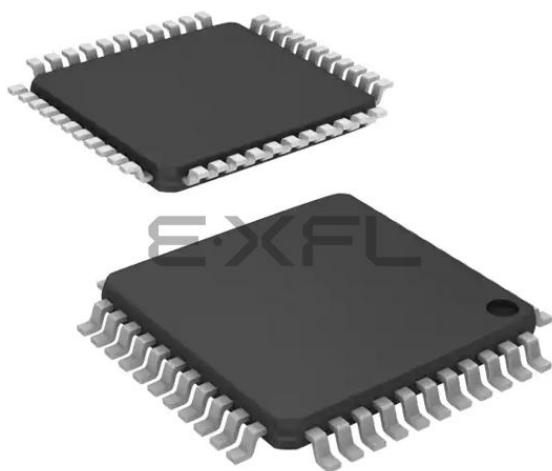




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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	16MHz
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	33
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	454 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c44-16i-pt

PIC17C4X

NOTES:

13.2 USART Asynchronous Mode

In this mode, the USART uses standard nonreturn-to-zero (NRZ) format (one start bit, eight or nine data bits, and one stop bit). The most common data format is 8-bits. An on-chip dedicated 8-bit baud rate generator can be used to derive standard baud rate frequencies from the oscillator. The USART's transmitter and receiver are functionally independent but use the same data format and baud rate. The baud rate generator produces a clock x64 of the bit shift rate. Parity is not supported by the hardware, but can be implemented in software (and stored as the ninth data bit). Asynchronous mode is stopped during SLEEP.

The asynchronous mode is selected by clearing the SYNC bit (TXSTA<4>).

The USART Asynchronous module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver

13.2.1 USART ASYNCHRONOUS TRANSMITTER

The USART transmitter block diagram is shown in Figure 13-3. The heart of the transmitter is the transmit shift register (TSR). The shift register obtains its data from the read/write transmit buffer (TXREG). TXREG is loaded with data in software. The TSR is not loaded until the stop bit has been transmitted from the previous load. As soon as the stop bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once TXREG transfers the data to the TSR (occurs in one Tcy at the end of the current BRG cycle), the TXREG is empty and an interrupt bit, TXIF (PIR<1>) is set. This interrupt can be enabled or disabled by the TXIE bit (PIE<1>). TXIF will be set regardless of TXIE and cannot be reset in software. It will reset only when new data is loaded into TXREG. While TXIF indicates the status of the TXREG, the TRMT (TXSTA<1>) bit shows the status of the TSR. TRMT is a read only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR is empty.

Note: The TSR is not mapped in data memory, so it is not available to the user.

Transmission is enabled by setting the TXEN (TXSTA<5>) bit. The actual transmission will not occur until TXREG has been loaded with data and the baud rate generator (BRG) has produced a shift clock (Figure 13-5). The transmission can also be started by first loading TXREG and then setting TXEN. Normally when transmission is first started, the TSR is empty, so a transfer to TXREG will result in an immediate transfer to TSR resulting in an empty TXREG. A back-to-back transfer is thus possible (Figure 13-6). Clearing TXEN during a transmission will cause the transmission to be aborted. This will reset the transmitter and the RA5/TX/CK pin will revert to hi-impedance.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG. This is because a data write to TXREG can result in an immediate transfer of the data to the TSR (if the TSR is empty).

Steps to follow when setting up an Asynchronous Transmission:

1. Initialize the SPBRG register for the appropriate baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. Load data to the TXREG register.
6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
7. Enable the transmission by setting TXEN (starts transmission).

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN) allows transmission to start sooner than doing these two events in the opposite order.

Note: To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

FIGURE 13-5: ASYNCHRONOUS MASTER TRANSMISSION

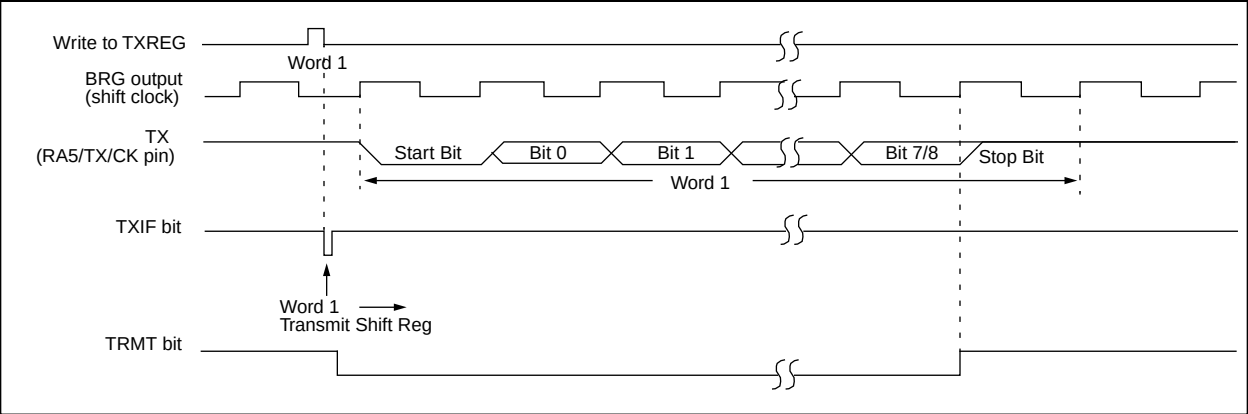


FIGURE 13-6: ASYNCHRONOUS MASTER TRANSMISSION (BACK TO BACK)

