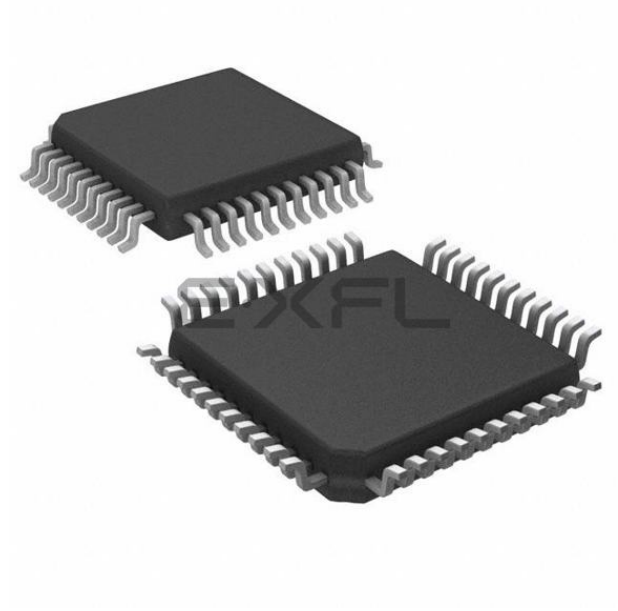

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Details	
Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
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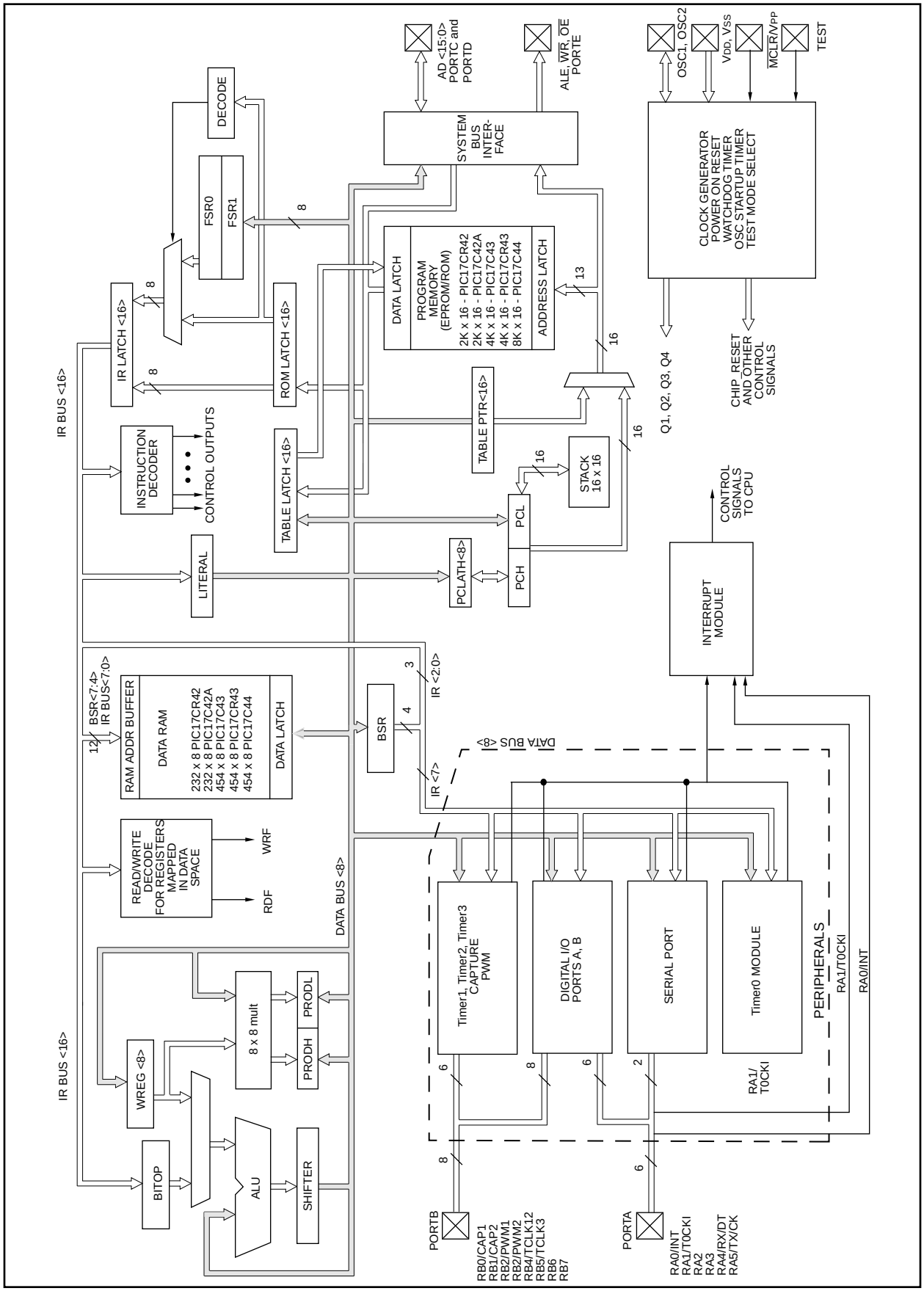
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	33
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	454 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-QFP
Supplier Device Package	44-MQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c44-25e-pq

FIGURE 3-2: PIC17CR42/42A/43/R43/44 BLOCK DIAGRAM



4.0 RESET

The PIC17CXX differentiates between various kinds of reset:

- Power-on Reset (POR)
- $\overline{MCLR}$ reset during normal operation
- WDT Reset (normal operation)

Some registers are not affected in any reset condition; their status is unknown on POR and unchanged in any other reset. Most other registers are forced to a "reset state" on Power-on Reset (POR), on $\overline{MCLR}$ or WDT Reset and on $\overline{MCLR}$ reset during SLEEP. They are not affected by a WDT Reset during SLEEP, since this reset is viewed as the resumption of normal operation. The $\overline{TO}$ and $\overline{PD}$ bits are set or cleared differently in different reset situations as indicated in Table 4-3. These bits are used in software to determine the nature of reset. See Table 4-4 for a full description of reset states of all registers.

Note:	While the device is in a reset state, the internal phase clock is held in the Q1 state. Any processor mode that allows external execution will force the RE0/ALE pin as a low output and the RE1/$\overline{OE}$ and RE2/$\overline{WR}$ pins as high outputs.

A simplified block diagram of the on-chip reset circuit is shown in Figure 4-1.

