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Details

Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	33MHz
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	33
Program Memory Size	8KB (4K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	454 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c43t-33e-pt

PIC17C4X

NOTES:

FIGURE 3-2: PIC17CR42/42A/43/R43/44 BLOCK DIAGRAM

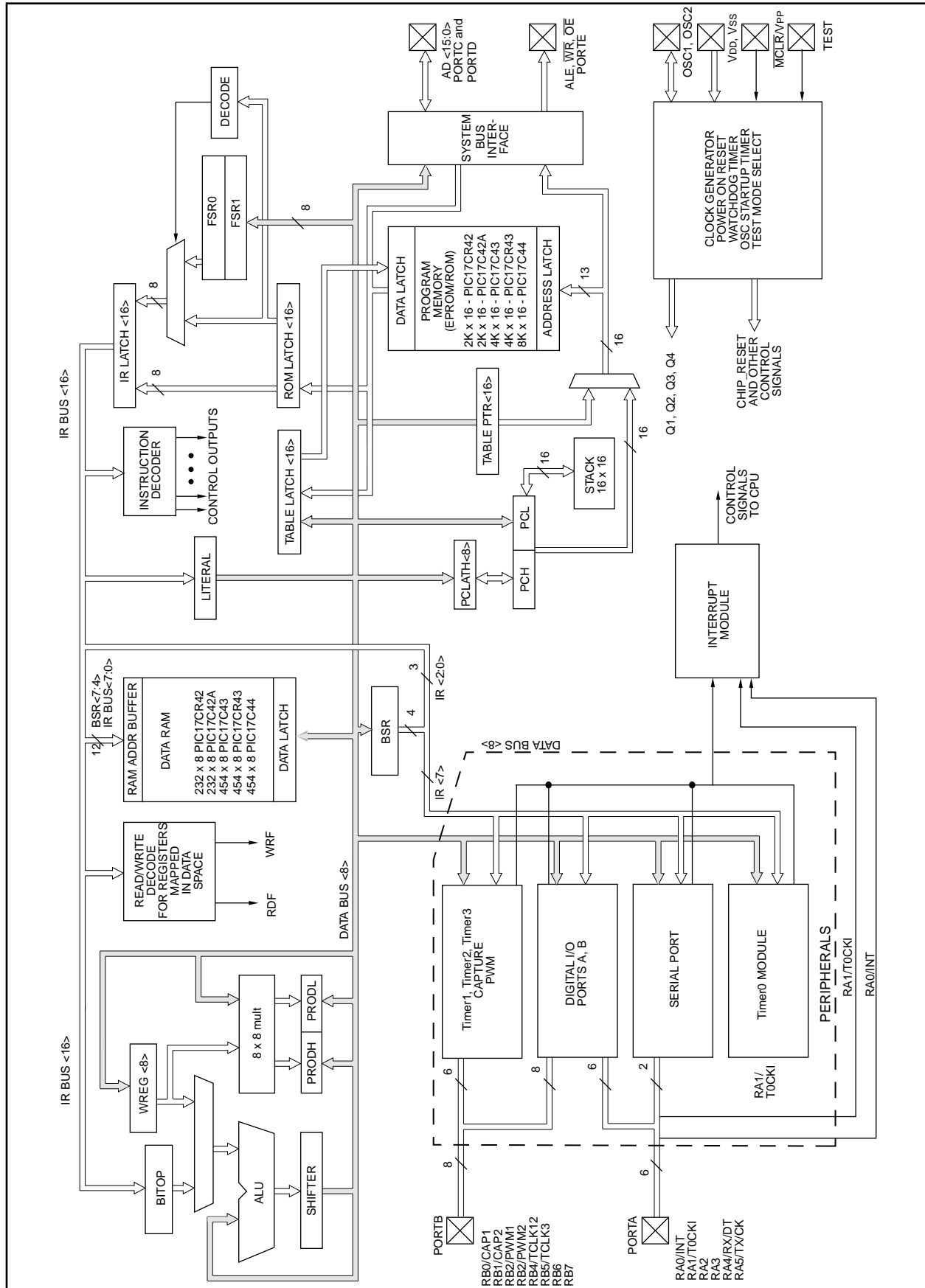


TABLE 4-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS

Register	Address	Power-on Reset	MCLR Reset WDT Reset	Wake-up from SLEEP through interrupt
Unbanked				
INDF0	00h	0000 0000	0000 0000	0000 0000
FSR0	01h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h	0000h	0000h	PC + 1 ⁽²⁾
PCLATH	03h	0000 0000	0000 0000	uuuu uuuu
ALUSTA	04h	1111 xxxx	1111 uuuu	1111 uuuu
T0STA	05h	0000 000-	0000 000-	0000 000-
CPUSTA ⁽³⁾	06h	--11 11--	--11 qq--	--uu qq--
INTSTA	07h	0000 0000	0000 0000	uuuu uuuu ⁽¹⁾
INDF1	08h	0000 0000	0000 0000	uuuu uuuu
FSR1	09h	xxxx xxxx	uuuu uuuu	uuuu uuuu
WREG	0Ah	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR0L	0Bh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR0H	0Ch	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBLPTRL ⁽⁴⁾	0Dh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBLPTRH ⁽⁴⁾	0Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBLPTRL ⁽⁵⁾	0Dh	0000 0000	0000 0000	uuuu uuuu
TBLPTRH ⁽⁵⁾	0Eh	0000 0000	0000 0000	uuuu uuuu
BSR	0Fh	0000 0000	0000 0000	uuuu uuuu
Bank 0				
PORTA	10h	0-xx xxxx	0-uu uuuu	uuuu uuuu
DDRB	11h	1111 1111	1111 1111	uuuu uuuu
PORTB	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
RCSTA	13h	0000 -00x	0000 -00u	uuuu -uuu
RCREG	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXSTA	15h	0000 --1x	0000 --1u	uuuu --uu
TXREG	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
SPBRG	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 1				
DDRC	10h	1111 1111	1111 1111	uuuu uuuu
PORTC	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
DDRD	12h	1111 1111	1111 1111	uuuu uuuu
PORTD	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
DDRE	14h	---- -111	---- -111	---- -uuu
PORTE	15h	---- -xxx	---- -uuu	---- -uuu
PIR	16h	0000 0010	0000 0010	uuuu uuuu ⁽¹⁾
PIE	17h	0000 0000	0000 0000	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented read as '0', q = value depends on condition.

Note 1: One or more bits in INTSTA, PIR will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 4-3 for reset value of specific condition.

4: Only applies to the PIC17C42.

5: Does not apply to the PIC17C42.

PIC17C4X

TABLE 4-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (Cont.'d)

Register	Address	Power-on Reset	MCLR Reset WDT Reset	Wake-up from SLEEP through interrupt
Bank 2				
TMR1	10h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR2	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3L	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3H	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR1	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR2	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1L	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1H	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 3				
PW1DCL	10h	xx-- ----	uu-- ----	uu-- ----
PW2DCL	11h	xx-- ----	uu-- ----	uu-- ----
PW1DCH	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PW2DCH	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2L	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2H	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TCON1	16h	0000 0000	0000 0000	uuuu uuuu
TCON2	17h	0000 0000	0000 0000	uuuu uuuu
Unbanked				
PRODL ⁽⁵⁾	18h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PRODH ⁽⁵⁾	19h	xxxx xxxx	uuuu uuuu	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented read as '0', q = value depends on condition.

Note 1: One or more bits in INTSTA, PIR will be affected (to cause wake-up).