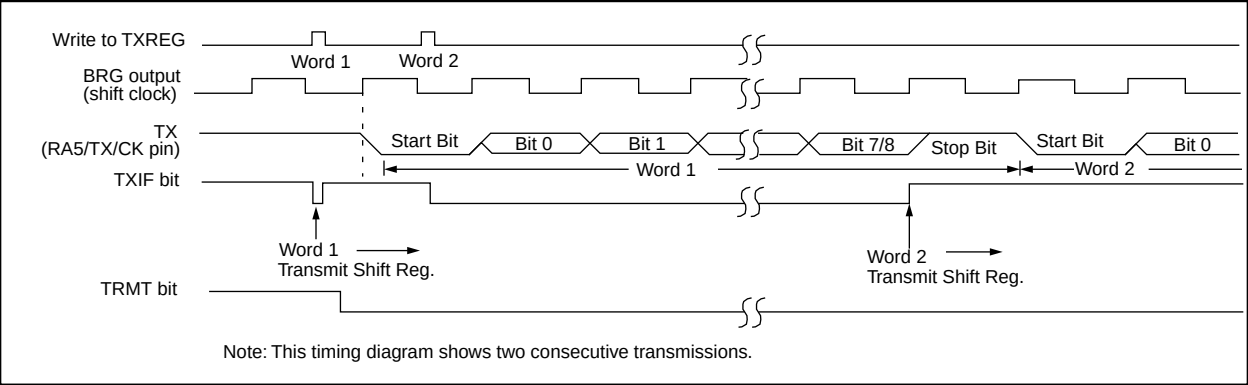


TABLE 13-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
16h, Bank 1	PIR	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TXIF	RCIF	0000 0010	0000 0010
13h, Bank 0	RCSTA	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
16h, Bank 0	TXREG	Serial port transmit register								xxxx xxxx	uuuu uuuu
17h, Bank 1	PIE	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TXIE	RCIE	0000 0000	0000 0000
15h, Bank 0	TXSTA	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG	Baud rate generator register								xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', shaded cells are not used for asynchronous transmission.

Note 1: Other (non power-up) resets include: external reset through $\overline{\text{MCLR}}$ and Watchdog Timer Reset.

14.1 Configuration Bits

The PIC17CXX has up to seven configuration locations (Table 14-1). These locations can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. Any write to a configuration location, regardless of the data, will program that configuration bit. A TABLWT instruction is required to write to program memory locations. The configuration bits can be read by using the TABLRD instructions. Reading any configuration location between FE00h and FE07h will read the low byte of the configuration word (Figure 14-1) into the TABLATL register. The TABLATH register will be FFh. Reading a configuration location between FE08h and FE0Fh will read the high byte of the configuration word into the TABLATL register. The TABLATH register will be FFh.

Addresses FE00h through FE0Fh are only in the program memory space for microcontroller and code protected microcontroller modes. A device programmer will be able to read the configuration word in any processor mode. See programming specifications for more detail.

TABLE 14-1: CONFIGURATION LOCATIONS

Bit	Address
FOSC0	FE00h
FOSC1	FE01h
WDTPS0	FE02h
WDTPS1	FE03h
PM0	FE04h
PM1	FE06h
PM2 ⁽¹⁾	FE0Fh ⁽¹⁾

Note 1: This location does not exist on the PIC17C42.

Note: When programming the desired configuration locations, they must be programmed in ascending order. Starting with address FE00h.

14.2 Oscillator Configurations

14.2.1 OSCILLATOR TYPES

The PIC17CXX can be operated in four different oscillator modes. The user can program two configuration bits (FOSC1:FOSC0) to select one of these four modes:

- LF: Low Power Crystal
- XT: Crystal/Resonator
- EC: External Clock Input
- RC: Resistor/Capacitor

14.2.2 CRYSTAL OSCILLATOR / CERAMIC RESONATORS

In XT or LF modes, a crystal or ceramic resonator is connected to the OSC1/CLKIN and OSC2/CLKOUT pins to establish oscillation (Figure 14-2). The PIC17CXX Oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturers specifications.

For frequencies above 20 MHz, it is common for the crystal to be an overtone mode crystal. Use of overtone mode crystals require a tank circuit to attenuate the gain at the fundamental frequency. Figure 14-3 shows an example of this.

FIGURE 14-2: CRYSTAL OR CERAMIC RESONATOR OPERATION (XT OR LF OSC CONFIGURATION)

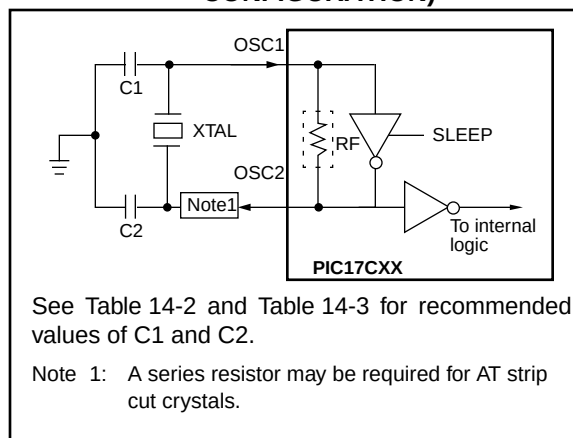


FIGURE 14-3: CRYSTAL OPERATION, OVERTONE CRYSTALS (XT OSC CONFIGURATION)

