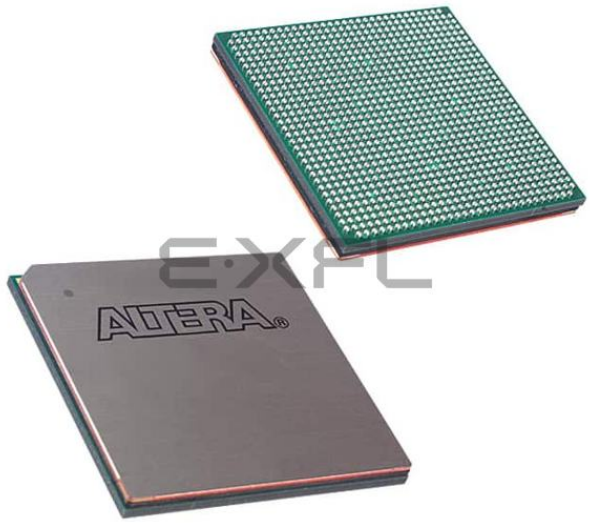

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Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	3022
Number of Logic Elements/Cells	60440
Total RAM Bits	2544192
Number of I/O	718
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1020-BBGA
Supplier Device Package	1020-FBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2s60f1020c5

Chapter Revision Dates

The chapters in this book, *Stratix II Device Handbook, Volume 1*, were revised on the following dates. Where chapters or groups of chapters are available separately, part numbers are listed.

Chapter 1. Introduction
Revised: *May 2007*
Part number: *SII51001-4.2*

Chapter 2. Stratix II Architecture
Revised: *May 2007*
Part number: *SII51002-4.3*

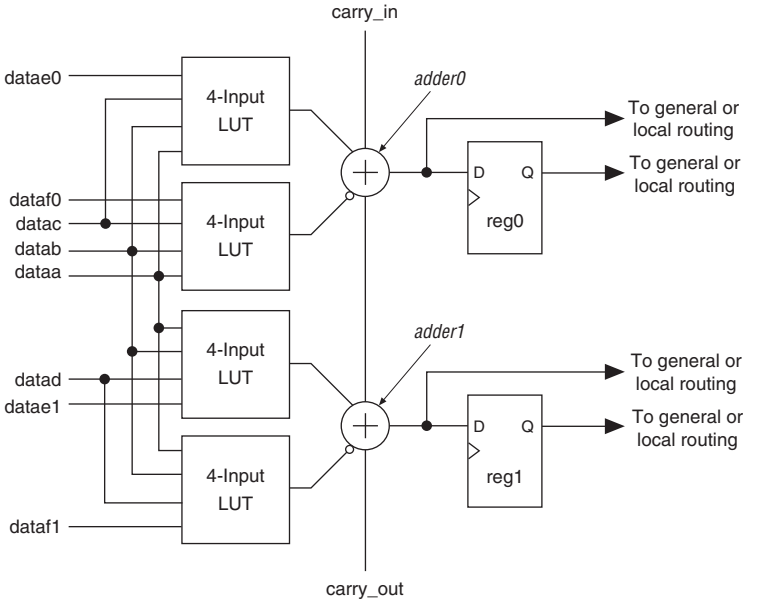
Chapter 3. Configuration & Testing
Revised: *May 2007*
Part number: *SII51003-4.2*

Chapter 4. Hot Socketing & Power-On Reset
Revised: *May 2007*
Part number: *SII51004-3.2*

Chapter 5. DC & Switching Characteristics
Revised: *April 2011*
Part number: *SII51005-4.5*

Chapter 6. Reference & Ordering Information
Revised: *April 2011*
Part number: *SII51006-2.2*

