Symbol	Parameter	-3 Speed Grade (1)		-3 Speed Grade (2)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (3)	Max	Min (3)	Max	Min (4)	Max	Min (3)	Max	
t_{CLKL}	Minimum clock low time	1,190		1,249		1,368 1,368		1,594		ps
t_{CLKH}	Minimum clock high time	1,190		1,249		1,368 1,368		1,594		ps

Notes to Table 5–39:

- (1) These numbers apply to -3 speed grade EP2S15, EP2S30, EP2S60, and EP2S90 devices.
- (2) These numbers apply to -3 speed grade EP2S130 and EP2S180 devices.
- (3) For the -3 and -5 speed grades, the minimum timing is for the commercial temperature grade. Only -4 speed grade devices offer the industrial temperature grade.
- (4) For the -4 speed grade, the first number is the minimum timing parameter for industrial devices. The second number is the minimum timing parameter for commercial devices.

Table 5–40. M512 Block Internal Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	-3 Speed Grade (2)		-3 Speed Grade (3)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (4)	Max	Min (4)	Max	Min (5)	Max	Min (4)	Max	
t_{M512RC}	Synchronous read cycle time	2,089	2,318	2,089	2,433	1,989 2,089	2,664	2,089	3,104	ps
$t_{M512WERESU}$	Write or read enable setup time before clock	22		23		25 25		29		ps
$t_{M512WEREH}$	Write or read enable hold time after clock	203		213		233 233		272		ps
$t_{M512DATASU}$	Data setup time before clock	22		23		25 25		29		ps
$t_{M512DATAH}$	Data hold time after clock	203		213		233 233		272		ps
$t_{M512WADDRSU}$	Write address setup time before clock	22		23		25 25		29		ps
$t_{M512WADDRH}$	Write address hold time after clock	203		213		233 233		272		ps
$t_{M512RADDRSU}$	Read address setup time before clock	22		23		25 25		29		ps
$t_{M512RADDRH}$	Read address hold time after clock	203		213		233 233		272		ps

Clock Network Skew Adders

The Quartus II software models skew within dedicated clock networks such as global and regional clocks. Therefore, intra-clock network skew adder is not specified. Table 5–68 specifies the clock skew between any two clock networks driving registers in the IOE.

Table 5–68. Clock Network Specifications

Name	Description	Min	Typ	Max	Unit
Clock skew adder EP2S15, EP2S30, EP2S60 (1)	Inter-clock network, same side			±50	ps
	Inter-clock network, entire chip			±100	ps
Clock skew adder EP2S90 (1)	Inter-clock network, same side			±55	ps
	Inter-clock network, entire chip			±110	ps
Clock skew adder EP2S130 (1)	Inter-clock network, same side			±63	ps
	Inter-clock network, entire chip			±125	ps
Clock skew adder EP2S180 (1)	Inter-clock network, same side			±75	ps
	Inter-clock network, entire chip			±150	ps

Note to Table 5–68:

(1) This is in addition to intra-clock network skew, which is modeled in the Quartus II software.

I/O Delays

See [Tables 5–72](#) through [5–76](#) for I/O delays.

Table 5–72. I/O Delay Parameters

Symbol	Parameter
t_{DIP}	Delay from I/O datain to output pad
t_{OP}	Delay from I/O output register to output pad
t_{PCOUT}	Delay from input pad to I/O dataout to core
t_{PI}	Delay from input pad to I/O input register

Table 5–73. Stratix II I/O Input Delay for Column Pins (Part 1 of 3)