- 2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.
- 3: See Table 4-3 for reset value of specific condition.
- 4: Only applies to the PIC17C42.
- 5: Does not apply to the PIC17C42.

PIC17C4X

6.2.2.3 TMR0 STATUS/CONTROL REGISTER (T0STA)

This register contains various control bits. Bit7 (INTEDG) is used to control the edge upon which a signal on the RA0/INT pin will set the RB0/INT interrupt flag. The other bits configure the Timer0 prescaler and clock source. (Figure 11-1).

FIGURE 6-9: T0STA REGISTER (ADDRESS: 05h, UNBANKED)

R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	U - 0
INTEDG	T0SE	T0CS	PS3	PS2	PS1	PS0	—
bit7							bit0

R = Readable bit
W = Writable bit
U = Unimplemented, reads as '0'
-n = Value at POR reset

bit 7: **INTEDG:** RA0/INT Pin Interrupt Edge Select bit
This bit selects the edge upon which the interrupt is detected.
1 = Rising edge of RA0/INT pin generates interrupt
0 = Falling edge of RA0/INT pin generates interrupt

bit 6: **T0SE:** Timer0 Clock Input Edge Select bit
This bit selects the edge upon which TMR0 will increment.
When T0CS = 0
1 = Rising edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
0 = Falling edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
When T0CS = 1
Don't care

bit 5: **T0CS:** Timer0 Clock Source Select bit
This bit selects the clock source for Timer0.
1 = Internal instruction clock cycle (TCY)
0 = T0CKI pin

bit 4-1: **PS3:PS0:** Timer0 Prescale Selection bits
These bits select the prescale value for Timer0.

PS3:PS0	Prescale Value
0000	1:1
0001	1:2
0010	1:4
0011	1:8
0100	1:16
0101	1:32
0110	1:64
0111	1:128
1xxx	1:256

bit 0: **Unimplemented:** Read as '0'

13.2 USART Asynchronous Mode

In this mode, the USART uses standard nonreturn-to-zero (NRZ) format (one start bit, eight or nine data bits, and one stop bit). The most common data format is 8-bits. An on-chip dedicated 8-bit baud rate generator can be used to derive standard baud rate frequencies from the oscillator. The USART's transmitter and receiver are functionally independent but use the same data format and baud rate. The baud rate generator produces a clock x64 of the bit shift rate. Parity is not supported by the hardware, but can be implemented in software (and stored as the ninth data bit). Asynchronous mode is stopped during SLEEP.

The asynchronous mode is selected by clearing the SYNC bit (TXSTA<4>).

The USART Asynchronous module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver

13.2.1 USART ASYNCHRONOUS TRANSMITTER

The USART transmitter block diagram is shown in Figure 13-3. The heart of the transmitter is the transmit shift register (TSR). The shift register obtains its data from the read/write transmit buffer (TXREG). TXREG is loaded with data in software. The TSR is not loaded until the stop bit has been transmitted from the previous load. As soon as the stop bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once TXREG transfers the data to the TSR (occurs in one Tcy at the end of the current BRG cycle), the TXREG is empty and an interrupt bit, TXIF (PIR<1>) is set. This interrupt can be enabled or disabled by the TXIE bit (PIE<1>). TXIF will be set regardless of TXIE and cannot be reset in software. It will reset only when new data is loaded into TXREG. While TXIF indicates the status of the TXREG, the TRMT (TXSTA<1>) bit shows the status of the TSR. TRMT is a read only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR is empty.

Note: The TSR is not mapped in data memory, so it is not available to the user.

Transmission is enabled by setting the TXEN (TXSTA<5>) bit. The actual transmission will not occur until TXREG has been loaded with data and the baud rate generator (BRG) has produced a shift clock (Figure 13-5). The transmission can also be started by first loading TXREG and then setting TXEN. Normally when transmission is first started, the TSR is empty, so a transfer to TXREG will result in an immediate transfer to TSR resulting in an empty TXREG. A back-to-back transfer is thus possible (Figure 13-6). Clearing TXEN during a transmission will cause the transmission to be aborted. This will reset the transmitter and the RA5/TX/CK pin will revert to hi-impedance.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG. This is because a data write to TXREG can result in an immediate transfer of the data to the TSR (if the TSR is empty).

Steps to follow when setting up an Asynchronous Transmission:

1. Initialize the SPBRG register for the appropriate baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. Load data to the TXREG register.
6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
7. Enable the transmission by setting TXEN (starts transmission).

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN) allows transmission to start sooner then doing these two events in the opposite order.

Note: To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

TABLE 13-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
16h, Bank 1	PIR	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TXIF	RCIF	0000 0010	0000 0010
13h, Bank 0	RCSTA	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
16h, Bank 0	TXREG	TX7	TX6	TX5	TX4	TX3	TX2	TX1	TX0	xxxx xxxx	uuuu uuuu
17h, Bank 1	PIE	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TXIE	RCIE	0000 0000	0000 0000
15h, Bank 0	TXSTA	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG	Baud rate generator register								xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', shaded cells are not used for synchronous master transmission.

Note 1: Other (non power-up) resets include: external reset through $\overline{\text{MCLR}}$ and Watchdog Timer Reset.

FIGURE 13-9: SYNCHRONOUS TRANSMISSION

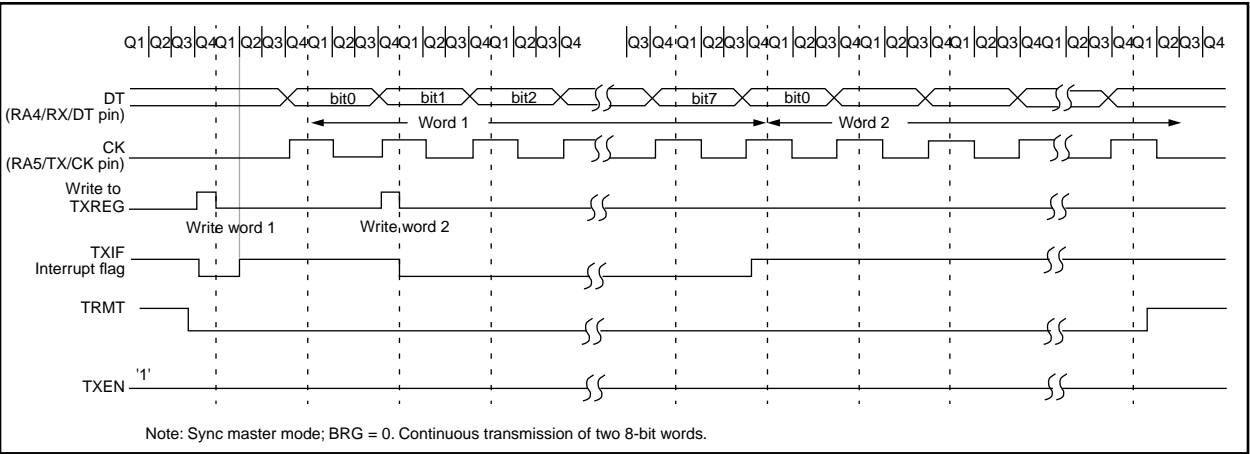
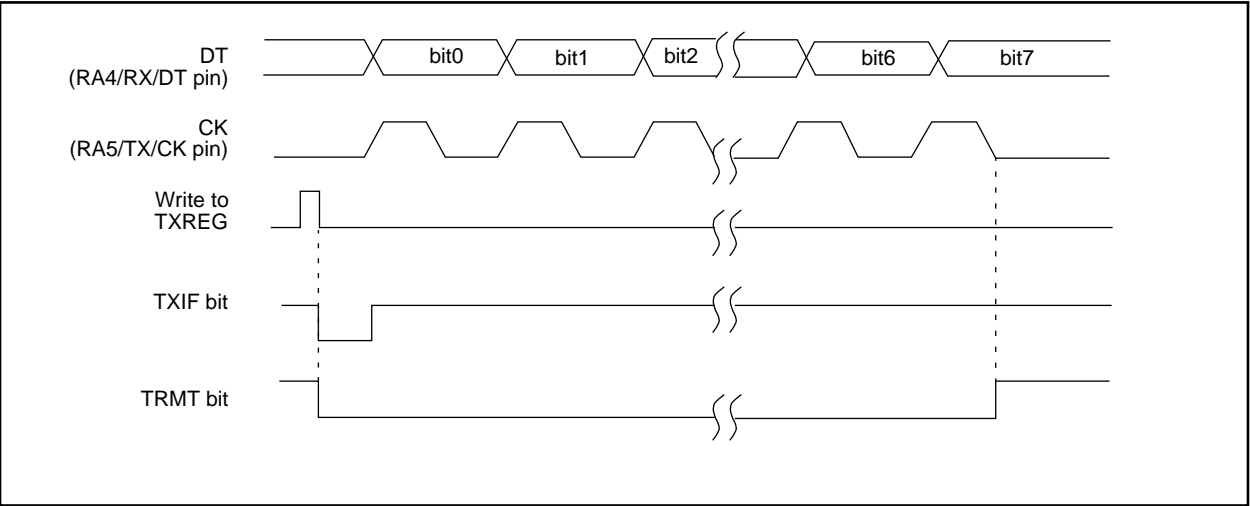


