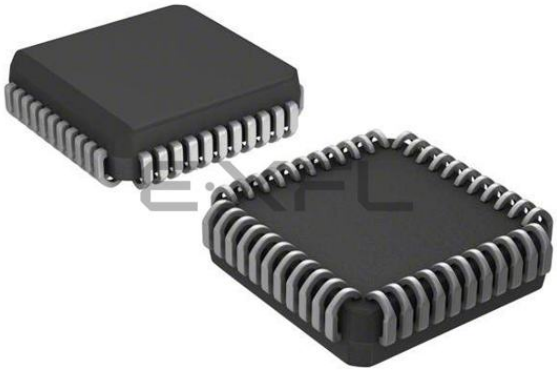

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	16MHz
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	33
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	454 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LCC (J-Lead)
Supplier Device Package	44-PLCC (16.59x16.59)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c44-16-l

5.4 Interrupt Operation

Global Interrupt Disable bit, GLINTD (CPUSTA<4>), enables all unmasked interrupts (if clear) or disables all interrupts (if set). Individual interrupts can be disabled through their corresponding enable bits in the INTSTA register. Peripheral interrupts need either the global peripheral enable PEIE bit disabled, or the specific peripheral enable bit disabled. Disabling the peripherals via the global peripheral enable bit, disables all peripheral interrupts. GLINTD is set on reset (interrupts disabled).

The RETFIE instruction allows returning from interrupt and re-enable interrupts at the same time.

When an interrupt is responded to, the GLINTD bit is automatically set to disable any further interrupt, the return address is pushed onto the stack and the PC is loaded with interrupt vector. There are four interrupt vectors to reduce interrupt latency.

The peripheral interrupt vector has multiple interrupt sources. Once in the peripheral interrupt service routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The peripheral interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid continuous interrupts.

The PIC17C4X devices have four interrupt vectors. These vectors and their hardware priority are shown in Table 5-1. If two enabled interrupts occur “at the same time”, the interrupt of the highest priority will be serviced first. This means that the vector address of that interrupt will be loaded into the program counter (PC).

TABLE 5-1: INTERRUPT VECTORS/ PRIORITIES

Address	Vector	Priority
0008h	External Interrupt on RA0/ INT pin (INTF)	1 (Highest)
0010h	TMR0 overflow interrupt (T0IF)	2
0018h	External Interrupt on T0CKI (T0CKIF)	3
0020h	Peripherals (PEIF)	4 (Lowest)

Note 1: Individual interrupt flag bits are set regardless of the status of their corresponding mask bit or the GLINTD bit.

Note 2: When disabling any of the INTSTA enable bits, the GLINTD bit should be set (disabled).

Note 3: For the PIC17C42 only:
If an interrupt occurs while the Global Interrupt Disable (GLINTD) bit is being set, the GLINTD bit may unintentionally be re-enabled by the user’s Interrupt Service Routine (the RETFIE instruction). The events that would cause this to occur are:

1. An interrupt occurs simultaneously with an instruction that sets the GLINTD bit.
2. The program branches to the Interrupt vector and executes the Interrupt Service Routine.
3. The Interrupt Service Routine completes with the execution of the RETFIE instruction. This causes the GLINTD bit to be cleared (enables interrupts), and the program returns to the instruction after the one which was meant to disable interrupts.

The method to ensure that interrupts are globally disabled is:

1. Ensure that the GLINTD bit was set by the instruction, as shown in the following code:

```
LOOP   BSF    CPUSTA, GLINTD ; Disable Global
                              ; Interrupt
       BTFSS  CPUSTA, GLINTD ; Global Interrupt
                              ; Disabled?
       GOTO   LOOP           ; NO, try again
                              ; YES, continue
                              ; with program
                              ; low
```

Example 8-4 shows the sequence to do an 16 x 16 signed multiply. Equation 8-2 shows the algorithm that used. The 32-bit result is stored in four registers RES3:RES0. To account for the sign bits of the arguments, each argument pairs most significant bit (MSb) is tested and the appropriate subtractions are done.

EQUATION 8-2: 16 x 16 SIGNED MULTIPLICATION ALGORITHM

RES3:RES0

= ARG1H:ARG1L * ARG2H:ARG2L

= (ARG1H * ARG2H * 2^{16}) +
(ARG1H * ARG2L * 2^{8}) +
(ARG1L * ARG2H * 2^{8}) +
(ARG1L * ARG2L) +
(-1 * ARG2H<7> * ARG1H:ARG1L * 2^{16}) +
(-1 * ARG1H<7> * ARG2H:ARG2L * 2^{16})

EXAMPLE 8-4: 16 x 16 SIGNED MULTIPLY ROUTINE

```
   MOVFP    ARG1L, WREG
   MULWF    ARG2L       ; ARG1L * ARG2L ->
                        ;   PRODH:PRODL
   MOVPF    PRODH, RES1 ;
   MOVPF    PRODL, RES0 ;
;
   MOVFP    ARG1H, WREG
   MULWF    ARG2H       ; ARG1H * ARG2H ->
                        ;   PRODH:PRODL
   MOVPF    PRODH, RES3 ;
   MOVPF    PRODL, RES2 ;
;
   MOVFP    ARG1L, WREG
   MULWF    ARG2H       ; ARG1L * ARG2H ->
                        ;   PRODH:PRODL
   MOVFP    PRODL, WREG ;
   ADDWF    RES1, F     ; Add cross
   MOVFP    PRODH, WREG ;    products
   ADDWFC   RES2, F     ;
   CLRF     WREG, F     ;
   ADDWFC   RES3, F     ;
;
   MOVFP    ARG1H, WREG ;
   MULWF    ARG2L       ; ARG1H * ARG2L ->
                        ;   PRODH:PRODL

   MOVFP    PRODL, WREG ;
   ADDWF    RES1, F     ; Add cross
   MOVFP    PRODH, WREG ;    products
   ADDWFC   RES2, F     ;
   CLRF     WREG, F     ;
   ADDWFC   RES3, F     ;
;
   BTFSS    ARG2H, 7    ; ARG2H:ARG2L neg?
   GOTO     SIGN_ARG1   ; no, check ARG1
   MOVFP    ARG1L, WREG ;
   SUBWF    RES2        ;
   MOVFP    ARG1H, WREG ;
   SUBWFB   RES3
;
SIGN_ARG1
   BTFSS    ARG1H, 7    ; ARG1H:ARG1L neg?
   GOTO     CONT_CODE   ; no, done
   MOVFP    ARG2L, WREG ;
   SUBWF    RES2        ;
   MOVFP    ARG2H, WREG ;
   SUBWFB   RES3
;
CONT_CODE
     :
```

NOTES:

9.4 PORTD and DDRD Registers

PORTD is an 8-bit bi-directional port. The corresponding data direction register is DDRD. A '1' in DDRD configures the corresponding port pin as an input. A '0' in the DDRC register configures the corresponding port pin as an output. Reading PORTD reads the status of the pins, whereas writing to it will write to the port latch. PORTD is multiplexed with the system bus. When operating as the system bus, PORTD is the high order byte of the address/data bus (AD15:AD8). The timing for the system bus is shown in the Electrical Characteristics section.