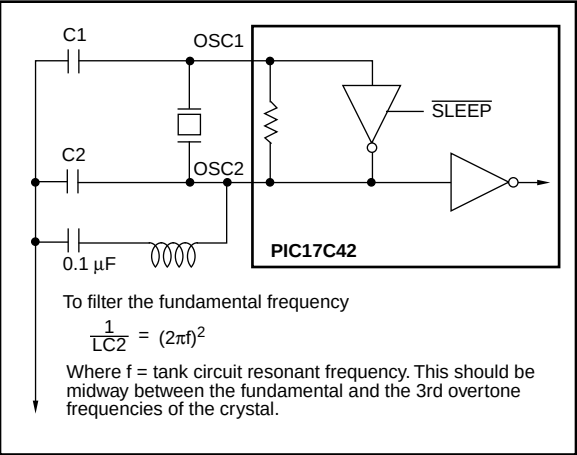


TABLE 14-2: CAPACITOR SELECTION FOR CERAMIC RESONATORS

Oscillator Type	Resonator Frequency	Capacitor Range C1 = C2
LF	455 kHz	15 - 68 pF
	2.0 MHz	10 - 33 pF
XT	4.0 MHz	22 - 68 pF
	8.0 MHz	33 - 100 pF
	16.0 MHz	33 - 100 pF

Higher capacitance increases the stability of the oscillator but also increases the start-up time. These values are for design guidance only. Since each resonator has its own characteristics, the user should consult the resonator manufacturer for appropriate values of external components.

Resonators Used:

455 kHz	Panasonic EFO-A455K04B	± 0.3%
2.0 MHz	Murata Erie CSA2.00MG	± 0.5%
4.0 MHz	Murata Erie CSA4.00MG	± 0.5%
8.0 MHz	Murata Erie CSA8.00MT	± 0.5%
16.0 MHz	Murata Erie CSA16.00MX	± 0.5%

Resonators used did not have built-in capacitors.

TABLE 14-3: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Osc Type	Freq	C1	C2
LF	32 kHz ⁽¹⁾	100-150 pF	100-150 pF
	1 MHz	10-33 pF	10-33 pF
	2 MHz	10-33 pF	10-33 pF
XT	2 MHz	47-100 pF	47-100 pF
	4 MHz	15-68 pF	15-68 pF
	8 MHz ⁽²⁾	15-47 pF	15-47 pF
	16 MHz	TBD	TBD
	25 MHz	15-47 pF	15-47 pF
	32 MHz ⁽³⁾	0 ⁽³⁾	0 ⁽³⁾

Higher capacitance increases the stability of the oscillator but also increases the start-up time and the oscillator current. These values are for design guidance only. Rs may be required in XT mode to avoid overdriving the crystals with low drive level specification. Since each crystal has its own characteristics, the user should consult the crystal manufacturer for appropriate values for external components.

Note 1: For VDD > 4.5V, C1 = C2 ≈ 30 pF is recommended.

2: Rs of 330Ω is required for a capacitor combination of 15/15 pF.

3: Only the capacitance of the board was present.

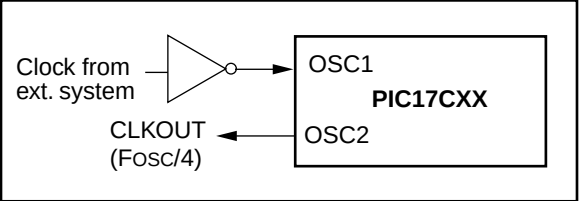
Crystals Used:

32.768 kHz	Epson C-001R32.768K-A	± 20 PPM
1.0 MHz	ECS-10-13-1	± 50 PPM
2.0 MHz	ECS-20-20-1	± 50 PPM
4.0 MHz	ECS-40-20-1	± 50 PPM
8.0 MHz	ECS ECS-80-S-4 ECS-80-18-1	± 50 PPM
16.0 MHz	ECS-160-20-1	TBD
25 MHz	CTS CTS25M	± 50 PPM
32 MHz	CRYSTEK HF-2	± 50 PPM

14.2.3 EXTERNAL CLOCK OSCILLATOR

In the EC oscillator mode, the OSC1 input can be driven by CMOS drivers. In this mode, the OSC1/CLKIN pin is hi-impedance and the OSC2/CLKOUT pin is the CLKOUT output (4 TOSC).

FIGURE 14-4: EXTERNAL CLOCK INPUT OPERATION (EC OSC CONFIGURATION)



14.4.2 MINIMIZING CURRENT CONSUMPTION

To minimize current consumption, all I/O pins should be either at V_{DD} , or V_{SS} , with no external circuitry drawing current from the I/O pin. I/O pins that are hi-impedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The $T0CKI$ input should be at V_{DD} or V_{SS} . The contributions from on-chip pull-ups on $PORTB$ should also be considered, and disabled when possible.

14.5 Code Protection

The code in the program memory can be protected by selecting the microcontroller in code protected mode ($PM2:PM0 = '000'$).

Note: PM2 does not exist on the PIC17C42. To select code protected microcontroller mode, $PM1:PM0 = '00'$.

In this mode, instructions that are in the on-chip program memory space, can continue to read or write the program memory. An instruction that is executed outside of the internal program memory range will be inhibited from writing to or reading from program memory.

Note: Microchip does not recommend code protecting windowed devices.