4.1 Power-on Reset (POR), Power-up Timer (PWRT), and Oscillator Start-up Timer (OST)

4.1.1 POWER-ON RESET (POR)

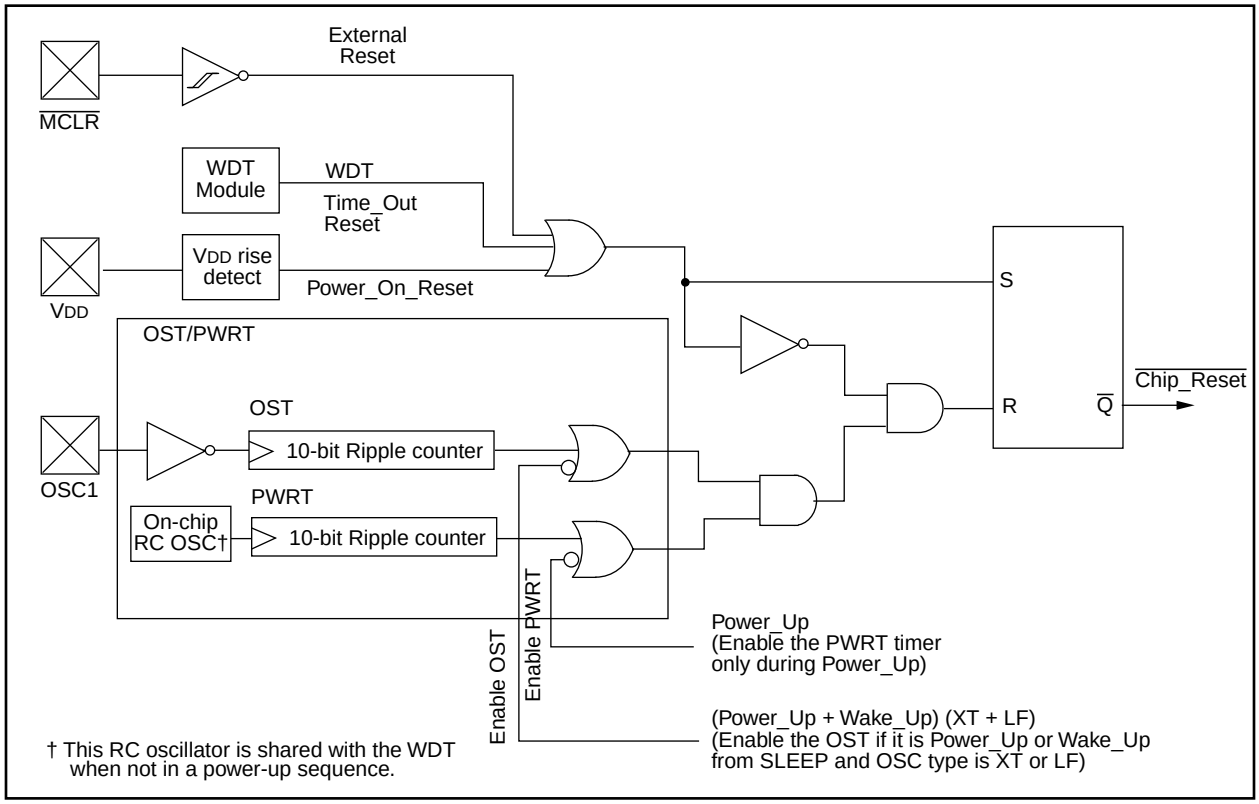
The Power-on Reset circuit holds the device in reset until V_{DD} is above the trip point (in the range of 1.4V - 2.3V). The PIC17C42 does not produce an internal reset when V_{DD} declines. All other devices will produce an internal reset for both rising and falling V_{DD}. To take advantage of the POR, just tie the $\overline{MCLR}/V_{PP}$ pin directly (or through a resistor) to V_{DD}. This will eliminate external RC components usually needed to create Power-on Reset. A minimum rise time for V_{DD} is required. See Electrical Specifications for details.

4.1.2 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 96 ms time-out (nominal) on power-up. This occurs from rising edge of the POR signal and after the first rising edge of $\overline{MCLR}$ (detected high). The Power-up Timer operates on an internal RC oscillator. The chip is kept in RESET as long as the PWRT is active. In most cases the PWRT delay allows the V_{DD} to rise to an acceptable level.

The power-up time delay will vary from chip to chip and to V_{DD} and temperature. See DC parameters for details.

FIGURE 4-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



6.2.2.3 TMR0 STATUS/CONTROL REGISTER (T0STA)

This register contains various control bits. Bit7 (INTEDG) is used to control the edge upon which a signal on the RA0/INT pin will set the RB0/INT interrupt flag. The other bits configure the Timer0 prescaler and clock source. (Figure 11-1).

FIGURE 6-9: T0STA REGISTER (ADDRESS: 05h, UNBANKED)

R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	U - 0
INTEDG	T0SE	T0CS	PS3	PS2	PS1	PS0	—
bit7							bit0

R = Readable bit
W = Writable bit
U = Unimplemented, reads as '0'
-n = Value at POR reset

bit 7: **INTEDG**: RA0/INT Pin Interrupt Edge Select bit
This bit selects the edge upon which the interrupt is detected.
1 = Rising edge of RA0/INT pin generates interrupt
0 = Falling edge of RA0/INT pin generates interrupt

bit 6: **T0SE**: Timer0 Clock Input Edge Select bit
This bit selects the edge upon which TMR0 will increment.
When T0CS = 0
1 = Rising edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
0 = Falling edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
When T0CS = 1
Don't care

bit 5: **T0CS**: Timer0 Clock Source Select bit
This bit selects the clock source for Timer0.
1 = Internal instruction clock cycle (T_{CY})
0 = T0CKI pin

bit 4-1: **PS3:PS0**: Timer0 Prescale Selection bits
These bits select the prescale value for Timer0.

PS3:PS0	Prescale Value
0000	1:1
0001	1:2
0010	1:4
0011	1:8
0100	1:16
0101	1:32
0110	1:64
0111	1:128
1xxx	1:256

bit 0: **Unimplemented**: Read as '0'

6.3 Stack Operation

The PIC17C4X devices have a 16 x 16-bit wide hardware stack (Figure 6-1). The stack is not part of either the program or data memory space, and the stack pointer is neither readable nor writable. The PC is "PUSHed" onto the stack when a CALL instruction is executed or an interrupt is acknowledged. The stack is "POPed" in the event of a RETURN, RETLW, or a RETFIE instruction execution. PCLATH is not affected by a "PUSH" or a "POP" operation.

The stack operates as a circular buffer, with the stack pointer initialized to '0' after all resets. There is a stack available bit (STKAV) to allow software to ensure that the stack has not overflowed. The STKAV bit is set after a device reset. When the stack pointer equals Fh, STKAV is cleared. When the stack pointer rolls over from Fh to 0h, the STKAV bit will be held clear until a device reset.

Note 1: There is not a status bit for stack underflow. The STKAV bit can be used to detect the underflow which results in the stack pointer being at the top of stack.

Note 2: There are no instruction mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW, and RETFIE instructions, or the vectoring to an interrupt vector.

Note 3: After a reset, if a "POP" operation occurs before a "PUSH" operation, the STKAV bit will be cleared. This will appear as if the stack is full (underflow has occurred). If a "PUSH" operation occurs next (before another "POP"), the STKAV bit will be locked clear. Only a device reset will cause this bit to set.

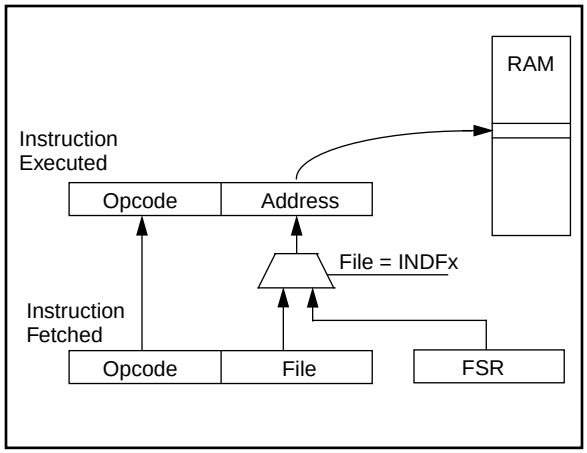
After the device is "PUSHed" sixteen times (without a "POP"), the seventeenth push overwrites the value from the first push. The eighteenth push overwrites the second push (and so on).

6.4 Indirect Addressing

Indirect addressing is a mode of addressing data memory where the data memory address in the instruction is not fixed. That is, the register that is to be read or written can be modified by the program. This can be useful for data tables in the data memory. Figure 6-10 shows the operation of indirect addressing. This shows the moving of the value to the data memory address specified by the value of the FSR register.

Example 6-1 shows the use of indirect addressing to clear RAM in a minimum number of instructions. A similar concept could be used to move a defined number of bytes (block) of data to the USART transmit register (TXREG). The starting address of the block of data to be transmitted could easily be modified by the program.

FIGURE 6-10: INDIRECT ADDRESSING



10.0 OVERVIEW OF TIMER RESOURCES

The PIC17C4X has four timer modules. Each module can generate an interrupt to indicate that an event has occurred. These timers are called:

- Timer0 - 16-bit timer with programmable 8-bit prescaler
- Timer1 - 8-bit timer
- Timer2 - 8-bit timer
- Timer3 - 16-bit timer

For enhanced time-base functionality, two input Captures and two Pulse Width Modulation (PWM) outputs are possible. The PWMs use the TMR1 and TMR2 resources and the input Captures use the TMR3 resource.

10.1 Timer0 Overview

The Timer0 module is a simple 16-bit overflow counter. The clock source can be either the internal system clock (Fosc/4) or an external clock.

The Timer0 module also has a programmable prescaler option. The PS3:PS0 bits (T0STA<4:1>) determine the prescaler value. TMR0 can increment at the following rates: 1:1, 1:2, 1:4, 1:8, 1:16, 1:32, 1:64, 1:128, 1:256.

When TImer0's clock source is an external clock, the Timer0 module can be selected to increment on either the rising or falling edge.