Figure 2–11. ALM in Arithmetic Mode

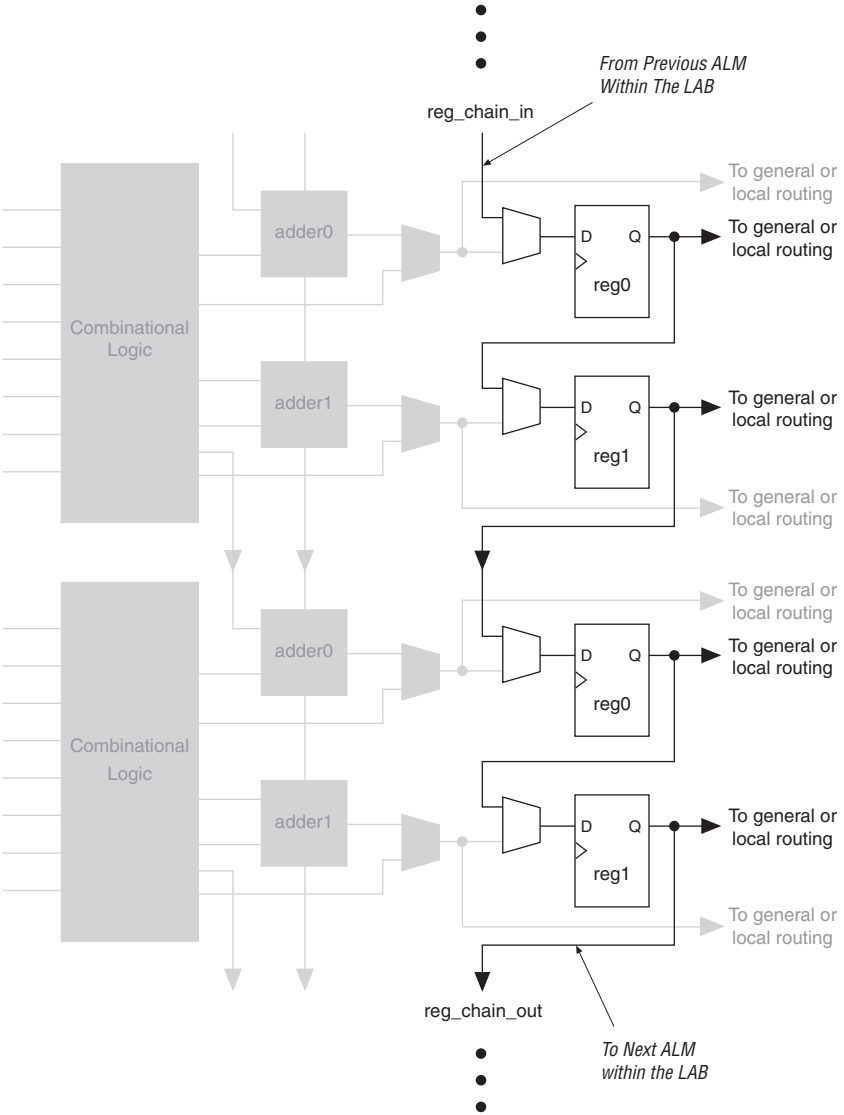


While operating in arithmetic mode, the ALM can support simultaneous use of the adder's carry output along with combinational logic outputs. In this operation, the adder output is ignored. This usage of the adder with the combinational logic output provides resource savings of up to 50% for functions that can use this ability. An example of such functionality is a conditional operation, such as the one shown in Figure 2–12. The equation for this example is:

$$R = (X < Y) ? Y : X$$

To implement this function, the adder is used to subtract 'Y' from 'X.' If 'X' is less than 'Y,' the `carry_out` signal is '1.' The `carry_out` signal is fed to an adder where it drives out to the LAB local interconnect. It then feeds to the LAB-wide `syncload` signal. When asserted, `syncload` selects the `syncdata` input. In this case, the data 'Y' drives the `syncdata` inputs to the registers. If 'X' is greater than or equal to 'Y,' the `syncload` signal is de-asserted and 'X' drives the data port of the registers.

Figure 2–15. Register Chain within an LAB *Note (1)*



Note to Figure 2–15:
(1) The combinational or adder logic can be utilized to implement an unrelated, un-registered function.

See the "MultiTrack Interconnect" on page 2–22 section for more information on register chain interconnect.

M512 RAM Block

The M512 RAM block is a simple dual-port memory block and is useful for implementing small FIFO buffers, DSP, and clock domain transfer applications. Each block contains 576 RAM bits (including parity bits). M512 RAM blocks can be configured in the following modes:

- Simple dual-port RAM
- Single-port RAM
- FIFO
- ROM
- Shift register

Violating the setup or hold time on the memory block address registers could corrupt memory contents. This applies to both read and write operations.

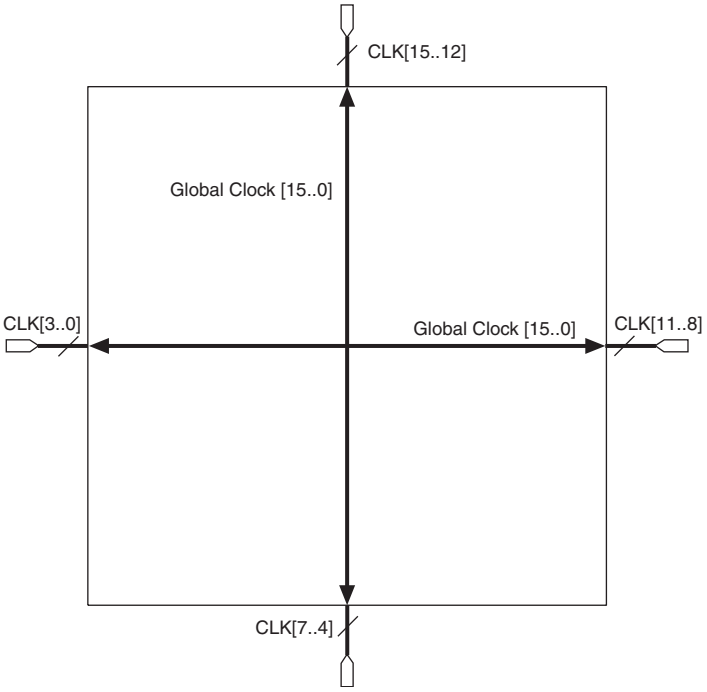
When configured as RAM or ROM, you can use an initialization file to pre-load the memory contents.

