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What is "Embedded - Microcontrollers"?

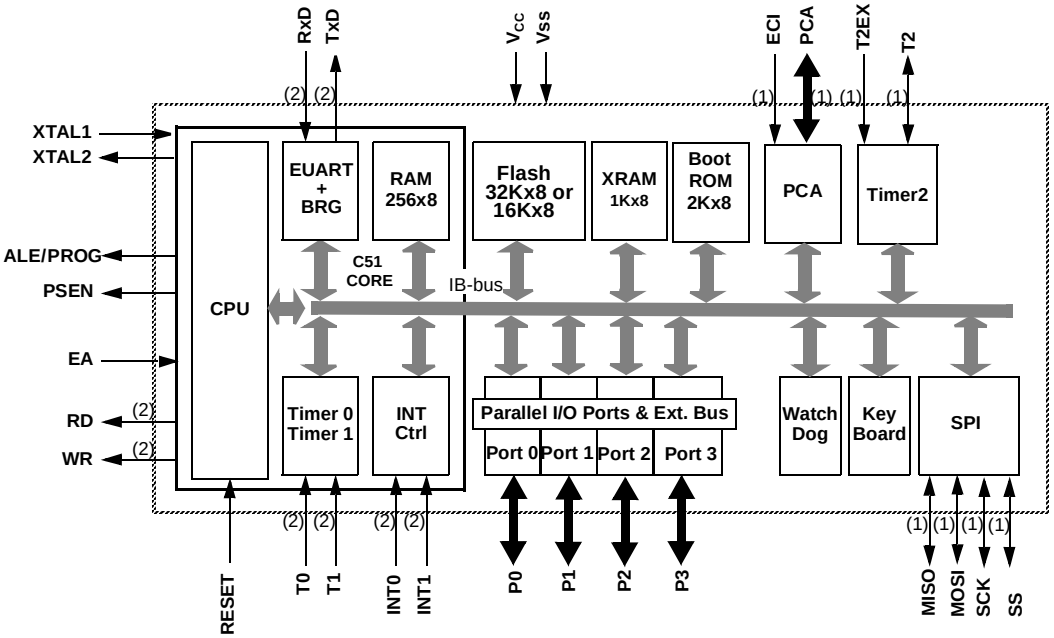
"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	80C51
Core Size	8-Bit
Speed	60MHz
Connectivity	SPI, UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	32
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LQFP
Supplier Device Package	44-VQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/at89c51rc2-rltim

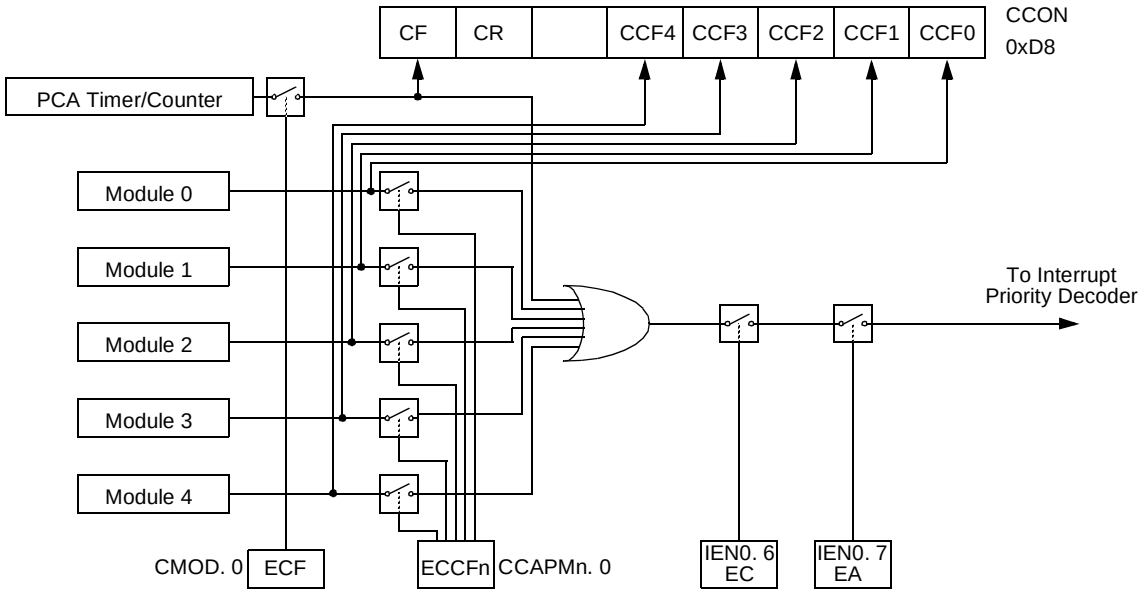
Block Diagram

Figure 1. Block Diagram



Notes: 1. Alternate function of Port 1.
2. Alternate function of Port 3.