Synchronization of the external clock occurs after the prescaler. When the prescaler is used, the external clock frequency may be higher then the device's frequency. The maximum frequency is 50 MHz, given the high and low time requirements of the clock.

10.2 Timer1 Overview

The TImer0 module is an 8-bit timer/counter with an 8-bit period register (PR1). When the TMR1 value rolls over from the period match value to 0h, the TMR1IF flag is set, and an interrupt will be generated when enabled. In counter mode, the clock comes from the RB4/TCLK12 pin, which can also be selected to be the clock for the Timer2 module.

TMR1 can be concatenated to TMR2 to form a 16-bit timer. The TMR1 register is the LSB and TMR2 is the MSB. When in the 16-bit timer mode, there is a corresponding 16-bit period register (PR2:PR1). When the TMR2:TMR1 value rolls over from the period match value to 0h, the TMR1IF flag is set, and an interrupt will be generated when enabled.

10.3 Timer2 Overview

The TMR2 module is an 8-bit timer/counter with an 8-bit period register (PR2). When the TMR2 value rolls over from the period match value to 0h, the TMR2IF flag is set, and an interrupt will be generated when enabled. In counter mode, the clock comes from the RB4/TCLK12 pin, which can also be selected to be the clock for the TMR1 module.

TMR1 can be concatenated to TMR2 to form a 16-bit timer. The TMR2 register is the MSB and TMR1 is the LSB. When in the 16-bit timer mode, there is a corresponding 16-bit period register (PR2:PR1). When the TMR2:TMR1 value rolls over from the period match value to 0h, the TMR1IF flag is set, and an interrupt will be generated when enabled.

10.4 Timer3 Overview

The TImer3 module is a 16-bit timer/counter with a 16-bit period register. When the TMR3H:TMR3L value rolls over to 0h, the TMR3IF bit is set and an interrupt will be generated when enabled. In counter mode, the clock comes from the RB5/TCLK3 pin.

When operating in the dual capture mode, the period registers become the second 16-bit capture register.

10.5 Role of the Timer/Counters

The timer modules are general purpose, but have dedicated resources associated with them. TImer1 and Timer2 are the time-bases for the two Pulse Width Modulation (PWM) outputs, while Timer3 is the time-base for the two input captures.

11.0 TIMER0

The Timer0 module consists of a 16-bit timer/counter, TMR0. The high byte is TMR0H and the low byte is TMR0L. A software programmable 8-bit prescaler makes an effective 24-bit overflow timer. The clock source is also software programmable as either the internal instruction clock or the RA1/T0CKI pin. The control bits for this module are in register T0STA (Figure 11-1).

FIGURE 11-1: T0STA REGISTER (ADDRESS: 05h, UNBANKED)

R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	R/W - 0	U - 0
INTEDG	T0SE	T0CS	PS3	PS2	PS1	PS0	—
bit7							bit0

R = Readable bit
W = Writable bit
U = Unimplemented, Read as '0'
-n = Value at POR reset

bit 7: **INTEDG**: RA0/INT Pin Interrupt Edge Select bit
This bit selects the edge upon which the interrupt is detected
1 = Rising edge of RA0/INT pin generates interrupt
0 = Falling edge of RA0/INT pin generates interrupt

bit 6: **T0SE**: Timer0 Clock Input Edge Select bit
This bit selects the edge upon which TMR0 will increment
When T0CS = 0
1 = Rising edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
0 = Falling edge of RA1/T0CKI pin increments TMR0 and/or generates a T0CKIF interrupt
When T0CS = 1
Don't care

bit 5: **T0CS**: Timer0 Clock Source Select bit
This bit selects the clock source for TMR0.
1 = Internal instruction clock cycle (T_{CY})
0 = T0CKI pin

bit 4-1: **PS3:PS0**: Timer0 Prescale Selection bits
These bits select the prescale value for TMR0.

PS3:PS0	Prescale Value
0000	1:1
0001	1:2
0010	1:4
0011	1:8
0100	1:16
0101	1:32
0110	1:64
0111	1:128
1xxx	1:256

bit 0: **Unimplemented**: Read as '0'

12.1.2 TIMER1 & TIMER2 IN 16-BIT MODE

To select 16-bit mode, the T16 bit must be set. In this mode TMR1 and TMR2 are concatenated to form a 16-bit timer (TMR2:TMR1). The 16-bit timer increments until it matches the 16-bit period register (PR2:PR1). On the following timer clock, the timer value is reset to 0h, and the TMR1IF bit is set.

When selecting the clock source for the16-bit timer, the TMR1CS bit controls the entire 16-bit timer and TMR2CS is a “don’t care.” When TMR1CS is clear, the timer increments once every instruction cycle (Fosc/4). When TMR1CS is set, the timer increments on every falling edge of the RB4/TCLK12 pin. For the 16-bit timer to increment, both TMR1ON and TMR2ON bits must be set (Table 12-1).

12.1.2.1 EXTERNAL CLOCK INPUT FOR TMR1:TMR2

When TMR1CS is set, the 16-bit TMR2:TMR1 increments on the falling edge of clock input TCLK12. The input on the RB4/TCLK12 pin is sampled and synchronized by the internal phase clocks twice every instruction cycle. This causes a delay from the time a falling edge appears on RB4/TCLK12 to the time TMR2:TMR1 is actually incremented. For the external clock input timing requirements, see the Electrical Specification section.

TABLE 12-1: TURNING ON 16-BIT TIMER

TMR2ON	TMR1ON	Result
1	1	16-bit timer (TMR2:TMR1) ON
0	1	Only TMR1 increments
x	0	16-bit timer OFF

FIGURE 12-4: TMR1 AND TMR2 IN 16-BIT TIMER/COUNTER MODE

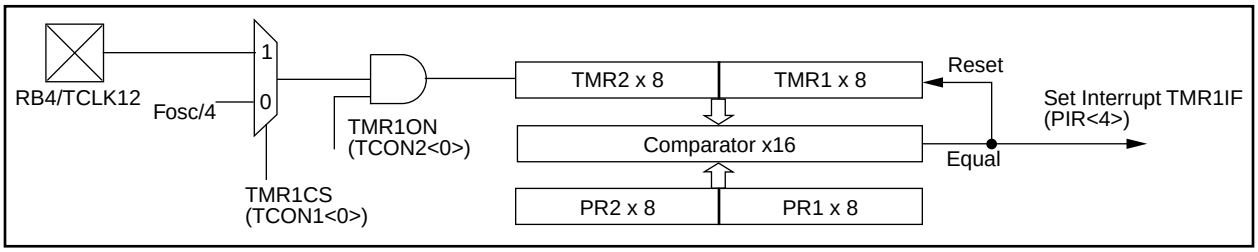


TABLE 12-2: SUMMARY OF TIMER1 AND TIMER2 REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
16h, Bank 3	TCON1	CA2ED1	CA2ED0	CA1ED1	CA1ED0	T16	TMR3CS	TMR2CS	TMR1CS	0000 0000	0000 0000
17h, Bank 3	TCON2	CA2OVF	CA1OVF	PWM2ON	PWM1ON	CA1/$\overline{PR3}$	TMR3ON	TMR2ON	TMR1ON	0000 0000	0000 0000
10h, Bank 2	TMR1	Timer1 register								xxxx xxxx	uuuu uuuu
11h, Bank 2	TMR2	Timer2 register								xxxx xxxx	uuuu uuuu
16h, Bank 1	PIR	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TXIF	RCIF	0000 0010	0000 0010
17h, Bank 1	PIE	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TXIE	RCIE	0000 0000	0000 0000
07h, Unbanked	INTSTA	PEIF	T0CKIF	T0IF	INTF	PEIE	T0CKIE	T0IE	INTE	0000 0000	0000 0000
06h, Unbanked	CPUSTA	—	—	STKAV	GLINTD	$\overline{TO}$	$\overline{PD}$	—	—	--11 11--	--11 qq--
14h, Bank 2	PR1	Timer1 period register								xxxx xxxx	uuuu uuuu
15h, Bank 2	PR2	Timer2 period register								xxxx xxxx	uuuu uuuu
10h, Bank 3	PW1DCL	DC1	DC0	—	—	—	—	—	—	xx-- ----	uu-- ----
11h, Bank 3	PW2DCL	DC1	DC0	TM2PW2	—	—	—	—	—	xx0- ----	uu0- ----
12h, Bank 3	PW1DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu
13h, Bank 3	PW2DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', q - value depends on condition, shaded cells are not used by Timer1 or Timer2.

