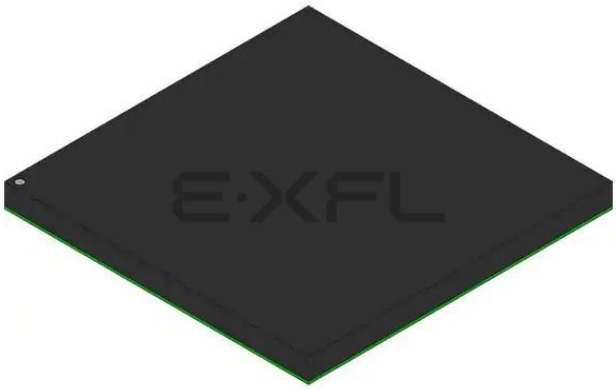

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

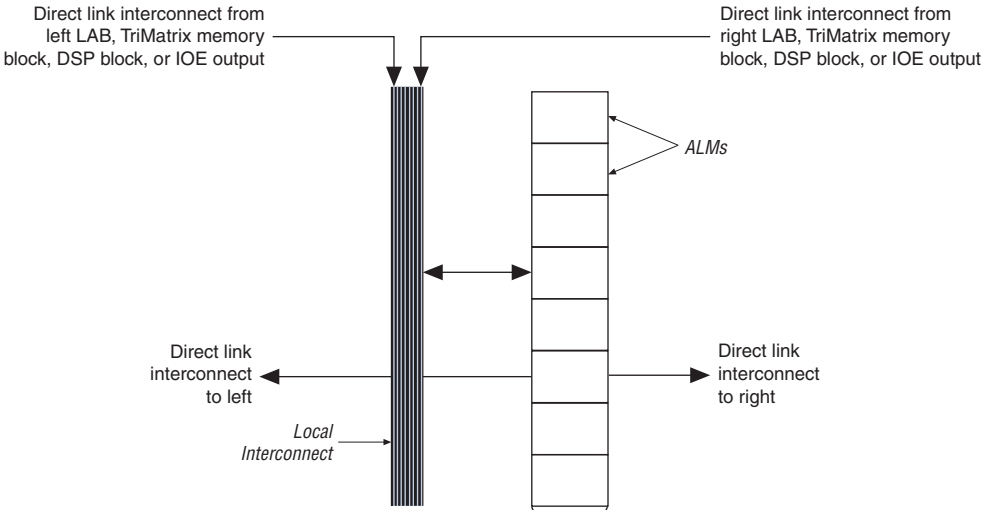
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	8970
Number of Logic Elements/Cells	179400
Total RAM Bits	9383040
Number of I/O	1170
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1508-BBGA, FCBGA
Supplier Device Package	1508-FBGA (40x40)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=ep2s180f1508c4

Figure 2–3. Direct Link Connection



LAB Control Signals

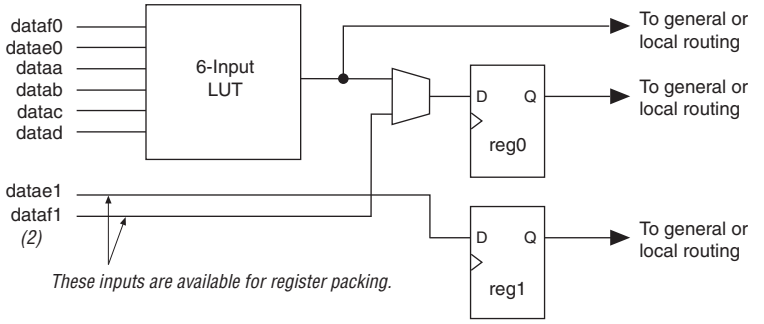
Each LAB contains dedicated logic for driving control signals to its ALMs. The control signals include three clocks, three clock enables, two asynchronous clears, synchronous clear, asynchronous preset/load, and synchronous load control signals. This gives a maximum of 11 control signals at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

Each LAB can use three clocks and three clock enable signals. However, there can only be up to two unique clocks per LAB, as shown in the LAB control signal generation circuit in Figure 2–4. Each LAB's clock and clock enable signals are linked. For example, any ALM in a particular LAB using the `labclk1` signal also uses `labclkena1`. If the LAB uses both the rising and falling edges of a clock, it also uses two LAB-wide clock signals. De-asserting the clock enable signal turns off the corresponding LAB-wide clock.

Each LAB can use two asynchronous clear signals and an asynchronous load/preset signal. By default, the Quartus II software uses a `NOT` gate push-back technique to achieve preset. If you disable the `NOT` gate push-up option or assign a given register to power up high using the Quartus II software, the preset is achieved using the asynchronous load

datae1 and dataf1 are utilized, the output drives to register1 and/or bypasses register1 and drives to the interconnect using the bottom set of output drivers. The Quartus II Compiler automatically selects the inputs to the LUT. Asynchronous load data for the register comes from the datae or dataf input of the ALM. ALMs in normal mode support register packing.