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

TABLE 15-2: PIC17CXX INSTRUCTION SET (Cont'd)

Mnemonic, Operands	Description	Cycles	16-bit Opcode				Status Affected	Notes
			MSb		LSb			
TABLWT t,i,f	Table Write	2	1010	11ti	ffff	ffff	None	5
TLRD t,f	Table Latch Read	1	1010	00tx	ffff	ffff	None	
TLWT t,f	Table Latch Write	1	1010	01tx	ffff	ffff	None	
TSTFSZ f	Test f, skip if 0	1 (2)	0011	0011	ffff	ffff	None	6,8
XORWF f,d	Exclusive OR WREG with f	1	0000	110d	ffff	ffff	Z	
BIT-ORIENTED FILE REGISTER OPERATIONS								
BCF f,b	Bit Clear f	1	1000	1bbb	ffff	ffff	None	
BSF f,b	Bit Set f	1	1000	0bbb	ffff	ffff	None	
BTFSC f,b	Bit test, skip if clear	1 (2)	1001	1bbb	ffff	ffff	None	6,8
BTFSS f,b	Bit test, skip if set	1 (2)	1001	0bbb	ffff	ffff	None	6,8
BTG f,b	Bit Toggle f	1	0011	1bbb	ffff	ffff	None	
LITERAL AND CONTROL OPERATIONS								
ADDLW k	ADD literal to WREG	1	1011	0001	kkkk	kkkk	OV,C,DC,Z	
ANDLW k	AND literal with WREG	1	1011	0101	kkkk	kkkk	Z	
CALL k	Subroutine Call	2	111k	kkkk	kkkk	kkkk	None	7
CLRWDT —	Clear Watchdog Timer	1	0000	0000	0000	0100	$\overline{TO}, \overline{PD}$	
GOTO k	Unconditional Branch	2	110k	kkkk	kkkk	kkkk	None	7
IORLW k	Inclusive OR literal with WREG	1	1011	0011	kkkk	kkkk	Z	
LCALL k	Long Call	2	1011	0111	kkkk	kkkk	None	4,7
MOVLB k	Move literal to low nibble in BSR	1	1011	1000	uuuu	kkkk	None	
MOVLR k	Move literal to high nibble in BSR	1	1011	101x	kkkk	uuuu	None	9
MOVLW k	Move literal to WREG	1	1011	0000	kkkk	kkkk	None	
MULLW k	Multiply literal with WREG	1	1011	1100	kkkk	kkkk	None	9
RETFIE —	Return from interrupt (and enable interrupts)	2	0000	0000	0000	0101	GLINTD	7
RETLW k	Return literal to WREG	2	1011	0110	kkkk	kkkk	None	7
RETURN —	Return from subroutine	2	0000	0000	0000	0010	None	7
SLEEP —	Enter SLEEP Mode	1	0000	0000	0000	0011	$\overline{TO}, \overline{PD}$	
SUBLW k	Subtract WREG from literal	1	1011	0010	kkkk	kkkk	OV,C,DC,Z	
XORLW k	Exclusive OR literal with WREG	1	1011	0100	kkkk	kkkk	Z	

Legend: Refer to Table 15-1 for opcode field descriptions.

Note 1: 2's Complement method.

2: Unsigned arithmetic.

3: If s = '1', only the file is affected; If s = '0', both the WREG register and the file are affected; If only the Working register (WREG) is required to be affected, then f = WREG must be specified.

4: During an LCALL, the contents of PCLATH are loaded into the MSB of the PC and kkkk kkkk is loaded into the LSB of the PC (PCL)

5: Multiple cycle instruction for EPROM programming when table pointer selects internal EPROM. The instruction is terminated by an interrupt event. When writing to external program memory, it is a two-cycle instruction.

6: Two-cycle instruction when condition is true, else single cycle instruction.

7: Two-cycle instruction except for TABLRD to PCL (program counter low byte) in which case it takes 3 cycles.

8: A "skip" means that instruction fetched during execution of current instruction is not executed, instead an NOP is executed.

9: These instructions are not available on the PIC17C42.

IORWF		Inclusive OR WREG with f						
Syntax:	[<i>label</i>] IORWF f,d							
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$							
Operation:	(WREG) .OR. (f) $\rightarrow$ (dest)							
Status Affected:	Z							
Encoding:	<table border="1"><tr><td>0000</td><td>100d</td><td>ffff</td><td>ffff</td></tr></table>				0000	100d	ffff	ffff
0000	100d	ffff	ffff					
Description:	Inclusive OR WREG with register 'f'. If 'd' is 0 the result is placed in WREG. If 'd' is 1 the result is placed back in register 'f'.							
Words:	1							
Cycles:	1							
Q Cycle Activity:								
	Q1	Q2	Q3	Q4				
	Decode	Read register 'f'	Execute	Write to destination				

Example: IORWF RESULT, 0

Before Instruction

RESULT = 0x13
WREG = 0x91

After Instruction

RESULT = 0x13
WREG = 0x93

LCALL		Long Call						
Syntax:	[<i>label</i>] LCALL k							
Operands:	$0 \leq k \leq 255$							
Operation:	PC + 1 $\rightarrow$ TOS; k $\rightarrow$ PCL, (PCLATH) $\rightarrow$ PCH							
Status Affected:	None							
Encoding:	<table border="1"><tr><td>1011</td><td>0111</td><td>kkkk</td><td>kkkk</td></tr></table>				1011	0111	kkkk	kkkk
1011	0111	kkkk	kkkk					
Description:	LCALL allows an unconditional subroutine call to anywhere within the 64k program memory space. First, the return address (PC + 1) is pushed onto the stack. A 16-bit destination address is then loaded into the program counter. The lower 8-bits of the destination address is embedded in the instruction. The upper 8-bits of PC is loaded from PC high holding latch, PCLATH.							
Words:	1							
Cycles:	2							
Q Cycle Activity:								
Q1		Q2		Q3		Q4		
Decode		Read literal 'k'		Execute		Write register PCL		
Forced NOP		NOP		Execute		NOP		

Example: MOVLW HIGH(SUBROUTINE)
MOVWF WREG, PCLATH
LCALL LOW(SUBROUTINE)

Before Instruction

SUBROUTINE = 16-bit Address
PC = ?

