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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details

Product Status	Obsolete
Number of LABs/CLBs	3022
Number of Logic Elements/Cells	60440
Total RAM Bits	2544192
Number of I/O	718
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1020-BBGA
Supplier Device Package	1020-FBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2s60f1020c3

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- Support for numerous single-ended and differential I/O standards
- High-speed differential I/O support with DPA circuitry for 1-Gbps performance
- Support for high-speed networking and communications bus standards including Parallel RapidIO, SPI-4 Phase 2 (POS-PHY Level 4), HyperTransport™ technology, and SFI-4
- Support for high-speed external memory, including DDR and DDR2 SDRAM, RLDRAM II, QDR II SRAM, and SDR SDRAM
- Support for multiple intellectual property megafunctions from Altera MegaCore® functions and Altera Megafunction Partners Program (AMPPSM) megafunctions
- Support for design security using configuration bitstream encryption
- Support for remote configuration updates

Table 1–1. Stratix II FPGA Family Features

Feature	EP2S15	EP2S30	EP2S60	EP2S90	EP2S130	EP2S180
ALMs	6,240	13,552	24,176	36,384	53,016	71,760
Adaptive look-up tables (ALUTs) (1)	12,480	27,104	48,352	72,768	106,032	143,520
Equivalent LEs (2)	15,600	33,880	60,440	90,960	132,540	179,400
M512 RAM blocks	104	202	329	488	699	930
M4K RAM blocks	78	144	255	408	609	768
M-RAM blocks	0	1	2	4	6	9
Total RAM bits	419,328	1,369,728	2,544,192	4,520,488	6,747,840	9,383,040
DSP blocks	12	16	36	48	63	96
18-bit × 18-bit multipliers (3)	48	64	144	192	252	384
Enhanced PLLs	2	2	4	4	4	4
Fast PLLs	4	4	8	8	8	8
Maximum user I/O pins	366	500	718	902	1,126	1,170

Notes to Table 1–1:

- (1) One ALM contains two ALUTs. The ALUT is the cell used in the Quartus® II software for logic synthesis.
- (2) This is the equivalent number of LEs in a Stratix device (four-input LUT-based architecture).
- (3) These multipliers are implemented using the DSP blocks.

