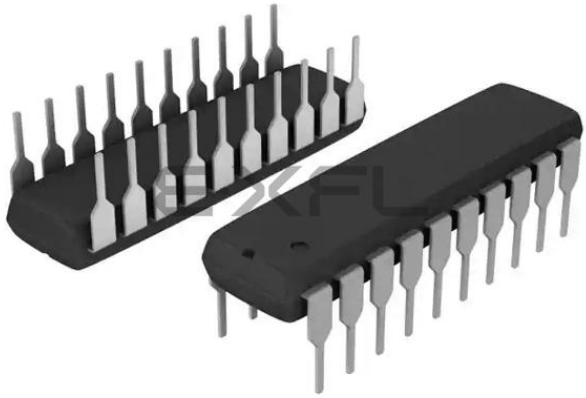

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	20MHz
Connectivity	I²C, UART/USART
Peripherals	Brown-out Detect/Reset, LED, POR, WDT
Number of I/O	18
Program Memory Size	4KB (4K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 6V
Data Converters	A/D 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	20-DIP (0.300", 7.62mm)
Supplier Device Package	20-DIP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/p87lpc767fn-112

Low power, low price, low pin count (20 pin) microcontroller with 4-kbyte OTP and 8-bit A/D converter

P87LPC767

Low power, low price, low pin count (20 pin) microcontroller with 4-kbyte OTP and 8-bit A/D converter

P87LPC767

GENERAL DESCRIPTION

The P87LPC767 is a 20-pin single-chip microcontroller designed for low pin count applications demanding high-integration, low cost solutions over a wide range of performance requirements. A member of the Philips low pin count family, the P87LPC767 offers programmable oscillator configurations for high and low speed crystals or RC operation, wide operating voltage range, programmable port output configurations, selectable Schmitt trigger inputs, LED drive outputs, and a built-in watchdog timer. The P87LPC767 is based on an accelerated 80C51 processor architecture that executes instructions at twice the rate of standard 80C51 devices.

FEATURES

- An accelerated 80C51 CPU provides instruction cycle times of 300–600 ns for all instructions except multiply and divide when executing at 20 MHz. Execution at up to 20 MHz when V_{DD} = 4.5 V to 6.0 V, 10 MHz when V_{DD} = 2.7 V to 6.0 V.
- Four-channel multiplexed 8-bit A/D converter. Conversion time of 9.3 μS at f_{OSC} = 20 MHz.
- 2.7 V to 6.0 V operating range for digital functions.
- 4 K bytes EPROM code memory.
- 128 byte RAM data memory.
- 32-byte customer code EPROM allows serialization of devices, storage of setup parameters, etc.
- Two 16-bit counter/timers. Each timer may be configured to toggle a port output upon timer overflow.
- Two analog comparators.
- Full duplex UART.
- I^2C communication port.
- Eight keypad interrupt inputs, plus two additional external interrupt inputs.
- Four interrupt priority levels.
- Watchdog timer with separate on-chip oscillator, requiring no external components. The watchdog timeout time is selectable from 8 values.
- Active low reset. On-chip power-on reset allows operation with no external reset components.
- Low voltage reset. One of two preset low voltage levels may be selected to allow a graceful system shutdown when power fails. May optionally be configured as an interrupt.
- Oscillator Fail Detect. The watchdog timer has a separate fully on-chip oscillator, allowing it to perform an oscillator fail detect function.
- Configurable on-chip oscillator with frequency range and RC oscillator options (selected by user programmed EPROM bits). The RC oscillator option allows operation with no external oscillator components.
- Programmable port output configuration options: quasi-bidirectional, open drain, push-pull, input-only.
- Selectable Schmitt trigger port inputs.
- LED drive capability (20 mA) on all port pins.
- Controlled slew rate port outputs to reduce EMI. Outputs have approximately 10 ns minimum ramp times.
- 15 I/O pins minimum. Up to 18 I/O pins using on-chip oscillator and reset options.
- Only power and ground connections are required to operate the P87LPC767 when fully on-chip oscillator and reset options are selected.
- Serial EPROM programming allows simple in-circuit production coding. Two EPROM security bits prevent reading of sensitive application programs.
- Idle and Power Down reduced power modes. Improved wakeup from Power Down mode (a low interrupt input starts execution). Typical Power Down current is 1 μA.
- 20-pin DIP and SO packages.

PIN DESCRIPTIONS

MNEMONIC	PIN NO.	TYPE	NAME AND FUNCTION		
P0.0–P0.7	1, 13, 14, 16–20	I/O	**Port 0**: Port 0 is an 8-bit I/O port with a user-configurable output type. Port 0 latches are configured in the quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of port 0 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to the section on I/O port configuration and the DC Electrical Characteristics for details. The Keyboard Interrupt feature operates with port 0 pins. Port 0 also provides various special functions as described below.		
	1	O	**P0.0**	**CMP2**	Comparator 2 output.
	20	I	**P0.1**	**CIN2B**	Comparator 2 positive input B.
	19	I	**P0.2**	**CIN2A**	Comparator 2 positive input A.
	18	I	**P0.3**	**CIN1B**	Comparator 1 positive input B.
		I		**AD0**	A/D channel 0 input.
	17	I	**P0.4**	**CIN1A**	Comparator 1 positive input A.
		I		**AD1**	A/D channel 1 input.
	16	I	**P0.5**	**CMPREF**	Comparator reference (negative) input.
		I		**AD2**	A/D channel 2 input.
	14	O	**P0.6**	**CMP1**	Comparator 1 output.
		I		**AD3**	A/D channel 3 input.
	13	I/O	**P0.7**	**T1**	Timer/counter 1 external count input or overflow output.
P1.0–P1.7	2–4, 8–12	I/O	**Port 1**: Port 1 is an 8-bit I/O port with a user-configurable output type, except for three pins as noted below. Port 1 latches are configured in the quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of the configurable port 1 pins as inputs and outputs depends upon the port configuration selected. Each of the configurable port pins are programmed independently. Refer to the section on I/O port configuration and the DC Electrical Characteristics for details. Port 1 also provides various special functions as described below.		
	12	O	**P1.0**	**TxD**	Transmitter output for the serial port.
	11	I	**P1.1**	**RxD**	Receiver input for the serial port.
	10	I/O	**P1.2**	**T0**	Timer/counter 0 external count input or overflow output.
		I/O		**SCL**	I^2C serial clock input/output. When configured as an output, P1.2 is open drain, in order to conform to I^2C specifications.
	9	I	**P1.3**	**$\overline{\text{INT0}}$**	External interrupt 0 input.
		I/O		**SDA**	I^2C serial data input/output. When configured as an output, P1.3 is open drain, in order to conform to I^2C specifications.
	8	I	**P1.4**	**$\overline{\text{INT1}}$**	External interrupt 1 input.
	4	I	**P1.5**	**$\overline{\text{RST}}$**	External Reset input (if selected via EPROM configuration). A low on this pin resets the microcontroller, causing I/O ports and peripherals to take on their default states, and the processor begins execution at address 0. When used as a port pin, P1.5 is a Schmitt trigger input only.

P2.0–P2.1	6, 7	I/O	**Port 2**: Port 2 is a 2-bit I/O port with a user-configurable output type. Port 2 latches are configured in the quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of port 2 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to the section on I/O port configuration and the DC Electrical Characteristics for details. Port 2 also provides various special functions as described below.		
	7	O	**P2.0**	**X2**	Output from the oscillator amplifier (when a crystal oscillator option is selected via the EPROM configuration).
				CLKOUT	CPU clock divided by 6 clock output when enabled via SFR bit and in conjunction with internal RC oscillator or external clock input.
	6	I	**P2.1**	**X1**	Input to the oscillator circuit and internal clock generator circuits (when selected via the EPROM configuration).
V_{SS}	5	I	**Ground**: 0 V reference.		
V_{DD}	15	I	**Power Supply**: This is the power supply voltage for normal operation as well as Idle and Power Down modes.		

