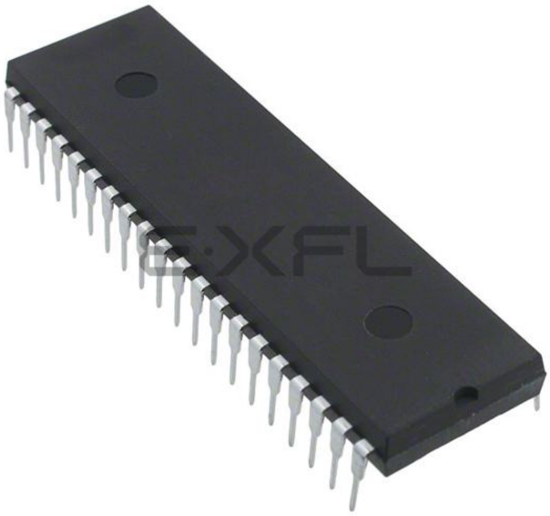

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Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	25MHz
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	33
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	454 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Through Hole
Package / Case	40-DIP (0.600", 15.24mm)
Supplier Device Package	40-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c44-25-p

3.0 ARCHITECTURAL OVERVIEW

The high performance of the PIC17C4X can be attributed to a number of architectural features commonly found in RISC microprocessors. To begin with, the PIC17C4X uses a modified Harvard architecture. This architecture has the program and data accessed from separate memories. So the device has a program memory bus and a data memory bus. This improves bandwidth over traditional von Neumann architecture, where program and data are fetched from the same memory (accesses over the same bus). Separating program and data memory further allows instructions to be sized differently than the 8-bit wide data word. PIC17C4X opcodes are 16-bits wide, enabling single word instructions. The full 16-bit wide program memory bus fetches a 16-bit instruction in a single cycle. A two-stage pipeline overlaps fetch and execution of instructions. Consequently, all instructions execute in a single cycle (121 ns @ 33 MHz), except for program branches and two special instructions that transfer data between program and data memory.

The PIC17C4X can address up to 64K x 16 of program memory space.

The **PIC17C42** and **PIC17C42A** integrate 2K x 16 of EPROM program memory on-chip, while the **PIC17CR42** has 2K x 16 of ROM program memory on-chip.

The **PIC17C43** integrates 4K x 16 of EPROM program memory, while the **PIC17CR43** has 4K x 16 of ROM program memory.

The **PIC17C44** integrates 8K x 16 EPROM program memory.

Program execution can be internal only (microcontroller or protected microcontroller mode), external only (microprocessor mode) or both (extended microcontroller mode). Extended microcontroller mode does not allow code protection.

The PIC17CXX can directly or indirectly address its register files or data memory. All special function registers, including the Program Counter (PC) and Working Register (WREG), are mapped in the data memory. The PIC17CXX has an orthogonal (symmetrical) instruction set that makes it possible to carry out any operation on any register using any addressing mode. This symmetrical nature and lack of 'special optimal situations' make programming with the PIC17CXX simple yet efficient. In addition, the learning curve is reduced significantly.

One of the PIC17CXX family architectural enhancements from the PIC16CXX family allows two file registers to be used in some two operand instructions. This allows data to be moved directly between two registers without going through the WREG register. This increases performance and decreases program memory usage.

The PIC17CXX devices contain an 8-bit ALU and working register. The ALU is a general purpose arithmetic unit. It performs arithmetic and Boolean functions between data in the working register and any register file.

The ALU is 8-bits wide and capable of addition, subtraction, shift, and logical operations. Unless otherwise mentioned, arithmetic operations are two's complement in nature.

The WREG register is an 8-bit working register used for ALU operations.

All PIC17C4X devices (except the PIC17C42) have an 8 x 8 hardware multiplier. This multiplier generates a 16-bit result in a single cycle.

Depending on the instruction executed, the ALU may affect the values of the Carry (C), Digit Carry (DC), and Zero (Z) bits in the STATUS register. The C and DC bits operate as a $\overline{\text{borrow}}$ and $\overline{\text{digit borrow}}$ out bit, respectively, in subtraction. See the `SUBLW` and `SUBWF` instructions for examples.

Although the ALU does not perform signed arithmetic, the Overflow bit (OV) can be used to implement signed math. Signed arithmetic is comprised of a magnitude and a sign bit. The overflow bit indicates if the magnitude overflows and causes the sign bit to change state. Signed math can have greater than 7-bit values (magnitude), if more than one byte is used. The use of the overflow bit only operates on bit6 (MSb of magnitude) and bit7 (sign bit) of the value in the ALU. That is, the overflow bit is not useful if trying to implement signed math where the magnitude, for example, is 11-bits. If the signed math values are greater than 7-bits (15-, 24- or 31-bit), the algorithm must ensure that the low order bytes ignore the overflow status bit.

Care should be taken when adding and subtracting signed numbers to ensure that the correct operation is executed. Example 3-1 shows an item that must be taken into account when doing signed arithmetic on an ALU which operates as an unsigned machine.

EXAMPLE 3-1: SIGNED MATH

```
Hex Value      Signed Value    Unsigned Value
               Math            Math

  FFh            -127            255
+ 01h          +    1          +   1
=  ?           = -126 (FEh)    =   0 (00h);
                               Carry bit = 1

Signed math requires the result in REG to
be FEh (-126). This would be accomplished
by subtracting one as opposed to adding
one.
```

Simplified block diagrams are shown in Figure 3-1 and Figure 3-2. The descriptions of the device pins are listed in Table 3-1.

TABLE 4-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (Cont.'d)

Register	Address	Power-on Reset	$\overline{MCLR}$ Reset WDT Reset	Wake-up from SLEEP through interrupt
Bank 2				
TMR1	10h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR2	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3L	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3H	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR1	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR2	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1L	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1H	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 3				
PW1DCL	10h	xx-- ----	uu-- ----	uu-- ----
PW2DCL	11h	xx-- ----	uu-- ----	uu-- ----
PW1DCH	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PW2DCH	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2L	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2H	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TCON1	16h	0000 0000	0000 0000	uuuu uuuu
TCON2	17h	0000 0000	0000 0000	uuuu uuuu
Unbanked				
PRODL (5)	18h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PRODH (5)	19h	xxxx xxxx	uuuu uuuu	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented read as '0', q = value depends on condition.

Note 1: One or more bits in INTSTA, PIR will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 4-3 for reset value of specific condition.

4: Only applies to the PIC17C42.

5: Does not apply to the PIC17C42.

9.3 PORTC and DDRC Registers

PORTC is an 8-bit bi-directional port. The corresponding data direction register is DDRC. A '1' in DDRC configures the corresponding port pin as an input. A '0' in the DDRC register configures the corresponding port pin as an output. Reading PORTC reads the status of the pins, whereas writing to it will write to the port latch. PORTC is multiplexed with the system bus. When operating as the system bus, PORTC is the low order byte of the address/data bus (AD7:AD0). The timing for the system bus is shown in the Electrical Characteristics section.