Figure 12. PCA Interrupt System



PCA Modules: each one of the five compare/capture Modules has six possible functions. It can perform:

- 16-bit Capture, positive-edge triggered
- 16-bit Capture, negative-edge triggered
- 16-bit Capture, both positive and negative-edge triggered
- 16-bit Software Timer
- 16-bit High-speed Output
- 8-bit Pulse Width Modulator

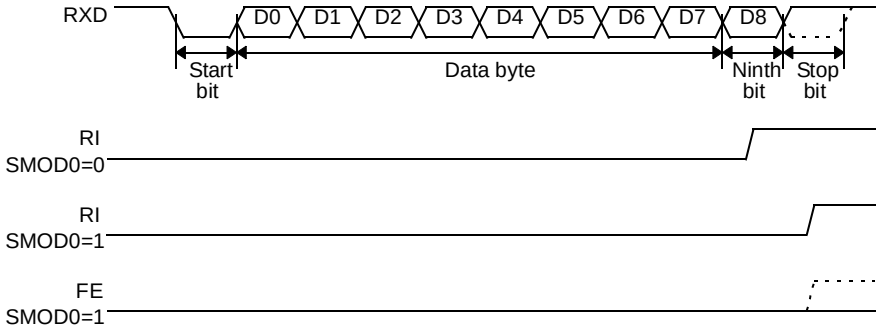
In addition, Module 4 can be used as a Watchdog Timer.

Each Module in the PCA has a special function register associated with it. These registers are: CCAPM0 for Module 0, CCAPM1 for Module 1, etc. (see Table 24). The registers contain the bits that control the mode that each Module will operate in.

- The ECCF bit (CCAPMn. 0 where n = 0, 1, 2, 3, or 4 depending on the Module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated Module.
- PWM (CCAPMn. 1) enables the pulse width modulation mode.
- The TOG bit (CCAPMn. 2) when set causes the CEX output associated with the Module to toggle when there is a match between the PCA counter and the Module's capture/compare register.
- The match bit MAT (CCAPMn. 3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the Module's capture/compare register.
- The next two bits CAPN (CCAPMn. 4) and CAPP (CCAPMn. 5) determine the edge that a capture input will be active on. The CAPN bit enables the negative edge, and the CAPP bit enables the positive edge. If both bits are set both edges will be enabled and a capture will occur for either transition.
- The last bit in the register ECOM (CCAPMn. 6) when set enables the comparator function.

Table 24 shows the CCAPMn settings for the various PCA functions.

Figure 19. UART Timings in Modes 2 and 3



Automatic Address Recognition

The automatic address recognition feature is enabled when the multiprocessor communication feature is enabled (SM2 bit in SCON register is set).

Implemented in hardware, automatic address recognition enhances the multiprocessor communication feature by allowing the serial port to examine the address of each incoming command frame. Only when the serial port recognizes its own address, the receiver sets RI bit in SCON register to generate an interrupt. This ensures that the CPU is not interrupted by command frames addressed to other devices.

If desired, the user may enable the automatic address recognition feature in mode 1.In this configuration, the stop bit takes the place of the ninth data bit. Bit RI is set only when the received command frame address matches the device's address and is terminated by a valid stop bit.

To support automatic address recognition, a device is identified by a given address and a broadcast address.

Note: The multiprocessor communication and automatic address recognition features cannot be enabled in mode 0 (i. e. setting SM2 bit in SCON register in mode 0 has no effect).

Given Address

Each device has an individual address that is specified in SADDR register; the SADEN register is a mask byte that contains don't-care bits (defined by zeros) to form the device's given address. The don't-care bits provide the flexibility to address one or more slaves at a time. The following example illustrates how a given address is formed.

To address a device by its individual address, the SADEN mask byte must be `1111 1111b`.

For example:

SADDR0101 0110b
SADEN1111 1100b
Given0101 01XXb

The following is an example of how to use given addresses to address different slaves:

Slave A:SADDR1111 0001b
SADEN1111 1010b
Given1111 0X0Xb

Slave B:SADDR1111 0011b
SADEN1111 1001b
Given1111 0XX1b

Slave C:SADDR1111 0010b
SADEN1111 1101b
Given1111 00X1b

Table 41. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0
SMOD1	**SMOD0**	**-**	**POF**	**GF1**	**GF0**	**PD**	**IDL**

Bit Number	Bit Mnemonic	Description
7	SMOD1	**Serial port Mode bit 1 for UART** Set to select double baud rate in mode 1, 2 or 3.
6	SMOD0	**Serial port Mode bit 0 for UART** Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.
5	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit.
4	POF	**Power-Off Flag** Cleared to recognize next reset type. Set by hardware when VCC rises from 0 to its nominal voltage. Can also be set by software.
3	GF1	**General purpose Flag** Cleared by user for general purpose usage. Set by user for general purpose usage.
2	GF0	**General purpose Flag** Cleared by user for general purpose usage. Set by user for general purpose usage.
1	PD	**Power-Down mode bit** Cleared by hardware when reset occurs. Set to enter power-down mode.
0	IDL	**Idle mode bit** Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.

Reset Value = 00X1 0000b
Not bit addressable

Power-off flag reset value will be 1 only after a power on (cold reset). A warm reset doesn't affect the value of this bit.

Table 42. BDRCON Register

BDRCON - Baud Rate Control Register (9Bh)

7	6	5	4	3	2	1	0
-	-	-	BRR	TBCK	RBCK	SPD	SRC

Bit Number	Bit Mnemonic	Description
7	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit
6	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit
5	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit.
4	BRR	**Baud Rate Run Control bit** Cleared to stop the internal Baud Rate Generator. Set to start the internal Baud Rate Generator.
3	TBCK	**Transmission Baud rate Generator Selection bit for UART** Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.
2	RBCK	**Reception Baud Rate Generator Selection bit for UART** Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.
1	SPD	**Baud Rate Speed Control bit for UART** Cleared to select the SLOW Baud Rate Generator. Set to select the FAST Baud Rate Generator.
0	SRC	**Baud Rate Source select bit in Mode 0 for UART** Cleared to select $F_{OSC}/12$ as the Baud Rate Generator ($F_{CLK\ PERIPH}/6$ in X2 mode). Set to select the internal Baud Rate Generator for UARTs in mode 0.