SPECIAL FUNCTION REGISTERS

Name	Description	SFR Address	Bit Functions and Addresses MSB							LSB	Reset Value
			E7	E6	E5	E4	E3	E2	E1	E0	
ACC*	Accumulator	E0h									00h
			C7	C6	C5	C4	C3	C2	C1	C0	
ADCON#*	A/D Control	C0h	ENADC	–	–	ADCI	ADCS	RCCLK	AADR1	AADR0	00h
AUXR1#	Auxiliary Function Register	A2h	KBF	BOD	BOI	LPEP	SRST	0	–	DPS	02h[1]
			F7	F6	F5	F4	F3	F2	F1	F0	
B*	B register	F0h									00h
CMP1#	Comparator 1 control register	ACh	–	–	CE1	CP1	CN1	OE1	CO1	CMF1	00h[1]
CMP2#	Comparator 2 control register	ADh	–	–	CE2	CP2	CN2	OE2	CO2	CMF2	00h[1]
DAC0#	A/D Result	C5h									00h
DIVM#	CPU clock divide-by-M control	95h									00h
DPTR:	Data pointer (2 bytes)										
DPH	Data pointer high byte	83h									00h
DPL	Data pointer low byte	82h									00h
			CF	CE	CD	CC	CB	CA	C9	C8	
I2CFG#*	I^2C configuration register	C8h/RD	SLAVEN	MASTRQ	0	TIRUN	–	–	CT1	CT0	00h[1]
		C8h/WR	SLAVEN	MASTRQ	CLRTI	TIRUN	–	–	CT1	CT0	
			DF	DE	DD	DC	DB	DA	D9	D8	
I2CON#*	I^2C control register	D8h/RD	RDAT	ATN	DRDY	ARL	STR	STP	MASTER	–	80h[1]
		D8h/WR	CXA	IDLE	CDR	CARL	CSTR	CSTP	XSTR	XSTP	
I2DAT#	I^2C data register	D9h/RD	RDAT	0	0	0	0	0	0	0	80h
		D9h/WR	XDAT	x	x	x	x	x	x	x	
			AF	AE	AD	AC	AB	AA	A9	A8	
IEN0*	Interrupt enable 0	A8h	EA	EWD	EBO	ES	ET1	EX1	ET0	EX0	00h
			EF	EE	ED	EC	EB	EA	E9	E8	
IEN1#*	Interrupt enable 1	E8h	ETI	–	EC1	EAD	–	EC2	EKB	EI2	00h[1]
			BF	BE	BD	BC	BB	BA	B9	B8	
IP0*	Interrupt priority 0	B8h	–	PWD	PBO	PS	PT1	PX1	PT0	PX0	00h[1]
IP0H#	Interrupt priority 0 high byte	B7h	–	PWDH	PBOH	PSH	PT1H	PX1H	PT0H	PX0H	00h[1]

Table 1. Example A/D Conversion Times

CPU Clock Rate	RCCLK = 0	RCCLK = 1		
		minimum	nominal	maximum
32 kHz	NA	563.4 µs	659 µs	757 µs
1 MHz	186 µs	32.4 µs	39.3 µs	48.9 µs
4 MHz	46.5 µs	18.9 µs	23.6 µs	30.1 µs
11.0592 MHz	16.8 µs	16 µs	20.2 µs	27.1 µs
12 MHz	15.5 µs			
16 MHz	11.6 µs			
20 MHz	9.3 µs			

Note: Do not clock ADC from the RC oscillator when MCU clock is greater than 4 MHz.

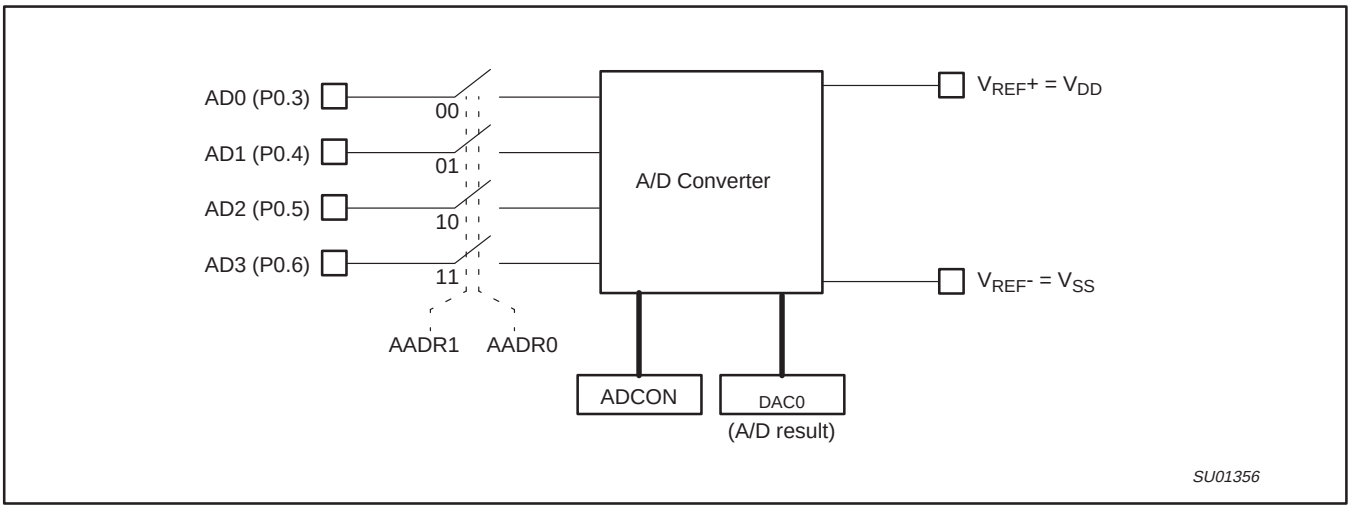

Figure 3. A/D Converter Connections

The A/D in Power Down and Idle Modes

While using the CPU clock as the A/D clock source, the Idle mode may be used to conserve power and/or to minimize system noise during the conversion. CPU operation will resume and Idle mode terminate automatically when a conversion is complete if the A/D interrupt is active. In Idle mode, noise from the CPU itself is eliminated, but noise from the oscillator and any other on-chip peripherals that are running will remain.

The CPU may be put into Power Down mode when the A/D is clocked by the on-chip RC oscillator (RCCLK = 1). This mode gives the best possible A/D accuracy by eliminating most on-chip noise sources.

If the Power Down mode is entered while the A/D is running from the CPU clock (RCCLK = 0), the A/D will abort operation and will not wake up the CPU. The contents of DAC0 will be invalid when operation does resume.

When an A/D conversion is started, Power Down or Idle mode must be activated within two machine cycles in order to have the most accurate A/D result. These two machine cycles are counted at the CPU clock rate. When using the A/D with either Power Down or Idle mode, care must be taken to insure that the CPU is not restarted by another interrupt until the A/D conversion is complete. The possible causes of wakeup are different in Power Down and Idle modes.

A/D accuracy is also affected by noise generated elsewhere in the application, power supply noise, and power supply regulation. Since the P87LPC767 power pins are also used as the A/D reference and supply, the power supply has a very direct affect on the accuracy of A/D readings. Using the A/D without Power Down mode while the clock is divided through the use of CLKR or DIVM has an adverse effect on A/D accuracy.