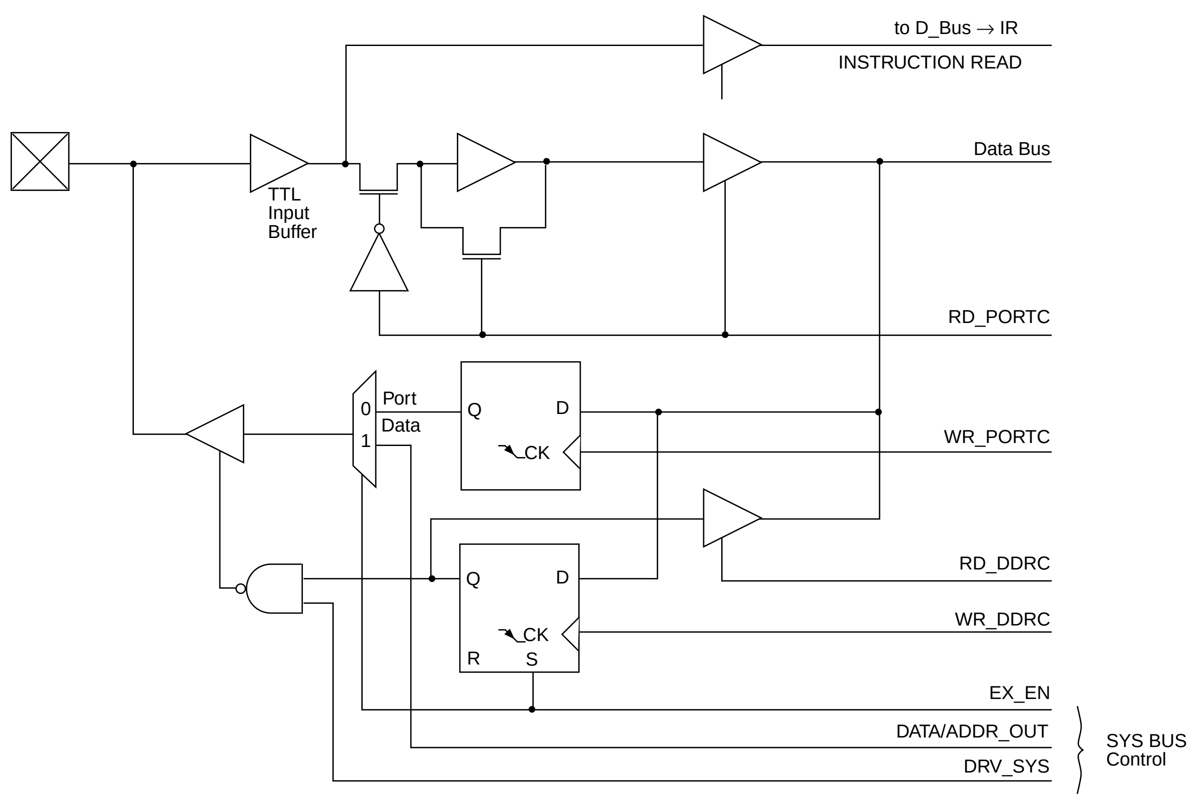
Note:	This port is configured as the system bus when the device's configuration bits are selected to Microprocessor or Extended Microcontroller modes. In the two other microcontroller modes, this port is a general purpose I/O.

Example 9-2 shows the instruction sequence to initialize PORTC. The Bank Select Register (BSR) must be selected to Bank 1 for the port to be initialized.

EXAMPLE 9-2: INITIALIZING PORTC

```
MOVLB  1            ; Select Bank 1
CLRF   PORTC        ; Initialize PORTC data
                    ;  latches before setting
                    ;  the data direction
                    ;  register
MOVLW  0xCF         ; Value used to initialize
                    ;  data direction
MOVWF  DDRC         ; Set RC<3:0> as inputs
                    ;  RC<5:4> as outputs
                    ;  RC<7:6> as inputs
```

FIGURE 9-6: BLOCK DIAGRAM OF RC<7:0> PORT PINS



Note: I/O pins have protection diodes to VDD and Vss.

9.4 PORTD and DDRD Registers

PORTD is an 8-bit bi-directional port. The corresponding data direction register is DDRD. A '1' in DDRD configures the corresponding port pin as an input. A '0' in the DDRC register configures the corresponding port pin as an output. Reading PORTD reads the status of the pins, whereas writing to it will write to the port latch. PORTD is multiplexed with the system bus. When operating as the system bus, PORTD is the high order byte of the address/data bus (AD15:AD8). The timing for the system bus is shown in the Electrical Characteristics section.

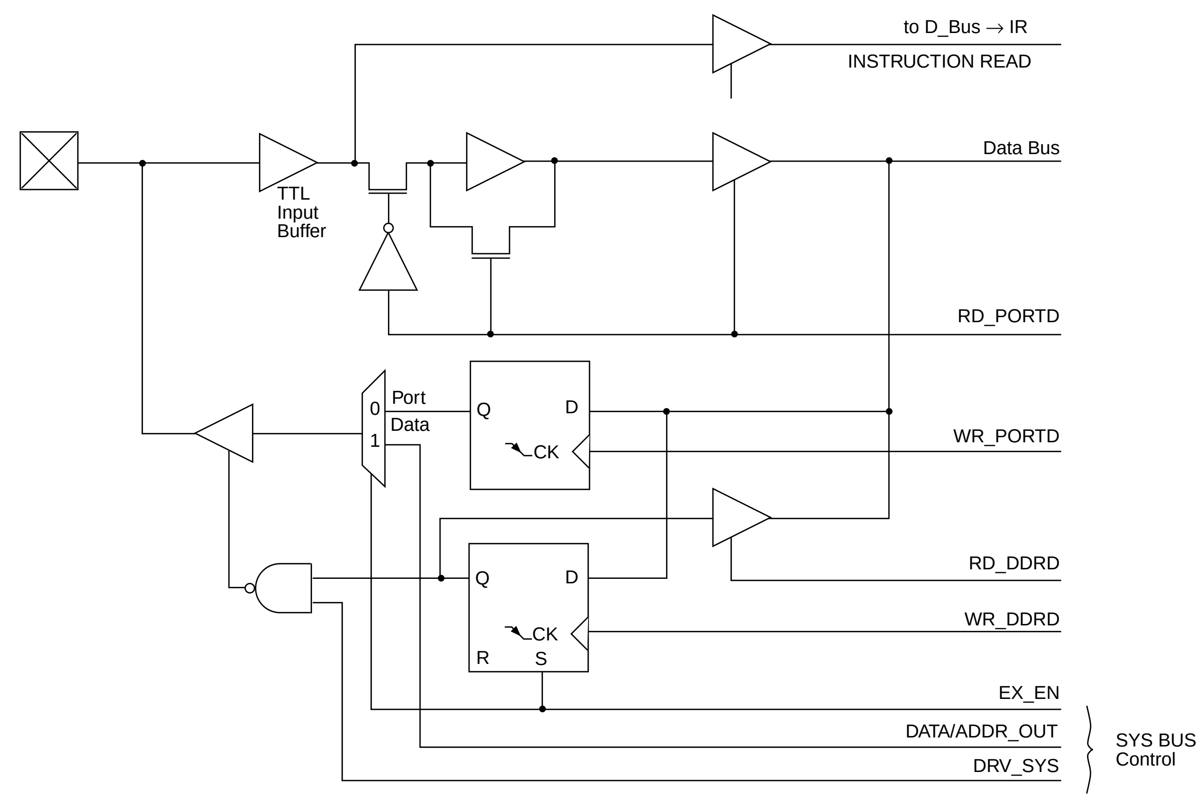
Note:	This port is configured as the system bus when the device's configuration bits are selected to Microprocessor or Extended Microcontroller modes. In the two other microcontroller modes, this port is a general purpose I/O.