M512 RAM blocks can have different clocks on its inputs and outputs. The wren, datain, and write address registers are all clocked together from one of the two clocks feeding the block. The read address, rden, and output registers can be clocked by either of the two clocks driving the block. This allows the RAM block to operate in read/write or input/output clock modes. Only the output register can be bypassed. The six `labclk` signals or local interconnect can drive the `inclock`, `outclock`, `wren`, `rden`, and `outclr` signals. Because of the advanced interconnect between the LAB and M512 RAM blocks, ALMs can also control the `wren` and `rden` signals and the RAM clock, clock enable, and asynchronous clear signals. Figure 2–19 shows the M512 RAM block control signal generation logic.

The RAM blocks in Stratix II devices have local interconnects to allow ALMs and interconnects to drive into RAM blocks. The M512 RAM block local interconnect is driven by the R4, C4, and direct link interconnects from adjacent LABs. The M512 RAM blocks can communicate with LABs on either the left or right side through these row interconnects or with LAB columns on the left or right side with the column interconnects. The M512 RAM block has up to 16 direct link input connections from the left adjacent LABs and another 16 from the right adjacent LAB. M512 RAM outputs can also connect to left and right LABs through direct link interconnect. The M512 RAM block has equal opportunity for access and performance to and from LABs on either its left or right side. Figure 2–20 shows the M512 RAM block to logic array interface.

global clock networks can also be driven by internal logic for internally generated global clocks and asynchronous clears, clock enables, or other control signals with large fanout. Figure 2–31 shows the 16 dedicated CLK pins driving global clock networks.

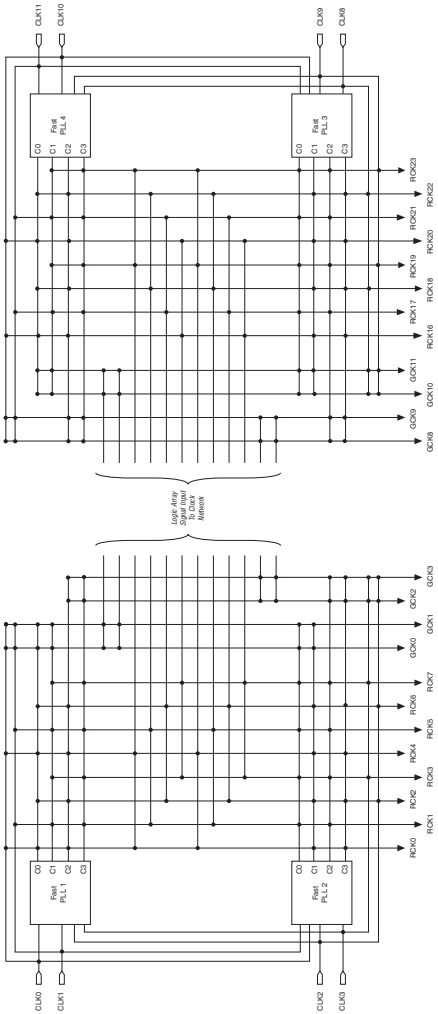
Figure 2–31. Global Clocking



Regional Clock Network

There are eight regional clock networks RCLK[7..0] in each quadrant of the Stratix II device that are driven by the dedicated CLK[15..0] input pins, by PLL outputs, or by internal logic. The regional clock networks provide the lowest clock delay and skew for logic contained in a single quadrant. The CLK clock pins symmetrically drive the RCLK networks in a particular quadrant, as shown in Figure 2–32.

Figure 2–41. Global & Regional Clock Connections from Center Clock Pins & Fast PLL Outputs *Note (1)*



Notes to Figure 2–41:

(1) EP2S15 and EP2S30 devices only have four fast PLLs (1, 2, 3, and 4), but the connectivity from these four PLLs to the global and regional clock networks remains the same as shown.

(2) The global or regional clocks in a fast PLL's quadrant can drive the fast PLL input. The global or regional clock input can be driven by an output from another PLL, a pin-driven dedicated global or regional clock, or through a clock control block, provided the clock control block is fed by an output from another PLL or a pin-driven dedicated global or regional clock. An internally generated global signal cannot drive the PLL.

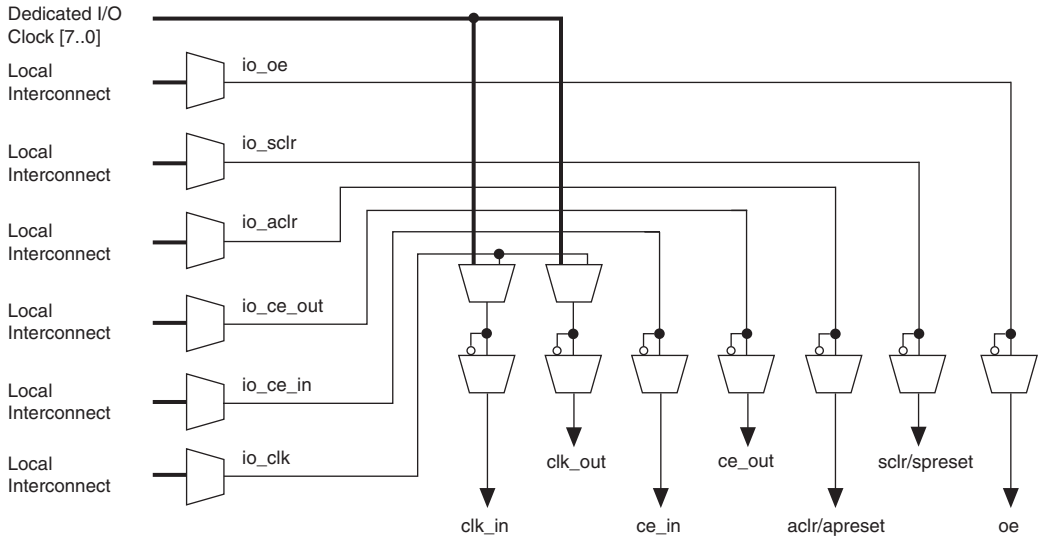
Table 2–11. Global & Regional Clock Connections from Top Clock Pins & Enhanced PLL Outputs (Part 2 of 2)