FIGURE 13-10: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



13.3.2 USART SYNCHRONOUS MASTER RECEPTION

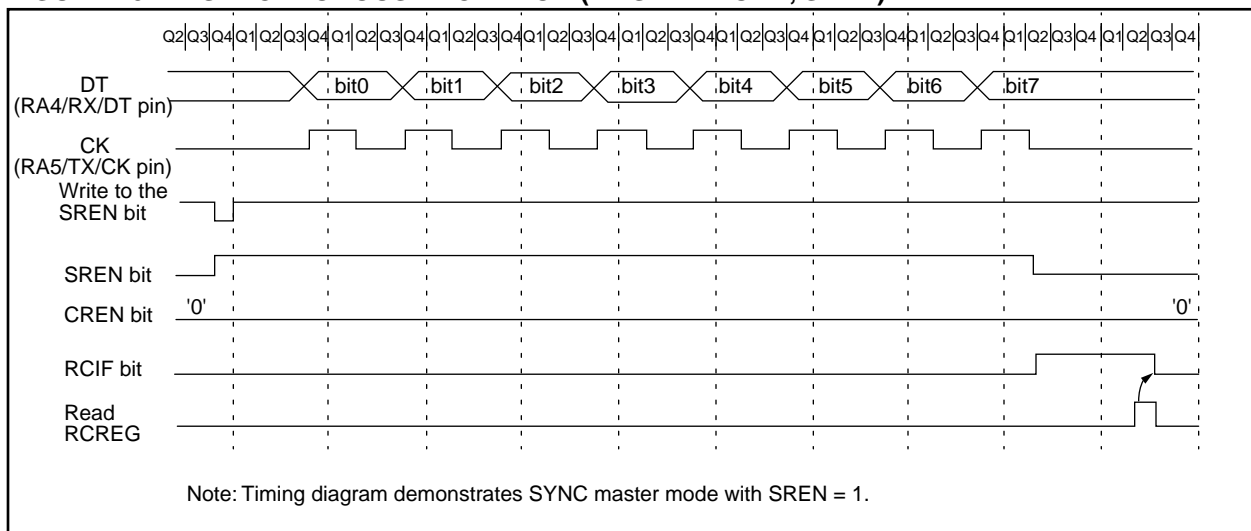
Once synchronous mode is selected, reception is enabled by setting either the SREN (RCSTA<5>) bit or the CREN (RCSTA<4>) bit. Data is sampled on the RA4/RX/DT pin on the falling edge of the clock. If SREN is set, then only a single word is received. If CREN is set, the reception is continuous until CREN is reset. If both bits are set, then CREN takes precedence. After clocking the last bit, the received data in the Receive Shift Register (RSR) is transferred to RCREG (if it is empty). If the transfer is complete, the interrupt bit RCIF (PIR<0>) is set. The actual interrupt can be enabled/disabled by setting/clearing the RCIE (PIE<0>) bit. RCIF is a read only bit which is RESET by the hardware. In this case it is reset when RCREG has been read and is empty. RCREG is a double buffered register; i.e., it is a two deep FIFO. It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte to begin shifting into the RSR. On the clocking of the last bit of the third byte, if RCREG is still full, then the overrun error bit OERR (RCSTA<1>) is set. The word in the RSR will be lost. RCREG can be read twice to retrieve the two bytes in the FIFO. The OERR bit has to be cleared in software. This is done by clearing the CREN bit. If OERR bit is set, transfers from RSR to RCREG are inhibited, so it is essential to clear OERR bit if it is set. The 9th receive bit is buffered the same way as the receive data. Reading the RCREG register will allow the RX9D and FERR bits to be loaded with values for the next received data; therefore, it is essential for the user to read the RCSTA register before reading RCREG in order not to lose the old FERR and RX9D information.

Steps to follow when setting up a Synchronous Master Reception:

1. Initialize the SPBRG register for the appropriate baud rate. See Section 13.1 for details.
2. Enable the synchronous master serial port by setting bits SYNC, SPEN, and CSRC.
3. If interrupts are desired, then set the RCIE bit.
4. If 9-bit reception is desired, then set the RX9 bit.
5. If a single reception is required, set bit SREN. For continuous reception set bit CREN.
6. The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
7. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading RCREG.
9. If any error occurred, clear the error by clearing CREN.

Note: To terminate a reception, either clear the SREN and CREN bits, or the SPEN bit. This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

FIGURE 13-11: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



MOVPF Move p to f

Syntax: `[label] MOVPF p,f`

Operands: $0 \leq f \leq 255$
 $0 \leq p \leq 31$

Operation: $(p) \rightarrow (f)$

Status Affected: Z

Encoding:

010p	pppp	ffff	ffff
------	------	------	------

Description: Move data from data memory location 'p' to data memory location 'f'. Location 'f' can be anywhere in the 256 byte data space (00h to FFh) while 'p' can be 00h to 1Fh.

Either 'p' or 'f' can be WREG (a useful special situation).