Example 9-3 shows the instruction sequence to initialize PORTD. The Bank Select Register (BSR) must be selected to Bank 1 for the port to be initialized.

EXAMPLE 9-3: INITIALIZING PORTD

```
MOVLB  1            ; Select Bank 1
CLRF   PORTD        ; Initialize PORTD data
                    ;   latches before setting
                    ;   the data direction
                    ;   register
MOVLW  0xCF         ; Value used to initialize
                    ;  data direction
MOVWF  DDRD         ; Set RD<3:0> as inputs
                    ;   RD<5:4> as outputs
                    ;   RD<7:6> as inputs
```

FIGURE 9-7: PORTD BLOCK DIAGRAM (IN I/O PORT MODE)



Note: I/O pins have protection diodes to VDD and VSS.

TABLE 9-9: PORTE FUNCTIONS

Name	Bit	Buffer Type	Function
RE0/ALE	bit0	TTL	Input/Output or system bus Address Latch Enable (ALE) control pin.
RE1/$\overline{\text{OE}}$	bit1	TTL	Input/Output or system bus Output Enable ($\overline{\text{OE}}$) control pin.
RE2/$\overline{\text{WR}}$	bit2	TTL	Input/Output or system bus Write ($\overline{\text{WR}}$) control pin.

Legend: TTL = TTL input.

TABLE 9-10: REGISTERS/BITS ASSOCIATED WITH PORTE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
15h, Bank 1	PORTE	—	—	—	—	—	RE2/$\overline{\text{WR}}$	RE1/$\overline{\text{OE}}$	RE0/ALE	---- -xxx	---- -uuu
14h, Bank 1	DDRE	Data direction register for PORTE								---- -111	---- -111

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used by PORTE.
Note 1: Other (non power-up) resets include: external reset through $\overline{\text{MCLR}}$ and the Watchdog Timer Reset.

11.3 Read/Write Consideration for TMR0

Although TMR0 is a 16-bit timer/counter, only 8-bits at a time can be read or written during a single instruction cycle. Care must be taken during any read or write.

11.3.1 READING 16-BIT VALUE

The problem in reading the entire 16-bit value is that after reading the low (or high) byte, its value may change from FFh to 00h.

Example 11-1 shows a 16-bit read. To ensure a proper read, interrupts must be disabled during this routine.

EXAMPLE 11-1: 16-BIT READ

```
MOVPF   TMR0L, TMPLO    ;read low tmr0
MOVPF   TMR0H, TMPHI    ;read high tmr0
MOVFP   TMPLO, WREG     ;tmplo —> wreg
CPFSLT  TMR0L           ;tmr0l < wreg?
RETURN                  ;no then return
MOVPF   TMR0L, TMPLO    ;read low tmr0
MOVPF   TMR0H, TMPHI    ;read high tmr0
RETURN                  ;return
```

11.3.2 WRITING A 16-BIT VALUE TO TMR0

Since writing to either TMR0L or TMR0H will effectively inhibit increment of that half of the TMR0 in the next cycle (following write), but not inhibit increment of the other half, the user must write to TMR0L first and TMR0H next in two consecutive instructions, as shown in Example 11-2. The interrupt must be disabled. Any write to either TMR0L or TMR0H clears the prescaler.

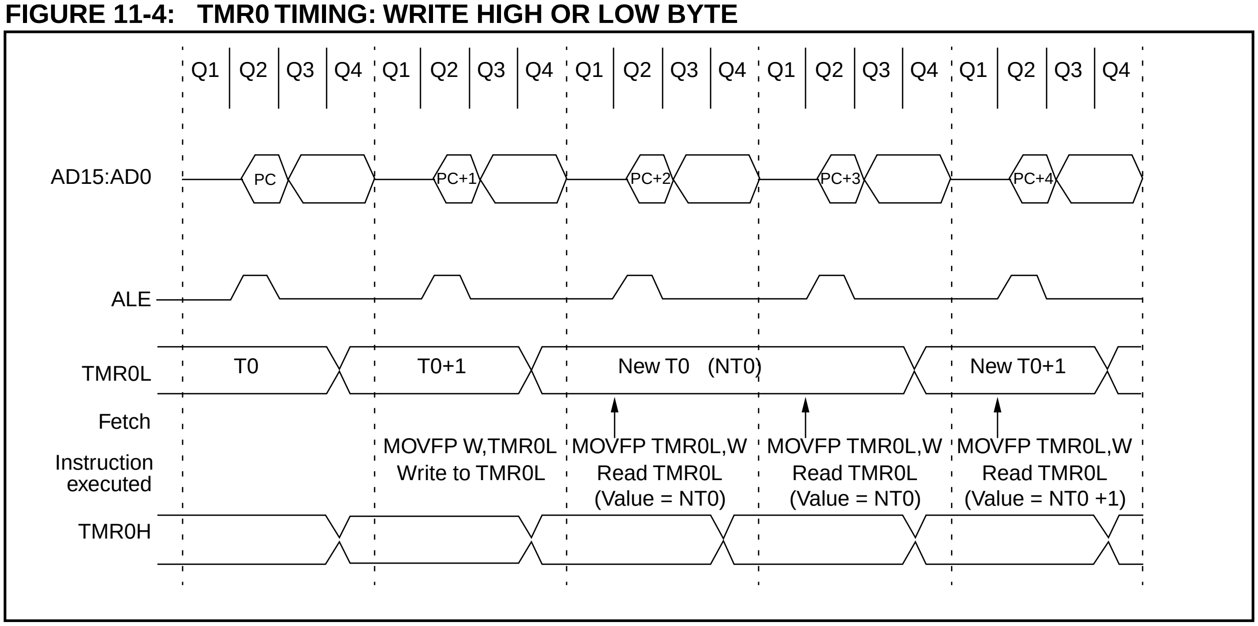
EXAMPLE 11-2: 16-BIT WRITE

```
BSF   CPUSTA, GLINTD ; Disable interrupt
MOVFP RAM_L, TMR0L   ;
MOVFP RAM_H, TMR0H   ;
BCF   CPUSTA, GLINTD ; Done, enable interrupt
```

11.4 Prescaler Assignments

Timer0 has an 8-bit prescaler. The prescaler assignment is fully under software control; i.e., it can be changed “on the fly” during program execution. When changing the prescaler assignment, clearing the prescaler is recommended before changing assignment. The value of the prescaler is “unknown,” and assigning a value that is less then the present value makes it difficult to take this unknown time into account.