Reset Value = XXX0 0000b
Not bit addressablef

Table 46. IPH0 Register

IPH0 - Interrupt Priority High Register (B7h)

7	6	5	4	3	2	1	0
-	PPCH	PT2H	PSH	PT1H	PX1H	PT0H	PX0H

Bit Number	Bit Mnemonic	Description
7	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit.
6	PPCH	**PCA Interrupt Priority High Bit** PPCH PPCL Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
5	PT2H	**Timer 2 Overflow Interrupt Priority High Bit** PT2H PT2L Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
4	PSH	**Serial Port Priority High Bit** PSH PSL Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
3	PT1H	**Timer 1 Overflow Interrupt Priority High Bit** PT1H PT1L Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
2	PX1H	**External Interrupt 1 Priority High Bit** PX1H PX1L Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
1	PT0H	**Timer 0 Overflow Interrupt Priority High Bit** PT0H PT0L Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest
0	PX0H	**External Interrupt 0 Priority High Bit** PX0H PX0L Priority Level 0 0 Lowest 0 1 1 0 1 1 Highest

Reset Value = X000 0000b
Not bit addressable

Serial Port Interface (SPI)

The Serial Peripheral Interface Module (SPI) allows full-duplex, synchronous, serial communication between the MCU and peripheral devices, including other MCUs.

Features

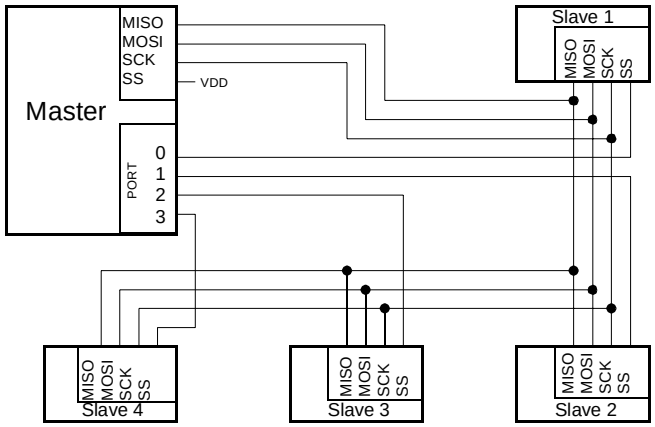
Features of the SPI Module include the following:

- Full-duplex, three-wire synchronous transfers
- Master or Slave operation
- Eight programmable Master clock rates
- Serial clock with programmable polarity and phase
- Master Mode fault error flag with MCU interrupt capability
- Write collision flag protection

Signal Description

Figure 25 shows a typical SPI bus configuration using one Master controller and many Slave peripherals. The bus is made of three wires connecting all the devices.

Figure 25. SPI Master/Slaves Interconnection



The Master device selects the individual Slave devices by using four pins of a parallel port to control the four $\overline{SS}$ pins of the Slave devices.

Master Output Slave Input (MOSI)

This 1-bit signal is directly connected between the Master Device and a Slave Device. The MOSI line is used to transfer data in series from the Master to the Slave. Therefore, it is an output signal from the Master, and an input signal to a Slave. A Byte (8-bit word) is transmitted most significant bit (MSB) first, least significant bit (LSB) last.

Master Input Slave Output (MISO)

This 1-bit signal is directly connected between the Slave Device and a Master Device. The MISO line is used to transfer data in series from the Slave to the Master. Therefore, it is an output signal from the Slave, and an input signal to the Master. A Byte (8-bit word) is transmitted most significant bit (MSB) first, least significant bit (LSB) last.

SPI Serial Clock (SCK)

This signal is used to synchronize the data movement both in and out of the devices through their MOSI and MISO lines. It is driven by the Master for eight clock cycles which allows to exchange one Byte on the serial lines.

Slave Select ($\overline{SS}$)

Each Slave peripheral is selected by one Slave Select pin ($\overline{SS}$). This signal must stay low for any message for a Slave. It is obvious that only one Master ($\overline{SS}$ high level) can

Figure 28. Data Transmission Format (CPHA = 0)

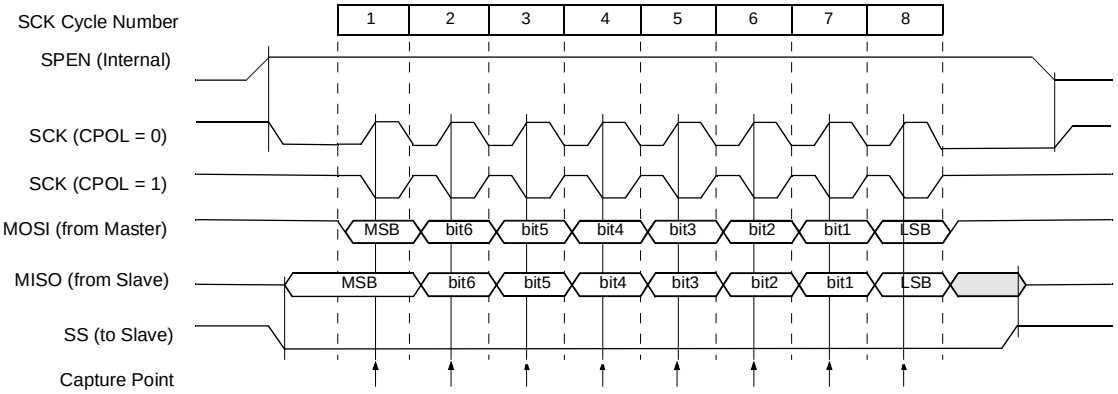


Figure 29. Data Transmission Format (CPHA = 1)