Top Side Global & Regional Clock Network Connectivity	DLLCLK	CLK12	CLK13	CLK14	CLK15	RCLK24	RCLK25	RCLK26	RCLK27	RCLK28	RCLK29	RCLK30	RCLK31
c4	✓					✓		✓		✓		✓	
c5	✓						✓		✓		✓		✓
Enhanced PLL 11 outputs													
c0		✓	✓			✓				✓			
c1		✓	✓				✓				✓		
c2				✓	✓			✓				✓	
c3				✓	✓				✓				✓
c4						✓		✓		✓		✓	
c5							✓		✓		✓		✓

Table 2–12. Global & Regional Clock Connections from Bottom Clock Pins & Enhanced PLL Outputs (Part 1 of 2)

Bottom Side Global & Regional Clock Network Connectivity	DLLCLK	CLK4	CLK5	CLK6	CLK7	RCLK8	RCLK9	RCLK10	RCLK11	RCLK12	RCLK13	RCLK14	RCLK15
Clock pins													
CLK4p	✓	✓	✓			✓				✓			
CLK5p	✓	✓	✓				✓				✓		
CLK6p	✓			✓	✓			✓				✓	
CLK7p	✓			✓	✓				✓				✓
CLK4n		✓				✓				✓			
CLK5n			✓				✓				✓		
CLK6n				✓				✓				✓	
CLK7n					✓				✓				✓
Drivers from internal logic													
GCLKDRV0		✓											
GCLKDRV1			✓										
GCLKDRV2				✓									

Figure 2–50. Control Signal Selection per IOE



Notes to Figure 2–50:

(1) Control signals `ce_in`, `ce_out`, `aclr/apreset`, `sclr/spreset`, and `oe` can be global signals even though their control selection multiplexers are not directly fed by the `ioe_clk[7..0]` signals. The `ioe_clk` signals can drive the I/O local interconnect, which then drives the control selection multiplexers.

In normal bidirectional operation, the input register can be used for input data requiring fast setup times. The input register can have its own clock input and clock enable separate from the OE and output registers. The output register can be used for data requiring fast clock-to-output performance. The OE register can be used for fast clock-to-output enable timing. The OE and output register share the same clock source and the same clock enable source from local interconnect in the associated LAB, dedicated I/O clocks, and the column and row interconnects.

3. Configuration & Testing

SII51003-4.2

IEEE Std. 1149.1 JTAG Boundary-Scan Support

All Stratix® II devices provide Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry that complies with the IEEE Std. 1149.1. JTAG boundary-scan testing can be performed either before or after, but not during configuration. Stratix II devices can also use the JTAG port for configuration with the Quartus® II software or hardware using either Jam Files (**.jam**) or Jam Byte-Code Files (**.jbc**).

Stratix II devices support IOE I/O standard setting reconfiguration through the JTAG BST chain. The JTAG chain can update the I/O standard for all input and output pins any time before or during user mode through the `CONFIG_IO` instruction. You can use this capability for JTAG testing before configuration when some of the Stratix II pins drive or receive from other devices on the board using voltage-referenced standards. Because the Stratix II device may not be configured before JTAG testing, the I/O pins may not be configured for appropriate electrical standards for chip-to-chip communication. Programming those I/O standards via JTAG allows you to fully test I/O connections to other devices.

A device operating in JTAG mode uses four required pins, `TDI`,`TDO`, `TMS`, and `TCK`, and one optional pin, `TRST`. The `TCK` pin has an internal weak pull-down resistor, while the `TDI`,`TMS` and `TRST` pins have weak internal pull-ups. The JTAG input pins are powered by the 3.3-V `VCCPD` pins. The TDO output pin is powered by the V_{CCIO} power supply of bank 4.

Stratix II devices also use the JTAG port to monitor the logic operation of the device with the SignalTap® II embedded logic analyzer. Stratix II devices support the JTAG instructions shown in Table 3–1.

Stratix II, Stratix, Cyclone® II, and Cyclone devices must be within the first 17 devices in a JTAG chain. All of these devices have the same JTAG controller. If any of the Stratix II, Stratix, Cyclone II, or Cyclone devices are in the 18th of further position, they fail configuration. This does not affect SignalTap II.