Note 1: Other (non power-up) resets include: external reset through $\overline{MCLR}$ and WDT Timer Reset.

12.2.1 ONE CAPTURE AND ONE PERIOD REGISTER MODE

In this mode registers PR3H/CA1H and PR3L/CA1L constitute a 16-bit period register. A block diagram is shown in Figure 12-7. The timer increments until it equals the period register and then resets to 0000h. TMR3 Interrupt Flag bit (TMR3IF) is set at this point. This interrupt can be disabled by clearing the TMR3 Interrupt Enable bit (TMR3IE). TMR3IF must be cleared in software.

This mode is selected if control bit CA1/$\overline{PR3}$ is clear. In this mode, the Capture1 register, consisting of high byte (PR3H/CA1H) and low byte (PR3L/CA1L), is configured as the period control register for TMR3. Capture1 is disabled in this mode, and the corresponding Interrupt bit CA1IF is never set. TMR3 increments until it equals the value in the period register and then resets to 0000h.

Capture2 is active in this mode. The CA2ED1 and CA2ED0 bits determine the event on which capture will occur. The possible events are:

- Capture on every falling edge
- Capture on every rising edge
- Capture every 4th rising edge
- Capture every 16th rising edge

When a capture takes place, an interrupt flag is latched into the CA2IF bit. This interrupt can be enabled by setting the corresponding mask bit CA2IE. The Peripheral Interrupt Enable bit (PEIE) must be set and the Global Interrupt Disable bit (GLINTD) must be cleared for the interrupt to be acknowledged. The CA2IF interrupt flag bit must be cleared in software.

When the capture prescale select is changed, the prescaler is not reset and an event may be generated. Therefore, the first capture after such a change will be ambiguous. However, it sets the time-base for the next capture. The prescaler is reset upon chip reset.

Capture pin RB1/CAP2 is a multiplexed pin. When used as a port pin, Capture2 is not disabled. However, the user can simply disable the Capture2 interrupt by clearing CA2IE. If RB1/CAP2 is used as an output pin, the user can activate a capture by writing to the port pin. This may be useful during development phase to emulate a capture interrupt.

The input on capture pin RB1/CAP2 is synchronized internally to internal phase clocks. This imposes certain restrictions on the input waveform (see the Electrical Specification section for timing).

The Capture2 overflow status flag bit is double buffered. The master bit is set if one captured word is already residing in the Capture2 register and another "event" has occurred on the RB1/CA2 pin. The new event will not transfer the Timer3 value to the capture register, protecting the previous unread capture value. When the user reads both the high and the low bytes (in any order) of the Capture2 register, the master overflow bit is transferred to the slave overflow bit (CA2OVF) and then the master bit is reset. The user can then read TCON2 to determine the value of CA2OVF.

The recommended sequence to read capture registers and capture overflow flag bits is shown in Example 12-1.

EXAMPLE 12-1: SEQUENCE TO READ CAPTURE REGISTERS

```
MOVLB 3               ;Select Bank 3
MOVPF CA2L,LO_BYTE    ;Read Capture2 low
                      ;byte, store in LO_BYTE
MOVPF CA2H,HI_BYTE    ;Read Capture2 high
                      ;byte, store in HI_BYTE
MOVPF TCON2,STAT_VAL  ;Read TCON2 into file
                      ;STAT_VAL
```

FIGURE 12-7: TIMER3 WITH ONE CAPTURE AND ONE PERIOD REGISTER BLOCK DIAGRAM

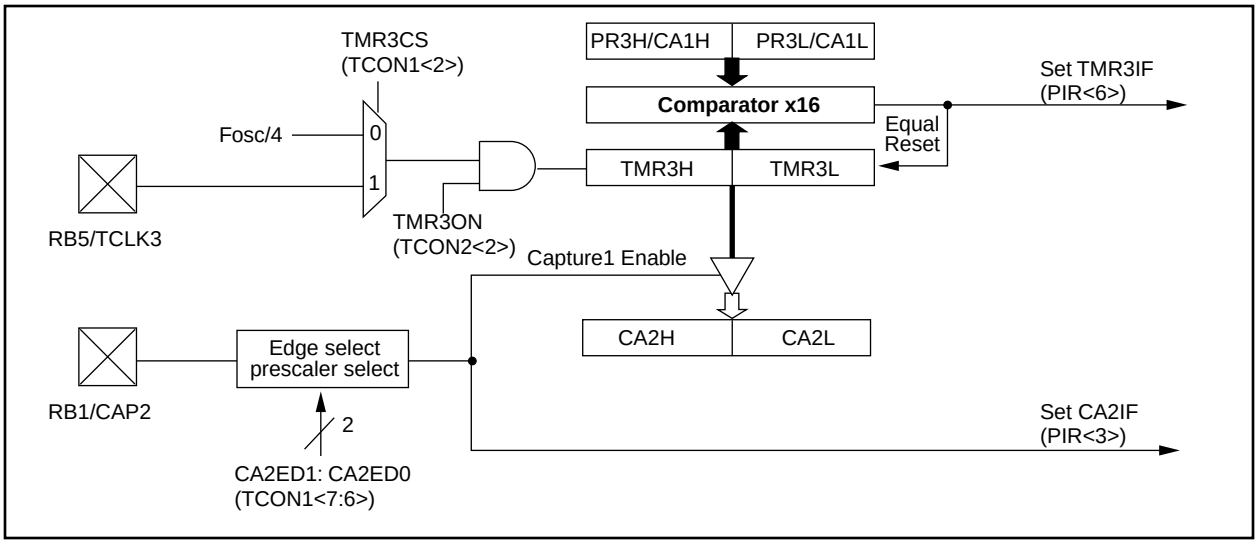


TABLE 13-4: BAUD RATES FOR ASYNCHRONOUS MODE

BAUD RATE (K)	FOSC = 33 MHz			FOSC = 25 MHz			FOSC = 20 MHz			FOSC = 16 MHz		
	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)
0.3	NA	—	—	NA	—	—	NA	—	—	NA	—	—
1.2	NA	—	—	NA	—	—	1.221	+1.73	255	1.202	+0.16	207
2.4	2.398	-0.07	214	2.396	0.14	162	2.404	+0.16	129	2.404	+0.16	103
9.6	9.548	-0.54	53	9.53	-0.76	40	9.469	-1.36	32	9.615	+0.16	25
19.2	19.09	-0.54	26	19.53	+1.73	19	19.53	+1.73	15	19.23	+0.16	12
76.8	73.66	-4.09	6	78.13	+1.73	4	78.13	+1.73	3	83.33	+8.51	2
96	103.12	+7.42	4	97.65	+1.73	3	104.2	+8.51	2	NA	—	—
300	257.81	-14.06	1	390.63	+30.21	0	312.5	+4.17	0	NA	—	—
500	515.62	+3.13	0	NA	—	—	NA	—	—	NA	—	—
HIGH	515.62	—	0	—	—	0	312.5	—	0	250	—	0
LOW	2.014	—	255	1.53	—	255	1.221	—	255	0.977	—	255

BAUD RATE (K)	FOSC = 10 MHz			FOSC = 7.159 MHz			FOSC = 5.068 MHz		
	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)
0.3	NA	—	—	NA	—	—	0.31	+3.13	255
1.2	1.202	+0.16	129	1.203	_0.23	92	1.2	0	65
2.4	2.404	+0.16	64	2.380	-0.83	46	2.4	0	32
9.6	9.766	+1.73	15	9.322	-2.90	11	9.9	-3.13	7
19.2	19.53	+1.73	7	18.64	-2.90	5	19.8	+3.13	3
76.8	78.13	+1.73	1	NA	—	—	79.2	+3.13	0
96	NA	—	—	NA	—	—	NA	—	—
300	NA	—	—	NA	—	—	NA	—	—
500	NA	—	—	NA	—	—	NA	—	—
HIGH	156.3	—	0	111.9	—	0	79.2	—	0
LOW	0.610	—	255	0.437	—	255	0.309	—	255