MOVPF is particularly useful for transferring a peripheral register (e.g. the timer or an I/O port) to a data memory location. Both 'f' and 'p' can be indirectly addressed.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'p'	Execute	Write register 'f'

Example: `MOVPF REG1, REG2`

Before Instruction

REG1 = 0x11
 REG2 = 0x33

After Instruction

REG1 = 0x11
 REG2 = 0x11

MOVWF Move WREG to f

Syntax: `[label] MOVWF f`

Operands: $0 \leq f \leq 255$

Operation: $(WREG) \rightarrow (f)$

Status Affected: None

Encoding:

0000	0001	ffff	ffff
------	------	------	------

Description: Move data from WREG to register 'f'. Location 'f' can be anywhere in the 256 word data space.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	Write register 'f'

Example: `MOVWF REG`

Before Instruction

WREG = 0x4F
 REG = 0xFF

After Instruction

WREG = 0x4F
 REG = 0x4F

RLNCF Rotate Left f (no carry)

Syntax: [label] RLNCF f,d

Operands: $0 \leq f \leq 255$
 $d \in [0,1]$

Operation: $f \langle n \rangle \rightarrow d \langle n+1 \rangle$;
 $f \langle 7 \rangle \rightarrow d \langle 0 \rangle$

Status Affected: None

Encoding:

0010	001d	ffff	ffff
------	------	------	------

Description: The contents of register 'f' are rotated one bit to the left. If 'd' is 0 the result is placed in WREG. If 'd' is 1 the result is stored back in register 'f'.



Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	Write to destination

Example: RLNCF REG, 1

Before Instruction

C = 0
 REG = 1110 1011

After Instruction

C =
 REG = 1101 0111

RRCF Rotate Right f through Carry

Syntax: [label] RRCF f,d

Operands: $0 \leq f \leq 255$
 $d \in [0,1]$

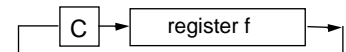
Operation: $f \langle n \rangle \rightarrow d \langle n-1 \rangle$;
 $f \langle 0 \rangle \rightarrow C$;
 $C \rightarrow d \langle 7 \rangle$

Status Affected: C

Encoding:

0001	100d	ffff	ffff
------	------	------	------

Description: The contents of register 'f' are rotated one bit to the right through the Carry Flag. If 'd' is 0 the result is placed in WREG. If 'd' is 1 the result is placed back in register 'f'.



Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	Write to destination

Example: RRCF REG1, 0

Before Instruction

REG1 = 1110 0110
 C = 0

After Instruction

REG1 = 1110 0110
 WREG = 0111 0011
 C = 0

MPASM allow full symbolic debugging from the Microchip Universal Emulator System (PICMASTER).

MPASM has the following features to assist in developing software for specific use applications.

- Provides translation of Assembler source code to object code for all Microchip microcontrollers.
- Macro assembly capability.
- Produces all the files (Object, Listing, Symbol, and special) required for symbolic debug with Microchip's emulator systems.
- Supports Hex (default), Decimal and Octal source and listing formats.

MPASM provides a rich directive language to support programming of the PIC16/17. Directives are helpful in making the development of your assemble source code shorter and more maintainable.

16.11 Software Simulator (MPLAB-SIM)

The MPLAB-SIM Software Simulator allows code development in a PC host environment. It allows the user to simulate the PIC16/17 series microcontrollers on an instruction level. On any given instruction, the user may examine or modify any of the data areas or provide external stimulus to any of the pins. The input/output radix can be set by the user and the execution can be performed in; single step, execute until break, or in a trace mode.

MPLAB-SIM fully supports symbolic debugging using MPLAB-C and MPASM. The Software Simulator offers the low cost flexibility to develop and debug code outside of the laboratory environment making it an excellent multi-project software development tool.

16.12 C Compiler (MPLAB-C)

The MPLAB-C Code Development System is a complete 'C' compiler and integrated development environment for Microchip's PIC16/17 family of microcontrollers. The compiler provides powerful integration capabilities and ease of use not found with other compilers.

For easier source level debugging, the compiler provides symbol information that is compatible with the MPLAB IDE memory display (PICMASTER emulator software versions 1.13 and later).

16.13 Fuzzy Logic Development System (fuzzyTECH-MP)

fuzzyTECH-MP fuzzy logic development tool is available in two versions - a low cost introductory version, MP Explorer, for designers to gain a comprehensive working knowledge of fuzzy logic system design; and a full-featured version, *fuzzyTECH-MP*, edition for implementing more complex systems.

Both versions include Microchip's *fuzzyLAB™* demonstration board for hands-on experience with fuzzy logic systems implementation.

16.14 MP-DriveWay™ – Application Code Generator

MP-DriveWay is an easy-to-use Windows-based Application Code Generator. With MP-DriveWay you can visually configure all the peripherals in a PIC16/17 device and, with a click of the mouse, generate all the initialization and many functional code modules in C language. The output is fully compatible with Microchip's MPLAB-C C compiler. The code produced is highly modular and allows easy integration of your own code. MP-DriveWay is intelligent enough to maintain your code through subsequent code generation.

16.15 SEEVAL® Evaluation and Programming System

The SEEVAL SEEPROM Designer's Kit supports all Microchip 2-wire and 3-wire Serial EEPROMs. The kit includes everything necessary to read, write, erase or program special features of any Microchip SEEPROM product including Smart Serials™ and secure serials. The Total Endurance™ Disk is included to aid in trade-off analysis and reliability calculations. The total kit can significantly reduce time-to-market and result in an optimized system.

16.16 TrueGauge® Intelligent Battery Management

The TrueGauge development tool supports system development with the MTA11200B TrueGauge Intelligent Battery Management IC. System design verification can be accomplished before hardware prototypes are built. User interface is graphically-oriented and measured data can be saved in a file for exporting to Microsoft Excel.

16.17 KEELOQ® Evaluation and Programming Tools

KEELOQ evaluation and programming tools support Microchips HCS Secure Data Products. The HCS evaluation kit includes an LCD display to show changing codes, a decoder to decode transmissions, and a programming interface to program test transmitters.

17.1 DC CHARACTERISTICS: PIC17C42-16 (Commercial, Industrial) PIC17C42-25 (Commercial, Industrial)

Standard Operating Conditions (unless otherwise stated)							
DC CHARACTERISTICS							
Operating temperature							
-40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial							
Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001	VDD	Supply Voltage	4.5	–	5.5	V	
D002	VDR	RAM Data Retention Voltage (Note 1)	1.5 *	–	–	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure internal Power-on Reset signal	–	VSS	–	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure internal Power-on Reset signal	0.060*	–	–	mV/ms	See section on Power-on Reset for details
D010 D011 D012 D013 D014	IDD	Supply Current (Note 2)	–	3 6 11 19 95	6 12 * 24 * 38 150	mA mA mA mA μA	FOSC = 4 MHz (Note 4) FOSC = 8 MHz FOSC = 16 MHz FOSC = 25 MHz FOSC = 32 kHz WDT enabled (EC osc configuration)
D020 D021	IPD	Power-down Current (Note 3)	–	10 < 1	40 5	μA μA	VDD = 5.5V, WDT enabled VDD = 5.5V, WDT disabled

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to VDD or VSS, T0CKI = VDD, MCLR = VDD; WDT enabled/disabled as specified.

Current consumed from the oscillator and I/O's driving external capacitive or resistive loads need to be considered.

For the RC oscillator, the current through the external pull-up resistor (R) can be estimated as: $V_{DD} / (2 \cdot R)$.

For capacitive loads, The current can be estimated (for an individual I/O pin) as $(C_L \cdot V_{DD}) \cdot f$

C_L = Total capacitive load on the I/O pin; f = average frequency on the I/O pin switches.

The capacitive currents are most significant when the device is configured for external execution (includes extended microcontroller mode).

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula $I_R = V_{DD}/2R_{ext}$ (mA) with Rext in kOhm.