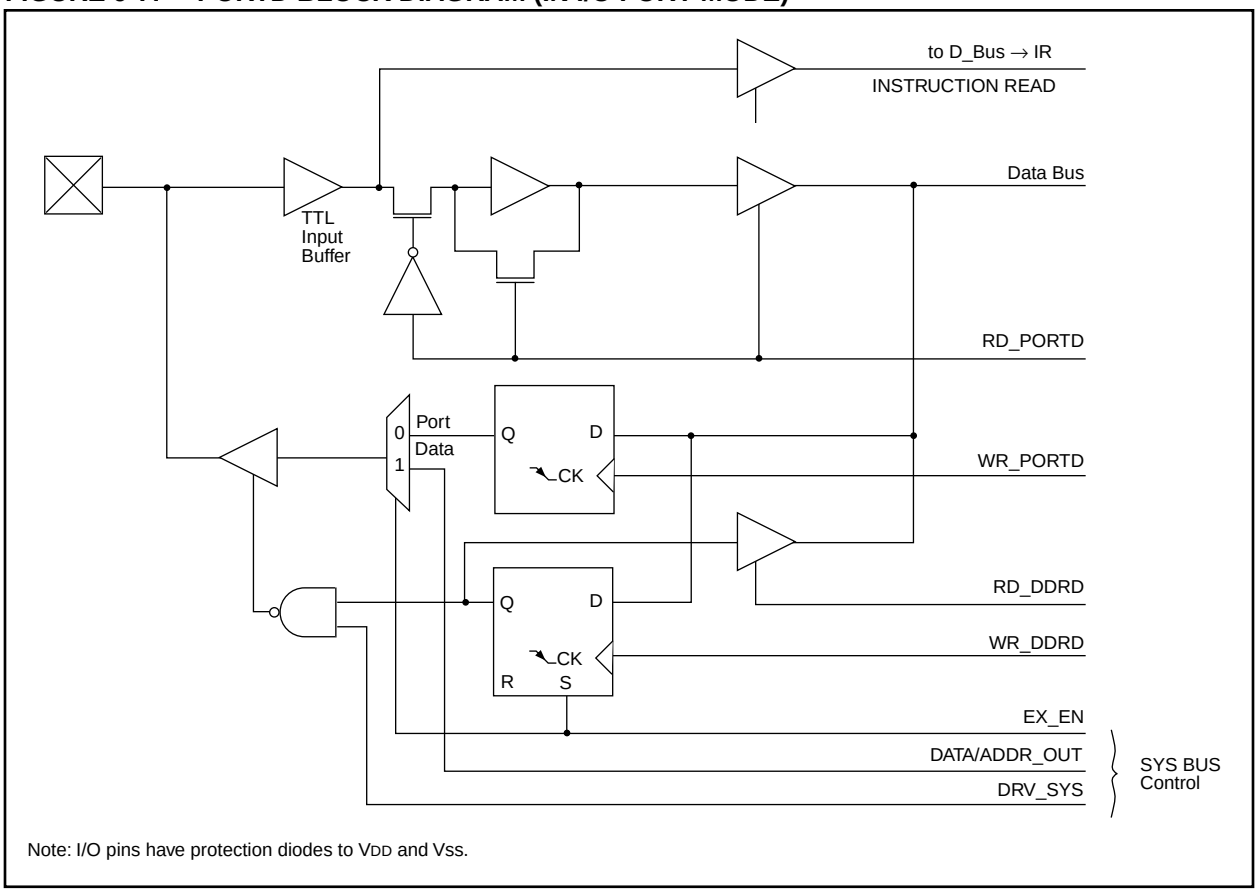
Note:	This port is configured as the system bus when the device's configuration bits are selected to Microprocessor or Extended Microcontroller modes. In the two other microcontroller modes, this port is a general purpose I/O.

Example 9-3 shows the instruction sequence to initialize PORTD. The Bank Select Register (BSR) must be selected to Bank 1 for the port to be initialized.

EXAMPLE 9-3: INITIALIZING PORTD

```
MOVLB  1          ;  Select Bank 1
CLRF   PORTD      ;  Initialize PORTD data
                  ;   latches before setting
                  ;   the data direction
                  ;   register
MOVLW  0xCF       ;  Value used to initialize
                  ;   data direction
MOVWF  DDRD       ;  Set RD<3:0> as inputs
                  ;   RD<5:4> as outputs
                  ;   RD<7:6> as inputs
```

FIGURE 9-7: PORTD BLOCK DIAGRAM (IN I/O PORT MODE)



Note: I/O pins have protection diodes to VDD and Vss.

13.2 USART Asynchronous Mode

In this mode, the USART uses standard nonreturn-to-zero (NRZ) format (one start bit, eight or nine data bits, and one stop bit). The most common data format is 8-bits. An on-chip dedicated 8-bit baud rate generator can be used to derive standard baud rate frequencies from the oscillator. The USART's transmitter and receiver are functionally independent but use the same data format and baud rate. The baud rate generator produces a clock x64 of the bit shift rate. Parity is not supported by the hardware, but can be implemented in software (and stored as the ninth data bit). Asynchronous mode is stopped during SLEEP.

The asynchronous mode is selected by clearing the SYNC bit (TXSTA<4>).

The USART Asynchronous module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver

13.2.1 USART ASYNCHRONOUS TRANSMITTER

The USART transmitter block diagram is shown in Figure 13-3. The heart of the transmitter is the transmit shift register (TSR). The shift register obtains its data from the read/write transmit buffer (TXREG). TXREG is loaded with data in software. The TSR is not loaded until the stop bit has been transmitted from the previous load. As soon as the stop bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once TXREG transfers the data to the TSR (occurs in one T_{CY} at the end of the current BRG cycle), the TXREG is empty and an interrupt bit, TXIF (PIR<1>) is set. This interrupt can be enabled or disabled by the TXIE bit (PIE<1>). TXIF will be set regardless of TXIE and cannot be reset in software. It will reset only when new data is loaded into TXREG. While TXIF indicates the status of the TXREG, the TRMT (TXSTA<1>) bit shows the status of the TSR. TRMT is a read only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR is empty.

Note:	The TSR is not mapped in data memory, so it is not available to the user.

Transmission is enabled by setting the TXEN (TXSTA<5>) bit. The actual transmission will not occur until TXREG has been loaded with data and the baud rate generator (BRG) has produced a shift clock (Figure 13-5). The transmission can also be started by first loading TXREG and then setting TXEN. Normally when transmission is first started, the TSR is empty, so a transfer to TXREG will result in an immediate transfer to TSR resulting in an empty TXREG. A back-to-back transfer is thus possible (Figure 13-6). Clearing TXEN during a transmission will cause the transmission to be aborted. This will reset the transmitter and the RA5/TX/CK pin will revert to hi-impedance.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG. This is because a data write to TXREG can result in an immediate transfer of the data to the TSR (if the TSR is empty).

Steps to follow when setting up an Asynchronous Transmission:

1. Initialize the SPBRG register for the appropriate baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. Load data to the TXREG register.
6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
7. Enable the transmission by setting TXEN (starts transmission).

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN) allows transmission to start sooner then doing these two events in the opposite order.

Note:	To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

13.2.2 USART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 13-4. The data comes in the RA4/RX/DT pin and drives the data recovery block. The data recovery block is actually a high speed shifter operating at 16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at FOSC.

Once asynchronous mode is selected, reception is enabled by setting bit CREN (RCSTA<4>).

The heart of the receiver is the receive (serial) shift register (RSR). After sampling the stop bit, the received data in the RSR is transferred to the RCREG (if it is empty). If the transfer is complete, the interrupt bit RCIF (PIR<0>) is set. The actual interrupt can be enabled/disabled by setting/clearing the RCIE (PIE<0>) bit. RCIF is a read only bit which is cleared by the hardware. It is cleared when RCREG has been read and is empty. RCREG is a double buffered register; (i.e. it is a two deep FIFO). It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte begin shifting to the RSR. On detection of the stop bit of the third byte, if the RCREG is still full, then the overrun error bit, OERR (RCSTA<1>) will be set. The word in the RSR will be lost. RCREG can be read twice to retrieve the two bytes in the FIFO. The OERR bit has to be cleared in software which is done by resetting the receive logic (CREN is set). If the OERR bit is set, transfers from the RSR to RCREG are inhibited, so it is essential to clear the OERR bit if it is set. The framing error bit FERR (RCSTA<2>) is set if a stop bit is not detected.

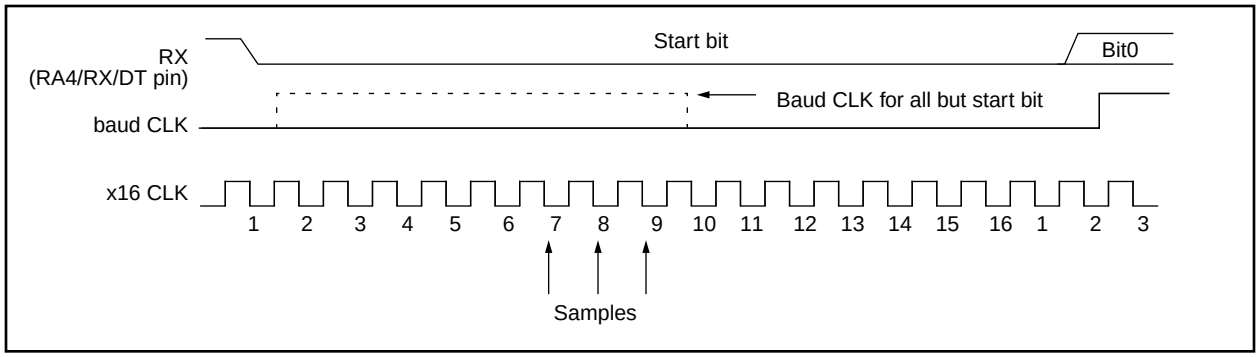
Note:	The FERR and the 9th receive bit are buffered the same way as the receive data. Reading the RCREG register will allow the RX9D and FERR bits to be loaded with values for the next received Received data; therefore, it is essential for the user to read the RCSTA register before reading RCREG in order not to lose the old FERR and RX9D information.