BAUD RATE (K)	FOSC = 3.579 MHz			FOSC = 1 MHz			FOSC = 32.768 kHz		
	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)	KBAUD	%ERROR	SPBRG value (decimal)
0.3	0.301	+0.23	185	0.300	+0.16	51	0.256	-14.67	1
1.2	1.190	-0.83	46	1.202	+0.16	12	NA	—	—
2.4	2.432	+1.32	22	2.232	-6.99	6	NA	—	—
9.6	9.322	-2.90	5	NA	—	—	NA	—	—
19.2	18.64	-2.90	2	NA	—	—	NA	—	—
76.8	NA	—	—	NA	—	—	NA	—	—
96	NA	—	—	NA	—	—	NA	—	—
300	NA	—	—	NA	—	—	NA	—	—
500	NA	—	—	NA	—	—	NA	—	—
HIGH	55.93	—	0	15.63	—	0	0.512	—	0
LOW	0.218	—	255	0.061	—	255	0.002	—	255

TABLE 13-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
16h, Bank 1	PIR	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TXIF	RCIF	0000 0010	0000 0010
13h, Bank 0	RCSTA	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
14h, Bank 0	RCREG	RX7	RX6	RX5	RX4	RX3	RX2	RX1	RX0	xxxx xxxx	uuuu uuuu
17h, Bank 1	PIE	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TXIE	RCIE	0000 0000	0000 0000
15h, Bank 0	TXSTA	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG	Baud rate generator register								xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented read as a '0', shaded cells are not used for synchronous master reception.

Note 1: Other (non power-up) resets include: external reset through $\overline{\text{MCLR}}$ and Watchdog Timer Reset.

FIGURE 14-8: WATCHDOG TIMER BLOCK DIAGRAM

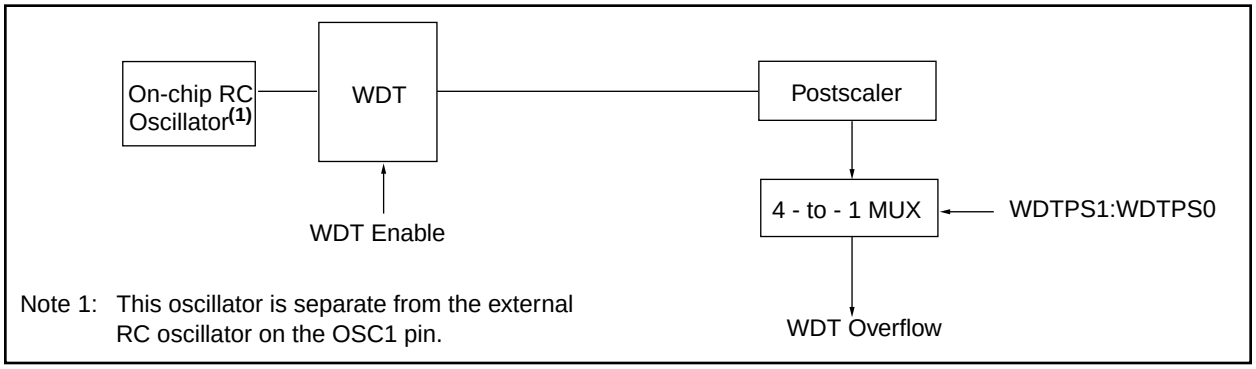


TABLE 14-4: REGISTERS/BITS ASSOCIATED WITH THE WATCHDOG TIMER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note1)
—	Config	—	PM1	—	PM0	WDTPS1	WDTPS0	FOSC1	FOSC0	(Note 2)	(Note 2)
06h, Unbanked	CPUSTA	—	—	STKAV	GLINTD	$\overline{TO}$	$\overline{PD}$	—	—	--11 11--	--11 qq--

Legend: - = unimplemented read as '0', q - value depends on condition, shaded cells are not used by the WDT.
Note 1: Other (non power-up) resets include: external reset through $\overline{MCLR}$ and Watchdog Timer Reset.
2: This value will be as the device was programmed, or if unprogrammed, will read as all '1's.

MOVPF **Move p to f**

Syntax: [*label*] MOVPF p,f

Operands: $0 \le f \le 255$
$0 \le p \le 31$

Operation: (p) → (f)

Status Affected: Z

Encoding:

010p	pppp	ffff	ffff

Description: Move data from data memory location 'p' to data memory location 'f'. Location 'f' can be anywhere in the 256 byte data space (00h to FFh) while 'p' can be 00h to 1Fh.

Either 'p' or 'f' can be WREG (a useful special situation).

MOVPF is particularly useful for transferring a peripheral register (e.g. the timer or an I/O port) to a data memory location. Both 'f' and 'p' can be indirectly addressed.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'p'	Execute	Write register 'f'

Example: MOVPF REG1, REG2

Before Instruction
REG1 = 0x11
REG2 = 0x33

After Instruction
REG1 = 0x11
REG2 = 0x11

MOVWF **Move WREG to f**

Syntax: [*label*] MOVWF f

Operands: $0 \le f \le 255$

Operation: (WREG) → (f)

Status Affected: None

Encoding:

0000	0001	ffff	ffff

Description: Move data from WREG to register 'f'. Location 'f' can be anywhere in the 256 word data space.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Execute	Write register 'f'

Example: MOVWF REG

Before Instruction
WREG = 0x4F
REG = 0xFF

After Instruction
WREG = 0x4F
REG = 0x4F

MPASM allow full symbolic debugging from the Microchip Universal Emulator System (PICMASTER).

MPASM has the following features to assist in developing software for specific use applications.

- Provides translation of Assembler source code to object code for all Microchip microcontrollers.
- Macro assembly capability.
- Produces all the files (Object, Listing, Symbol, and special) required for symbolic debug with Microchip's emulator systems.
- Supports Hex (default), Decimal and Octal source and listing formats.

MPASM provides a rich directive language to support programming of the PIC16/17. Directives are helpful in making the development of your assemble source code shorter and more maintainable.

16.11 Software Simulator (MPLAB-SIM)

The MPLAB-SIM Software Simulator allows code development in a PC host environment. It allows the user to simulate the PIC16/17 series microcontrollers on an instruction level. On any given instruction, the user may examine or modify any of the data areas or provide external stimulus to any of the pins. The input/output radix can be set by the user and the execution can be performed in; single step, execute until break, or in a trace mode.

MPLAB-SIM fully supports symbolic debugging using MPLAB-C and MPASM. The Software Simulator offers the low cost flexibility to develop and debug code outside of the laboratory environment making it an excellent multi-project software development tool.

16.12 C Compiler (MPLAB-C)

The MPLAB-C Code Development System is a complete 'C' compiler and integrated development environment for Microchip's PIC16/17 family of microcontrollers. The compiler provides powerful integration capabilities and ease of use not found with other compilers.

For easier source level debugging, the compiler provides symbol information that is compatible with the MPLAB IDE memory display (PICMASTER emulator software versions 1.13 and later).

16.13 Fuzzy Logic Development System (*fuzzy*TECH-MP)

*fuzzy*TECH-MP fuzzy logic development tool is available in two versions - a low cost introductory version, MP Explorer, for designers to gain a comprehensive working knowledge of fuzzy logic system design; and a full-featured version, *fuzzy*TECH-MP, edition for implementing more complex systems.

Both versions include Microchip's *fuzzy*LAB™ demonstration board for hands-on experience with fuzzy logic systems implementation.

16.14 MP-DriveWay™ – Application Code Generator

MP-DriveWay is an easy-to-use Windows-based Application Code Generator. With MP-DriveWay you can visually configure all the peripherals in a PIC16/17 device and, with a click of the mouse, generate all the initialization and many functional code modules in C language. The output is fully compatible with Microchip's MPLAB-C C compiler. The code produced is highly modular and allows easy integration of your own code. MP-DriveWay is intelligent enough to maintain your code through subsequent code generation.

16.15 SEEVAL® Evaluation and Programming System

The SEEVAL SEEPROM Designer's Kit supports all Microchip 2-wire and 3-wire Serial EEPROMs. The kit includes everything necessary to read, write, erase or program special features of any Microchip SEEPROM product including Smart Serials™ and secure serials. The Total Endurance™ Disk is included to aid in trade-off analysis and reliability calculations. The total kit can significantly reduce time-to-market and result in an optimized system.

16.16 TrueGauge® Intelligent Battery Management

The TrueGauge development tool supports system development with the MTA11200B TrueGauge Intelligent Battery Management IC. System design verification can be accomplished before hardware prototypes are built. User interface is graphically-oriented and measured data can be saved in a file for exporting to Microsoft Excel.