Figure 2–9. 6-Input Function in Normal Mode *Notes (1), (2)*



Notes to Figure 2–9:

(1) If datae1 and dataf1 are used as inputs to the six-input function, then datae0 and dataf0 are available for register packing.

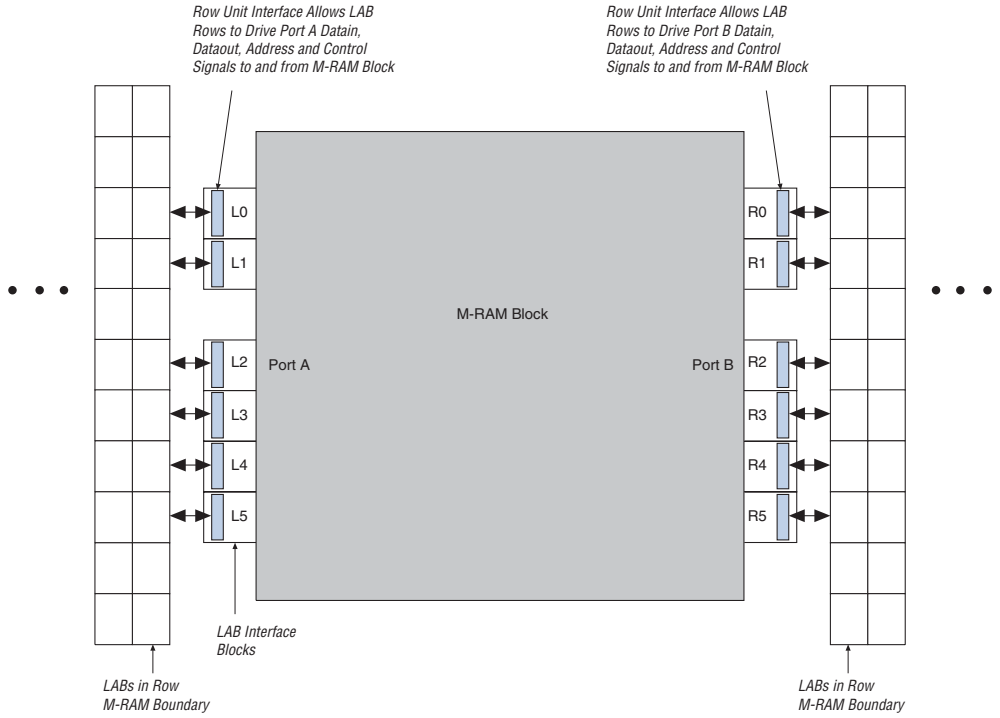
(2) The dataf1 input is available for register packing only if the six-input function is un-registered.

Extended LUT Mode

The extended LUT mode is used to implement a specific set of seven-input functions. The set must be a 2-to-1 multiplexer fed by two arbitrary five-input functions sharing four inputs. Figure 2–10 shows the template of supported seven-input functions utilizing extended LUT mode. In this mode, if the seven-input function is unregistered, the unused eighth input is available for register packing.

Functions that fit into the template shown in Figure 2–10 occur naturally in designs. These functions often appear in designs as "if-else" statements in Verilog HDL or VHDL code.

Figure 2–25. M-RAM Block LAB Row Interface *Note (1)*



Note to Figure 2–25:
(1) Only R24 and C16 interconnects cross the M-RAM block boundaries.