The Stratix II device instruction register length is 10 bits and the `USERCODE` register length is 32 bits. Tables 3–2 and 3–3 show the boundary-scan register length and device `IDCODE` information for Stratix II devices.

Table 5–25. 1.5-V HSTL Class I & II Differential Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	I/O supply voltage		1.425	1.500	1.575	V
V_{DIF} (DC)	DC input differential voltage		0.2			V
V_{CM} (DC)	DC common mode input voltage		0.68		0.90	V
V_{DIF} (AC)	AC differential input voltage		0.4			V
V_{OX} (AC)	AC differential cross point voltage		0.68		0.90	V

Table 5–26. 1.8-V HSTL Class I Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.71	1.80	1.89	V
V_{REF}	Input reference voltage		0.85	0.90	0.95	V
V_{TT}	Termination voltage		0.85	0.90	0.95	V
V_{IH} (DC)	DC high-level input voltage		V_{REF} + 0.1			V
V_{IL} (DC)	DC low-level input voltage		–0.3		V_{REF} – 0.1	V
V_{IH} (AC)	AC high-level input voltage		V_{REF} + 0.2			V
V_{IL} (AC)	AC low-level input voltage				V_{REF} – 0.2	V
V_{OH}	High-level output voltage	I_{OH} = 8 mA *(1)*	V_{CCIO} – 0.4			V
V_{OL}	Low-level output voltage	I_{OH} = –8 mA *(1)*			0.4	V

Note to Table 5–26:

(1) This specification is supported across all the programmable drive settings available for this I/O standard as shown in the *Stratix II Architecture* chapter in volume 1 of the *Stratix II Device Handbook*.

Table 5–31. Series & Differential On-Chip Termination Specification for Left & Right I/O Banks

Symbol	Description	Conditions	Resistance Tolerance		
			Commercial Max	Industrial Max	Unit
25-Ω R_S 3.3/2.5	Internal series termination without calibration (25-Ω setting)	V_{CCIO} = 3.3/2.5 V	±30	±30	%
50-Ω R_S 3.3/2.5/1.8	Internal series termination without calibration (50-Ω setting)	V_{CCIO} = 3.3/2.5/1.8 V	±30	±30	%
50-Ω R_S 1.5	Internal series termination without calibration (50-Ω setting)	V_{CCIO} = 1.5 V	±36	±36	%
R_D	Internal differential termination for LVDS or HyperTransport technology (100-Ω setting)	V_{CCIO} = 2.5 V	±20	±25	%

Pin Capacitance

Table 5–32 shows the Stratix II device family pin capacitance.

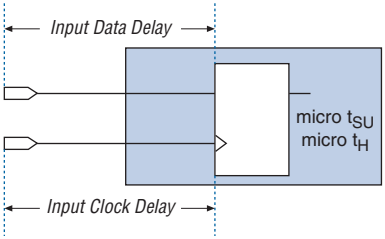
Table 5–32. Stratix II Device Capacitance *Note (1)*

Symbol	Parameter	Typical	Unit
C_{IOTB}	Input capacitance on I/O pins in I/O banks 3, 4, 7, and 8.	5.0	pF
C_{IOLR}	Input capacitance on I/O pins in I/O banks 1, 2, 5, and 6, including high-speed differential receiver and transmitter pins.	6.1	pF
C_{CLKTB}	Input capacitance on top/bottom clock input pins: `CLK[4..7]` and `CLK[12..15]`.	6.0	pF
C_{CLKLR}	Input capacitance on left/right clock inputs: `CLK0`, `CLK2`, `CLK8`, `CLK10`.	6.1	pF
C_{CLKLR+}	Input capacitance on left/right clock inputs: `CLK1`, `CLK3`, `CLK9`, and `CLK11`.	3.3	pF
C_{OUTFB}	Input capacitance on dual-purpose clock output/feedback pins in PLL banks 9, 10, 11, and 12.	6.7	pF

Note to Table 5–32:

(1) Capacitance is sample-tested only. Capacitance is measured using time-domain reflections (TDR). Measurement accuracy is within ±0.5pF

Figure 5–3. Input Register Setup & Hold Timing Diagram



For output timing, different I/O standards require different baseline loading techniques for reporting timing delays. Altera characterizes timing delays with the required termination for each I/O standard and with 0 pF (except for PCI and PCI-X which use 10 pF) loading and the timing is specified up to the output pin of the FPGA device. The Quartus II software calculates the I/O timing for each I/O standard with a default baseline loading as specified by the I/O standards.

The following measurements are made during device characterization. Altera measures clock-to-output delays (t_{CO}) at worst-case process, minimum voltage, and maximum temperature (PVT) for default loading conditions shown in Table 5–34. Use the following equations to calculate clock pin to output pin timing for Stratix II devices.

> t_{CO} from clock pin to I/O pin = delay from clock pad to I/O output register + IOE output register clock-to-output delay + delay from output register to output pin + I/O output delay
>
> t_{xz}/t_{zx} from clock pin to I/O pin = delay from clock pad to I/O output register + IOE output register clock-to-output delay + delay from output register to output pin + I/O output delay + output enable pin delay

Simulation using IBIS models is required to determine the delays on the PCB traces in addition to the output pin delay timing reported by the Quartus II software and the timing model in the device handbook.