FIGURE 11-4: TMR0 TIMING: WRITE HIGH OR LOW BYTE



12.1.2 TIMER1 & TIMER2 IN 16-BIT MODE

To select 16-bit mode, the T16 bit must be set. In this mode TMR1 and TMR2 are concatenated to form a 16-bit timer (TMR2:TMR1). The 16-bit timer increments until it matches the 16-bit period register (PR2:PR1). On the following timer clock, the timer value is reset to 0h, and the TMR1IF bit is set.

When selecting the clock source for the16-bit timer, the TMR1CS bit controls the entire 16-bit timer and TMR2CS is a “don’t care.” When TMR1CS is clear, the timer increments once every instruction cycle (Fosc/4). When TMR1CS is set, the timer increments on every falling edge of the RB4/TCLK12 pin. For the 16-bit timer to increment, both TMR1ON and TMR2ON bits must be set (Table 12-1).

12.1.2.1 EXTERNAL CLOCK INPUT FOR TMR1:TMR2

When TMR1CS is set, the 16-bit TMR2:TMR1 increments on the falling edge of clock input TCLK12. The input on the RB4/TCLK12 pin is sampled and synchronized by the internal phase clocks twice every instruction cycle. This causes a delay from the time a falling edge appears on RB4/TCLK12 to the time TMR2:TMR1 is actually incremented. For the external clock input timing requirements, see the Electrical Specification section.

TABLE 12-1: TURNING ON 16-BIT TIMER

TMR2ON	TMR1ON	Result
1	1	16-bit timer (TMR2:TMR1) ON
0	1	Only TMR1 increments
x	0	16-bit timer OFF

FIGURE 12-4: TMR1 AND TMR2 IN 16-BIT TIMER/COUNTER MODE

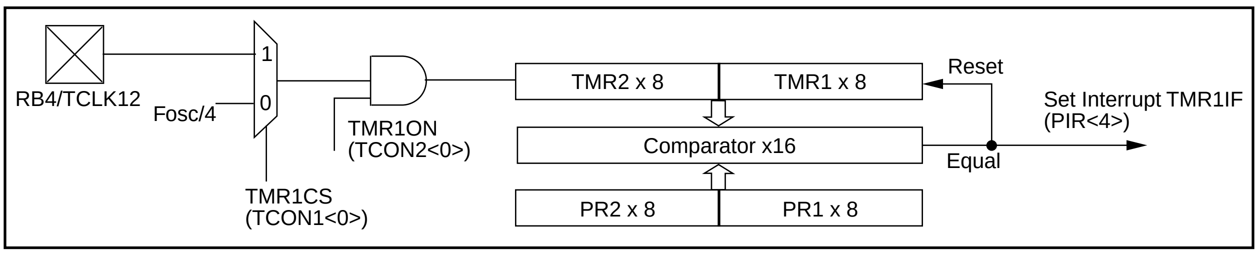


TABLE 12-2: SUMMARY OF TIMER1 AND TIMER2 REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
16h, Bank 3	TCON1	CA2ED1	CA2ED0	CA1ED1	CA1ED0	T16	TMR3CS	TMR2CS	TMR1CS	0000 0000	0000 0000
17h, Bank 3	TCON2	CA2OVF	CA1OVF	PWM2ON	PWM1ON	CA1/$\overline{PR3}$	TMR3ON	TMR2ON	TMR1ON	0000 0000	0000 0000
10h, Bank 2	TMR1	Timer1 register								xxxx xxxx	uuuu uuuu
11h, Bank 2	TMR2	Timer2 register								xxxx xxxx	uuuu uuuu
16h, Bank 1	PIR	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TXIF	RCIF	0000 0010	0000 0010
17h, Bank 1	PIE	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TXIE	RCIE	0000 0000	0000 0000
07h, Unbanked	INTSTA	PEIF	T0CKIF	T0IF	INTF	PEIE	T0CKIE	T0IE	INTE	0000 0000	0000 0000
06h, Unbanked	CPUSTA	—	—	STKAV	GLINTD	$\overline{TO}$	$\overline{PD}$	—	—	--11 11--	--11 qq--
14h, Bank 2	PR1	Timer1 period register								xxxx xxxx	uuuu uuuu
15h, Bank 2	PR2	Timer2 period register								xxxx xxxx	uuuu uuuu
10h, Bank 3	PW1DCL	DC1	DC0	—	—	—	—	—	—	xx-- ----	uu-- ----
11h, Bank 3	PW2DCL	DC1	DC0	TM2PW2	—	—	—	—	—	xx0- ----	uu0- ----
12h, Bank 3	PW1DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu
13h, Bank 3	PW2DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', q - value depends on condition, shaded cells are not used by Timer1 or Timer2.

Note 1: Other (non power-up) resets include: external reset through $\overline{MCLR}$ and WDT Timer Reset.

13.1 USART Baud Rate Generator (BRG)

The BRG supports both the Asynchronous and Synchronous modes of the USART. It is a dedicated 8-bit baud rate generator. The SPBRG register controls the period of a free running 8-bit timer. Table 13-1 shows the formula for computation of the baud rate for different USART modes. These only apply when the USART is in synchronous master mode (internal clock) and asynchronous mode.

Given the desired baud rate and Fosc, the nearest integer value between 0 and 255 can be calculated using the formula below. The error in baud rate can then be determined.

TABLE 13-1: BAUD RATE FORMULA

SYNC	Mode	Baud Rate
0	Asynchronous	FOSC/(64(X+1))
1	Synchronous	FOSC/(4(X+1))

X = value in SPBRG (0 to 255)

Example 13-1 shows the calculation of the baud rate error for the following conditions:

FOSC = 16 MHz
Desired Baud Rate = 9600
SYNC = 0

EXAMPLE 13-1: CALCULATING BAUD RATE ERROR

Desired Baud rate=Fosc / (64 (X + 1))

9600 = 16000000 /(64 (X + 1))

X = 25.042 = 25

Calculated Baud Rate=16000000 / (64 (25 + 1))

= 9615

Error = $\frac{\text{(Calculated Baud Rate - Desired Baud Rate)}}{\text{Desired Baud Rate}}$

= (9615 - 9600) / 9600

= 0.16%

Writing a new value to the SPBRG, causes the BRG timer to be reset (or cleared), this ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

TABLE 13-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
13h, Bank 0	RCSTA	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
15h, Bank 0	TXSTA	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG	Baud rate generator register								xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', shaded cells are not used by the Baud Rate Generator.
Note 1: Other (non power-up) resets include: external reset through $\overline{\text{MCLR}}$ and Watchdog Timer Reset.