Figure 2–26. M-RAM Row Unit Interface to Interconnect

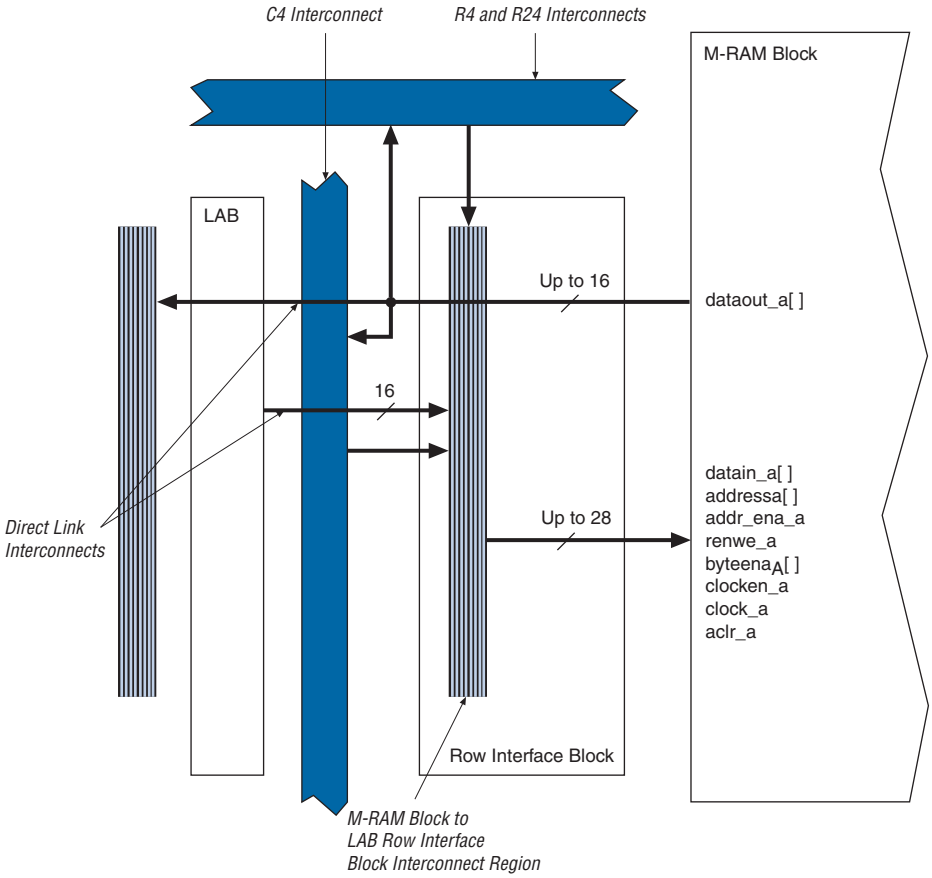


Table 2–4 shows the input and output data signal connections along with the address and control signal input connections to the row unit interfaces (L0 to L5 and R0 to R5).

Modes of Operation

The adder, subtractor, and accumulate functions of a DSP block have four modes of operation:

- Simple multiplier
- Multiply-accumulator
- Two-multipliers adder
- Four-multipliers adder

Table 2–6 shows the different number of multipliers possible in each DSP block mode according to size. These modes allow the DSP blocks to implement numerous applications for DSP including FFTs, complex FIR, FIR, and 2D FIR filters, equalizers, IIR, correlators, matrix multiplication and many other functions. The DSP blocks also support mixed modes and mixed multiplier sizes in the same block. For example, half of one DSP block can implement one 18 × 18-bit multiplier in multiply-accumulator mode, while the other half of the DSP block implements four 9 × 9-bit multipliers in simple multiplier mode.

Table 2–6. Multiplier Size & Configurations per DSP Block

DSP Block Mode	9 × 9	18 × 18	36 × 36
Multiplier	Eight multipliers with eight product outputs	Four multipliers with four product outputs	One multiplier with one product output
Multiply-accumulator	-	Two 52-bit multiply-accumulate blocks	-
Two-multipliers adder	Four two-multiplier adder (two 9 × 9 complex multiply)	Two two-multiplier adder (one 18 × 18 complex multiply)	-
Four-multipliers adder	Two four-multiplier adder	One four-multiplier adder	-

DSP Block Interface

Stratix II device DSP block input registers can generate a shift register that can cascade down in the same DSP block column. Dedicated connections between DSP blocks provide fast connections between the shift register inputs to cascade the shift register chains. You can cascade registers within multiple DSP blocks for 9 × 9- or 18 × 18-bit FIR filters larger than four taps, with additional adder stages implemented in ALMs. If the DSP block is configured as 36 × 36 bits, the adder, subtractor, or accumulator stages are implemented in ALMs. Each DSP block can route the shift register chain out of the block to cascade multiple columns of DSP blocks.

PLLs & Clock Networks

Stratix II devices provide a hierarchical clock structure and multiple PLLs with advanced features. The large number of clocking resources in combination with the clock synthesis precision provided by enhanced and fast PLLs provides a complete clock management solution.

Global & Hierarchical Clocking

Stratix II devices provide 16 dedicated global clock networks and 32 regional clock networks (eight per device quadrant). These clocks are organized into a hierarchical clock structure that allows for up to 24 clocks per device region with low skew and delay. This hierarchical clocking scheme provides up to 48 unique clock domains in Stratix II devices.

There are 16 dedicated clock pins (CLK[15..0]) to drive either the global or regional clock networks. Four clock pins drive each side of the device, as shown in Figures 2–31 and 2–32. Internal logic and enhanced and fast PLL outputs can also drive the global and regional clock networks. Each global and regional clock has a clock control block, which controls the selection of the clock source and dynamically enables/disables the clock to reduce power consumption. Table 2–8 shows global and regional clock features.

Table 2–8. Global & Regional Clock Features

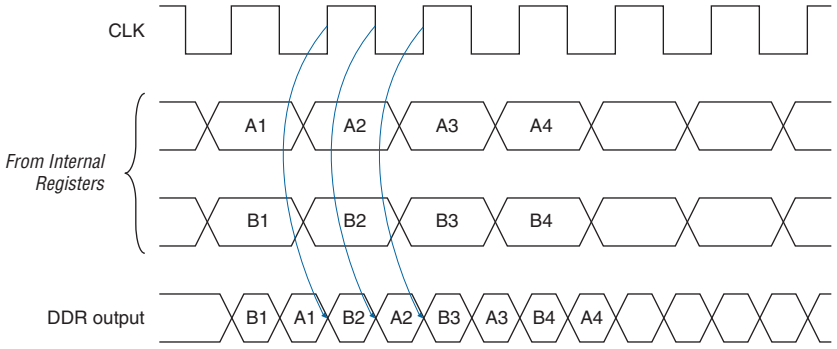
Feature	Global Clocks	Regional Clocks
Number per device	16	32
Number available per quadrant	16	8
Sources	CLK pins, PLL outputs, or internal logic	CLK pins, PLL outputs, or internal logic
Dynamic clock source selection	✓ (1)	
Dynamic enable/disable	✓	✓

Note to Table 2–8:
(1) Dynamic source clock selection is supported for selecting between CLKp pins and PLL outputs only.