16.17 KEELOQ® Evaluation and Programming Tools

KEELOQ evaluation and programming tools support Microchips HCS Secure Data Products. The HCS evaluation kit includes an LCD display to show changing codes, a decoder to decode transmissions, and a programming interface to program test transmitters.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 17-12: MEMORY INTERFACE READ TIMING

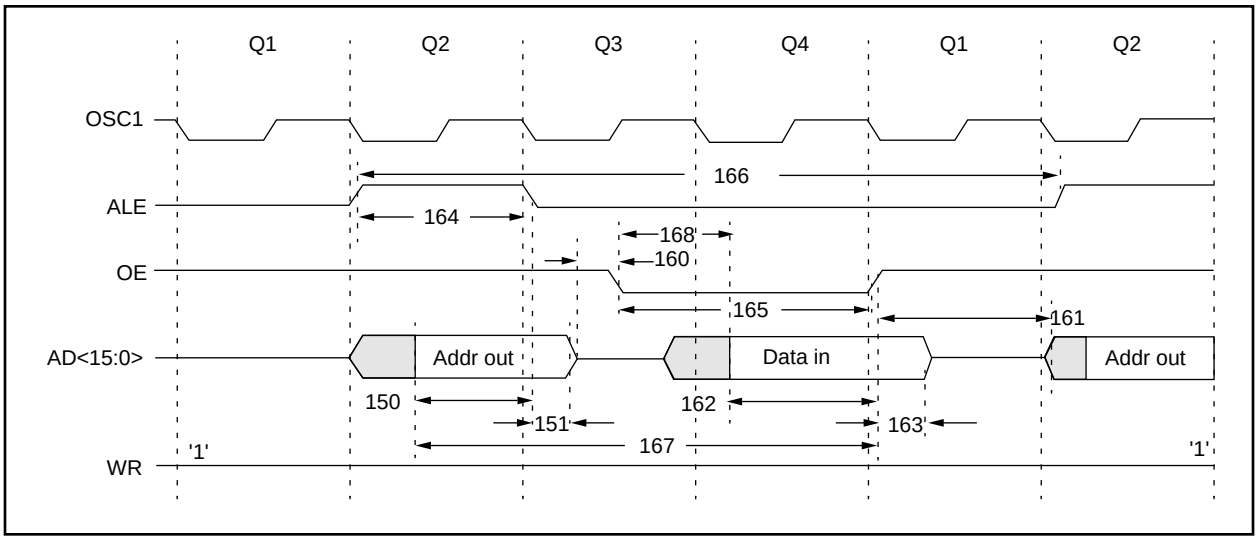


TABLE 17-12: MEMORY INTERFACE READ REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
150	TadV2alL	AD<15:0> (address) valid to ALE↓ (address setup time)	0.25Tcy - 30	—	—	ns	
151	TalL2adI	ALE↓ to address out invalid (address hold time)	5*	—	—	ns	
160	TadZ2oeL	AD<15:0> high impedance to $\overline{OE}$↓	0*	—	—	ns	
161	ToeH2adD	$\overline{OE}$↑ to AD<15:0> driven	0.25Tcy - 15	—	—	ns	
162	TadV2oeH	Data in valid before $\overline{OE}$↑ (data setup time)	35	—	—	ns	
163	ToeH2adI	$\overline{OE}$↑to data in invalid (data hold time)	0	—	—	ns	
164	TalH	ALE pulse width	—	0.25TCY §	—	ns	
165	ToeL	$\overline{OE}$ pulse width	0.5Tcy - 35 §	—	—	ns	
166	TalH2alH	ALE↑ to ALE↑ (cycle time)	—	TCY §	—	ns	
167	Tacc	Address access time	—	—	0.75 TCY-40	ns	
168	Toe	Output enable access time ($\overline{OE}$ low to Data Valid)	—	—	0.5 TCY - 60	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification guaranteed by design.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 18-19: V_{TH}, V_{IL} of I/O PINS (SCHMITT TRIGGER) vs. V_{DD}

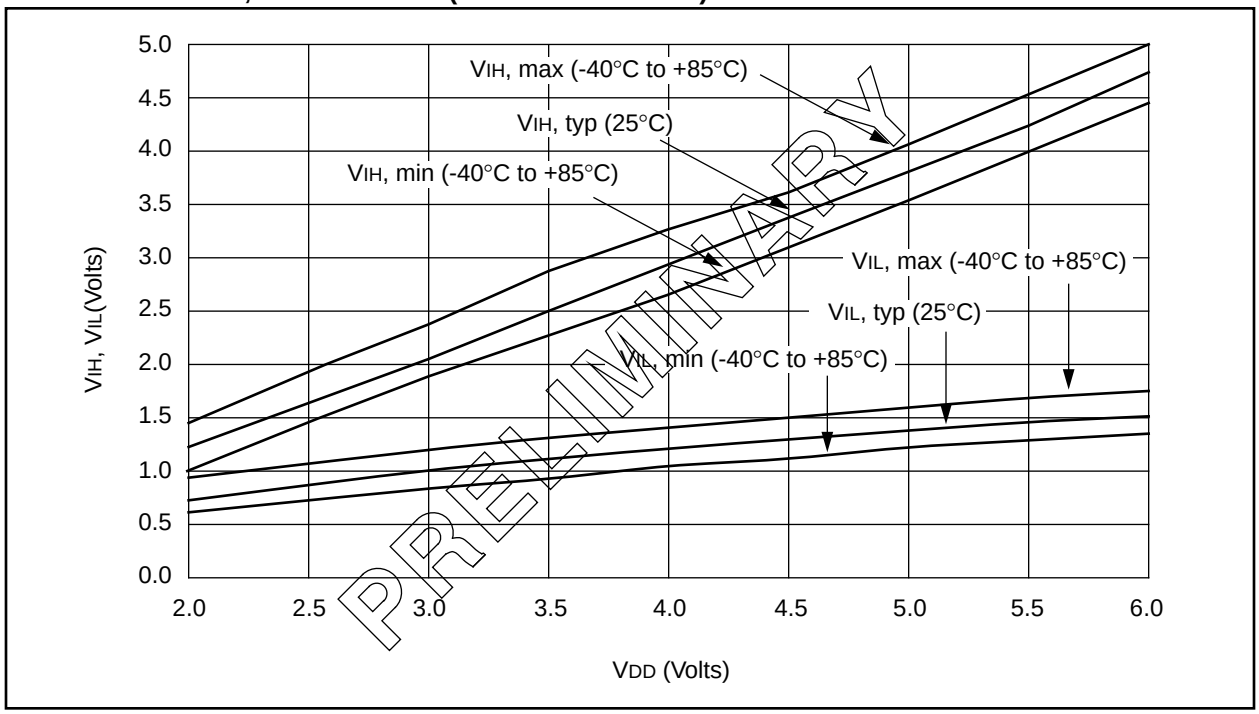
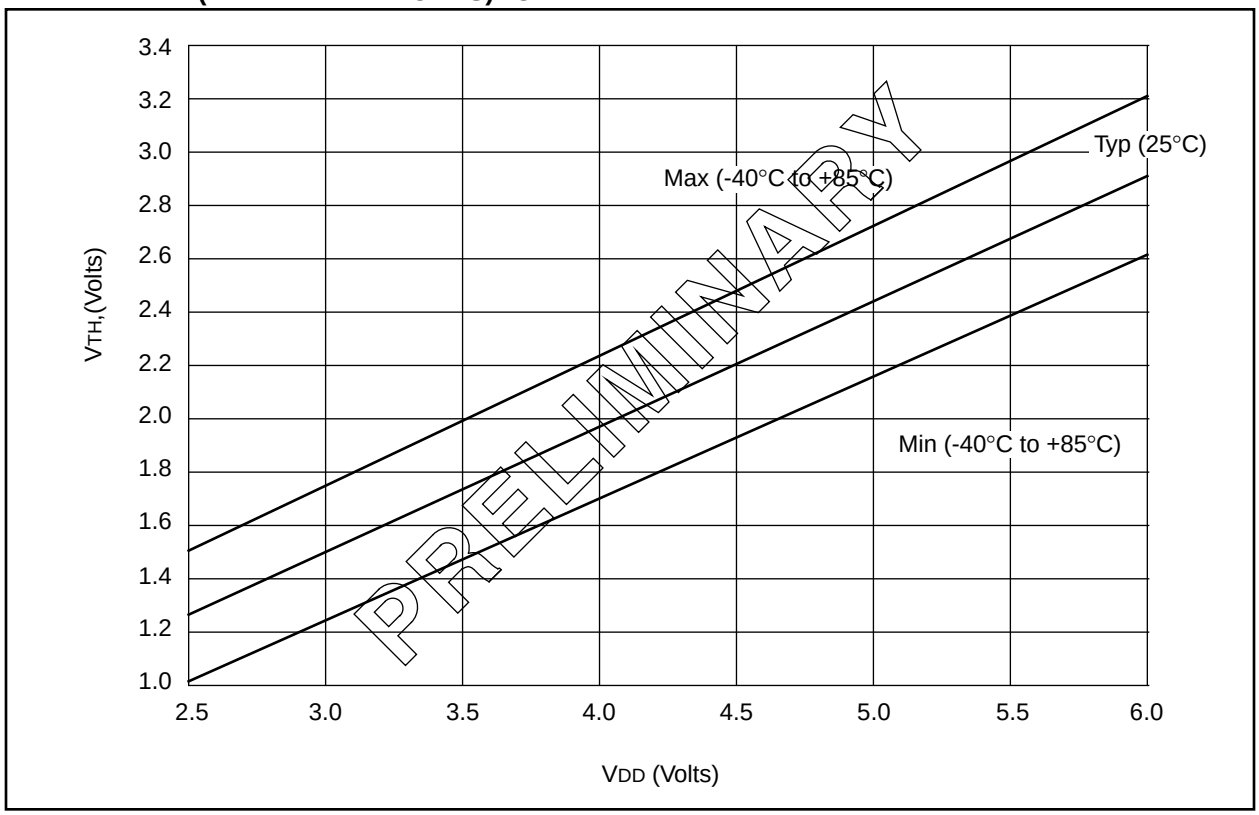