TABLE 13-3: BAUD RATES FOR SYNCHRONOUS MODE

BAUD RATE (K)	FOSC = 33 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 25 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 20 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 16 MHz KBAUD	%ERROR	SPBRG value (decimal)
0.3	NA	—	—	NA	—	—	NA	—	—	NA	—	—
1.2	NA	—	—	NA	—	—	NA	—	—	NA	—	—
2.4	NA	—	—	NA	—	—	NA	—	—	NA	—	—
9.6	NA	—	—	NA	—	—	NA	—	—	NA	—	—
19.2	NA	—	—	NA	—	—	19.53	+1.73	255	19.23	+0.16	207
76.8	77.10	+0.39	106	77.16	+0.47	80	76.92	+0.16	64	76.92	+0.16	51
96	95.93	-0.07	85	96.15	+0.16	64	96.15	+0.16	51	95.24	-0.79	41
300	294.64	-1.79	27	297.62	-0.79	20	294.1	-1.96	16	307.69	+2.56	12
500	485.29	-2.94	16	480.77	-3.85	12	500	0	9	500	0	7
HIGH	8250	—	0	6250	—	0	5000	—	0	4000	—	0
LOW	32.22	—	255	24.41	—	255	19.53	—	255	15.625	—	255

BAUD RATE (K)	FOSC = 10 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 7.159 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 5.068 MHz KBAUD	%ERROR	SPBRG value (decimal)
0.3	NA	—	—	NA	—	—	NA	—	—
1.2	NA	—	—	NA	—	—	NA	—	—
2.4	NA	—	—	NA	—	—	NA	—	—
9.6	9.766	+1.73	255	9.622	+0.23	185	9.6	0	131
19.2	19.23	+0.16	129	19.24	+0.23	92	19.2	0	65
76.8	75.76	-1.36	32	77.82	+1.32	22	79.2	+3.13	15
96	96.15	+0.16	25	94.20	-1.88	18	97.48	+1.54	12
300	312.5	+4.17	7	298.3	-0.57	5	316.8	+5.60	3
500	500	0	4	NA	—	—	NA	—	—
HIGH	2500	—	0	1789.8	—	0	1267	—	0
LOW	9.766	—	255	6.991	—	255	4.950	—	255

BAUD RATE (K)	FOSC = 3.579 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 1 MHz KBAUD	%ERROR	SPBRG value (decimal)	FOSC = 32.768 kHz KBAUD	%ERROR	SPBRG value (decimal)
0.3	NA	—	—	NA	—	—	0.303	+1.14	26
1.2	NA	—	—	1.202	+0.16	207	1.170	-2.48	6
2.4	NA	—	—	2.404	+0.16	103	NA	—	—
9.6	9.622	+0.23	92	9.615	+0.16	25	NA	—	—
19.2	19.04	-0.83	46	19.24	+0.16	12	NA	—	—
76.8	74.57	-2.90	11	83.34	+8.51	2	NA	—	—
96	99.43	_3.57	8	NA	—	—	NA	—	—
300	298.3	-0.57	2	NA	—	—	NA	—	—
500	NA	—	—	NA	—	—	NA	—	—
HIGH	894.9	—	0	250	—	0	8.192	—	0
LOW	3.496	—	255	0.976	—	255	0.032	—	255

13.4 USART Synchronous Slave Mode

The synchronous slave mode differs from the master mode in the fact that the shift clock is supplied externally at the RA5/TX/CK pin (instead of being supplied internally in the master mode). This allows the device to transfer or receive data in the SLEEP mode. The slave mode is entered by clearing the CSRC (TXSTA<7>) bit.

13.4.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the sync master and slave modes are identical except in the case of the SLEEP mode.

If two words are written to TXREG and then the SLEEP instruction executes, the following will occur. The first word will immediately transfer to the TSR and will transmit as the shift clock is supplied. The second word will remain in TXREG. TXIF will not be set. When the first word has been shifted out of TSR, TXREG will transfer the second word to the TSR and the TXIF flag will now be set. If TXIE is enabled, the interrupt will wake the chip from SLEEP and if the global interrupt is enabled, then the program will branch to interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. Clear the CREN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. Start transmission by loading data to TXREG.
6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
7. Enable the transmission by setting TXEN.

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN) allows transmission to start sooner then doing these two events in the reverse order.

Note:	To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

13.4.2 USART SYNCHRONOUS SLAVE RECEPTION

Operation of the synchronous master and slave modes are identical except in the case of the SLEEP mode. Also, SREN is a don't care in slave mode.

If receive is enabled (CREN) prior to the SLEEP instruction, then a word may be received during SLEEP. On completely receiving the word, the RSR will transfer the data to RCREG (setting RCIF) and if the RCIE bit is set, the interrupt generated will wake the chip from SLEEP. If the global interrupt is enabled, the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Reception:

1. Enable the synchronous master serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. If interrupts are desired, then set the RCIE bit.
3. If 9-bit reception is desired, then set the RX9 bit.
4. To enable reception, set the CREN bit.
5. The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
6. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
7. Read the 8-bit received data by reading RCREG.
8. If any error occurred, clear the error by clearing the CREN bit.

Note:	To abort reception, either clear the SPEN bit, the SREN bit (when in single receive mode), or the CREN bit (when in continuous receive mode). This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

14.4.2 MINIMIZING CURRENT CONSUMPTION

To minimize current consumption, all I/O pins should be either at V_{DD}, or V_{SS}, with no external circuitry drawing current from the I/O pin. I/O pins that are hi-impedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The T0CKI input should be at V_{DD} or V_{SS}. The contributions from on-chip pull-ups on PORTB should also be considered, and disabled when possible.

14.5 Code Protection

The code in the program memory can be protected by selecting the microcontroller in code protected mode (PM2:PM0 = '000').

> **Note:** PM2 does not exist on the PIC17C42. To select code protected microcontroller mode, PM1:PM0 = '00'.

In this mode, instructions that are in the on-chip program memory space, can continue to read or write the program memory. An instruction that is executed outside of the internal program memory range will be inhibited from writing to or reading from program memory.

> **Note:** Microchip does not recommend code protecting windowed devices.

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

CPFSEQ — Compare f with WREG, skip if f = WREG

Syntax:	[*label*] CPFSEQ f
Operands:	$0 \le f \le 255$
Operation:	(f) – (WREG), skip if (f) = (WREG) (unsigned comparison)
Status Affected:	None
Encoding:	`0011` `0001` `ffff` `ffff`
Description:	Compares the contents of data memory location 'f' to the contents of WREG by performing an unsigned subtraction. If 'f' = WREG then the fetched instruction is discarded and an NOP is executed instead making this a two-cycle instruction.
Words:	1
Cycles:	1 (2)

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	NOP

If skip:

Q1	Q2	Q3	Q4
Forced NOP	NOP	Execute	NOP

Example:

```
HERE    CPFSEQ REG
NEQUAL  :
EQUAL   :
```

Before Instruction

PC Address	=	`HERE`
WREG	=	?
REG	=	?

After Instruction

If REG	=	WREG;
PC	=	Address (`EQUAL`)
If REG	≠	WREG;
PC	=	Address (`NEQUAL`)

CPFSGT — Compare f with WREG, skip if f > WREG

Syntax:	[*label*] CPFSGT f
Operands:	$0 \le f \le 255$
Operation:	(f) – (WREG), skip if (f) > (WREG) (unsigned comparison)
Status Affected:	None
Encoding:	`0011` `0010` `ffff` `ffff`
Description:	Compares the contents of data memory location 'f' to the contents of the WREG by performing an unsigned subtraction. If the contents of 'f' > the contents of WREG then the fetched instruction is discarded and an NOP is executed instead making this a two-cycle instruction.
Words:	1
Cycles:	1 (2)

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	NOP

If skip:

Q1	Q2	Q3	Q4
Forced NOP	NOP	Execute	NOP

Example:

```
HERE      CPFSGT REG
NGREATER  :
GREATER   :
```

Before Instruction

PC	=	Address (`HERE`)
WREG	=	?