Global Clock Network

These clocks drive throughout the entire device, feeding all device quadrants. The global clock networks can be used as clock sources for all resources in the device-IOEs, ALMs, DSP blocks, and all memory blocks. These resources can also be used for control signals, such as clock enables and synchronous or asynchronous clears fed from the external pin. The

Figure 2–55. Output TIming Diagram in DDR Mode



The Stratix II IOE operates in bidirectional DDR mode by combining the DDR input and DDR output configurations. The negative-edge-clocked OE register holds the OE signal inactive until the falling edge of the clock. This is done to meet DDR SDRAM timing requirements.

External RAM Interfacing

In addition to the six I/O registers in each IOE, Stratix II devices also have dedicated phase-shift circuitry for interfacing with external memory interfaces. Stratix II devices support DDR and DDR2 SDRAM, QDR II SRAM, RLDRAM II, and SDR SDRAM memory interfaces. In every Stratix II device, the I/O banks at the top (banks 3 and 4) and bottom (banks 7 and 8) of the device support DQ and DQS signals with DQ bus modes of ×4, ×8/×9, ×16/×18, or ×32/×36. Table 2–14 shows the number of DQ and DQS buses that are supported per device.

Table 2–14. DQS & DQ Bus Mode Support (Part 1 of 2) *Note (1)*

Device	Package	Number of ×4 Groups	Number of ×8/×9 Groups	Number of ×16/×18 Groups	Number of ×32/×36 Groups
EP2S15	484-pin FineLine BGA	8	4	0	0
	672-pin FineLine BGA	18	8	4	0
EP2S30	484-pin FineLine BGA	8	4	0	0
	672-pin FineLine BGA	18	8	4	0
EP2S60	484-pin FineLine BGA	8	4	0	0
	672-pin FineLine BGA	18	8	4	0
	1,020-pin FineLine BGA	36	18	8	4

Table 2–19. Board Design Recommendations for nCEO

nCE Input Buffer Power in I/O Bank 3	Stratix II nCEO V_{CCIO} Voltage Level in I/O Bank 7				
	V_{CCIO} = 3.3 V	V_{CCIO} = 2.5 V	V_{CCIO} = 1.8 V	V_{CCIO} = 1.5 V	V_{CCIO} = 1.2 V
VCCSEL high (V_{CCIO} Bank 3 = 1.5 V)	✓ *(1), (2)*	✓ *(3), (4)*	✓ *(5)*	✓	✓
VCCSEL high (V_{CCIO} Bank 3 = 1.8 V)	✓ *(1), (2)*	✓ *(3), (4)*	✓	✓	Level shifter required
VCCSEL low (nCE Powered by V_{CCPD} = 3.3V)	✓	✓ *(4)*	✓ *(6)*	Level shifter required	Level shifter required

Notes to Table 2–19:

(1) Input buffer is 3.3-V tolerant.
(2) The nCEO output buffer meets V_{OH} (MIN) = 2.4 V.
(3) Input buffer is 2.5-V tolerant.
(4) The nCEO output buffer meets V_{OH} (MIN) = 2.0 V.
(5) Input buffer is 1.8-V tolerant.
(6) An external 250-Ω pull-up resistor is not required, but recommended if signal levels on the board are not optimal.

For JTAG chains, the TDO pin of the first device drives the TDI pin of the second device in the chain. The V_{CCSEL} input on JTAG input I/O cells (TCK, TMS, TDI, and TRST) is internally hardwired to GND selecting the 3.3-V/2.5-V input buffer powered by V_{CCPD}. The ideal case is to have the V_{CCIO} of the TDO bank from the first device to match the V_{CCSEL} settings for TDI on the second device, but that may not be possible depending on the application. Table 2–20 contains board design recommendations to ensure proper JTAG chain operation.