13.2.3 SAMPLING

The data on the RA4/RX/DT pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RA4/RX/DT pin. The sampling is done on the seventh, eighth and ninth falling edges of a x16 clock (Figure 11-3).

The x16 clock is a free running clock, and the three sample points occur at a frequency of every 16 falling edges.

FIGURE 13-7: RX PIN SAMPLING SCHEME



13.3.2 USART SYNCHRONOUS MASTER RECEPTION

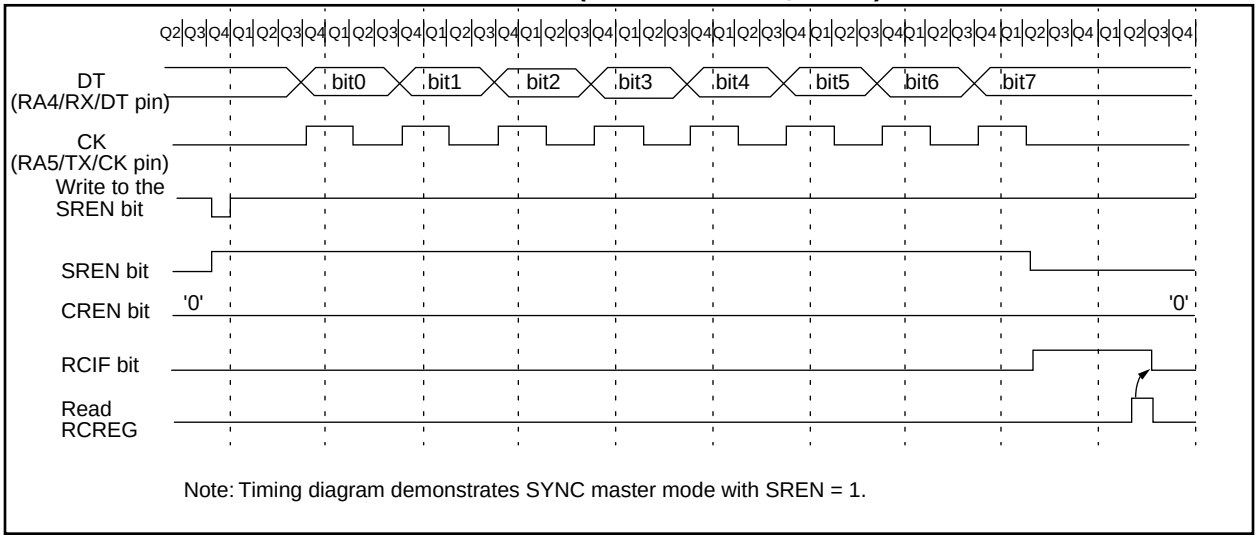
Once synchronous mode is selected, reception is enabled by setting either the SREN (RCSTA<5>) bit or the CREN (RCSTA<4>) bit. Data is sampled on the RA4/RX/DT pin on the falling edge of the clock. If SREN is set, then only a single word is received. If CREN is set, the reception is continuous until CREN is reset. If both bits are set, then CREN takes precedence. After clocking the last bit, the received data in the Receive Shift Register (RSR) is transferred to RCREG (if it is empty). If the transfer is complete, the interrupt bit RCIF (PIR<0>) is set. The actual interrupt can be enabled/disabled by setting/clearing the RCIE (PIE<0>) bit. RCIF is a read only bit which is RESET by the hardware. In this case it is reset when RCREG has been read and is empty. RCREG is a double buffered register; i.e., it is a two deep FIFO. It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte to begin shifting into the RSR. On the clocking of the last bit of the third byte, if RCREG is still full, then the overrun error bit OERR (RCSTA<1>) is set. The word in the RSR will be lost. RCREG can be read twice to retrieve the two bytes in the FIFO. The OERR bit has to be cleared in software. This is done by clearing the CREN bit. If OERR bit is set, transfers from RSR to RCREG are inhibited, so it is essential to clear OERR bit if it is set. The 9th receive bit is buffered the same way as the receive data. Reading the RCREG register will allow the RX9D and FERR bits to be loaded with values for the next received data; therefore, it is essential for the user to read the RCSTA register before reading RCREG in order not to lose the old FERR and RX9D information.

Steps to follow when setting up a Synchronous Master Reception:

1. Initialize the SPBRG register for the appropriate baud rate. See Section 13.1 for details.
2. Enable the synchronous master serial port by setting bits SYNC, SPEN, and CSRC.
3. If interrupts are desired, then set the RCIE bit.
4. If 9-bit reception is desired, then set the RX9 bit.
5. If a single reception is required, set bit SREN. For continuous reception set bit CREN.
6. The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
7. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading RCREG.
9. If any error occurred, clear the error by clearing CREN.

Note:	To terminate a reception, either clear the SREN and CREN bits, or the SPEN bit. This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

FIGURE 13-11: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note: Timing diagram demonstrates SYNC master mode with SREN = 1.

13.4 USART Synchronous Slave Mode

The synchronous slave mode differs from the master mode in the fact that the shift clock is supplied externally at the RA5/TX/CK pin (instead of being supplied internally in the master mode). This allows the device to transfer or receive data in the SLEEP mode. The slave mode is entered by clearing the CSRC (TXSTA<7>) bit.

13.4.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the sync master and slave modes are identical except in the case of the SLEEP mode.

If two words are written to TXREG and then the SLEEP instruction executes, the following will occur. The first word will immediately transfer to the TSR and will transmit as the shift clock is supplied. The second word will remain in TXREG. TXIF will not be set. When the first word has been shifted out of TSR, TXREG will transfer the second word to the TSR and the TXIF flag will now be set. If TXIE is enabled, the interrupt will wake the chip from SLEEP and if the global interrupt is enabled, then the program will branch to interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. Clear the CREN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. Start transmission by loading data to TXREG.
6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
7. Enable the transmission by setting TXEN.

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN) allows transmission to start sooner then doing these two events in the reverse order.

Note:	To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

13.4.2 USART SYNCHRONOUS SLAVE RECEPTION

Operation of the synchronous master and slave modes are identical except in the case of the SLEEP mode. Also, SREN is a don't care in slave mode.

If receive is enabled (CREN) prior to the SLEEP instruction, then a word may be received during SLEEP. On completely receiving the word, the RSR will transfer the data to RCREG (setting RCIF) and if the RCIE bit is set, the interrupt generated will wake the chip from SLEEP. If the global interrupt is enabled, the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Reception:

1. Enable the synchronous master serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. If interrupts are desired, then set the RCIE bit.
3. If 9-bit reception is desired, then set the RX9 bit.
4. To enable reception, set the CREN bit.
5. The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
6. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
7. Read the 8-bit received data by reading RCREG.
8. If any error occurred, clear the error by clearing the CREN bit.

Note:	To abort reception, either clear the SPEN bit, the SREN bit (when in single receive mode), or the CREN bit (when in continuous receive mode). This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

15.2 Q Cycle Activity

Each instruction cycle (Tcy) is comprised of four Q cycles (Q1-Q4). The Q cycles provide the timing/designation for the Decode, Read, Execute, Write etc., of each instruction cycle. The following diagram shows the relationship of the Q cycles to the instruction cycle.