Figure 2–6. Stratix II ALM Details

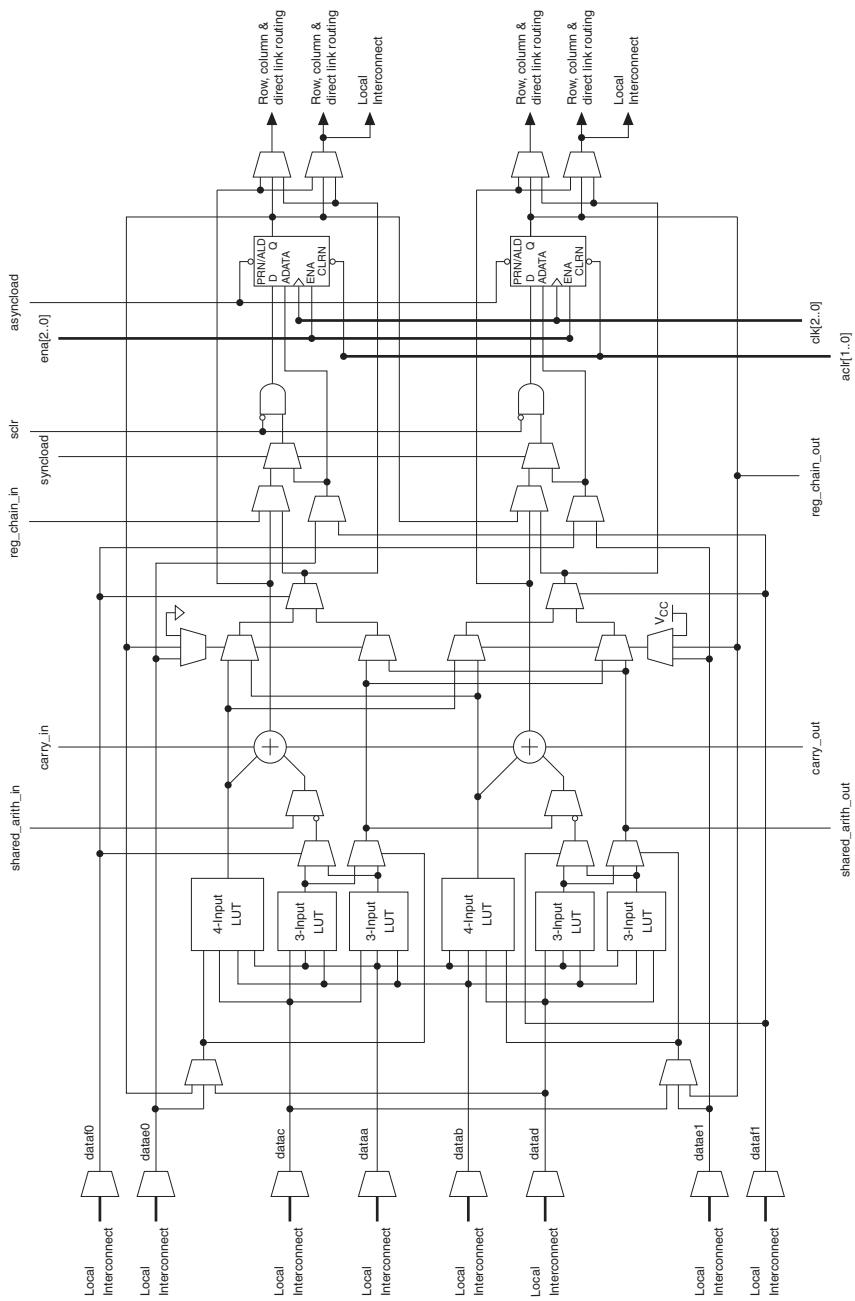
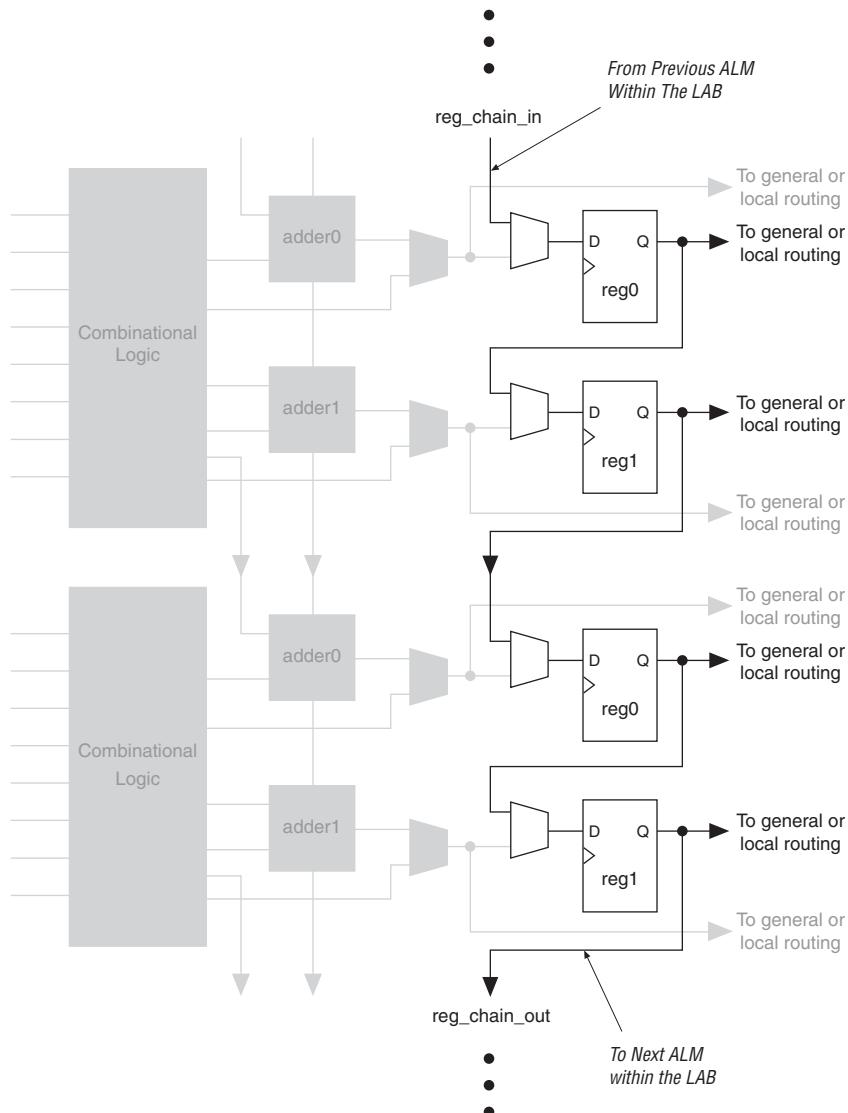


Figure 2–15. Register Chain within an LAB Note (1)**Note to Figure 2–15:**

- (1) The combinational or adder logic can be utilized to implement an unrelated, un-registered function.