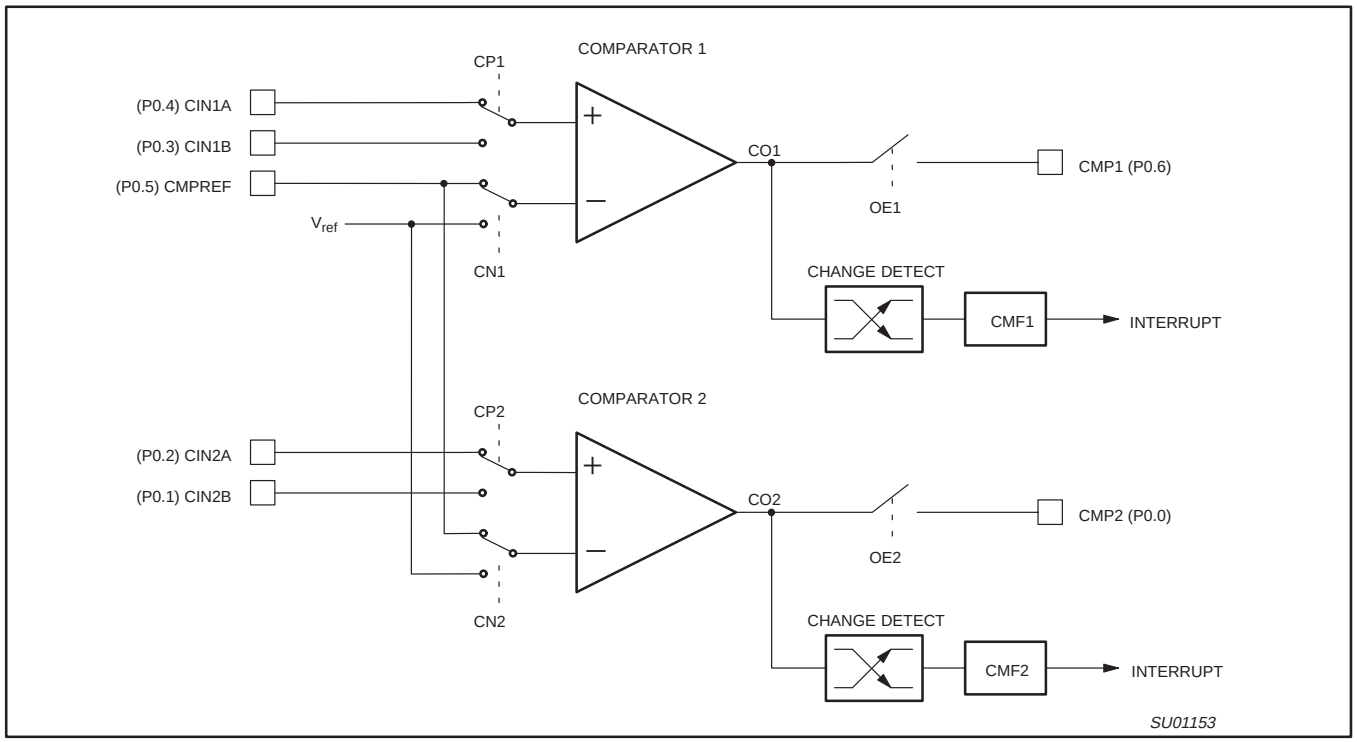

Figure 5. Comparator Input and Output Connections

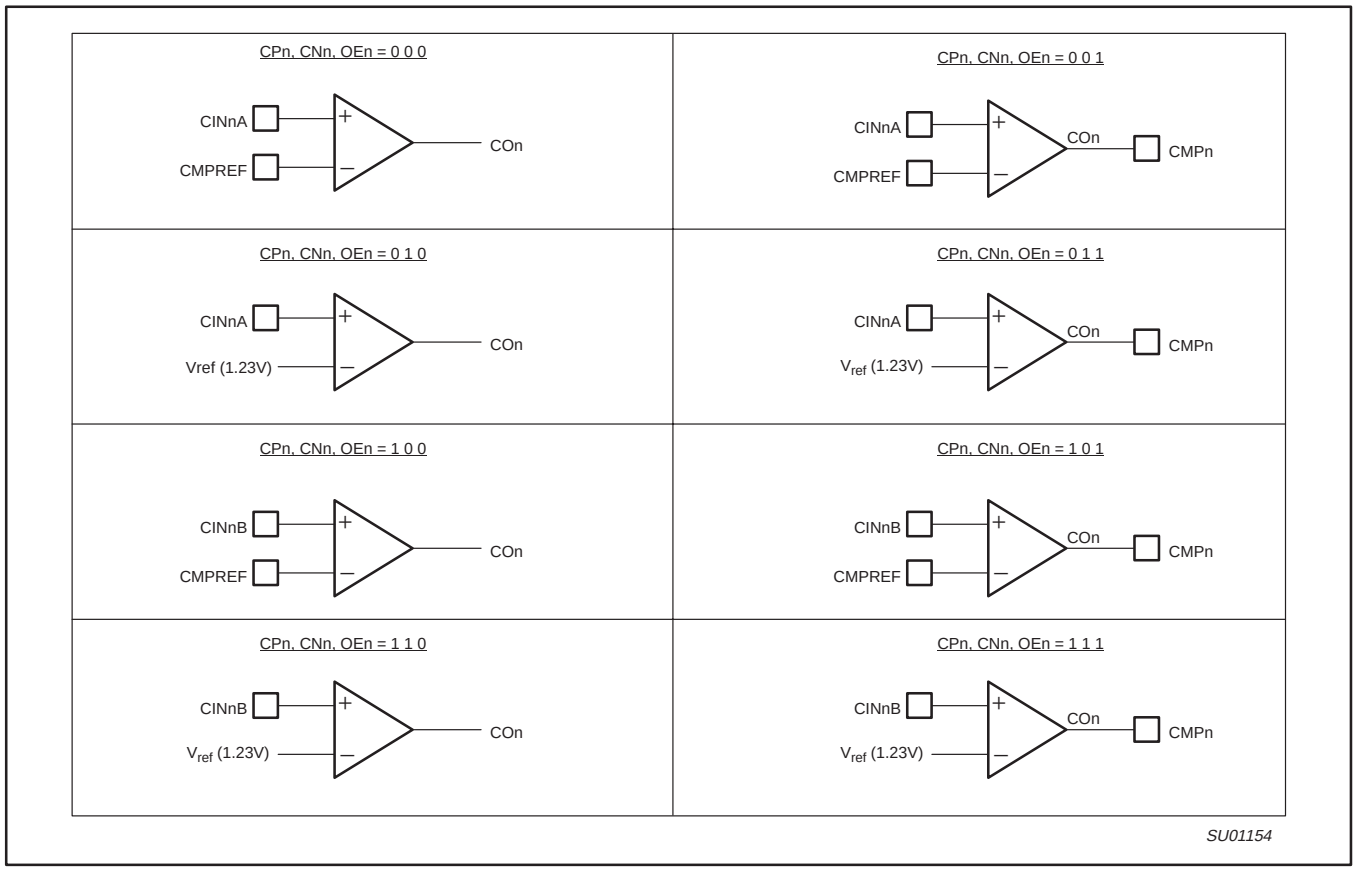

Figure 6. Comparator Configurations

I²C Serial Interface

The I²C bus uses two wires (SDA and SCL) to transfer information between devices connected to the bus. The main features of the bus are:

- Bidirectional data transfer between masters and slaves.
- Serial addressing of slaves (no added wiring).
- Acknowledgment after each transferred byte.
- Multimaster bus.
- Arbitration between simultaneously transmitting masters without corruption of serial data on bus.

The I²C subsystem includes hardware to simplify the software required to drive the I²C bus. The hardware is a single bit interface which in addition to including the necessary arbitration and framing error checks, includes clock stretching and a bus timeout timer. The interface is synchronized to software either through polled loops or interrupts.

Refer to the application note AN422, entitled “Using the 8XC751 Microcontroller as an I²C Bus Master” for additional discussion of the 8xC76x I²C interface and sample driver routines.

The P87LPC767 I2C implementation duplicates that of the 87C751 and 87C752 except for the following details:

- The interrupt vector addresses for both the I²C interrupt and the Timer I interrupt.
- The I²C SFR addresses (I2CON, !2CFG, I2DAT).
- The location of the I²C interrupt enable bit and the name of the SFR it is located within (EI2 is Bit 0 in IEN1).
- The location of the Timer I interrupt enable bit and the name of the SFR it is located within (ETI is Bit 7 in IEN1).
- The I²C and Timer I interrupts have a settable priority.

Timer I is used to both control the timing of the I²C bus and also to detect a “bus locked” condition, by causing an interrupt when nothing happens on the I²C bus for an inordinately long period of time while a transmission is in progress. If this interrupt occurs, the program has the opportunity to attempt to correct the fault and resume I²C operation.

Six time spans are important in I²C operation and are insured by timer I:

- The MINIMUM HIGH time for SCL when this device is the master.
- The MINIMUM LOW time for SCL when this device is a master. This is not very important for a single-bit hardware interface like this one, because the SCL low time is stretched until the software responds to the I²C flags. The software response time normally meets or exceeds the MIN LO time. In cases where the software responds within MIN HI + MIN LO) time, timer I will ensure that the minimum time is met.
- The MINIMUM SCL HIGH TO SDA HIGH time in a stop condition.
- The MINIMUM SDA HIGH TO SDA LOW time between I²C stop and start conditions (4.7ms, see I²C specification).
- The MINIMUM SDA LOW TO SCL LOW time in a start condition.
- The MAXIMUM SCL CHANGE time while an I²C frame is in progress. A frame is in progress between a start condition and the following stop condition. This time span serves to detect a lack of software response on this device as well as external I²C problems. SCL “stuck low” indicates a faulty master or slave. SCL “stuck high” may mean a faulty device, or that noise induced onto the I²C bus caused all masters to withdraw from I²C arbitration.

The first five of these times are 4.7 ms (see I²C specification) and are covered by the low order three bits of timer I. Timer I is clocked by the P87LPC767 CPU clock. Timer I can be pre-loaded with one of four values to optimize timing for different oscillator frequencies. At lower frequencies, software response time is increased and will degrade maximum performance of the I²C bus. See special function register I2CFG description for prescale values (CT0, CT1).

The MAXIMUM SCL CHANGE time is important, but its exact span is not critical. The complete 10 bits of timer I are used to count out the maximum time. When I²C operation is enabled, this counter is cleared by transitions on the SCL pin. The timer does not run between I²C frames (i.e., whenever reset or stop occurred more recently than the last start). When this counter is running, it will carry out after 1020 to 1023 machine cycles have elapsed since a change on SCL. A carry out causes a hardware reset of the I²C interface and generates an interrupt if the Timer I interrupt is enabled. In cases where the bus hang-up is due to a lack of software response by this device, the reset releases SCL and allows I²C operation among other devices to continue.

Timer I is enabled to run, and will reset the I²C interface upon overflow, if the TIRUN bit in the I2CFG register is set. The Timer I interrupt may be enabled via the ETI bit in IEN1, and its priority set by the PTIH and PTI bits in the Ip1H and IP1 registers respectively.

I²C Interrupts

If I²C interrupts are enabled (EA and EI2 are both set to 1), an I²C interrupt will occur whenever the ATN flag is set by a start, stop, arbitration loss, or data ready condition (refer to the description of ATN following). In practice, it is not efficient to operate the I²C interface in this fashion because the I²C interrupt service routine would somehow have to distinguish between hundreds of possible conditions. Also, since I²C can operate at a fairly high rate, the software may execute faster if the code simply waits for the I²C interface.

Typically, the I²C interrupt should only be used to indicate a start condition at an idle slave device, or a stop condition at an idle master device (if it is waiting to use the I²C bus). This is accomplished by enabling the I²C interrupt only during the aforementioned conditions.

Reading I2CON

RDAT — The data from SDA is captured into “Receive DATa” whenever a rising edge occurs on SCL. RDAT is also available (with seven low-order zeros) in the I2DAT register. The difference between reading it here and there is that reading I2DAT clears DRDY, allowing the I²C to proceed on to another bit. Typically, the first seven bits of a received byte are read from I2DAT, while the 8th is read here. Then I2DAT can be written to send the Acknowledge bit and clear DRDY.