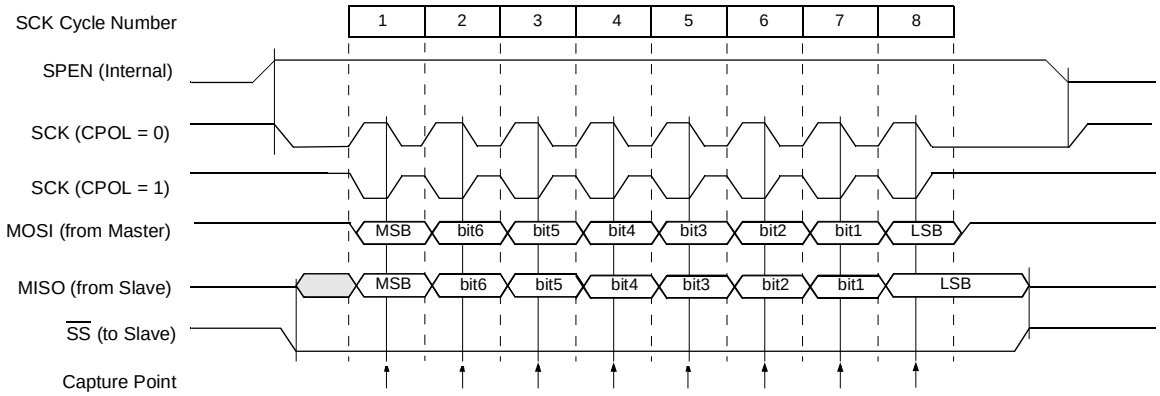
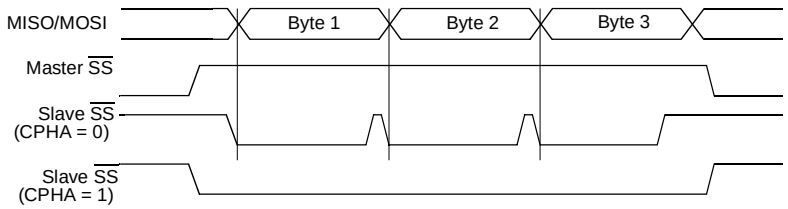


Figure 30. CPHA/SS Timing



As shown in Figure 28, the first SCK edge is the MSB capture strobe. Therefore, the Slave must begin driving its data before the first SCK edge, and a falling edge on the SS pin is used to start the transmission. The SS pin must be toggled high and then low between each Byte transmitted (Figure 30).

Figure 29 shows an SPI transmission in which CPHA is '1'. In this case, the Master begins driving its MOSI pin on the first SCK edge. Therefore, the Slave uses the first SCK edge as a start transmission signal. The SS pin can remain low between transmissions (Figure 30). This format may be preffered in systems having only one Master and only one Slave driving the MISO data line.

Bit Number	Bit Mnemonic	Description
1	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit.
0	-	**Reserved** The value read from this bit is indeterminate. Do not set this bit.

Reset Value = 00X0 XXXXb

Not Bit addressable

Serial Peripheral DATa Register (SPDAT)

The Serial Peripheral Data Register (Table 58) is a read/write buffer for the receive data register. A write to SPDAT places data directly into the shift register. No transmit buffer is available in this model.

A Read of the SPDAT returns the value located in the receive buffer and not the content of the shift register.

Table 58. SPDAT Register

SPDAT - Serial Peripheral Data Register (0C5H)

7	6	5	4	3	2	1	0
R7	**R6**	**R5**	**R4**	**R3**	**R2**	**R1**	**R0**

Reset Value = Indeterminate

R7:R0: Receive data bits

SPCON, SPSTA and SPDAT registers may be read and written at any time while there is no on-going exchange. However, special care should be taken when writing to them while a transmission is on-going:

- Do not change SPR2, SPR1 and SPR0
- Do not change CPHA and CPOL
- Do not change MSTR
- Clearing SPEN would immediately disable the peripheral
- Writing to the SPDAT will cause an overflow.

Table 60. WDTPRG Register

WDTPRG - Watchdog Timer Out Register (0A7h)

7	6	5	4	3	2	1	0
-	-	-	-	-	S2	S1	S0

Bit Number	Bit Mnemonic	Description
7	-	**Reserved** The value read from this bit is undetermined. Do not try to set this bit.
6	-	
5	-	
4	-	
3	-	
2	S2	**WDT Time-out Select Bit 2**
1	S1	**WDT Time-out Select Bit 1**
0	S0	**WDT Time-out Select Bit 0**
		S2 S1 S0Selected Time-out 0 0 0(2^{14} - 1) machine cycles, 16. 3 ms @ F_{OSCA} = 12 MHz 0 0 1(2^{15} - 1) machine cycles, 32.7 ms @ F_{OSCA} = 12 MHz 0 1 0 (2^{16} - 1) machine cycles, 65. 5 ms @ F_{OSCA} = 12 MHz 0 1 1(2^{17} - 1) machine cycles, 131 ms @ F_{OSCA} = 12 MHz 1 0 0(2^{18} - 1) machine cycles, 262 ms @ F_{OSCA} = 12 MHz 1 0 1 (2^{19} - 1) machine cycles, 542 ms @ F_{OSCA} = 12 MHz 1 1 0(2^{20} - 1) machine cycles, 1.05 s @ F_{OSCA} = 12 MHz 1 1 1 (2^{21} - 1) machine cycles, 2.09 s @ F_{OSCA} = 12 MHz

Reset Value = XXXX X000

WDT During Power-down and Idle

In Power-down mode the oscillator stops, which means the WDT also stops. While in Power-down mode the user does not need to service the WDT. There are two methods of exiting Power-down mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power-down mode. When Power-down is exited with hardware reset, servicing the WDT should occur as it normally should whenever the AT89C51RB2/RC2 is reset. Exiting Power-down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service routine.