The 4 Q cycles that make up an instruction cycle (Tcy) can be generalized as:

Q1: Instruction Decode Cycle or forced NOP

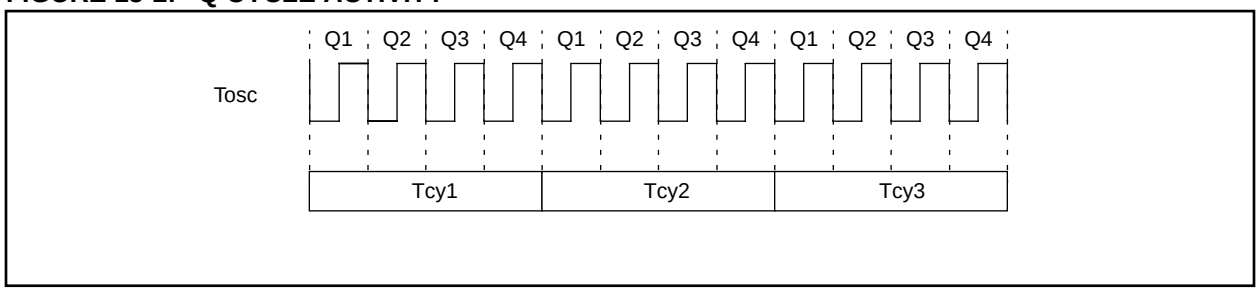
Q2: Instruction Read Cycle or NOP

Q3: Instruction Execute

Q4: Instruction Write Cycle or NOP

Each instruction will show the detailed Q cycle operation for the instruction.

FIGURE 15-2: Q CYCLE ACTIVITY



Applicable Devices	42	R42	42A	43	R43	44

FIGURE 17-5: TIMER0 CLOCK TIMINGS

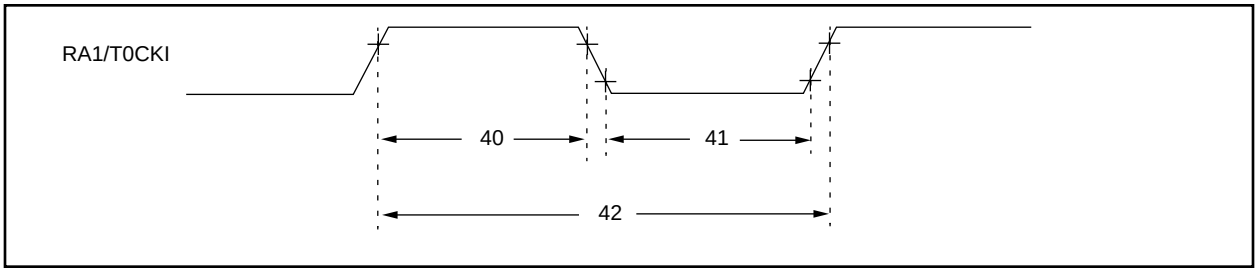


TABLE 17-5: TIMER0 CLOCK REQUIREMENTS

Parameter No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
40	Tt0H	T0CKI High Pulse Width	No Prescaler	0.5TCY + 20 §	—	—	ns	
			With Prescaler	10*	—	—	ns	
41	Tt0L	T0CKI Low Pulse Width	No Prescaler	0.5TCY + 20 §	—	—	ns	
			With Prescaler	10*	—	—	ns	
42	Tt0P	T0CKI Period		TCY + 40 § / N	—	—	ns	N = prescale value (1, 2, 4, ..., 256)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification ensured by design.

FIGURE 17-6: TIMER1, TIMER2, AND TIMER3 CLOCK TIMINGS

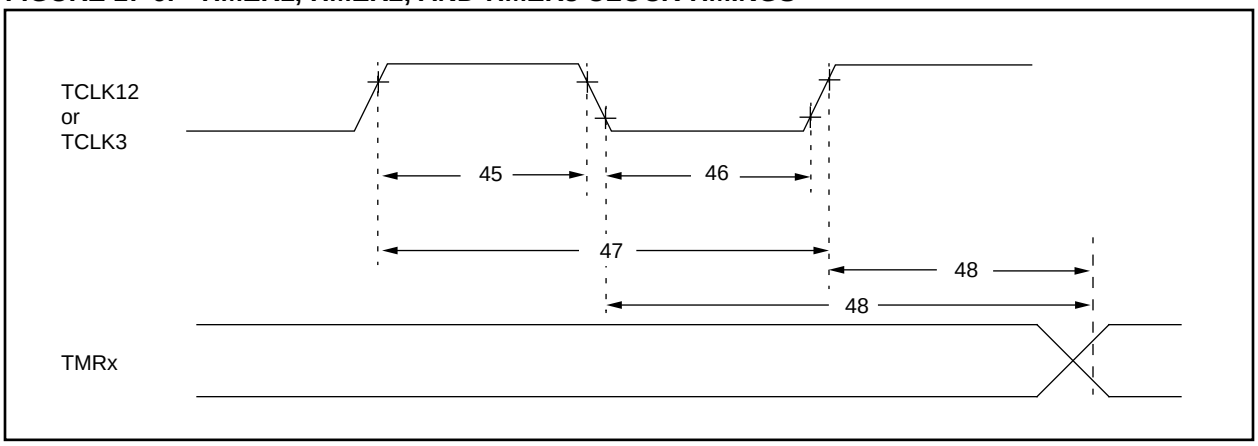


TABLE 17-6: TIMER1, TIMER2, AND TIMER3 CLOCK REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ †	Max	Units	Conditions
45	Tt123H	TCLK12 and TCLK3 high time	0.5 TCY + 20 §	—	—	ns	
46	Tt123L	TCLK12 and TCLK3 low time	0.5 TCY + 20 §	—	—	ns	
47	Tt123P	TCLK12 and TCLK3 input period	TCY + 40 § / N	—	—	ns	N = prescale value (1, 2, 4, 8)
48	TckE2tmrI	Delay from selected External Clock Edge to Timer increment	2TOSC §	—	6 Tosc §	—	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification ensured by design.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 18-7: TYPICAL IDD vs. FREQUENCY (EXTERNAL CLOCK 25°C)