Table 2–20. Supported TDO/TDI Voltage Combinations (Part 1 of 2)

Device	TDI Input Buffer Power	Stratix II TDO V_{CCIO} Voltage Level in I/O Bank 4				
		V_{CCIO} = 3.3 V	V_{CCIO} = 2.5 V	V_{CCIO} = 1.8 V	V_{CCIO} = 1.5 V	V_{CCIO} = 1.2 V
Stratix II	Always V_{CCPD} (3.3V)	✓ *(1)*	✓ *(2)*	✓ *(3)*	Level shifter required	Level shifter required

Table 2–23. EP2S60 Differential Channels *Note (1)*

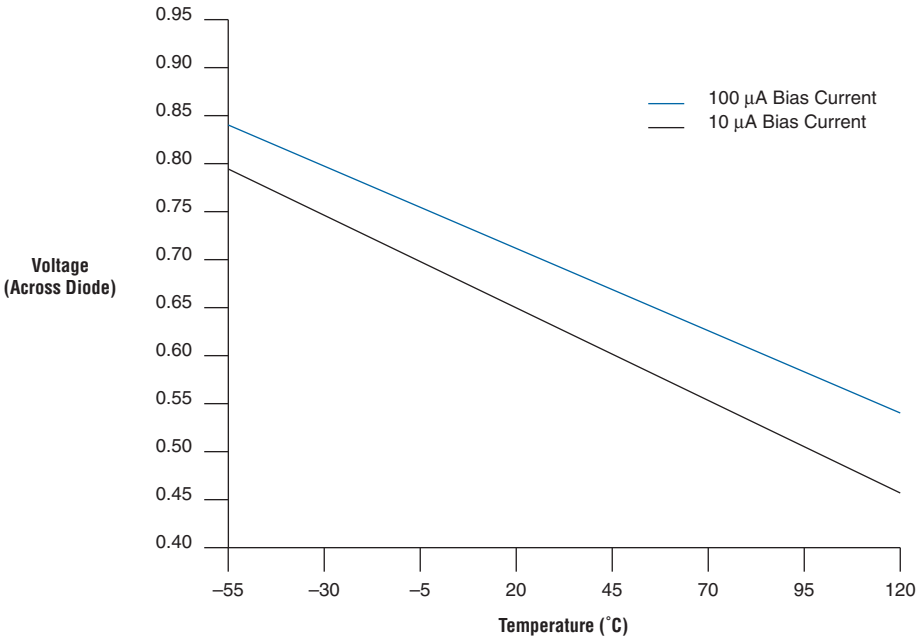
Package	Transmitter/ Receiver	Total Channels	Center Fast PLLs				Corner Fast PLLs *(4)*			
			PLL 1	PLL 2	PLL 3	PLL 4	PLL 7	PLL 8	PLL 9	PLL 10
484-pin FineLine BGA	Transmitter	38 *(2)*	10	9	9	10	10	9	9	10
		(3)	19	19	19	19	-	-	-	-
	Receiver	42 *(2)*	11	10	10	11	11	10	10	11
		(3)	21	21	21	21	-	-	-	-
672-pin FineLine BGA	Transmitter	58 *(2)*	16	13	13	16	16	13	13	16
		(3)	29	29	29	29	-	-	-	-
	Receiver	62 *(2)*	17	14	14	17	17	14	14	17
		(3)	31	31	31	31	-	-	-	-
1,020-pin FineLine BGA	Transmitter	84 *(2)*	21	21	21	21	21	21	21	21
		(3)	42	42	42	42	-	-	-	-
	Receiver	84 *(2)*	21	21	21	21	21	21	21	21
		(3)	42	42	42	42	-	-	-	-

Table 2–24. EP2S90 Differential Channels *Note (1)*

Package	Transmitter/ Receiver	Total Channels	Center Fast PLLs				Corner Fast PLLs *(4)*			
			PLL 1	PLL 2	PLL 3	PLL 4	PLL 7	PLL 8	PLL 9	PLL 10
484-pin Hybrid FineLine BGA	Transmitter	38 *(2)*	10	9	9	10	-	-	-	-
		(3)	19	19	19	19	-	-	-	-
	Receiver	42 *(2)*	11	10	10	11	-	-	-	-
		(3)	21	21	21	21	-	-	-	-
780-pin FineLine BGA	Transmitter	64 *(2)*	16	16	16	16	-	-	-	
		(3)	32	32	32	32	-	-	-	-
	Receiver	68 *(2)*	17	17	17	17	-	-	-	-
		(3)	34	34	34	34	-	-	-	
1,020-pin FineLine BGA	Transmitter	90 *(2)*	23	22	22	23	23	22	22	23
		(3)	45	45	45	45	-	-	-	-
	Receiver	94 *(2)*	23	24	24	23	23	24	24	23
		(3)	46	46	46	46	-	-	-	-
1,508-pin FineLine BGA	Transmitter	118 *(2)*	30	29	29	30	30	29	29	30
		(3)	59	59	59	59	-	-	-	-
	Receiver	118 *(2)*	30	29	29	30	30	29	29	30
		(3)	59	59	59	59	-	-	-	-

The temperature-sensing diode works for the entire operating range, as shown in Figure 3–2.

Figure 3–2. Temperature vs. Temperature-Sensing Diode Voltage



The temperature sensing diode is a very sensitive circuit which can be influenced by noise coupled from other traces on the board, and possibly within the device package itself, depending on device usage. The interfacing device registers temperature based on milivolts of difference as seen at the TSD. Switching I/O near the TSD pins can affect the temperature reading. Altera recommends you take temperature readings during periods of no activity in the device (for example, standby mode where no clocks are toggling in the device), such as when the nearby I/Os are at a DC state, and disable clock networks in the device.