FIGURE 18-20: V_{TH} (INPUT THRESHOLD VOLTAGE) OF OSC1 INPUT (IN XT AND LF MODES) vs. V_{DD}



Applicable Devices	42	R42	42A	43	R43	44

19.0 PIC17CR42/42A/43/R43/44 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Ambient temperature under bias ..... -55 to +125˚C

Storage temperature ..... -65˚C to +150˚C

Voltage on V_{DD} with respect to V_{SS} ..... 0 to +7.5V

Voltage on $\overline{MCLR}$ with respect to V_{SS} (Note 2) ..... -0.6V to +14V

Voltage on RA2 and RA3 with respect to V_{SS} ..... -0.6V to +14V

Voltage on all other pins with respect to V_{SS} ..... -0.6V to V_{DD} + 0.6V

Total power dissipation (Note 1) 1.0W

Maximum current out of V_{SS} pin(s) - total ..... 250 mA

Maximum current into V_{DD} pin(s) - total ..... 200 mA

Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ±20 mA

Output clamp current, I_{OK} (V_O < 0 or V_O > V_{DD}) ..... ±20 mA

Maximum output current sunk by any I/O pin (except RA2 and RA3) ..... 35 mA

Maximum output current sunk by RA2 or RA3 pins ..... 60 mA

Maximum output current sourced by any I/O pin ..... 20 mA

Maximum current sunk by PORTA and PORTB (combined) ..... 150 mA

Maximum current sourced by PORTA and PORTB (combined) ..... 100 mA

Maximum current sunk by PORTC, PORTD and PORTE (combined) ..... 150 mA

Maximum current sourced by PORTC, PORTD and PORTE (combined) ..... 100 mA

Note 1: Power dissipation is calculated as follows: $Pdis = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD}-V_{OH}) \times I_{OH}\} + \sum(V_{OL} \times I_{OL})$

Note 2: Voltage spikes below V_{SS} at the $\overline{MCLR}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the $\overline{MCLR}$ pin rather than pulling this pin directly to V_{SS}.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 19-9: USART MODULE: SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

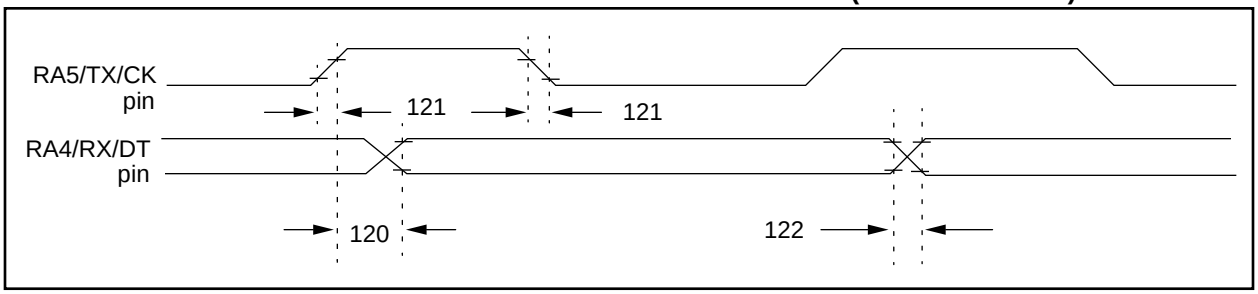


TABLE 19-9: SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
120	TckH2dtV	SYNC XMIT (MASTER & SLAVE) Clock high to data out valid	PIC17CR42/42A/43/R43/44	—	—	50	ns	
			PIC17LCR42/42A/43/R43/44	—	—	75	ns	
121	TckRF	Clock out rise time and fall time (Master Mode)	PIC17CR42/42A/43/R43/44	—	—	25	ns	
			PIC17LCR42/42A/43/R43/44	—	—	40	ns	
122	TdtRF	Data out rise time and fall time	PIC17CR42/42A/43/R43/44	—	—	25	ns	
			PIC17LCR42/42A/43/R43/44	—	—	40	ns	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 19-10: USART MODULE: SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

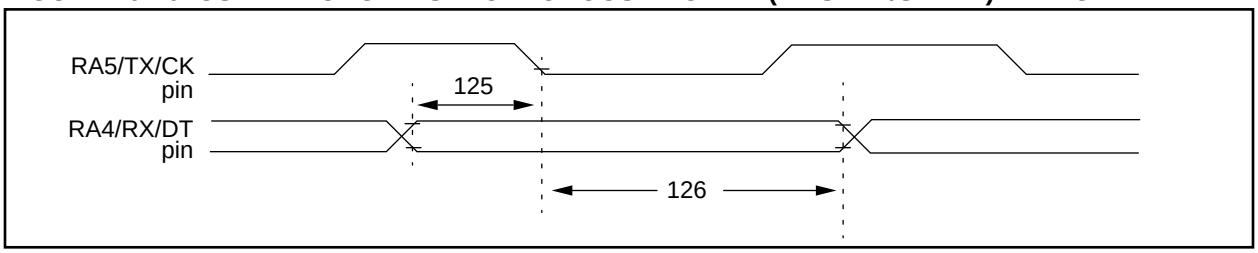


TABLE 19-10: SYNCHRONOUS RECEIVE REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
125	TdtV2ckL	SYNC RCV (MASTER & SLAVE) Data hold before CK↓ (DT hold time)	15	—	—	ns	
126	TckL2dtl	Data hold after CK↓ (DT hold time)	15	—	—	ns	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 20-4: TYPICAL RC OSCILLATOR FREQUENCY vs. VDD