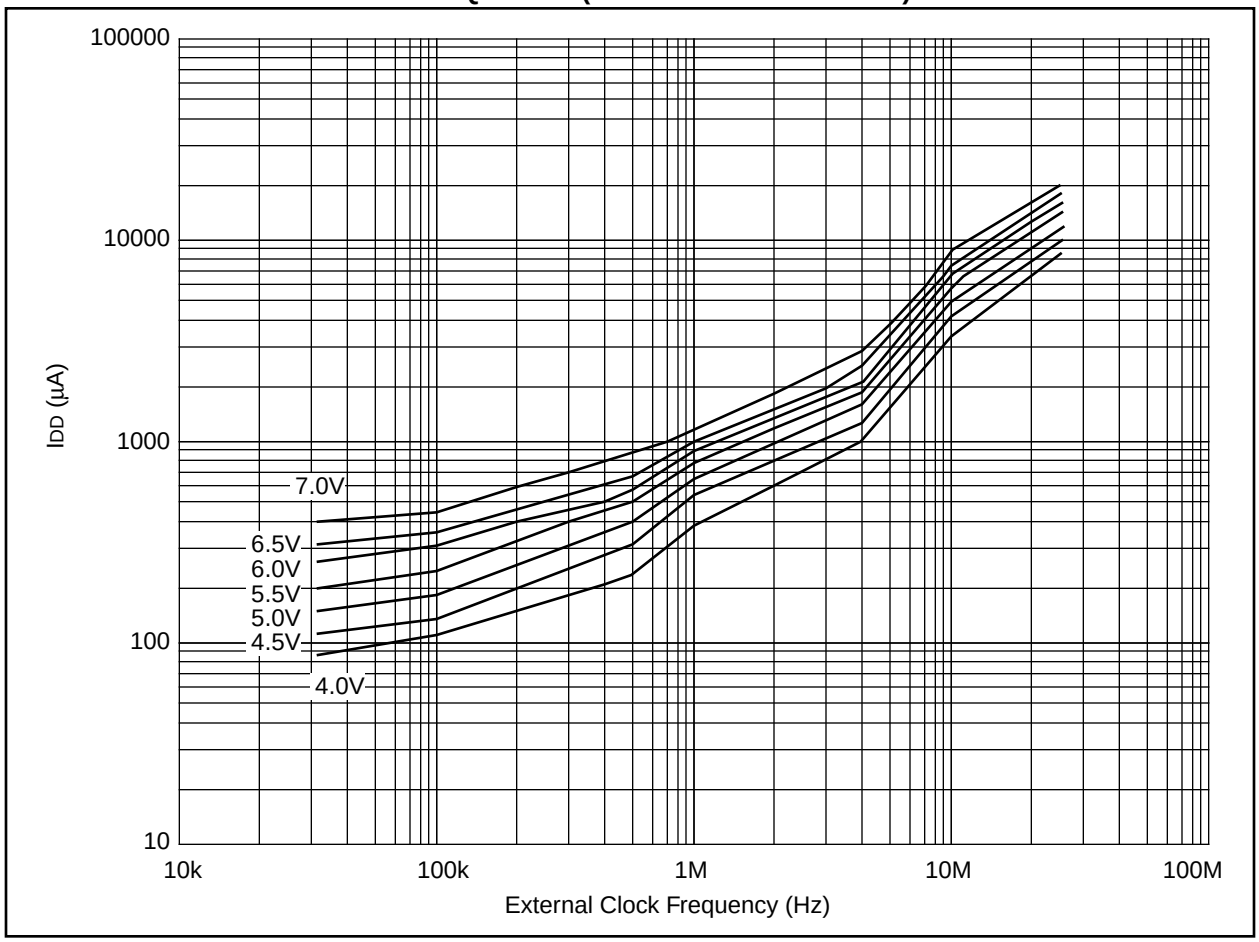
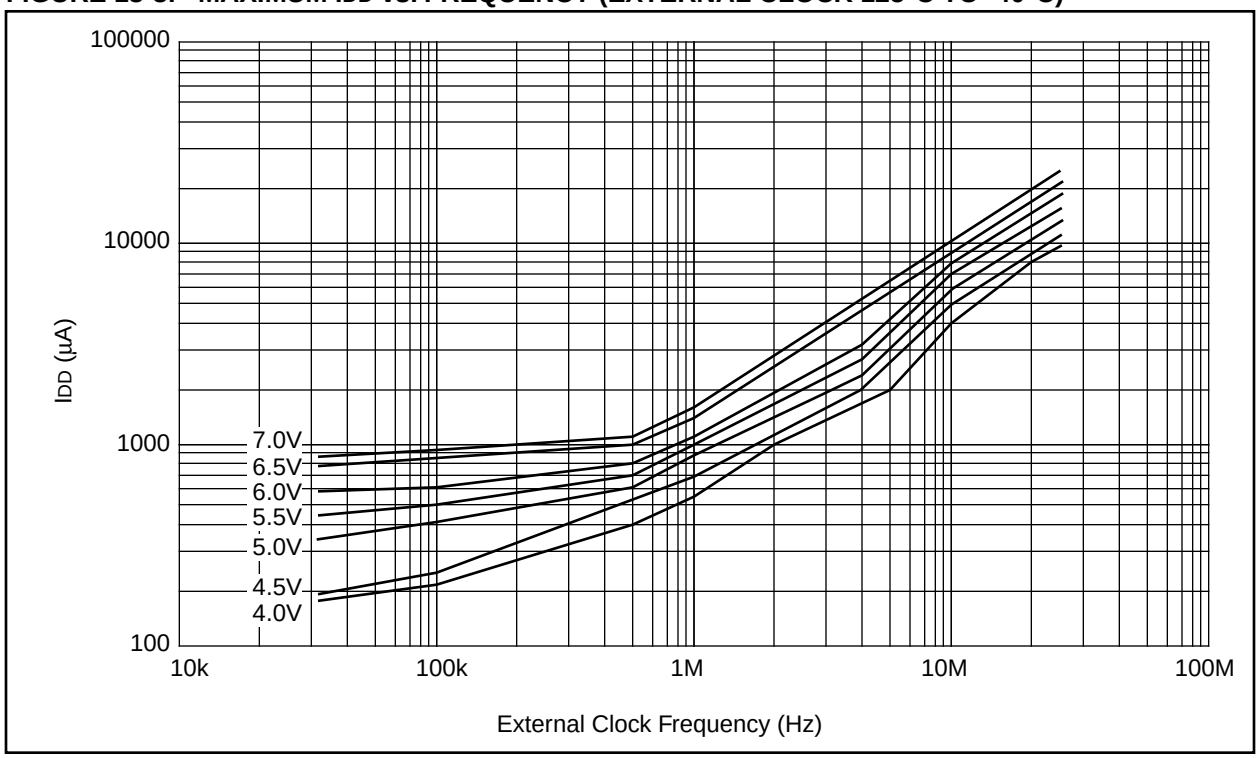


FIGURE 18-8: MAXIMUM IDD vs. FREQUENCY (EXTERNAL CLOCK 125°C TO -40°C)



Applicable Devices	42	R42	42A	43	R43	44

19.0 PIC17CR42/42A/43/R43/44 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Ambient temperature under bias ..... -55 to +125˚C

Storage temperature ..... -65˚C to +150˚C

Voltage on V_{DD} with respect to V_{SS} ..... 0 to +7.5V

Voltage on $\overline{MCLR}$ with respect to V_{SS} (Note 2) ..... -0.6V to +14V

Voltage on RA2 and RA3 with respect to V_{SS} ..... -0.6V to +14V

Voltage on all other pins with respect to V_{SS} ..... -0.6V to V_{DD} + 0.6V

Total power dissipation (Note 1) 1.0W

Maximum current out of V_{SS} pin(s) - total ..... 250 mA

Maximum current into V_{DD} pin(s) - total ..... 200 mA

Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ±20 mA

Output clamp current, I_{OK} (V_O < 0 or V_O > V_{DD}) ..... ±20 mA

Maximum output current sunk by any I/O pin (except RA2 and RA3) ..... 35 mA

Maximum output current sunk by RA2 or RA3 pins ..... 60 mA

Maximum output current sourced by any I/O pin ..... 20 mA

Maximum current sunk by PORTA and PORTB (combined) ..... 150 mA

Maximum current sourced by PORTA and PORTB (combined) ..... 100 mA

Maximum current sunk by PORTC, PORTD and PORTE (combined) ..... 150 mA

Maximum current sourced by PORTC, PORTD and PORTE (combined) ..... 100 mA

Note 1: Power dissipation is calculated as follows: $Pdis = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD}-V_{OH}) \times I_{OH}\} + \sum(V_{OL} \times I_{OL})$

Note 2: Voltage spikes below V_{SS} at the $\overline{MCLR}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the $\overline{MCLR}$ pin rather than pulling this pin directly to V_{SS}.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Applicable Devices	42	R42	42A	43	R43	44

19.3 DC CHARACTERISTICS: PIC17CR42/42A/43/R43/44-16 (Commercial, Industrial) PIC17CR42/42A/43/R43/44-25 (Commercial, Industrial) PIC17CR42/42A/43/R43/44-33 (Commercial, Industrial) PIC17LCR42/42A/43/R43/44-08 (Commercial, Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial Operating voltage VDD range as described in Section 19.1				
Parameter No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
		Input Low Voltage					
	VIL	I/O ports					
D030		with TTL buffer	VSS	–	0.8	V	4.5V ≤ VDD ≤ 5.5V
			VSS	–	0.2VDD	V	2.5V ≤ VDD ≤ 4.5V
D031		with Schmitt Trigger buffer	VSS	–	0.2VDD	V	
D032		$\overline{MCLR}$, OSC1 (in EC and RC mode)	Vss	–	0.2VDD	V	Note1
D033		OSC1 (in XT, and LF mode)	–	0.5VDD	–	V	
		Input High Voltage					
	VIH	I/O ports					
D040		with TTL buffer	2.0	–	VDD	V	4.5V ≤ VDD ≤ 5.5V
			1 + 0.2VDD	–	VDD	V	2.5V ≤ VDD ≤ 4.5V
D041		with Schmitt Trigger buffer	0.8VDD	–	VDD	V	
D042		$\overline{MCLR}$	0.8VDD	–	VDD	V	Note1
D043		OSC1 (XT, and LF mode)	–	0.5VDD	–	V	
D050	VHYS	Hysteresis of Schmitt Trigger inputs	0.15VDD *	–	–	V	
		Input Leakage Current (Notes 2, 3)					
D060	IIL	I/O ports (except RA2, RA3)	–	–	±1	µA	Vss ≤ VPIN ≤ VDD, I/O Pin at hi-impedance PORTB weak pull-ups disabled
D061		$\overline{MCLR}$	–	–	±2	µA	VPIN = Vss or VPIN = VDD
D062		RA2, RA3			±2	µA	Vss ≤ VRA2, VRA3 ≤ 12V
D063		OSC1, TEST (EC, RC modes)	–	–	±1	µA	Vss ≤ VPIN ≤ VDD
D063B		OSC1, TEST (XT, LF modes)	–	–	VPIN	µA	RF ≥ 1 MΩ, see Figure 14.2
D064		$\overline{MCLR}$	–	–	10	µA	VMCLR = VPP = 12V (when not programming)
D070	IPURB	PORTB weak pull-up current	60	200	400	µA	VPIN = VSS, $\overline{RBPU}$ = 0 4.5V ≤ VDD ≤ 6.0V

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

‡ These parameters are for design guidance only and are not tested, nor characterized.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended that the PIC17CXX devices be driven with external clock in RC mode.