ATN — “ATteNtion” is 1 when one or more of DRDY, ARL, STR, or STP is 1. Thus, ATN comprises a single bit that can be tested to release the I²C service routine from a “wait loop.”

DRDY — “Data ReaDY” (and thus ATN) is set when a rising edge occurs on SCL, except at idle slave. DRDY is cleared by writing CDR = 1, or by writing or reading the I2DAT register. The following low period on SCL is stretched until the program responds by clearing DRDY.

Table 2. Interaction of TIRUN with SLAVEN, MASTRQ, and MASTER

SLAVEN, MASTRQ, MASTER	TIRUN	OPERATING MODE
All 0	0	The I^2C interface is disabled. Timer I is cleared and does not run. This is the state assumed after a reset. If an I^2C application wants to ignore the I^2C at certain times, it should write SLAVEN, MASTRQ, and TIRUN all to zero.
All 0	1	The I^2C interface is disabled.
Any or all 1	0	The I^2C interface is enabled. The 3 low-order bits of Timer I run for min-time generation, but the hi-order bits do not, so that there is no checking for I^2C being “hung.” This configuration can be used for very slow I^2C operation.
Any or all 1	1	The I^2C interface is enabled. Timer I runs during frames on the I^2C, and is cleared by transitions on SCL, and by Start and Stop conditions. This is the normal state for I^2C operation.

Table 3. CT1, CT0 Values

CT1, CT0	Min Time Count (Machine Cycles)	CPU Clock Max (for 100 kHz I^2C)	Timeout Period (Machine Cycles)
1 0	7	8.4 MHz	1023
0 1	6	7.2 MHz	1022
0 0	5	6.0 MHz	1021
1 1	4	4.8 MHz	1020

Interrupts

The P87LPC767 uses a four priority level interrupt structure. This allows great flexibility in controlling the handling of the P87LPC767’s many interrupt sources. The P87LPC767 supports up to 13 interrupt sources.

Each interrupt source can be individually enabled or disabled by setting or clearing a bit in registers IEN0 or IEN1. The IEN0 register also contains a global disable bit, EA, which disables all interrupts at once.

Each interrupt source can be individually programmed to one of four priority levels by setting or clearing bits in the IP0, IP0H, IP1, and IP1H registers. An interrupt service routine in progress can be interrupted by a higher priority interrupt, but not by another interrupt of the same or lower priority. The highest priority interrupt service cannot be interrupted by any other interrupt source. So, if two requests of different priority levels are received simultaneously, the request of higher priority level is serviced.

If requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. This is called the arbitration ranking. Note that the arbitration ranking is only used to resolve simultaneous requests of the same priority level.

Table 3 summarizes the interrupt sources, flag bits, vector addresses, enable bits, priority bits, arbitration ranking, and whether each interrupt may wake up the CPU from Power Down mode.

Table 4. Summary of Interrupts

Description	Interrupt Flag Bit(s)	Vector Address	Interrupt Enable Bit(s)	Interrupt Priority	Arbitration Ranking	Power Down Wakeup
External Interrupt 0	IE0	0003h	EX0 (IEN0.0)	IP0H.0, IP0.0	1 (highest)	Yes
Timer 0 Interrupt	TF0	000Bh	ET0 (IEN0.1)	IP0H.1, IP0.1	4	No
External Interrupt 1	IE1	0013h	EX1 (IEN0.2)	IP0H.2, IP0.2	7	Yes
Timer 1 Interrupt	TF1	001Bh	ET1 (IEN0.3)	IP0H.3, IP0.3	10	No
Serial Port Tx and Rx	TI & RI	0023h	ES (IEN0.4)	IP0H.4, IP0.4	12	No
Brownout Detect	BOF	002Bh	EBO (IEN0.5)	IP0H.5, IP0.5	2	Yes
I^2C Interrupt	ATN	0033h	EI2 (IEN1.0)	IP1H.0, IP1.0	5	No
KBI Interrupt	KBF	003Bh	EKB (IEN1.1)	IP1H.1, IP1.1	8	Yes
Comparator 2 interrupt	CMF2	0043h	EC2 (IEN1.2)	IP1H.2, IP1.2	11	Yes
Watchdog Timer	WDOVF	0053h	EWD (IEN0.6)	IP0H.6, IP0.6	3	Yes
A/D Converter	ADCI	005Bh	EAD (IEN1.4)	IP1H.4, IP1.4	6	Yes
Comparator 1 interrupt	CMF1	0063h	EC1 (IEN1.5)	IP1H.5, IP1.5	9	Yes
Timer I	–	0073h	ETI (IEN1.7)	IP1H.7, IP1.7	13 (lowest)	No

Open Drain Output Configuration

The open drain output configuration turns off all pull-ups and only drives the pull-down transistor of the port driver when the port latch contains a logic 0. To be used as a logic output, a port configured in this manner must have an external pull-up, typically a resistor tied to V_{DD}. The pull-down for this mode is the same as for the quasi-bidirectional mode.

The open drain port configuration is shown in Figure 13.

Push-Pull Output Configuration

The push-pull output configuration has the same pull-down structure as both the open drain and the quasi-bidirectional output modes, but provides a continuous strong pull-up when the port latch contains a logic 1. The push-pull mode may be used when more source current is needed from a port output.

The push-pull port configuration is shown in Figure 14.

The three port pins that cannot be configured are P1.2, P1.3, and P1.5. The port pins P1.2 and P1.3 are permanently configured as open drain outputs. They may be used as inputs by writing ones to their respective port latches. P1.5 may be used as a Schmitt trigger input if the P87LPC767 has been configured for an internal reset and is not using the external reset input function $\overline{RST}$.

Additionally, port pins P2.0 and P2.1 are disabled for both input and output if one of the crystal oscillator options is chosen. Those options are described in the Oscillator section.

The value of port pins at reset is determined by the PRHI bit in the UCFG1 register. Ports may be configured to reset high or low as needed for the application. When port pins are driven high at reset, they are in quasi-bidirectional mode and therefore do not source large amounts of current.

Every output on the P87LPC767 may potentially be used as a 20 mA sink LED drive output. However, there is a maximum total output current for all ports which must not be exceeded.

All ports pins of the P87LPC767 have slew rate controlled outputs. This is to limit noise generated by quickly switching output signals. The slew rate is factory set to approximately 10 ns rise and fall times.

The bits in the P2M1 register that are not used to control configuration of P2.1 and P2.0 are used for other purposes. These bits can enable Schmitt trigger inputs on each I/O port, enable toggle outputs from Timer 0 and Timer 1, and enable a clock output if either the internal RC oscillator or external clock input is being used. The last two functions are described in the Timer/Counters and Oscillator sections respectively. The enable bits for all of these functions are shown in Figure 15.

Each I/O port of the P87LPC767 may be selected to use TTL level inputs or Schmitt inputs with hysteresis. A single configuration bit determines this selection for the entire port. Port pins P1.2, P1.3, and P1.5 always have a Schmitt trigger input.