Automated Single Event Upset (SEU) Detection

Stratix II devices offer on-chip circuitry for automated checking of single event upset (SEU) detection. Some applications that require the device to operate error free at high elevations or in close proximity to Earth's North or South Pole require periodic checks to ensure continued data integrity. The error detection cyclic redundancy check (CRC) feature controlled by

Preliminary status means the timing model is subject to change. Initially, timing numbers are created using simulation results, process data, and other known parameters. These tests are used to make the preliminary numbers as close to the actual timing parameters as possible.

Final timing numbers are based on actual device operation and testing. These numbers reflect the actual performance of the device under worst-case voltage and junction temperature conditions.

Table 5–33. Stratix II Device Timing Model Status

Device	Preliminary	Final
EP2S15		✓
EP2S30		✓
EP2S60		✓
EP2S90		✓
EP2S130		✓
EP2S180		✓

I/O Timing Measurement Methodology

Altera characterizes timing delays at the worst-case process, minimum voltage, and maximum temperature for input register setup time (t_{SU}) and hold time (t_H). The Quartus II software uses the following equations to calculate t_{SU} and t_H timing for Stratix II devices input signals.

t_{SU} = + data delay from input pin to input register
+ micro setup time of the input register
– clock delay from input pin to input register

t_H = – data delay from input pin to input register
+ micro hold time of the input register
+ clock delay from input pin to input register

Figure 5–3 shows the setup and hold timing diagram for input registers.

Table 5–35. Timing Measurement Methodology for Input Pins (Part 2 of 2) *Notes (1)–(4)*

I/O Standard	Measurement Conditions			Measurement Point
	V_{CCIO} (V)	V_{REF} (V)	Edge Rate (ns)	V_{MEAS} (V)
1.8-V HSTL Class II	1.660	0.830	1.660	0.83
1.5-V HSTL Class I	1.375	0.688	1.375	0.6875
1.5-V HSTL Class II	1.375	0.688	1.375	0.6875
1.2-V HSTL with OCT	1.140	0.570	1.140	0.570
Differential SSTL-2 Class I	2.325	1.163	2.325	1.1625
Differential SSTL-2 Class II	2.325	1.163	2.325	1.1625
Differential SSTL-18 Class I	1.660	0.830	1.660	0.83
Differential SSTL-18 Class II	1.660	0.830	1.660	0.83
1.5-V Differential HSTL Class I	1.375	0.688	1.375	0.6875
1.5-V Differential HSTL Class II	1.375	0.688	1.375	0.6875
1.8-V Differential HSTL Class I	1.660	0.830	1.660	0.83
1.8-V Differential HSTL Class II	1.660	0.830	1.660	0.83
LVDS	2.325		0.100	1.1625
HyperTransport	2.325		0.400	1.1625
LVPECL	3.135		0.100	1.5675

Notes to Table 5–35:

(1) Input buffer sees no load at buffer input.
(2) Input measuring point at buffer input is $0.5 \times V_{CCIO}$.
(3) Output measuring point is $0.5 \times V_{CC}$ at internal node.
(4) Input edge rate is 1 V/ns.
(5) Less than 50-mV ripple on V_{CCIO} and V_{CCPD}, V_{CCINT} = 1.15 V with less than 30-mV ripple
(6) V_{CCPD} = 2.97 V, less than 50-mV ripple on V_{CCIO} and V_{CCPD}, V_{CCINT} = 1.15 V

Performance

Table 5–36 shows Stratix II performance for some common designs. All performance values were obtained with the Quartus II software compilation of library of parameterized modules (LPM), or MegaCore® functions for the finite impulse response (FIR) and fast Fourier transform (FFT) designs.

Table 5–59. EP2S90 Row Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.585	1.658	2.757	3.154	3.665	ns
t_{COUT}	1.590	1.663	2.753	3.150	3.660	ns
t_{PLLCIN}	-0.341	-0.341	-0.193	-0.235	-0.278	ns
$t_{PLLCOUT}$	-0.336	-0.336	-0.197	-0.239	-0.283	ns

EP2S130 Clock Timing Parameters

Tables 5–60 through 5–63 show the maximum clock timing parameters for EP2S130 devices.