1. Simulate the output driver of choice into the generalized test setup, using values from Table 5–34.
2. Record the time to V_{MEAS}.
3. Simulate the output driver of choice into the actual PCB trace and load, using the appropriate IBIS model or capacitance value to represent the load.

Table 5–39. DSP Block Internal Timing Microparameters (Part 2 of 2)

Symbol	Parameter	-3 Speed Grade (1)		-3 Speed Grade (2)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (3)	Max	Min (3)	Max	Min (4)	Max	Min (3)	Max	
t_{CLKL}	Minimum clock low time	1,190		1,249		1,368 1,368		1,594		ps
t_{CLKH}	Minimum clock high time	1,190		1,249		1,368 1,368		1,594		ps

Notes to Table 5–39:

(1) These numbers apply to -3 speed grade EP2S15, EP2S30, EP2S60, and EP2S90 devices.
(2) These numbers apply to -3 speed grade EP2S130 and EP2S180 devices.
(3) For the -3 and -5 speed grades, the minimum timing is for the commercial temperature grade. Only -4 speed grade devices offer the industrial temperature grade.
(4) For the -4 speed grade, the first number is the minimum timing parameter for industrial devices. The second number is the minimum timing parameter for commercial devices.

Table 5–40. M512 Block Internal Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	-3 Speed Grade (2)		-3 Speed Grade (3)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (4)	Max	Min (4)	Max	Min (5)	Max	Min (4)	Max	
t_{M512RC}	Synchronous read cycle time	2,089	2,318	2,089	2.433	1,989 2,089	2,664	2,089	3,104	ps
$t_{M512WERESU}$	Write or read enable setup time before clock	22		23		25 25		29		ps
$t_{M512WEREH}$	Write or read enable hold time after clock	203		213		233 233		272		ps
$t_{M512DATASU}$	Data setup time before clock	22		23		25 25		29		ps
$t_{M512DATAH}$	Data hold time after clock	203		213		233 233		272		ps
$t_{M512WADDRSU}$	Write address setup time before clock	22		23		25 25		29		ps
$t_{M512WADDRH}$	Write address hold time after clock	203		213		233 233		272		ps
$t_{M512RADDRSU}$	Read address setup time before clock	22		23		25 25		29		ps
$t_{M512RADDRH}$	Read address hold time after clock	203		213		233 233		272		ps

Table 5–40. M512 Block Internal Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	-3 Speed Grade (2)		-3 Speed Grade (3)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (4)	Max	Min (4)	Max	Min (5)	Max	Min (4)	Max	
$t_{M512DATACO1}$	Clock-to-output delay when using output registers	298	478	298	501	284 298	548	298	640	ps
$t_{M512DATACO2}$	Clock-to-output delay without output registers	2,102	2,345	2,102	2,461	2,003 2,102	2,695	2,102	3,141	ps
$t_{M512CLKL}$	Minimum clock low time	1,315		1,380		1,512 1,512		1,762		ps
$t_{M512CLKH}$	Minimum clock high time	1,315		1,380		1,512 1,512		1,762		ps
$t_{M512CLR}$	Minimum clear pulse width	144		151		165 165		192		ps

Notes to Table 5–40:

(1) F_{MAX} of M512 block obtained using the Quartus II software does not necessarily equal to 1/TM512RC.
(2) These numbers apply to -3 speed grade EP2S15, EP2S30, EP2S60, and EP2S90 devices.
(3) These numbers apply to -3 speed grade EP2S130 and EP2S180 devices.
(4) For the -3 and -5 speed grades, the minimum timing is for the commercial temperature grade. Only -4 speed grade devices offer the industrial temperature grade.
(5) For the -4 speed grade, the first number is the minimum timing parameter for industrial devices. The second number is the minimum timing parameter for commercial devices.

Table 5–41. M4K Block Internal Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	-3 Speed Grade (2)		-3 Speed Grade (3)		-4 Speed Grade		-5 Speed Grade		Unit
		Min (4)	Max	Min (4)	Max	Min (5)	Max	Min (4)	Max	
t_{M4KRC}	Synchronous read cycle time	1,462	2,240	1,462	2,351	1,393 1,462	2,575	1,462	3,000	ps
$t_{M4KWERESU}$	Write or read enable setup time before clock	22		23		25 25		29		ps
$t_{M4KWEREH}$	Write or read enable hold time after clock	203		213		233 233		272		ps
$t_{M4KBESU}$	Byte enable setup time before clock	22		23		25 25		29		ps
t_{M4KBEH}	Byte enable hold time after clock	203		213		233 233		272		ps

Table 5–71. Default Loading of Different I/O Standards for Stratix II (Part 2 of 2)