To ensure that the WDT does not overflow within a few states of exiting of power-down, it is better to reset the WDT just before entering power-down.

In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting the AT89C51RB2/RC2 while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.

Flash EEPROM Memory

The Flash memory increases EPROM and ROM functionality with in-circuit electrical erasure and programming. It contains 16K or 32K Bytes of program memory organized in 128 or 256 pages of 128 Bytes. This memory is both parallel and serial In-system Programmable (ISP). ISP allows devices to alter their own program memory in the actual end product under software control. A default serial loader (bootloader) program allows ISP of the Flash.

The programming does not require external dedicated programming voltage. The necessary high programming voltage is generated on-chip using the standard V_{CC} pins of the microcontroller.

Features

- Flash EEPROM internal program memory.
- Boot vector allows user provided Flash loader code to reside anywhere in the Flash memory space. This configuration provides flexibility to the user.
- Default loader in Boot ROM allows programming via the serial port without the need of a user-provided loader.
- Up to 64K Byte external program memory if the internal program memory is disabled (EA = 0).
- Programming and erase voltage with standard 5V or 3V V_{CC} supply.
- Read/Programming/Erase:
  - Byte-wise read without wait state
  - Byte or page erase and programming (10 ms)
- Typical programming time (32K Bytes) in 10 s
- Parallel programming with 87C51 compatible hardware interface to programmer
- Programmable security for the code in the Flash
- 100K write cycles
- 10 years data retention

Flash Programming and Erasure

The 16K or 32K Bytes Flash is programmed by Bytes or by pages of 128 Bytes. It is not necessary to erase a Byte or a page before programming. The programming of a Byte or a page includes a self erase before programming.

There are three methods of programming the Flash memory:

- First, the on-chip ISP bootloader may be invoked which will use low level routines to program the pages. The interface used for serial downloading of Flash is the UART.
- Second, the Flash may be programmed or erased in the end-user application by calling low-level routines through a common entry point in the Boot ROM.
- Third, the Flash may be programmed using the parallel method by using a conventional EPROM programmer. The parallel programming method used by these devices is similar to that used by EPROM 87C51 but it is not identical and the commercially available programmers need to have support for the AT89C51RB2/RC2. The bootloader and the Application Programming Interface (API) routines are located in the BOOT ROM.

Bootloader Functionality

Introduction

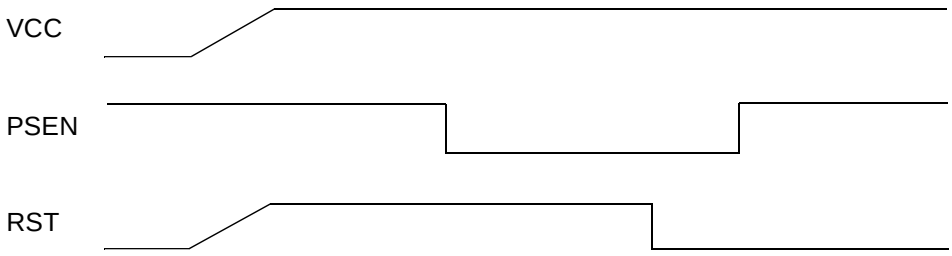
The bootloader can be activated by two means: Hardware conditions or regular boot process.

The Hardware conditions (EA = 1, PSEN = 0) during the Reset# falling edge force the on-chip bootloader execution. This allows an application to be built that will normally execute the end user's code but can be manually forced into default ISP operation.

As PSEN is an output port in normal operating mode (running user application or boorloader code) after reset, it is recommended to release PSEN after falling edge of reset signal. The hardware conditions are sampled at reset signal falling edge, thus they can be released at any time when reset input is low.

To ensure correct microcontroller startup, the PSEN pin should not be tied to ground during power-on (See Figure 38).

Figure 38. Hardware conditions typical sequence during power-on.



The on-chip bootloader boot process is shown in Figure 39.

	Purpose
Hardware Conditions	The Hardware Conditions force the bootloader execution whatever BLJB, BSB and SBV values.
BLJB	The Boot Loader Jump Bit forces the application execution. BLJB = 0 => Boot loader execution. BLJB = 1 => Application execution. The BLJB is a fuse bit in the Hardware Byte. That can be modified by hardware (programmer) or by software (API). Note: The BLJB test is perform by hardware to prevent any program execution.
SBV	The Software Boot Vector contains the high address of custumer bootloader stored in the application. SBV = FCh (default value) if no custumer bootloader in user Flash. Note: The costumer bootloader is called by JMP [SBV]00h instruction.