After Instruction

PC = Address (SUBROUTINE)

17.1 DC CHARACTERISTICS: PIC17C42-16 (Commercial, Industrial) PIC17C42-25 (Commercial, Industrial)

Standard Operating Conditions (unless otherwise stated)							
DC CHARACTERISTICS							
Operating temperature							
-40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial							
Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001	VDD	Supply Voltage	4.5	–	5.5	V	
D002	VDR	RAM Data Retention Voltage (Note 1)	1.5 *	–	–	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure internal Power-on Reset signal	–	VSS	–	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure internal Power-on Reset signal	0.060*	–	–	mV/ms	See section on Power-on Reset for details
D010 D011 D012 D013 D014	IDD	Supply Current (Note 2)	–	3 6 11 19 95	6 12 * 24 * 38 150	mA mA mA mA μA	FOSC = 4 MHz (Note 4) FOSC = 8 MHz FOSC = 16 MHz FOSC = 25 MHz FOSC = 32 kHz WDT enabled (EC osc configuration)
D020 D021	IPD	Power-down Current (Note 3)	–	10 < 1	40 5	μA μA	VDD = 5.5V, WDT enabled VDD = 5.5V, WDT disabled

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to VDD or VSS, T0CKI = VDD, MCLR = VDD; WDT enabled/disabled as specified.

Current consumed from the oscillator and I/O's driving external capacitive or resistive loads need to be considered.

For the RC oscillator, the current through the external pull-up resistor (R) can be estimated as: $V_{DD} / (2 \cdot R)$.

For capacitive loads, The current can be estimated (for an individual I/O pin) as $(C_L \cdot V_{DD}) \cdot f$

CL = Total capacitive load on the I/O pin; f = average frequency on the I/O pin switches.

The capacitive currents are most significant when the device is configured for external execution (includes extended microcontroller mode).

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula $I_R = V_{DD}/2R_{ext}$ (mA) with Rext in kOhm.

PIC17C4X

Applicable Devices 42 R42 42A 43 R43 44

FIGURE 19-11: MEMORY INTERFACE WRITE TIMING (NOT SUPPORTED IN PIC17LC4X DEVICES)

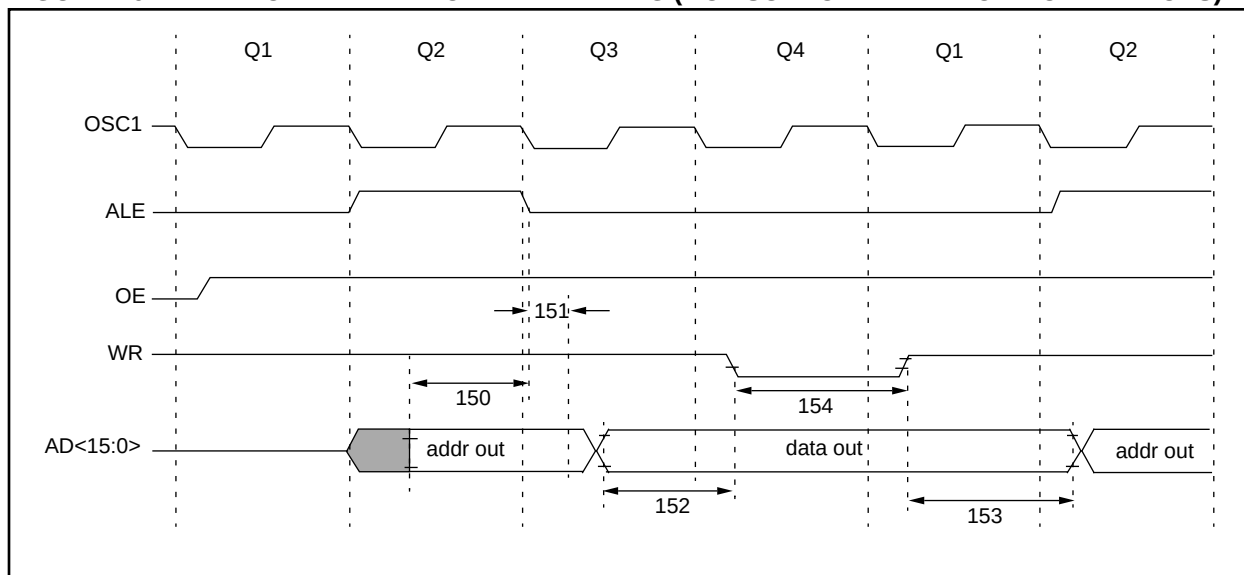


TABLE 19-11: MEMORY INTERFACE WRITE REQUIREMENTS (NOT SUPPORTED IN PIC17LC4X DEVICES)

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
150	TadV2aLL	AD<15:0> (address) valid to ALE↓ (address setup time)	0.25Tcy - 10	—	—	ns	
151	TaLL2adI	ALE↓ to address out invalid (address hold time)	0	—	—	ns	
152	TadV2wrL	Data out valid to WR↓ (data setup time)	0.25Tcy - 40	—	—	ns	
153	TwrH2adI	WR↑ to data out invalid (data hold time)	—	0.25Tcy §	—	ns	
154	TwrL	WR pulse width	—	0.25Tcy §	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification ensured by design.