Table 5–60. EP2S130 Column Pins Regional Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.889	1.981	3.405	3.722	4.326	ns
t_{COUT}	1.732	1.816	3.151	3.444	4.002	ns
t_{PLLCIN}	0.105	0.106	0.226	0.242	0.277	ns
$t_{PLLCOUT}$	-0.052	-0.059	-0.028	-0.036	-0.047	ns

Table 5–61. EP2S130 Column Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.907	1.998	3.420	3.740	4.348	ns
t_{COUT}	1.750	1.833	3.166	3.462	4.024	ns
t_{PLLCIN}	0.134	0.136	0.276	0.296	0.338	ns
$t_{PLLCOUT}$	-0.023	-0.029	0.022	0.018	0.014	ns

Table 5–62. EP2S130 Row Pins Regional Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.680	1.760	3.070	3.351	3.892	ns
t_{COUT}	1.685	1.765	3.066	3.347	3.887	ns
t_{PLLCIN}	-0.113	-0.124	-0.12	-0.138	-0.168	ns
$t_{PLLCOUT}$	-0.108	-0.119	-0.124	-0.142	-0.173	ns

Table 5–63. EP2S130 Row Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.690	1.770	3.075	3.362	3.905	ns
t_{COUT}	1.695	1.775	3.071	3.358	3.900	ns
t_{PLLCIN}	-0.087	-0.097	-0.075	-0.089	-0.11	ns
$t_{PLLCOUT}$	-0.082	-0.092	-0.079	-0.093	-0.115	ns

EP2S180 Clock Timing Parameters

Tables 5–64 through 5–67 show the maximum clock timing parameters for EP2S180 devices.

Table 5–64. EP2S180 Column Pins Regional Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	2.001	2.095	3.643	3.984	4.634	ns
t_{COUT}	1.844	1.930	3.389	3.706	4.310	ns
t_{PLLCIN}	-0.307	-0.297	0.053	0.046	0.048	ns
$t_{PLLCOUT}$	-0.464	-0.462	-0.201	-0.232	-0.276	ns

Table 5–65. EP2S180 Column Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	2.003	2.100	3.652	3.993	4.648	ns
t_{COUT}	1.846	1.935	3.398	3.715	4.324	ns
t_{PLLCIN}	-0.3	-0.29	0.053	0.054	0.058	ns
$t_{PLLCOUT}$	-0.457	-0.455	-0.201	-0.224	-0.266	ns

Table 5–66. EP2S180 Row Pins Regional Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.759	1.844	3.273	3.577	4.162	ns
t_{COUT}	1.764	1.849	3.269	3.573	4.157	ns
t_{PLLCIN}	-0.542	-0.541	-0.317	-0.353	-0.414	ns
$t_{PLLCOUT}$	-0.537	-0.536	-0.321	-0.357	-0.419	ns

Table 5–67. EP2S180 Row Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.763	1.850	3.285	3.588	4.176	ns
t_{COUT}	1.768	1.855	3.281	3.584	4.171	ns
t_{PLLCIN}	-0.542	-0.542	-0.319	-0.355	-0.42	ns
$t_{PLLCOUT}$	-0.537	-0.537	-0.323	-0.359	-0.425	ns

Clock Network Skew Adders

The Quartus II software models skew within dedicated clock networks such as global and regional clocks. Therefore, intra-clock network skew adder is not specified. Table 5–68 specifies the clock skew between any two clock networks driving registers in the IOE.

Table 5–68. Clock Network Specifications

Name	Description	Min	Typ	Max	Unit
Clock skew adder EP2S15, EP2S30, EP2S60 *(1)*	Inter-clock network, same side			±50	ps
	Inter-clock network, entire chip			±100	ps
Clock skew adder EP2S90 *(1)*	Inter-clock network, same side			±55	ps
	Inter-clock network, entire chip			±110	ps
Clock skew adder EP2S130 *(1)*	Inter-clock network, same side			±63	ps
	Inter-clock network, entire chip			±125	ps
Clock skew adder EP2S180 *(1)*	Inter-clock network, same side			±75	ps
	Inter-clock network, entire chip			±150	ps

Note to Table 5–68:
(1) This is in addition to intra-clock network skew, which is modeled in the Quartus II software.

I/O Delays

See Tables 5–72 through 5–76 for I/O delays.

Table 5–72. I/O Delay Parameters

Symbol	Parameter
t_{DIP}	Delay from I/O datain to output pad
t_{OP}	Delay from I/O output register to output pad
t_{PCOUT}	Delay from input pad to I/O dataout to core
t_{PI}	Delay from input pad to I/O input register

Table 5–73. Stratix II I/O Input Delay for Column Pins (Part 1 of 3)

I/O Standard	Parameter	Minimum Timing		-3 Speed Grade (2)	-3 Speed Grade (3)	-4 Speed Grade	-5 Speed Grade	Unit
		Industrial	Commercial					
LVTTL	t_{PI}	674	707	1223	1282	1405	1637	ps
	t_{PCOUT}	408	428	787	825	904	1054	ps
2.5 V	t_{PI}	684	717	1210	1269	1390	1619	ps
	t_{PCOUT}	418	438	774	812	889	1036	ps
1.8 V	t_{PI}	747	783	1366	1433	1570	1829	ps
	t_{PCOUT}	481	504	930	976	1069	1246	ps
1.5 V	t_{PI}	749	786	1436	1506	1650	1922	ps
	t_{PCOUT}	483	507	1000	1049	1149	1339	ps
LVCMOS	t_{PI}	674	707	1223	1282	1405	1637	ps
	t_{PCOUT}	408	428	787	825	904	1054	ps
SSTL-2 Class I	t_{PI}	507	530	818	857	939	1094	ps
	t_{PCOUT}	241	251	382	400	438	511	ps
SSTL-2 Class II	t_{PI}	507	530	818	857	939	1094	ps
	t_{PCOUT}	241	251	382	400	438	511	ps
SSTL-18 Class I	t_{PI}	543	569	898	941	1031	1201	ps
	t_{PCOUT}	277	290	462	484	530	618	ps
SSTL-18 Class II	t_{PI}	543	569	898	941	1031	1201	ps
	t_{PCOUT}	277	290	462	484	530	618	ps
1.5-V HSTL Class I	t_{PI}	560	587	993	1041	1141	1329	ps
	t_{PCOUT}	294	308	557	584	640	746	ps