Applicable Devices	42	R42	42A	43	R43	44
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FIGURE 17-11: MEMORY INTERFACE WRITE TIMING

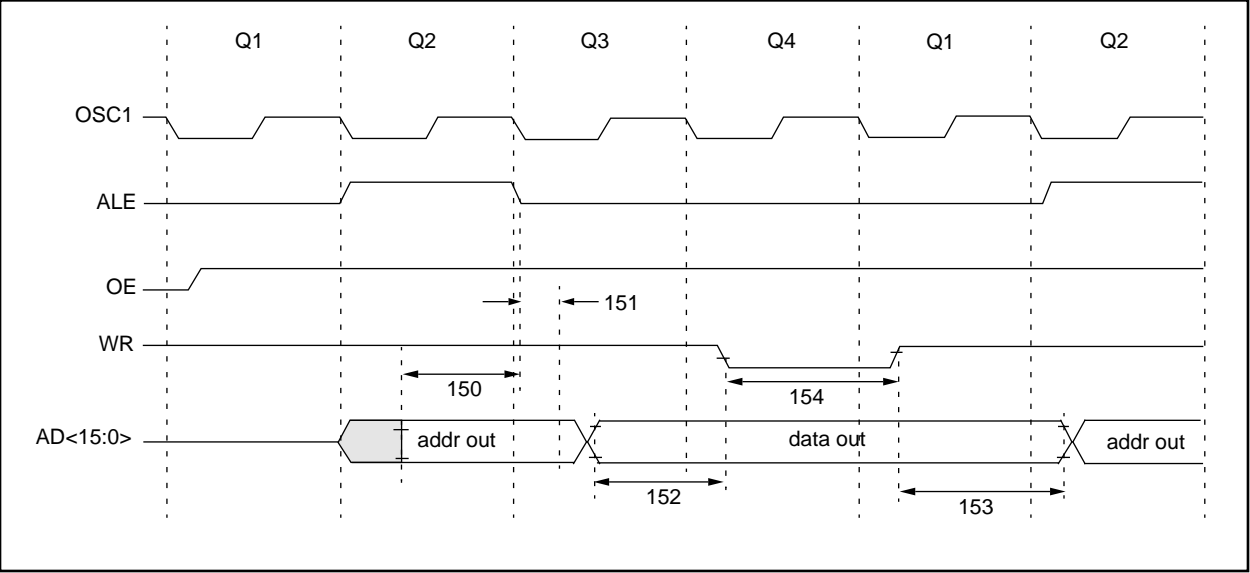


TABLE 17-11: MEMORY INTERFACE WRITE REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
150	TadV2aLL	AD<15:0> (address) valid to ALE↓ (address setup time)	0.25Tcy - 30	—	—	ns	
151	TaLL2adl	ALE↓ to address out invalid (address hold time)	0	—	—	ns	
152	TadV2wrL	Data out valid to WR↓ (data setup time)	0.25Tcy - 40	—	—	ns	
153	TwrH2adl	WR↑ to data out invalid (data hold time)	—	0.25Tcy §	—	ns	
154	TwrL	WR pulse width	—	0.25Tcy §	—	ns	

* These parameters are characterized but not tested.
† Data in “Typ” column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
§ This specification is guaranteed by design.

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FIGURE 18-9: TYPICAL I_{PD} vs. V_{DD} WATCHDOG DISABLED 25°C

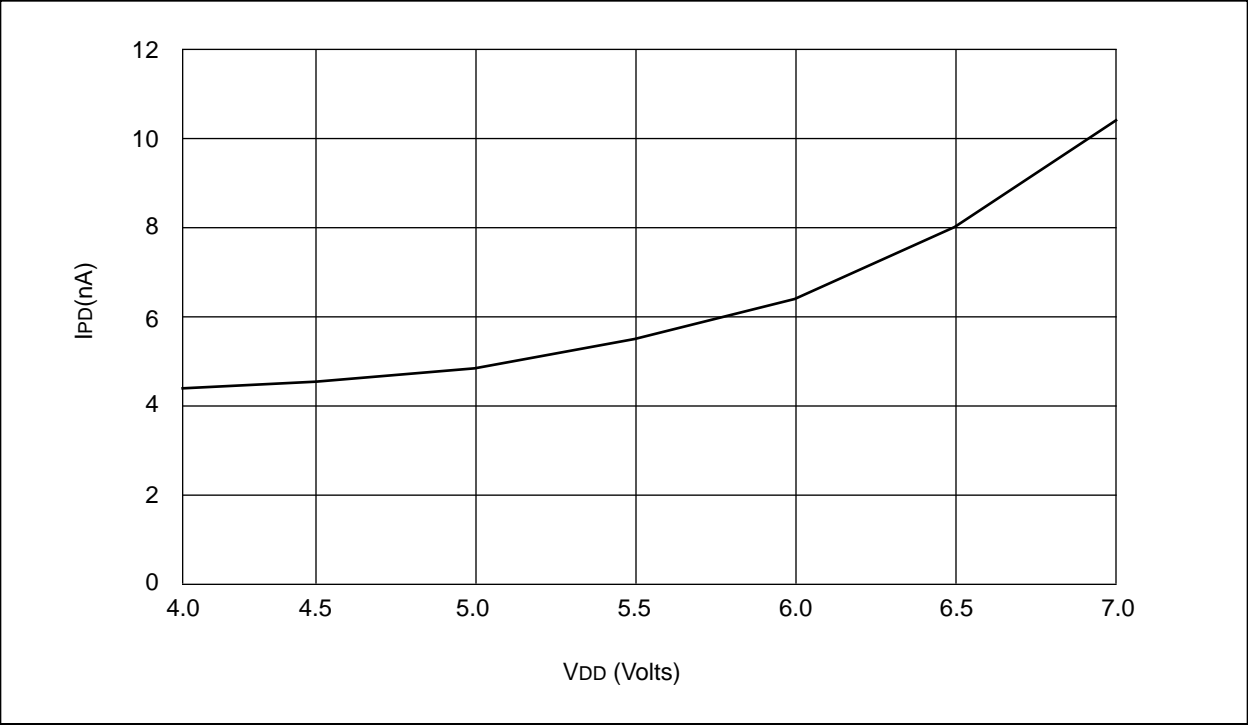
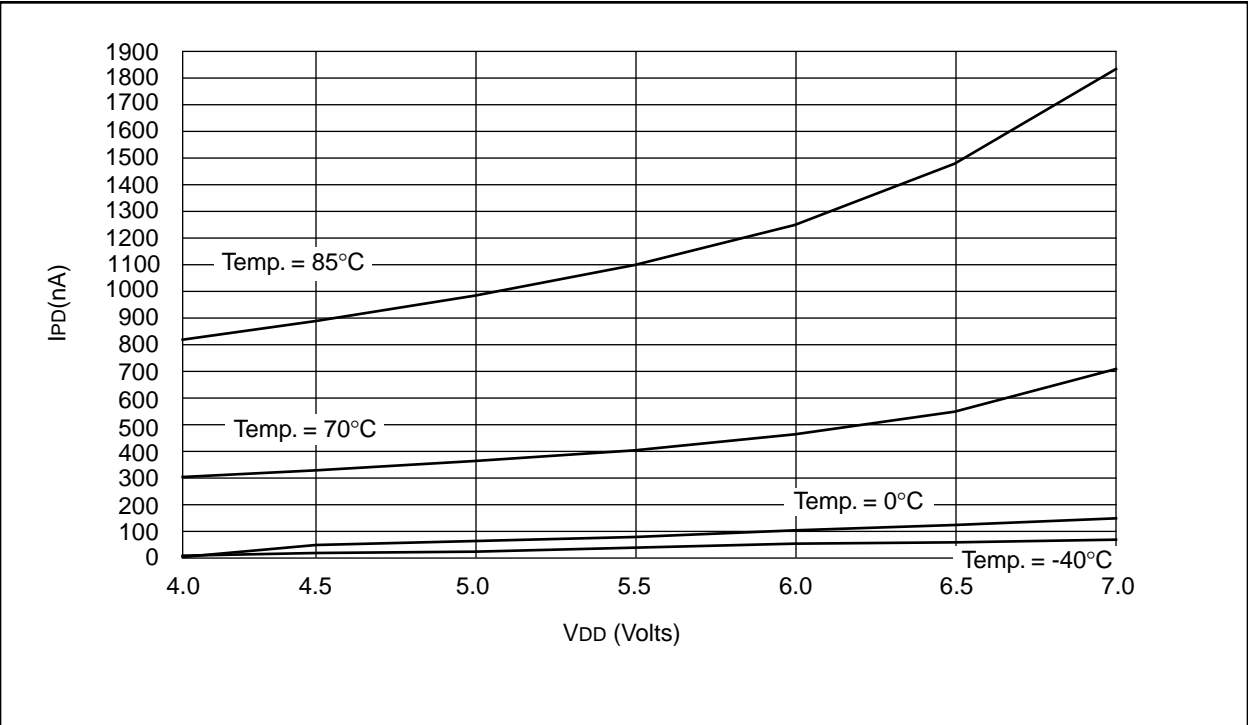