Autobaud Performances

The ISP feature allows a wide range of baud rates in the user application. It is also adaptable to a wide range of oscillator frequencies. This is accomplished by measuring the bit-time of a single bit in a received character. This information is then used to program the baud rate in terms of timer counts based on the oscillator frequency. The ISP feature requires that an initial character (an uppercase U) be sent to the AT89C51RB2/RC2 to establish the baud rate. Table 72 shows the autobaud capability.

Table 72. Autobaud Performances

Frequency (MHz) **Baudrate (bit/s)**	**1.8432**	**2**	**2.4576**	**3**	**3.6864**	**4**	**5**	**6**	**7.3728**	**8**
2400	OK	OK	OK	OK	OK	OK	OK	OK	OK	OK
4800	OK	-	OK	OK	OK	OK	OK	OK	OK	OK
9600	OK	-	OK	OK	OK	OK	OK	OK	OK	OK
19200	OK	-	OK	OK	OK	-	-	OK	OK	OK
38400	-	-	OK		OK	-	OK	OK	OK	
57600	-	-	-	-	OK	-	-	-	OK	
115200	-	-	-	-	-	-	-	-	OK	
Frequency (MHz) **Baudrate (bit/s)**	**10**	**11.0592**	**12**	**14.318**	**14.746**	**16**	**20**	**24**	**26.6**	
2400	OK	OK	OK	OK	OK	OK	OK	OK	OK	
4800	OK	OK	OK	OK	OK	OK	OK	OK	OK	
9600	OK	OK	OK	OK	OK	OK	OK	OK	OK	
19200	OK	OK	OK	OK	OK	OK	OK	OK	OK	
38400	-	OK	OK	OK	OK	OK	OK	OK	OK	
57600	-	OK	-	OK	OK	OK	OK	OK	OK	
115200	-	OK	-	OK	OK	-	-	-	-	

Command Data Stream Protocol

All commands are sent using the same flow. Each frame sent by the host is echoed by the bootloader.

T_A = -40°C to +85°C; V_{SS} = 0V;

V_{CC} =2.7V to 5.5V and F = 0 to 40 MHz (both internal and external code execution)

V_{CC} =4.5V to 5.5V and F = 0 to 60 MHz (internal code execution only) (Continued)

Symbol	Parameter	Min	Typ	Max	Unit	Test Conditions
V_{OH1}	Output High Voltage, port 0, ALE, $\overline{PSEN}$	V_{CC} - 0.3 V_{CC} - 0.7 V_{CC} - 1.5			V V V	V_{CC} = 5V ± 10% I_{OH} = -200 µA I_{OH} = -3.2 mA I_{OH} = -7.0 mA
		0.9 V_{CC}			V	VCC = 2.7V to 5.5V I_{OH} = -10 µA
R_{RST}	RST Pulldown Resistor	50	200[5]	250	kΩ	
I_{IL}	Logical 0 Input Current ports 1, 2, 3, 4 and 5			-50	µA	V_{IN} = 0.45V
I_{LI}	Input Leakage Current for P0 only			±10	µA	0.45V < V_{IN} < V_{CC}
I_{TL}	Logical 1 to 0 Transition Current, ports 1, 2, 3, 4			-650	µA	V_{IN} = 2.0V
C_{IO}	Capacitance of I/O Buffer			10	pF	Fc = 3 MHz T_A = 25°C
I_{PD}	Power Down Current		100	150	µA	4.5V < V_{CC} < 5.5V[3]
I_{CCOP}	Power Supply Current on normal mode			0.4 x Frequency (MHz) + 5	mA	V_{CC} = 5.5V[1]
I_{CCIDLE}	Power Supply Current on idle mode			0.3 x Frequency (MHz) + 5	mA	V_{CC} = 5.5V[1]
I_{CCProg}	Power Supply Current during flash Write / Erase		0.4 x Frequency (MHz) + 20		mA	V_{CC} = 5.5V[8]

Notes:
1. Operating I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH}, T_{CHCL} = 5 ns (see Figure 49.), V_{IL} = V_{SS} + 0.5V,
   V_{IH} = V_{CC} - 0.5V; XTAL2 N.C.; $\overline{EA}$ = RST = Port 0 = V_{CC}. I_{CC} would be slightly higher if a crystal oscillator used (see Figure 46).
2. Idle I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH}, T_{CHCL} = 5 ns, V_{IL} = V_{SS} + 0.5V, V_{IH} = V_{CC} - 0.5V; XTAL2 N.C; Port 0 = V_{CC}; $\overline{EA}$ = RST = V_{SS} (see Figure 47).
3. Power Down I_{CC} is measured with all output pins disconnected; $\overline{EA}$ = V_{SS}, PORT 0 = V_{CC}; XTAL2 NC.; RST = V_{SS} (see Figure 48).
4. Capacitance loading on Ports 0 and 2 may cause spurious noise pulses to be superimposed on the V_{OL}s of ALE and Ports 1 and 3. The noise is due to external bus capacitance discharging into the Port 0 and Port 2 pins when these pins make 1 to 0 transitions during bus operation. In the worst cases (capacitive loading 100pF), the noise pulse on the ALE line may exceed 0.45V with maxi V_{OL} peak 0.6V. A Schmitt Trigger use is not necessary.
5. Typical are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 5V.
6. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 10 mA
   Maximum I_{OL} per 8-bit port:
   Port 0: 26 mA
   Ports 1, 2 and 3: 15 mA
   Maximum total I_{OL} for all output pins: 71 mA
   If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
7. For other values, please contact your sales office.
8. Icc Flash Write operation current while an on-chip flash page write is on going.
9. Flash Retention is guaranteed with the same formula for V_{CC} Min down to 0.