PIC17C4X

Applicable Devices 42 R42 42A 43 R43 44

FIGURE 20-2: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

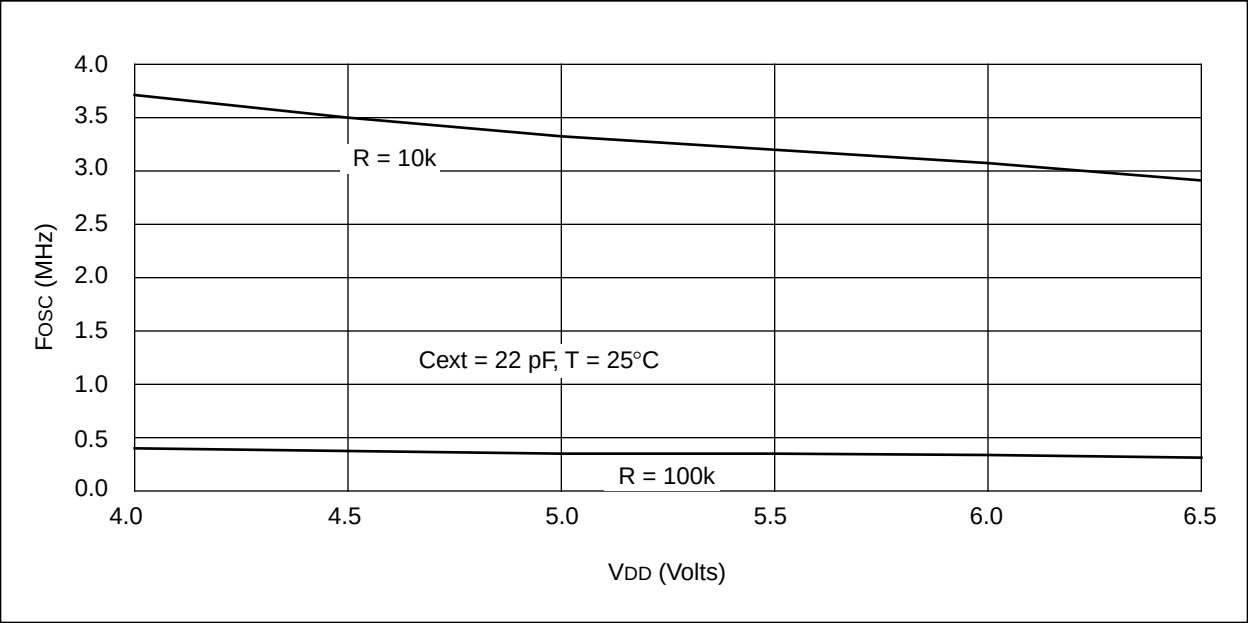
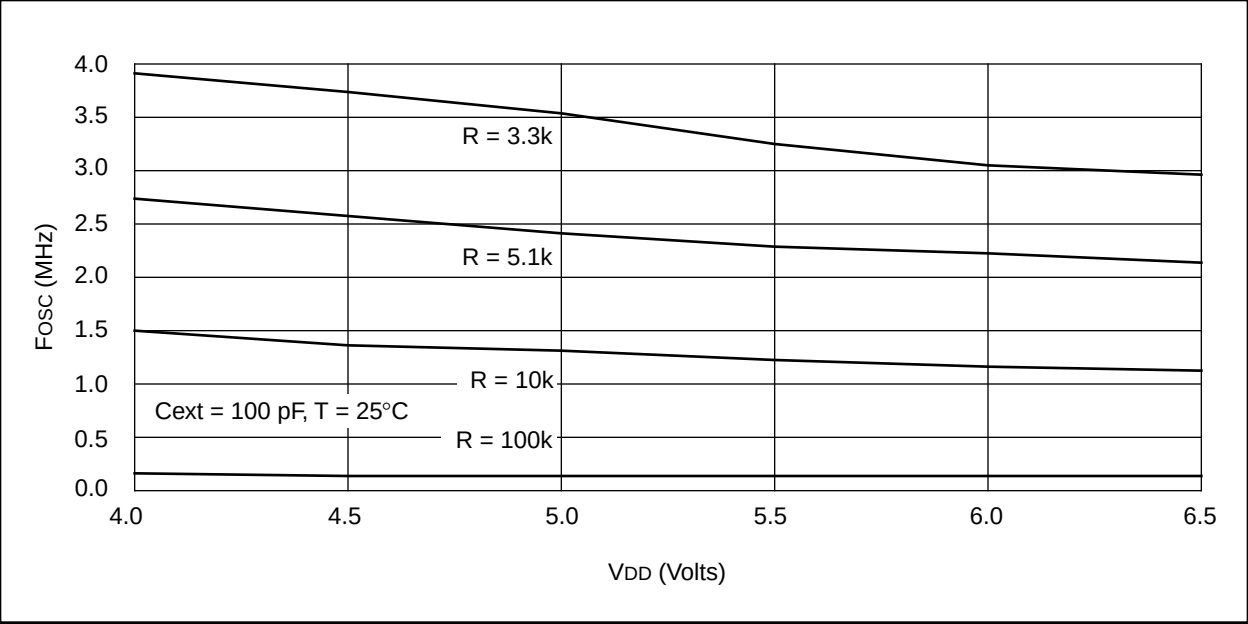