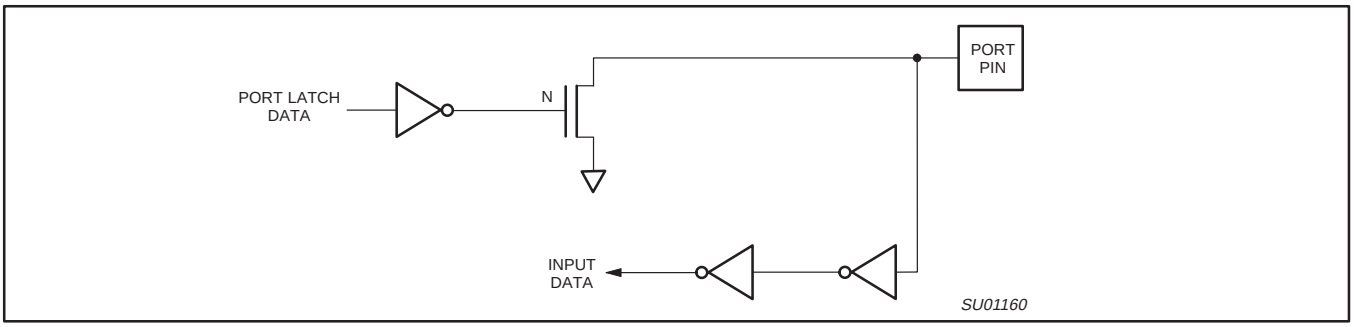

Figure 13. Open Drain Output

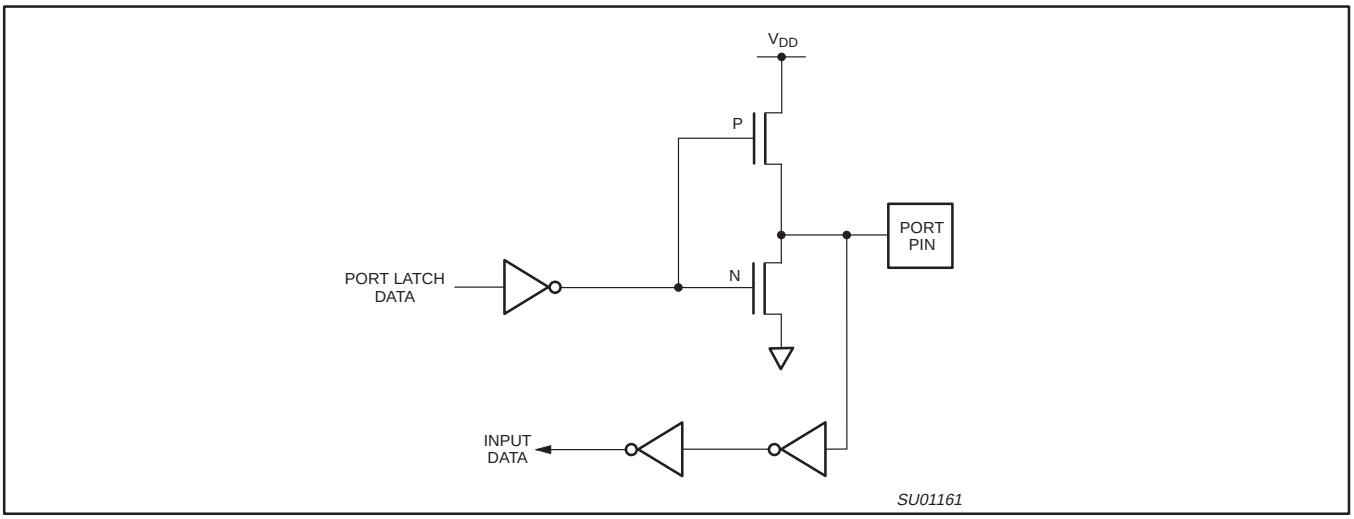

Figure 14. Push-Pull Output

Mode 0

Putting either Timer into Mode 0 makes it look like an 8048 Timer, which is an 8-bit Counter with a divide-by-32 prescaler. Figure 26 shows Mode 0 operation.

In this mode, the Timer register is configured as a 13-bit register. As the count rolls over from all 1s to all 0s, it sets the Timer interrupt flag TFn. The count input is enabled to the Timer when TRn = 1 and either GATE = 0 or $\overline{INTn}$ = 1. (Setting GATE = 1 allows the Timer to be controlled by external input $\overline{INTn}$, to facilitate pulse width measurements). TRn is a control bit in the Special Function Register TCON (Figure 25). The GATE bit is in the TMOD register.

The 13-bit register consists of all 8 bits of THn and the lower 5 bits of TLn. The upper 3 bits of TLn are indeterminate and should be ignored. Setting the run flag (TRn) does not clear the registers.

Mode 0 operation is the same for Timer 0 and Timer 1. See Figure 26. There are two different GATE bits, one for Timer 1 (TMOD.7) and one for Timer 0 (TMOD.3).

TCON Address: 88h Reset Value: 00h

Bit Addressable

7	6	5	4	3	2	1	0
TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0

BIT	SYMBOL	FUNCTION
TCON.7	TF1	Timer 1 overflow flag. Set by hardware on Timer/Counter overflow. Cleared by hardware when the interrupt is processed, or by software.
TCON.6	TR1	Timer 1 Run control bit. Set/cleared by software to turn Timer/Counter 1 on/off.
TCON.5	TF0	Timer 0 overflow flag. Set by hardware on Timer/Counter overflow. Cleared by hardware when the processor vectors to the interrupt routine, or by software.
TCON.4	TR0	Timer 0 Run control bit. Set/cleared by software to turn Timer/Counter 0 on/off.
TCON.3	IE1	Interrupt 1 Edge flag. Set by hardware when external interrupt 1 edge is detected. Cleared by hardware when the interrupt is processed, or by software.
TCON.2	IT1	Interrupt 1 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.
TCON.1	IE0	Interrupt 0 Edge flag. Set by hardware when external interrupt 0 edge is detected. Cleared by hardware when the interrupt is processed, or by software.
TCON.0	IT0	Interrupt 0 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.

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Figure 25. Timer/Counter Control Register (TCON)

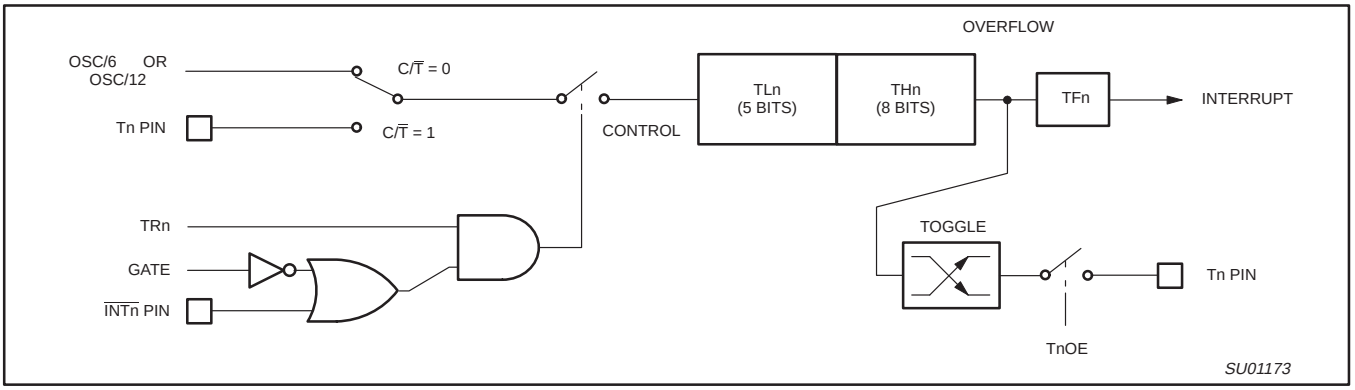

Figure 26. Timer/Counter 0 or 1 in Mode 0 (13-Bit Counter)

Serial Port Control Register (SCON)

The serial port control and status register is the Special Function Register SCON, shown in Figure 30. This register contains not only the mode selection bits, but also the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

The Framing Error bit (FE) allows detection of missing stop bits in the received data stream. The FE bit shares the bit position SCON.7 with the SM0 bit. Which bit appears in SCON at any particular time is determined by the SMOD0 bit in the PCON register. If SMOD0 = 0, SCON.7 is the SM0 bit. If SMOD0 = 1, SCON.7 is the FE bit. Once set, the FE bit remains set until it is cleared by software. This allows detection of framing errors for a group of characters without the need for monitoring it for every character individually.

SCON Address: 98h Reset Value: 00h

Bit Addressable

7	6	5	4	3	2	1	0
SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI

BIT	SYMBOL	FUNCTION
SCON.7	FE	Framing Error. This bit is set by the UART receiver when an invalid stop bit is detected. Must be cleared by software. The SMOD0 bit in the PCON register must be 1 for this bit to be accessible. See SM0 bit below.
SCON.7	SM0	With SM1, defines the serial port mode. The SMOD0 bit in the PCON register must be 0 for this bit to be accessible. See FE bit above.
SCON. 6	SM1	With SM0, defines the serial port mode (see table below).

SM0, SM1	UART Mode	Baud Rate
0 0	0: shift register	CPU clock/6
0 1	1: 8-bit UART	Variable (see text)
1 0	2: 9-bit UART	CPU clock/32 or CPU clock/16
1 1	3: 9-bit UART	Variable (see text)

SCON.5	SM2	Enables the multiprocessor communication feature in Modes 2 and 3. In Mode 2 or 3, if SM2 is set to 1, then RI will not be activated if the received 9th data bit (RB8) is 0. In Mode 1, if SM2=1 then RI will not be activated if a valid stop bit was not received. In Mode 0, SM2 should be 0.
SCON.4	REN	Enables serial reception. Set by software to enable reception. Clear by software to disable reception.
SCON.3	TB8	The 9th data bit that will be transmitted in Modes 2 and 3. Set or clear by software as desired.
SCON.2	RB8	In Modes 2 and 3, is the 9th data bit that was received. In Mode 1, it SM2=0, RB8 is the stop bit that was received. In Mode 0, RB8 is not used.
SCON.1	TI	Transmit interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or at the beginning of the stop bit in the other modes, in any serial transmission. Must be cleared by software.
SCON.0	RI	Receive interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or halfway through the stop bit time in the other modes, in any serial reception (except see SM2). Must be cleared by software.

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Figure 30. Serial Port Control Register (SCON)

Baud Rates

The baud rate in Mode 0 is fixed: Mode 0 Baud Rate = CPU clock/6. The baud rate in Mode 2 depends on the value of bit SMOD1 in Special Function Register PCON. If SMOD1 = 0 (which is the value on reset), the baud rate is 1/32 of the CPU clock frequency. If SMOD1 = 1, the baud rate is 1/16 of the CPU clock frequency.

$$\text{Mode 2 Baud Rate} = \frac{1 + \text{SMOD1}}{32} \times \text{CPU clock frequency}$$

Using Timer 1 to Generate Baud Rates

When Timer 1 is used as the baud rate generator, the baud rates in Modes 1 and 3 are determined by the Timer 1 overflow rate and the value of SMOD1. The Timer 1 interrupt should be disabled in this application. The Timer itself can be configured for either “timer” or “counter” operation, and in any of its 3 running modes. In the most typical applications, it is configured for “timer” operation, in the auto-reload mode (high nibble of TMOD = 0010b). In that case the baud rate is given by the formula:

$$\text{Mode 1, 3 Baud Rate} = \frac{\text{CPU clock frequency}/\text{192 (or 96 if SMOD1 = 1)}}{256 - \text{(TH1)}}$$

Tables 6 and 7 list various commonly used baud rates and how they can be obtained using Timer 1 as the baud rate generator.