See the “[MultiTrack Interconnect](#)” on page 2–22 section for more information on register chain interconnect.

Table 2–5 shows the number of DSP blocks in each Stratix II device.

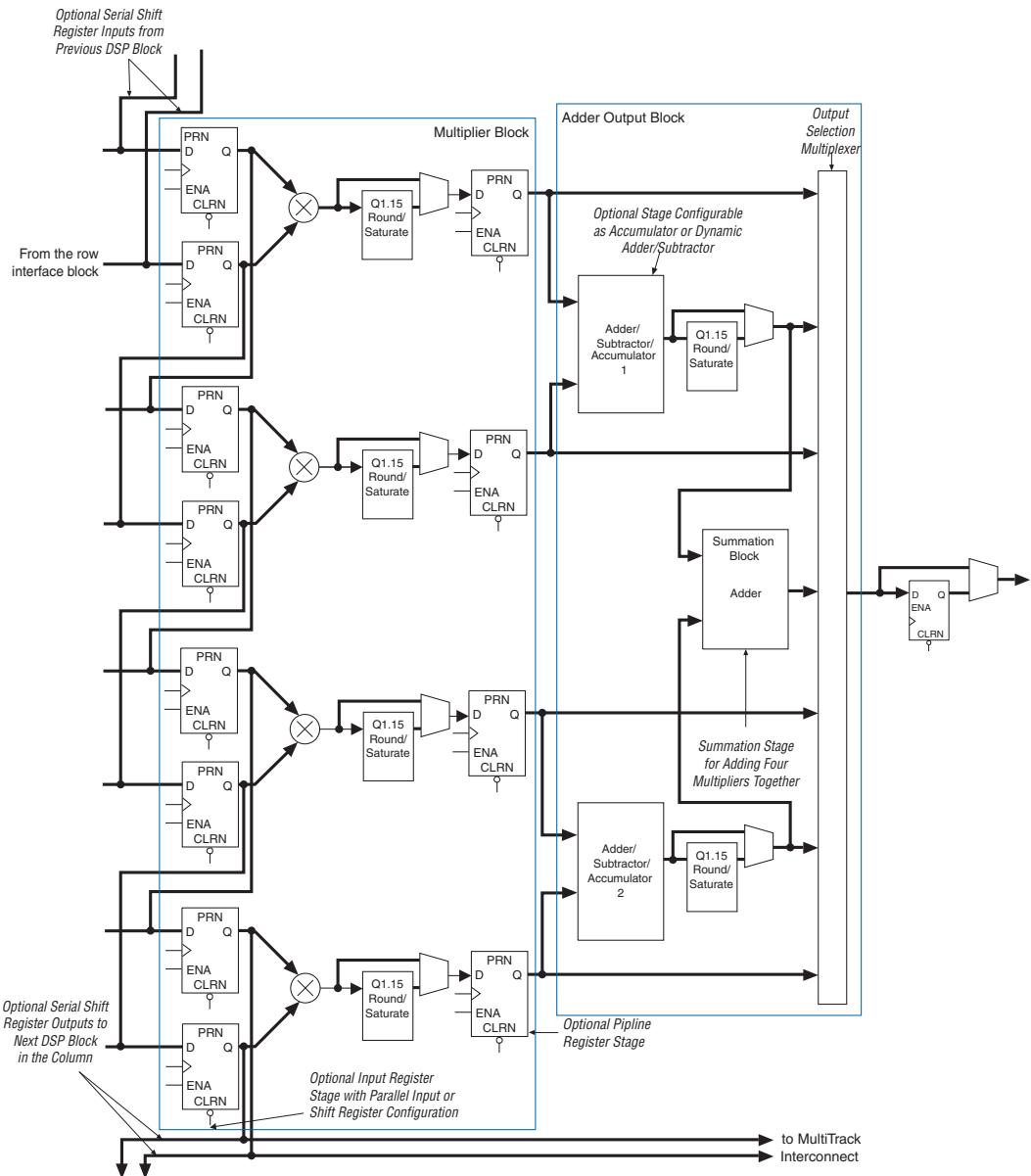
Table 2–5. DSP Blocks in Stratix II Devices Note (1)				
Device	DSP Blocks	Total 9×9 Multipliers	Total 18×18 Multipliers	Total 36×36 Multipliers
EP2S15	12	96	48	12
EP2S30	16	128	64	16
EP2S60	36	288	144	36
EP2S90	48	384	192	48
EP2S130	63	504	252	63
EP2S180	96	768	384	96

Note to Table 2–5:

- (1) Each device has either the numbers of 9×9 -, 18×18 -, or 36×36 -bit multipliers shown. The total number of multipliers for each device is not the sum of all the multipliers.