After Instruction

If REG	>	WREG;
PC	=	Address (`GREATER`)
If REG	≤	WREG;
PC	=	Address (`NGREATER`)

MOVFP — Move f to p

Syntax:	[*label*] MOVFP f,p
Operands:	$0 \le f \le 255$ $0 \le p \le 31$
Operation:	(f) → (p)
Status Affected:	None

Encoding:

`011p`	`pppp`	`ffff`	`ffff`

Description: Move data from data memory location 'f' to data memory location 'p'. Location 'f' can be anywhere in the 256 word data space (00h to FFh) while 'p' can be 00h to 1Fh.

Either 'p' or 'f' can be WREG (a useful special situation).

`MOVFP` is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port). Both 'f' and 'p' can be indirectly addressed.

Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	Write register 'p'

Example: `MOVFP REG1, REG2`

Before Instruction
REG1 = 0x33,
REG2 = 0x11

After Instruction
REG1 = 0x33,
REG2 = 0x33

MOVLB — Move Literal to low nibble in BSR

Syntax:	[*label*] MOVLB k
Operands:	$0 \le k \le 15$
Operation:	k → (BSR<3:0>)
Status Affected:	None

Encoding:

`1011`	`1000`	`uuuu`	`kkkk`

Description: The four bit literal 'k' is loaded in the Bank Select Register (BSR). Only the low 4-bits of the Bank Select Register are affected. The upper half of the BSR is unchanged. The assembler will encode the “u” fields as '0'.

Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'u:k'	Execute	Write literal 'k' to BSR<3:0>

Example: `MOVLB 0x5`

Before Instruction
BSR register = 0x22

After Instruction
BSR register = 0x25

Note: For the PIC17C42, only the low four bits of the BSR register are physically implemented. The upper nibble is read as '0'.

16.0 DEVELOPMENT SUPPORT

16.1 Development Tools

The PIC16/17 microcontrollers are supported with a full range of hardware and software development tools:

- PICMASTER/PICMASTER CE Real-Time In-Circuit Emulator
- ICEPIC Low-Cost PIC16C5X and PIC16CXXX In-Circuit Emulator
- PRO MATE® II Universal Programmer
- PICSTART® Plus Entry-Level Prototype Programmer
- PICDEM-1 Low-Cost Demonstration Board
- PICDEM-2 Low-Cost Demonstration Board
- PICDEM-3 Low-Cost Demonstration Board
- MPASM Assembler
- MPLAB-SIM Software Simulator
- MPLAB-C (C Compiler)
- Fuzzy logic development system (fuzzyTECH®–MP)

16.2 PICMASTER: High Performance Universal In-Circuit Emulator with MPLAB IDE

The PICMASTER Universal In-Circuit Emulator is intended to provide the product development engineer with a complete microcontroller design tool set for all microcontrollers in the PIC12C5XX, PIC14000, PIC16C5X, PIC16CXXX and PIC17CXX families. PICMASTER is supplied with the MPLAB™ Integrated Development Environment (IDE), which allows editing, “make” and download, and source debugging from a single environment.

Interchangeable target probes allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the PICMASTER allows expansion to support all new Microchip microcontrollers.

The PICMASTER Emulator System has been designed as a real-time emulation system with advanced features that are generally found on more expensive development tools. The PC compatible 386 (and higher) machine platform and Microsoft Windows® 3.x environment were chosen to best make these features available to you, the end user.

A CE compliant version of PICMASTER is available for European Union (EU) countries.

16.3 ICEPIC: Low-cost PIC16CXXX In-Circuit Emulator

ICEPIC is a low-cost in-circuit emulator solution for the Microchip PIC16C5X and PIC16CXXX families of 8-bit OTP microcontrollers.

ICEPIC is designed to operate on PC-compatible machines ranging from 286-AT® through Pentium™ based machines under Windows 3.x environment. ICEPIC features real time, non-intrusive emulation.

16.4 PRO MATE II: Universal Programmer

The PRO MATE II Universal Programmer is a full-featured programmer capable of operating in stand-alone mode as well as PC-hosted mode.

The PRO MATE II has programmable V_{DD} and V_{PP} supplies which allows it to verify programmed memory at V_{DD} min and V_{DD} max for maximum reliability. It has an LCD display for displaying error messages, keys to enter commands and a modular detachable socket assembly to support various package types. In stand-alone mode the PRO MATE II can read, verify or program PIC16C5X, PIC16CXXX, PIC17CXX and PIC14000 devices. It can also set configuration and code-protect bits in this mode.

16.5 PICSTART Plus Entry Level Development System

The PICSTART programmer is an easy-to-use, low-cost prototype programmer. It connects to the PC via one of the COM (RS-232) ports. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. PICSTART Plus is not recommended for production programming.

PICSTART Plus supports all PIC12C5XX, PIC14000, PIC16C5X, PIC16CXXX and PIC17CXX devices with up to 40 pins. Larger pin count devices such as the PIC16C923 and PIC16C924 may be supported with an adapter socket.

MPASM allow full symbolic debugging from the Microchip Universal Emulator System (PICMASTER).

MPASM has the following features to assist in developing software for specific use applications.

- Provides translation of Assembler source code to object code for all Microchip microcontrollers.
- Macro assembly capability.
- Produces all the files (Object, Listing, Symbol, and special) required for symbolic debug with Microchip's emulator systems.
- Supports Hex (default), Decimal and Octal source and listing formats.

MPASM provides a rich directive language to support programming of the PIC16/17. Directives are helpful in making the development of your assemble source code shorter and more maintainable.

16.11 Software Simulator (MPLAB-SIM)

The MPLAB-SIM Software Simulator allows code development in a PC host environment. It allows the user to simulate the PIC16/17 series microcontrollers on an instruction level. On any given instruction, the user may examine or modify any of the data areas or provide external stimulus to any of the pins. The input/output radix can be set by the user and the execution can be performed in; single step, execute until break, or in a trace mode.

MPLAB-SIM fully supports symbolic debugging using MPLAB-C and MPASM. The Software Simulator offers the low cost flexibility to develop and debug code outside of the laboratory environment making it an excellent multi-project software development tool.

16.12 C Compiler (MPLAB-C)

The MPLAB-C Code Development System is a complete 'C' compiler and integrated development environment for Microchip's PIC16/17 family of microcontrollers. The compiler provides powerful integration capabilities and ease of use not found with other compilers.

For easier source level debugging, the compiler provides symbol information that is compatible with the MPLAB IDE memory display (PICMASTER emulator software versions 1.13 and later).