I/O Standard	Parameter	Minimum Timing		-3 Speed Grade (2)	-3 Speed Grade (3)	-4 Speed Grade	-5 Speed Grade	Unit
		Industrial	Commercial					
LVTTTL	t_{PI}	674	707	1223	1282	1405	1637	ps
	t_{PCOUT}	408	428	787	825	904	1054	ps
2.5 V	t_{PI}	684	717	1210	1269	1390	1619	ps
	t_{PCOUT}	418	438	774	812	889	1036	ps
1.8 V	t_{PI}	747	783	1366	1433	1570	1829	ps
	t_{PCOUT}	481	504	930	976	1069	1246	ps
1.5 V	t_{PI}	749	786	1436	1506	1650	1922	ps
	t_{PCOUT}	483	507	1000	1049	1149	1339	ps
LVCMOS	t_{PI}	674	707	1223	1282	1405	1637	ps
	t_{PCOUT}	408	428	787	825	904	1054	ps
SSTL-2 Class I	t_{PI}	507	530	818	857	939	1094	ps
	t_{PCOUT}	241	251	382	400	438	511	ps
SSTL-2 Class II	t_{PI}	507	530	818	857	939	1094	ps
	t_{PCOUT}	241	251	382	400	438	511	ps
SSTL-18 Class I	t_{PI}	543	569	898	941	1031	1201	ps
	t_{PCOUT}	277	290	462	484	530	618	ps
SSTL-18 Class II	t_{PI}	543	569	898	941	1031	1201	ps
	t_{PCOUT}	277	290	462	484	530	618	ps
1.5-V HSTL Class I	t_{PI}	560	587	993	1041	1141	1329	ps
	t_{PCOUT}	294	308	557	584	640	746	ps

Table 5–73. Stratix II I/O Input Delay for Column Pins (Part 2 of 3)

I/O Standard	Parameter	Minimum Timing		-3 Speed Grade (2)	-3 Speed Grade (3)	-4 Speed Grade	-5 Speed Grade	Unit
		Industrial	Commercial					
1.5-V HSTL Class II	t _{PI}	560	587	993	1041	1141	1329	ps
	t _{PCOUT}	294	308	557	584	640	746	ps
1.8-V HSTL Class I	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
1.8-V HSTL Class II	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
PCI	t _{PI}	679	712	1214	1273	1395	1625	ps
	t _{PCOUT}	413	433	778	816	894	1042	ps
PCI-X	t _{PI}	679	712	1214	1273	1395	1625	ps
	t _{PCOUT}	413	433	778	816	894	1042	ps
Differential SSTL-2 Class I (1)	t _{PI}	507	530	818	857	939	1094	ps
	t _{PCOUT}	241	251	382	400	438	511	ps
Differential SSTL-2 Class II (1)	t _{PI}	507	530	818	857	939	1094	ps
	t _{PCOUT}	241	251	382	400	438	511	ps
Differential SSTL-18 Class I (1)	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
Differential SSTL-18 Class II (1)	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
1.8-V Differential HSTL Class I (1)	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
1.8-V Differential HSTL Class II (1)	t _{PI}	543	569	898	941	1031	1201	ps
	t _{PCOUT}	277	290	462	484	530	618	ps
1.5-V Differential HSTL Class I (1)	t _{PI}	560	587	993	1041	1141	1329	ps
	t _{PCOUT}	294	308	557	584	640	746	ps
1.5-V Differential HSTL Class II (1)	t _{PI}	560	587	993	1041	1141	1329	ps
	t _{PCOUT}	294	308	557	584	640	746	ps

Table 5–102 shows the JTAG timing parameters and values for Stratix II devices.

Table 5–102. Stratix II JTAG Timing Parameters & Values				
Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	30		ns
t_{JCH}	TCK clock high time	13		ns
t_{JCL}	TCK clock low time	13		ns
t_{JPSU}	JTAG port setup time	3		ns
t_{JPH}	JTAG port hold time	5		ns
t_{JPCO}	JTAG port clock to output		11 (1)	ns
t_{JPZX}	JTAG port high impedance to valid output		14 (1)	ns
t_{JPXZ}	JTAG port valid output to high impedance		14 (1)	ns

Note to Table 5–102:

- (1) A 1 ns adder is required for each V_{CCIO} voltage step down from 3.3 V. For example, $t_{JPCO} = 12$ ns if V_{CCIO} of the TDO I/O bank = 2.5 V, or 13 ns if it equals 1.8 V.

Document Revision History

Table 5–103 shows the revision history for this chapter.

Table 5–103. Document Revision History (Part 1 of 3)		
Date and Document Version	Changes Made	Summary of Changes
April 2011, v4.5	Updated Table 5–3.	Added operating junction temperature for military use.
July 2009, v4.4	Updated Table 5–92.	Updated the spread spectrum modulation frequency (f_{SS}) from (100 kHz–500 kHz) to (30 kHz–150 kHz).
May 2007, v4.3	<ul style="list-style-type: none"> Updated R_{CONF} in Table 5–4. Updated f_{IN} (min) in Table 5–92. Updated f_{IN} and f_{INPFD} in Table 5–93. 	—
	Moved the Document Revision History section to the end of the chapter.	—