DSP block multipliers can optionally feed an adder/subtractor or accumulator in the block depending on the configuration. This makes routing to ALMs easier, saves ALM routing resources, and increases performance, because all connections and blocks are in the DSP block. Additionally, the DSP block input registers can efficiently implement shift registers for FIR filter applications, and DSP blocks support Q1.15 format rounding and saturation.

Figure 2–28 shows the top-level diagram of the DSP block configured for 18×18 -bit multiplier mode.

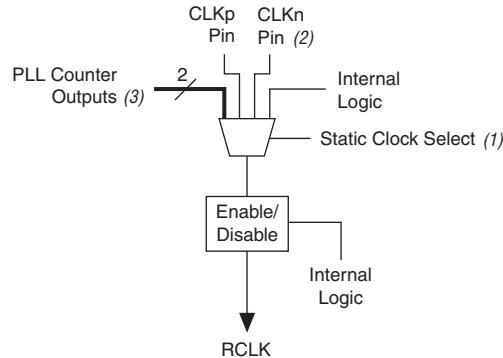
Figure 2–28. DSP Block Diagram for 18 × 18-Bit Configuration

The LAB row source for control signals, data inputs, and outputs is shown in [Table 2-7](#).

Table 2-7. DSP Block Signal Sources & Destinations			
LAB Row at Interface	Control Signals Generated	Data Inputs	Data Outputs
0	clock0 aclr0 ena0 mult01_saturate addnsub1_round/ accum_round addnsub1 signa sourcea sourceb	A1[17..0] B1[17..0]	OA[17..0] OB[17..0]
1	clock1 aclr1 ena1 accum_saturate mult01_round accum_sload sourcea sourceb mode0	A2[17..0] B2[17..0]	OC[17..0] OD[17..0]
2	clock2 aclr2 ena2 mult23_saturate addnsub3_round/ accum_round addnsub3 sign_b sourcea sourceb	A3[17..0] B3[17..0]	OE[17..0] OF[17..0]
3	clock3 aclr3 ena3 accum_saturate mult23_round accum_sload sourcea sourceb mode1	A4[17..0] B4[17..0]	OG[17..0] OH[17..0]

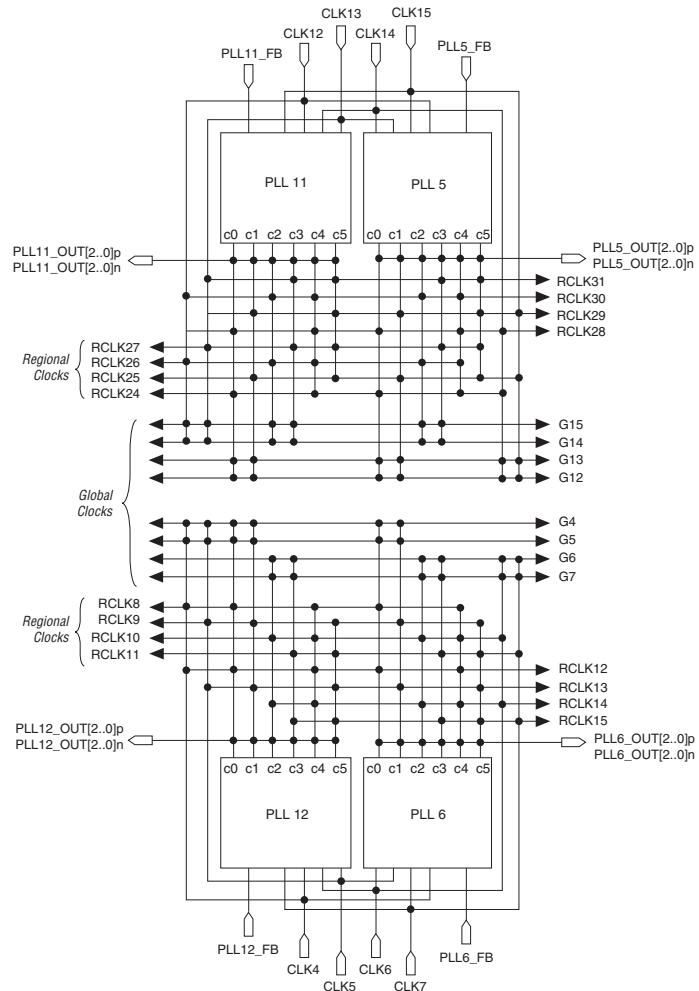


See the *DSP Blocks in Stratix II & Stratix II GX Devices* chapter in volume 2 of the *Stratix II Device Handbook* or the *Stratix II GX Device Handbook*, for more information on DSP blocks.