I/O Standard	Capacitive Load	Unit
SSTL-2 Class II	0	pF
SSTL-18 Class I	0	pF
SSTL-18 Class II	0	pF
1.5-V HSTL Class I	0	pF
1.5-V HSTL Class II	0	pF
1.8-V HSTL Class I	0	pF
1.8-V HSTL Class II	0	pF
1.2-V HSTL with OCT	0	pF
Differential SSTL-2 Class I	0	pF
Differential SSTL-2 Class II	0	pF
Differential SSTL-18 Class I	0	pF
Differential SSTL-18 Class II	0	pF
1.5-V Differential HSTL Class I	0	pF
1.5-V Differential HSTL Class II	0	pF
1.8-V Differential HSTL Class I	0	pF
1.8-V Differential HSTL Class II	0	pF
LVDS	0	pF
HyperTransport	0	pF
LVPECL	0	pF

Table 5–78. Maximum Output Toggle Rate on Stratix II Devices (Part 1 of 5) *Note (1)*

I/O Standard	Drive Strength	Column I/O Pins (MHz)			Row I/O Pins (MHz)			Clock Outputs (MHz)		
		-3	-4	-5	-3	-4	-5	-3	-4	-5
3.3-V LVTTL	4 mA	270	225	210	270	225	210	270	225	210
	8 mA	435	355	325	435	355	325	435	355	325
	12 mA	580	475	420	580	475	420	580	475	420
	16 mA	720	594	520	-	-	-	720	594	520
	20 mA	875	700	610	-	-	-	875	700	610
	24 mA	1,030	794	670	-	-	-	1,030	794	670
3.3-V LVCMOS	4 mA	290	250	230	290	250	230	290	250	230
	8 mA	565	480	440	565	480	440	565	480	440
	12 mA	790	710	670	-	-	-	790	710	670
	16 mA	1,020	925	875	-	-	-	1,020	925	875
	20 mA	1,066	985	935	-	-	-	1,066	985	935
	24 mA	1,100	1,040	1,000	-	-	-	1,100	1,040	1,000
2.5-V LVTTL/LVCMOS	4 mA	230	194	180	230	194	180	230	194	180
	8 mA	430	380	380	430	380	380	430	380	380
	12 mA	630	575	550	630	575	550	630	575	550
	16 mA	930	845	820	-	-	-	930	845	820
1.8-V LVTTL/LVCMOS	2 mA	120	109	104	120	109	104	120	109	104
	4 mA	285	250	230	285	250	230	285	250	230
	6 mA	450	390	360	450	390	360	450	390	360
	8 mA	660	570	520	660	570	520	660	570	520
	10 mA	905	805	755	-	-	-	905	805	755
	12 mA	1,131	1,040	990	-	-	-	1,131	1,040	990
1.5-V LVTTL/LVCMOS	2 mA	244	200	180	244	200	180	244	200	180
	4 mA	470	370	325	470	370	325	470	370	325
	6 mA	550	430	375	-	-	-	550	430	375
	8 mA	625	495	420	-	-	-	625	495	420
SSTL-2 Class I	8 mA	400	300	300	-	-	-	400	300	300
	12 mA	400	400	350	400	350	350	400	400	350
SSTL-2 Class II	16 mA	350	350	300	350	350	300	350	350	300
	20 mA	400	350	350	-	-	-	400	350	350
	24 mA	400	400	350	-	-	-	400	400	350

Table 5–79. Maximum Output Clock Toggle Rate Derating Factors (Part 5 of 5)

I/O Standard	Drive Strength	Maximum Output Clock Toggle Rate Derating Factors (ps/pF)								
		Column I/O Pins			Row I/O Pins			Dedicated Clock Outputs		
		-3	-4	-5	-3	-4	-5	-3	-4	-5
3.3-V LVTTL	OCT 50 Ω	133	152	152	133	152	152	147	152	152
2.5-V LVTTL	OCT 50 Ω	207	274	274	207	274	274	235	274	274
1.8-V LVTTL	OCT 50 Ω	151	165	165	151	165	165	153	165	165
3.3-V LVCMOS	OCT 50 Ω	300	316	316	300	316	316	263	316	316
1.5-V LVCMOS	OCT 50 Ω	157	171	171	157	171	171	174	171	171
SSTL-2 Class I	OCT 50 Ω	121	134	134	121	134	134	77	134	134
SSTL-2 Class II	OCT 25 Ω	56	101	101	56	101	101	58	101	101
SSTL-18 Class I	OCT 50 Ω	100	123	123	100	123	123	106	123	123
SSTL-18 Class II	OCT 25 Ω	61	110	110	-	-	-	59	110	110
1.2-V HSTL *(2)*	OCT 50 Ω	95	-	-	-	-	-	-	-	95

Notes to Table 5–79:

(1) For LVDS and HyperTransport technology output on row I/O pins, the toggle rate derating factors apply to loads larger than 5 pF. In the derating calculation, subtract 5 pF from the intended load value in pF for the correct result. For a load less than or equal to 5 pF, refer to Table 5–78 for output toggle rates.
(2) 1.2-V HSTL is only supported on column I/O pins in I/O banks 4,7, and 8.
(3) Differential HSTL and SSTL is only supported on column clock and DQS outputs.
(4) LVPECL is only supported on column clock outputs.