Table 10. Baud Rates, Timer Values, and CPU Clock Frequencies for SMOD1 = 0

Timer Count	Baud Rate					
	2400	4800	9600	19.2k	38.4k	57.6k
–1	0.4608	0.9216	* 1.8432	* 3.6864	* 7.3728	* 11.0592
–2	0.9216	1.8432	* 3.6864	* 7.3728	* 14.7456	
–3	1.3824	2.7648	5.5296	* 11.0592	–	–
–4	* 1.8432	* 3.6864	* 7.3728	* 14.7456	–	–
–5	2.3040	4.6080	9.2160	* 18.4320	–	–
–6	2.7648	5.5296	* 11.0592	–	–	–
–7	3.2256	6.4512	12.9024	–	–	–
–8	* 3.6864	* 7.3728	* 14.7456	–	–	–
–9	4.1472	8.2944	16.5888	–	–	–
–10	4.6080	9.2160	* 18.4320	–	–	–

Table 11. Baud Rates, Timer Values, and CPU Clock Frequencies for SMOD1 = 1

Timer Count	Baud Rate						
	2400	4800	9600	19.2k	38.4k	57.6k	115.2k
–1	0.2304	0.4608	0.9216	* 1.8432	* 3.6864	5.5296	* 11.0592
–2	0.4608	0.9216	* 1.8432	* 3.6864	* 7.3728	* 11.0592	–
–3	0.6912	1.3824	2.7648	5.5296	* 11.0592	16.5888	–
–4	0.9216	* 1.8432	* 3.6864	* 7.3728	* 14.7456	–	–
–5	1.1520	2.3040	4.6080	9.2160	* 18.4320	–	–
–6	1.3824	2.7648	5.5296	* 11.0592	–	–	–
–7	1.6128	3.2256	6.4512	12.9024	–	–	–
–8	* 1.8432	* 3.6864	* 7.3728	* 14.7456	–	–	–
–9	2.0736	4.1472	8.2944	16.5888	–	–	–
–10	2.3040	4.6080	9.2160	* 18.4320	–	–	–
–11	2.5344	5.0688	10.1376	–	–	–	–
–12	2.7648	5.5296	* 11.0592	–	–	–	–
–13	2.9952	5.9904	11.9808	–	–	–	–
–14	3.2256	6.4512	12.9024	–	–	–	–
–15	3.4560	6.9120	13.8240	–	–	–	–
–16	* 3.6864	* 7.3728	* 14.7456	–	–	–	–
–17	3.9168	7.8336	15.6672	–	–	–	–
–18	4.1472	8.2944	16.5888	–	–	–	–
–19	4.3776	8.7552	17.5104	–	–	–	–
–20	4.6080	9.2160	* 18.4320	–	–	–	–
–21	4.8384	9.6768	19.3536	–	–	–	–

NOTES TO TABLES 10 AND 11:

1. Tables 6 and 7 apply to UART modes 1 and 3 (variable rate modes), and show CPU clock rates in MHz for standard baud rates from 2400 to 115.2k baud.
2. Table 6 shows timer settings and CPU clock rates with the SMOD1 bit in the PCON register = 0 (the default after reset), while Table 7 reflects the SMOD1 bit = 1.
3. The tables show all potential CPU clock frequencies up to 20 MHz that may be used for baud rates from 9600 baud to 115.2k baud. Other CPU clock frequencies that would give only lower baud rates are not shown.
4. Table entries marked with an asterisk (*) indicate standard crystal and ceramic resonator frequencies that may be obtained from many sources without special ordering.

More About UART Mode 0

Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received: 8 data bits (LSB first). The baud rate is fixed at 1/6 the CPU clock frequency. Figure 31 shows a simplified functional diagram of the serial port in Mode 0, and associated timing.

Transmission is initiated by any instruction that uses SBUF as a destination register. The “write to SBUF” signal at S6P2 also loads a 1 into the 9th position of the transmit shift register and tells the TX Control block to commence a transmission. The internal timing is such that one full machine cycle will elapse between “write to SBUF” and activation of SEND.

SEND enables the output of the shift register to the alternate output function line of P1.1 and also enable SHIFT CLOCK to the alternate output function line of P1.0. SHIFT CLOCK is low during S3, S4, and S5 of every machine cycle, and high during S6, S1, and S2. At S6P2 of every machine cycle in which SEND is active, the contents of the transmit shift are shifted to the right one position.

As data bits shift out to the right, zeros come in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position, is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control block to do one last shift and then deactivate SEND and set T1. Both of these actions occur at S1P1 of the 10th machine cycle after “write to SBUF.” Reception is initiated by the condition REN = 1 and R1 = 0. At S6P2 of the next machine cycle, the RX Control unit writes the bits 11111110 t o the receive shift register, and in the next clock phase activates RECEIVE.

RECEIVE enable SHIFT CLOCK to the alternate output function line of P1.0. SHIFT CLOCK makes transitions at S3P1 and S6P1 of every machine cycle. At S6P2 of every machine cycle in which RECEIVE is active, the contents of the receive shift register are shifted to the left one position. The value that comes in from the right is the value that was sampled at the P1.1 pin at S5P2 of the same machine cycle.

As data bits come in from the right, 1s shift out to the left. When the 0 that was initially loaded into the rightmost position arrives at the leftmost position in the shift register, it flags the RX Control block to do one last shift and load SBUF. At S1P1 of the 10th machine cycle after the write to SCON that cleared RI, RECEIVE is cleared as RI is set.

More About UART Mode 1

Ten bits are transmitted (through TxD), or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in SCON. In the P87LPC767 the baud rate is determined by the Timer 1 overflow rate. Figure 32 shows a simplified functional diagram of the serial port in Mode 1, and associated timings for transmit receive.

Transmission is initiated by any instruction that uses SBUF as a destination register. The “write to SBUF” signal also loads a 1 into the 9th bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission actually commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. (Thus, the bit times are synchronized to the divide-by-16 counter, not to the “write to SBUF” signal.)

The transmission begins with activation of SEND which puts the start bit at TxD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TxD. The first shift pulse occurs one bit time after that.

As data bits shift out to the right, zeros are clocked in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control unit to do one last shift and then deactivate SEND and set TI. This occurs at the 10th divide-by-16 rollover after “write to SBUF.”

Reception is initiated by a detected 1-to-0 transition at RxD. For this purpose RxD is sampled at a rate of 16 times whatever baud rate has been established. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written into the input shift register. Resetting the divide-by-16 counter aligns its rollovers with the boundaries of the incoming bit times.

The 16 states of the counter divide each bit time into 16ths. At the 7th, 8th, and 9th counter states of each bit time, the bit detector samples the value of RxD. The value accepted is the value that was seen in at least 2 of the 3 samples. This is done for noise rejection. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit goes back to looking for another 1-to-0 transition. This is to provide rejection of false start bits. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame will proceed.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in mode 1 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI. The signal to load SBUF and RB8, and to set RI, will be generated if, and only if, the following conditions are met at the time the final shift pulse is generated.: 1. R1 = 0, and 2. Either SM2 = 0, or the received stop bit = 1.

If either of these two conditions is not met, the received frame is irretrievably lost. If both conditions are met, the stop bit goes into RB8, the 8 data bits go into SBUF, and RI is activated. At this time, whether the above conditions are met or not, the unit goes back to looking for a 1-to-0 transition in RxD.

More About UART Modes 2 and 3

Eleven bits are transmitted (through TxD), or received (through RxD): a start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). On transmit, the 9th data bit (TB8) can be assigned the value of 0 or 1. On receive, the 9the data bit goes into RB8 in SCON. The baud rate is programmable to either 1/16 or 1/32 of the CPU clock frequency in Mode 2. Mode 3 may have a variable baud rate generated from Timer 1.

Figures 33 and 34 show a functional diagram of the serial port in Modes 2 and 3. The receive portion is exactly the same as in Mode 1. The transmit portion differs from Mode 1 only in the 9th bit of the transmit shift register.

Transmission is initiated by any instruction that uses SBUF as a destination register. The “write to SBUF” signal also loads TB8 into the 9th bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. (Thus, the bit times are synchronized to the divide-by-16 counter, not to the “write to SBUF” signal.)

The transmission begins with activation of SEND, which puts the start bit at TxD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TxD. The first shift pulse occurs one bit time after that. The first shift clocks a 1 (the stop bit) into the 9th bit position of the shift register. Thereafter, only zeros are clocked in. Thus, as data bits shift out to the right, zeros are clocked in from the left. When TB8 is at the output position of the shift register, then the stop bit is just to the left of TB8, and all positions to the left of that contain zeros. This condition flags the TX Control unit to do one last shift and then deactivate SEND and set TI. This occurs at the 11th divide-by-16 rollover after “write to SBUF.”