16.13 Fuzzy Logic Development System (*fuzzy*TECH-MP)

*fuzzy*TECH-MP fuzzy logic development tool is available in two versions - a low cost introductory version, MP Explorer, for designers to gain a comprehensive working knowledge of fuzzy logic system design; and a full-featured version, *fuzzy*TECH-MP, edition for implementing more complex systems.

Both versions include Microchip's *fuzzy*LAB™ demonstration board for hands-on experience with fuzzy logic systems implementation.

16.14 MP-DriveWay™ – Application Code Generator

MP-DriveWay is an easy-to-use Windows-based Application Code Generator. With MP-DriveWay you can visually configure all the peripherals in a PIC16/17 device and, with a click of the mouse, generate all the initialization and many functional code modules in C language. The output is fully compatible with Microchip's MPLAB-C C compiler. The code produced is highly modular and allows easy integration of your own code. MP-DriveWay is intelligent enough to maintain your code through subsequent code generation.

16.15 SEEVAL® Evaluation and Programming System

The SEEVAL SEEPROM Designer's Kit supports all Microchip 2-wire and 3-wire Serial EEPROMs. The kit includes everything necessary to read, write, erase or program special features of any Microchip SEEPROM product including Smart Serials™ and secure serials. The Total Endurance™ Disk is included to aid in trade-off analysis and reliability calculations. The total kit can significantly reduce time-to-market and result in an optimized system.

16.16 TrueGauge® Intelligent Battery Management

The TrueGauge development tool supports system development with the MTA11200B TrueGauge Intelligent Battery Management IC. System design verification can be accomplished before hardware prototypes are built. User interface is graphically-oriented and measured data can be saved in a file for exporting to Microsoft Excel.

16.17 KEELOQ® Evaluation and Programming Tools

KEELOQ evaluation and programming tools support Microchips HCS Secure Data Products. The HCS evaluation kit includes an LCD display to show changing codes, a decoder to decode transmissions, and a programming interface to program test transmitters.

Applicable Devices	42	R42	42A	43	R43	44

17.1 DC CHARACTERISTICS: PIC17C42-16 (Commercial, Industrial) PIC17C42-25 (Commercial, Industrial)

DC CHARACTERISTICS			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40˚C ≤ TA ≤ +85˚C for industrial and 0˚C ≤ TA ≤ +70˚C for commercial				
Parameter No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D001	VDD	Supply Voltage	4.5	–	5.5	V	
D002	VDR	RAM Data Retention Voltage (Note 1)	1.5 *	–	–	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure internal Power-on Reset signal	–	VSS	–	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure internal Power-on Reset signal	0.060*	–	–	mV/ms	See section on Power-on Reset for details
D010 D011 D012 D013 D014	IDD	Supply Current (Note 2)	– – – – –	3 6 11 19 95	6 12 * 24 * 38 150	mA mA mA mA µA	FOSC = 4 MHz (Note 4) FOSC = 8 MHz FOSC = 16 MHz FOSC = 25 MHz FOSC = 32 kHz WDT enabled (EC osc configuration)
D020 D021	IPD	Power-down Current (Note 3)	– –	10 < 1	40 5	µA µA	VDD = 5.5V, WDT enabled VDD = 5.5V, WDT disabled

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:
$\overline{\text{OSC1}}$ = external square wave, from rail to rail; all I/O pins tristated, pulled to VDD or VSS, T0CKI = VDD,
$\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.
Current consumed from the oscillator and I/O's driving external capacitive or resistive loads need to be considered.
For the RC oscillator, the current through the external pull-up resistor (R) can be estimated as: VDD / (2 • R).
For capacitive loads, The current can be estimated (for an individual I/O pin) as (CL • VDD) • f
CL = Total capacitive load on the I/O pin; f = average frequency on the I/O pin switches.
The capacitive currents are most significant when the device is configured for external execution (includes extended microcontroller mode).

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula IR = VDD/2Rext (mA) with Rext in kOhm.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 18-13: WDT TIMER TIME-OUT PERIOD vs. VDD

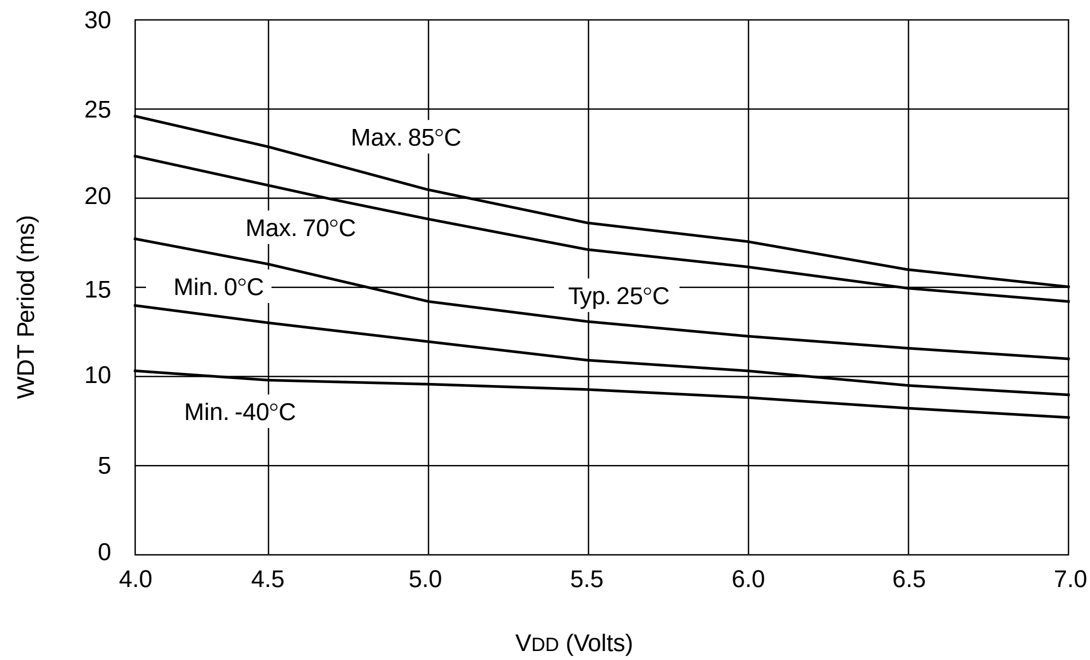
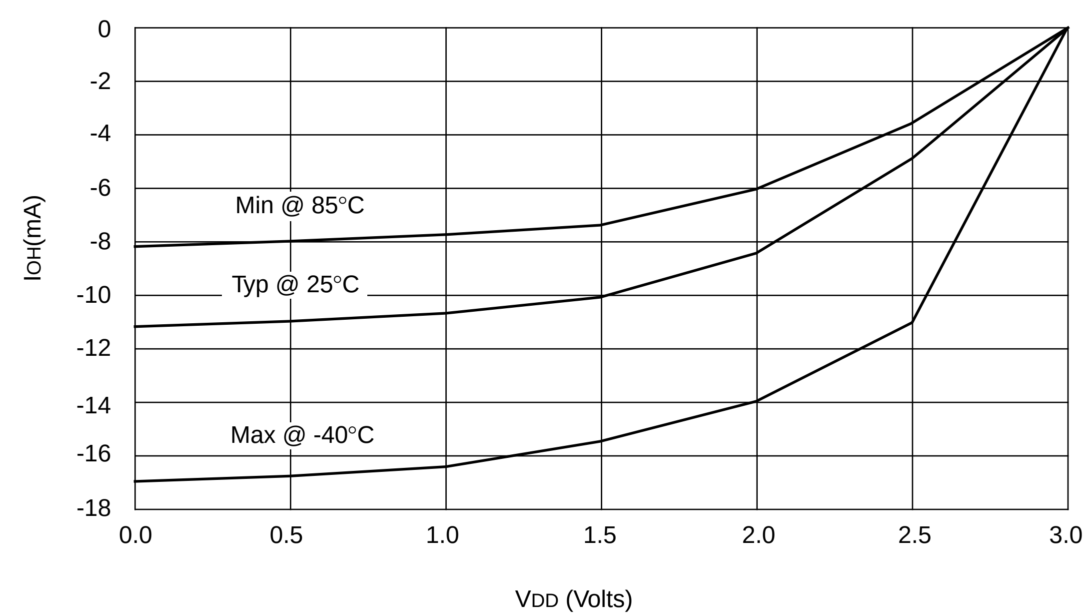