Table 5–79. Maximum Output Clock Toggle Rate Derating Factors (Part 2 of 5)

I/O Standard	Drive Strength	Maximum Output Clock Toggle Rate Derating Factors (ps/pF)								
		Column I/O Pins			Row I/O Pins			Dedicated Clock Outputs		
		-3	-4	-5	-3	-4	-5	-3	-4	-5
1.8-V LVTTL/LVCMOS	2 mA	951	1421	1421	951	1421	1421	904	1421	1421
	4 mA	405	516	516	405	516	516	393	516	516
	6 mA	261	325	325	261	325	325	253	325	325
	8 mA	223	274	274	223	274	274	224	274	274
	10 mA	194	236	236	-	-	-	199	236	236
	12 mA	174	209	209	-	-	-	180	209	209
1.5-V LVTTL/LVCMOS	2 mA	652	963	963	652	963	963	618	963	963
	4 mA	333	347	347	333	347	347	270	347	347
	6 mA	182	247	247	-	-	-	198	247	247
	8 mA	135	194	194	-	-	-	155	194	194
SSTL-2 Class I	8 mA	364	680	680	364	680	680	350	680	680
	12 mA	163	207	207	163	207	207	188	207	207
SSTL-2 Class II	16 mA	118	147	147	118	147	147	94	147	147
	20 mA	99	122	122	-	-	-	87	122	122
	24 mA	91	116	116	-	-	-	85	116	116
SSTL-18 Class I	4 mA	458	570	570	458	570	570	505	570	570
	6 mA	305	380	380	305	380	380	336	380	380
	8 mA	225	282	282	225	282	282	248	282	282
	10 mA	167	220	220	167	220	220	190	220	220
	12 mA	129	175	175	-	-	-	148	175	175
SSTL-18 Class II	8 mA	173	206	206	-	-	-	155	206	206
	16 mA	150	160	160	-	-	-	140	160	160
	18 mA	120	130	130	-	-	-	110	130	130
	20 mA	109	127	127	-	-	-	94	127	127
SSTL-2 Class I	8 mA	364	680	680	364	680	680	350	680	680
	12 mA	163	207	207	163	207	207	188	207	207
SSTL-2 Class II	16 mA	118	147	147	118	147	147	94	147	147
	20 mA	99	122	122	-	-	-	87	122	122
	24 mA	91	116	116	-	-	-	85	116	116

Table 5–89. High-Speed I/O Specifications for -3 Speed Grade (Part 2 of 2) *Notes (1), (2)*

Symbol	Conditions			-3 Speed Grade			Unit
				Min	Typ	Max	
f_{HSDR} (data rate)	J = 4 to 10 (LVDS, HyperTransport technology)			150		1,040	Mbps
	J = 2 (LVDS, HyperTransport technology)			*(4)*		760	Mbps
	J = 1 (LVDS only)			*(4)*		500	Mbps
$f_{HSDRDPA}$ (DPA data rate)	J = 4 to 10 (LVDS, HyperTransport technology)			150		1,040	Mbps
TCCS	All differential standards			-		200	ps
SW	All differential standards			330		-	ps
Output jitter						190	ps
Output t_{RISE}	All differential I/O standards					160	ps
Output t_{FALL}	All differential I/O standards					180	ps
t_{DUTY}				45	50	55	%
DPA run length						6,400	UI
DPA jitter tolerance	Data channel peak-to-peak jitter			0.44			UI
DPA lock time	**Standard**	**Training Pattern**	**Transition Density**				Number of repetitions
	SPI-4	0000000000 1111111111	10%	256			
	Parallel Rapid I/O	00001111	25%	256			
		10010000	50%	256			
	Miscellaneous	10101010	100%	256			
		01010101		256			

Notes to Table 5–89:

(1) When J = 4 to 10, the SERDES block is used.

(2) When J = 1 or 2, the SERDES block is bypassed.

(3) The input clock frequency and the W factor must satisfy the following fast PLL VCO specification: 150 ≤ input clock frequency × W ≤ 1,040.

(4) The minimum specification is dependent on the clock source (fast PLL, enhanced PLL, clock pin, and so on) and the clock routing resource (global, regional, or local) utilized. The I/O differential buffer and input register do not have a minimum toggle rate.