Reception is initiated by a detected 1-to-0 transition at RxD. For this purpose RxD is sampled at a rate of 16 times whatever baud rate has been established. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written to the input shift register.

At the 7th, 8th, and 9th counter states of each bit time, the bit detector samples the value of R–D. The value accepted is the value that was seen in at least 2 of the 3 samples. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit goes back to looking for another 1-to-0 transition. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame will proceed.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in Modes 2 and 3 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI.

The signal to load SBUF and RB8, and to set RI, will be generated if, and only if, the following conditions are met at the time the final shift pulse is generated. 1. RI = 0, and 2. Either SM2 = 0, or the received 9th data bit = 1.

If either of these conditions is not met, the received frame is irretrievably lost, and RI is not set. If both conditions are met, the received 9th data bit goes into RB8, and the first 8 data bits go into SBUF. One bit time later, whether the above conditions were met or not, the unit goes back to looking for a 1-to-0 transition at the RxD input.

Multiprocessor Communications

UART modes 2 and 3 have a special provision for multiprocessor communications. In these modes, 9 data bits are received or transmitted. When data is received, the 9th bit is stored in RB8. The UART can be programmed such that when the stop bit is received, the serial port interrupt will be activated only if RB8 = 1. This feature is enabled by setting bit SM2 in SCON. One way to use this feature in multiprocessor systems is as follows:

When the master processor wants to transmit a block of data to one of several slaves, it first sends out an address byte which identifies the target slave. An address byte differs from a data byte in that the 9th bit is 1 in an address byte and 0 in a data byte. With SM2 = 1, no slave will be interrupted by a data byte. An address byte, however, will interrupt all slaves, so that each slave can examine the received byte and see if it is being addressed. The addressed slave will clear its SM2 bit and prepare to receive the data bytes that follow. The slaves that weren’t being addressed leave their SM2 bits set and go on about their business, ignoring the subsequent data bytes.

SM2 has no effect in Mode 0, and in Mode 1 can be used to check the validity of the stop bit, although this is better done with the Framing Error flag. In a Mode 1 reception, if SM2 = 1, the receive interrupt will not be activated unless a valid stop bit is received.

Automatic Address Recognition

Automatic Address Recognition is a feature which allows the UART to recognize certain addresses in the serial bit stream by using hardware to make the comparisons. This feature saves a great deal of software overhead by eliminating the need for the software to examine every serial address which passes by the serial port. This feature is enabled by setting the SM2 bit in SCON. In the 9 bit UART modes, mode 2 and mode 3, the Receive Interrupt flag (RI) will be automatically set when the received byte contains either the “Given” address or the “Broadcast” address. The 9 bit mode requires that the 9th information bit is a 1 to indicate that the received information is an address and not data.

Using the Automatic Address Recognition feature allows a master to selectively communicate with one or more slaves by invoking the Given slave address or addresses. All of the slaves may be contacted by using the Broadcast address. Two special Function Registers are used to define the slave’s address, SADDR, and the address mask, SADEN. SADEN is used to define which bits in the SADDR are to be used and which bits are “don’t care”. The SADEN mask can be logically ANDed with the SADDR to create the “Given” address which the master will use for addressing each of the slaves. Use of the Given address allows multiple slaves to be recognized while excluding others. The following examples will help to show the versatility of this scheme:

Slave 0 SADDR = 1100 0000
SADEN = 1111 1101
Given = 1100 00X0

Slave 1 SADDR = 1100 0000
SADEN = 1111 1110
Given = 1100 000X

In the above example SADDR is the same and the SADEN data is used to differentiate between the two slaves. Slave 0 requires a 0 in bit 0 and it ignores bit 1. Slave 1 requires a 0 in bit 1 and bit 0 is ignored. A unique address for Slave 0 would be 1100 0010 since slave 1 requires a 0 in bit 1. A unique address for slave 1 would be 1100 0001 since a 1 in bit 0 will exclude slave 0. Both slaves can be selected at the same time by an address which has bit 0 = 0 (for slave 0) and bit 1 = 0 (for slave 1). Thus, both could be addressed with 1100 0000.

In a more complex system the following could be used to select slaves 1 and 2 while excluding slave 0:

Slave 0 SADDR = 1100 0000
SADEN = 1111 1001
Given = 1100 0XX0

Slave 1 SADDR = 1110 0000
SADEN = 1111 1010
Given = 1110 0X0X

Slave 2 SADDR = 1110 0000
SADEN = 1111 1100
Given = 1110 00XX

In the above example the differentiation among the 3 slaves is in the lower 3 address bits. Slave 0 requires that bit 0 = 0 and it can be uniquely addressed by 1110 0110. Slave 1 requires that bit 1 = 0 and it can be uniquely addressed by 1110 and 0101. Slave 2 requires that bit 2 = 0 and its unique address is 1110 0011. To select Slaves 0 and 1 and exclude Slave 2 use address 1110 0100, since it is necessary to make bit 2 = 1 to exclude slave 2. The Broadcast Address for each slave is created by taking the logical OR of SADDR and SADEN. Zeros in this result are treated as don’t-cares. In most cases, interpreting the don’t-cares as ones, the broadcast address will be FF hexadecimal. Upon reset SADDR and SADEN are loaded with 0s. This produces a given address of all “don’t cares” as well as a Broadcast address of all “don’t cares”. This effectively disables the Automatic Addressing mode and allows the microcontroller to use standard UART drivers which do not make use of this feature.

Watchdog Timer

When enabled via the WDTE configuration bit, the watchdog timer is operated from an independent, fully on-chip oscillator in order to provide the greatest possible dependability. When the watchdog feature is enabled, the timer must be fed regularly by software in order to prevent it from resetting the CPU, and it cannot be turned off. When disabled as a watchdog timer (via the WDTE bit in the UCFG1 configuration register), it may be used as an interval timer and may generate an interrupt. The watchdog timer is shown in Figure 35.

The watchdog timeout time is selectable from one of eight values, nominal times range from 16 milliseconds to 2.1 seconds. The frequency tolerance of the independent watchdog RC oscillator is ±37%. The timeout selections and other control bits are shown in Figure 36. When the watchdog function is enabled, the WDCON register may be written once during chip initialization in order to set the watchdog timeout time. The recommended method of initializing the WDCON register is to first feed the watchdog, then write to WDCON to configure the WDS2–0 bits. Using this method, the watchdog initialization may be done any time within 10 milliseconds after startup without a watchdog overflow occurring before the initialization can be completed.

Since the watchdog timer oscillator is fully on-chip and independent of any external oscillator circuit used by the CPU, it intrinsically serves as an oscillator fail detection function. If the watchdog feature is enabled and the CPU oscillator fails for any reason, the watchdog timer will time out and reset the CPU.

When the watchdog function is enabled, the timer is deactivated temporarily when a chip reset occurs from another source, such as a power on reset, brownout reset, or external reset.

Watchdog Feed Sequence

If the watchdog timer is running, it must be fed before it times out in order to prevent a chip reset from occurring. The watchdog feed sequence consists of first writing the value 1Eh, then the value E1h to the WDRST register. An example of a watchdog feed sequence is shown below.

```
WDFeed:
   mov   WDRST,#1eh   ; First part of watchdog feed sequence.
   mov   WDRST,#0e1h  ; Second part of watchdog feed sequence.
```

The two writes to WDRST do not have to occur in consecutive instructions. An incorrect watchdog feed sequence does not cause any immediate response from the watchdog timer, which will still time out at the originally scheduled time if a correct feed sequence does not occur prior to that time.

After a chip reset, the user program has a limited time in which to either feed the watchdog timer or change the timeout period. When a low CPU clock frequency is used in the application, the number of instructions that can be executed before the watchdog overflows may be quite small.

Watchdog Reset

If a watchdog reset occurs, the internal reset is active for approximately one microsecond. If the CPU clock was still running, code execution will begin immediately after that. If the processor was in Power Down mode, the watchdog reset will start the oscillator and code execution will resume after the oscillator is stable.