FIGURE 18-14: IOH vs. VOH, VDD = 3V



NOTES:

Applicable Devices	42	R42	42A	43	R43	44

19.1 DC CHARACTERISTICS: PIC17CR42/42A/43/R43/44-16 (Commercial, Industrial) PIC17CR42/42A/43/R43/44-25 (Commercial, Industrial) PIC17CR42/42A/43/R43/44-33 (Commercial, Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40˚C ≤ TA ≤ +85˚C for industrial and 0˚C ≤ TA ≤ +70˚C for commercial				
Parameter No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D001	V_{DD}	Supply Voltage	4.5	–	6.0	V	
D002	V_{DR}	RAM Data Retention Voltage (Note 1)	1.5 *	–	–	V	Device in SLEEP mode
D003	V_{POR}	V_{DD} start voltage to ensure internal Power-on Reset signal	–	V_{SS}	–	V	See section on Power-on Reset for details
D004	S_{VDD}	V_{DD} rise rate to ensure internal Power-on Reset signal	0.060 *	–	–	mV/ms	See section on Power-on Reset for details
D010 D011 D012 D013 D015 D014	I_{DD}	Supply Current (Note 2)	– – – – – –	3 6 11 19 25 95	6 12 * 24 * 38 50 150	mA mA mA mA mA µA	F_{OSC} = 4 MHz (Note 4) F_{OSC} = 8 MHz F_{OSC} = 16 MHz F_{OSC} = 25 MHz F_{OSC} = 33 MHz F_{OSC} = 32 kHz, WDT enabled (EC osc configuration)
D020 D021	I_{PD}	Power-down Current (Note 3)	– –	10 < 1	40 5	µA µA	V_{DD} = 5.5V, WDT enabled V_{DD} = 5.5V, WDT disabled

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which V_{DD} can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all I_{DD} measurements in active operation mode are:
OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to V_{DD} or V_{SS}, T0CKI = V_{DD}, $\overline{MCLR}$ = V_{DD}; WDT enabled/disabled as specified.
Current consumed from the oscillator and I/O's driving external capacitive or resistive loads needs to be considered.
For the RC oscillator, the current through the external pull-up resistor (R) can be estimated as: V_{DD} / (2 • R).
For capacitive loads, the current can be estimated (for an individual I/O pin) as (C_L • V_{DD}) • f
C_L = Total capacitive load on the I/O pin; f = average frequency the I/O pin switches.
The capacitive currents are most significant when the device is configured for external execution (includes extended microcontroller mode).

3: The power down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to V_{DD} and V_{SS}.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula I_R = V_{DD}/2Rext (mA) with Rext in kOhm.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 20-11: TYPICAL IPD vs. VDD WATCHDOG ENABLED 25°C

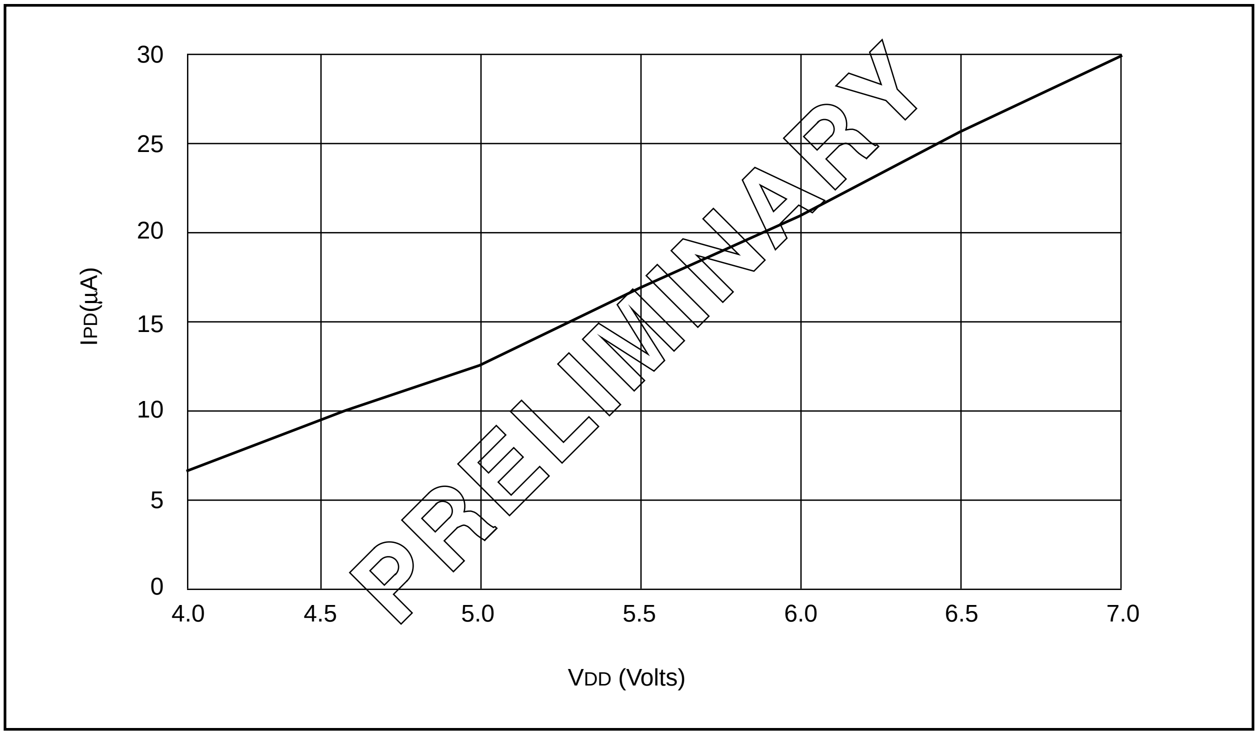


FIGURE 20-12: MAXIMUM IPD vs. VDD WATCHDOG ENABLED