Figure 2–38. Regional Clock Control Blocks**Notes to Figure 2–38:**

- (1) These clock select signals can only be set through a configuration file (.sof or .pof) and cannot be dynamically controlled during user mode operation.
 - (2) Only the CLKn pins on the top and bottom of the device feed to regional clock select blocks. The clock outputs from corner PLLs cannot be dynamically selected through the global clock control block.
 - (3) The clock outputs from corner PLLs cannot be dynamically selected through the global clock control block.
-

Figure 2–43. Global & Regional Clock Connections from Top & Bottom Clock Pins & Enhanced PLL Outputs
 Notes (1), (2), and (3)



Notes to Figure 2–43:

- (1) EP2S15 and EP2S30 devices only have two enhanced PLLs (5 and 6), but the connectivity from these two PLLs to the global and regional clock networks remains the same as shown.
- (2) If the design uses the feedback input, you lose one (or two, if FBIN is differential) external clock output pin.
- (3) The enhanced PLLs can also be driven through the global or regional clock netwrks. The global or regional clock input can be driven by an output from another PLL, a pin-driven dedicated global or regional clock, or through a clock control block provided the clock control block is fed by an output from another PLL or a pin-driven dedicated global or regional clock. An internally generated global signal cannot drive the PLL.

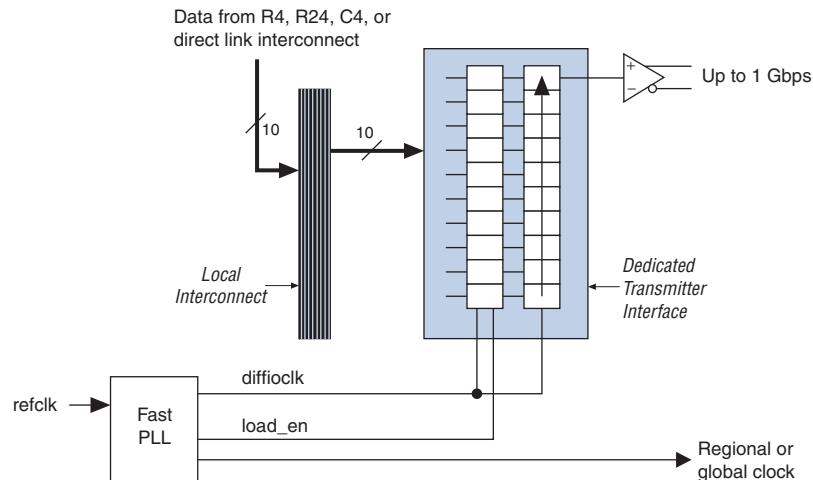
Table 2–17 shows the Stratix II on-chip termination support per I/O bank.

Table 2–17. On-Chip Termination Support by I/O Banks (Part 1 of 2)			
On-Chip Termination Support	I/O Standard Support	Top & Bottom Banks	Left & Right Banks
Series termination without calibration	3.3-V LVTTL	✓	✓
	3.3-V LVC MOS	✓	✓
	2.5-V LVTTL	✓	✓
	2.5-V LVC MOS	✓	✓
	1.8-V LVTTL	✓	✓
	1.8-V LVC MOS	✓	✓
	1.5-V LVTTL	✓	✓
	1.5-V LVC MOS	✓	✓
	SSTL-2 Class I and II	✓	✓
	SSTL-18 Class I	✓	✓
	SSTL-18 Class II	✓	
	1.8-V HSTL Class I	✓	✓
	1.8-V HSTL Class II	✓	
	1.5-V HSTL Class I	✓	✓
	1.2-V HSTL	✓	