2: The leakage current on the $\overline{MCLR}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

3: Negative current is defined as coming out of the pin.

4: These specifications are for the programming of the on-chip program memory EPROM through the use of the table write instructions. The complete programming specifications can be found in: PIC17CXX Programming Specifications (Literature number DS30139).

5: The $\overline{MCLR}$/VPP pin may be kept in this range at times other than programming, but is not recommended.

6: For TTL buffers, the better of the two specifications may be used.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 19-11: MEMORY INTERFACE WRITE TIMING (NOT SUPPORTED IN PIC17LC4X DEVICES)

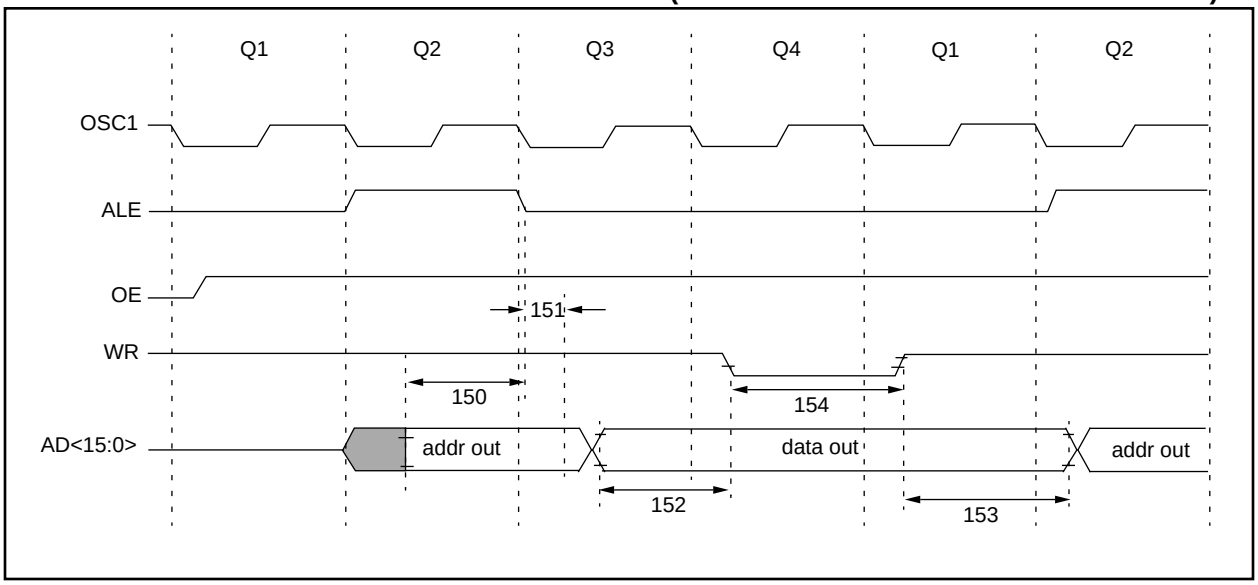


TABLE 19-11: MEMORY INTERFACE WRITE REQUIREMENTS (NOT SUPPORTED IN PIC17LC4X DEVICES)

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
150	TadV2alL	AD<15:0> (address) valid to ALE↓ (address setup time)	0.25Tcy - 10	—	—	ns	
151	TalL2adI	ALE↓ to address out invalid (address hold time)	0	—	—	ns	
152	TadV2wrL	Data out valid to $\overline{WR}$↓ (data setup time)	0.25Tcy - 40	—	—	ns	
153	TwrH2adI	$\overline{WR}$↑ to data out invalid (data hold time)	—	0.25TCY §	—	ns	
154	TwrL	$\overline{WR}$ pulse width	—	0.25TCY §	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification ensured by design.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 20-9: TYPICAL IPD vs. VDD WATCHDOG DISABLED 25°C

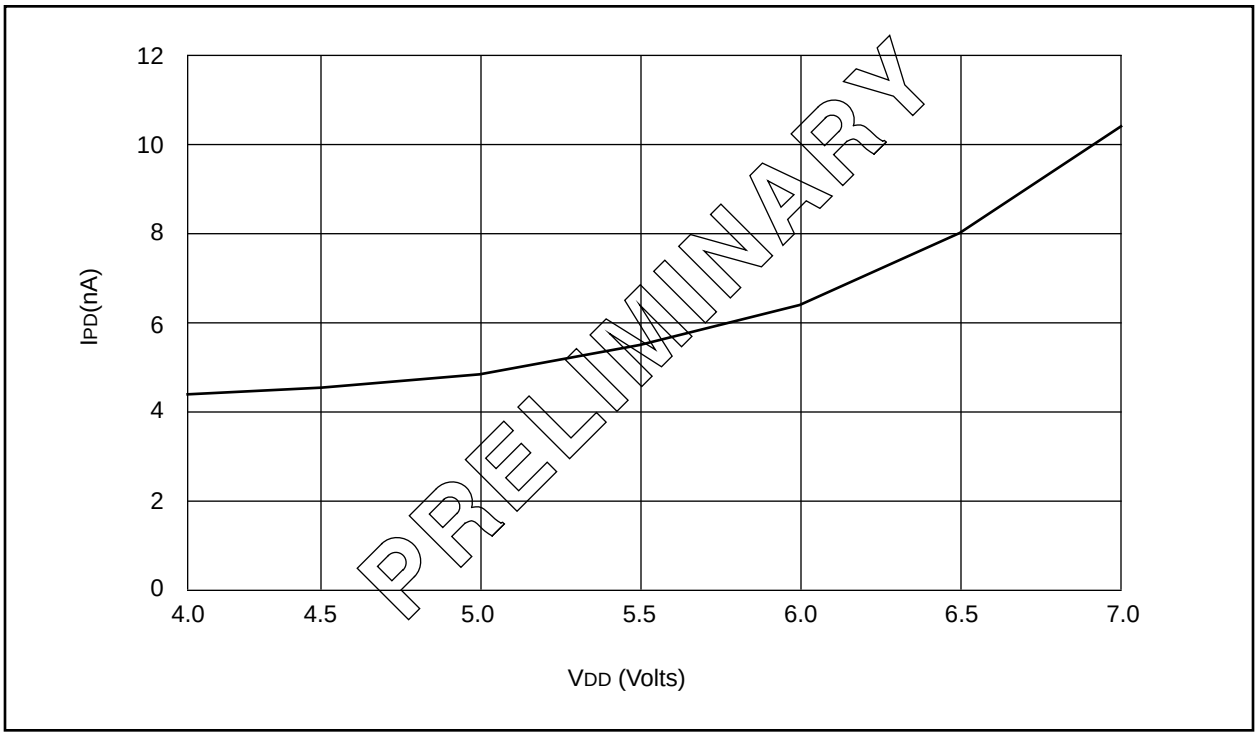
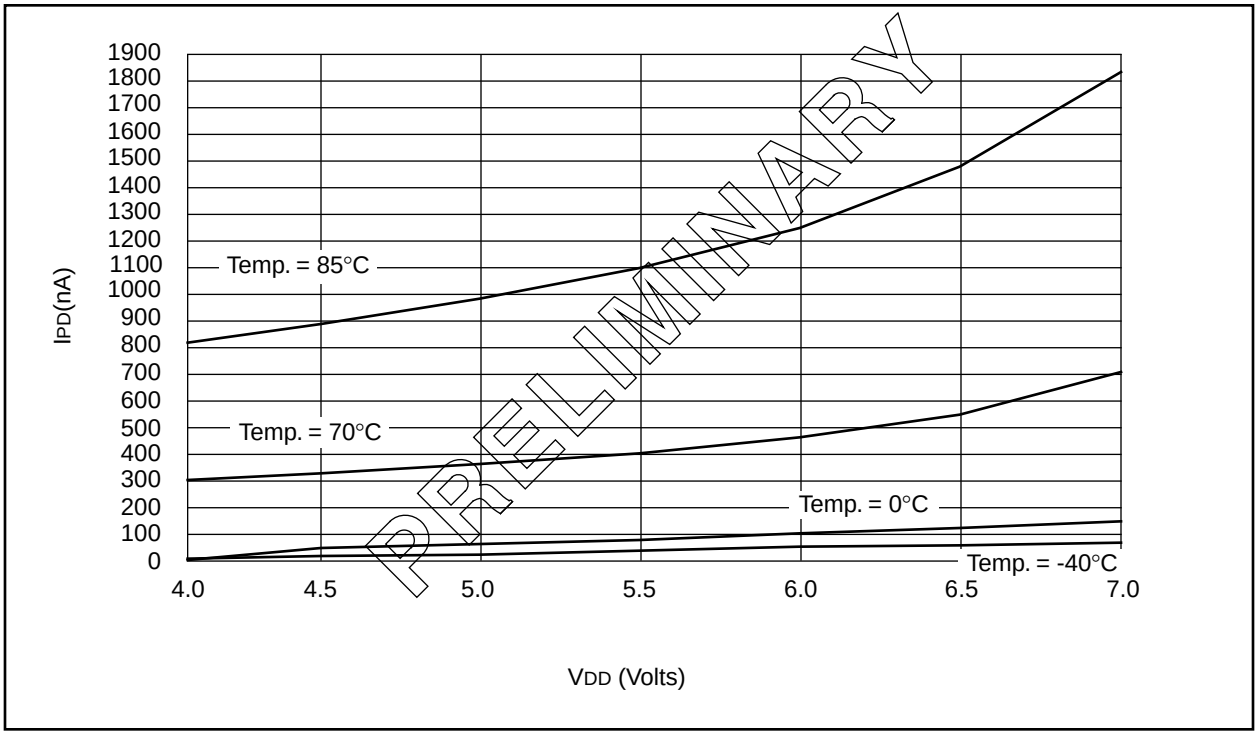