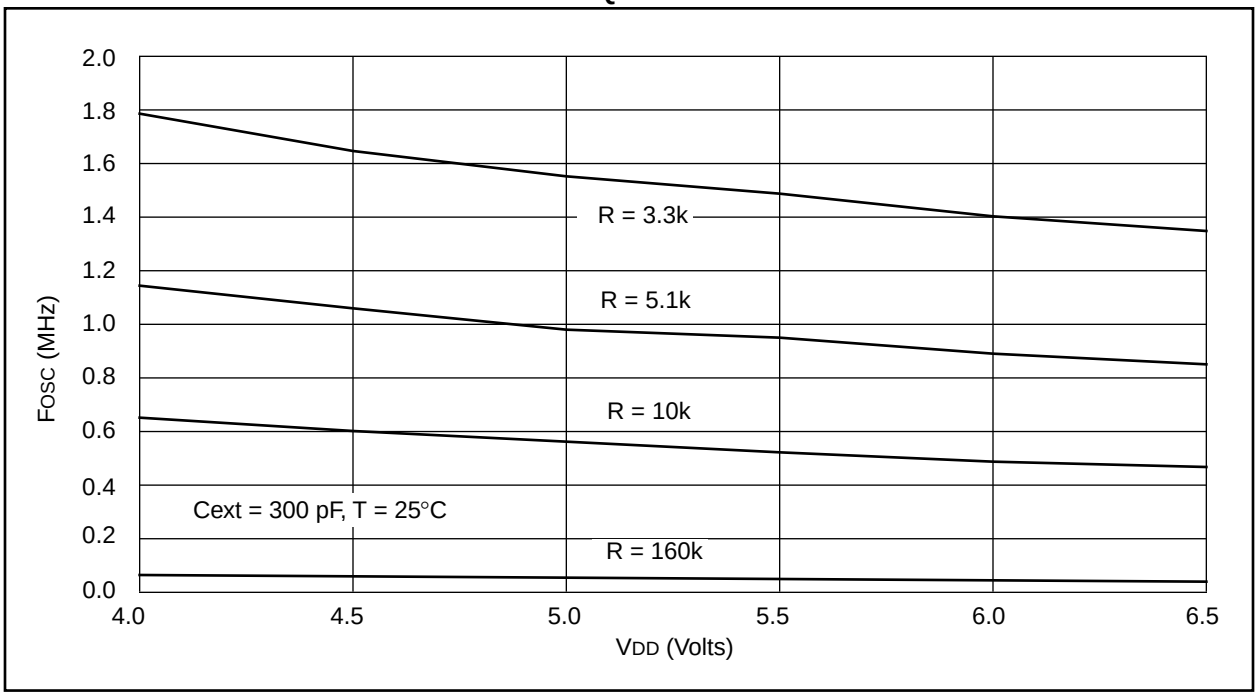


TABLE 20-2: RC OSCILLATOR FREQUENCIES

Cext	Rext	Average Fosc @ 5V, 25°C	
22 pF	10k	3.33 MHz	± 12%
	100k	353 kHz	± 13%
100 pF	3.3k	3.54 MHz	± 10%
	5.1k	2.43 MHz	± 14%
	10k	1.30 MHz	± 17%
	100k	129 kHz	± 10%
300 pF	3.3k	1.54 MHz	± 14%
	5.1k	980 kHz	± 12%
	10k	564 kHz	± 16%
	160k	35 kHz	± 18%

Applicable Devices	42	R42	42A	43	R43	44

FIGURE 20-17: I_{OL} vs. V_{OL}, V_{DD} = 5V

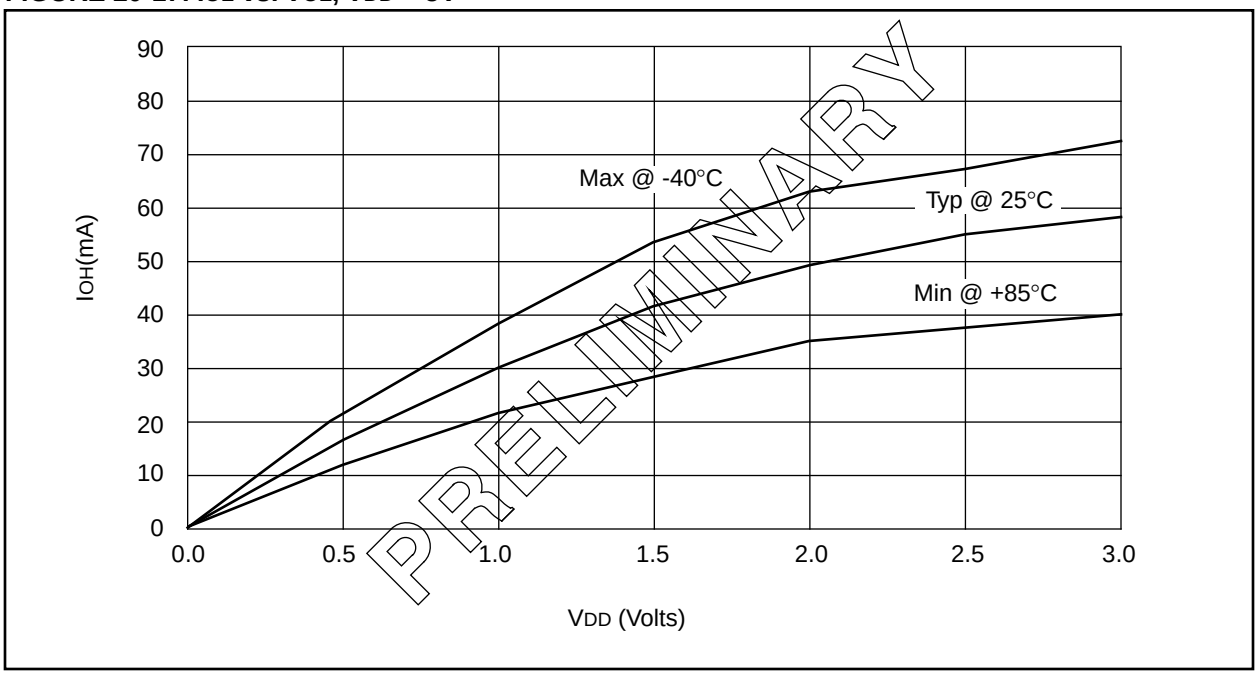
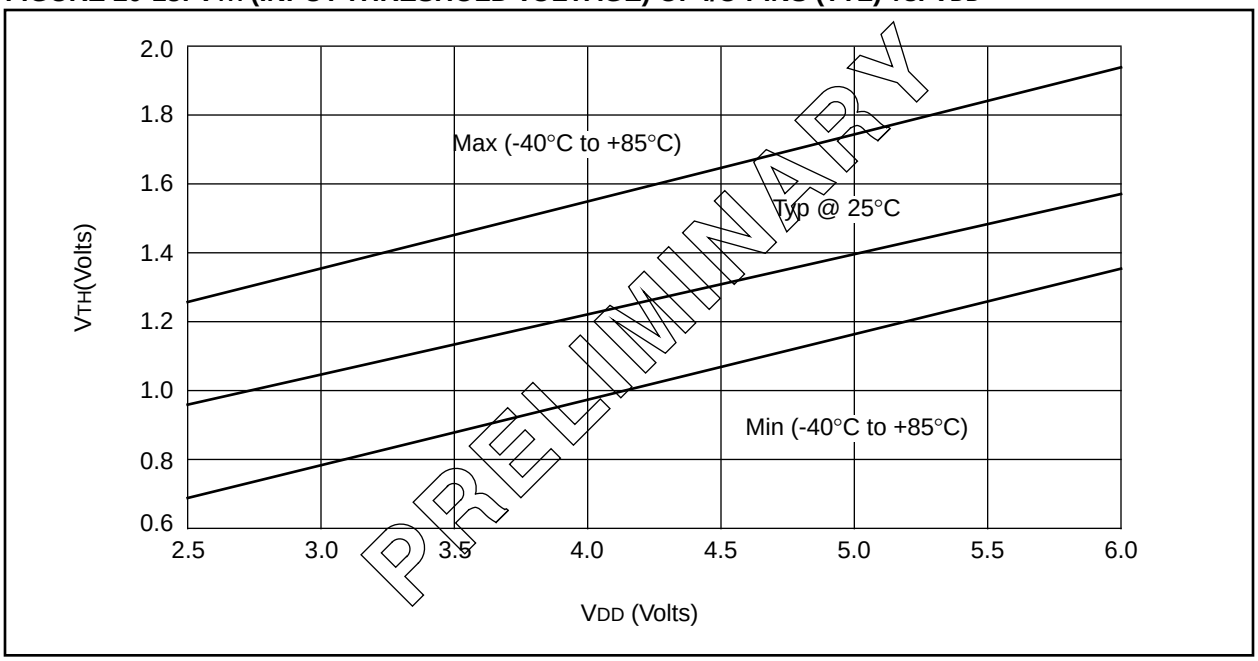


FIGURE 20-18: V_{TH} (INPUT THRESHOLD VOLTAGE) OF I/O PINS (TTL) vs. V_{DD}



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