Dedicated Circuitry with DPA Support

Stratix II devices support source-synchronous interfacing with LVDS or HyperTransport signaling at up to 1 Gbps. Stratix II devices can transmit or receive serial channels along with a low-speed or high-speed clock. The receiving device PLL multiplies the clock by an integer factor $W = 1$ through 32. For example, a HyperTransport technology application where the data rate is 1,000 Mbps and the clock rate is 500 MHz would require that W be set to 2. The SERDES factor J determines the parallel data width to deserialize from receivers or to serialize for transmitters. The SERDES factor J can be set to 4, 5, 6, 7, 8, 9, or 10 and does not have to equal the PLL clock-multiplication W value. A design using the dynamic phase aligner also supports all of these J factor values. For a J factor of 1, the Stratix II device bypasses the SERDES block. For a J factor of 2, the Stratix II device bypasses the SERDES block, and the DDR input and output registers are used in the IOE. Figure 2–58 shows the block diagram of the Stratix II transmitter channel.

Figure 2–58. Stratix II Transmitter Channel



Each Stratix II receiver channel features a DPA block for phase detection and selection, a SERDES, a synchronizer, and a data realigner circuit. You can bypass the dynamic phase aligner without affecting the basic source-synchronous operation of the channel. In addition, you can dynamically switch between using the DPA block or bypassing the block via a control signal from the logic array. Figure 2–59 shows the block diagram of the Stratix II receiver channel.

The Quartus II software has an Auto Usercode feature where you can choose to use the checksum value of a programming file as the JTAG user code. If selected, the checksum is automatically loaded to the USERCODE register. Turn on the **Auto Usercode** option by clicking **Device & Pin Options**, then **General**, in the **Settings** dialog box (Assignments menu).

Table 3–2. Stratix II Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EP2S15	1,140
EP2S30	1,692
EP2S60	2,196
EP2S90	2,748
EP2S130	3,420
EP2S180	3,948

Table 3–3. 32-Bit Stratix II Device IDCODE

Device	IDCODE (32 Bits) (1)			
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer Identity (11 Bits)	LSB (1 Bit) (2)
EP2S15	0000	0010 0000 1001 0001	000 0110 1110	1
EP2S30	0000	0010 0000 1001 0010	000 0110 1110	1
EP2S60	0001	0010 0000 1001 0011	000 0110 1110	1
EP2S90	0000	0010 0000 1001 0100	000 0110 1110	1
EP2S130	0000	0010 0000 1001 0101	000 0110 1110	1
EP2S180	0000	0010 0000 1001 0110	000 0110 1110	1

Notes to Table 3–3:

- (1) The most significant bit (MSB) is on the left.
- (2) The IDCODE's least significant bit (LSB) is always 1.



Stratix, Stratix II, Cyclone, and Cyclone II devices must be within the first 17 devices in a JTAG chain. All of these devices have the same JTAG controller. If any of the Stratix, Stratix II, Cyclone, and Cyclone II devices are in the 18th or after they fail configuration. This does not affect SignalTap II.

Table 5–17. SSTL-18 Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.71	1.80	1.89	V
V_{REF}	Reference voltage		0.855	0.900	0.945	V
V_{TT}	Termination voltage		$V_{REF} - 0.04$	V_{REF}	$V_{REF} + 0.04$	V
V_{IH} (DC)	High-level DC input voltage		$V_{REF} + 0.125$			V
V_{IL} (DC)	Low-level DC input voltage				$V_{REF} - 0.125$	V
V_{IH} (AC)	High-level AC input voltage		$V_{REF} + 0.25$			V
V_{IL} (AC)	Low-level AC input voltage				$V_{REF} - 0.25$	V
V_{OH}	High-level output voltage	$I_{OH} = -13.4 \text{ mA}$ (1)	$V_{CCIO} - 0.28$			V
V_{OL}	Low-level output voltage	$I_{OL} = 13.4 \text{ mA}$ (1)			0.28	V

Note to Table 5–17:

- (1) This specification is supported across all the programmable drive settings available for this I/O standard as shown in the *Stratix II Architecture* chapter in volume 1 of the *Stratix II Device Handbook*.