FIGURE 20-3: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

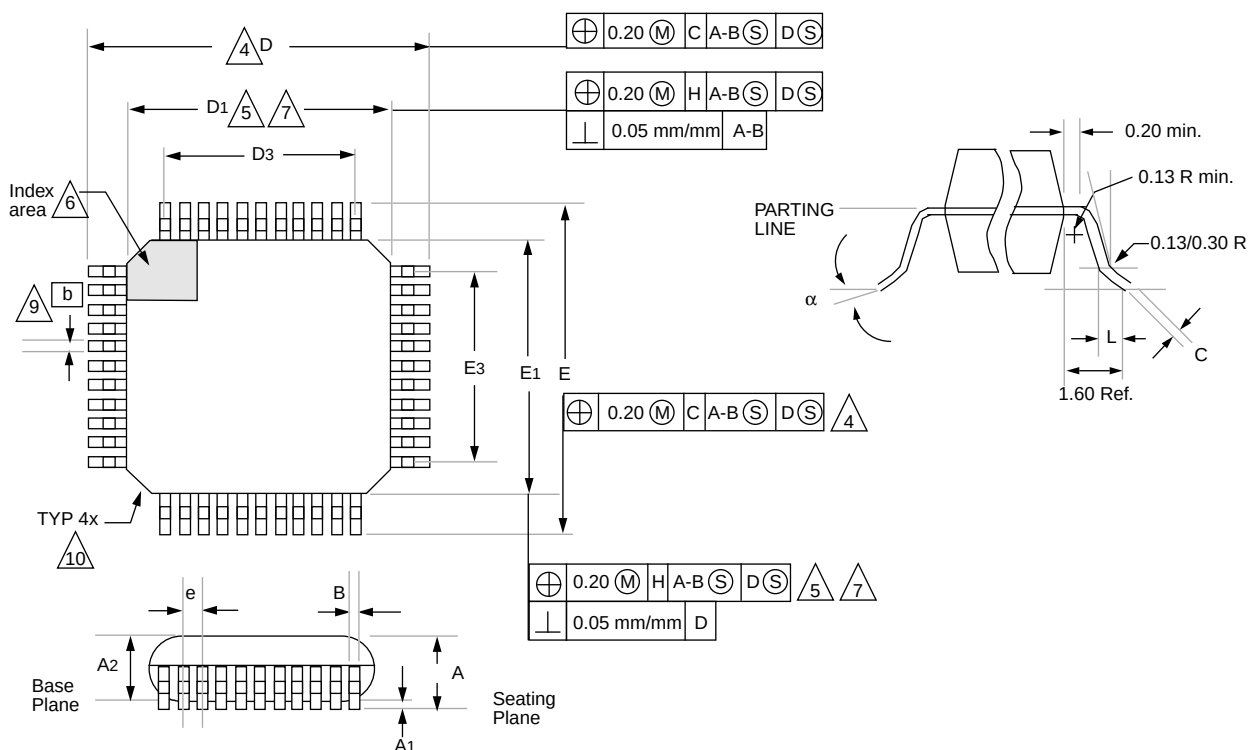


PIC17C4X

NOTES:

PIC17C4X

21.4 44-Lead Plastic Surface Mount (MQFP 10x10 mm Body 1.6/0.15 mm Lead Form)



Package Group: Plastic MQFP						
Symbol	Millimeters			Inches		
	Min	Max	Notes	Min	Max	Notes
α	0°	7°		0°	7°	
A	2.000	2.350		0.078	0.093	
A1	0.050	0.250		0.002	0.010	
A2	1.950	2.100		0.768	0.083	
b	0.300	0.450	Typical	0.011	0.018	Typical
C	0.150	0.180		0.006	0.007	
D	12.950	13.450		0.510	0.530	
D1	9.900	10.100		0.390	0.398	
D3	8.000	8.000	Reference	0.315	0.315	Reference
E	12.950	13.450		0.510	0.530	
E1	9.900	10.100		0.390	0.398	
E3	8.000	8.000	Reference	0.315	0.315	Reference
e	0.800	0.800		0.031	0.032	
L	0.730	1.030		0.028	0.041	
N	44	44		44	44	
CP	0.102	—		0.004	—	

APPENDIX E: PIC16/17 MICROCONTROLLERS

E.1 PIC14000 Devices

	Clock		Memory		Peripherals				Features			
	Maximum Frequency of Operation (MHz)	EPROM Program Memory (K14 words)	Data Memory (bytes)	Timer Module(s)	Serial Ports (SPI/I ² C, USART)	Slope A/D Converter (high-res) Channels	Interrupt Sources	I/O Pins	Voltage Range (Volts)	In-Circuit Serial Programming	Additional On-chip Features	Packages
PIC14000	20	4K	192	TMR0 ADTMR	I ² C/ SMBus	14	11	22	2.7-6.0	Yes	Internal Oscillator, Bandgap Reference, Temperature Sensor, Calibration Factors, Low Voltage Detector, SLEEP, HIBERNATE, Comparators with Programmable References (2)	28-pin DIP, SOIC, SSOP (.300 mil)

PIC17C4X

E.6 PIC16C8X Family of Devices

	Clock			Memory			Peripherals		Features		
	Maximum Frequency of Operation (MHz)			Program Memory			Data EEPROM (bytes)			Timer Modules	
PIC16C84	10	—	1K	—	36	64	TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16F84 ⁽¹⁾	10	1K	—	—	68	64	TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16CR84 ⁽¹⁾	10	—	—	1K	68	64	TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16F83 ⁽¹⁾	10	512	—	—	36	64	TMR0	4	13	2.0-6.0	18-pin DIP, SOIC
PIC16CR83 ⁽¹⁾	10	—	—	512	36	64	TMR0	4	13	2.0-6.0	18-pin DIP, SOIC

All PIC16/17 family devices have Power-on Reset, selectable Watchdog Timer, selectable code protect, and high I/O current capability.

All PIC16C8X family devices use serial programming with clock pin RB6 and data pin RB7.

Note 1: Please contact your local sales office for availability of these devices.

APPENDIX F: ERRATA FOR PIC17C42 SILICON

The PIC17C42 devices that you have received have the following anomalies. At present there is no intention for future revisions to the present PIC17C42 silicon. If these cause issues for the application, it is recommended that you select the PIC17C42A device.

Note: New designs should use the PIC17C42A.

Design considerations

The device must not be operated outside of the specified voltage range. An external reset circuit must be used to ensure the device is in reset when a brown-out occurs or the V_{DD} rise time is too long. Failure to ensure that the device is in reset when device voltage is out of specification may cause the device to lock-up and ignore the MCLR pin.