DC Parameters for Low Voltage

T_A = 0°C to +70°C; V_{SS} = 0V; V_{CC} = 2.7V to 3.6V; F = 0to 40 MHz
T_A = -40°C to +85°C; V_{SS} = 0V; V_{CC} = 2.7V to 3.6V; F = 0 to 40 MHz

Symbol	Parameter	Min	Typ	Max	Unit	Test Conditions
V_{IL}	Input Low Voltage	-0.5		0.2 V_{CC} - 0.1	V	
V_{IH}	Input High Voltage except RST, XTAL1	0.2 V_{CC} + 0.9		V_{CC} + 0.5	V	
V_{IH1}	Input High Voltage, RST, XTAL1	0.7 V_{CC}		V_{CC} + 0.5	V	
V_{OL}	Output Low Voltage, ports 1, 2, 3, 4 (6)			0.45	V	I_{OL} = 0.8 mA (4)
V_{OL1}	Output Low Voltage, port 0, ALE, $\overline{PSEN}$ (6)			0.45	V	I_{OL} = 1.6 mA (4)
V_{OH}	Output High Voltage, ports 1, 2, 3, 4	0.9 V_{CC}			V	I_{OH} = -10 µA
V_{OH1}	Output High Voltage, port 0, ALE, $\overline{PSEN}$	0.9 V_{CC}			V	I_{OH} = -40 µA
I_{IL}	Logical 0 Input Current ports 1, 2, 3, 4			-50	µA	V_{IN} = 0.45 V
I_{LI}	Input Leakage Current for P0 only			±10	µA	0.45V < V_{IN} < V_{CC}
I_{TL}	Logical 1 to 0 Transition Current, ports 1, 2, 3,			-650	µA	V_{IN} = 2.0V
R_{RST}	RST Pulldown Resistor	50	200 (5)	250	kΩ	
C_{IO}	Capacitance of I/O Buffer			10	pF	Fc = 3 MHz T_A = 25°C
I_{PD}	Power Down Current		10 (5)	50	µA	V_{CC} = 2.7V to 3.6V (3)
I_{CCOP}	Power Supply Current on normal mode			0.4 x Frequency (MHz) + 5	mA	V_{CC} = 3.6 V (1)
I_{CCIDLE}	Power Supply Current on idle mode			0.3 x Frequency (MHz) + 5	mA	V_{CC} = 3.6 V (2)
I_{CCProg}	Power Supply Current during flash Write / Erase		0.4 x Frequency (MHz) + 20		mA	V_{CC} = 5.5V (8)

Notes:
1. Operating I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH}, T_{CHCL} = 5 ns (see Figure 49.), V_{IL} = V_{SS} + 0.5V,
   V_{IH} = V_{CC} - 0.5V; XTAL2 N.C.; $\overline{EA}$ = RST = Port 0 = V_{CC}. I_{CC} would be slightly higher if a crystal oscillator used (see Figure 46).
2. Idle I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH}, T_{CHCL} = 5 ns, V_{IL} = V_{SS} + 0.5V, V_{IH} = V_{CC} - 0.5V; XTAL2 N.C; Port 0 = V_{CC}; $\overline{EA}$ = RST = V_{SS} (see Figure 47).
3. Power Down I_{CC} is measured with all output pins disconnected; $\overline{EA}$ = V_{SS}, PORT 0 = V_{CC}; XTAL2 NC.; RST = V_{SS} (see Figure 48).
4. Capacitance loading on Ports 0 and 2 may cause spurious noise pulses to be superimposed on the V_{OL}s of ALE and Ports 1 and 3. The noise is due to external bus capacitance discharging into the Port 0 and Port 2 pins when these pins make 1 to 0 transitions during bus operation. In the worst cases (capacitive loading 100pF), the noise pulse on the ALE line may exceed 0.45V with maxi V_{OL} peak 0.6V. A Schmitt Trigger use is not necessary.
5. Typical are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 5V.
6. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 10 mA
   Maximum I_{OL} per 8-bit port:
   Port 0: 26 mA
   Ports 1, 2 and 3: 15 mA
   Maximum total I_{OL} for all output pins: 71 mA

AC Parameters

Explanation of the AC Symbols

Each timing symbol has 5 characters. The first character is always a “T” (stands for time). The other characters, depending on their positions, stand for the name of a signal or the logical status of that signal. The following is a list of all the characters and what they stand for.

Example: T_{AVLL} = Time for Address Valid to ALE Low.
T_{LLPL} = Time for ALE Low to $\overline{PSEN}$ Low.

(Load Capacitance for port 0, ALE and PSEN = 100 pF; Load Capacitance for all other outputs = 80 pF.)

Table 75 Table 78, and Table 80 give the description of each AC symbols.

Table 77, Table 79 and Table 81 give the AC parameterfor each range.

Table 76, Table 77 and Table 82 gives the frequency derating formula of the AC parameter for each speed range description. To calculate each AC symbols, take the x value in the correponding column (-M or -L) and use this value in the formula.