Table 5–18. SSTL-18 Class I & II Differential Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.71	1.80	1.89	V
V_{SWING} (DC)	DC differential input voltage		0.25			V
V_x (AC)	AC differential input cross point voltage		$(V_{CCIO}/2) - 0.175$		$(V_{CCIO}/2) + 0.175$	V
V_{SWING} (AC)	AC differential input voltage		0.5			V
V_{ISO}	Input clock signal offset voltage			$0.5 \times V_{CCIO}$		V
ΔV_{ISO}	Input clock signal offset voltage variation			± 200		mV
V_{ox} (AC)	AC differential cross point voltage		$(V_{CCIO}/2) - 0.125$		$(V_{CCIO}/2) + 0.125$	V

Table 5–59. EP2S90 Row Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.585	1.658	2.757	3.154	3.665	ns
t_{COUT}	1.590	1.663	2.753	3.150	3.660	ns
t_{PLLCIN}	-0.341	-0.341	-0.193	-0.235	-0.278	ns
$t_{PLLCOUT}$	-0.336	-0.336	-0.197	-0.239	-0.283	ns

EP2S130 Clock Timing Parameters

Tables 5–60 through 5–63 show the maximum clock timing parameters for EP2S130 devices.

Table 5–60. EP2S130 Column Pins Regional Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.889	1.981	3.405	3.722	4.326	ns
t_{COUT}	1.732	1.816	3.151	3.444	4.002	ns
t_{PLLCIN}	0.105	0.106	0.226	0.242	0.277	ns
$t_{PLLCOUT}$	-0.052	-0.059	-0.028	-0.036	-0.047	ns

Table 5–61. EP2S130 Column Pins Global Clock Timing Parameters

Parameter	Minimum Timing		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
	Industrial	Commercial				
t_{CIN}	1.907	1.998	3.420	3.740	4.348	ns
t_{COUT}	1.750	1.833	3.166	3.462	4.024	ns
t_{PLLCIN}	0.134	0.136	0.276	0.296	0.338	ns
$t_{PLLCOUT}$	-0.023	-0.029	0.022	0.018	0.014	ns

Clock Network Skew Adders

The Quartus II software models skew within dedicated clock networks such as global and regional clocks. Therefore, intra-clock network skew adder is not specified. **Table 5–68** specifies the clock skew between any two clock networks driving registers in the IOE.

Table 5–68. Clock Network Specifications					
Name	Description	Min	Typ	Max	Unit
Clock skew adder EP2S15, EP2S30, EP2S60 (1)	Inter-clock network, same side			± 50	ps
	Inter-clock network, entire chip			± 100	ps
Clock skew adder EP2S90 (1)	Inter-clock network, same side			± 55	ps
	Inter-clock network, entire chip			± 110	ps
Clock skew adder EP2S130 (1)	Inter-clock network, same side			± 63	ps
	Inter-clock network, entire chip			± 125	ps
Clock skew adder EP2S180 (1)	Inter-clock network, same side			± 75	ps
	Inter-clock network, entire chip			± 150	ps

Note to Table 5–68:

- (1) This is in addition to intra-clock network skew, which is modeled in the Quartus II software.

Table 5–75. Stratix II I/O Output Delay for Column Pins (Part 2 of 8)

I/O Standard	Drive Strength	Parameter	Minimum Timing		-3 Speed Grade (3)	-3 Speed Grade (4)	-4 Speed Grade	-5 Speed Grade	Unit
			Industrial	Commercial					
LVCMOS	4 mA	t_{OP}	1041	1091	2036	2136	2340	2448	ps
		t_{DIP}	1061	1113	2102	2206	2416	2538	ps
	8 mA	t_{OP}	952	999	1786	1874	2053	2153	ps
		t_{DIP}	972	1021	1852	1944	2129	2243	ps
	12 mA	t_{OP}	926	971	1720	1805	1977	2075	ps
		t_{DIP}	946	993	1786	1875	2053	2165	ps
	16 mA	t_{OP}	933	978	1693	1776	1946	2043	ps
		t_{DIP}	953	1000	1759	1846	2022	2133	ps
	20 mA	t_{OP}	921	965	1677	1759	1927	2025	ps
		t_{DIP}	941	987	1743	1829	2003	2115	ps
	24 mA (1)	t_{OP}	909	954	1659	1741	1906	2003	ps
		t_{DIP}	929	976	1725	1811	1982	2093	ps
2.5 V	4 mA	t_{OP}	1004	1053	2063	2165	2371	2480	ps
		t_{DIP}	1024	1075	2129	2235	2447	2570	ps
	8 mA	t_{OP}	955	1001	1841	1932	2116	2218	ps
		t_{DIP}	975	1023	1907	2002	2192	2308	ps
	12 mA	t_{OP}	934	980	1742	1828	2002	2101	ps
		t_{DIP}	954	1002	1808	1898	2078	2191	ps
	16 mA (1)	t_{OP}	918	962	1679	1762	1929	2027	ps
		t_{DIP}	938	984	1745	1832	2005	2117	ps