FIGURE 20-10: MAXIMUM IPD vs. VDD WATCHDOG DISABLED



Applicable Devices	42	R42	42A	43	R43	44

FIGURE 20-13: WDT TIMER TIME-OUT PERIOD vs. VDD

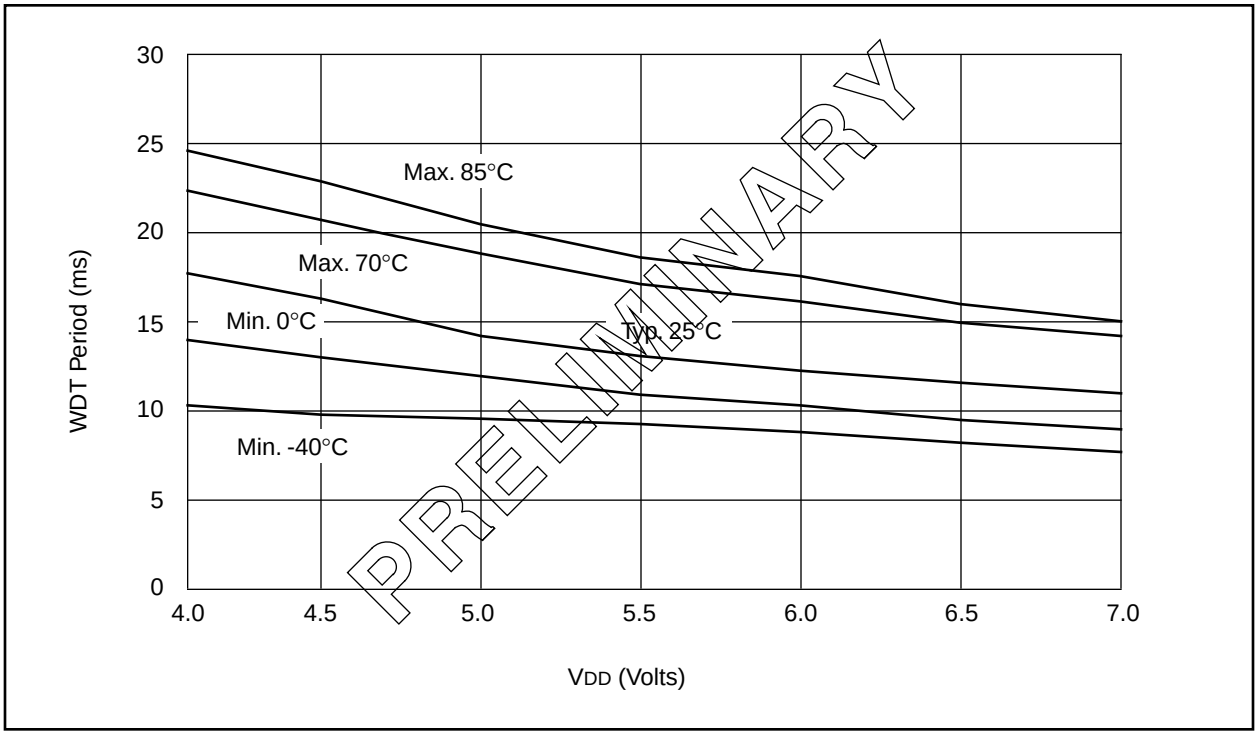
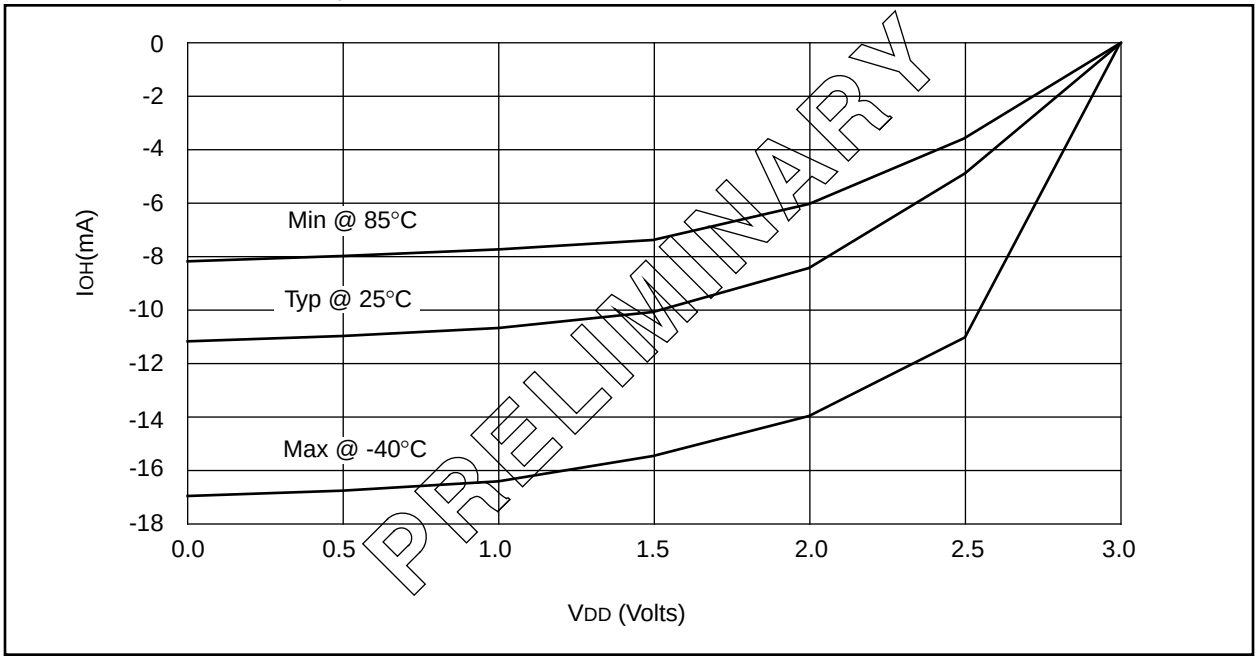
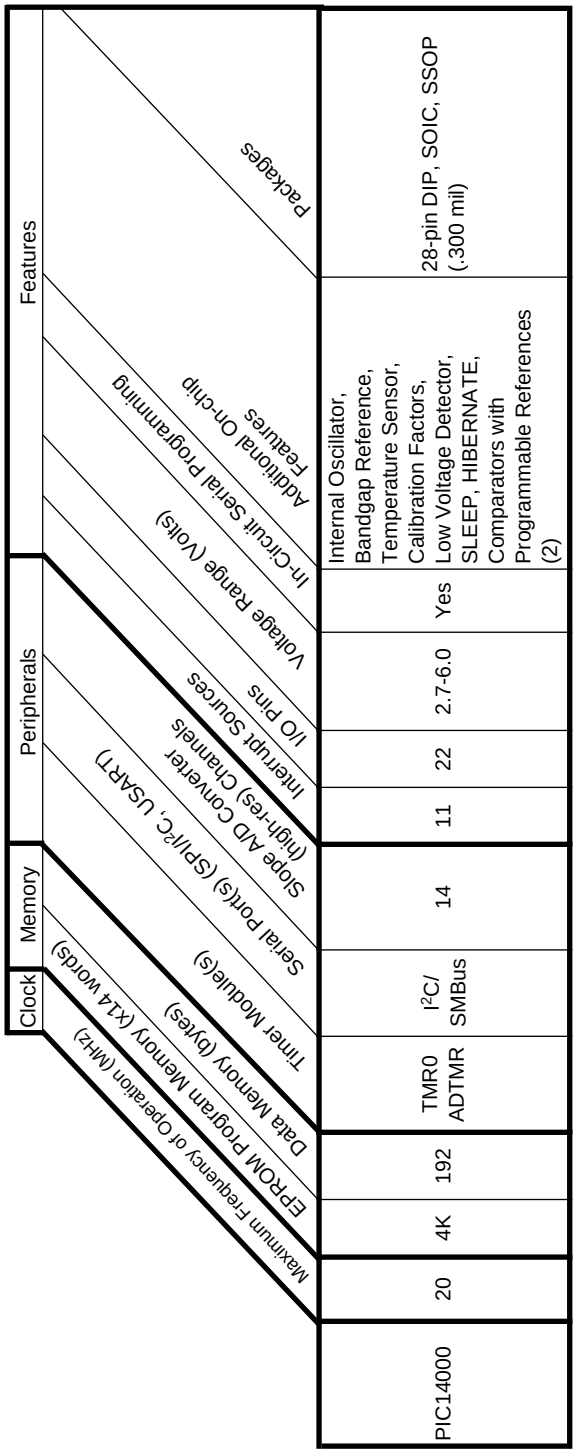