1. When the Oscillator Start-Up Timer (OST) is enabled (in LF or XT oscillator modes), any interrupt that wakes the processor may cause a WDT reset. This occurs when the WDT is greater than or equal to 50% time-out period when the SLEEP instruction is executed. This will not occur in either the EC or RC oscillator modes.

Work-arounds

- a) Always ensure that the CLRWDT instruction is executed before the WDT increments past 50% of the WDT period. This will keep the "false" WDT reset from occurring.
- b) When using the WDT as a normal timer (WDT disabled), ensure that the WDT is less than or equal to 50% time-out period when the SLEEP instruction is executed. This can be done by monitoring the $\overline{TO}$ bit for changing state from set to clear. Example 1 shows putting the PIC17C42 to sleep.

EXAMPLE F-1: PIC17C42 TO SLEEP

```

      BTFSS    CPUSTA, TO    ; TO = 0?
      CLRWDT                      ; YES, WDT = 0
LOOP   BTFSC    CPUSTA, TO    ; WDT rollover?
      GOTO     LOOP          ; NO, wait
      SLEEP                      ; YES, goto Sleep
  
```

2. When the clock source of Timer1 or Timer2 is selected to external clock, the overflow interrupt flag will be set twice, once when the timer equals the period, and again when the timer value is reset to 0h. If the latency to clear TMRxIF is greater than the time to the next clock pulse, no problems will be noticed. If the latency is less than the time to the next timer clock pulse, the interrupt will be serviced twice.

Work-arounds

- a) Ensure that the timer has rolled over to 0h before clearing the flag bit.
- b) Clear the timer in software. Clearing the timer in software causes the period to be one count less than expected.

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New York

150 Motor Parkway, Suite 202
Hauppauge, NY 11788
Tel: 631-273-5305 Fax: 631-273-5335

San Jose

Microchip Technology Inc.
2107 North First Street, Suite 590
San Jose, CA 95131
Tel: 408-436-7950 Fax: 408-436-7955

Toronto

6285 Northam Drive, Suite 108
Mississauga, Ontario L4V 1X5, Canada
Tel: 905-673-0699 Fax: 905-673-6509

ASIA/PACIFIC

Australia

Microchip Technology Australia Pty Ltd
Suite 22, 41 Rawson Street
Epping 2121, NSW
Australia
Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing

Microchip Technology Consulting (Shanghai)
Co., Ltd., Beijing Liaison Office
Unit 915
Bei Hai Wan Tai Bldg.
No. 6 Chaoyangmen Beidajie
Beijing, 100027, No. China
Tel: 86-10-85282100 Fax: 86-10-85282104

China - Chengdu

Microchip Technology Consulting (Shanghai)
Co., Ltd., Chengdu Liaison Office
Rm. 2401, 24th Floor,
Ming Xing Financial Tower
No. 88 TIDU Street
Chengdu 610016, China
Tel: 86-28-6766200 Fax: 86-28-6766599

China - Fuzhou

Microchip Technology Consulting (Shanghai)
Co., Ltd., Fuzhou Liaison Office
Unit 28F, World Trade Plaza
No. 71 Wusi Road
Fuzhou 350001, China
Tel: 86-591-7503506 Fax: 86-591-7503521

China - Shanghai

Microchip Technology Consulting (Shanghai)
Co., Ltd.
Room 701, Bldg. B
Far East International Plaza
No. 317 Xian Xia Road
Shanghai, 200051
Tel: 86-21-6275-5700 Fax: 86-21-6275-5060

China - Shenzhen

Microchip Technology Consulting (Shanghai)
Co., Ltd., Shenzhen Liaison Office
Rm. 1315, 13/F, Shenzhen Kerry Centre,
Renminnan Lu
Shenzhen 518001, China
Tel: 86-755-2350361 Fax: 86-755-2366086

Hong Kong

Microchip Technology Hongkong Ltd.
Unit 901-6, Tower 2, Metroplaza
223 Hing Fong Road
Kwai Fong, N.T., Hong Kong
Tel: 852-2401-1200 Fax: 852-2401-3431

India

Microchip Technology Inc.
India Liaison Office
Divyasree Chambers
1 Floor, Wing A (A3/A4)
No. 11, O'Shaugnessey Road
Bangalore, 560 025, India
Tel: 91-80-2290061 Fax: 91-80-2290062

Japan

Microchip Technology Japan K.K.
Benex S-1 6F
3-18-20, Shinyokohama
Kohoku-Ku, Yokohama-shi
Kanagawa, 222-0033, Japan
Tel: 81-45-471- 6166 Fax: 81-45-471-6122

Korea

Microchip Technology Korea
168-1, Youngbo Bldg. 3 Floor
Samsung-Dong, Kangnam-Ku
Seoul, Korea 135-882
Tel: 82-2-554-7200 Fax: 82-2-558-5934

Singapore

Microchip Technology Singapore Pte Ltd.
200 Middle Road
#07-02 Prime Centre
Singapore, 188980
Tel: 65-334-8870 Fax: 65-334-8850

Taiwan

Microchip Technology Taiwan
11F-3, No. 207
Tung Hua North Road
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Tel: 886-2-2717-7175 Fax: 886-2-2545-0139

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Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

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Microchip Technology GmbH
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D-81739 Munich, Germany
Tel: 49-89-627-144 0 Fax: 49-89-627-144-44

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Microchip Technology SRL
Centro Direzionale Colleoni
Palazzo Taurus 1 V. Le Colleoni 1
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Milan, Italy
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United Kingdom

Arizona Microchip Technology Ltd.
505 Eskdale Road
Winnersh Triangle
Wokingham
Berkshire, England RG41 5TU
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