Table 5–82. Maximum DCD for DDIO Output on Row I/O Pins Without PLL in the Clock Path for -3 Devices *Notes (1), (2)*

Row DDIO Output I/O Standard	Maximum DCD Based on I/O Standard of Input Feeding the DDIO Clock Port (No PLL in Clock Path)					Unit
	TTL/CMOS		SSTL-2	SSTL/HSTL	LVDS/ HyperTransport Technology	
	3.3 & 2.5 V	1.8 & 1.5 V	2.5 V	1.8 & 1.5 V	3.3 V	
3.3-V LVTTL	260	380	145	145	110	ps
3.3-V LVCMOS	210	330	100	100	65	ps
2.5 V	195	315	85	85	75	ps
1.8 V	150	265	85	85	120	ps
1.5-V LVCMOS	255	370	140	140	105	ps
SSTL-2 Class I	175	295	65	65	70	ps
SSTL-2 Class II	170	290	60	60	75	ps
SSTL-18 Class I	155	275	55	50	90	ps
1.8-V HSTL Class I	150	270	60	60	95	ps
1.5-V HSTL Class I	150	270	55	55	90	ps
LVDS/ HyperTransport technology	180	180	180	180	180	ps

Notes to Table 5–82:

- (1) The information in Table 5–82 assumes the input clock has zero DCD.
- (2) The DCD specification is based on a no logic array noise condition.

Here is an example for calculating the DCD in percentage for a DDIO output on a row I/O on a -3 device:

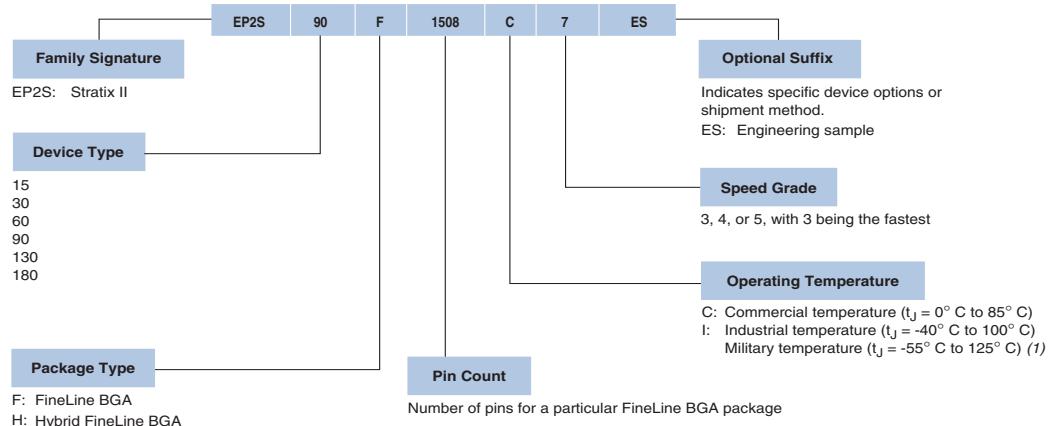
If the input I/O standard is SSTL-2 and the DDIO output I/O standard is SSTL-2 Class II, the maximum DCD is 60 ps (see Table 5–82). If the clock frequency is 267 MHz, the clock period T is:

$$T = 1 / f = 1 / 267 \text{ MHz} = 3.745 \text{ ns} = 3745 \text{ ps}$$

Calculate the DCD as a percentage:

$$(T/2 - DCD) / T = (3745\text{ps}/2 - 60\text{ps}) / 3745\text{ps} = 48.4\% \text{ (for low boundary)}$$

$$(T/2 + DCD) / T = (3745 \text{ ps}/2 + 60 \text{ ps}) / 3745\text{ps} = 51.6\% \text{ (for high boundary)}$$

Figure 6–1. Stratix II Device Packaging Ordering Information**Note to Figure 6–1:**

- (1) Applicable to I4 devices. For more information, refer to the *Stratix II Military Temperature Range Support* technical brief.

Document Revision History

Table 6–1 shows the revision history for this chapter.

Table 6–1. Document Revision History

Date and Document Version	Changes Made	Summary of Changes
April 2011, v2.2	Updated Figure 6–1.	Added operating junction temperature for military use.
May 2007, v2.1	Moved the Document Revision History section to the end of the chapter.	—
January 2005, v2.0	Contact information was removed.	—
October 2004, v1.1	Updated Figure 6–1.	—
February 2004, v1.0	Added document to the Stratix II Device Handbook.	—