DC ELECTRICAL CHARACTERISTICS

V_{DD} = 2.7 V to 6.0 V unless otherwise specified; T_{amb} = 0 °C to +70 °C, –40 °C to +85 °C, or –40 °C to +125 °C, unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP[1,2]	MAX	
I_{DD}	Power supply current, operating	5.0 V, 20 MHz[11]		15	25	mA
		3.0 V, 10 MHz[11]		4	7	mA
I_{ID}	Power supply current, Idle mode	5.0 V, 20 MHz[11]		6	10	mA
		3.0 V, 10 MHz[11]		2	4	mA
I_{PD}	Power supply current, Power Down mode	5.0 V[11]		1	10	µA
		3.0 V[11]		1	5	µA
V_{RAM}	RAM keep-alive voltage		1.5			V
V_{IL}	Input low voltage (TTL input)	4.0 V < V_{DD} < 6.0 V	–0.5		0.2 V_{DD}–0.1	V
		2.7 V < V_{DD} < 4.0 V	–0.5		0.7	V
V_{IL1}	Negative going threshold (Schmitt input)		–0.5 V_{DD}	0.4 V_{DD}	0.3 V_{DD}	V
V_{IH}	Input high voltage (TTL input)		0.2 V_{DD}+0.9		V_{DD}+0.5	V
V_{IH1}	Positive going threshold (Schmitt input)		0.7 V_{DD}	0.6 V_{DD}	V_{DD}+0.5	V
HYS	Hysteresis voltage			0.2 V_{DD}		V
V_{OL}	Output low voltage all ports[5, 9]	I_{OL} = 3.2 mA, V_{DD} = 2.7 V			0.4	V
V_{OL1}	Output low voltage all ports[5, 9]	I_{OL} = 20 mA, V_{DD} = 2.7 V			1.0	V
V_{OH}	Output high voltage, all ports[3]	I_{OH} = –20 µA, V_{DD} = 2.7 V	V_{DD}–0.7			V
		I_{OH} = –30 µA, V_{DD} = 4.5 V	V_{DD}–0.7			V
V_{OH1}	Output high voltage, all ports[4]	I_{OH} = –1.0 mA, V_{DD} = 2.7 V	V_{DD}–0.7			V
C_{IO}	Input/Output pin capacitance[10]				15	pF
I_{IL}	Logical 0 input current, all ports[8]	V_{IN} = 0.4 V			–50	µA
I_{LI}	Input leakage current, all ports[7]	V_{IN} = V_{IL} or V_{IH}			±2	µA
I_{TL}	Logical 1 to 0 transition current, all ports[3, 6]	V_{IN} = 1.5 V at V_{DD} = 3.0 V	–30		–250	µA
		V_{IN} = 2.0 V at V_{DD} = 5.5 V	–150		–650	µA
R_{RST}	Internal reset pull-up resistor		40		225	kΩ
$V_{BO2.5}$	Brownout trip voltage with BOV = 1[12]	T_{amb} = 0 °C to +70 °C	2.45	2.5	2.65	V
$V_{BO3.8}$	Brownout trip voltage with BOV = 0		3.45	3.8	3.90	V
V_{REF}	Reference voltage		1.11	1.26	1.41	V

NOTES:

1. Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
2. See other Figures for details.
3. Ports in quasi-bidirectional mode with weak pull-up (applies to all port pins with pull-ups). Does not apply to open drain pins.
4. Ports in PUSH-PULL mode. Does not apply to open drain pins.
5. In all output modes except high impedance mode.
6. Port pins source a transition current when used in quasi-bidirectional mode and externally driven from 1 to 0. This current is highest when V_{IN} is approximately 2 V.
7. Measured with port in high impedance mode. Parameter is guaranteed but not tested at cold temperature.
8. Measured with port in quasi-bidirectional mode.
9. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 20 mA
   Maximum total I_{OL} for all outputs: 80 mA
   Maximum total I_{OH} for all outputs: 5 mA
   If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
10. Pin capacitance is characterized but not tested.
11. The I_{DD}, I_{ID}, and I_{PD} specifications are measured using an external clock with the following functions disabled: comparators, brownout detect, and watchdog timer. For V_{DD} = 3 V, LPEP = 1. Refer to the appropriate figures on the following pages for additional current drawn by each of these functions and detailed graphs for other frequency and voltage combinations.
12. Devices initially operating at V_{DD} = 2.7 V or above and at f_{OSC} = 10 MHz or less are guaranteed to continue to execute instructions correctly at the brownout trip point. Initial power-on operation below V_{DD} = 2.7 V is not guaranteed.

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V_{DD} = 3.0 V to 6.0 V unless otherwise specified; T_{amb} = 0 °C to +70 °C, –40 °C to +85 °C, or –40 °C to +125 °C, unless otherwise specified

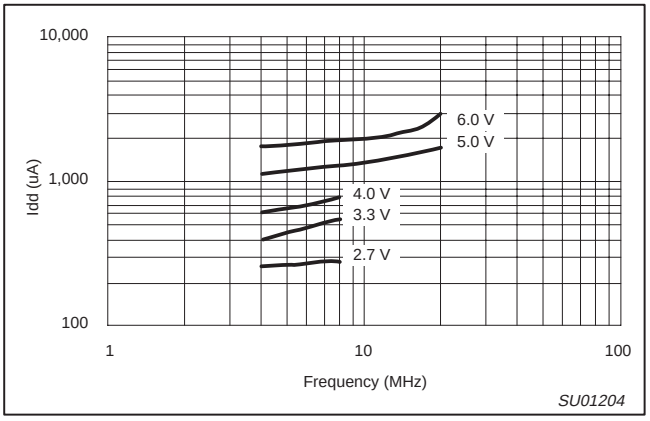

Figure 45. Typical Idd versus frequency (high frequency oscillator, 25 °C)

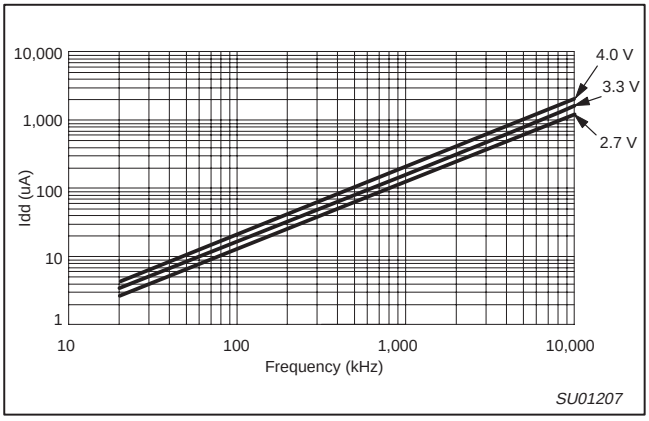

Figure 48. Typical Idle Idd versus frequency (external clock, 25 °C, LPEP = 1)

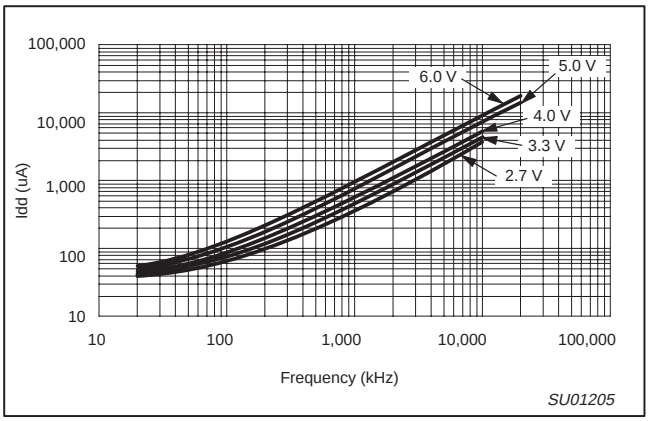

Figure 46. Typical Active Idd versus frequency (external clock, 25 °C, LPEP = 0)

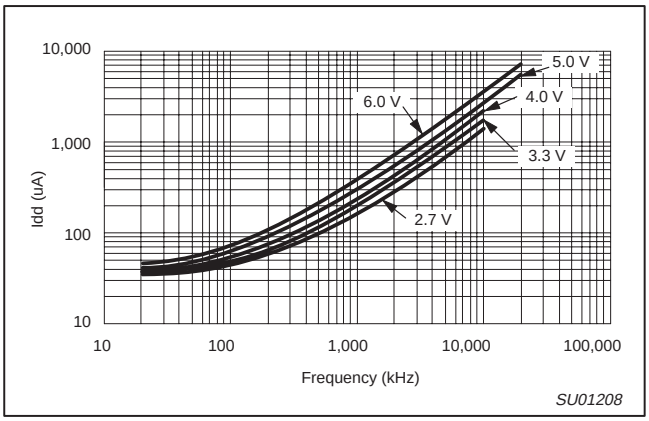

Figure 49. Typical Idle Idd versus frequency (external clock, 25 °C, LPEP = 0)

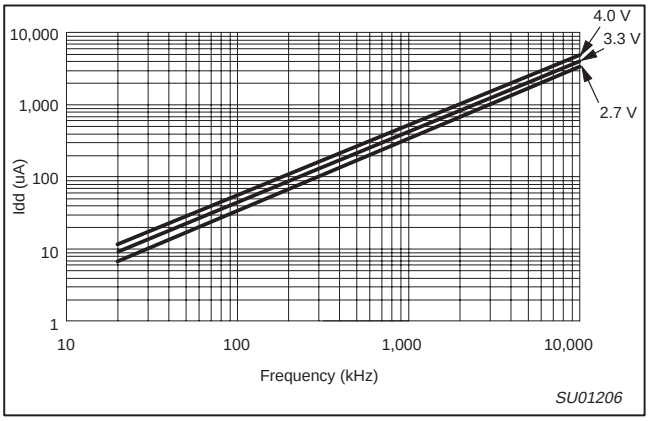

Figure 47. Typical Active Idd versus frequency (external clock, 25 °C, LPEP = 1)

REVISION HISTORY

Date	CPCN	Description
2002 Mar 25	9397 750 09557	– Added revision history – Updated Reset section
2001 Aug 07	9397 750 08675	Previous release