Duty Cycle Distortion

Duty cycle distortion (DCD) describes how much the falling edge of a clock is off from its ideal position. The ideal position is when both the clock high time (CLKH) and the clock low time (CLKL) equal half of the clock period (T), as shown in Figure 5–7. DCD is the deviation of the non-ideal falling edge from the ideal falling edge, such as D1 for the falling edge A and D2 for the falling edge B (Figure 5–7). The maximum DCD for a clock is the larger value of D1 and D2.

Table 5–80. Maximum DCD for Non-DDIO Output on Row I/O Pins (Part 2 of 2) *Note (1)*

Row I/O Output Standard	Maximum DCD for Non-DDIO Output		
	-3 Devices	-4 & -5 Devices	Unit
1.8 V	180	180	ps
1.5-V LVCMOS	165	195	ps
SSTL-2 Class I	115	145	ps
SSTL-2 Class II	95	125	ps
SSTL-18 Class I	55	85	ps
1.8-V HSTL Class I	80	100	ps
1.5-V HSTL Class I	85	115	ps
LVDS/ HyperTransport technology	55	80	ps

Note to Table 5–80:
(1) The DCD specification is based on a no logic array noise condition.

Here is an example for calculating the DCD as a percentage for a non-DDIO output on a row I/O on a -3 device:

If the non-DDIO output I/O standard is SSTL-2 Class II, the maximum DCD is 95 ps (see Table 5–80). If the clock frequency is 267 MHz, the clock period T is:

T = 1/ f = 1 / 267 MHz = 3.745 ns = 3745 ps

To calculate the DCD as a percentage:

(T/2 – DCD) / T = (3745ps/2 – 95ps) / 3745ps = 47.5% (for low boundary)

(T/2 + DCD) / T = (3745ps/2 + 95ps) / 3745ps = 52.5% (for high boundary)

Table 5–86. Maximum DCD for DDIO Output on Row I/O Pins with PLL in the Clock Path (Part 2 of 2) *Note (1)*

Row DDIO Output I/O Standard	Maximum DCD (PLL Output Clock Feeding DDIO Clock Port)		Unit
	-3 Device	-4 & -5 Device	
LVDS/ HyperTransport technology	180	180	ps

Note to Table 5–86:
(1) The DCD specification is based on a no logic array noise condition.

Table 5–87. Maximum DCD for DDIO Output on Column I/O with PLL in the Clock Path *Note (1)*

Column DDIO Output I/O Standard	Maximum DCD (PLL Output Clock Feeding DDIO Clock Port)		Unit
	-3 Device	-4 & -5 Device	
3.3-V LVTTL	145	160	ps
3.3-V LVCMOS	100	110	ps
2.5V	85	95	ps
1.8V	85	100	ps
1.5-V LVCMOS	140	155	ps
SSTL-2 Class I	65	75	ps
SSTL-2 Class II	60	70	ps
SSTL-18 Class I	50	65	ps
SSTL-18 Class II	70	80	ps
1.8-V HSTL Class I	60	70	ps
1.8-V HSTL Class II	60	70	ps
1.5-V HSTL Class I	55	70	ps
1.5-V HSTL Class II	85	100	ps
1.2-V HSTL	155	-	ps
LVPECL	180	180	ps

Notes to Table 5–87:
(1) The DCD specification is based on a no logic array noise condition.
(2) 1.2-V HSTL is only supported in -3 devices.

Table 5–89. High-Speed I/O Specifications for -3 Speed Grade (Part 2 of 2) *Notes (1), (2)*

Symbol	Conditions			-3 Speed Grade			Unit
				Min	Typ	Max	
f_{HSDR} (data rate)	J = 4 to 10 (LVDS, HyperTransport technology)			150		1,040	Mbps
	J = 2 (LVDS, HyperTransport technology)			*(4)*		760	Mbps
	J = 1 (LVDS only)			*(4)*		500	Mbps
$f_{HSDRDPA}$ (DPA data rate)	J = 4 to 10 (LVDS, HyperTransport technology)			150		1,040	Mbps
TCCS	All differential standards			-		200	ps
SW	All differential standards			330		-	ps
Output jitter						190	ps
Output t_{RISE}	All differential I/O standards					160	ps
Output t_{FALL}	All differential I/O standards					180	ps
t_{DUTY}				45	50	55	%
DPA run length						6,400	UI
DPA jitter tolerance	Data channel peak-to-peak jitter			0.44			UI
DPA lock time	**Standard**	**Training Pattern**	**Transition Density**				Number of repetitions
	SPI-4	0000000000 1111111111	10%	256			
	Parallel Rapid I/O	00001111	25%	256			
		10010000	50%	256			
	Miscellaneous	10101010	100%	256			
		01010101		256			

Notes to Table 5–89:

(1) When J = 4 to 10, the SERDES block is used.

(2) When J = 1 or 2, the SERDES block is bypassed.

(3) The input clock frequency and the W factor must satisfy the following fast PLL VCO specification: 150 ≤ input clock frequency × W ≤ 1,040.

(4) The minimum specification is dependent on the clock source (fast PLL, enhanced PLL, clock pin, and so on) and the clock routing resource (global, regional, or local) utilized. The I/O differential buffer and input register do not have a minimum toggle rate.