FIGURE 18-10: MAXIMUM I_{PD} vs. V_{DD} WATCHDOG DISABLED



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Applicable Devices 42 R42 42A 43 R43 44

FIGURE 18-17: I_{OH} vs. V_{OL} , $V_{DD} = 5V$

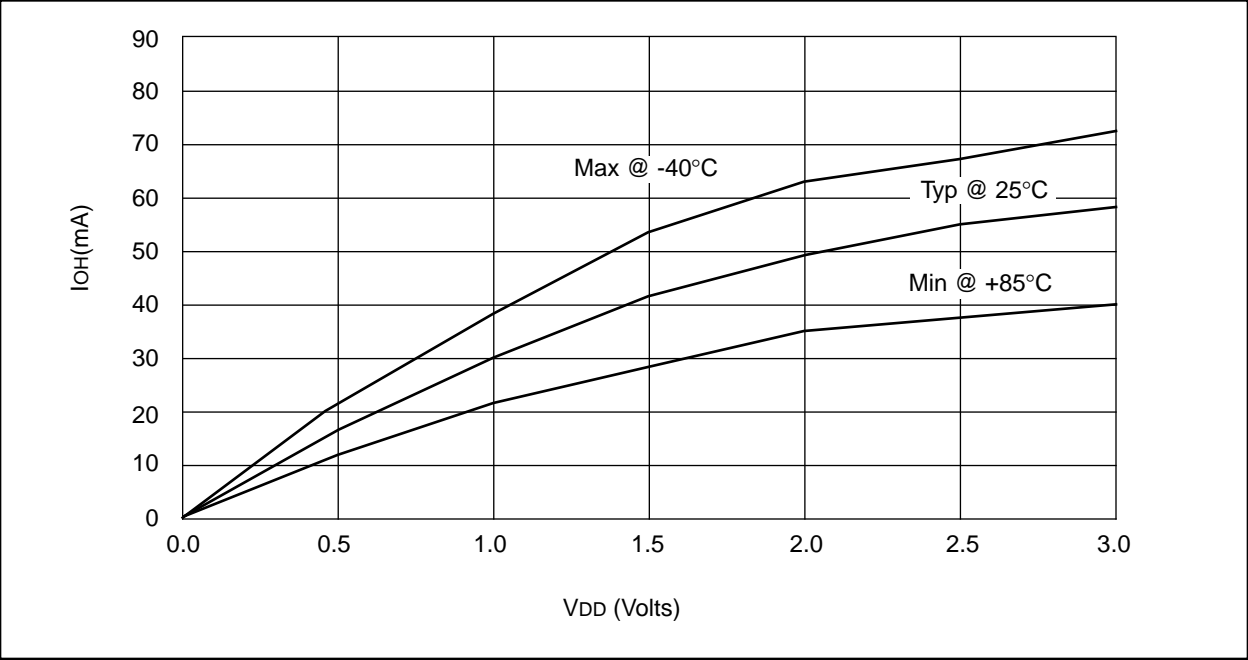
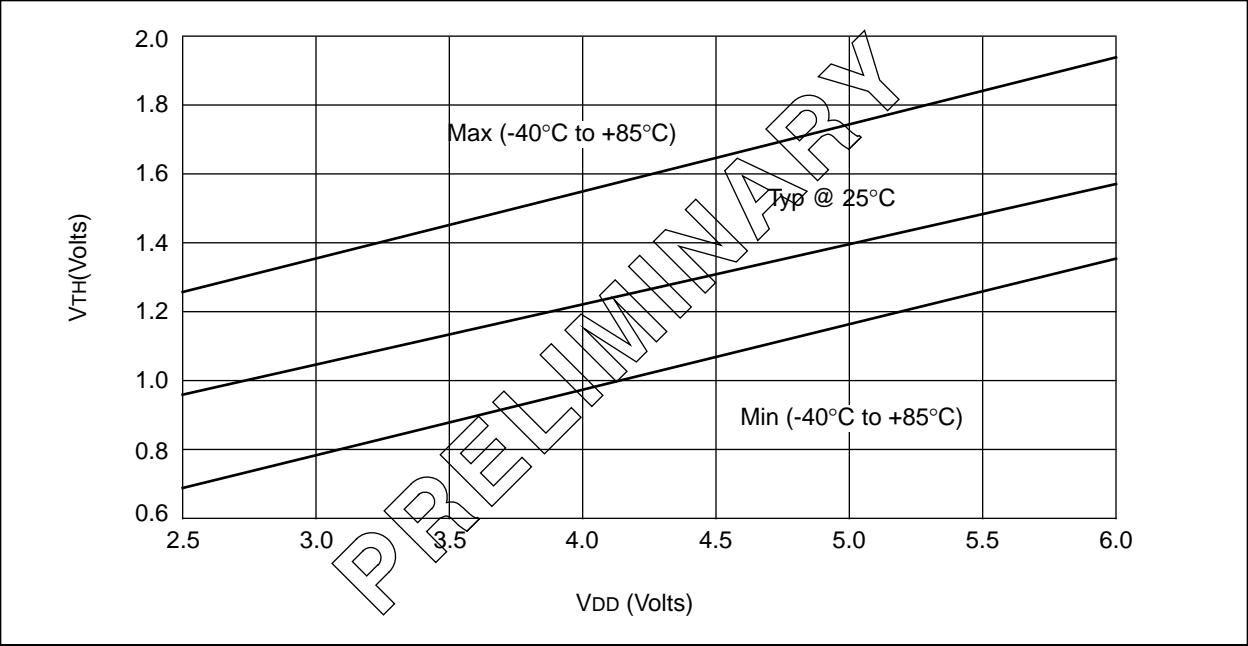


FIGURE 18-18: V_{TH} (INPUT THRESHOLD VOLTAGE) OF I/O PINS (TTL) vs. V_{DD}



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FIGURE 20-2: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