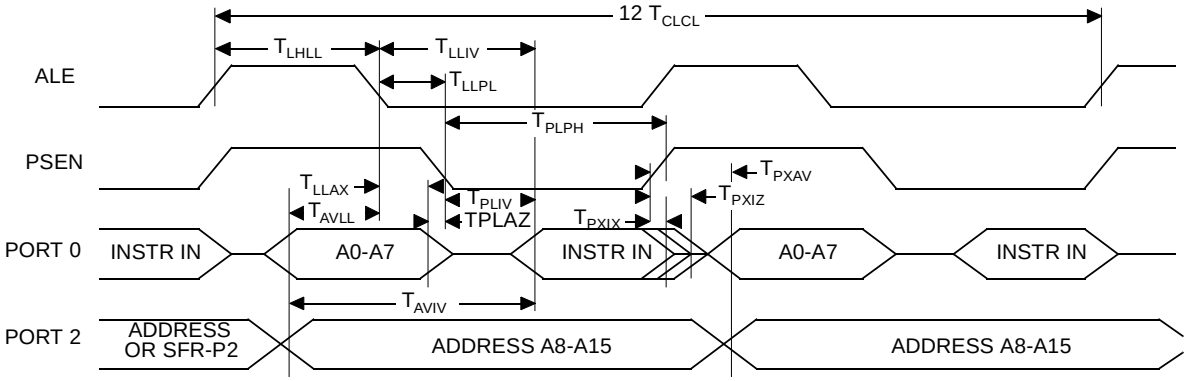
Example: T_{LLIU} for -M and 20 MHz, Standard clock.
x = 35 ns
T 50 ns
T_{CCIV} = 4T - x = 165 ns

External Program Memory Characteristics

Table 75. Symbol Description

Symbol	Parameter
T	Oscillator clock period
T_{LHLL}	ALE pulse width
T_{AVLL}	Address Valid to ALE
T_{LLAX}	Address Hold after ALE
T_{LLIV}	ALE to Valid Instruction In
T_{LLPL}	ALE to $\overline{PSEN}$
T_{PLPH}	$\overline{PSEN}$ Pulse Width
T_{PLIV}	$\overline{PSEN}$ to Valid Instruction In
T_{PXIX}	Input Instruction Hold after $\overline{PSEN}$
T_{PXIZ}	Input Instruction Float after $\overline{PSEN}$
T_{AVIV}	Address to Valid Instruction In
T_{PLAZ}	$\overline{PSEN}$ Low to Address Float

External Program Memory Read Cycle



External Data Memory Characteristics

Table 78. Symbol Description

Symbol	Parameter
T_{RLRH}	$\overline{RD}$ Pulse Width
T_{WLWH}	$\overline{WR}$ Pulse Width
T_{RLDV}	$\overline{RD}$ to Valid Data In
T_{RHDX}	Data Hold After $\overline{RD}$
T_{RHDZ}	Data Float After $\overline{RD}$
T_{LLDV}	ALE to Valid Data In
T_{AVDV}	Address to Valid Data In
T_{LLWL}	ALE to $\overline{WR}$ or $\overline{RD}$
T_{AVWL}	Address to $\overline{WR}$ or $\overline{RD}$
T_{QVWX}	Data Valid to $\overline{WR}$ Transition
T_{QVWH}	Data set-up to $\overline{WR}$ High
T_{WHQX}	Data Hold After $\overline{WR}$
T_{RLAZ}	$\overline{RD}$ Low to Address Float
T_{WHLH}	$\overline{RD}$ or $\overline{WR}$ High to ALE high

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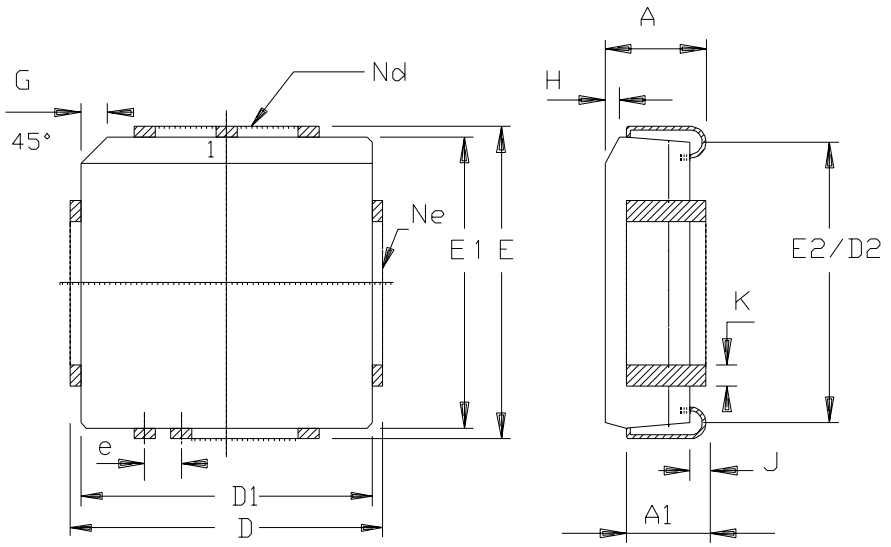

	MM		INCH	
A	4.20	4.57	.165	.180
A1	2.29	3.04	.090	.120
D	17.40	17.65	.685	.695
D1	16.44	16.66	.647	.656
D2	14.99	16.00	.590	.630
E	17.40	17.65	.685	.695
E1	16.44	16.66	.647	.656
E2	14.99	16.00	.590	.630
e	1.27	BSC	.050	BSC
G	1.07	1.22	.042	.048
H	1.07	1.42	.042	.056
J	0.51	–	.020	–
K	0.33	0.53	.013	.021
Nd	11		11	
Ne	11		11	
PKG STD	00			