FIGURE 20-14: IOH vs. VOH, VDD = 3V



APPENDIX E: PIC16/17 MICROCONTROLLERS

E.1 PIC14000 Devices



E.2 PIC16C5X Family of Devices

	Clock	Memory				Peripherals		Features	
	Maximum Frequency of Operation (MHz)	Program Memory (x12 words)	EPROM	ROM	RAM Data Memory (bytes)	Timer Module(s)	I/O Pins	Voltage Range (Volts)	Number of Instructions	Packages
PIC16C52	4	384	—	25	TMR0	12	2.5-6.25	33	18-pin DIP, SOIC
PIC16C54	20	512	—	25	TMR0	12	2.5-6.25	33	18-pin DIP, SOIC; 20-pin SSOP
PIC16C54A	20	512	—	25	TMR0	12	2.0-6.25	33	18-pin DIP, SOIC; 20-pin SSOP
PIC16CR54A	20	—	512	25	TMR0	12	2.0-6.25	33	18-pin DIP, SOIC; 20-pin SSOP
PIC16C55	20	512	—	24	TMR0	20	2.5-6.25	33	28-pin DIP, SOIC, SSOP
PIC16C56	20	1K	—	25	TMR0	12	2.5-6.25	33	18-pin DIP, SOIC; 20-pin SSOP
PIC16C57	20	2K	—	72	TMR0	20	2.5-6.25	33	28-pin DIP, SOIC, SSOP
PIC16CR57B	20	—	2K	72	TMR0	20	2.5-6.25	33	28-pin DIP, SOIC, SSOP
PIC16C58A	20	2K	—	73	TMR0	12	2.0-6.25	33	18-pin DIP, SOIC; 20-pin SSOP
PIC16CR58A	20	—	2K	73	TMR0	12	2.5-6.25	33	18-pin DIP, SOIC; 20-pin SSOP

All PIC16/17 Family devices have Power-On Reset, selectable Watchdog Timer, selectable code protect and high I/O current capability.

E.5 PIC16C7X Family of Devices

Clock		Memory		Peripherals					Features			
Maximum Frequency of Operation (MHz)	EPROM Program Memory (x14 words)	Data Memory (bytes)	Timer Module(s)	Capture/Compare/PWM Module(s)	Serial Port(s) (SPI/I²C, USART)	Parallel Slave Port	A/D Converter (8-bit) Channels	Interrupt Sources	I/O Pins	Voltage Range (Volts)	In-Circuit Serial Programming	Brown-out Reset	Packages
PIC16C710	20	512	36	TMR0	—	—	—	4	4	13	3.0-6.0	Yes	Yes	18-pin DIP, SOIC; 20-pin SSOP
PIC16C71	20	1K	36	TMR0	—	—	—	4	4	13	3.0-6.0	Yes	—	18-pin DIP, SOIC
PIC16C711	20	1K	68	TMR0	—	—	—	4	4	13	3.0-6.0	Yes	Yes	18-pin DIP, SOIC; 20-pin SSOP
PIC16C72	20	2K	128	TMR0, TMR1, TMR2	1	SPI/I²C	—	5	8	22	2.5-6.0	Yes	Yes	28-pin SDIP, SOIC, SSOP
PIC16C73	20	4K	192	TMR0, TMR1, TMR2	2	SPI/I²C, USART	—	5	11	22	3.0-6.0	Yes	—	28-pin SDIP, SOIC
PIC16C73A(1)	20	4K	192	TMR0, TMR1, TMR2	2	SPI/I²C, USART	—	5	11	22	2.5-6.0	Yes	Yes	28-pin SDIP, SOIC
PIC16C74	20	4K	192	TMR0, TMR1, TMR2	2	SPI/I²C, USART	Yes	8	12	33	3.0-6.0	Yes	—	40-pin DIP; 44-pin PLCC, MQFP
PIC16C74A(1)	20	4K	192	TMR0, TMR1, TMR2	2	SPI/I²C, USART	Yes	8	12	33	2.5-6.0	Yes	Yes	40-pin DIP; 44-pin PLCC, MQFP, TQFP

All PIC16/17 Family devices have Power-on Reset, selectable Watchdog Timer, selectable code protect and high I/O current capability.

All PIC16C7X Family devices use serial programming with clock pin RB6 and data pin RB7.

Note 1: Please contact your local sales office for availability of these devices.

E.6 PIC16C8X Family of Devices

	Clock	Memory				Peripherals		Features		
	Maximum Frequency of Operation (MHz)	Program Memory	Flash	EEPROM	ROM	Data Memory (bytes)	Data EEPROM (bytes)	Timer Module(s)	Interrupt Sources	I/O Pins	Voltage Range (Volts)	Packages
PIC16C84	10	—	1K	—	36	64		TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16F84(1)	10	1K	—	—	68	64		TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16CR84(1)	10	—	—	1K	68	64		TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16F83(1)	10	512	—	—	36	64		TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16CR83(1)	10	—	—	512	36	64		TMR0	4	13	2.0-6.0	18-pin DIP, SOIC

All PIC16/17 family devices have Power-on Reset, selectable Watchdog Timer, selectable code protect, and high I/O current capability.

All PIC16C8X family devices use serial programming with clock pin RB6 and data pin RB7.

Note 1: Please contact your local sales office for availability of these devices.