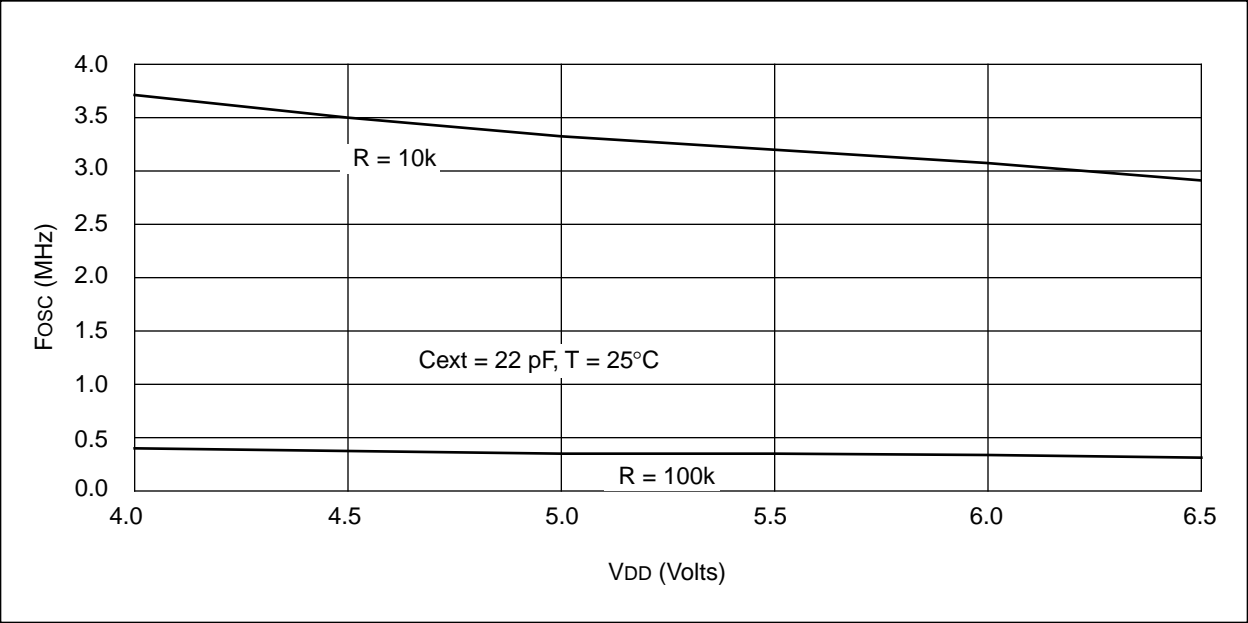


FIGURE 20-3: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

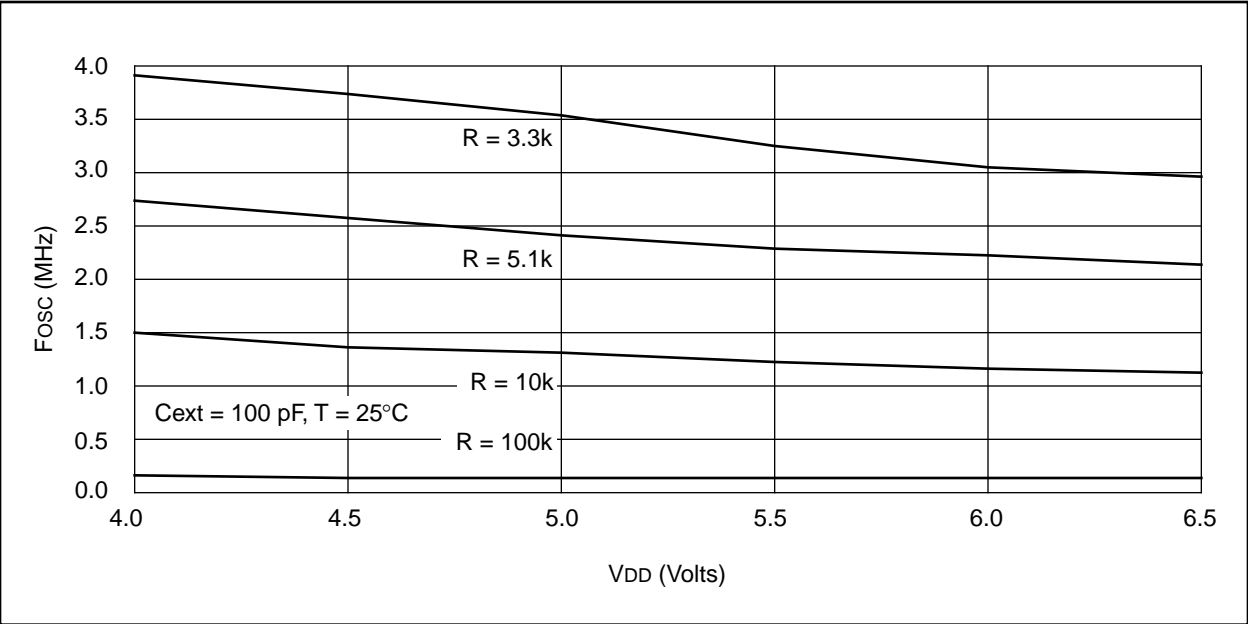


FIGURE 20-11: TYPICAL I_{PD} vs. V_{DD} WATCHDOG ENABLED 25°C

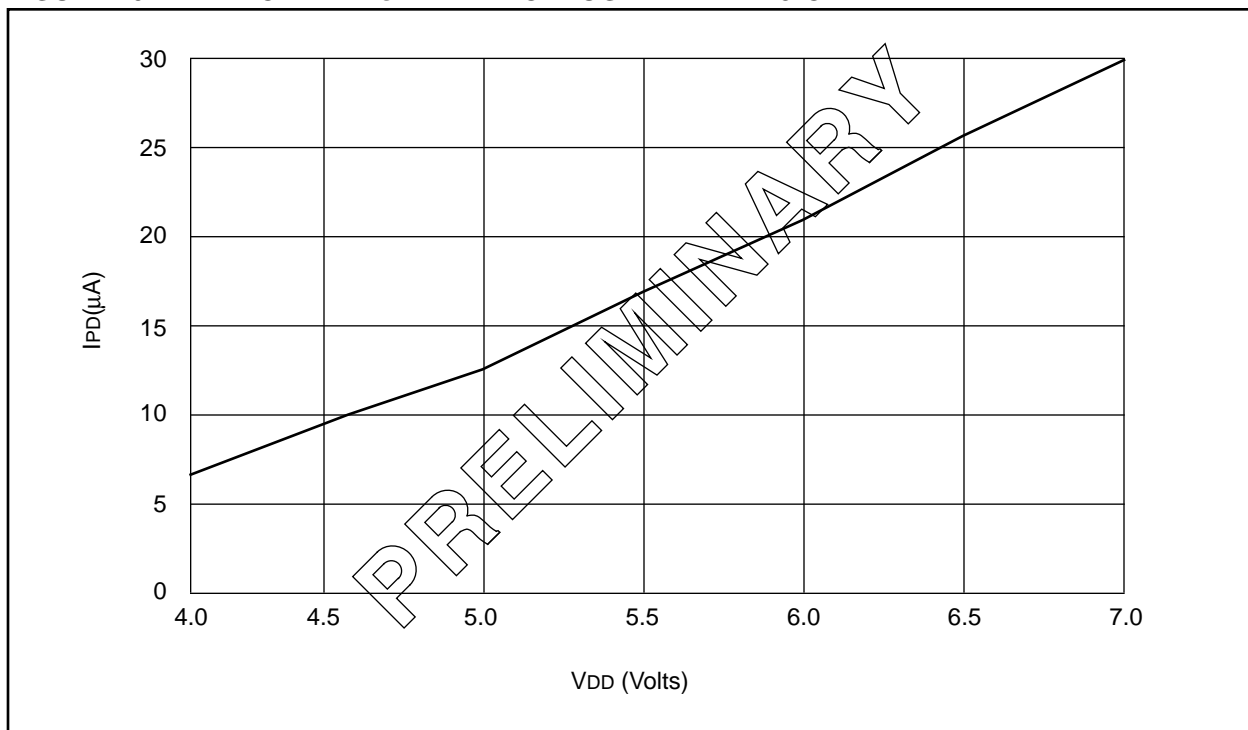
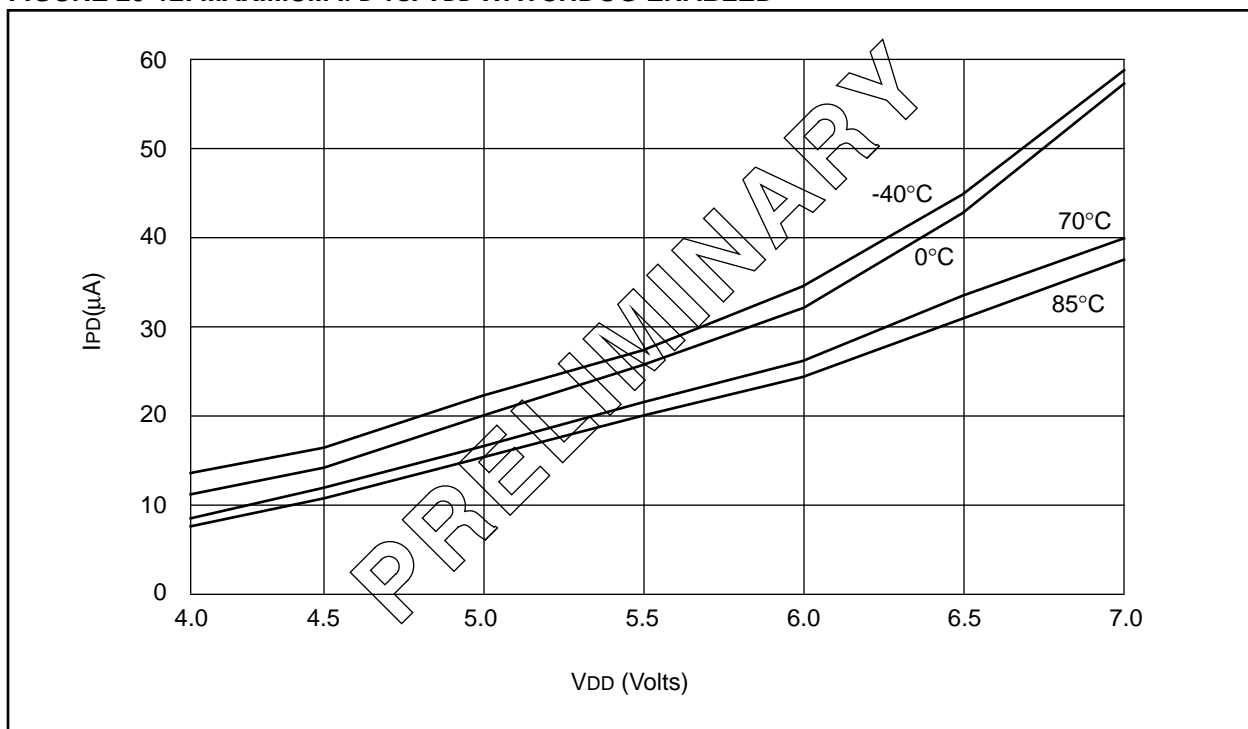
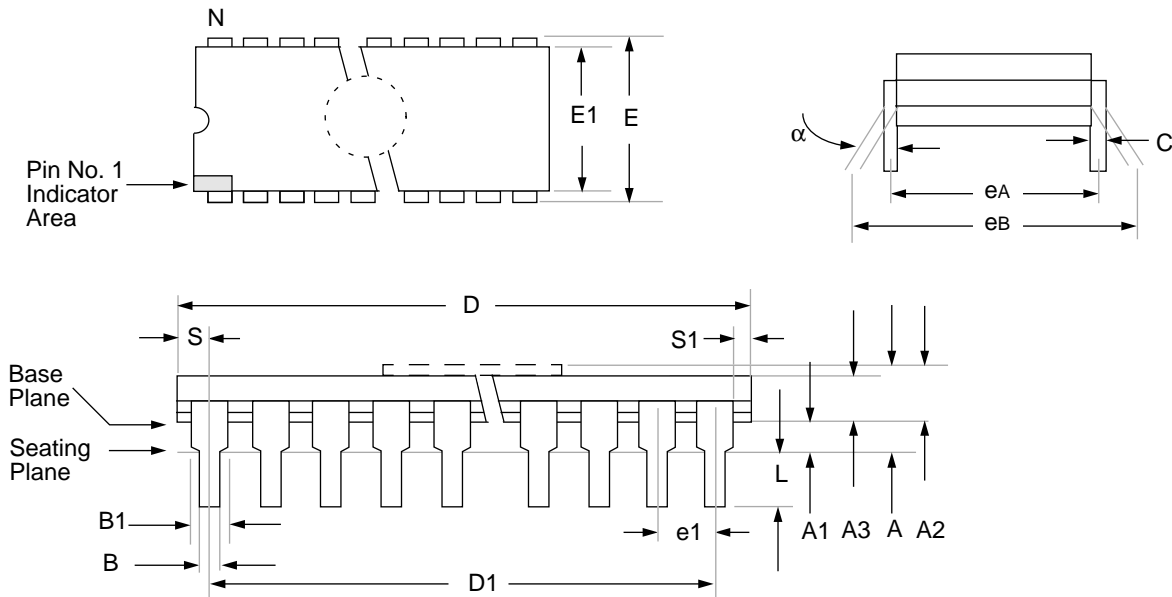


FIGURE 20-12: MAXIMUM I_{PD} vs. V_{DD} WATCHDOG ENABLED



21.0 PACKAGING INFORMATION

21.1 40-Lead Ceramic Cerdip Dual In-line, and Cerdip Dual In-line with Window (600 mil)



Package Group: Ceramic Cerdip Dual In-Line (CDP)						
Symbol	Millimeters			Inches		
	Min	Max	Notes	Min	Max	Notes
α	0°	10°		0°	10°	
A	4.318	5.715		0.170	0.225	
A1	0.381	1.778		0.015	0.070	
A2	3.810	4.699		0.150	0.185	
A3	3.810	4.445		0.150	0.175	
B	0.355	0.585		0.014	0.023	
B1	1.270	1.651	Typical	0.050	0.065	Typical
C	0.203	0.381	Typical	0.008	0.015	Typical
D	51.435	52.705		2.025	2.075	
D1	48.260	48.260	Reference	1.900	1.900	Reference
E	15.240	15.875		0.600	0.625	
E1	12.954	15.240		0.510	0.600	
e1	2.540	2.540	Reference	0.100	0.100	Reference
eA	14.986	16.002	Typical	0.590	0.630	Typical
eB	15.240	18.034		0.600	0.710	
L	3.175	3.810		0.125	0.150	
N	40	40		40	40	
S	1.016	2.286		0.040	0.090	
S1	0.381	1.778		0.015	0.070	

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NOTES:

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